

EPSON®



S1D13A04 LCD/USB Companion Chip

S1D13A04 TECHNICAL MANUAL

Document Number: X37A-Q-001-01

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S1D13A04 LCD/USB Companion Chip

September 2001

The S1D13A04 is an LCD/USB solution designed for seamless connection to a wide variety of microprocessors. The S1D13A04 integrates a USB slave controller and an LCD graphics controller with an embedded 160K byte SRAM display buffer. The LCD controller, based on the popular S1D13706, supports all standard panel types including the Sharp HR-TFT family of products. In addition to the S1D13706 feature set, the S1D13A04 includes a Hardware Acceleration Engine to greatly improve screen drawing functions. The USB controller provides revision 1.1 compliance for applications requiring a USB client. This high level of integration provides a low cost, low power, single chip solution to meet the demands of embedded markets requiring USB client support, such as Mobile Communications devices and Palm-size PCs.

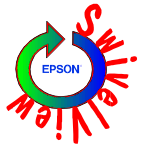
The S1D13A04 utilizes a guaranteed low-latency CPU architecture that provides support for microprocessors without READY/WAIT# handshaking signals. The 32-bit internal data path, write buffer and the Hardware Acceleration Engine provide high performance bandwidth into display memory allowing for fast display updates. 'Direct' support for the Sharp HR-TFT removes the requirement of an external Timing Control IC.

Additionally, products requiring a rotated display can take advantage of the SwivelView™ feature which provides hardware rotation of the display memory transparent to the software application. The S1D13A04 also provides support for "Picture-in-Picture Plus" (a variable size Overlay window).

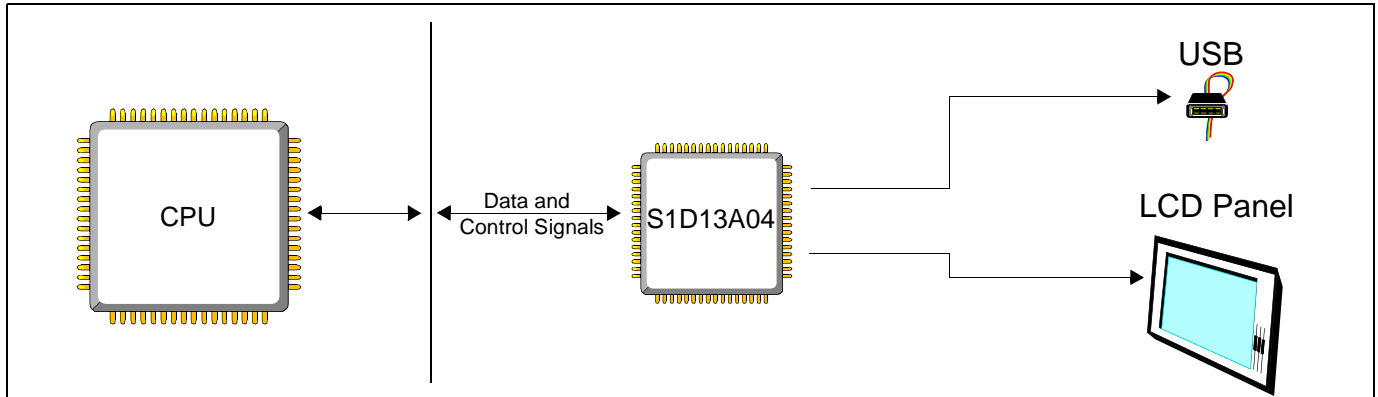
The S1D13A04, with its integrated USB client, provides impressive support for Palm OS® handhelds. However, its impartiality to CPU type or operating system makes it an ideal display solution for a wide variety of applications.

■ **FEATURES**

- Embedded 160KB Display Buffer.
- Low Operating Voltage.
- Low-latency CPU interface.
- Direct support for multiple CPU types.
- Programmable resolutions and color depths.
- Passive LCD support.
- Active Matrix LCD support.
- 'Direct' Sharp HR-TFT support.
- USB Client, Revision 1.1 compliant.
- SwivelView™ (90°, 180°, 270° hardware rotation of displayed image).
- "Picture-in-Picture Plus".
- Pixel Doubling.
- Hardware Acceleration Engine.
- Software Initiated Power Save Mode.
- Software Video Invert.
- 121-pin PFBGA or TQFP15 128-pin package.



■ **SYSTEM BLOCK DIAGRAM**



S1D13A04

DESCRIPTION

CPU Interface

- 'Fixed' low-latency CPU access times.
- Direct support for:
 - Hitachi SH-4 / SH-3.
 - Motorola M68xxx (REDCAP2, DragonBall, ColdFire).
 - MPU bus interface with programmable READY.

Memory Interface

- Embedded 160K byte SRAM display buffer.

Power Down Modes

- Software Initiated Power Save Mode.

Operating Voltage

- $CORE_{VDD}$ $2.0 \pm 10\%$ or $2.5 \pm 10\%$ volts.
- IO_{VDD} $3.0 \pm 10\%$ volts.

Clock Source

- Three independent clock inputs including dedicated USB clock (single clock possible if USB not required).
- Flexible clock source selection and divides.

Package

- 121-pin PFBGA.
- 128-pin TQFP15.

Integrated USB Features

- USB Client, Revision 1.1 Compliant.

Integrated LCD Controller Features

- 1/2/4/8/16 bit-per-pixel (bpp) support.
- Up to 64 gray shades on monochrome passive panels.
- Up to 64K colors on passive/active matrix panels.
- Single-panel, single-drive passive displays.
 - 4/8-bit monochrome LCD interface.
 - 4/8/16-bit color passive LCD interface.
- 9/12/18-bit Active matrix TFT interface.
- 18-bit 'Direct' HR-TFT interface.
- SwivelView: hardware rotation by 90°, 180°, 270°.
- "Picture-in-Picture Plus": displays a variable size window overlaid over background image.
- Pixel Doubling: horizontal and vertical resolutions can be doubled without any additional memory.
- Software video invert.
- Typical resolutions supported:
 - 320x240@16 bpp
 - 320x320@8 bpp
 - 160x160@16 bpp (2 pages)
 - 160x240@16 bpp
- 2D BitBLT Engine.

Write BLT	Transparent Write BLT
Move BLT	Transparent Move BLT
Solid Fill BLT	Read BLT
Pattern Fill	Color Expansion BLT
Move BLT with Color Expansion	

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S1D13A04 LCD/USB Companion Chip

Hardware Functional Specification

Document Number: X37A-A-001-06

Status: Revision 6.0

Issue Date: 2003/05/01

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1 Introduction

1.1 Scope

This is the Hardware Functional Specification for the S1D13A04 LCD/USB Companion Chip. Included in this document are timing diagrams, AC and DC characteristics, register descriptions, and power management descriptions. This document is intended for two audiences: Video Subsystem Designers and Software Developers.

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We appreciate your comments on our documentation. Please contact us via email at documentation@erd.epson.com.

1.2 Overview Description

The S1D13A04 is an LCD/USB solution designed for seamless connection to a wide variety of microprocessors. The S1D13A04 integrates a USB slave controller and an LCD graphics controller with an embedded 160K byte SRAM display buffer. The LCD controller, based on the popular S1D13706, supports all standard panel types including the Sharp HR-TFT family of products. In addition to the S1D13706 feature set, the S1D13A04 includes a Hardware Acceleration Engine to greatly improve screen drawing functions. The USB controller provides revision 1.1 compliance for applications requiring a USB client. This high level of integration provides a low cost, low power, single chip solution to meet the demands of embedded markets requiring USB client support, such as Mobile Communications devices and Palm-size PCs.

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Additionally, products requiring a rotated display can take advantage of the SwivelView™ feature which provides hardware rotation of the display memory transparent to the software application. The S1D13A04 also provides support for "Picture-in-Picture Plus" (a variable size Overlay window).

The S1D13A04, with its integrated USB client, provides impressive support for Palm OS® handhelds. However, its impartiality to CPU type or operating system makes it an ideal display solution for a wide variety of applications.

2 Features

2.1 Integrated Frame Buffer

- Embedded 160k byte SRAM display buffer.

2.2 CPU Interface

- Direct support of the following interfaces:
 - Generic MPU bus interface with programmable ready (WAIT#).
 - Hitachi SH-4 / SH-3.
 - Motorola M68K.
 - Motorola MC68EZ328/MC68VZ328 DragonBall.
 - Motorola "REDCAP2" - no WAIT# signal.
- "Fixed" low-latency CPU access times.
- Registers are memory-mapped - M/R# input selects between memory and register address space.
- The complete 160k byte display buffer is directly and contiguously available through the 18-bit address bus.

2.3 Display Support

- Single-panel, single drive passive displays.
 - 4/8-bit monochrome LCD interface.
 - 4/8/16-bit color LCD interface.
- Active Matrix TFT interface.
 - 9/12/18-bit interface.
- 'Direct' support for 18-bit Sharp HR-TFT LCD or compatible interface.

2.4 Display Modes

- 1/2/4/8/16 bit-per-pixel (bpp) color depths.
- Up to 64 gray shades on monochrome passive LCD panels.
- Up to 64K colors on passive panels.
- Up to 64K colors on active matrix LCD panels.
- Example resolutions:
 - 320x240 at a color depth of 16 bpp
 - 320x320 at a color depth of 8 bpp
 - 160x160 at a color depth of 16 bpp (2 pages)
 - 160x240 at a color depth of 16 bpp

2.5 Display Features

- SwivelView™: 90°, 180°, 270° counter-clockwise hardware rotation of display image.
- Virtual display support: displays images larger than the panel size through the use of panning and scrolling.
- Picture-in-Picture Plus (PIP⁺): displays a variable size window overlaid over background image.
- Pixel Doubling: independent control of both horizontal and vertical pixel doubling.
 - example usage: 160x160 8 bpp can be expanded to 320x320 8 bpp without any additional memory.
- Double Buffering/Multi-pages: provides smooth animation and instantaneous screen updates.

2.6 Clock Source

- Three independent clock inputs: CLKI, CLKI2 and USBCLK.
- Flexible clock source selection:
 - internal Bus Clock (BCLK) selected from CLKI or CLKI/2 (CNF6)
 - internal Memory Clock (MCLK) selected from BCLK or BCLK divide ratio (REG[04h])
 - internal Pixel Clock (PCLK) selected from CLKI, CLKI2, MCLK, or BCLK. PCLK can also be divided down from source (REG[08h])
- Single clock input possible if USB support not required.

2.7 USB Device

- USB Client, revision 1.1 compliant.
- Dedicated clock input: USBCLK.

2.8 2D Acceleration

- 2D BitBLT engine including:

Write BitBLT	Transparent Write BitBLT
Move BitBLT	Transparent Move BitBLT
Solid Fill BitBLT	Read BitBLT
Pattern Fill BitBLT	Color Expansion BitBLT
Move BitBLT with Color Expansion	

2.9 Miscellaneous

- Software Video Invert.
- Software initiated Power Save mode.
- General Purpose Input/Output pins are available.
- IO Operates at 3.3 volts \pm 10%.
- Core operates at 2.0 volts \pm 10% or 2.5 volts \pm 10%.
- 121-pin PFBGA package.
- 128-pin TQFP15 package.

3 Typical System Implementation Diagrams

3.1 Typical System Diagrams.

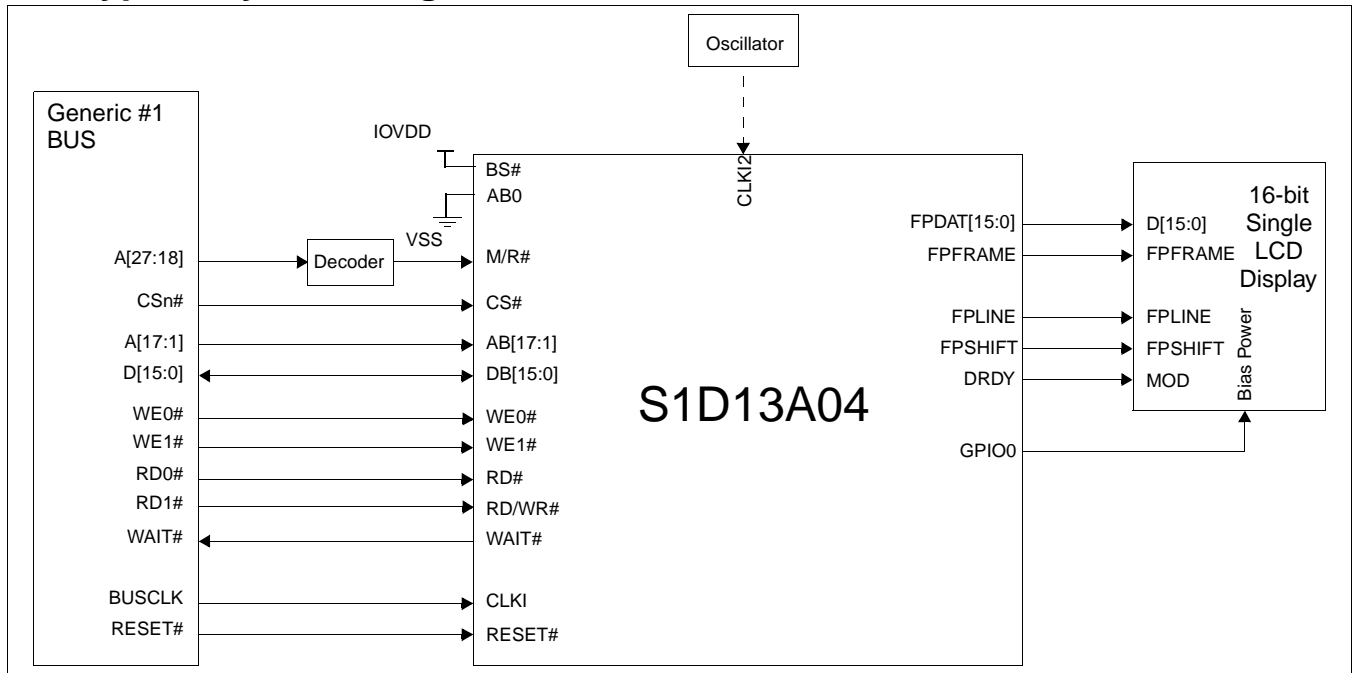


Figure 3-1: Typical System Diagram (Generic #1 Bus)

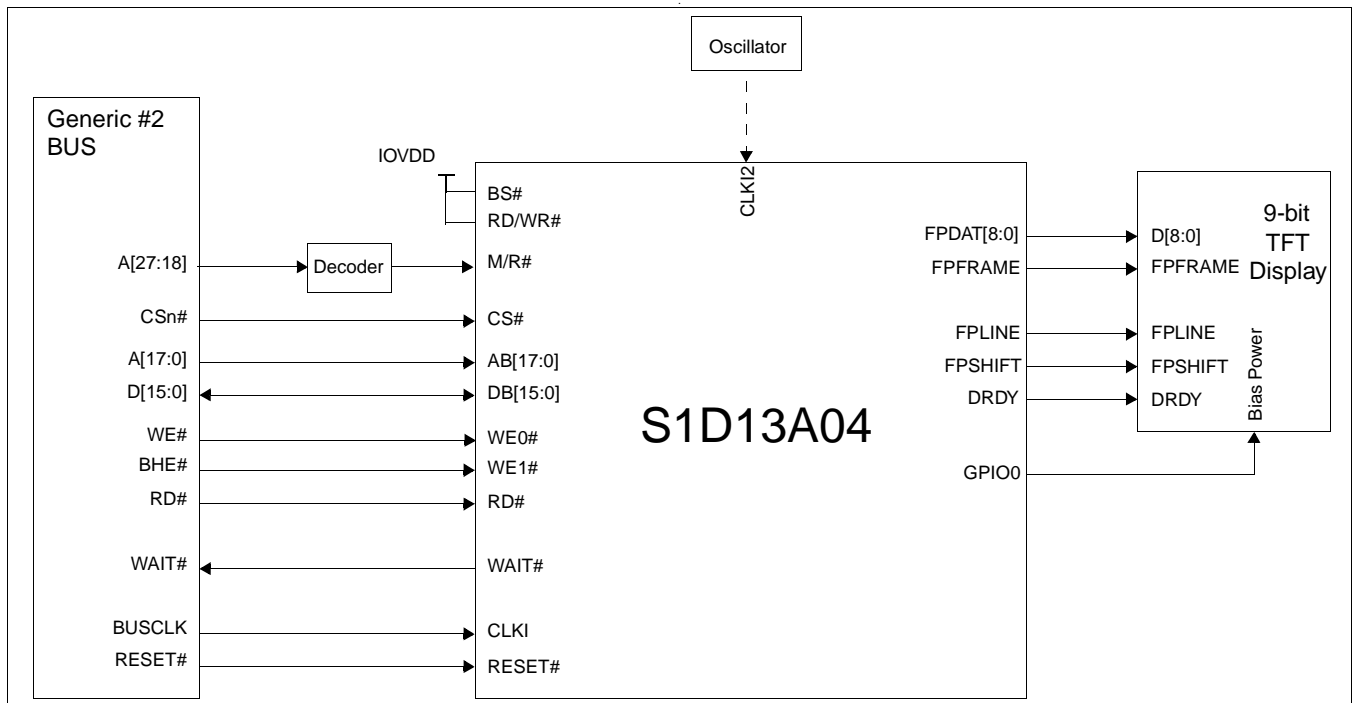


Figure 3-2: Typical System Diagram (Generic #2 Bus)

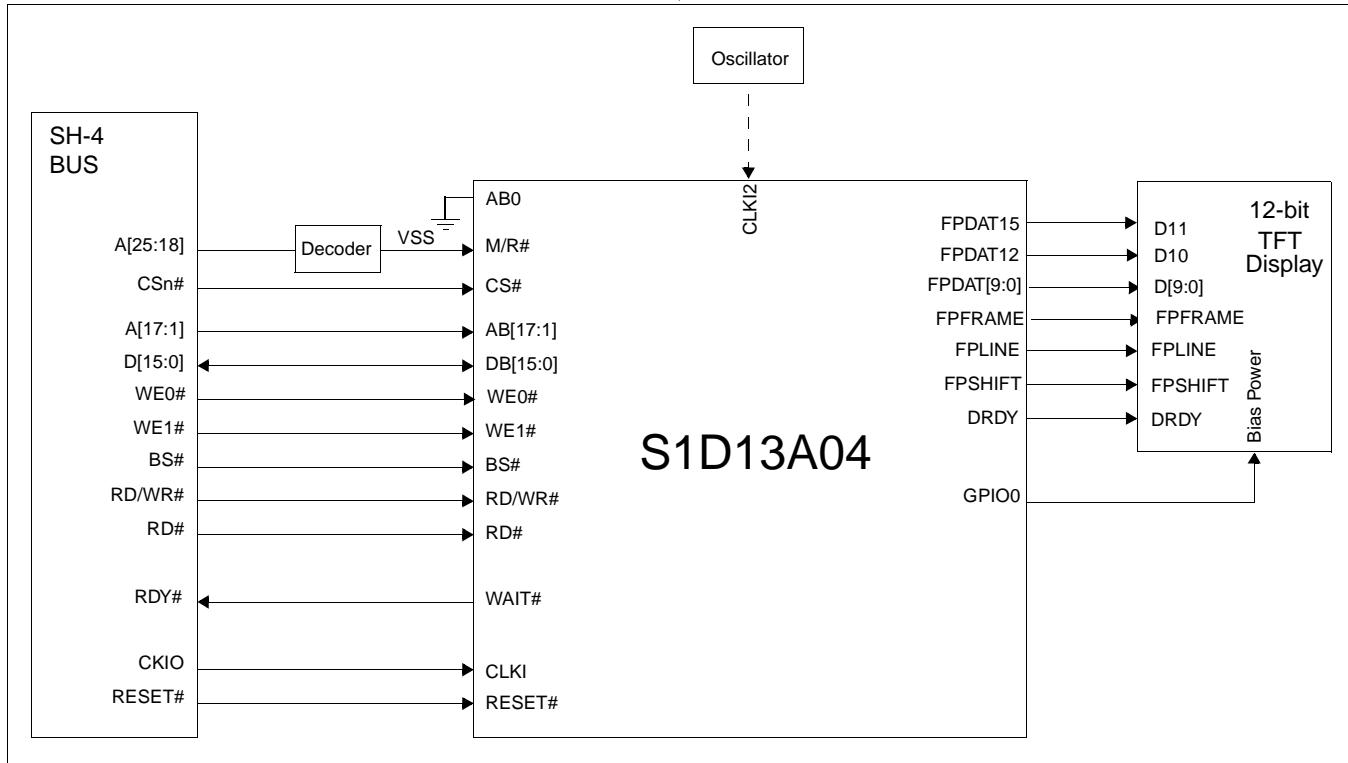


Figure 3-3: Typical System Diagram (Hitachi SH-4 Bus)

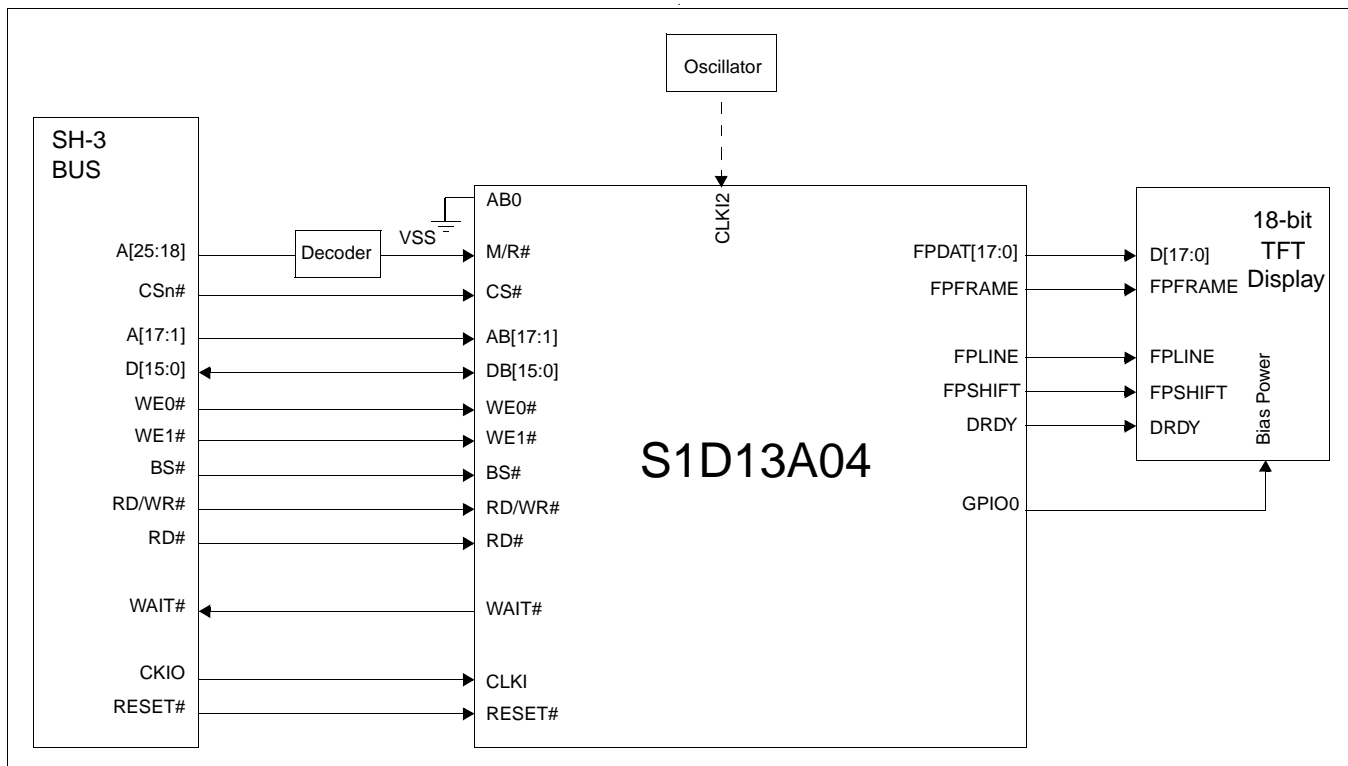


Figure 3-4: Typical System Diagram (Hitachi SH-3 Bus)

