HARRIS
S E M I C O N D U C T O R

HGTG40N60B3

PRELIMINARY

May 1995

70A, 600V, UFS Series N-Channel IGBT

Features

- **70A, 600V at T_C = +25°C**

- **Square Switching SOA Capability**

- **Typical Fall Time - 160ns at +150°C**

- **Short Circuit Rating**

- **Low Conduction Loss**

Description

The HGTG40N60B3 is a MOS gated high voltage switching device combining the best features of MOSFETs and bipolar transistors. The device has the high input impedance of a MOSFET and the low on-state conduction loss of a bipolar transistor. The much lower on-state voltage drop varies only moderately between +25°C and +150°C.

The IGBT is ideal for many high voltage switching applications operating at moderate frequencies where low conduction losses are essential, such as: AC and DC motor controls, power supplies and drivers for solenoids, relays and contactors.
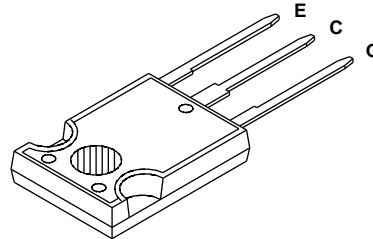
PACKAGING AVAILABILITY

PART NUMBER	PACKAGE	BRAND
HGTG40N60B3	TO-247	G40N60B3

NOTE: When ordering, use the entire part number.
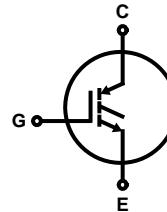
Formerly Developmental Type TA49052

Package

JEDEC STYLE TO-247

Terminal Diagram

N-CHANNEL ENHANCEMENT MODE

Absolute Maximum Ratings T_C = +25°C, Unless Otherwise Specified

	HGTG40N60B3	UNITS
Collector-Emitter Voltage . . . . BV$_{CES}$	600	V
Collector-Gate Voltage, R_{GE} = 1MΩ. . . . . BV$_{CGR}$	600	V
Collector Current Continuous		
At T_C = +25°C (Package Limited) . . . . I$_{C25}$	70	A
At T_C = +110°C . . . . I$_{C110}$	40	A
Collector Current Pulsed (Note 1) . . . . I$_{CM}$	330	A
Gate-Emitter Voltage Continuous. . . . . V$_{GES}$	±20	V
Gate-Emitter Voltage Pulsed . . . . V$_{GEM}$	±30	V
Switching Safe Operating Area at T_C = +150°C. . . . . .SSOA	160A at 0.8 BV$_{CES}$	
Power Dissipation Total at T_C = +25°C . . . . P$_D$	290	W
Power Dissipation Derating T_C > +25°C	2.33	W/^{o}C
Operating and Storage Junction Temperature Range . . . . .T$_J$, T$_{STG}$	-40 to +150	^{o}C
Maximum Lead Temperature for Soldering . . . . .T$_L$	260	^{o}C
Short Circuit Withstand Time (Note 2) at V$_{GE}$ = 15V. . . . . t$_{SC}$	2	µs
Short Circuit Withstand Time (Note 2) at V$_{GE}$ = 10V. . . . . t$_{SC}$	10	µs

NOTE:

1. Repetitive Rating: Pulse width limited by maximum junction temperature.

2. $V_{CE(PK)}$ = 360V, T_C = +125°C, R_{GE} = 25Ω.

