

Dual 145 μ A, 9.5nV/ $\sqrt{\text{Hz}}$,
 $A_V \geq 5$, Rail-to-Rail Output
 Precision Op Amp

FEATURES

- 60 μ V Maximum Offset Voltage
- Low 1/f Noise: 200nV_{P-P} (0.1Hz to 10Hz)
 40nV_{RMS} (0.1Hz to 10Hz)
- Low White Noise: 9.5nV/ $\sqrt{\text{Hz}}$ (1kHz)
- Rail-to-Rail Output Swing
- 145 μ A Supply Current per Amplifier
- 400pA Maximum Input Bias Current
- $A_V \geq 5$ Stable; Up to 500pF C_{LOAD}
- 0.2V/ μ s Slew Rate
- 1.4MHz Gain Bandwidth Product
- 120dB Minimum Voltage Gain, V_S = ± 15 V
- 0.8 μ V/ $^{\circ}$ C Maximum V_{OS} Drift
- 2.7V to ± 18 V Supply Voltage Operation
- Operating Temperature Range: -40 $^{\circ}$ C to 85 $^{\circ}$ C
- Available in SO-8 and Space Saving 3mm \times 3mm DFN Packages

APPLICATIONS

- Thermocouple Amplifiers
- Precision Photodiode Amplifiers
- Instrumentation Amplifiers
- Battery-Powered Precision Systems
- Low-Voltage Precision Systems
- Micro-Power Sensor Interface

DESCRIPTION

The LT[®]6014 dual op amp combines low noise and high precision input performance with low power consumption and rail-to-rail output swing. The amplifier is stable in a gain of 5 or more and features greatly improved CMRR and PSRR versus frequency compared to other precision op amps.

Input offset voltage is factory-trimmed to less than 60 μ V. The low drift and excellent long-term stability ensure a high accuracy over temperature and time. The 400pA maximum input bias current and 120dB minimum voltage gain further maintain this precision over operating conditions.

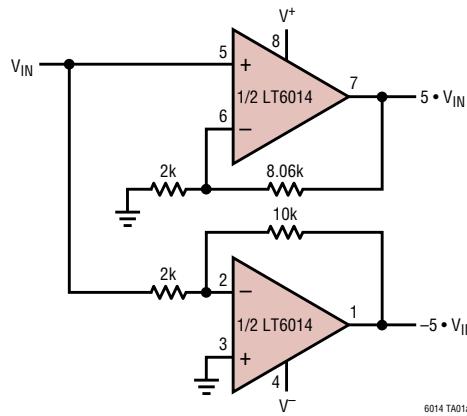
The LT6014 operates from any supply voltage from 2.7V to 36V and draws only 145 μ A of supply current per amplifier on a 5V supply. The output swings to within 40mV of either supply rail, making the amplifier very useful for low voltage single supply operation.

The LT6014 is fully specified at 5V and ± 15 V supplies and from -40 $^{\circ}$ C to 85 $^{\circ}$ C. The device is available in SO-8 and space saving 3mm \times 3mm DFN packages. For a unity gain stable version, refer to the LT6011 data sheet.

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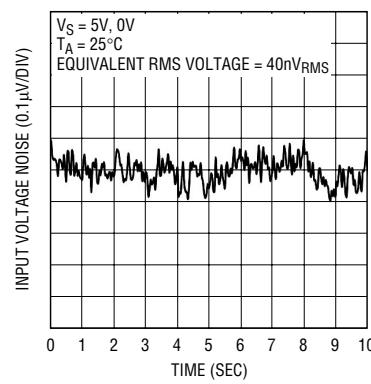
TYPICAL APPLICATION

Gain of 5 Single Ended to Differential Converter



6014 TA01a

LT6014 0.1Hz to 10Hz Voltage Noise

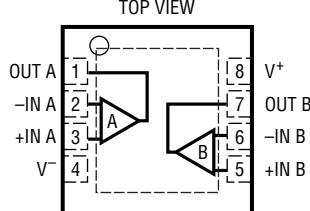


6014 TA01b

ABSOLUTE MAXIMUM RATINGS (Note 1)

Total Supply Voltage (V^+ to V^-)	40V	Maximum Junction Temperature	
Differential Input Voltage (Note 2)	10V	DD Package	125°C
Input Voltage	V^+ to V^-	S8 Package	150°C
Input Current (Note 2)	$\pm 10\text{mA}$	Storage Temperature Range	
Output Short-Circuit Duration (Note 3)	Indefinite	DD Package	-65°C to 125°C
Operating Temperature Range (Note 4) ..	-40°C to 85°C	S8 Package	-65°C to 150°C
Specified Temperature Range (Note 5) ...	-40°C to 85°C	Lead Temperature (Soldering, 10 sec)	300°C

PACKAGE/ORDER INFORMATION

TOP VIEW	ORDER PART NUMBER	ORDER PART NUMBER
	LT6014CDD LT6014IDD LT6014ACDD LT6014AIDD	LT6014CS8 LT6014IS8 LT6014ACS8 LT6014AIS8
	DD PART MARKING*	S8 PART MARKING
OUT A [1] V+ -IN A [2] OUT B [7] +IN A [3] -IN B [6] V- [4] +IN B [5]	LBCB	6014 6014I 6014A 6014AI

*Temperature and electrical grades are identified by a label on the shipping container.

Consult LTC Marketing for parts specified with wider operating temperature ranges.

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_S = 5\text{V}$, 0V ; $V_{CM} = 2.5\text{V}$; R_L to 0V ; unless otherwise specified. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage (Note 8)	LT6014AS8 $T_A = 0^\circ\text{C}$ to 70°C $T_A = -40^\circ\text{C}$ to 85°C	● ●	20 85 110	60 μV	μV
		LT6014S8 $T_A = 0^\circ\text{C}$ to 70°C $T_A = -40^\circ\text{C}$ to 85°C	● ●	25 100 125	75 μV	μV
		LT6014ADD $T_A = 0^\circ\text{C}$ to 70°C $T_A = -40^\circ\text{C}$ to 85°C	● ●	25 135 170	85 μV	μV
		LT6014DD $T_A = 0^\circ\text{C}$ to 70°C $T_A = -40^\circ\text{C}$ to 85°C	● ●	30 175 210	125 μV	μV
$\Delta V_{OS}/\Delta T$	Input Offset Voltage Drift (Note 6)	S8 Package DD Package	● ●	0.2 0.2	0.8 1.2	μV/°C μV/°C
I_{OS}	Input Offset Current (Note 8)	LT6014AS8, LT6014ADD $T_A = 0^\circ\text{C}$ to 70°C $T_A = -40^\circ\text{C}$ to 85°C	● ●	100 600 700	500 pA	pA
		LT6014S8, LT6014DD $T_A = 0^\circ\text{C}$ to 70°C $T_A = -40^\circ\text{C}$ to 85°C	● ●	150 1000 1200	800 pA	pA
I_B	Input Bias Current (Note 8)	LT6014AS8, LT6014ADD $T_A = 0^\circ\text{C}$ to 70°C $T_A = -40^\circ\text{C}$ to 85°C	● ●	100 ±600 ±700	±400 pA	pA
		LT6014S8, LT6014DD $T_A = 0^\circ\text{C}$ to 70°C $T_A = -40^\circ\text{C}$ to 85°C	● ●	150 ±1000 ±1200	±800 pA	pA
e_n	Input Noise Voltage Density	$f = 1\text{kHz}$, LT6014 $f = 1\text{kHz}$, LT6014A		9.5 9.5	13	nV/√Hz nV/√Hz
	Input Noise Voltage (Low Frequency)	Bandwidth = 0.01Hz to 1Hz		200 50		nV _{P-P} nV _{RMS}
		Bandwidth = 0.1Hz to 10Hz		200 40		nV _{P-P} nV _{RMS}
i_n	Input Noise Current Density	$f = 1\text{kHz}$		0.15		pA/√Hz
	Input Noise Current (Low Frequency)	Bandwidth = 0.01Hz to 1Hz		7 1.3		pA _{P-P} pA _{RMS}
		Bandwidth = 0.1Hz to 10Hz		5 0.4		pA _{P-P} pA _{RMS}
R_{IN}	Input Resistance	Common Mode, $V_{CM} = 1\text{V}$ to 3.8V Differential		120 20		GΩ MΩ
C_{IN}	Input Capacitance			4		pF
V_{CM}	Input Voltage Range (Positive) Input Voltage Range (Negative)	Guaranteed by CMRR Guaranteed by CMRR	● ●	3.8 0.7	4 1	V V
CMRR	Common Mode Rejection Ratio	$V_{CM} = 1\text{V}$ to 3.8V	●	107	135	dB
	Minimum Supply Voltage	Guaranteed by PSRR	●		2.4 2.7	V
PSRR	Power Supply Rejection Ratio	$V_S = 2.7\text{V}$ to 36V , $V_{CM} = 1/2V_S$	●	112	135	dB
A_{VOL}	Large-Signal Voltage Gain	$R_L = 10\text{k}$, $V_{OUT} = 1\text{V}$ to 4V $R_L = 2\text{k}$, $V_{OUT} = 1\text{V}$ to 4V	● ●	300 250	2000 2000	V/mV V/mV
	Channel Separation	$V_{OUT} = 1\text{V}$ to 4V	●	110	140	dB