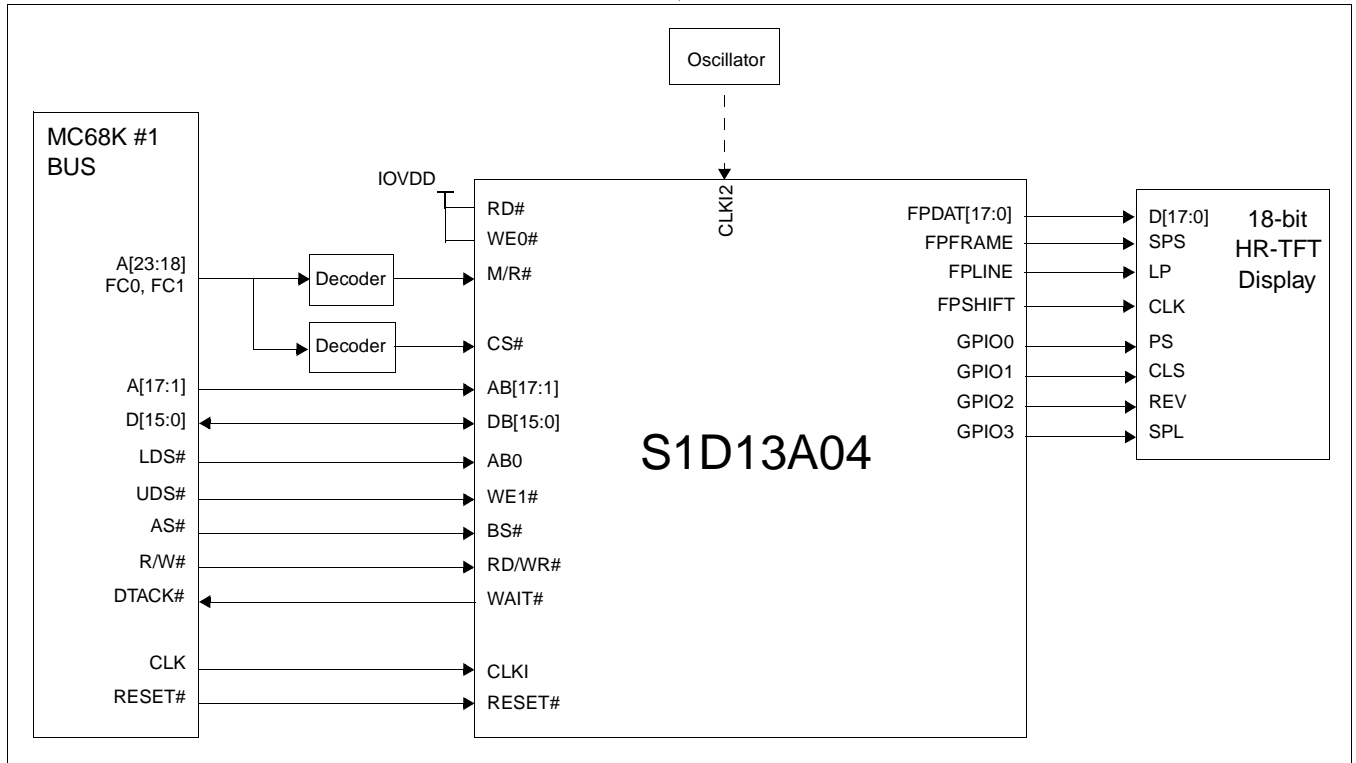


Figure 3-5: Typical System Diagram (MC68K #1, Motorola 16-Bit 68000)

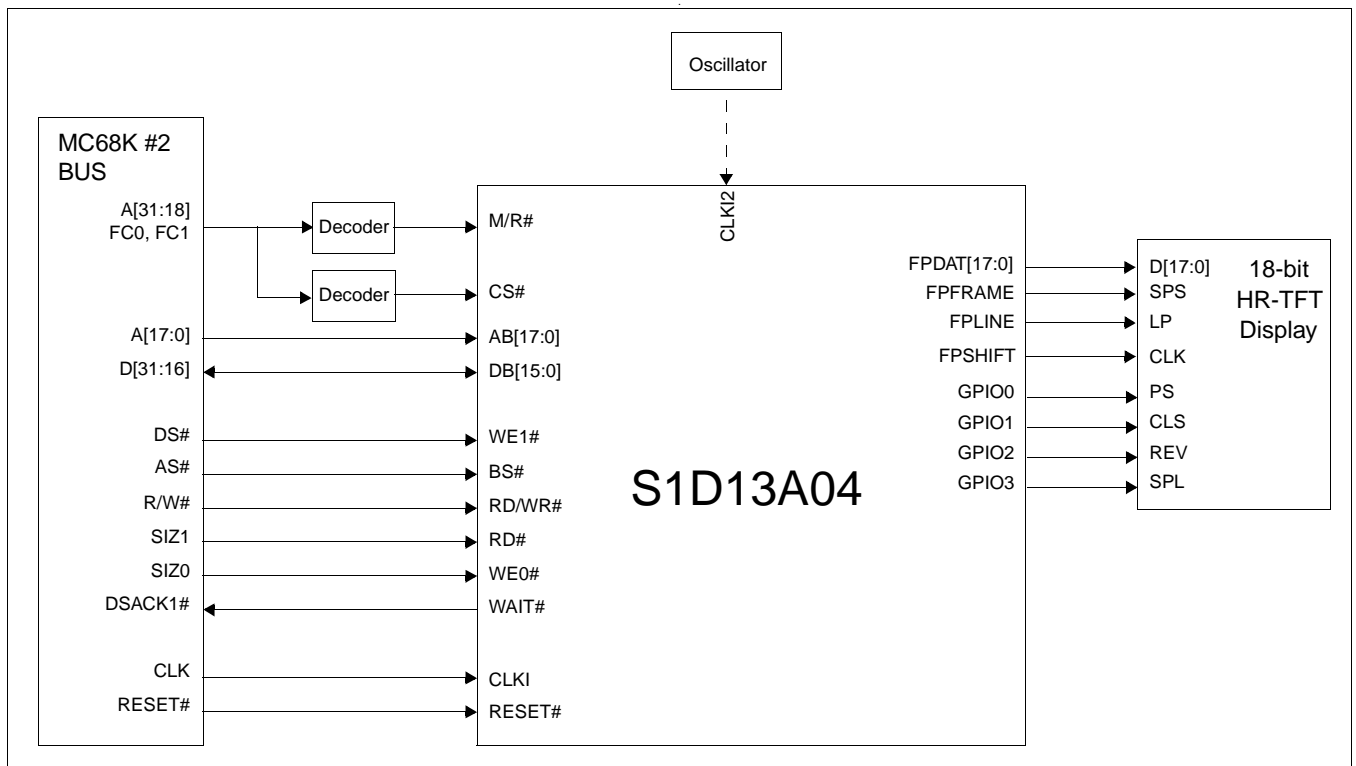


Figure 3-6: Typical System Diagram (MC68K #2, Motorola 32-Bit 68030)

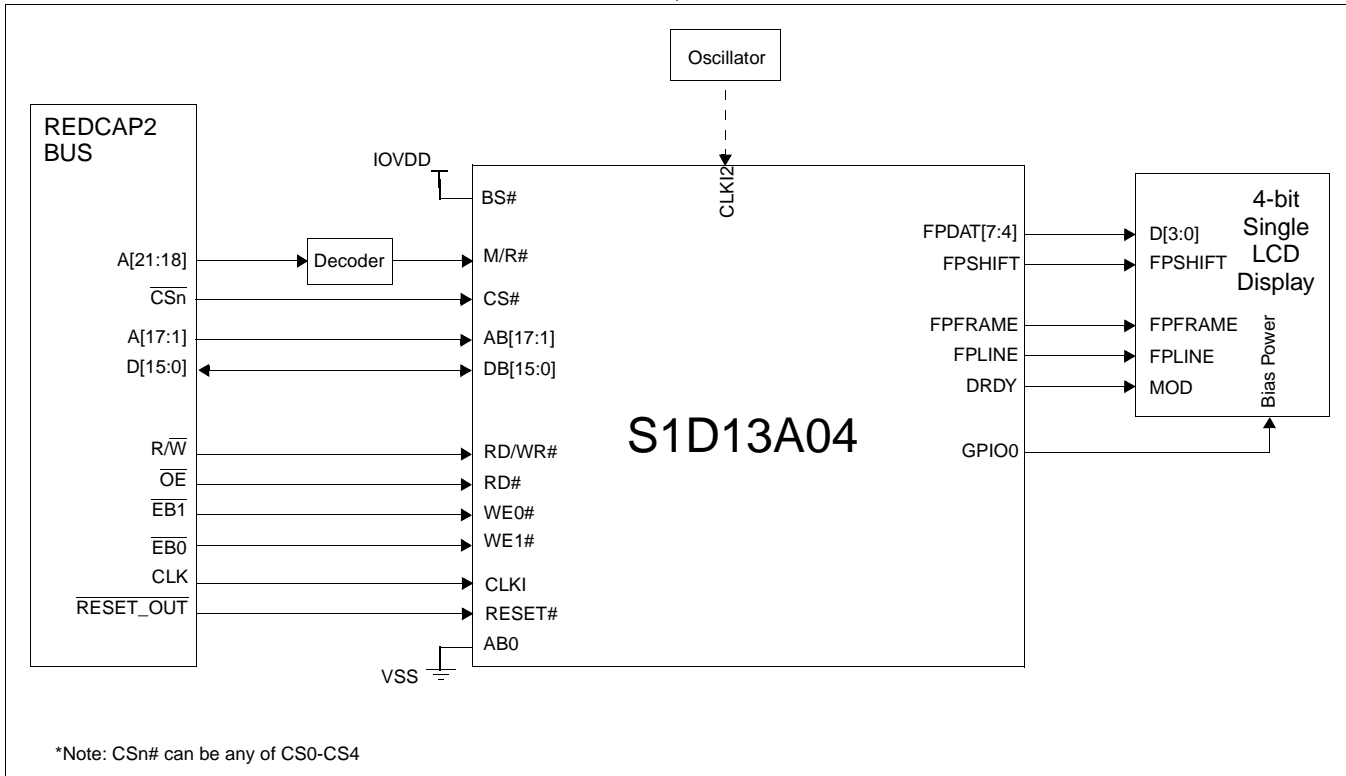


Figure 3-7: Typical System Diagram (Motorola REDCAP2 Bus)

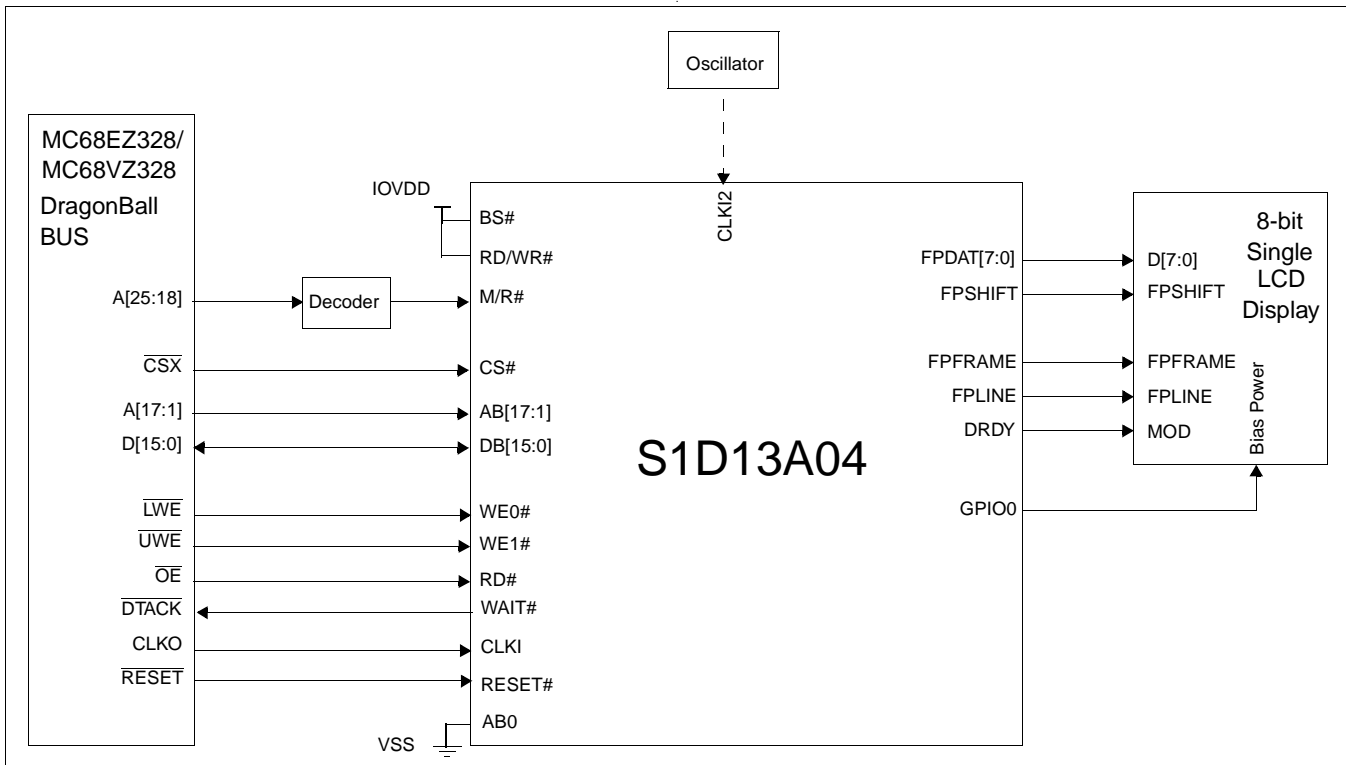


Figure 3-8: Typical System Diagram (Motorola MC68EZ328/MC68VZ328 "DragonBall" Bus)

3.2 USB Interface

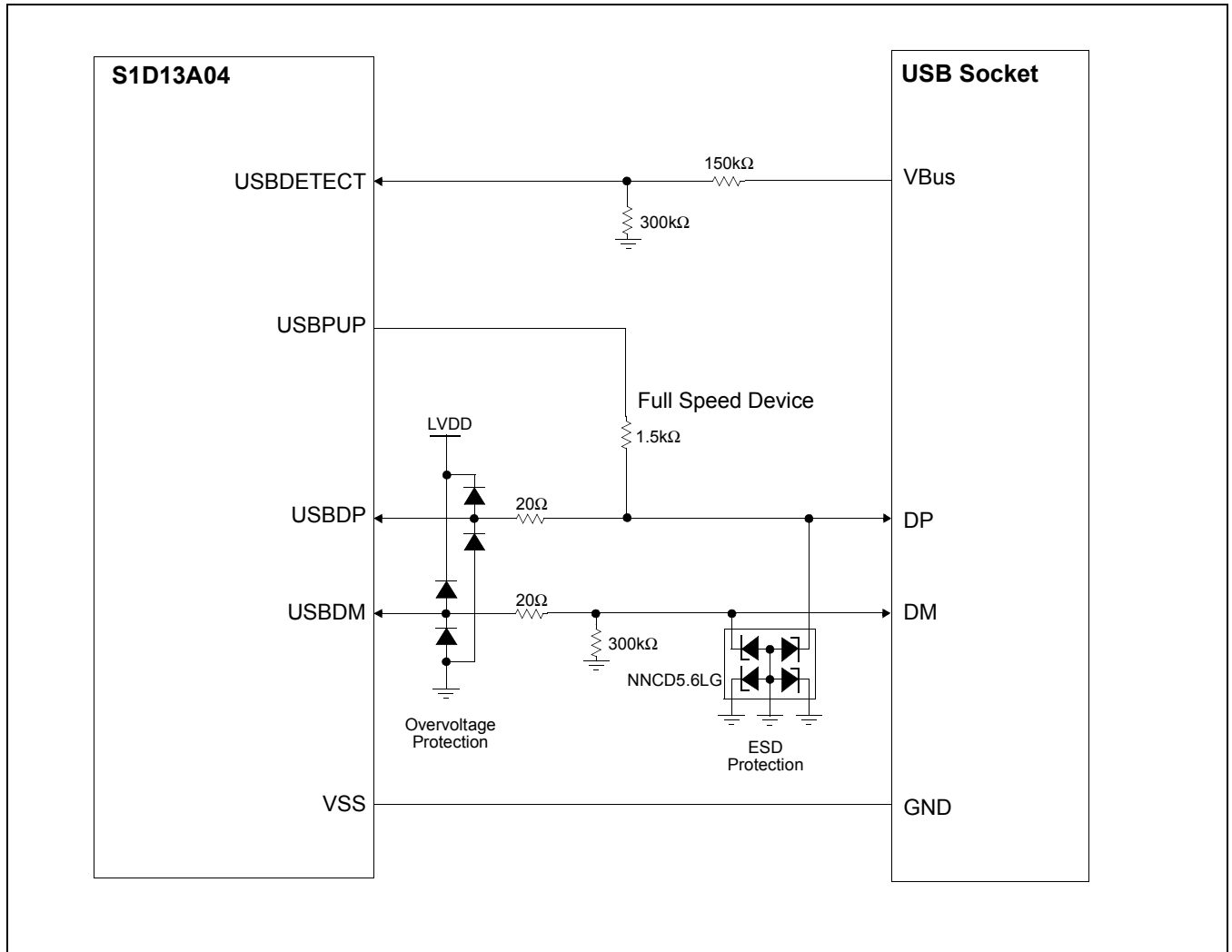


Figure 3-9: USB Typical Implementation

4 Pins

4.1 Pinout Diagram - PFBGA - 121-pin

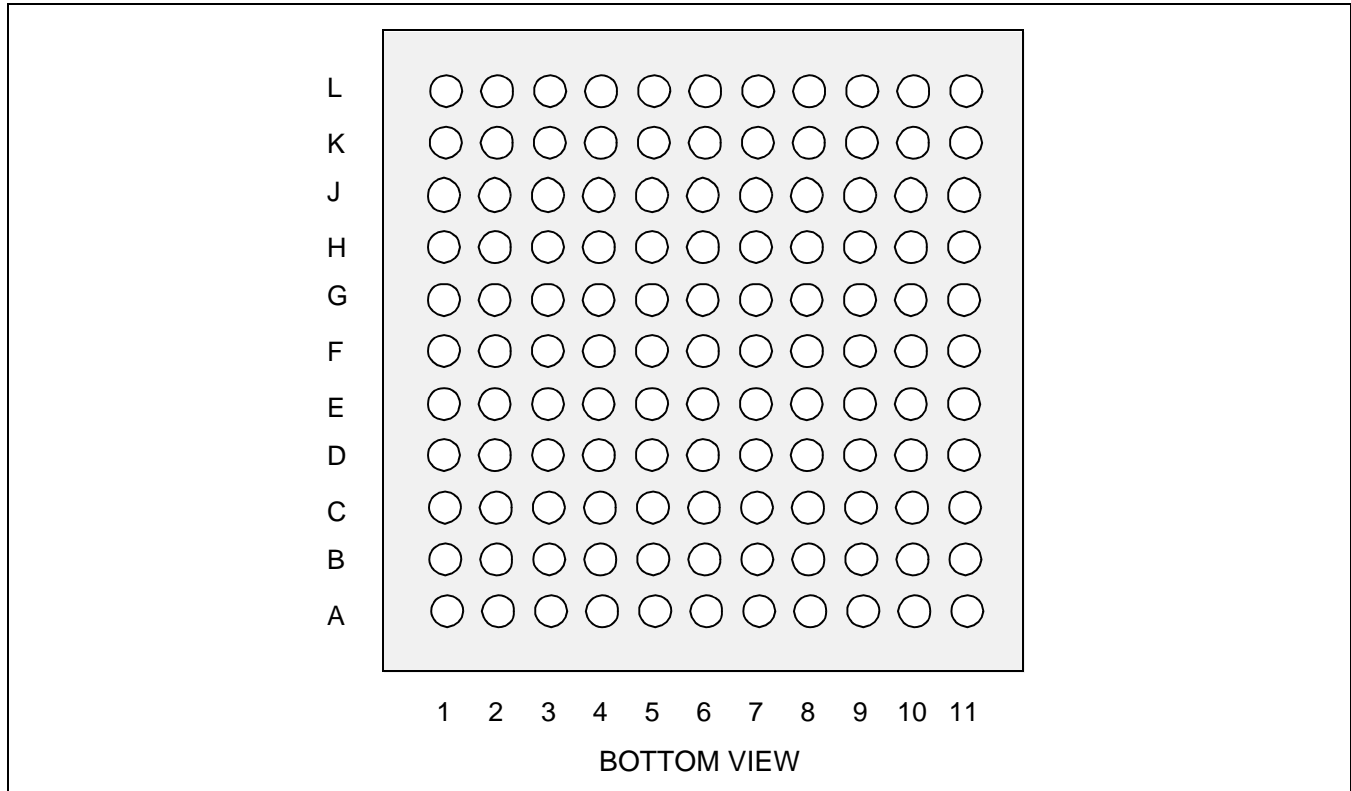


Figure 4-1: Pinout Diagram - PFBGA 121-pin

Table 4-1: PFBGA 121-pin Mapping

L	NC	IOVDD	DB7	DB3	DB0	GPIO7	GPIO3	GPIO0	IOVDD	COREVDD	NC
K	NC	VSS	DB8	DB4	DB1	GPIO6	GPIO2	IRQ	DRDY	VSS	NC
J	NC	DB9	DB6	DB5	DB2	NC	GPIO1	USBCLK	FPFRAME	COREVDD	NC
H	DB12	DB11	DB10	DB13	NC	IOVDD	GPIO4	NC	FPLINE	FPSHIFT	FPDAT0
G	WAIT#	DB15	DB14	IOVDD	VSS	GPIO5	FPDAT5	FPDAT1	FPDAT2	FPDAT3	FPDAT4
F	RESET#	VSS	RD/WR#	WE1#	CLKI	NC	FPDAT8	FPDAT6	VSS	FPDAT7	IOVDD
E	RD#	BS#	M/R#	CS#	WE0#	AB13	TESTEN	FPDAT9	FPDAT12	FPDAT11	FPDAT10
D	AB0	AB1	AB2	AB8	AB12	AB17	CNF3	FPDAT13	FPDAT16	FPDAT15	FPDAT14
C	NC	COREVDD	AB3	AB6	AB9	AB16	CNF2	CNF5	CNF6	FPDAT17	NC
B	NC	VSS	AB5	NC	AB10	AB14	CNF1	CNF4	CLKI2	VSS	NC
A	NC	COREVDD	AB4	AB7	AB11	AB15	CNF0	NC	PWMOUT	IOVDD	NC
	1	2	3	4	5	6	7	8	9	10	11