File Number **3943**

Electrical Specifications T_C = +25°C, Unless Otherwise Specified

PARAMETERS	SYMBOL	TEST CONDITIONS		LIMITS			UNITS
				MIN	TYP	MAX	
Collector-Emitter Breakdown Voltage	BV_{CES}	I_{CE} = 250µA, V_{GE} = 0V		600	-	-	V
Collector-Emitter Leakage Current	I_{CES}	V_{CE} = BV_{CES}	T_J = +25°C	-	-	250	µA
		V_{CE} = BV_{CES}	T_J = +150°C	-	-	7.5	mA
Collector-Emitter Saturation Voltage	$V_{CE(SAT)}$	I_{CE} = 40A V_{GE} = 15V	T_J = +25°C	-	1.4	2.0	V
			T_J = +150°C	-	1.5	2.3	V
Gate-Emitter Threshold Voltage	$V_{GE(TH)}$	I_{CE} = 250A, V_{CE} = V_{GE}	T_J = +25°C	3.0	5	6.0	V
Gate-Emitter Leakage Current	I_{GES}	V_{GE} = ±20V		-	-	±300	nA
Latching Current	I_L	T_J = +150°C $V_{CE(PK)}$ = 0.8 BV_{CES} V_{GE} = 15V R_G = 3Ω L = 45µH		160	-	-	A
Gate-Emitter Plateau Voltage	V_{GEP}	I_{CE} = 40A, V_{CE} = 0.5 BV_{CES}		-	8.0	-	V
On-State Gate Charge	$Q_{G(ON)}$	I_{CE} = 40A, V_{CE} = 0.5 BV_{CES}	V_{GE} = 15V	-	240	320	nC
			V_{GE} = 20V	-	350	450	nC
Current Turn-On Delay Time	$t_{D(ON)I}$	T_J = +150°C I_{CE} = 40A $V_{CE(PK)}$ = 0.8 BV_{CES} V_{GE} = 15V R_G = 3Ω L = 100µH		-	50	-	ns
Current Rise Time	t_{RI}			-	40	-	ns
Current Turn-Off Delay Time	$t_{D(OFF)I}$			-	350	435	ns
Current Fall Time	t_{FI}			-	160	200	ns
Turn-On Energy	E_{ON}			-	1400	-	µJ
Turn-Off Energy (Note 1)	E_{OFF}			-	3300	-	µJ
Thermal Resistance	$R_{\theta JC}$			-	-	0.43	°C/W

NOTE:

1. Turn-Off Energy Loss (E_{OFF}) is defined as the integral of the instantaneous power loss starting at the trailing edge of the input pulse and ending at the point where the collector current equals zero (I_{CE} = 0A). The HGTG40N60B3 was tested per JEDEC standard No. 24-1 Method for Measurement of Power Device Turn-Off Switching Loss. This test method produces the true total Turn-Off Energy Loss.

Typical Performance Curves

PULSE DURATION = 250µs, DUTY CYCLE <0.5%, V_{CE} = 10V

FIGURE 1. TRANSFER CHARACTERISTICS

PULSE DURATION = 250µs, DUTY CYCLE <0.5%, T_C = +25°C
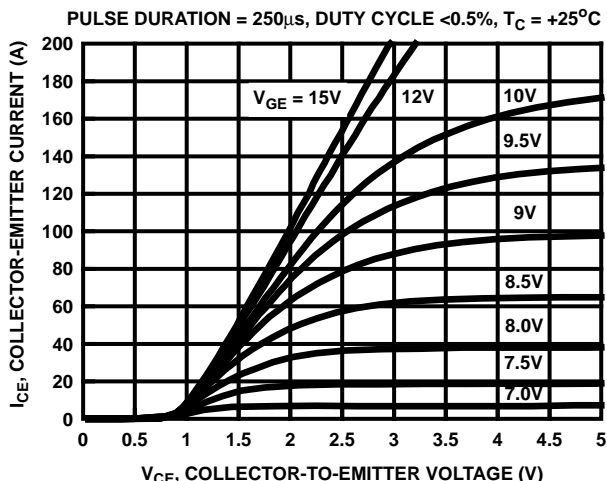
FIGURE 2. SATURATION CHARACTERISTICS
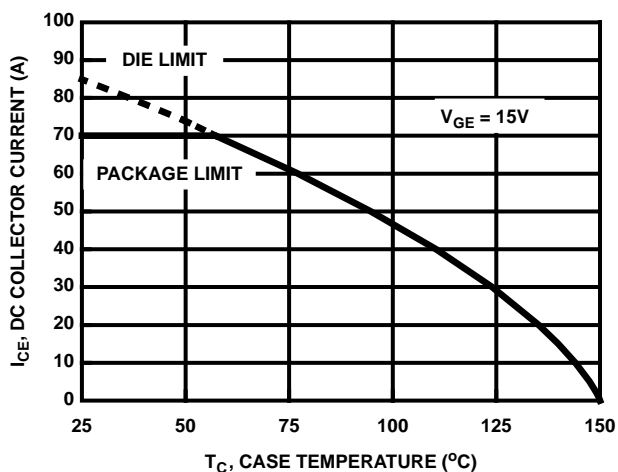
FIGURE 3. DC COLLECTOR CURRENT vs CASE TEMPERATURE