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_S = 5\text{V}$, 0V ; $V_{CM} = 2.5\text{V}$; R_L to 0V ; unless otherwise specified. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OUT}	Maximum Output Swing (Positive, Referred to V^+)	No Load, 50mV Overdrive		35	55	mV
		$I_{SOURCE} = 1\text{mA}$, 50mV Overdrive	●	65	120	mV
	Maximum Output Swing (Negative, Referred to 0V)	No Load, 50mV Overdrive	●	170	40	mV
		$I_{SINK} = 1\text{mA}$, 50mV Overdrive	●	220	225	mV
I_{SC}	Output Short-Circuit Current (Note 3)	$V_{OUT} = 0\text{V}$, 1V Overdrive, Source	●	8	14	mA
		$V_{OUT} = 5\text{V}$, -1V Overdrive, Sink	●	4	21	mA
SR	Slew Rate	$A_V = -10$, $R_F = 50\text{k}$, $R_G = 5\text{k}$	●	0.15	0.2	$\text{V}/\mu\text{s}$
		$T_A = 0^\circ\text{C}$ to 70°C	●	0.12		$\text{V}/\mu\text{s}$
		$T_A = -40^\circ\text{C}$ to 85°C	●	0.1		$\text{V}/\mu\text{s}$
GBW	Gain Bandwidth Product	$f = 10\text{kHz}$	●	1	1.4	MHz
			●	0.9		MHz
t_s	Settling Time	$A_V = -4$, 0.01%, $V_{OUT} = 1.5\text{V}$ to 3.5V		20		μs
t_r, t_f	Rise Time, Fall Time	$A_V = 5$, 10% to 90%, 0.1V Step		1		μs
ΔV_{OS}	Offset Voltage Match (Note 7)	LT6014AS8		50	120	μV
		$T_A = 0^\circ\text{C}$ to 70°C	●	170		μV
		$T_A = -40^\circ\text{C}$ to 85°C	●	220		μV
	LT6014ADD			50	170	μV
		$T_A = 0^\circ\text{C}$ to 70°C	●	270		μV
		$T_A = -40^\circ\text{C}$ to 85°C	●	340		μV
	LT6014S8			50	150	μV
		$T_A = 0^\circ\text{C}$ to 70°C	●	200		μV
		$T_A = -40^\circ\text{C}$ to 85°C	●	250		μV
	LT6014DD			60	250	μV
		$T_A = 0^\circ\text{C}$ to 70°C	●	350		μV
		$T_A = -40^\circ\text{C}$ to 85°C	●	420		μV
ΔI_B	Input Bias Current Match (Note 7)	LT6014AS8, LT6014ADD		200	800	pA
		$T_A = 0^\circ\text{C}$ to 70°C	●	1200		pA
		$T_A = -40^\circ\text{C}$ to 85°C	●	1400		pA
		LT6014S8, LT6014DD		300	1600	pA
$\Delta CMRR$	Common Mode Rejection Ratio Match (Note 7)	$T_A = 0^\circ\text{C}$ to 70°C	●	2000		pA
		$T_A = -40^\circ\text{C}$ to 85°C	●	2400		pA
$\Delta PSRR$	Power Supply Rejection Ratio Match (Note 7)		●	106	135	
			●	101		dB
I_S	Supply Current	per Amplifier		145	165	μA
		$T_A = 0^\circ\text{C}$ to 70°C	●	210		μA
		$T_A = -40^\circ\text{C}$ to 85°C	●	230		μA

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_S = \pm 15\text{V}$, $V_{CM} = 0\text{V}$, R_L to 0V , unless otherwise specified. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage (Note 8)	LT6014AS8 $T_A = 0^\circ\text{C}$ to 70°C $T_A = -40^\circ\text{C}$ to 85°C		30	135	μV
		●		160		μV
		●		185		μV
	LT6014S8 $T_A = 0^\circ\text{C}$ to 70°C $T_A = -40^\circ\text{C}$ to 85°C		35	150		μV
		●		175		μV
		●		200		μV
	LT6014ADD $T_A = 0^\circ\text{C}$ to 70°C $T_A = -40^\circ\text{C}$ to 85°C		35	160		μV
		●		210		μV
		●		225		μV
	LT6014DD $T_A = 0^\circ\text{C}$ to 70°C $T_A = -40^\circ\text{C}$ to 85°C		40	200		μV
		●		250		μV
		●		275		μV
$\Delta V_{OS}/\Delta T$	Input Offset Voltage Drift (Note 6)	S8 Package DD Package	●	0.2	0.8	$\mu\text{V}/^\circ\text{C}$
		●	0.2	1.3		$\mu\text{V}/^\circ\text{C}$
I_{OS}	Input Offset Current (Note 8)	LT6014AS8, LT6014ADD $T_A = 0^\circ\text{C}$ to 70°C $T_A = -40^\circ\text{C}$ to 85°C		100	500	pA
		●		600		pA
		●		700		pA
	LT6014S8, LT6014DD $T_A = 0^\circ\text{C}$ to 70°C $T_A = -40^\circ\text{C}$ to 85°C		150	800		pA
		●		1000		pA
		●		1200		pA
I_B	Input Bias Current (Note 8)	LT6014AS8, LT6014ADD $T_A = 0^\circ\text{C}$ to 70°C $T_A = -40^\circ\text{C}$ to 85°C		100	± 400	pA
		●		± 600		pA
		●		± 700		pA
	LT6014S8, LT6014DD $T_A = 0^\circ\text{C}$ to 70°C $T_A = -40^\circ\text{C}$ to 85°C		150	± 800		pA
		●		± 1000		pA
		●		± 1200		pA
e_n	Input Noise Voltage Density	$f = 1\text{kHz}$, LT6014 $f = 1\text{kHz}$, LT6014A		9.5		$\text{nV}/\sqrt{\text{Hz}}$
				9.5	13	$\text{nV}/\sqrt{\text{Hz}}$
	Input Noise Voltage (Low Frequency)	Bandwidth = 0.01Hz to 1Hz		200		$\text{nV}_{\text{P-P}}$
				50		nV_{RMS}
		Bandwidth = 0.1Hz to 10Hz		200		$\text{nV}_{\text{P-P}}$
				40		nV_{RMS}
i_n	Input Noise Current Density	$f = 1\text{kHz}$		0.15		$\text{pA}/\sqrt{\text{Hz}}$
	Input Noise Current (Low Frequency)	Bandwidth = 0.01Hz to 1Hz		7		$\text{pA}_{\text{P-P}}$
				1.3		pA_{RMS}
		Bandwidth = 0.1Hz to 10Hz		5		$\text{pA}_{\text{P-P}}$
				0.4		pA_{RMS}
R_{IN}	Input Resistance	Common Mode, $V_{CM} = \pm 13.5\text{V}$ Differential		400		$\text{G}\Omega$
				20		$\text{M}\Omega$
C_{IN}	Input Capacitance			4		pF
V_{CM}	Input Voltage Range	Guaranteed by CMRR	●	± 13.5	± 14	V
CMRR	Common Mode Rejection Ratio	$V_{CM} = -13.5\text{V}$ to 13.5V		115	135	dB
		●		112	135	dB
	Minimum Supply Voltage	Guaranteed by PSRR	●	± 1.2	± 1.35	V
PSRR	Power Supply Rejection Ratio	$V_S = \pm 1.35\text{V}$ to $\pm 18\text{V}$	●	112	135	dB
A_{VOL}	Large-Signal Voltage Gain	$R_L = 10\text{k}$, $V_{OUT} = -13.5\text{V}$ to 13.5V		1000	2000	V/mV
		●		600		V/mV
		$R_L = 5\text{k}$, $V_{OUT} = -13.5\text{V}$ to 13.5V		500	1500	V/mV
		●		300		V/mV
	Channel Separation	$V_{OUT} = -13.5\text{V}$ to 13.5V	●	120	140	dB