4.2 Pinout Diagram - TQFP15 - 128-pin

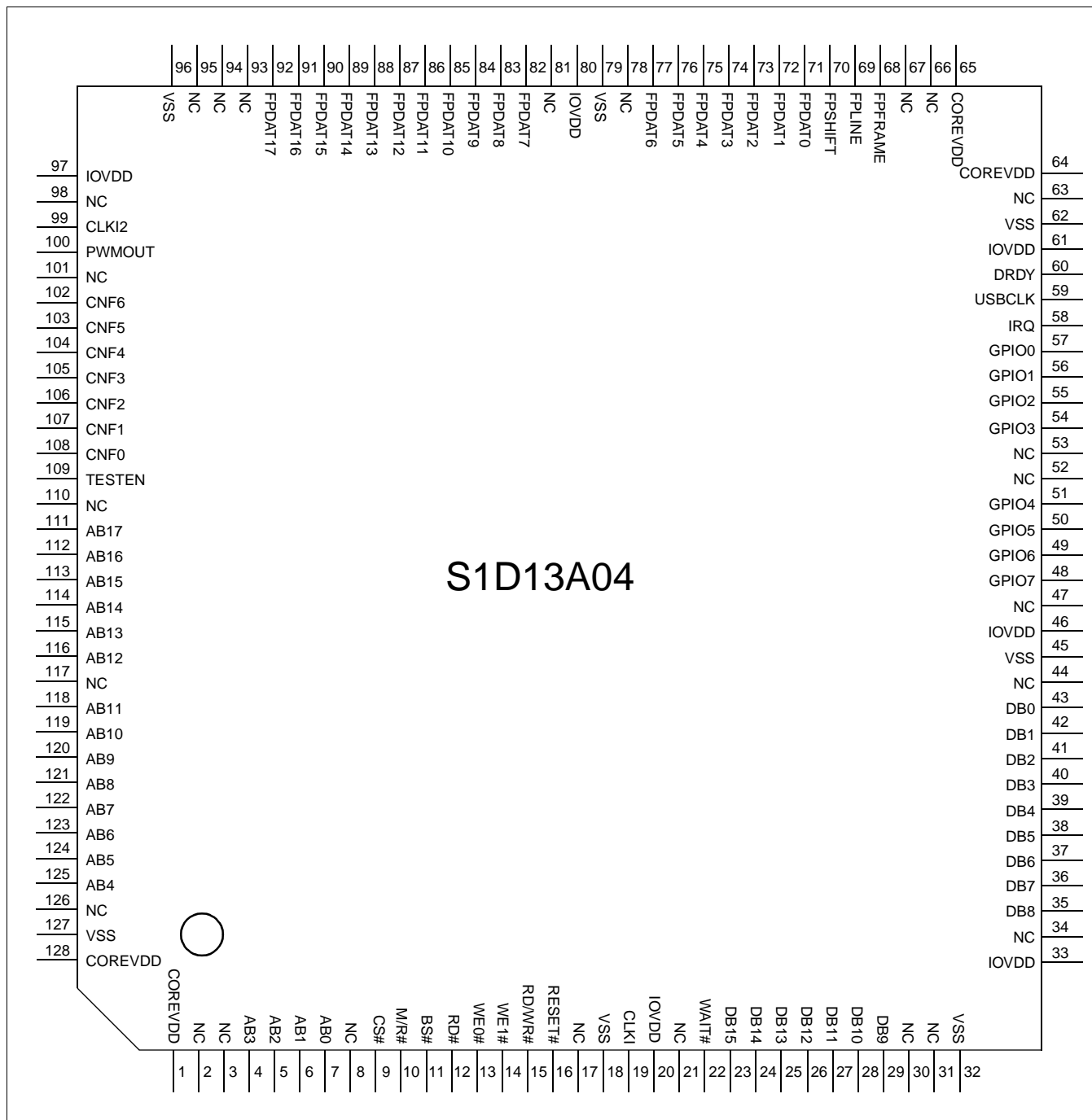


Figure 4-2: Pinout Diagram - TQFP15 128-pin

4.3 Pin Descriptions

Key:

I	=	Input
O	=	Output
IO	=	Bi-Directional (Input/Output)
P	=	Power pin
CI	=	CMOS input
LI	=	LVTTL input
LB2A	=	LVTTL IO buffer (6mA/-6mA@3.3V)
LB3P	=	Low noise LVTTL IO buffer (6mA/-6mA@3.3V)
LO3	=	Low noise LVTTL Output buffer (3mA/-3mA@3.3V)
LB3M	=	Low noise LVTTL IO buffer with input mask (3mA/-3mA@3.3V)
T1	=	Test mode control input with pull-down resistor (typical value of 50KΩ at 3.3V)
Hi-Z	=	High Impedance
CUS	=	Custom Cell Type

^a LVTTL is Low Voltage TTL (see Section 5, “D.C. Characteristics” on page 33).

4.3.1 Host Interface

Table 4-2: Host Interface Pin Descriptions

Pin Name	Type	PFBGA Pin #	TQFP15 Pin#	Cell	RESET# State	Description
AB0	I	D1	7	LI	—	<p>This input pin has multiple functions.</p> <ul style="list-style-type: none"> For Generic #1, this pin is not used and should be connected to VSS. For Generic #2, this pin inputs system address bit 0 (A0). For SH-3/SH-4, this pin is not used and should be connected to VSS. For MC68K #1, this pin inputs the lower data strobe (LDS#). For MC68K #2, this pin inputs system address bit 0 (A0). For REDCAP2, this pin is not used and should be connected to VSS. For DragonBall, this pin is not used and should be connected to VSS. <p>See Table 4-8: “Host Bus Interface Pin Mapping,” on page 31 for summary.</p>
AB[17:1]	I	C6,A6,B6, E6,D5,A5, B5,C5,D4 ,A4,C4,B3 ,A3,C3,D 3,D2,D6	4-6, 111-116, 118-125	CI	—	System address bus bits 17-1.

Table 4-2: Host Interface Pin Descriptions

Pin Name	Type	PFBGA Pin #	TQFP15 Pin#	Cell	RESET# State	Description
DB[15:0]	IO	L5,K5,J5, L4,K4,J4, J3,L3,K3, J2,H3,H2, H1,H4,G3 ,G2	23-29, 35-43	LB2A	Hi-Z	<p>Input data from the system data bus.</p> <ul style="list-style-type: none"> • For Generic #1, these pins are connected to D[15:0]. • For Generic #2, these pins are connected to D[15:0]. • For SH-3/SH-4, these pins are connected to D[15:0]. • For MC68K #1, these pins are connected to D[15:0]. • For MC68K #2, these pins are connected to D[31:16] for a 32-bit device (e.g. MC68030) or D[15:0] for a 16-bit device (e.g. MC68340). • For REDCAP2, these pins are connected to D[15:0]. • For DragonBall, these pins are connected to D[15:0]. <p>See Table 4-8: “Host Bus Interface Pin Mapping,” on page 31 for summary.</p>
WE0#	I	E5	13	LI	—	<p>This input pin has multiple functions.</p> <ul style="list-style-type: none"> • For Generic #1, this pin inputs the write enable signal for the lower data byte (WE0#). • For Generic #2, this pin inputs the write enable signal (WE#). • For SH-3/SH-4, this pin inputs the write enable signal for data byte 0 (WE0#). • For MC68K #1, this pin must be tied to IO V_{DD}. • For MC68K #2, this pin inputs the bus size bit 0 (SIZ0). • For REDCAP2, this pin inputs the byte enable signal for the D[7:0] data byte (EB1). • For DragonBall, this pin inputs the byte enable signal for the D[7:0] data byte (LWE). <p>See Table 4-8: “Host Bus Interface Pin Mapping,” on page 31 for summary.</p>
WE1#	I	F4	14	LI	—	<p>This input pin has multiple functions.</p> <ul style="list-style-type: none"> • For Generic #1, this pin inputs the write enable signal for the upper data byte (WE1#). • For Generic #2, this pin inputs the byte enable signal for the high data byte (BHE#). • For SH-3/SH-4, this pin inputs the write enable signal for data byte 1 (WE1#). • For MC68K #1, this pin inputs the upper data strobe (UDS#). • For MC68K #2, this pin inputs the data strobe (DS#). • For REDCAP2, this pin inputs the byte enable signal for the D[15:8] data byte (EB0). • For DragonBall, this pin inputs the byte enable signal for the D[15:8] data byte (UWE). <p>See Table 4-8: “Host Bus Interface Pin Mapping,” on page 31 for summary.</p>
CS#	I	E4	9	CI	—	<p>Chip select input. See Table 4-8: “Host Bus Interface Pin Mapping,” on page 31 for summary.</p>

Table 4-2: Host Interface Pin Descriptions

Pin Name	Type	PFBGA Pin #	TQFP15 Pin#	Cell	RESET# State	Description
M/R#	I	E3	10	LI	—	This input pin is used to select between the display buffer and register address spaces of the S1D13A04. M/R# is set high to access the display buffer and low to access the registers. See Table 4-8: "Host Bus Interface Pin Mapping," on page 31 for summary.
BS#	I	E2	11	LI	—	This input pin has multiple functions. <ul style="list-style-type: none"> • For Generic #1, this pin must be tied to IO V_{DD}. • For Generic #2, this pin must be tied to IO V_{DD}. • For SH-3/SH-4, this pin inputs the bus start signal (BS#). • For MC68K #1, this pin inputs the address strobe (AS#). • For MC68K #2, this pin inputs the address strobe (AS#). • For REDCAP2, this pin must be tied to IO V_{DD}. • For DragonBall, this pin must be tied to IO V_{DD}. See Table 4-8: "Host Bus Interface Pin Mapping," on page 31 for summary.
RD/WR#	I	F3	15	LI	—	This input pin has multiple functions. <ul style="list-style-type: none"> • For Generic #1, this pin inputs the read command for the upper data byte (RD1#). • For Generic #2, this pin must be tied to IO V_{DD}. • For SH-3/SH-4, this pin inputs the RD/WR# signal. The S1D13A04 needs this signal for early decode of the bus cycle. • For MC68K #1, this pin inputs the R/W# signal. • For MC68K #2, this pin inputs the R/W# signal. • For REDCAP2, this pin inputs the R/\overline{W} signal. • For DragonBall, this pin must be tied to IO V_{DD}. See Table 4-8: "Host Bus Interface Pin Mapping," on page 31 for summary.
RD#	I	E1	12	LI	—	This input pin has multiple functions. <ul style="list-style-type: none"> • For Generic #1, this pin inputs the read command for the lower data byte (RD0#). • For Generic #2, this pin inputs the read command (RD#). • For SH-3/SH-4, this pin inputs the read signal (RD#). • For MC68K #1, this pin must be tied to IO V_{DD}. • For MC68K #2, this pin inputs the bus size bit 1 (SIZ1). • For REDCAP2, this pin inputs the output enable (\overline{OE}). • For DragonBall, this pin inputs the output enable (\overline{OE}). See Table 4-8: "Host Bus Interface Pin Mapping," on page 31 for summary.

Table 4-2: Host Interface Pin Descriptions

Pin Name	Type	PFBGA Pin #	TQFP15 Pin#	Cell	RESET# State	Description
WAIT#	IO	G1	22	LB2A	Hi-Z	<p>During a data transfer, this output pin is driven active to force the system to insert wait states. It is driven inactive to indicate the completion of a data transfer. WAIT# is released to the high impedance state after the data transfer is complete. Its active polarity is configurable. See Table 4-7: "Summary of Power-On/Reset Options," on page 30.</p> <ul style="list-style-type: none"> • For Generic #1, this pin outputs the wait signal (WAIT#). • For Generic #2, this pin outputs the wait signal (WAIT#). • For SH-3 mode, this pin outputs the wait request signal (WAIT#). • For SH-4 mode, this pin outputs the device ready signal (RDY#). • For MC68K #1, this pin outputs the data transfer acknowledge signal (DTACK#). • For MC68K #2, this pin outputs the data transfer and size acknowledge bit 1 (DSACK1#). • For REDCAP2, this pin is unused (Hi-Z). • For DragonBall, this pin outputs the data transfer acknowledge signal (DTACK). <p>See Table 4-8: "Host Bus Interface Pin Mapping," on page 31 for summary.</p> <p>Note: This pin should be tied to the inactive voltage level as selected by CNF5, using a pull-up or pull-down resistor. If CNF5 = 1, the WAIT# pin should be tied low using a pull-down resistor. If CNF5 = 0, the WAIT# pin should be tied high using a pull-up resistor. If WAIT# is not used, this pin should be tied either high or low using a pull-up or pull-down resistor.</p>
RESET#	I	F1	16	LI	—	Active low input to set all internal registers to the default state and to force all signals to their inactive states.

4.3.2 LCD Interface

Table 4-3: LCD Interface Pin Descriptions

Pin Name	Type	PFBGA Pin#	TQFP15 Pin#	Cell	RESET# State	Description
FPDAT[17:0]	O	C10,D9,D10,D,11,D8,E9,E10,E11,E8,F7,F10,F8,G7,G11,G10,G9,G8,H11	71-77, 82-92	LB3P	0	Panel Data bits 17-0.
FPFRAME	O	J9	68	LB3P	0	This output pin has multiple functions. <ul style="list-style-type: none"> • Frame Pulse • SPS for 'Direct' HR-TFT See Table 4.6 "LCD Interface Pin Mapping," on page 32 for summary.
FPLINE	O	H9	69	LB3P	0	This output pin has multiple functions. <ul style="list-style-type: none"> • Line Pulse • LP for 'Direct' HR-TFT See Table 4.6 "LCD Interface Pin Mapping," on page 32 for summary.
FPSHIFT	O	H10	70	LB3P	0	This output pin has multiple functions. <ul style="list-style-type: none"> • Shift Clock • CLK for 'Direct' HR-TFT See Table 4.6 "LCD Interface Pin Mapping," on page 32 for summary.
DRDY	O	K9	60	LO3	0	This output pin has multiple functions. <ul style="list-style-type: none"> • Display enable (DRDY) for TFT panels • 2nd shift clock (FPSHIFT2) for passive LCD with Format 1 interface • LCD backplane bias signal (MOD) for all other LCD panels • General Purpose Output See Table 4.6 "LCD Interface Pin Mapping," on page 32 for summary.
GPIO0	IO	L8	57	LB3M	—	This pin has multiple functions. <ul style="list-style-type: none"> • PS for 'Direct' HR-TFT • General purpose IO pin 0 (GPIO0) GPIO0 defaults to a Hi-Z state during every RESET and defaults to an input after every RESET. When this pin is used for HR-TFT, it must be configured as an output using REG[64h]. Otherwise, it must either be configured as an output or be pulled high or low externally to avoid unnecessary current drain. <p>See Table 4.6 "LCD Interface Pin Mapping," on page 32 for summary.</p>

Table 4-3: LCD Interface Pin Descriptions

Pin Name	Type	PFBGA Pin#	TQFP15 Pin#	Cell	RESET# State	Description
GPIO1	IO	J7	56	LB3M	—	<p>This pin has multiple functions.</p> <ul style="list-style-type: none"> • CLS for 'Direct' HR-TFT • General purpose IO pin 1 (GPIO1) <p>GPIO1 defaults to a Hi-Z state during every RESET and defaults to an input after every RESET. When this pin is used for HR-TFT, it must be configured as an output using REG[64h]. Otherwise, it must either be configured as an output or be pulled high or low externally to avoid unnecessary current drain.</p> <p>See Table 4.6 "LCD Interface Pin Mapping," on page 32 for summary.</p>
GPIO2	IO	K7	55	LB3M	—	<p>This pin has multiple functions.</p> <ul style="list-style-type: none"> • REV for 'Direct' HR-TFT • General purpose IO pin 2 (GPIO2) <p>GPIO2 defaults to a Hi-Z state during every RESET and defaults to an input after every RESET. When this pin is used for HR-TFT, it must be configured as an output using REG[64h]. Otherwise, it must either be configured as an output or be pulled high or low externally to avoid unnecessary current drain.</p> <p>See Table 4.6 "LCD Interface Pin Mapping," on page 32 for summary.</p>
GPIO3	IO	L7	54	LB3M	—	<p>This pin has multiple functions.</p> <ul style="list-style-type: none"> • SPL for 'Direct' HR-TFT • General purpose IO pin 3 (GPIO3) <p>GPIO3 defaults to a Hi-Z state during every RESET and defaults to an input after every RESET. When this pin is used for HR-TFT, it must be configured as an output using REG[64h]. Otherwise, it must either be configured as an output or be pulled high or low externally to avoid unnecessary current drain.</p> <p>See Table 4.6 "LCD Interface Pin Mapping," on page 32 for summary.</p>
GPIO4	IO	H7	51	LB3M	—	<p>This pin has multiple functions.</p> <ul style="list-style-type: none"> • USBPUP • General purpose IO pin 4 (GPIO4) <p>GPIO4 defaults to a Hi-Z state during every RESET and defaults to an input after every RESET. When this pin is not used for USB, it must either be configured as an output using REG[64h] or be pulled high or low externally to avoid unnecessary current drain.</p>

Table 4-3: LCD Interface Pin Descriptions

Pin Name	Type	PFBGA Pin#	TQFP15 Pin#	Cell	RESET# State	Description
GPIO5	IO	G6	50	LB3M	—	<p>This pin has multiple functions.</p> <ul style="list-style-type: none"> • USBDETECT • General purpose IO pin 5 (GPIO5) <p>GPIO5 defaults to a Hi-Z state during every RESET and defaults to an input after every RESET. When this pin is not used for USB, it must either be configured as an output using REG[64h] or be pulled high or low externally to avoid unnecessary current drain.</p>
GPIO6	IO	K6	49	CUS	—	<p>This pin has multiple functions.</p> <ul style="list-style-type: none"> • USBDM • General purpose IO pin 6 (GPIO6) <p>GPIO6 defaults to a Hi-Z state during every RESET and defaults to an input after every RESET. When this pin is not used for USB, it must either be configured as an output using REG[64h] or be pulled high or low externally to avoid unnecessary current drain.</p>
GPIO7	IO	L6	48	CUS	—	<p>This pin has multiple functions.</p> <ul style="list-style-type: none"> • USBDP • General purpose IO pin 7 <p>GPIO7 defaults to a Hi-Z state during every RESET and defaults to an input after every RESET. When this pin is not used for USB, it must either be configured as an output using REG[64h] or be pulled high or low externally to avoid unnecessary current drain.</p>
IRQ	O	K8	58	LO3	0	<p>This output pin is the IRQ pin for USB. When IRQ is activated, an active high pulse is generated and stays high until the IRQ is serviced by software at REG[404Ah] or REG[404Ch].</p>
PWMOUT	O	A9	100	LO3	0	<p>This pin has multiple functions.</p> <ul style="list-style-type: none"> • PWM Clock output • General purpose output

4.3.3 Clock Input

Table 4-4: Clock Input Pin Descriptions

Pin Name	Type	PFBGA Pin#	TQFP15 Pin#	Cell	RESET# State	Description
CLKI	I	F5	19	CI	—	Typically used as input clock source for bus clock and memory clock
CLKI2	I	B9	99	CI	—	Typically used as input clock source for pixel clock
USBCLK	I	J8	59	LI	—	Typically used as input clock source for USB

4.3.4 Miscellaneous

Table 4-5: Miscellaneous Pin Descriptions

Pin Name	Type	PFBGA Pin#	TQFP15 Pin#	Cell	RESET# State	Description
CNF[6:0]	I	C9,C8,B8, D7,C7,B7, A7	102-108	CI	—	These inputs are used to configure the S1D13A04 - see Table 4-7: "Summary of Power-On/Reset Options," on page 30. Note: These pins are used for configuration of the S1D13A04 and must be connected directly to IO V_{DD} or V_{SS}.
TESTEN	I	E7	109	T1	—	Test Enable input used for production test only (has type 1 pull-down resistor with a typical value of 50K Ω at 3.3V). Note: This pin must not be connected.

4.3.5 Power And Ground

Table 4-6: Power And Ground Pin Descriptions

Pin Name	Type	PFBGA Pin#	TQFP15 Pin#	Cell	RESET# State	Description
IOVDD	P	L2,G4, H6,L9, A10,F11	20, 33, 46, 61, 80, 97	P	—	6 IO V _{DD} pins.
COREVDD	P	A2,C2, L10,J10	1, 64-65, 128	P	—	2 double-bonded Core V _{DD} pins on TQFP package. 4 Core V _{DD} pins on PFBGA package.
VSS	P	B2,F2, K2,G5, F9,B10,K 10	18, 32, 45, 62, 79, 96, 127	P	—	7 V _{SS} pins.

4.4 Summary of Configuration Options

These pins are used for configuration of the S1D13A04 and must be connected directly to IOV_{DD} or V_{SS}. The state of CNF[6:0] are latched on the rising edge of RESET#. Changing state at any other time has no effect.

Table 4-7: Summary of Power-On/Reset Options

S1D13A04 Configuration Input	Power-On/Reset State				
	1 (connected to IO V _{DD})	0 (connected to V _{SS})			
CNF4,CNF[2:0]	Select host bus interface as follows:				
	CNF4	CNF2	CNF1	CNF0	Host Bus
	1	0	0	0	SH-4/SH-3 interface, Big Endian
	0	0	0	0	SH-4/SH-3 interface, Little Endian
	1	0	0	1	MC68K #1, Big Endian
	0	0	0	1	Reserved
	1	0	1	0	MC68K #2, Big Endian
	0	0	1	0	Reserved
	1	0	1	1	Generic #1, Big Endian
	0	0	1	1	Generic #1, Little Endian
	1	1	0	0	Reserved
	0	1	0	0	Generic #2, Little Endian
	1	1	0	1	REDCAP2, Big Endian
	0	1	0	1	Reserved
1	1	1	0	DragonBall (MC68EZ328/MC68VZ328), Big Endian	
0	1	1	0	Reserved	
X	1	1	1	Reserved	
CNF3	Reserved. Must be set to 1.				
CNF5 (see note)	WAIT# is active high		WAIT# is active low		
CNF6	CLKI to BCLK divide ratio 2:1		CLKI to BCLK divide ratio 1:1		

Note

If CNF5 = 1, the WAIT# pin should be tied low using a pull-down resistor. If CNF5 = 0, the WAIT# pin should be tied high using a pull-up resistor. If WAIT# is not used, this pin should be tied either high or low using a pull-up or pull-down resistor.