PULSE DURATION = 250µs, DUTY CYCLE <0.5%, V_{GE} = 15V
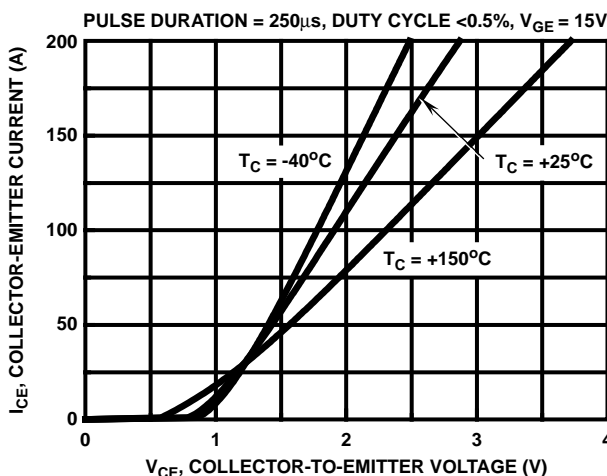
FIGURE 4. COLLECTOR-EMITTER ON-STATE VOLTAGE

FREQUENCY = 1MHz
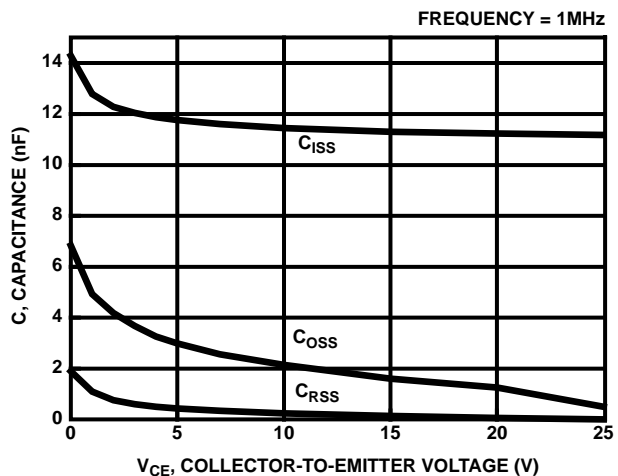
FIGURE 5. CAPACITANCE vs COLLECTOR-EMITTER VOLTAGE

I_{G(REF)} = 4.06mA, R_L = 7.5Ω, T_C = +25°C
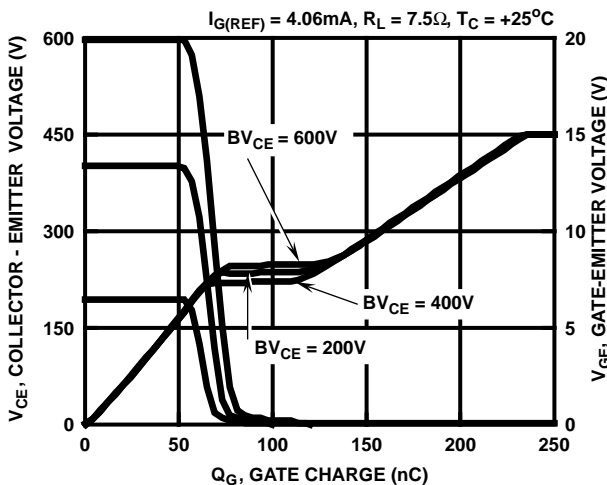
FIGURE 6. GATE CHARGE WAVEFORMS