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_S = \pm 15\text{V}$, $V_{CM} = 0\text{V}$, R_L to 0V , unless otherwise specified. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OUT}	Maximum Output Swing (Positive, Referred to V^+)	No Load, 50mV Overdrive		45	80	mV
		$I_{SOURCE} = 1\text{mA}$, 50mV Overdrive	●		100	mV
	Maximum Output Swing (Negative, Referred to V^-)	No Load, 50mV Overdrive		140	195	mV
		$I_{SINK} = 1\text{mA}$, 50mV Overdrive	●		240	mV
I_{SC}	Output Short-Circuit Current (Note 3)	$V_{OUT} = 0\text{V}$, 1V Overdrive (Source)		45	80	mV
		$V_{OUT} = 0\text{V}$, -1V Overdrive (Sink)	●		100	mV
SR	Slew Rate	$A_V = -10$, $R_F = 50\text{k}$, $R_G = 5\text{k}$		8	15	mA
		$T_A = 0^\circ\text{C}$ to 70°C	●	5		mA
		$T_A = -40^\circ\text{C}$ to 85°C	●		20	mA
GBW	Gain Bandwidth Product	$f = 10\text{kHz}$		0.15	0.2	V/ μs
			●	0.12		V/ μs
t_s	Settling Time	$A_V = -4$, 0.01%, $V_{OUT} = 0\text{V}$ to 10V		0.1		V/ μs
			●		40	μs
t_r, t_f	Rise Time, Fall Time	$A_V = 5$, 10% to 90%, 0.1V Step		0.9		μs
			●		0.9	μs
ΔV_{OS}	Offset Voltage Match (Note 7)	LT6014AS8		50	270	μV
		$T_A = 0^\circ\text{C}$ to 70°C	●		320	μV
		$T_A = -40^\circ\text{C}$ to 85°C	●		370	μV
	LT6014ADD			50	320	μV
		$T_A = 0^\circ\text{C}$ to 70°C	●		420	μV
		$T_A = -40^\circ\text{C}$ to 85°C	●		450	μV
	LT6014S8			70	300	μV
		$T_A = 0^\circ\text{C}$ to 70°C	●		350	μV
		$T_A = -40^\circ\text{C}$ to 85°C	●		400	μV
	LT6014DD			80	400	μV
		$T_A = 0^\circ\text{C}$ to 70°C	●		500	μV
		$T_A = -40^\circ\text{C}$ to 85°C	●		550	μV
ΔI_B	Input Bias Current Match (Note 7)	LT6014AS8, LT6014ADD		200	800	pA
		$T_A = 0^\circ\text{C}$ to 70°C	●		1200	pA
		$T_A = -40^\circ\text{C}$ to 85°C	●		1400	pA
		LT6014S8, LT6014DD		300	1600	pA
$\Delta CMRR$	Common Mode Rejection Ratio Match (Note 7)	$T_A = 0^\circ\text{C}$ to 70°C	●		2000	pA
		$T_A = -40^\circ\text{C}$ to 85°C	●		2400	pA
$\Delta PSRR$	Power Supply Rejection Ratio Match (Note 7)		●	106	135	dB
						dB
I_S	Supply Current	per Amplifier		200	250	μA
		$T_A = 0^\circ\text{C}$ to 70°C	●		290	μA
		$T_A = -40^\circ\text{C}$ to 85°C	●		310	μA

ELECTRICAL CHARACTERISTICS

Note 1: Absolute Maximum Ratings are those beyond which the life of the device may be impaired.

Note 2: The inputs are protected by back-to-back diodes and internal series resistors. If the differential input voltage exceeds 10V, the input current must be limited to less than 10mA.

Note 3: A heat sink may be required to keep the junction temperature below absolute maximum ratings.

Note 4: Both the LT6014C and LT6014I are guaranteed functional over the operating temperature range of -40°C to 85°C .

Note 5: The LT6014C is guaranteed to meet the specified performance from 0°C to 70°C and is designed, characterized and expected to meet specified performance from -40°C to 85°C but is not tested or QA sampled at these temperatures. The LT6014I is guaranteed to meet specified performance from -40°C to 85°C .