4.5 Host Bus Interface Pin Mapping

Table 4-8: Host Bus Interface Pin Mapping

S1D13A04 Pin Name	Generic #1	Generic #2	Hitachi SH-3 /SH-4	Motorola MC68K #1	Motorola MC68K #2	Motorola REDCAP2	Motorola MC68EZ328/ MC68VZ328 DragonBall
AB[17:1]	A[17:1]	A[17:1]	A[17:1]	A[17:1]	A[17:1]	A[17:1]	A[17:1]
AB0	A0 ¹	A0	A0 ¹	LDS#	A0	A0 ¹	A0 ¹
DB[15:0]	D[15:0]	D[15:0]	D[15:0]	D[15:0]	D[15:0] ²	D[15:0]	D[15:0]
CS#	External Decode		CSn#	External Decode		\overline{CSn}	\overline{CSX}
M/R#	External Decode						
CLKI	BUSCLK	BUSCLK	CKIO	CLK	CLK	CLK	CLKO
BS#	Connected to IO V _{DD}		BS#	AS#	AS#	Connected to IO V _{DD}	
RD/WR#	RD1#	Connected to IO V _{DD}	RD/WR#	R/W#	R/W#	R \overline{W}	Connected to IO V _{DD}
RD#	RD0#	RD#	RD#	Connected to IO V _{DD}	SIZ1	\overline{OE}	\overline{OE}
WE0#	WE0#	WE#	WE0#	Connected to IO V _{DD}	SIZ0	$\overline{EB1}$	\overline{LWE}
WE1#	WE1#	BHE#	WE1#	UDS#	DS#	$\overline{EB0}$	\overline{UWE}
WAIT#	WAIT#	WAIT#	WAIT#/ RDY#	DTACK#	DSACK1#	N/A	\overline{DTACK}
RESET#	RESET#	RESET#	RESET#	RESET#	RESET#	$\overline{RESET_OUT}$	\overline{RESET}

Note

¹ A0 for these busses is not used internally by the S1D13A04 and should be connected to V_{SS}.

² If the target MC68K bus is 32-bit, then these signals should be connected to D[31:16].

4.6 LCD Interface Pin Mapping

Table 4-9: LCD Interface Pin Mapping

Pin Name	Monochrome Passive Panel		Color Passive Panel				Color TFT Panel			
	Single		Single				Others			'Direct' HR-TFT ¹
	4-bit	8-bit	4-bit	Format 1 8-bit	Format 2 8-bit	16-Bit	9-bit	12-bit	18-bit	18-bit
FPFRAME	FPFRAME									SPS
FPLINE	FPLINE									LP
FPSHIFT	FPSHIFT									DCLK
DRDY	MOD			FPSHIFT2	MOD		DRDY			GPO ²
FPDAT0	driven 0	D0	driven 0	D0 (B5) ³	D0 (G3) ³	D0 (R6) ³	R2	R3	R5	R5
FPDAT1	driven 0	D1	driven 0	D1 (R5) ³	D1 (R3) ³	D1 (G5) ³	R1	R2	R4	R4
FPDAT2	driven 0	D2	driven 0	D2 (G4) ³	D2 (B2) ³	D2 (B4) ³	R0	R1	R3	R3
FPDAT3	driven 0	D3	driven 0	D3 (B3) ³	D3 (G2) ³	D3 (R4) ³	G2	G3	G5	G5
FPDAT4	D0	D4	D0 (R2) ³	D4 (R3) ³	D4 (R2) ³	D8 (B5) ³	G1	G2	G4	G4
FPDAT5	D1	D5	D1 (B1) ³	D5 (G2) ³	D5 (B1) ³	D9 (R5) ³	G0	G1	G3	G3
FPDAT6	D2	D6	D2 (G1) ³	D6 (B1) ³	D6 (G1) ³	D10 (G4) ³	B2	B3	B5	B5
FPDAT7	D3	D7	D3 (R1) ³	D7 (R1) ³	D7 (R1) ³	D11 (B3) ³	B1	B2	B4	B4
FPDAT8	driven 0	driven 0	driven 0	driven 0	driven 0	D4 (G3) ³	B0	B1	B3	B3
FPDAT9	driven 0	driven 0	driven 0	driven 0	driven 0	D5 (B2) ³	driven 0	R0	R2	R2
FPDAT10	driven 0	driven 0	driven 0	driven 0	driven 0	D6 (R2) ³	driven 0	driven 0	R1	R1
FPDAT11	driven 0	driven 0	driven 0	driven 0	driven 0	D7 (G1) ³	driven 0	driven 0	R0	R0
FPDAT12	driven 0	driven 0	driven 0	driven 0	driven 0	D12 (R3) ³	driven 0	G0	G2	G2
FPDAT13	driven 0	driven 0	driven 0	driven 0	driven 0	D13 (G2) ³	driven 0	driven 0	G1	G1
FPDAT14	driven 0	driven 0	driven 0	driven 0	driven 0	D14 (B1) ³	driven 0	driven 0	G0	G0
FPDAT15	driven 0	driven 0	driven 0	driven 0	driven 0	D15 (R1) ³	driven 0	B0	B2	B2
FPDAT16	driven 0	driven 0	driven 0	driven 0	driven 0	driven 0	driven 0	driven 0	B1	B1
FPDAT17	driven 0	driven 0	driven 0	driven 0	driven 0	driven 0	driven 0	driven 0	B0	B0
GPIO0	GPIO0									PS
GPIO1	GPIO1									CLS
GPIO2	GPIO2									REV
GPIO3	GPIO3									SPL

Note

- ¹ GPIO pins default to inputs at reset and require special configuration using REG[64h] when the 'Direct' HR-TFT interface is desired.
- ² When the 'Direct' HR-TFT interface is selected (REG[0Ch] bits 1-0 = 10), DRDY becomes a general purpose output (GPO) controllable using the 'Direct' HR-TFT LCD Interface GPO Control bit (REG[14h] bit 0). This GPO can be used to control the HR-TFT MOD signal if required. For further information, see the bit description for REG[14h] bit 0.
- ³ These pin mappings use signal names commonly used for each panel type, however signal names may differ between panel manufacturers. The values shown in brackets represent the color components as mapped to the corresponding FPDATxx signals at the first valid edge of FPSHIFT. For further FPDATxx to LCD interface mapping, see Section 6.4, "Display Interface" on page 56.

5 D.C. Characteristics

Note

When applying Supply Voltages to the S1D13A04, Core V_{DD} must be applied to the chip before, or simultaneously with IO V_{DD} , or damage to the chip may result.

Table 5-1: Absolute Maximum Ratings

Symbol	Parameter	Rating	Units
Core V_{DD}	Supply Voltage	$V_{SS} - 0.3$ to 3.0	V
IO V_{DD}	Supply Voltage	$V_{SS} - 0.3$ to 4.0	V
V_{IN}	Input Voltage	$V_{SS} - 0.3$ to IO $V_{DD} + 0.5$	V
V_{OUT}	Output Voltage	$V_{SS} - 0.3$ to IO $V_{DD} + 0.5$	V
T_{STG}	Storage Temperature	-65 to 150	°C
T_{SOL}	Solder Temperature/Time	260 for 10 sec. max at lead	°C

Table 5-2: Recommended Operating Conditions

Symbol	Parameter	Condition	Min	Typ	Max	Units
Core V_{DD}	Supply Voltage	$V_{SS} = 0$ V	1.8 (note 1)	2.0 (note 1)	2.2 (note 1)	V
		$V_{SS} = 0$ V	2.25	2.5	2.75	V
IO V_{DD}	Supply Voltage	$V_{SS} = 0$ V	3.0	3.3	3.6	V
V_{IN}	Input Voltage		V_{SS}		IO V_{DD}	V
			V_{SS}		CORE V_{DD}	
T_{OPR}	Operating Temperature		-40	25	85	°C

- When Core V_{DD} is $2.0V \pm 10\%$, the MCLK must be less than or equal to 30MHz ($MCLK \leq 30MHz$)

Table 5-3: Electrical Characteristics for $V_{DD} = 3.3V$ typical

Symbol	Parameter	Condition	Min	Typ	Max	Units
I_{DDs}	Quiescent Current	Quiescent Conditions			170	μA
I_{IZ}	Input Leakage Current		-1		1	μA
I_{OZ}	Output Leakage Current		-1		1	μA
V_{OH}	High Level Output Voltage	$V_{DD} = \text{min.}$ $I_{OH} = -3mA$ (Type 1) $-6mA$ (Type 2)	$V_{DD} - 0.4$			V
V_{OL}	Low Level Output Voltage	$V_{DD} = \text{min.}$ $I_{OL} = 3mA$ (Type 1) $6mA$ (Type 2)			0.4	V
V_{IH}	High Level Input Voltage	LVTTL Level, $V_{DD} = \text{max}$	2.0			V
V_{IL}	Low Level Input Voltage	LVTTL Level, $V_{DD} = \text{min}$			0.8	V
R_{PD}	Pull Down Resistance	$V_{IN} = V_{DD}$	20	50	120	$k\Omega$
C_I	Input Pin Capacitance				10	pF
C_O	Output Pin Capacitance				10	pF
C_{IO}	Bi-Directional Pin Capacitance				10	pF

6 A.C. Characteristics

Conditions: IO $V_{DD} = 3.3V \pm 10\%$
 $T_A = -40^\circ C$ to $85^\circ C$
 T_{rise} and T_{fall} for all inputs must be ≤ 5 nsec (10% ~ 90%)
 $C_L = 50pF$ (Bus/MPU Interface)
 $C_L = 0pF$ (LCD Panel Interface)

6.1 Clock Timing

6.1.1 Input Clocks

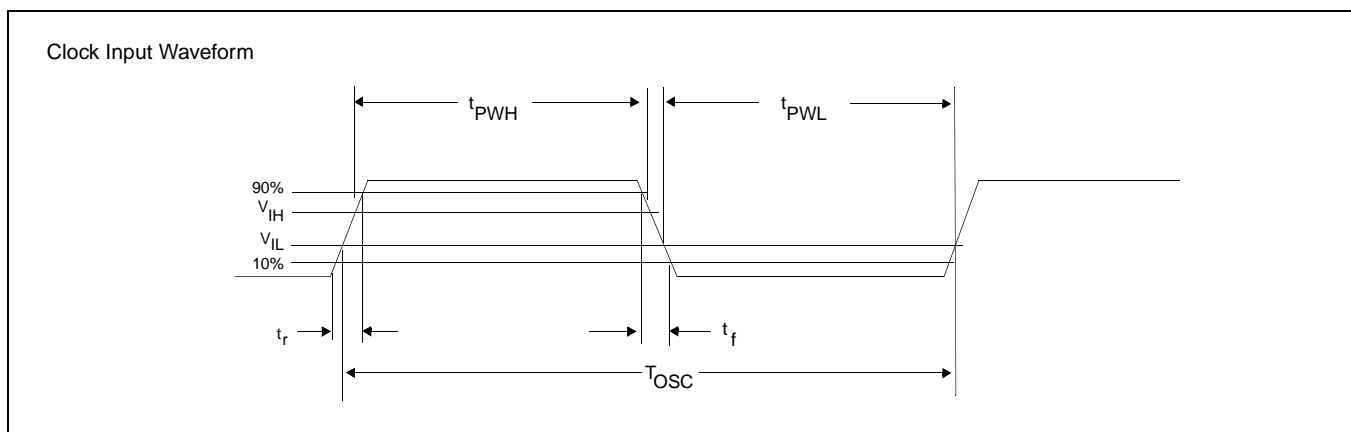


Figure 6-1: Clock Input Requirements

Table 6-1: Clock Input Requirements for CLKI when CLKI to BCLK divide > 1

Symbol	Parameter	Min	Max	Units
f_{osc}	Input Clock Frequency (CLKI)		100	MHz
T_{osc}	Input Clock period (CLKI)	$1/f_{osc}$		ns
t_{pwh}	Input Clock Pulse Width High (CLKI)	4.5		ns
t_{pwl}	Input Clock Pulse Width Low (CLKI)	4.5		ns
t_f	Input Clock Fall Time (10% - 90%)		5	ns
t_r	Input Clock Rise Time (10% - 90%)		5	ns

Note

Maximum internal requirements for clocks derived from CLKI must be considered when determining the frequency of CLKI. See Section 6.1.2, "Internal Clocks" on page 35 for internal clock requirements.

Table 6-2: Clock Input Requirements for CLKI when CLKI to BCLK divide = 1

Symbol	Parameter	Min	Max	Units
f_{OSC}	Input Clock Frequency (CLKI)		66	MHz
T_{OSC}	Input Clock period (CLKI)	$1/f_{OSC}$		ns
t_{PWH}	Input Clock Pulse Width High (CLKI)	3		ns
t_{PWL}	Input Clock Pulse Width Low (CLKI)	3		ns
t_f	Input Clock Fall Time (10% - 90%)		5	ns
t_r	Input Clock Rise Time (10% - 90%)		5	ns

Note

Maximum internal requirements for clocks derived from CLKI must be considered when determining the frequency of CLKI. See Section 6.1.2, “Internal Clocks” on page 35 for internal clock requirements.

Table 6-3: Clock Input Requirements for CLKI2

Symbol	Parameter	Min	Max	Units
f_{OSC}	Input Clock Frequency (CLKI2)		66	MHz
T_{OSC}	Input Clock period (CLKI2)	$1/f_{OSC}$		ns
t_{PWH}	Input Clock Pulse Width High (CLKI2)	3		ns
t_{PWL}	Input Clock Pulse Width Low (CLKI2)	3		ns
t_f	Input Clock Fall Time (10% - 90%)		5	ns
t_r	Input Clock Rise Time (10% - 90%)		5	ns

Note

Maximum internal requirements for clocks derived from CLKI2 must be considered when determining the frequency of CLKI2. See Section 6.1.2, “Internal Clocks” on page 35 for internal clock requirements.

6.1.2 Internal Clocks

Table 6-4: Internal Clock Requirements

Symbol	Parameter	Min	Max	Units
f_{BCLK}	Bus Clock frequency		66	MHz
f_{MCLK}	Memory Clock frequency		50 (note 1,note 2)	MHz
f_{PCLK}	Pixel Clock frequency		50	MHz
f_{PWMCLK}	PWM Clock frequency		66	MHz

1. When COREVDD = 2.0V \pm 10% f_{MCLK} max = 30MHz.
2. MCLK is derived from BCLK, therefore when BCLK is greater than 50MHz, MCLK must be divided using REG[04h] bits 5-4.

Note

For further information on internal clocks, refer to Section 7, “Clocks” on page 84.

6.2 CPU Interface Timing

6.2.1 Generic #1 Interface Timing (e.g. Epson EOC33)

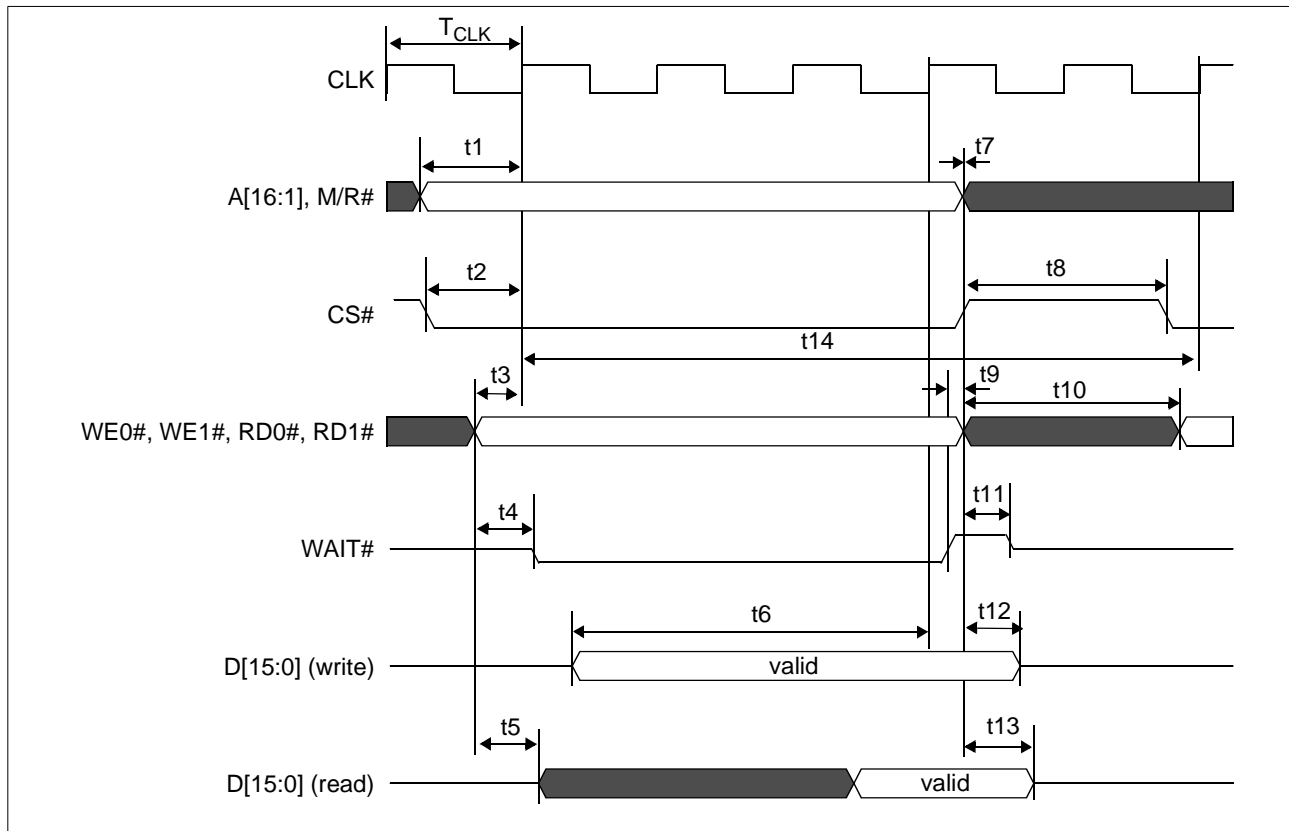


Figure 6-2: Generic #1 Interface Timing

Table 6-5: Generic #1 Interface Timing

Symbol	Parameter	Min	Max	Unit
f_{CLK}	Bus clock frequency		50	MHz
T_{CLK}	Bus clock period	$1/f_{CLK}$		ns
t1	A[16:1], M/R# setup to first CLK rising edge where CS# = 0 and either RD0#, RD1# = 0 or WE0#, WE1# = 0	9		ns
t2	CS# setup to CLK rising edge	9		ns
t3	RD0#, RD1#, WE0#, WE1# setup to CLK rising edge	1		ns
t4	RD0#, RD1# or WE0#, WE1# state change to WAIT# driven low	1	10	ns
t5	RD0#, RD1# falling edge to D[15:0] driven (read cycle)	2	10	ns
t6	D[15:0] setup to 4th rising CLK edge after CS#=0 and WE0#, WE1#=0	1		T_{CLK}
t7	A[16:1], M/R# and CS# hold from RD0#, RD1#, WE0#, WE1# rising edge	0		ns
t8	CS# deasserted to reasserted	0		ns
t9	WAIT# rising edge to RD0#, RD1#, WE0#, WE1# rising edge	0		ns
t10	WE0#, WE1#, RD0#, RD1# deasserted to reasserted	1		T_{CLK}
t11	Rising edge of either RD0#, RD1# or WE0#, WE1# to WAIT# high impedance		0.5	T_{CLK}
t12	D[15:0] hold from WE0#, WE1# rising edge (write cycle)	2		ns
t13	D[15:0] hold from RD0#, RD1# rising edge (read cycle)	1		ns
t14	Cycle Length	5		T_{CLK}

Table 6-6: Generic #1 Interface Truth Table for Little Endian

WE0#	WE1#	RD0#	RD1#	D[15:8]	D[7:0]	Comments
0	0	1	1	valid	valid	16-bit write
0	1	1	1	-	valid	8-bit write; data on low byte (even byte address ¹)
1	0	1	1	valid	-	8-bit write; data on high byte (odd byte address ¹)
1	1	0	0	valid	valid	16-bit read
1	1	0	1	-	valid	8-bit read; data on low byte (even byte address ¹)
1	1	1	0	valid	-	8-bit read; data on high byte (odd byte address ¹)

Table 6-7: Generic #1 Interface Truth Table for Big Endian

WE0#	WE1#	RD0#	RD1#	D[15:8]	D[7:0]	Comments
0	0	1	1	valid	valid	16-bit write
0	1	1	1	-	valid	8-bit write; data on low byte (odd byte address ¹)
1	0	1	1	valid	-	8-bit write; data on high byte (even byte address ¹)
1	1	0	0	valid	valid	16-bit read
1	1	0	1	-	valid	8-bit read; data on low byte (odd byte address ¹)
1	1	1	0	valid	-	8-bit read; data on high byte (even byte address ¹)

1. Because A0 is not used internally, all addresses are seen by the S1D13A04 as even addresses (16-bit word address aligned on even byte addresses).

6.2.2 Generic #2 Interface Timing (e.g. ISA)

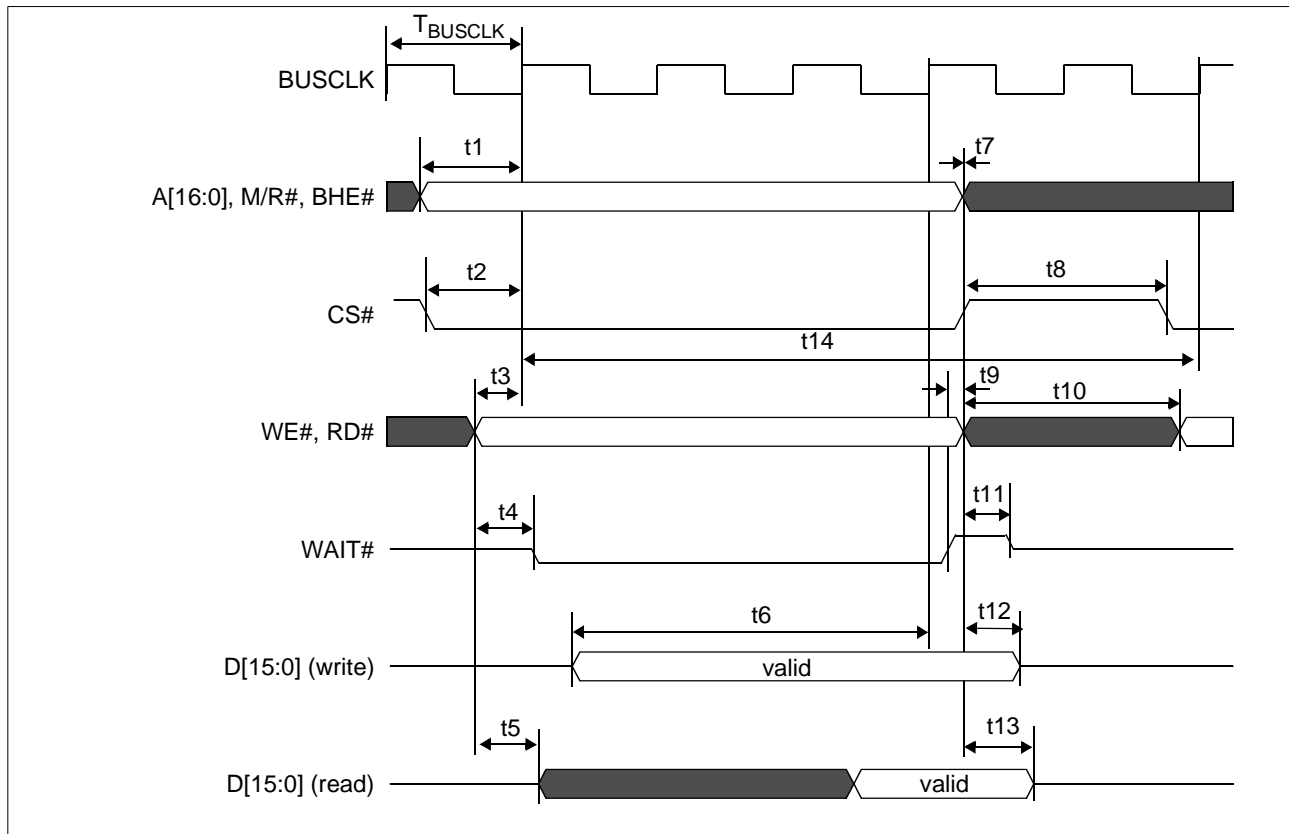


Figure 6-3: Generic #2 Interface Timing

Table 6-8: Generic #2 Interface Timing

Symbol	Parameter	Min	Max	Unit
f _{BUSCLK}	Bus clock frequency		50	MHz
T _{BUSCLK}	Bus clock period	1/f _{BUSCLK}		ns
t1	A[16:0], M/R#, BHE# setup to first BUSCLK rising edge where CS# = 0 and either RD# = 0 or WE# = 0	9		ns
t2	CS# setup to BUSCLK rising edge	9		ns
t3	RD#, WE# setup to BUSCLK rising edge	1		ns
t4	RD# or WE# state change to WAIT# driven low	1	10	ns
t5	RD# falling edge to D[15:0] driven (read cycle)	2	10	ns
t6	D[15:0] setup to 4th rising BUSCLK edge after CS#=0 and WE#=0	1		T _{BUSCLK}
t7	A[16:0], M/R#, BHE# and CS# hold from RD#, WE# rising edge	0		ns
t8	CS# deasserted to reasserted	0		ns
t9	WAIT# rising edge to RD#, WE# rising edge	0		ns
t10	WE#, RD# deasserted to reasserted	1		T _{BUSCLK}
t11	Rising edge of either RD# or WE# to WAIT# high impedance		0.5	T _{BUSCLK}
t12	D[15:0] hold from WE# rising edge (write cycle)	2		ns
t13	D[15:0] hold from RD# rising edge (read cycle)	1		ns
t14	Cycle Length	6		T _{BUSCLK}

Table 6-9: Generic #2 Interface Truth Table for Little Endian

WE#	RD#	BHE#	A0	D[15:8]	D[7:0]	Comments
0	1	0	0	valid	valid	16-bit write
0	1	1	0	-	valid	8-bit write at even address
0	1	0	1	valid	-	8-bit write at odd address
1	0	0	0	valid	valid	16-bit read
1	0	1	0	-	valid	8-bit read at even address
1	0	0	1	valid	-	8-bit read at odd address

Note

Generic #2 interface only supports Little Endian mode.