Typical Performance Curves (Continued)

$T_J = +150^oC$, $R_G = 3\Omega$, $L = 100\mu H$
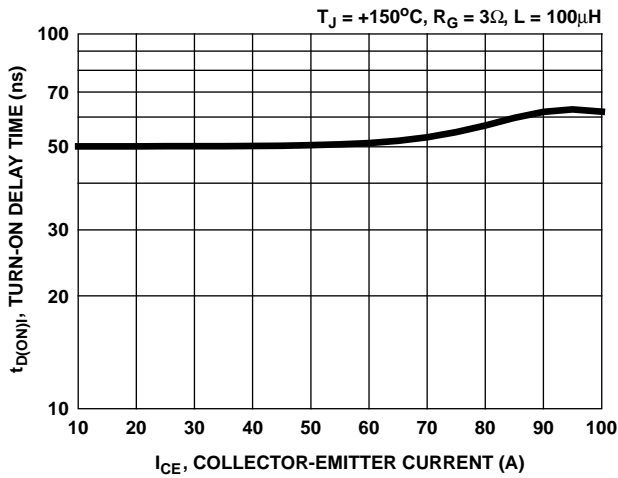
FIGURE 7. TURN-ON DELAY TIME AS A FUNCTION OF COLLECTOR-EMITTER CURRENT

$T_J = +150^oC$, $R_G = 3\Omega$, $L = 100\mu H$

$V_{CE(PK)} = 480V$, $V_{GE} = 15V$
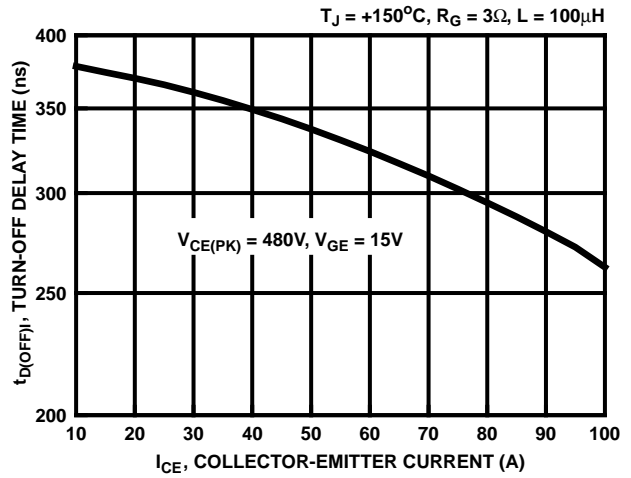
FIGURE 8. TURN-OFF DELAY TIME AS A FUNCTION OF COLLECTOR-EMITTER CURRENT

$T_J = +150^oC$, $R_G = 3\Omega$, $L = 100\mu H$

$V_{CE(PK)} = 480V$, $V_{GE} = 15V$
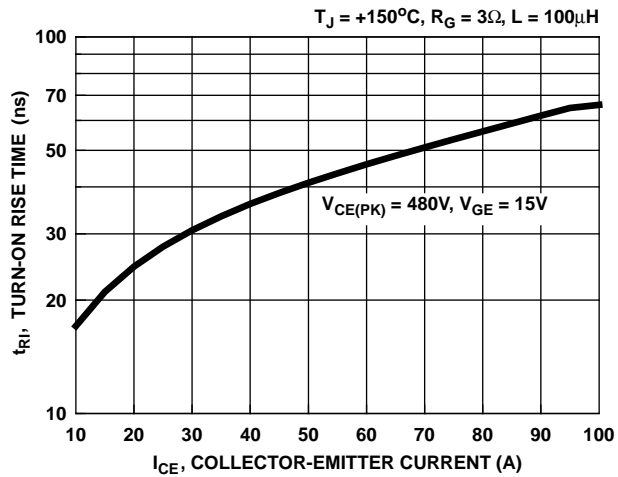
FIGURE 9. TURN-ON RISE TIME AS A FUNCTION OF COLLECTOR-EMITTER CURRENT