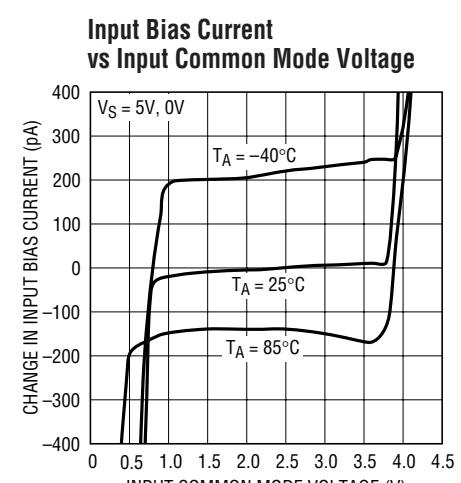
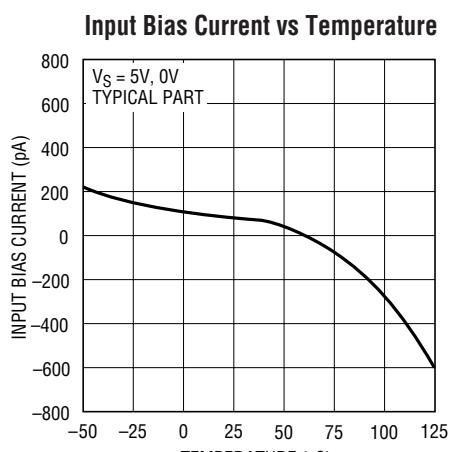
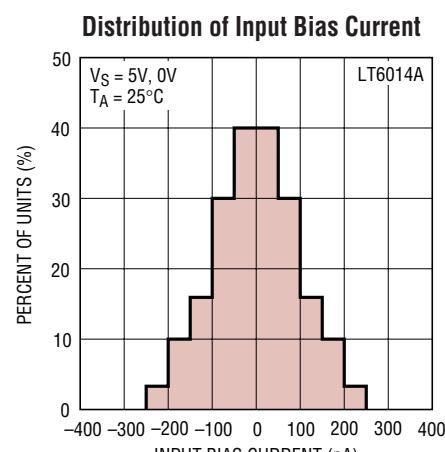
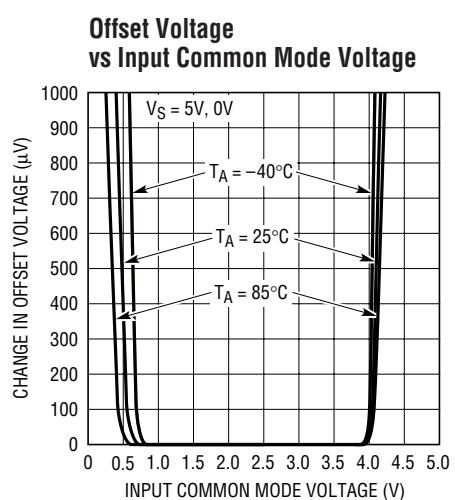
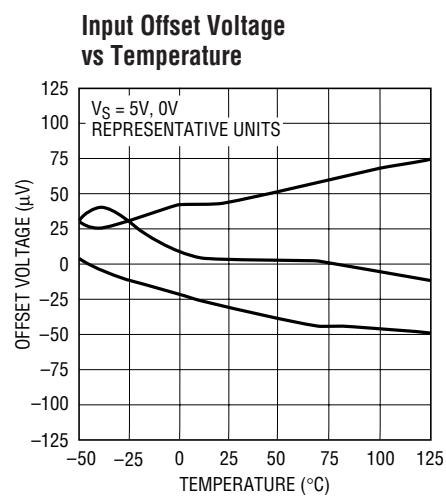
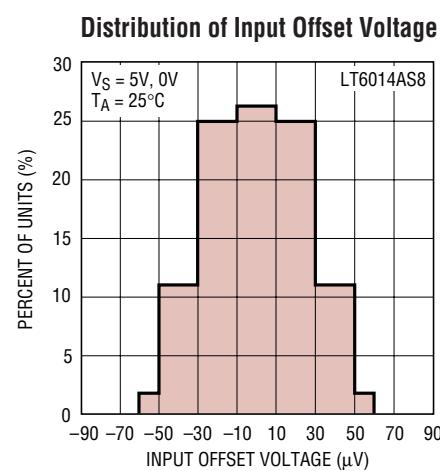
Note 6: This parameter is not 100% tested.

Note 7: Matching parameters are the difference between the two amplifiers. ΔCMRR and ΔPSRR are defined as follows: (1) CMRR and PSRR are measured in $\mu\text{V/V}$ for the individual amplifiers. (2) The difference between matching amplifiers is calculated in $\mu\text{V/V}$. (3) The result is converted to dB.

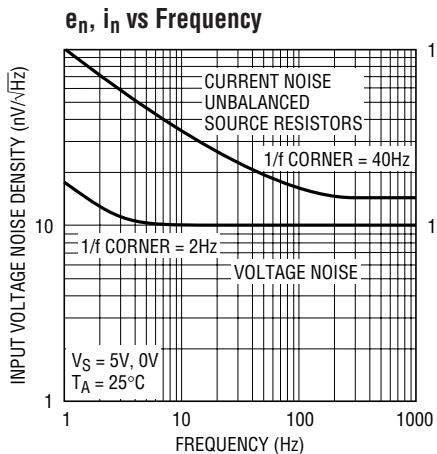
Note 8: The specifications for V_{OS} , I_B , and I_{OS} depend on the grade and on the package. The following table clarifies the notations.

	STANDARD GRADE	A GRADE
S8 Package	LT6014S8	LT6014AS8
DFN Package	LT6014DD	LT6014ADD