6.2.3 Hitachi SH-3 Interface Timing

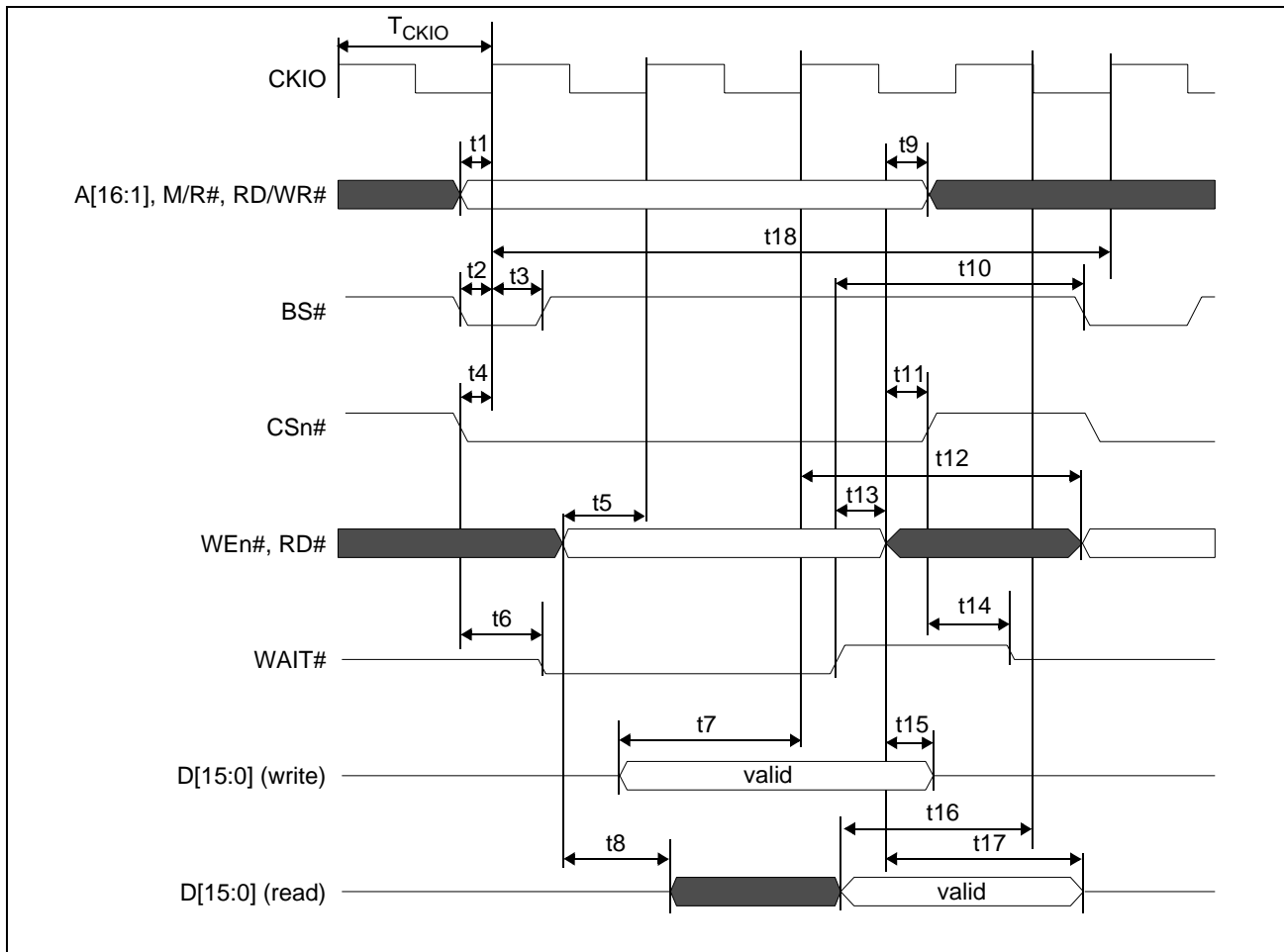


Figure 6-4: Hitachi SH-3 Interface Timing

Note

For this interface, the following formula must apply:

$$MCLK = BCLK \leq 33\text{MHz}$$

If a BCLK greater than 33MHz is desired, MCLK must be divided such that MCLK is not greater than 33MHz (see REG[04h] bits 5-4).

Table 6-10: Hitachi SH-3 Interface Timing

Symbol	Parameter	Min	Max	Unit
f_{CKIO}	Bus clock frequency		66	MHz
T_{CKIO}	Bus clock period	$1/f_{CKIO}$		ns
t1	A[16:1], RD/WR# setup to CKIO	1		ns
t2	BS# setup	1		ns
t3	BS# hold	5		ns
t4	CSn# setup	1		ns
t5	WEn#, RD# setup to next CKIO after BS# low	0		ns
t6	Falling edge CSn# to WAIT# driven low	3	10	ns
t7	D[15:0] setup to 3rd CKIO rising edge after BS# deasserted (write cycle)	1		T_{CKIO}
t8	Falling edge of RD# to D[15:0] driven (read cycle)	2	12	ns
t9	WE#, RD# deasserted to A[16:1], M/R#, RD/WR# deasserted	0		ns
t10	Rising edge of WAIT# to BS# falling	$T_{CKIO} + 16$		ns
t11	WE#, RD# deasserted to CS# high	0		ns
t12	CKIO rising edge before WAIT# deasserted to WEn#, RD# asserted for next cycle	2		T_{CKIO}
t13	Rising edge of WAIT# to WE#, RD# deasserted	0		ns
t14	Rising edge of CSn# to WAIT# high impedance		0.5	T_{CKIO}
t15	D[15:0] hold from WEn# deasserted (write cycle)	2 (note 1)		ns
t16	D[15:0] setup to CKIO falling edge (read cycle)	12		ns
t17	Rising edge of RD# to D[15:0] high impedance (read cycle)	1	5	ns
t18	Cycle Length	4		T_{CKIO}

1. The S1D13A04 requires 2ns of write data hold time.

6.2.4 Hitachi SH-4 Interface Timing

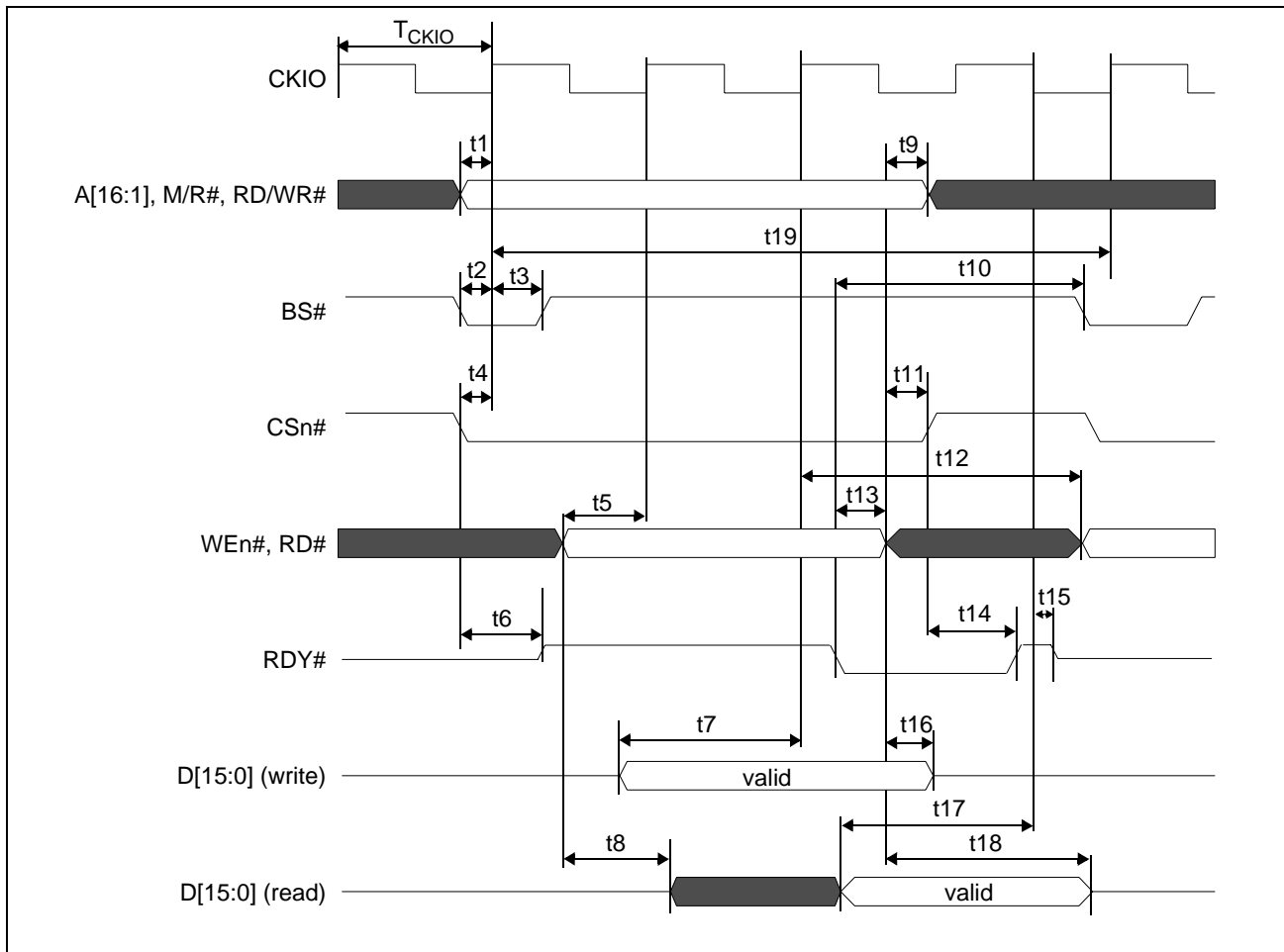


Figure 6-5: Hitachi SH-4 Interface Timing

Table 6-11: Hitachi SH-4 Interface Timing

Symbol	Parameter	Min	Max	Unit
f_{CKIO}	Bus clock frequency		66	MHz
T_{CKIO}	Bus clock period	$1/f_{CKIO}$		ns
t1	A[16:1], M/R#, RD/WR# setup to CKIO	1		ns
t2	BS# setup	1		ns
t3	BS# hold	5		ns
t4	CSn# setup	1		ns
t5	WEn#, RD# setup to 2nd CKIO rising edge after BS# low	0		ns
t6	Falling edge CSn# to RDY driven high	3	10	ns
t7	D[15:0] setup to 3rd CKIO rising edge after BS# deasserted (write cycle)	1		T_{CKIO}
t8	Falling edge RD# to D[15:0] driven (read cycle)	2	12	ns
t9	WE#,RD# deasserted to A[16:1],M/R#,RD/WR# deasserted	0		ns
t10	RDY falling edge to BS# falling	$T_{CKIO} + 11$		ns
t11	WE#,RD# deasserted to CS# high	0		ns
t12	CKIO rising edge before RDY deasserted to WEn#, RD# asserted for next cycle	2		T_{CKIO}
t13	RDY falling edge to WE#,RD# deasserted	0		ns
t14	Rising edge CSn# to RDY rising edge	3	10	ns
t15	CKIO falling edge to RDY tristate	3	9	ns
t16	D[15:0] hold from WEn# deasserted (write cycle)	3		ns
t17	D[15:0] setup to CKIO falling edge (read cycle)	12		ns
t18	Rising edge of RD# to D[15:0] high impedance (read cycle)	1	4	ns
t19	Cycle Length	4		T_{CKIO}

6.2.5 Motorola MC68K #1 Interface Timing (e.g. MC68000)

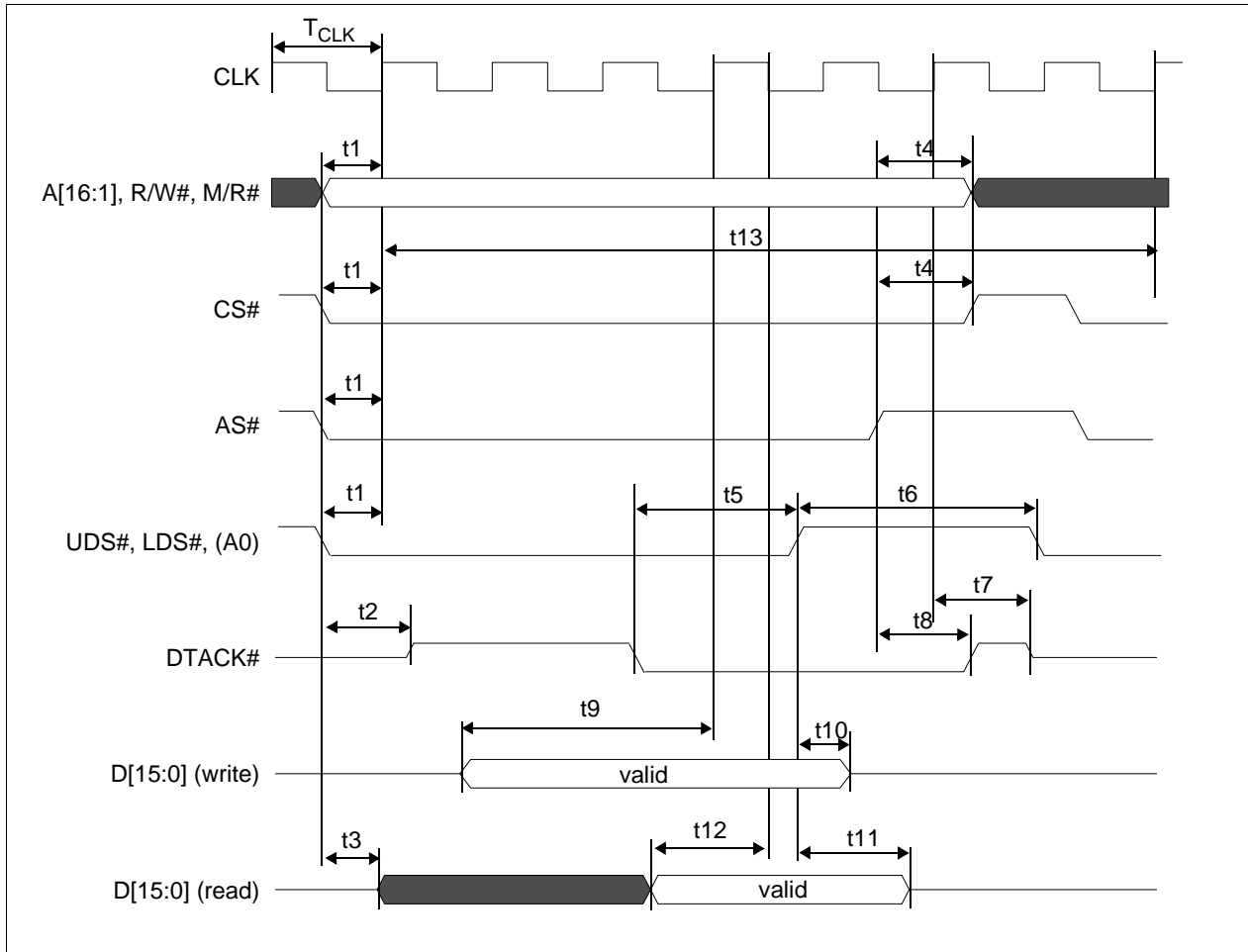


Figure 6-6: Motorola MC68K #1 Interface Timing

Table 6-12: Motorola MC68K#1 Interface Timing

Symbol	Parameter	Min	Max	Unit
f_{CLK}	Bus clock frequency		66	MHz
T_{CLK}	Bus clock period	$1/f_{CLK}$		ns
t1	A[16:1], M/R#, R/W# and CS# and AS# and UDS#, LDS# setup to first CLK rising edge	1		ns
t2	CS# and AS# asserted to DTACK# driven	3	10	ns
t3	UDS# = 0 or LDS# = 0 to D[15:0] driven (read cycle)		10	ns
t4	A[16:1], M/R#, R/W# and CS# hold from AS# rising edge	0		ns
t5	DTACK# falling edge to UDS#, LDS# rising edge	0		ns
t6	UDS#, LDS# deasserted high to reasserted low	1		T_{CLK}
t7	CLK rising edge to DTACK# high impedance		$T_{CLK} - 2$	ns
t8	AS# rising edge to DTACK# rising edge	3	11	ns
t9	D[15:0] valid to 4th CLK rising edge where CS# = 0, AS# = 0 and either UDS# = 0 or LDS# = 0 (write cycle)	1		T_{CLK}
t10	D[15:0] hold from UDS#, LDS# falling edge (write cycle)	4		ns
t11	UDS#, LDS# rising edge to D[15:0] high impedance (read cycle)	2		ns
t12	D[15:0] valid setup time to 2nd CLK falling edge after DTACK# goes low (read cycle)	10		ns
t13	Cycle Length	7		T_{CLK}

6.2.6 Motorola MC68K #2 Interface Timing (e.g. MC68030)

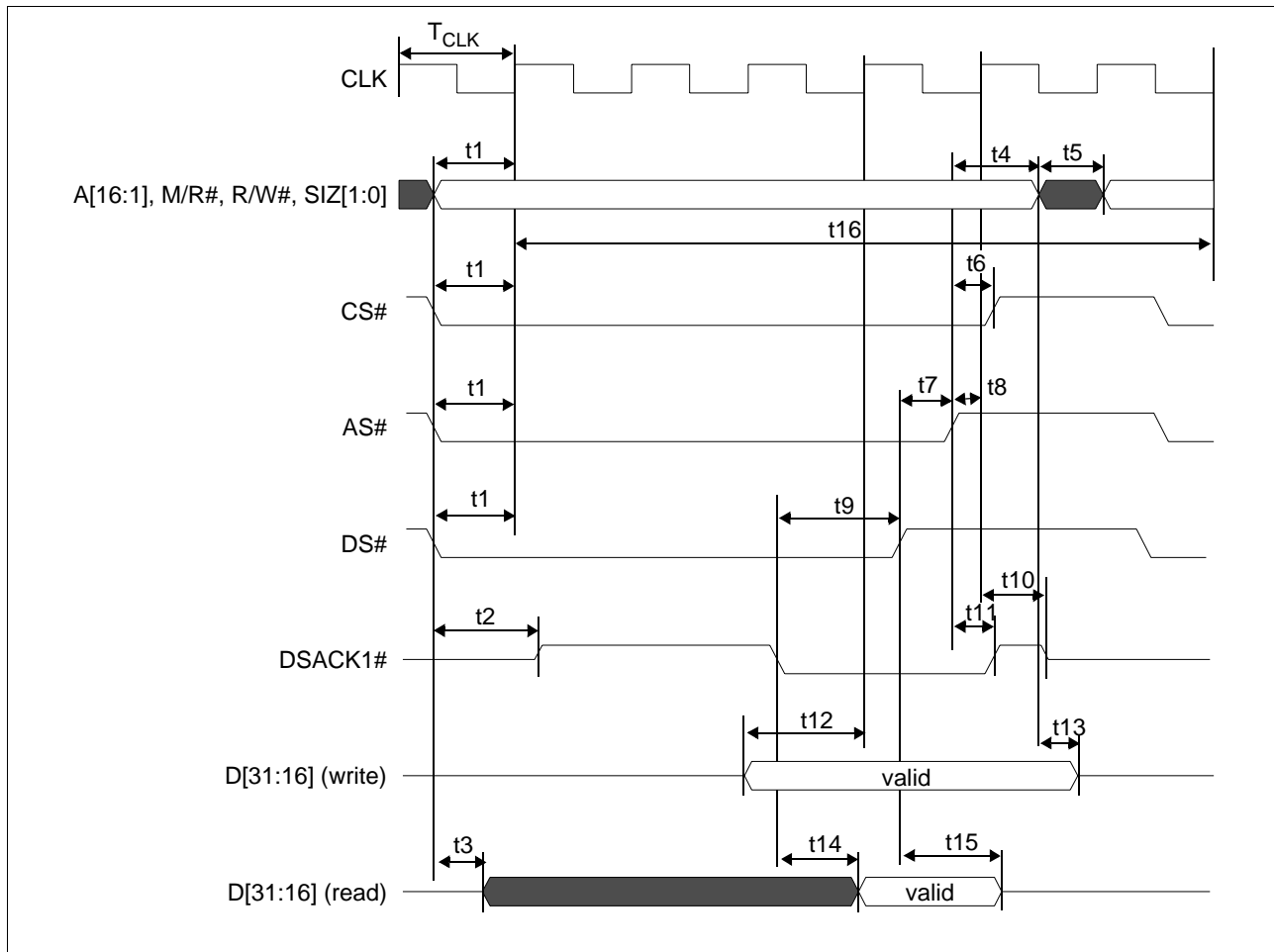


Figure 6-7: Motorola MC68K #2 Interface Timing

Table 6-13: Motorola MC68K#2 Interface Timing

Symbol	Parameter	Min	Max	Unit
f_{CLK}	Bus clock frequency		50	MHz
T_{CLK}	Bus clock period	$1/f_{CLK}$		ns
t1	A[16:0], M/R#, R/W#, SIZ[1:0] and CS# and AS# and DS# setup to first CLK rising edge	1		ns
t2	CS# and AS# asserted low to DSACK1# driven	3	10	ns
t3	A[1:0], R/W#, SIZ[1:0], DS# asserted to D[15:0] driven		8	ns
t4	A[16:1], M/R#, R/W#, SIZ[1:0] hold from AS# rising edge	0		ns
t5	R/W#, A[1:0], SIZ[1:0] deasserted to R/W#, A[1:0], SIZ[1:0] asserted for next cycle	1		T_{CLK}
t6	CS# hold from AS# rising edge	0		ns
t7	DS# rising edge to AS# rising edge	0		ns
t8	AS# setup to CLK rising edge	1		ns
t9	DSACK1# falling edge to DS# rising edge	0		ns
t10	CLK rising edge to DSACK1# high impedance		$T_{CLK} - 2$	ns
t11	AS# rising edge to DSACK1# rising edge	3	9	ns
t12	D[15:0] setup to 4th CLK rising edge after CS#=0, AS#=0, and UDS#=0 or LDS#=0	1		T_{CLK}
t13	D[15:0] hold from A[1:0], R/W#, SIZ[1:0] (write cycle)	1		ns
t14	DSACK1# falling edge to D[15:0] valid (read cycle)		3	ns
t15	DS# rising edge to D[15:0] high impedance (read cycle)	2	7	ns
t16	Cycle Length	6		T_{CLK}