$T_J = +150^oC$, $R_G = 3\Omega$, $L = 100\mu H$

$V_{CE(PK)} = 480V$, $V_{GE} = 15V$
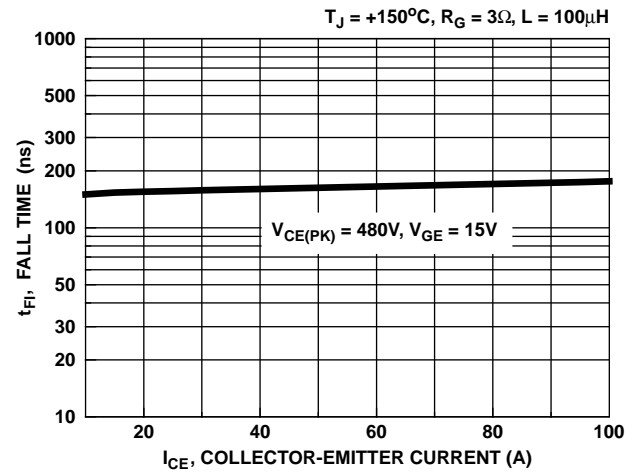
FIGURE 10. TURN-OFF FALL TIME AS A FUNCTION OF COLLECTOR-EMITTER CURRENT

$T_J = +150^oC$, $R_G = 3\Omega$, $L = 100\mu H$

$V_{CE(PK)} = 480V$, $V_{GE} = 15V$
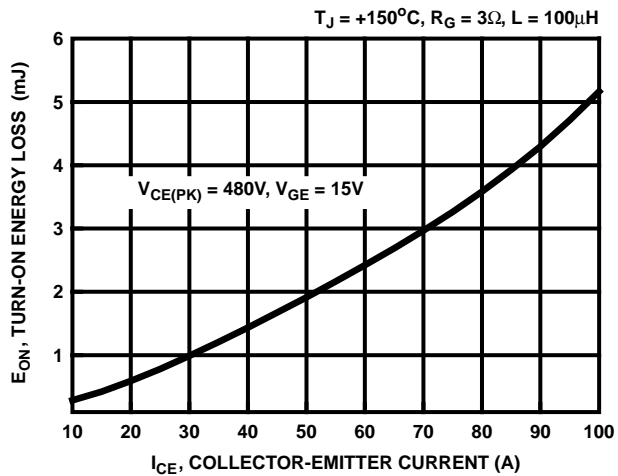
FIGURE 11. TURN-ON ENERGY LOSS AS A FUNCTION OF COLLECTOR-EMITTER CURRENT

$T_J = +150^oC$, $R_G = 3\Omega$, $L = 100\mu H$

$V_{CE(PK)} = 480V$, $V_{GE} = 15V$
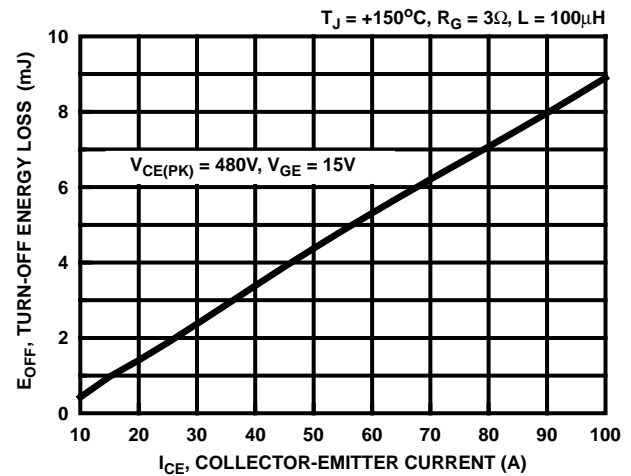
FIGURE 12. TURN-OFF ENERGY LOSS AS A FUNCTION OF COLLECTOR-EMITTER CURRENT