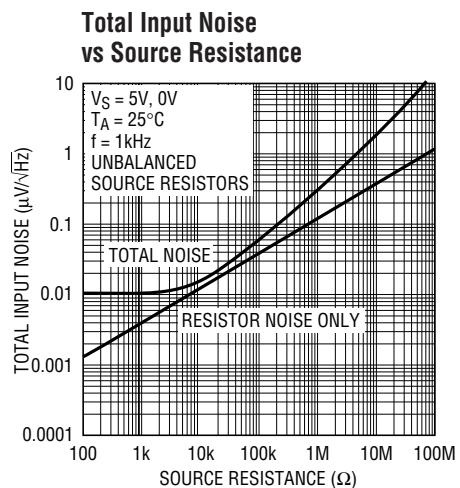
TYPICAL PERFORMANCE CHARACTERISTICS



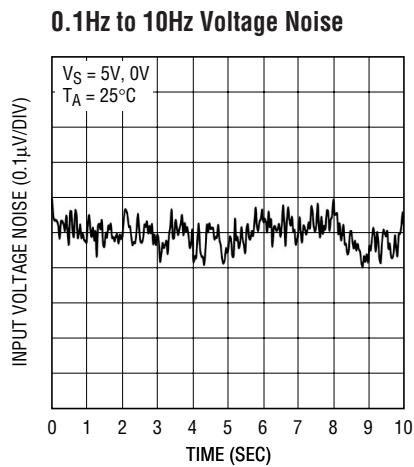
TYPICAL PERFORMANCE CHARACTERISTICS



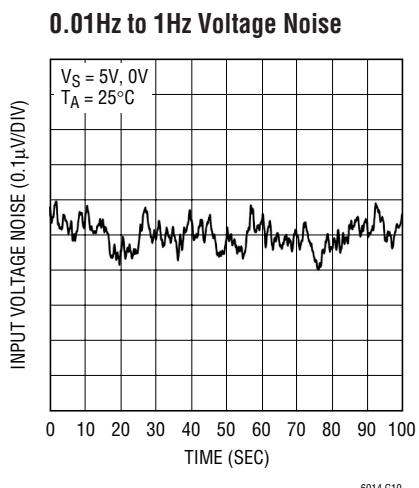
6014 G07



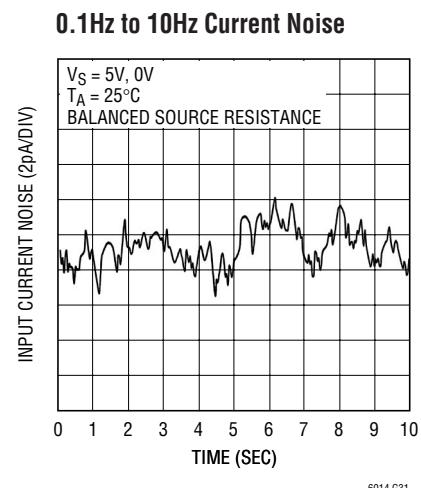
6014 G08



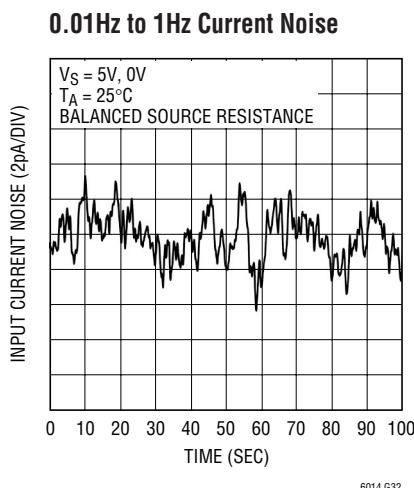
6014 G09



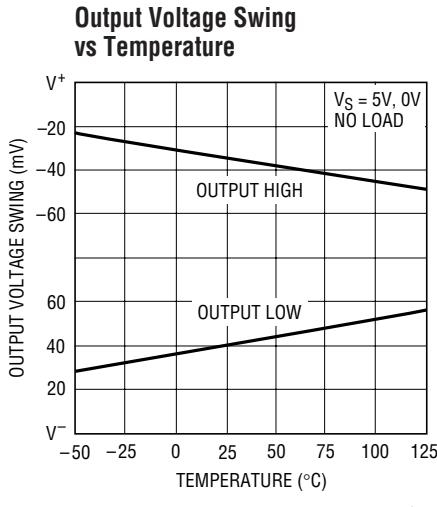
6014 G10



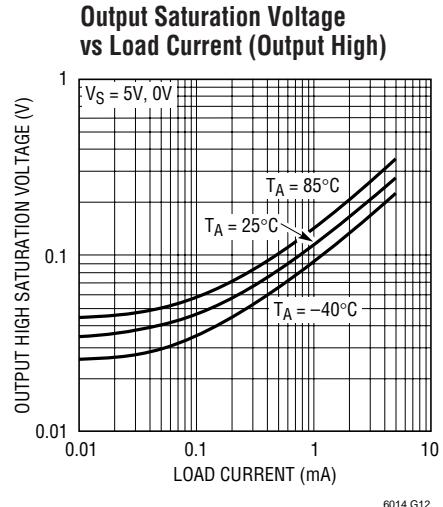
6014 G31



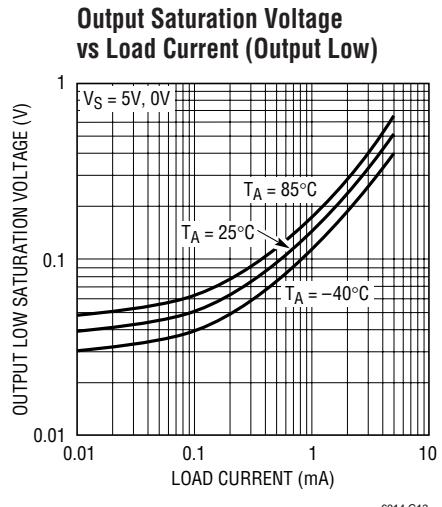
6014 G32



6014 G11

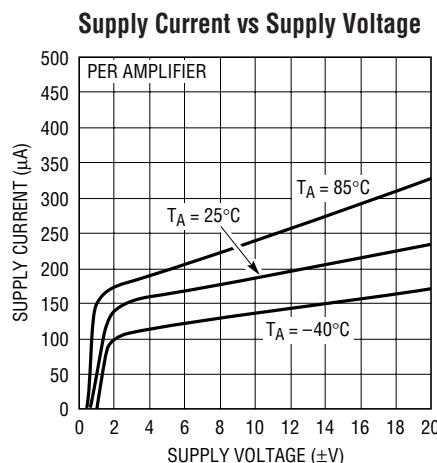


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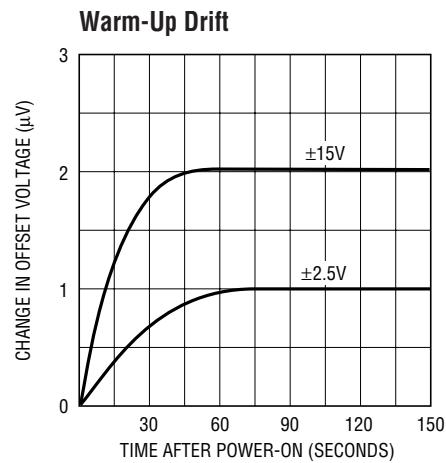


6014 G13

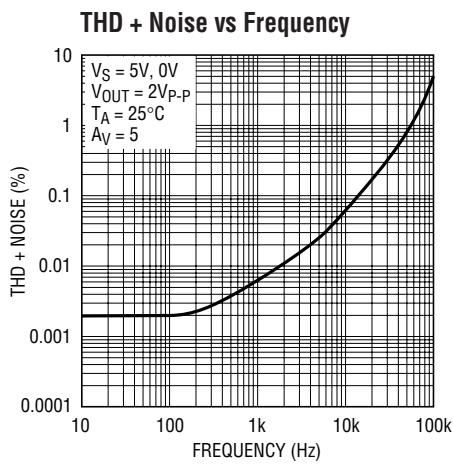
TYPICAL PERFORMANCE CHARACTERISTICS



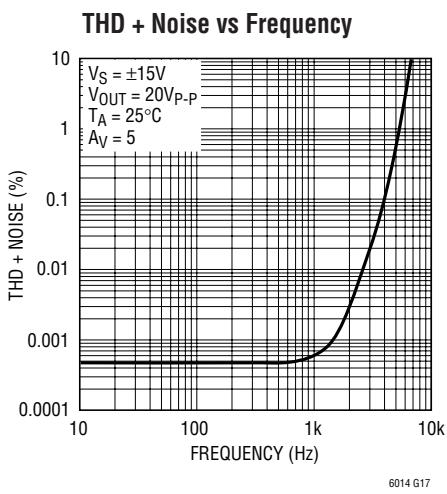
6014 G14



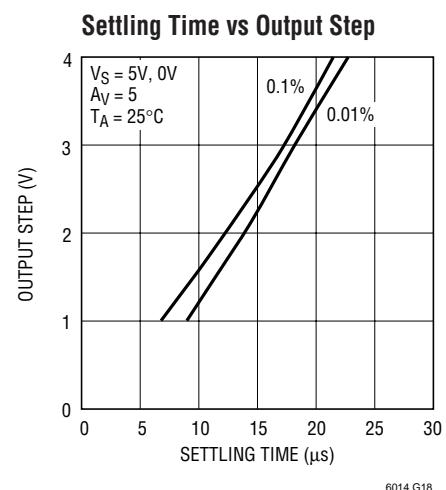
6014 G15



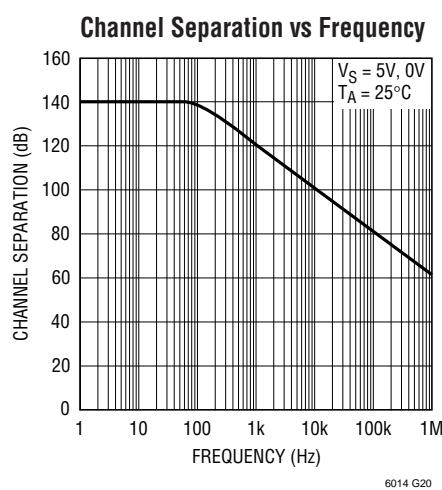
6014 G16



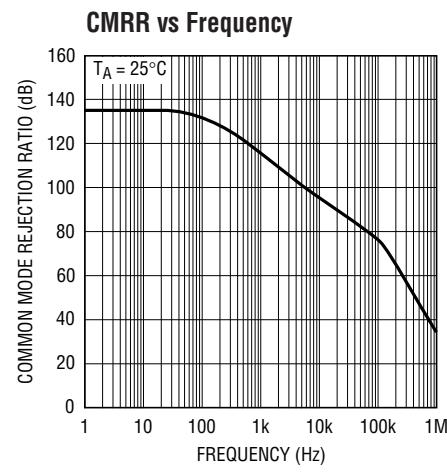
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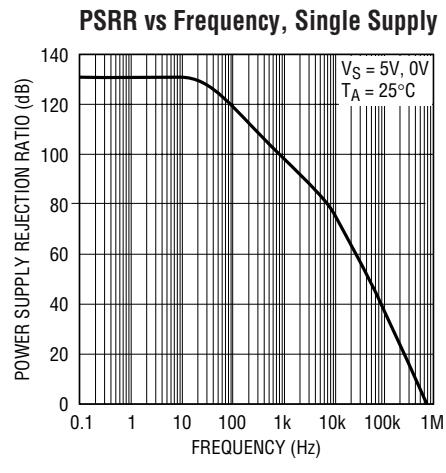
6014 G18