6.2.7 Motorola REDCAP2 Interface Timing

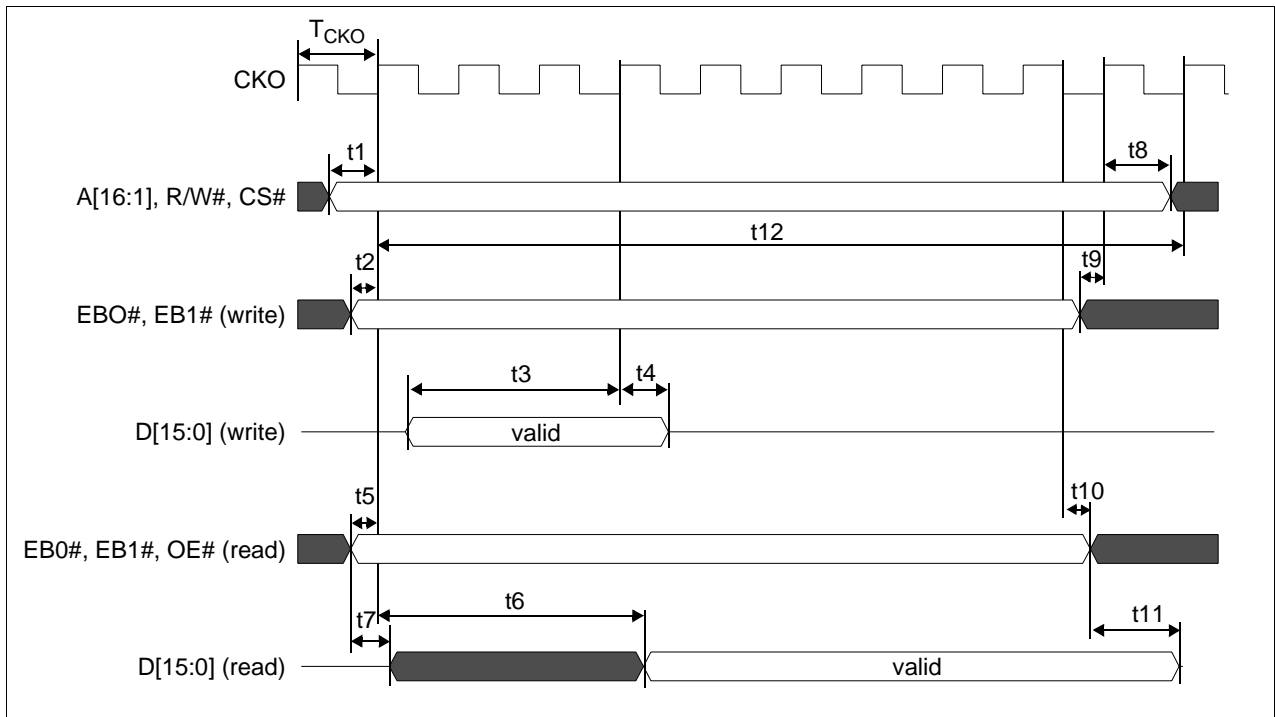


Figure 6-8: Motorola Redcap2 Interface Timing

Table 6-14: Motorola Redcap2 Interface Timing

Symbol	Parameter	Min	Max	Unit
f_{CKO}	Bus clock frequency		17	MHz
T_{CKO}	Bus clock period	$1/f_{CKO}$		ns
t1	A[16:1], R/W, CSn# setup to CKO rising edge	1		ns
t2	$\overline{EB0}, \overline{EB1}$ setup to CKO rising edge (write)	1		ns
t3	D[15:0] input setup to 4th CKO rising edge after CSn# and $\overline{EB0}$ or $\overline{EB1}$ asserted low (write cycle)	1		T_{CKO}
t4	D[15:0] input hold from 4th CKO rising edge after CSn# and $\overline{EB0}$ or $\overline{EB1}$ asserted low (write cycle)	11		ns
t5	$\overline{EB0}, \overline{EB1}, \overline{OE}$ setup to CKO rising edge (read cycle)	1		ns
t6a	1st CKO rising edge after CSn#, EB0 or EB1, OE asserted low to D[15:0] valid for MCLK = BCLK (read cycle)		5	T_{CKO}
t6b	1st CKO rising edge after CSn#, EB0 or EB1, OE asserted low to D[15:0] valid for MCLK = BCLK ÷ 2 (read cycle)		8	T_{CKO}
t6c	1st CKO rising edge after CSn#, EB0 or EB1, OE asserted low to D[15:0] valid for MCLK = BCLK ÷ 3 (read cycle)		10	T_{CKO}
t6d	1st CKO rising edge after CSn#, EB0 or EB1, OE asserted low to D[15:0] valid for MCLK = BCLK ÷ 4 (read cycle)		13	T_{CKO}
t7	$\overline{EB0}, \overline{EB1}, \overline{OE}$ falling edge to D[15:0] driven (read cycle)	4	11	ns
t8	A[16:1], R/W, CSn hold from CKO rising edge	0		ns
t9	EB0, EB1 setup to CKO rising edge (write cycle)	1		ns
t10	CKO falling edge to EB0, EB1, OE deasserted (read)	0		ns
t11	OE, EB0, EB1 deasserted to D[15:0] output high impedance (read)	1	7	ns
t12	Cycle Length (note 1)	10	10	T_{CKO}

1. The cycle length for the REDCAP interface is fixed.
2. The Read and Write 2D BitBLT functions are not available when using the REDCAP interface.

6.2.8 Motorola Dragonball Interface Timing with \overline{DTACK} (e.g. MC68EZ328/MC68VZ328)

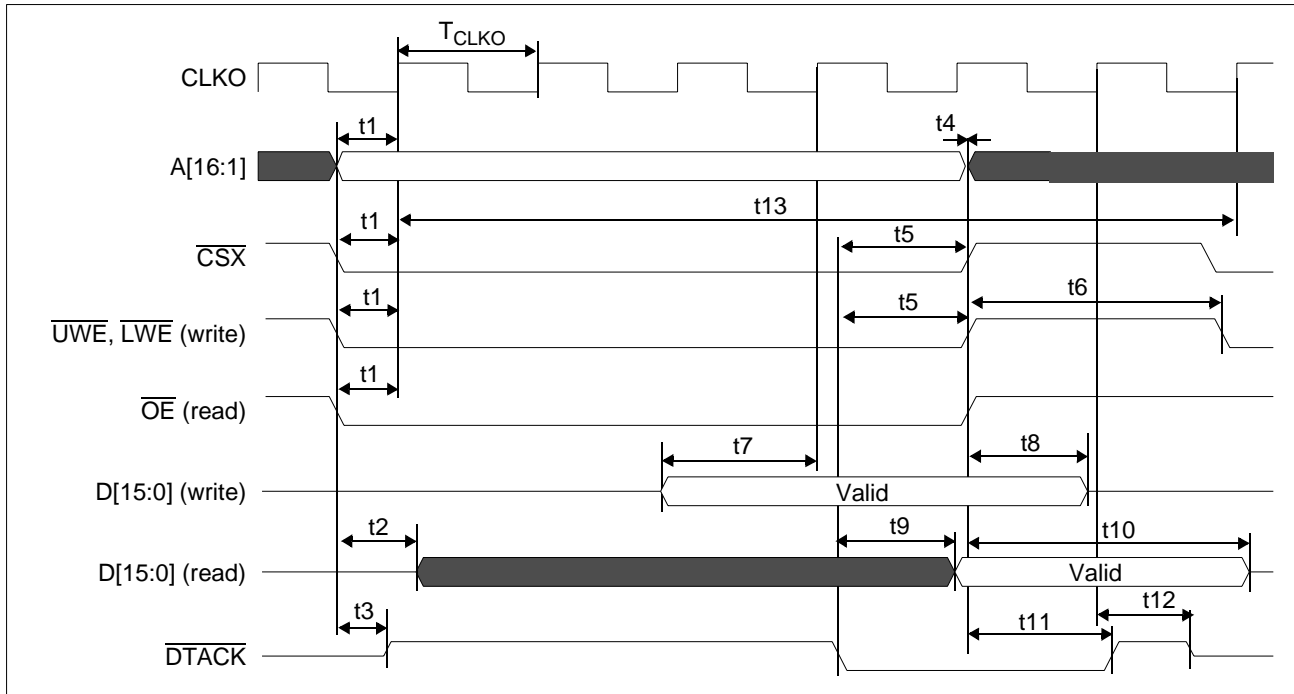


Figure 6-9: Motorola Dragonball Interface Timing with \overline{DTACK}

Table 6-15: Motorola Dragonball Interface Timing with \overline{DTACK}

Symbol	Parameter	Min	Max	Unit
f_{CLKO}	Clock frequency		33 (note 1)	MHz
T_{CLKO}	Clock period	$1/f_{CLKO}$		ns
t1	A[16:1], \overline{CSX} , \overline{UWE} , \overline{LWE} , \overline{OE} setup to CLKO rising edge	1		ns
t2	\overline{CSX} and \overline{OE} asserted low to D[15:0] driven (read cycle)		11	ns
t3	\overline{CSX} asserted low to \overline{DTACK} driven		11	ns
t4	A[16:1] hold from \overline{CSX} rising edge	0		ns
t5	\overline{DTACK} falling edge to \overline{UWE} , \overline{LWE} and \overline{CSX} rising edge	0		ns
t6	\overline{UWE} , \overline{LWE} deasserted to reasserted	2		T_{CLKO}
t7	D[15:0] valid to fourth CLKO rising edge where $\overline{CSX} = 0$ and $\overline{UWE} = 0$ or $\overline{LWE} = 0$ (write cycle)	1		T_{CLKO}
t8	D[15:0] hold from \overline{UWE} , \overline{LWE} rising edge (write cycle)	2		ns
t9	\overline{DTACK} falling edge to D[15:0] valid (read cycle)		$T_{CLKO} + 4$	ns
t10	\overline{CSX} rising edge to D[15:0] high impedance (read cycle)	2	6	ns
t11	\overline{CSX} rising edge to \overline{DTACK} rising edge	3	9	ns
t12	CLKO rising edge to \overline{DTACK} high impedance		9	ns
t13	Cycle Length	5		T_{CLKO}

1. The MC68VZ328 has a maximum clock frequency of 33MHz.
The MC68EZ328 has a maximum clock frequency of 16MHz.

6.2.9 Motorola Dragonball Interface Timing w/o \overline{DTACK} (e.g. MC68EZ328/MC68VZ328)

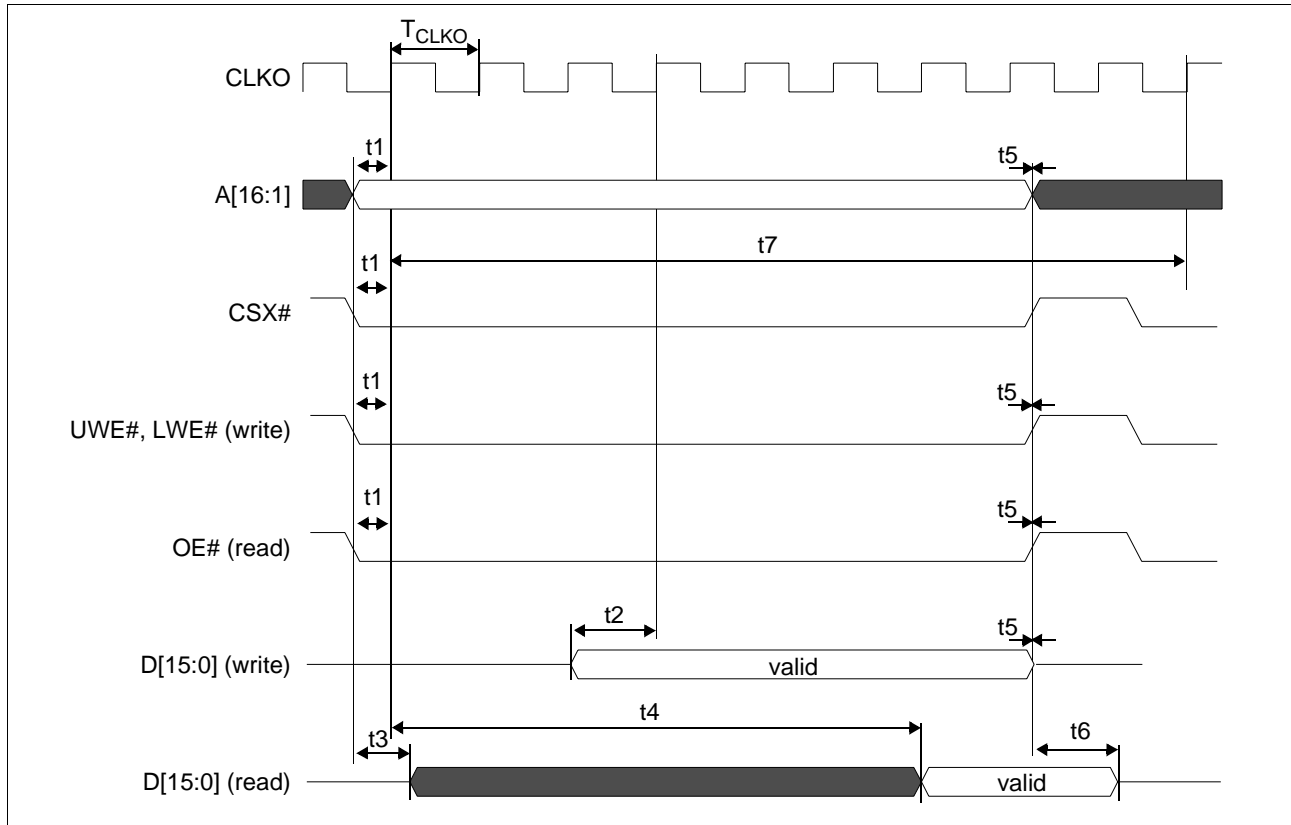


Figure 6-10: Motorola Dragonball Interface Timing w/o \overline{DTACK}

Table 6-16: Motorola Dragonball Interface Timing w/o \overline{DTACK}

Symbol	Parameter	Min	Max	Unit
f_{CLKO}	Bus clock frequency		33 (note 1)	MHz
T_{CLKO}	Bus clock period	$1/f_{CLKO}$		ns
t1	A[16:1] and CSX# and UWE#, LWE# and OE# setup to CLKO rising edge	1		ns
t2	D[15:0] valid to 4th CLK rising edge where CSX# = 0 and UWE# = 0 or LWE# = 0 (write cycle)	1		T_{CLKO}
t3	CSX# and OE# asserted low to D[15:0] driven (read cycle)		11	ns
t4a	1st CLKO rising edge after CSX# and OE# asserted to D[15:0] valid for MCLK=BCLK (read cycle)		5	T_{CLKO}
t4b	1st CLKO rising edge after CSX# and OE# asserted to D[15:0] valid for MCLK=BCLK + 2 (read cycle)		8	T_{CLKO}
t4c	1st CLKO rising edge after CSX# and OE# asserted to D[15:0] valid for MCLK=BCLK + 3 (read cycle) (see note 2)		10	T_{CLKO}
t4d	1st CLKO rising edge after CSX# and OE# asserted to D[15:0] valid for MCLK=BCLK + 4 (read cycle) (see note 2)		13	T_{CLKO}
t5	A[16:1] and UWE#, LWE# and OE# and D[15:0] (write) hold from CSX# rising edge	0		ns
t6	CSX# rising edge to D[15:0] high impedance	2	6	ns
t7	Cycle Length	9	9	T_{CLKO}

1. The MC68VZ328 has a maximum clock frequency of 33MHz.
The MC68EZ328 has a maximum clock frequency of 16MHz.
2. The MC68EZ328 does not support the MCLK = BCLK + 3 and MCLK = BCLK + 4 options.
3. The cycle length for the Dragonball w/o \overline{DTACK} interface is fixed.
4. The Read and Write 2D BitBLT functions are not available when using the Dragonball w/o \overline{DTACK} interface.

6.3 LCD Power Sequencing

6.3.1 Passive/TFT Power-On Sequence

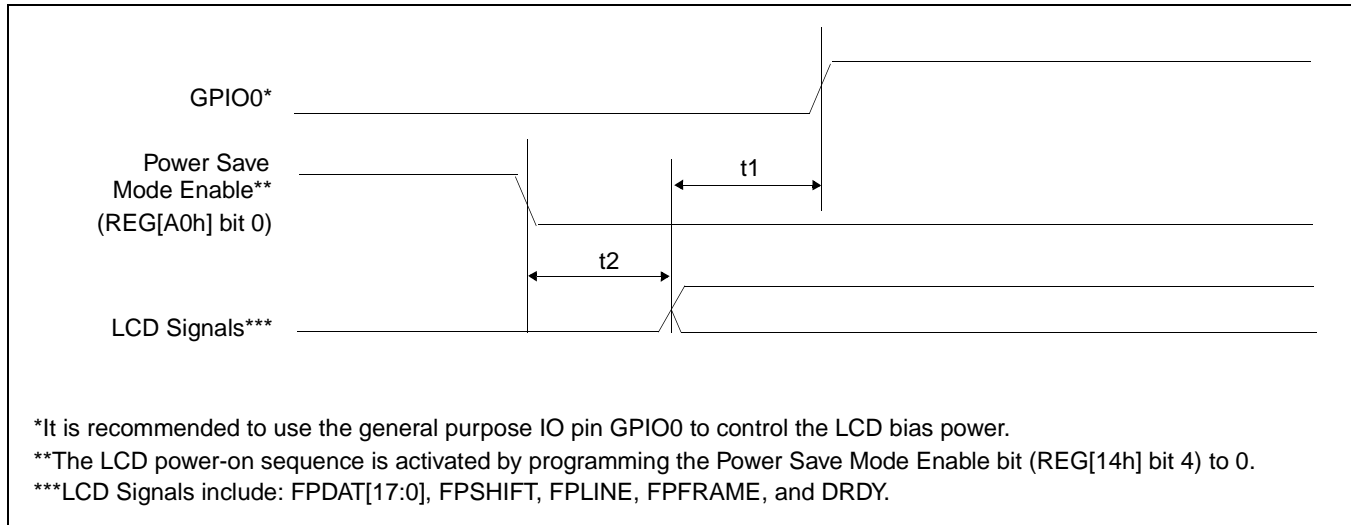


Figure 6-11: Passive/TFT Power-On Sequence Timing

Table 6-17: Passive/TFT Power-On Sequence Timing

Symbol	Parameter	Min	Max	Units
t1	LCD signals active to LCD bias active	Note 1	Note 1	
t2	Power Save Mode disabled to LCD signals active	0	1	BCLK

- t1 is controlled by software and must be determined from the bias power supply delay requirements of the panel connected.

6.3.2 Passive/TFT Power-Off Sequence

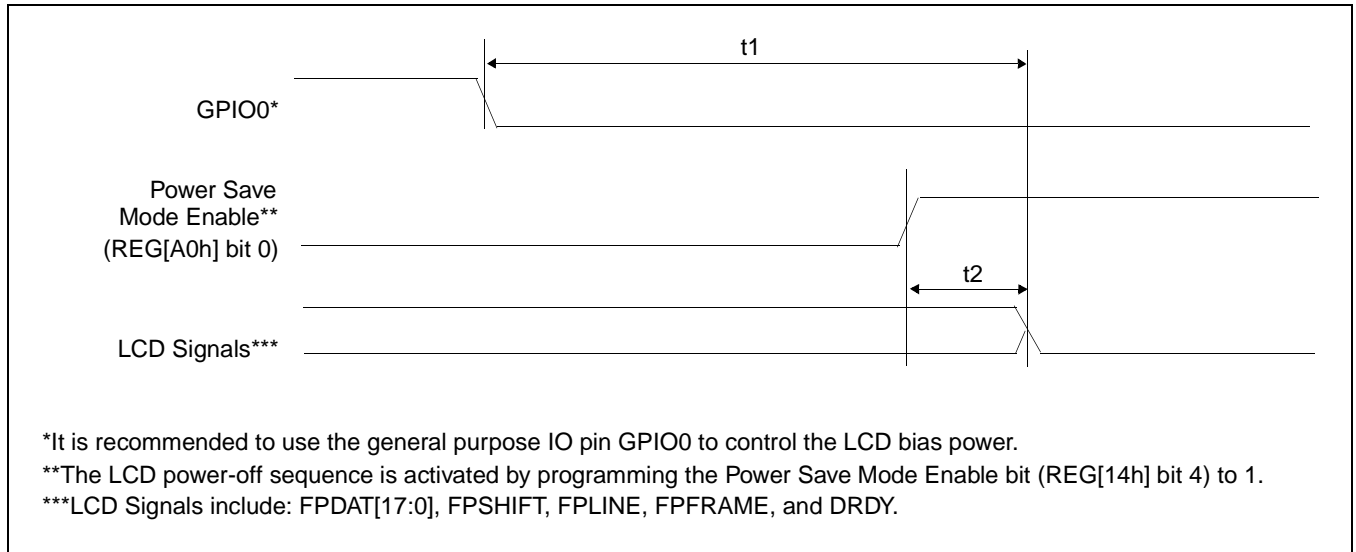


Figure 6-12: Passive/TFT Power-Off Sequence Timing

Table 6-18: Passive/TFT Power-Off Sequence Timing

Symbol	Parameter	Min	Max	Units
t1	LCD bias deactivated to LCD signals inactive	Note 1	Note 1	
t2	Power Save Mode enabled to LCD signals low	0	1	BCLK

- t1 is controlled by software and must be determined from the bias power supply delay requirements of the panel connected.

6.3.3 'Direct' HR-TFT Interface Power-On/Off Sequence

For 'Direct' HR-TFT Interface Power-On/Off sequence information, see *Connecting to the Sharp HR-TFT Panels*, document number X37A-G-011-xx.

6.4 Display Interface

The timing parameters required to drive a flat panel display are shown below. Timing details for each supported panel type are provided in the remainder of this section.