Typical Performance Curves (Continued)

$T_J = +150^oC$, $T_C = +75^oC$, $V_{GE} = +15V$, $R_G = 3\Omega$, $L = 100\mu H$

$f_{MAX1} = 0.05/(t_{D(OFF)I} + t_{D(ON)I})$
$f_{MAX2} = (P_D - P_C)/(E_{ON} + E_{OFF})$
P_D = ALLOWABLE DISSIPATION
P_C = CONDUCTION DISSIPATION
    (DUTY FACTOR = 50%)
$R_{\theta JC} = 0.43^oC/W$

f_{MAX}, OPERATING FREQUENCY (kHz)

I_{CE}, COLLECTOR-EMITTER CURRENT (A)
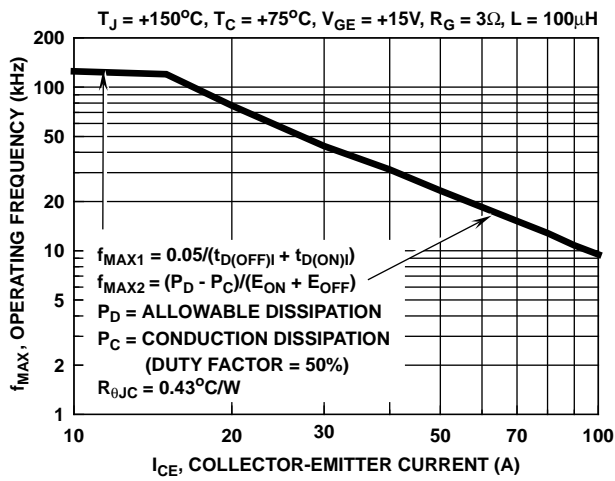
FIGURE 13. OPERATING FREQUENCY AS A FUNCTION OF COLLECTOR-EMITTER CURRENT

$T_C = +150^oC$, $V_{GE} = 15V$, $R_G = 3\Omega$, $L = 45\mu H$

I_{CE}, COLLECTOR-EMITTER CURRENT (A)

V_{CE}, COLLECTOR-EMITTER VOLTAGE (V)
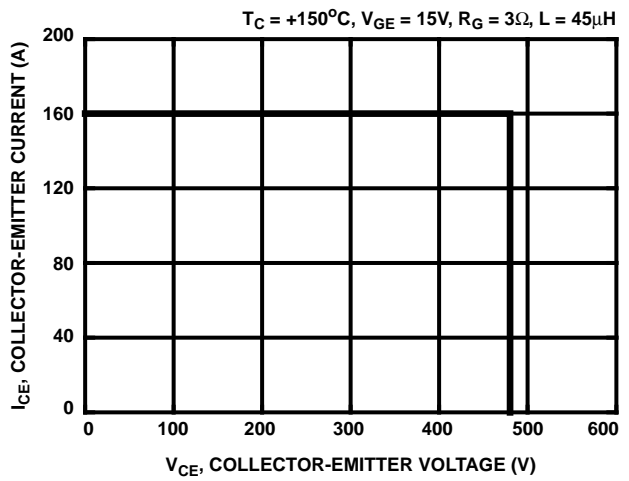
FIGURE 14. SWITCHING SAFE OPERATING AREA

Z_{\theta JC}, NORMALIZED THERMAL RESPONSE (^oC/W)