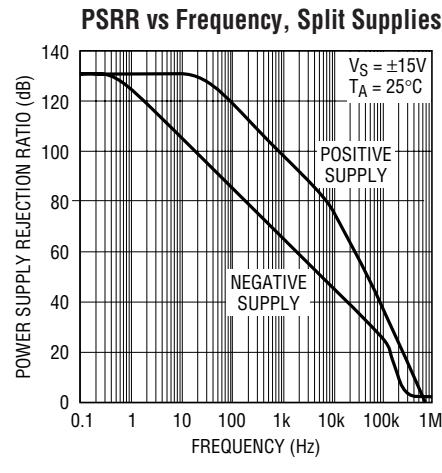
6014 G20



6014 G21



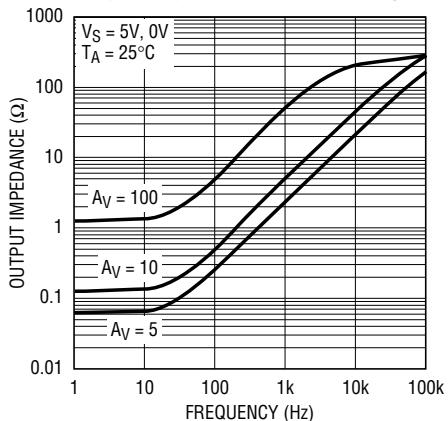
6014 G19



6014 G22

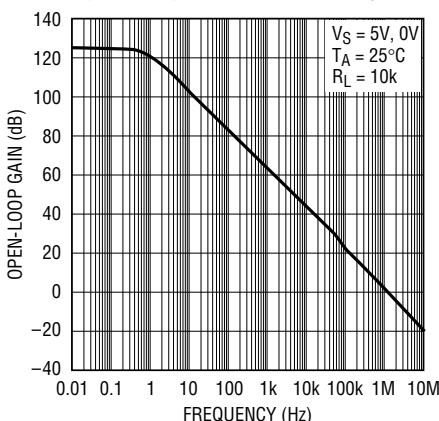
TYPICAL PERFORMANCE CHARACTERISTICS

Output Impedance vs Frequency



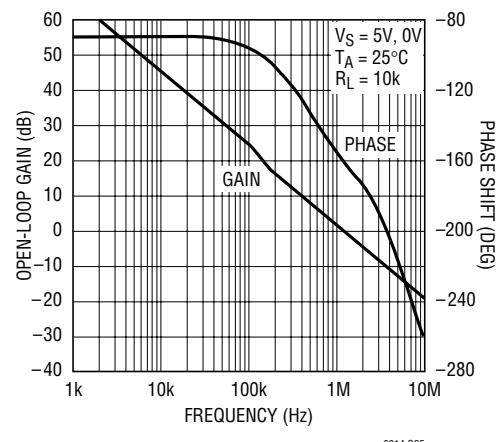
6014 G23

Open-Loop Gain vs Frequency



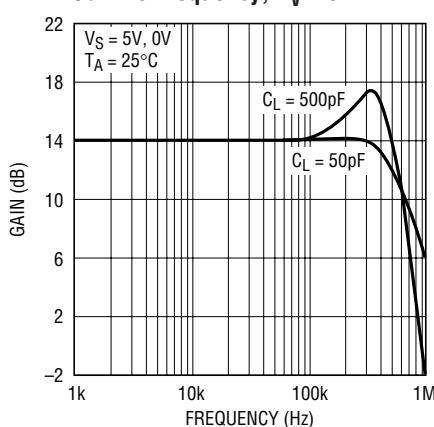
6014 G24

Gain and Phase vs Frequency



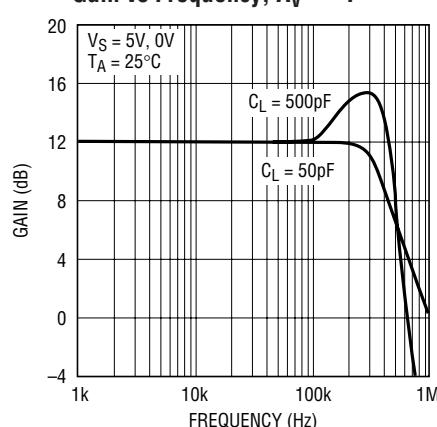
6014 G25

Gain vs Frequency, $A_V = 5$



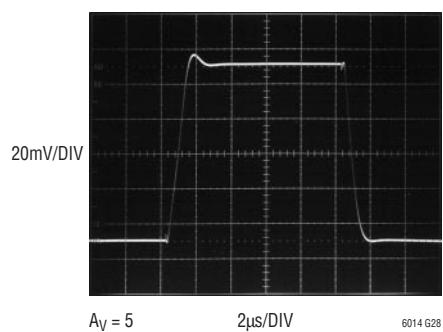
6014 G26

Gain vs Frequency, $A_V = -4$



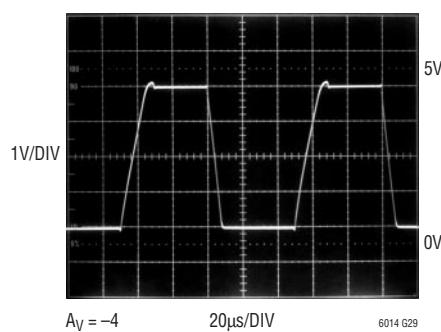
6014 G27

Small-Signal Transient Response



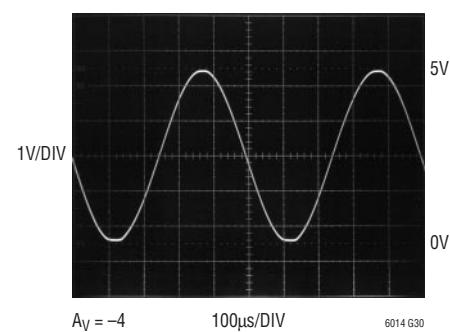
6014 G28

Large-Signal Transient Response



6014 G29

Rail-to-Rail Output Swing



6014 G30

APPLICATIONS INFORMATION

Not Unity-Gain Stable

The LT6014 amplifier is optimized for the lowest possible noise and small package size, and is intentionally compensated to be stable in a gain configuration of 5 or greater. Do not connect the amplifier in a gain less than 5 (such as unity-gain). For a unity-gain stable amplifier with similar performance though slightly higher noise and lower bandwidth, see the LT6011/LT6012 datasheet.

Figure 1 shows simple inverting and non-inverting op amp configurations and indicates how to achieve a gain of 5 or greater. For more general feedback networks, determine the gain that the op amp “sees” as follows:

1. Suppose the op amp is removed from the circuit.
2. Apply a small-signal voltage at the output node of the op amp.
3. Find the differential voltage that would appear across the two inputs of the op amp.
4. The ratio of the output voltage to the input voltage is the gain that the op amp “sees”. This ratio must be 5 or greater.

Do not place a capacitor bigger than 200pF between the output to the inverting input unless there is a 5 times larger capacitor from that input to AC ground. Otherwise, the op amp gain would drop to less than 5 at high frequencies, and the stability of the loop would be compromised.

Preserving Input Precision

Preserving the input accuracy of the LT6014 requires that the applications circuit and PC board layout do not introduce errors comparable to or greater than the 25 μ V typical offset of the amplifiers. Temperature differentials across the input connections can generate thermocouple voltages of 10's of microvolts so the connections to the input leads should be short, close together and away from heat dissipating components. Air currents across the board can also generate temperature differentials.