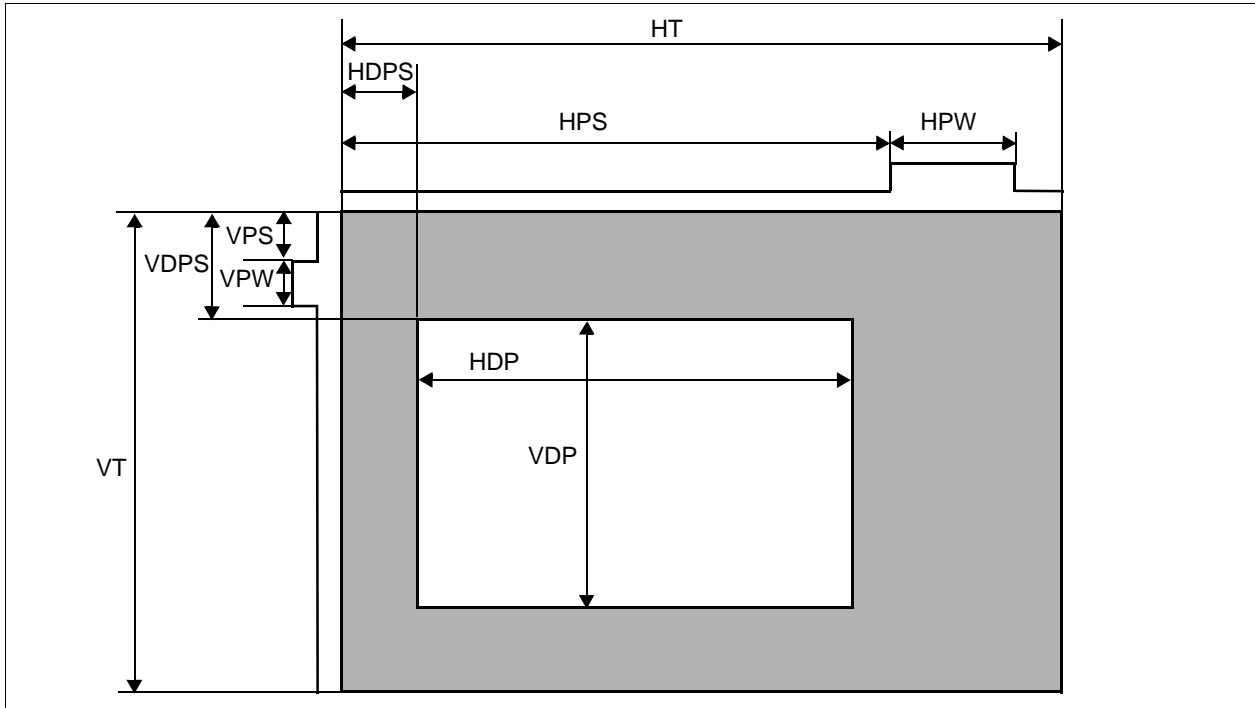


Figure 6-13: Panel Timing Parameters

Table 6-19: Panel Timing Parameter Definition and Register Summary

Symbol	Description	Derived From	Units
HT	Horizontal Total	$((\text{REG}[20\text{h}] \text{ bits } 6-0) + 1) \times 8$	Ts
HDP ¹	Horizontal Display Period ¹	$((\text{REG}[24\text{h}] \text{ bits } 6-0) + 1) \times 8$	
HDPS	Horizontal Display Period Start Position	For STN panels: $((\text{REG}[28\text{h}] \text{ bits } 9-0) + 22)$ For TFT panels: $((\text{REG}[28\text{h}] \text{ bits } 9-0) + 5)$	
HPS	FPLINE Pulse Start Position	$(\text{REG}[2\text{Ch}] \text{ bits } 9-0) + 1$	
HPW	FPLINE Pulse Width	$(\text{REG}[2\text{Ch}] \text{ bits } 22-16) + 1$	
VT	Vertical Total	$(\text{REG}[30\text{h}] \text{ bits } 9-0) + 1$	Lines (HT)
VDP	Vertical Display Period	$(\text{REG}[34\text{h}] \text{ bits } 9-0) + 1$	
VDPS	Vertical Display Period Start Position	$\text{REG}[38\text{h}] \text{ bits } 9-0$	
VPS	FPFRAME Pulse Start Position	$\text{REG}[2\text{Ch}] \text{ bits } 9-0$	
VPW	FPFRAME Pulse Width	$(\text{REG}[3\text{Ch}] \text{ bits } 18-16) + 1$	

- For passive panels, the HDP must be a minimum of 32 pixels and must be increased by multiples of 16.
For TFT panels, the HDP must be a minimum of 8 pixels and must be increased by multiples of 8.
- The following formulas must be valid for all panel timings:

$$\text{HDPS} + \text{HDP} < \text{HT}$$

$$\text{VDPS} + \text{VDP} < \text{VT}$$

6.4.1 Generic STN Panel Timing

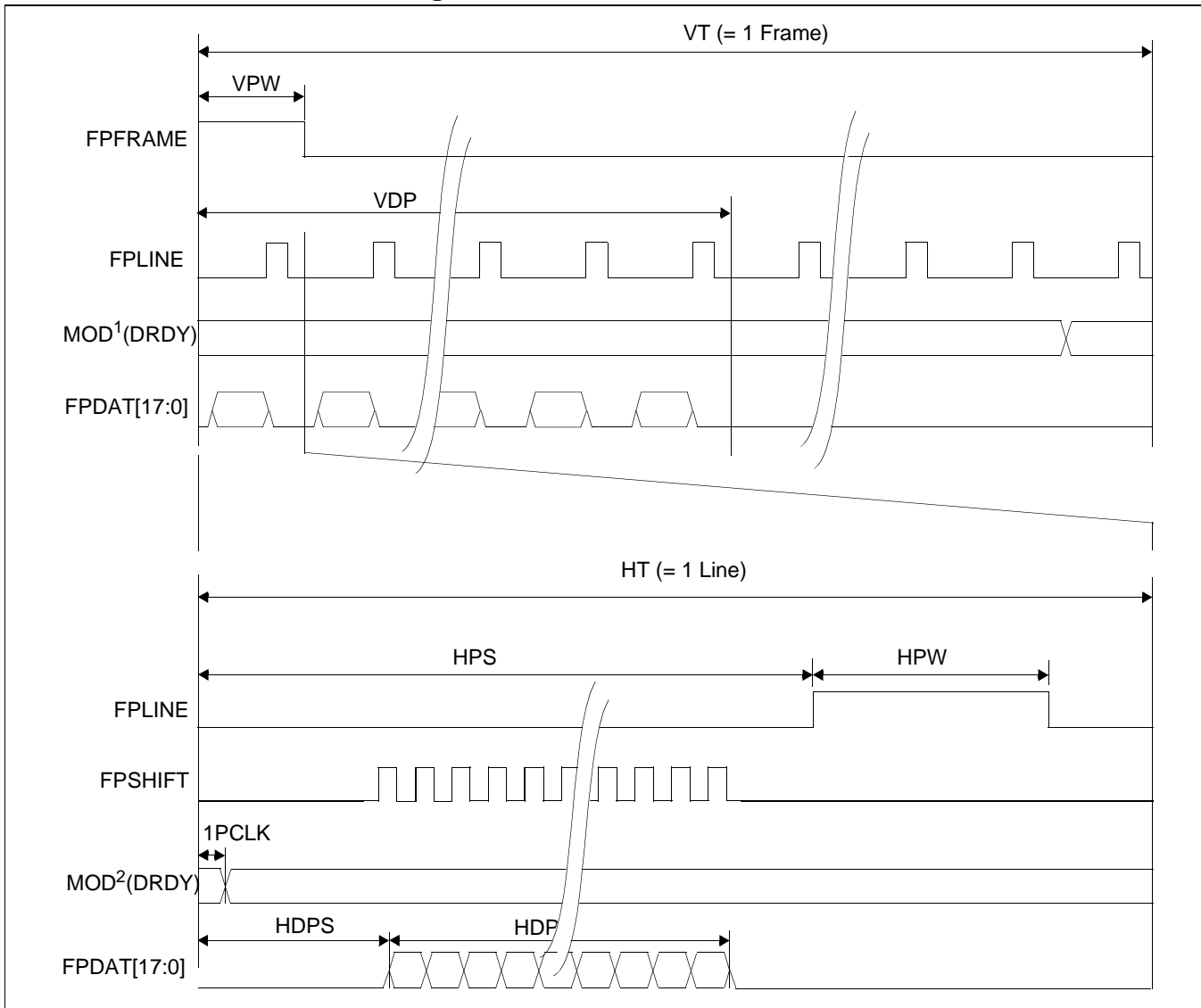


Figure 6-14: Generic STN Panel Timing

VT	= Vertical Total	= [(REG[30h] bits 9-0) + 1] lines
VPS	= FPFRAME Pulse Start Position	= 0 lines, because REG[2Ch] bits 9-0 = 0
VPW	= FPFRAME Pulse Width	= [(REG[3Ch] bits 18-16) + 1] lines
VDPS	= Vertical Display Period Start Position	= 0 lines, because REG[38h] bits 9-0 = 0
VDP	= Vertical Display Period	= [(REG[34h] bits 9-0) + 1] lines
HT	= Horizontal Total	= [((REG[20h] bits 6-0) + 1) x 8] pixels
HPS	= FPLINE Pulse Start Position	= [(REG[2Ch] bits 9-0) + 1] pixels
HPW	= FPLINE Pulse Width	= [(REG[2Ch] bits 22-16) + 1] pixels
HDPS	= Horizontal Display Period Start Position	= 22 pixels, because REG[28h] bits 9-0 = 0
HDP	= Horizontal Display Period	= [((REG[24h] bits 6-0) + 1) x 8] pixels

*For passive panels, the HDP must be a minimum of 32 pixels and must be increased by multiples of 16.

*HPS must comply with the following formula:

$$HPS > HDP + 22$$

$$HPS + HPW < HT$$

*Panel Type Bits (REG[0Ch] bits 1-0) = 00b (STN)

*FPFRAME Pulse Polarity Bit (REG[3Ch] bit 23) = 1 (active high)

*FPLINE Polarity Bit (REG[2Ch] bit 23) = 1 (active high)

*MOD¹ is the MOD signal when REG[0Ch] bits 21-16 = 0 (MOD toggles every FPFRAME)

*MOD² is the MOD signal when REG[0Ch] bits 21-16 = n (MOD toggles every n FPLINE)

6.4.2 Single Monochrome 4-Bit Panel Timing

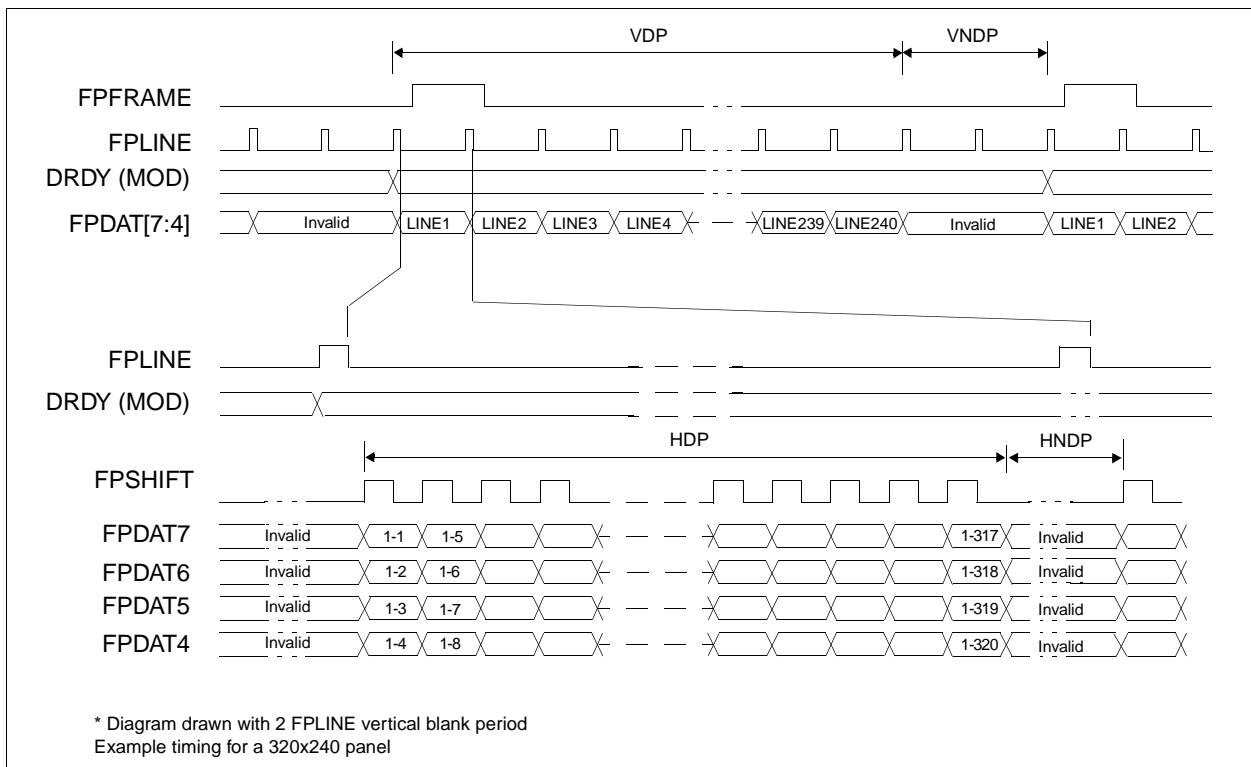


Figure 6-15: Single Monochrome 4-Bit Panel Timing

- VDP = Vertical Display Period
= (REG[34h] bits 9:0) + 1 Lines
- VNDP = Vertical Non-Display Period
= VT - VDP
= (REG[30h] bits 9:0) - (REG[34h] bits 9:0) Lines
- HDP = Horizontal Display Period
= ((REG[24h] bits 6:0) + 1) x 8Ts
- HNDP = Horizontal Non-Display Period
= HT - HDP
= (((REG[20h] bits 6:0) + 1) x 8Ts) - (((REG[24h] bits 6:0) + 1) x 8Ts)

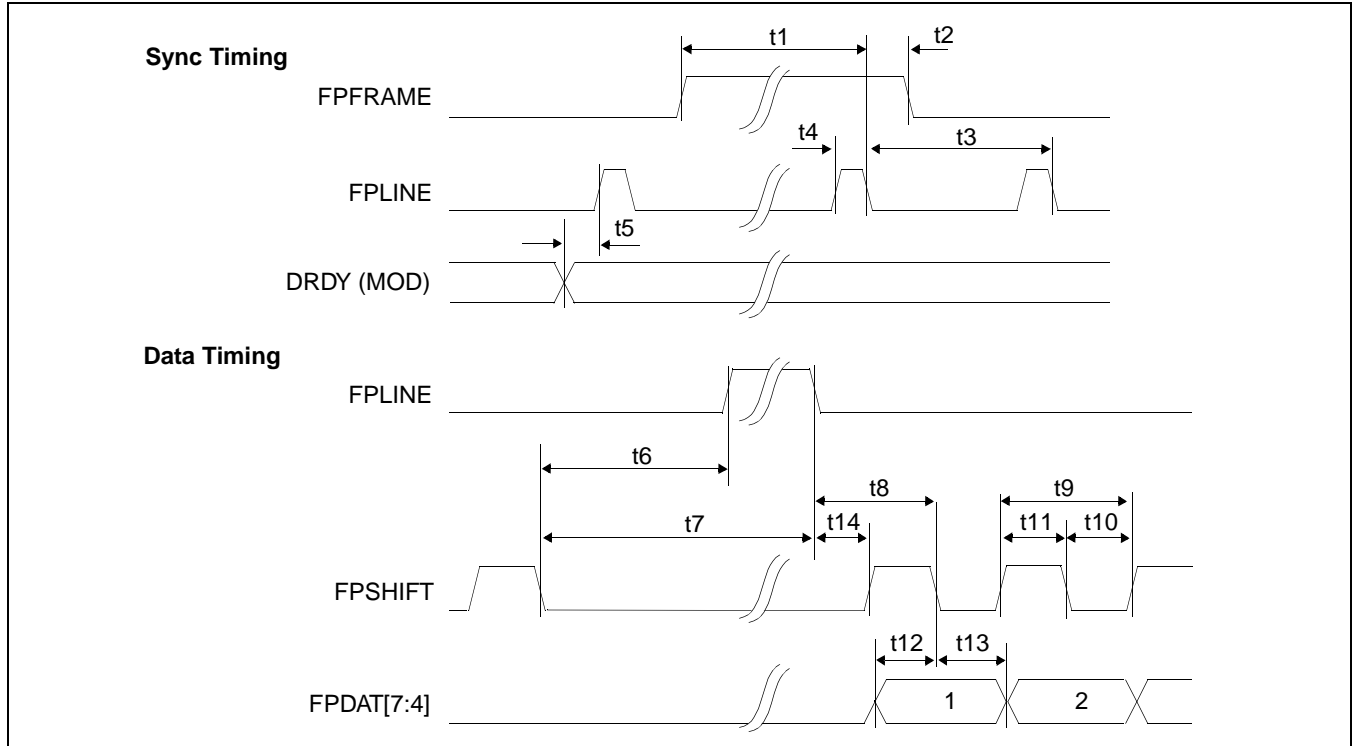


Figure 6-16: Single Monochrome 4-Bit Panel A.C. Timing

Table 6-20: Single Monochrome 4-Bit Panel A.C. Timing

Symbol	Parameter	Min	Typ	Max	Units
t1	FFRAME setup to FPLINE falling edge	note 2			Ts (note 1)
t2	FFRAME hold from FPLINE falling edge	note 3			Ts
t3	FPLINE period	note 4			Ts
t4	FPLINE pulse width	note 5			Ts
t5	MOD transition to FPLINE rising edge	note 6			Ts
t6	FPSHIFT falling edge to FPLINE rising edge	note 7			Ts
t7	FPSHIFT falling edge to FPLINE falling edge	t6 + t4			Ts
t8	FPLINE falling edge to FPSHIFT falling edge	t14 + 2			Ts
t9	FPSHIFT period	4			Ts
t10	FPSHIFT pulse width low	2			Ts
t11	FPSHIFT pulse width high	2			Ts
t12	FPDAT[7:4] setup to FPSHIFT falling edge	1			Ts
t13	FPDAT[7:4] hold to FPSHIFT falling edge	2			Ts
t14	FPLINE falling edge to FPSHIFT rising edge	note 8			Ts

1. Ts = pixel clock period
2. $t1_{min} = HPS + t4_{min}$
3. $t2_{min} = t3_{min} - (HPS + t4_{min})$
4. $t3_{min} = HT$
5. $t4_{min} = HPW$
6. $t5_{min} = HPS - 1$
7. $t6_{min} = HPS - (HDP + HDPS) + 2$, if negative add $t3_{min}$
8. $t14_{min} = HDPS - (HPS + t4_{min})$, if negative add $t3_{min}$

6.4.3 Single Monochrome 8-Bit Panel Timing

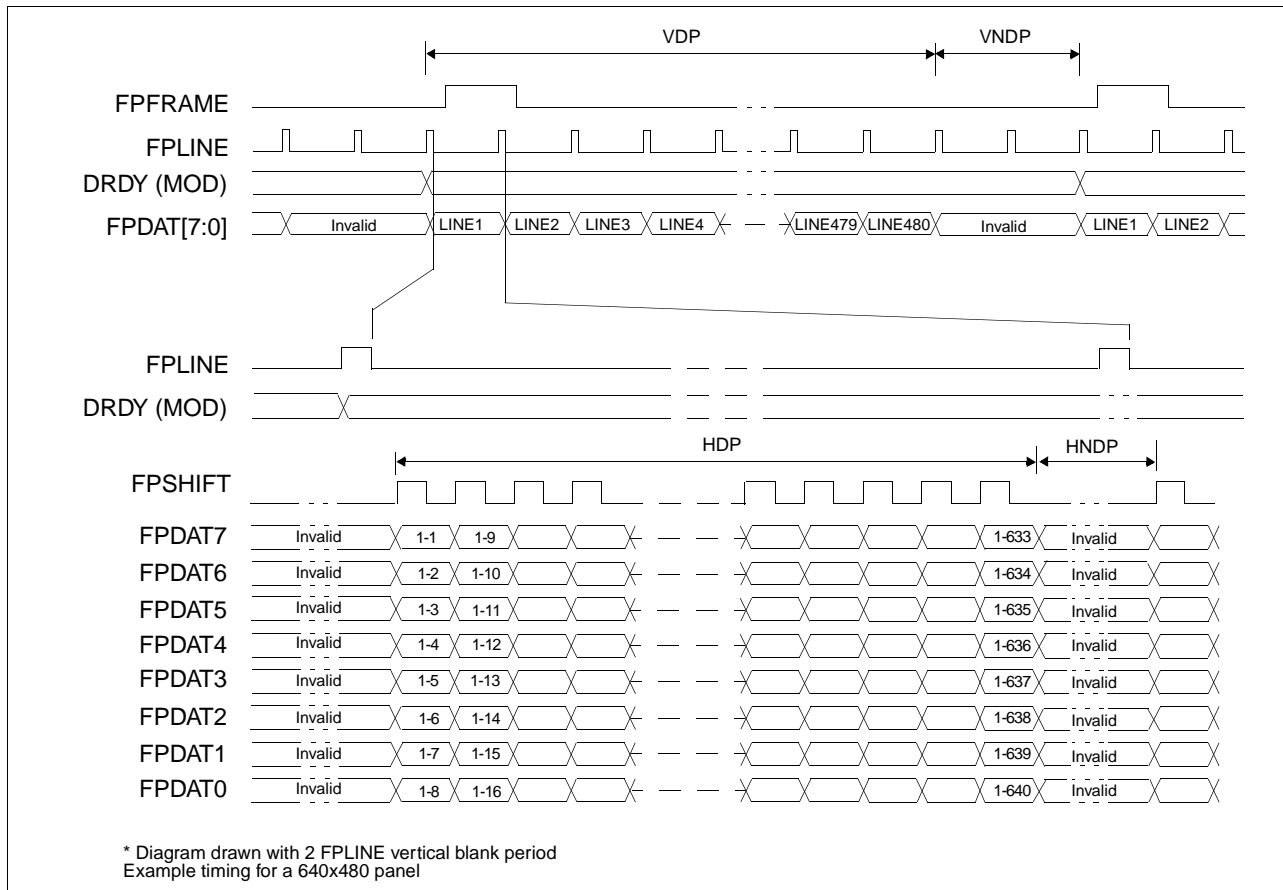


Figure 6-17: Single Monochrome 8-Bit Panel Timing

- VDP = Vertical Display Period
= (REG[34h] bits 9:0) + 1 Lines
- VNDP = Vertical Non-Display Period
= VT - VDP
= (REG[30h] bits 9:0) - (REG[34h] bits 9:0) Lines
- HDP = Horizontal Display Period
= ((REG[24h] bits 6:0) + 1) x 8Ts
- HNDP = Horizontal Non-Display Period
= HT - HDP
= (((REG[20h] bits 6:0) + 1) x 8Ts) - (((REG[24h] bits 6:0) + 1) x 8Ts)