0.5
0.2
0.1
0.05
0.02
0.01

SINGLE PULSE

NOTES:
DUTY FACTOR, D = t_1/t_2
PEAK $T_J = (P_D \times Z_{\theta JC} \times R_{\theta JC}) + T_C$

t_1, RECTANGULAR PULSE DURATION (s)
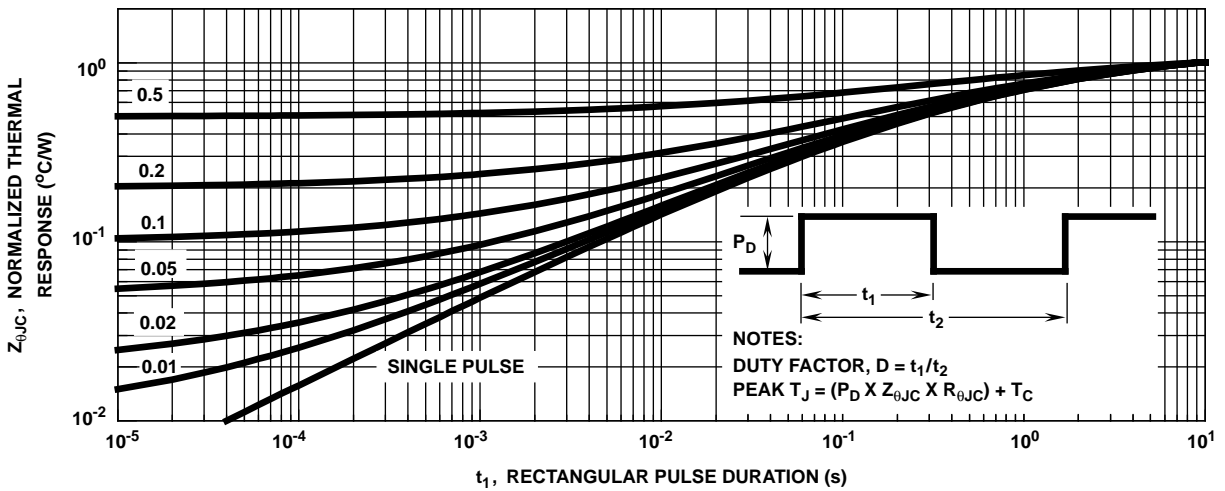
FIGURE 15. IGBT NORMALIZED TRANSIENT THERMAL IMPEDANCE, JUNCTION TO CASE