The extremely low input bias currents allow high accuracy to be maintained with high impedance sources and feedback resistors. The LT6014 low input bias currents are obtained by a cancellation circuit on-chip. This causes the resulting I_B^+ and I_B^- to be uncorrelated, as implied by the I_{OS} specification being comparable to I_B . Do not try to balance the input resistances in each input lead; instead keep the resistance at either input as low as possible for maximum accuracy.

Leakage currents on the PC board can be higher than the input bias current. For example, 10G Ω of leakage between a 15V supply lead and an input lead will generate 1.5nA! Surround the input leads with a guard ring driven to the same potential as the input common mode to avoid excessive leakage in high impedance applications.

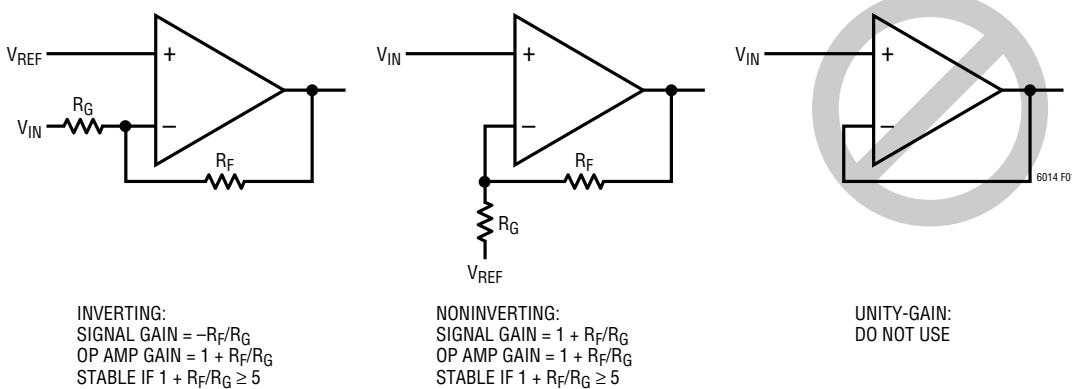


Figure 1. Use LT6014 in a Gain of 5 or Greater

APPLICATIONS INFORMATION

Input Protection

The LT6014 features on-chip back-to-back diodes between the input devices, along with 500Ω resistors in series with either input. This internal protection limits the input current to approximately 10mA (the maximum allowed) for a 10V differential input voltage. Use additional external series resistors to limit the input current to 10mA in applications where differential inputs of more than 10V are expected. For example, a 1k resistor in series with each input provides protection against 30V differential voltage.

Input Common Mode Range

The LT6014 output is able to swing close to each power supply rail (rail-to-rail out), but the input stage is limited to operating between $V^- + 1V$ and $V^+ - 1.2V$. Exceeding this common mode range will cause the gain to drop to zero, however, no phase reversal will occur.

Total Input Noise

The LT6014 amplifier contributes negligible noise to the system when driven by sensors (sources) with impedance

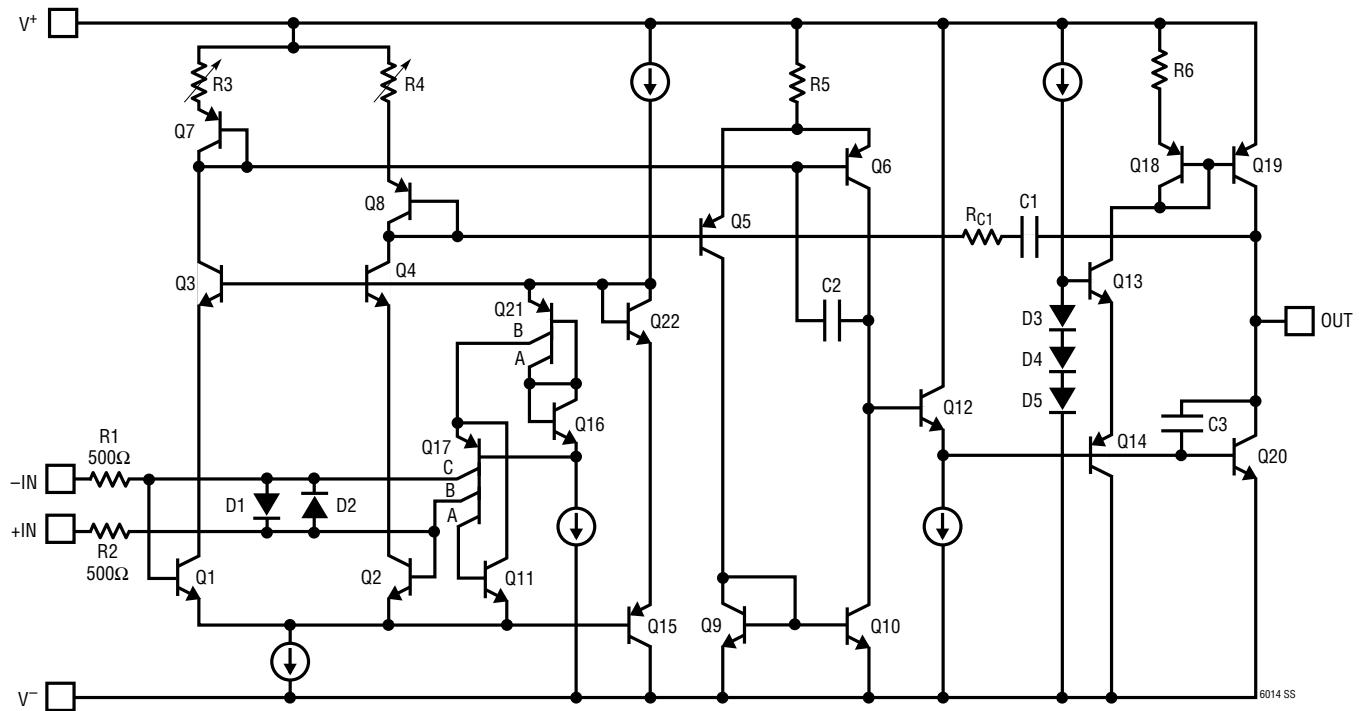
between $10k\Omega$ and $1M\Omega$. Throughout this range, total input noise is dominated by the $4kTR_S$ noise of the source. If the source impedance is less than $10k\Omega$, the input voltage noise of the amplifier starts to contribute with a minimum noise of $9.5nV/\sqrt{Hz}$ for very low source impedance. If the source impedance is more than $1M\Omega$, the input current noise of the amplifier, multiplied by this high impedance, starts to contribute and eventually dominate. Total input noise spectral density can be calculated as:

$$v_{n(TOTAL)} = \sqrt{e_n^2 + 4kTR_S + (i_n R_S)^2}$$

where $e_n = 9.5nV/\sqrt{Hz}$, $i_n = 0.15pA/\sqrt{Hz}$ and R_S is the total impedance at the input, including the source impedance.

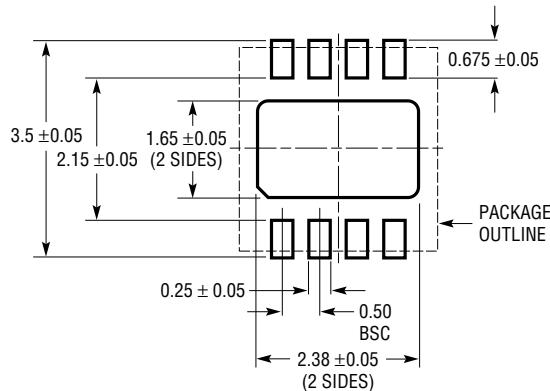
Capacitive Loads

The LT6014 can drive capacitive loads up to $500pF$ at a gain of 5. The capacitive load driving capability increases as the amplifier is used in higher gain configurations. A small series resistance between the output and the load further increases the amount of capacitance that the amplifier can drive.