Test Circuit and Waveforms

L = 100μH

RHRP3060

$R_G = 3\Omega$

$V_{DD} = 480V$
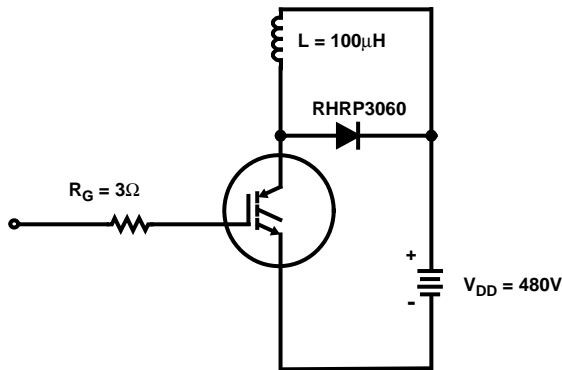
FIGURE 16. INDUCTIVE SWITCHING TEST CIRCUIT

90%

V_{GE}

10%

E_{OFF} E_{ON}

V_{CE}

90%

10%

I_{CE}

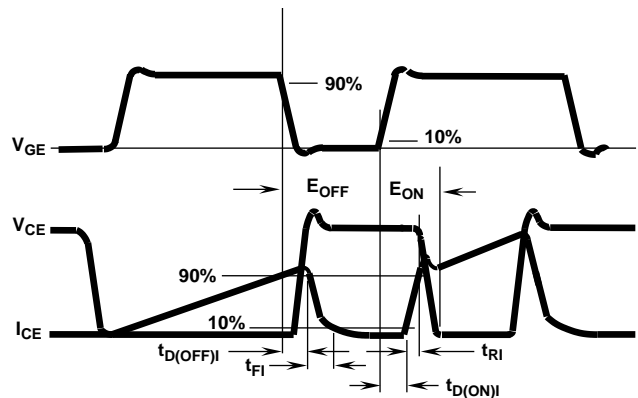$t_{D(OFF)I}$ t_{FI} t_{RI} $t_{D(ON)I}$

FIGURE 17. SWITCHING TEST WAVEFORMS

Operating Frequency Information

Operating frequency information for a typical device (Figure 13) is presented as a guide for estimating device performance for a specific application. Other typical frequency vs collector current (I_{CE}) plots are possible using the information shown for a typical unit in Figures 4, 7, 8, 11 and 12. The operating frequency plot (Figure 13) of a typical device shows f_{MAX1} or f_{MAX2} whichever is smaller at each point. The information is based on measurements of a typical device and is bounded by the maximum rated junction temperature.

f_{MAX1} is defined by $f_{MAX1} = 0.05/(t_{D(OFF)I} + t_{D(ON)I})$. Dead-time (the denominator) has been arbitrarily held to 10% of the on-state time for a 50% duty factor. Other definitions are possible. $t_{D(OFF)I}$ and $t_{D(ON)I}$ are defined in Figure 17.

Device turn-off delay can establish an additional frequency limiting condition for an application other than T_{JMAX}. $t_{D(OFF)I}$ is important when controlling output ripple under a lightly loaded condition.

f_{MAX2} is defined by $f_{MAX2} = (P_D - P_C)/(E_{OFF} + E_{ON})$. The allowable dissipation (P_D) is defined by $P_D = (T_{JMAX} - T_C)/R_{\theta JC}$.

The sum of device switching and conduction losses must not exceed P_D. A 50% duty factor was used (Figure 13) and the conduction losses (P_C) are approximated by $P_C = (V_{CE} \times I_{CE})/2$.

E_{ON} and E_{OFF} are defined in the switching waveforms shown in Figure 17. E_{ON} is the integral of the instantaneous power loss ($I_{CE} \times V_{CE}$) during turn-on and E_{OFF} is the integral of the instantaneous power loss ($I_{CE} \times V_{CE}$) during turn-off. All tail losses are included in the calculation of E_{OFF}; i.e. the collector current equals zero ($I_{CE} = 0$).

Handling Precautions for IGBT's

Insulated Gate Bipolar Transistors are susceptible to gate-insulation damage by the electrostatic discharge of energy through the devices. When handling these devices, care should be exercised to assure that the static charge built in the handler's body capacitance is not discharged through the device. With proper handling and application procedures, however, IGBT's are currently being extensively used in military, industrial and consumer applications, with virtually no damage problems due to electrostatic discharge. IGBT's can be handled safely if the following basic precautions are taken:

1. Prior to assembly into a circuit, all leads should be kept shorted together either by the use of metal shorting springs or by the insertion into conductive material such as †"ECCOSORBD LD26" or equivalent.

2. When devices are removed by hand from their carriers, the hand being used should be grounded by any suitable means - for example, with a metallic wristband.

3. Tips of soldering irons should be grounded.

4. Devices should never be inserted into or removed from circuits with power on.

5. **Gate Voltage Rating** - Never exceed the gate-voltage rating of V_{GEM}. Exceeding the rated V_{GE} can result in permanent damage to the oxide layer in the gate region.

6. **Gate Termination** - The gates of these devices are essentially capacitors. Circuits that leave the gate open-circuited or floating should be avoided. These conditions can result in turn-on of the device due to voltage buildup on the input capacitor due to leakage currents or pickup.

7. **Gate Protection** - These devices do not have an internal monolithic zener diode from gate to emitter. If gate protection is required an external zener is recommended.

† Trademark Emerson and Cumming, Inc.