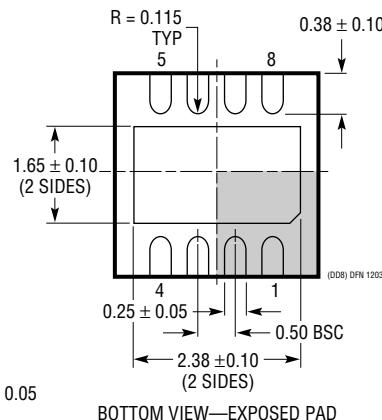
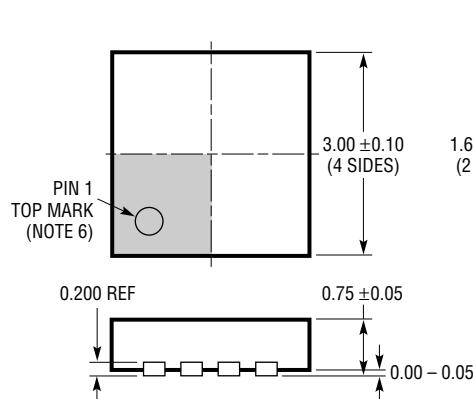
SIMPLIFIED SCHEMATIC (One Amplifier)


PACKAGE DESCRIPTION

DD Package
8-Lead Plastic DFN (3mm × 3mm)
(Reference LTC DWG # 05-08-1698)



RECOMMENDED SOLDER PAD PITCH AND DIMENSIONS

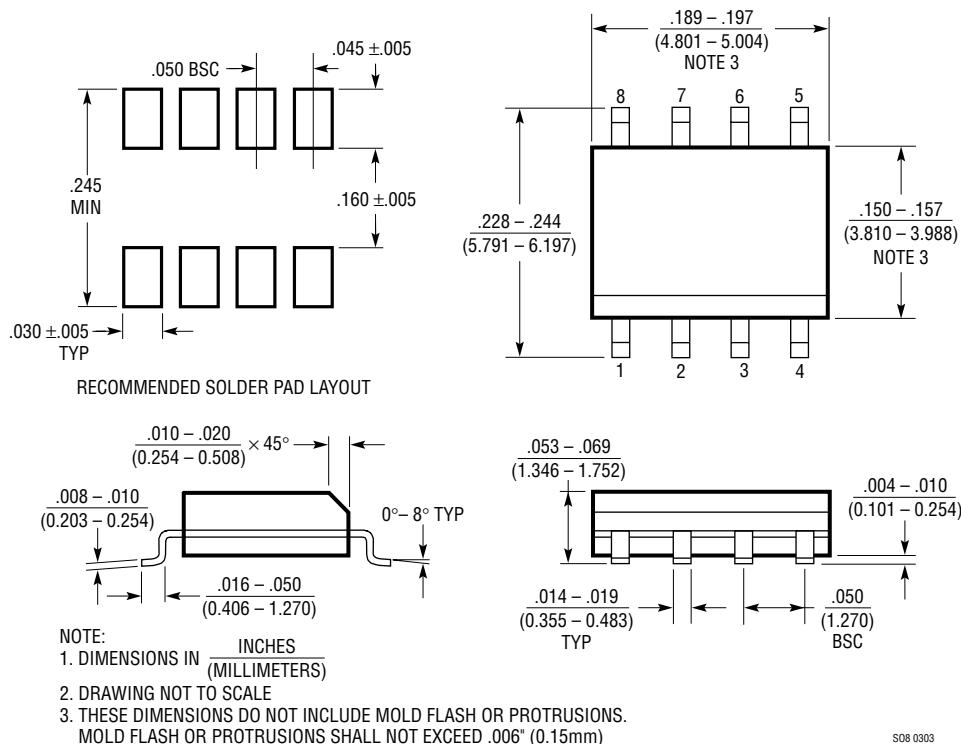


NOTE:

1. DRAWING TO BE MADE A JEDEC PACKAGE OUTLINE M0-229 VARIATION OF (WEED-1)
2. DRAWING NOT TO SCALE
3. ALL DIMENSIONS ARE IN MILLIMETERS
4. DIMENSIONS OF EXPOSED PAD ON BOTTOM OF PACKAGE DO NOT INCLUDE MOLD FLASH. MOLD FLASH, IF PRESENT, SHALL NOT EXCEED 0.15mm ON ANY SIDE
5. EXPOSED PAD SHALL BE SOLDER PLATED
6. SHADeD AREA IS ONLY A REFERENCE FOR PIN 1 LOCATION ON TOP AND BOTTOM OF PACKAGE

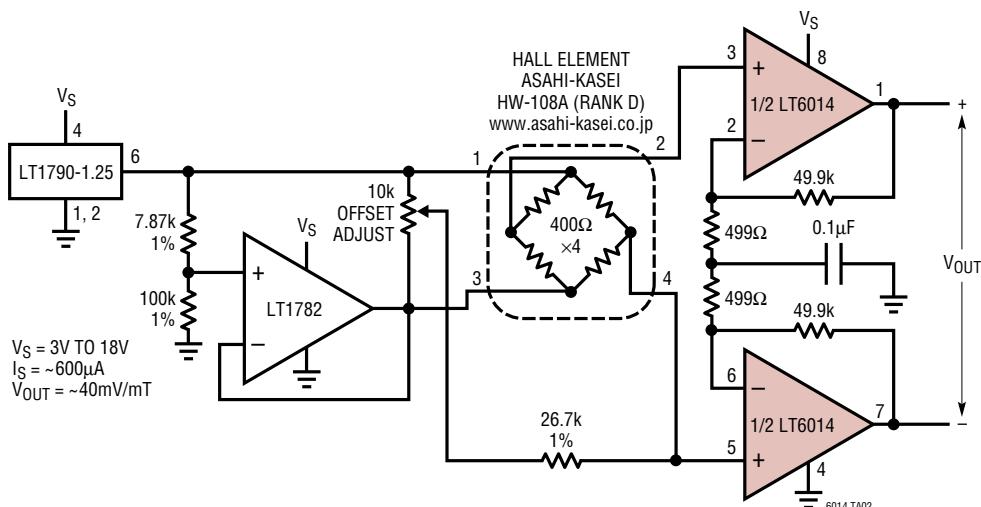
PACKAGE DESCRIPTION

S8 Package
8-Lead Plastic Small Outline (Narrow .150 Inch)
(Reference LTC DWG # 05-08-1610)



TYPICAL APPLICATION

Low Power Hall Sensor Amplifier



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1112/LT1114	Dual/Quad Low Power, Picoamp Input Precision Op Amps	250pA Input Bias Current
LT1880	Rail-to-Rail Output, Picoamp Input Precision Op Amp	SOT-23
LT1881/LT1882	Dual/Quad Rail-to-Rail Output, Picoamp Input Precision Op Amps	C_{LOAD} Up to 1000pF
LT1884/LT1885	Dual/Quad Rail-to-Rail Output, Picoamp Input Precision Op Amps	9.5nV/ $\sqrt{\text{Hz}}$ Input Noise
LT6011/LT6012	Dual/Quad Low Power Rail-to-Rail Output, Precision Op Amps	14nV/ $\sqrt{\text{Hz}}$, Unity-Gain Stable Version of LT6014
LT6010	Single Low Power Rail-to-Rail Output, Precision Op Amp	200pA Input Bias Current, Shutdown Feature