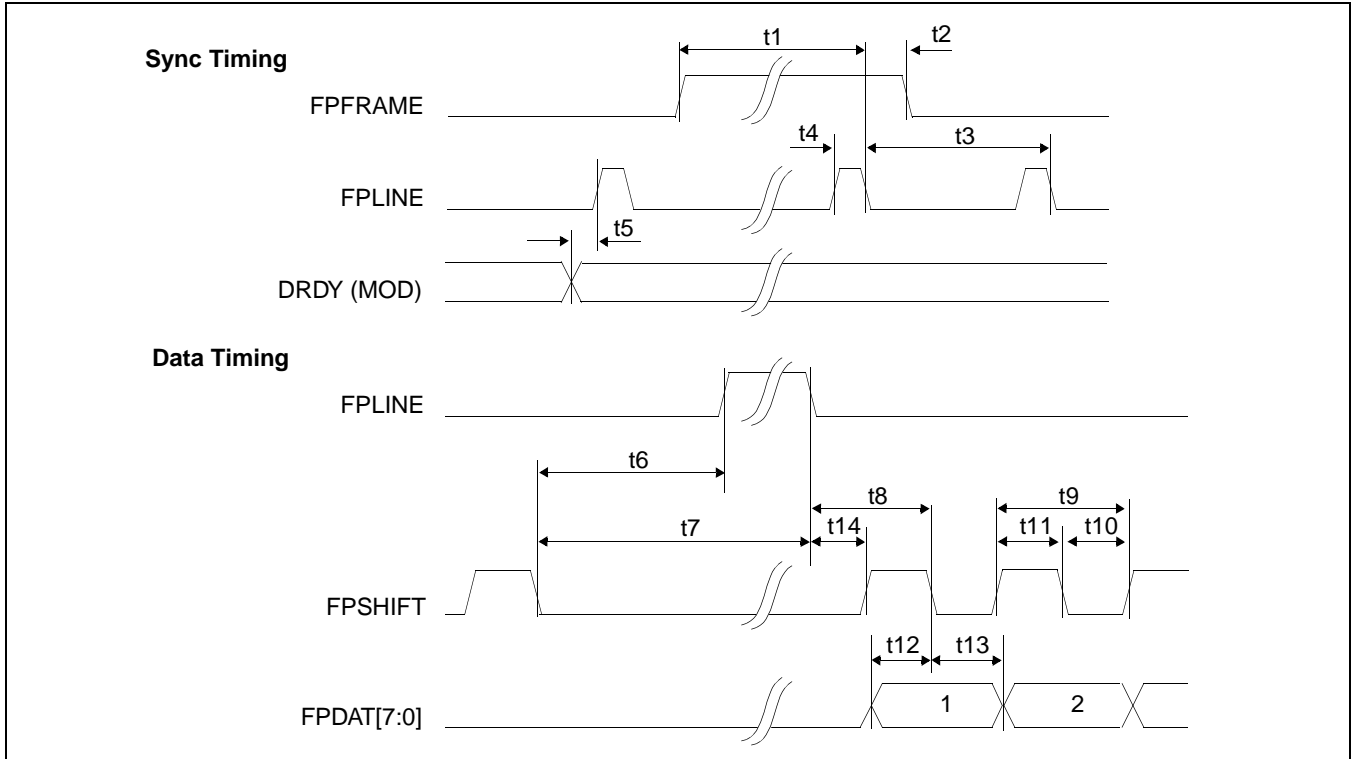


Figure 6-18: Single Monochrome 8-Bit Panel A.C. Timing

Table 6-21: Single Monochrome 8-Bit Panel A.C. Timing

Symbol	Parameter	Min	Typ	Max	Units
t1	FPFRAME setup to FPLINE falling edge	note 2			Ts (note 1)
t2	FPFRAME hold from FPLINE falling edge	note 3			Ts
t3	FPLINE period	note 4			Ts
t4	FPLINE pulse width	note 5			Ts
t5	MOD transition to FPLINE rising edge	note 6			Ts
t6	FPSHIFT falling edge to FPLINE rising edge	note 7			Ts
t7	FPSHIFT falling edge to FPLINE falling edge	t6 + t4			Ts
t8	FPLINE falling edge to FPSHIFT falling edge	t14 + 4			Ts
t9	FPSHIFT period	8			Ts
t10	FPSHIFT pulse width low	4			Ts
t11	FPSHIFT pulse width high	4			Ts
t12	FPDAT[7:0] setup to FPSHIFT falling edge	4			Ts
t13	FPDAT[7:0] hold to FPSHIFT falling edge	4			Ts
t14	FPLINE falling edge to FPSHIFT rising edge	note 8			Ts

1. Ts = pixel clock period
2. $t1_{min} = HPS + t4_{min}$
3. $t2_{min} = t3_{min} - (HPS + t4_{min})$
4. $t3_{min} = HT$
5. $t4_{min} = HPW$
6. $t5_{min} = HPS - 1$
7. $t6_{min} = HPS - (HDP + HDPS) + 4$, if negative add $t3_{min}$
8. $t14_{min} = HDPS - (HPS + t4_{min})$, if negative add $t3_{min}$

6.4.4 Single Color 4-Bit Panel Timing

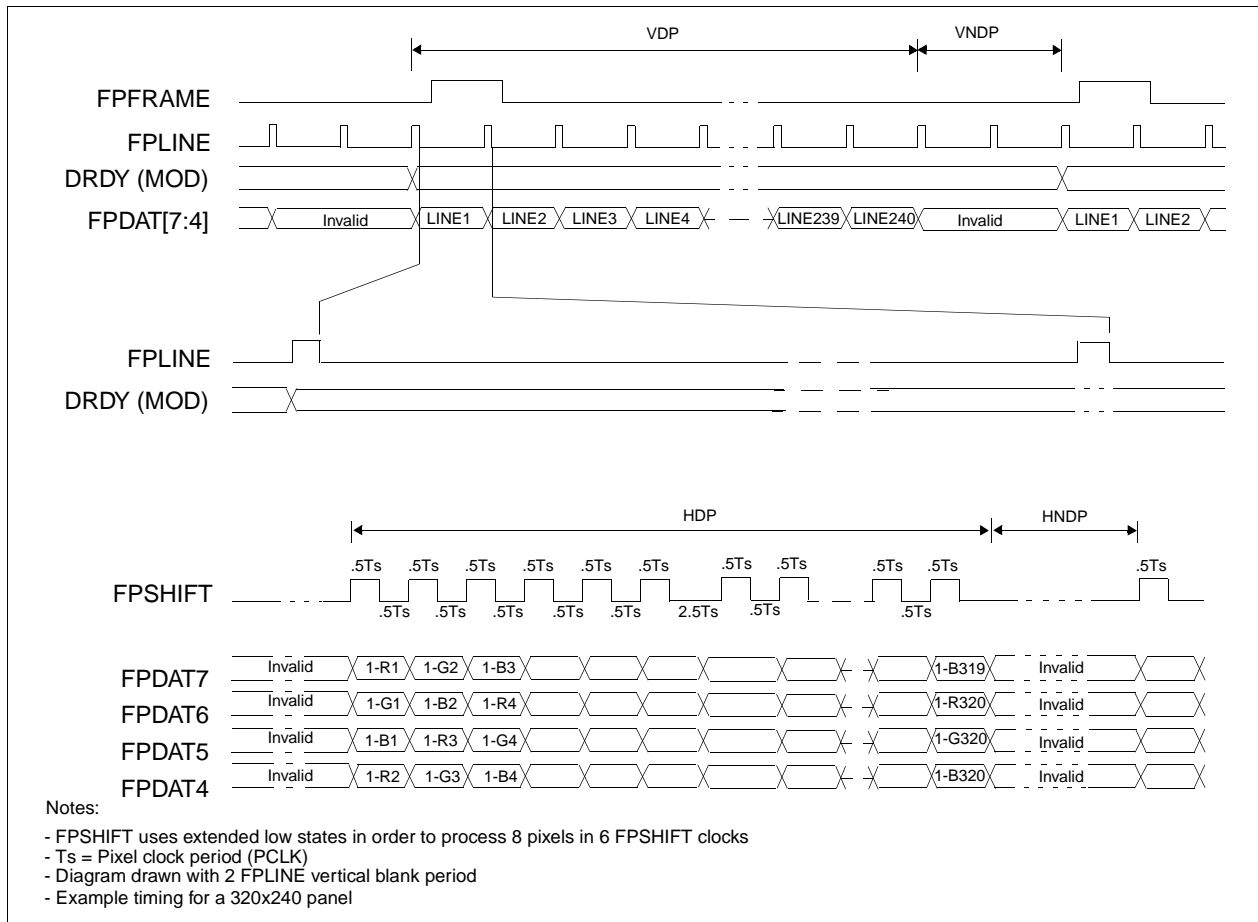


Figure 6-19: Single Color 4-Bit Panel Timing

- VDP = Vertical Display Period
= (REG[34h] bits 9:0) + 1 Lines
- VNDP = Vertical Non-Display Period
= VT - VDP
= (REG[30h] bits 9:0) - (REG[34h] bits 9:0) Lines
- HDP = Horizontal Display Period
= ((REG[24h] bits 6:0) + 1) x 8Ts
- HNDP = Horizontal Non-Display Period
= HT - HDP
= (((REG[20h] bits 6:0) + 1) x 8Ts) - (((REG[24h] bits 6:0) + 1) x 8Ts)

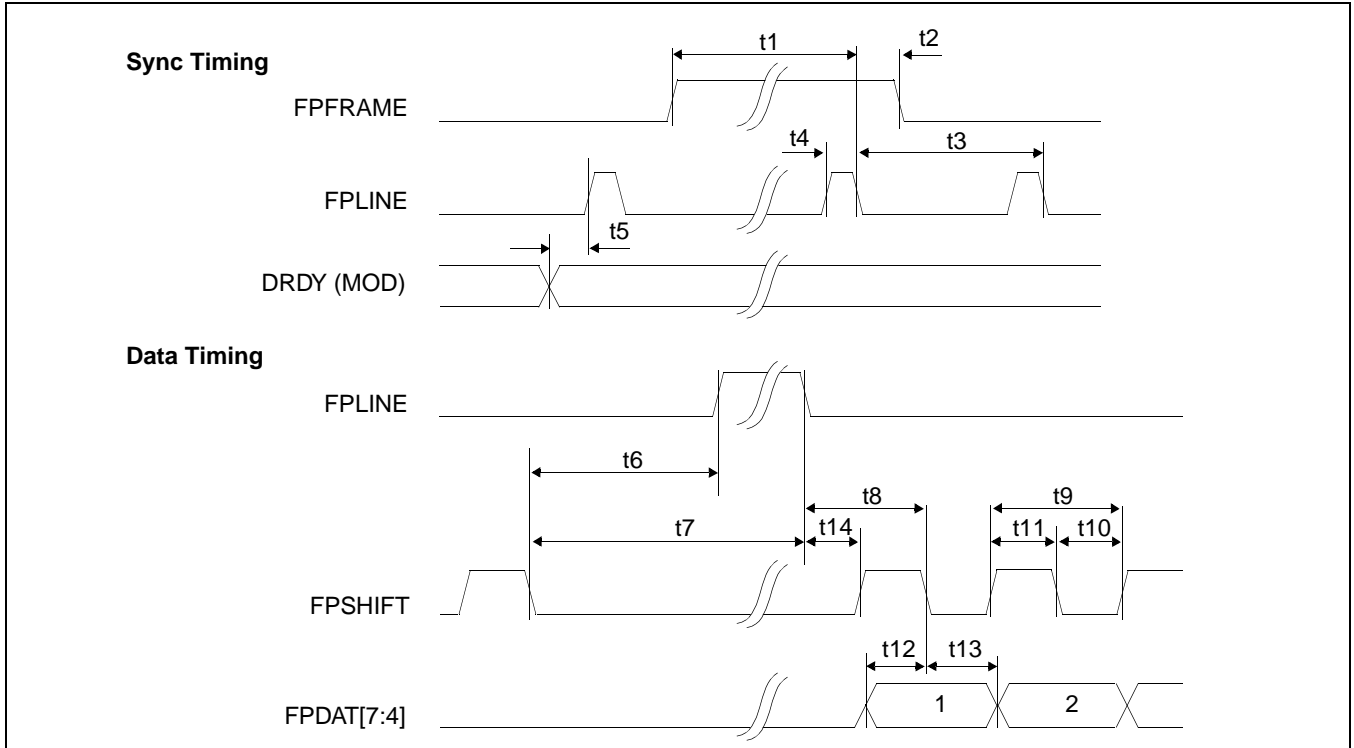


Figure 6-20: Single Color 4-Bit Panel A.C. Timing

Table 6-22: Single Color 4-Bit Panel A.C. Timing

Symbol	Parameter	Min	Typ	Max	Units
t1	FPFRAME setup to FPLINE falling edge	note 2			Ts (note 1)
t2	FPFRAME hold from FPLINE falling edge	note 3			Ts
t3	FPLINE period	note 4			Ts
t4	FPLINE pulse width	note 5			Ts
t5	MOD transition to FPLINE rising edge	note 6			Ts
t6	FPSHIFT falling edge to FPLINE rising edge	note 7			Ts
t7	FPSHIFT falling edge to FPLINE falling edge	t6 + t4			Ts
t8	FPLINE falling edge to FPSHIFT falling edge	t14 + 0.5			Ts
t9	FPSHIFT period	1			Ts
t10	FPSHIFT pulse width low	0.5			Ts
t11	FPSHIFT pulse width high	0.5			Ts
t12	FPDAT[7:4] setup to FPSHIFT falling edge	0.5			Ts
t13	FPDAT[7:4] hold to FPSHIFT falling edge	0.5			Ts
t14	FPLINE falling edge to FPSHIFT rising edge	note 8			Ts

1. Ts = pixel clock period
2. t1_{min} = HPS + t4_{min}
3. t2_{min} = t3_{min} - (HPS + t4_{min})
4. t3_{min} = HT
5. t4_{min} = HPW
6. t5_{min} = HPS - 1
7. t6_{min} = HPS - (HDP + HDPS) + 1.5, if negative add t3_{min}
8. t14_{min} = HDPS - (HPS + t4_{min}) + 1, if negative add t3_{min}

6.4.5 Single Color 8-Bit Panel Timing (Format 1)

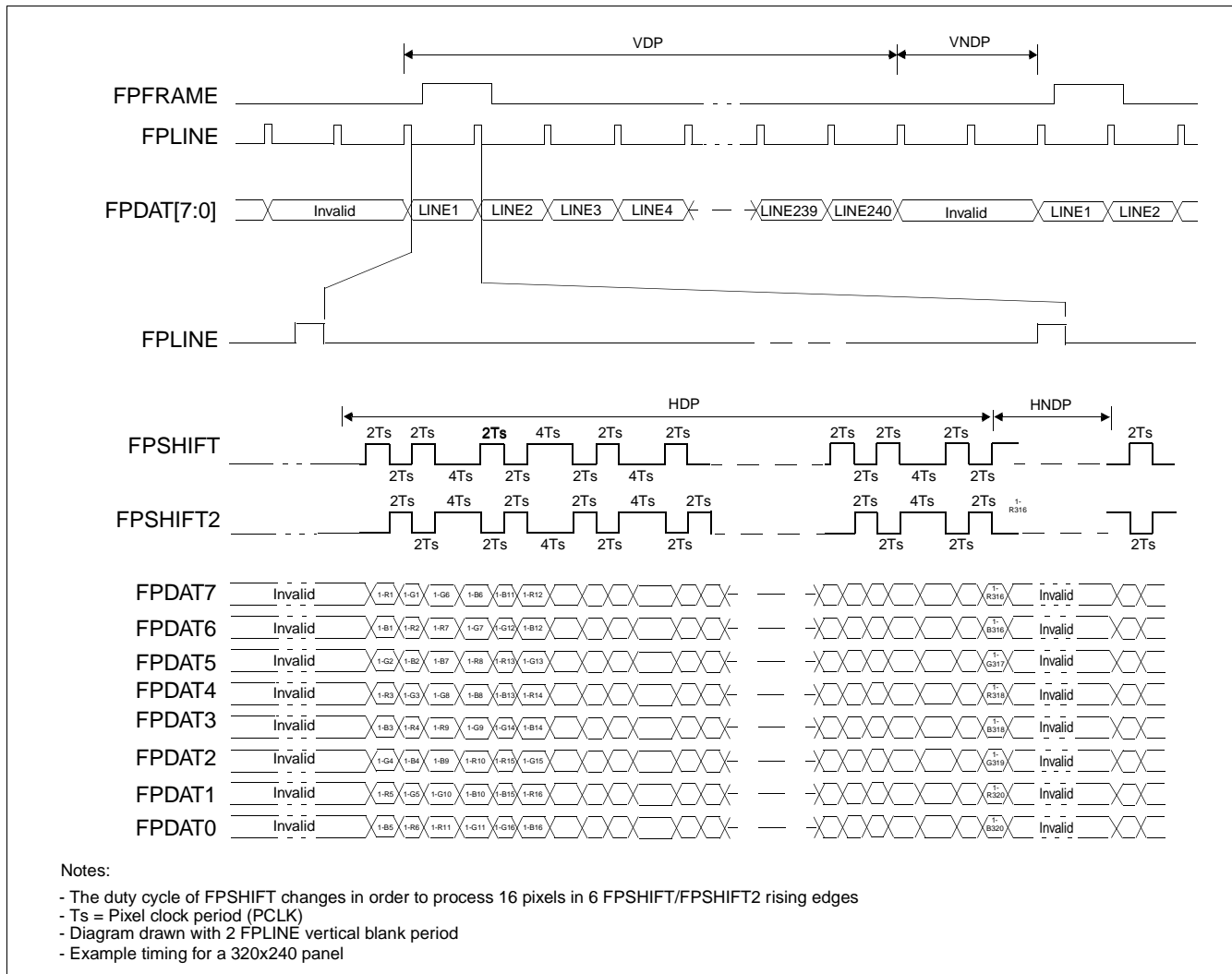


Figure 6-21: Single Color 8-Bit Panel Timing (Format 1)

- VDP = Vertical Display Period
= (REG[34h] bits 9:0) + 1 Lines
- VNDP = Vertical Non-Display Period
= VT - VDP
= (REG[30h] bits 9:0) - (REG[34h] bits 9:0) Lines
- HDP = Horizontal Display Period
= ((REG[24h] bits 6:0) + 1) x 8Ts
- HNDP = Horizontal Non-Display Period
= HT - HDP
= (((REG[20h] bits 6:0) + 1) x 8Ts) - (((REG[24h] bits 6:0) + 1) x 8Ts)

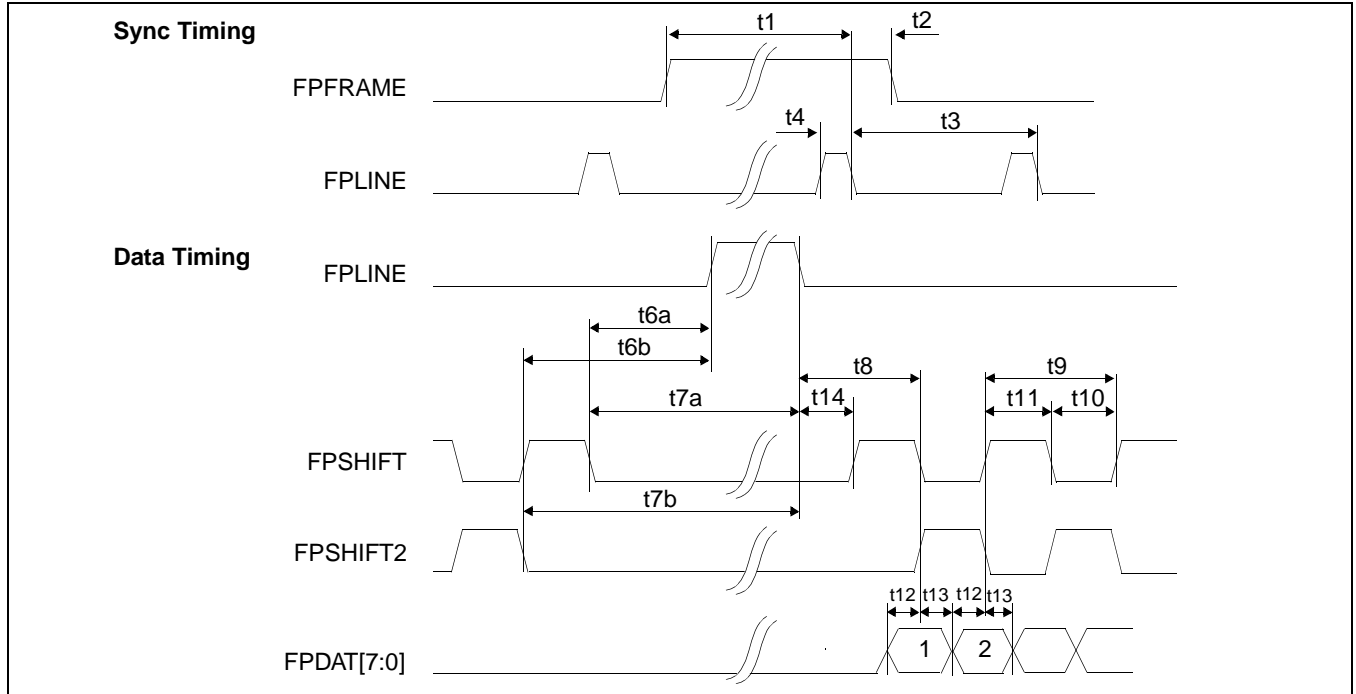


Figure 6-22: Single Color 8-Bit Panel A.C. Timing (Format 1)

Table 6-23: Single Color 8-Bit Panel A.C. Timing (Format 1)

Symbol	Parameter	Min	Typ	Max	Units
t1	FPFRAME setup to FPLINE falling edge	note 2			Ts (note 1)
t2	FPFRAME hold from FPLINE falling edge	note 3			Ts
t3	FPLINE period	note 4			Ts
t4	FPLINE pulse width	note 5			Ts
t6a	FPSHIFT falling edge to FPLINE rising edge	note 6			Ts
t6b	FPSHIFT2 falling edge to FPLINE rising edge	note 7			Ts
t7a	FPSHIFT falling edge to FPLINE falling edge	t6a + t4			Ts
t7b	FPSHIFT2 falling edge to FPLINE falling edge	t6b + t4			Ts
t8	FPLINE falling edge to FPSHIFT rising, FPSHIFT2 falling edge	t14 + 2			Ts
t9	FPSHIFT2, FPSHIFT period	4		6	Ts
t10	FPSHIFT2, FPSHIFT pulse width low	2			Ts
t11	FPSHIFT2, FPSHIFT pulse width high	2			Ts
t12	FPDAT[7:0] setup to FPSHIFT2, FPSHIFT falling edge	1			Ts
t13	FPDAT[7:0] hold from FPSHIFT2, FPSHIFT falling edge	1			Ts
t14	FPLINE falling edge to FPSHIFT rising edge	note 8			Ts

1. Ts = pixel clock period
2. $t1_{min} = HPS + t4_{min}$
3. $t2_{min} = t3_{min} - (HPS + t4_{min})$
4. $t3_{min} = HT$
5. $t4_{min} = HPW$
6. $t6a_{min} = HPS - (HDP + HDPS)$, if negative add $t3_{min}$
7. $t6b_{min} = HPS - (HDP + HDPS) + 2$, if negative add $t3_{min}$
8. $t14_{min} = HDPS - (HPS + t4_{min})$, if negative add $t3_{min}$

6.4.6 Single Color 8-Bit Panel Timing (Format 2)

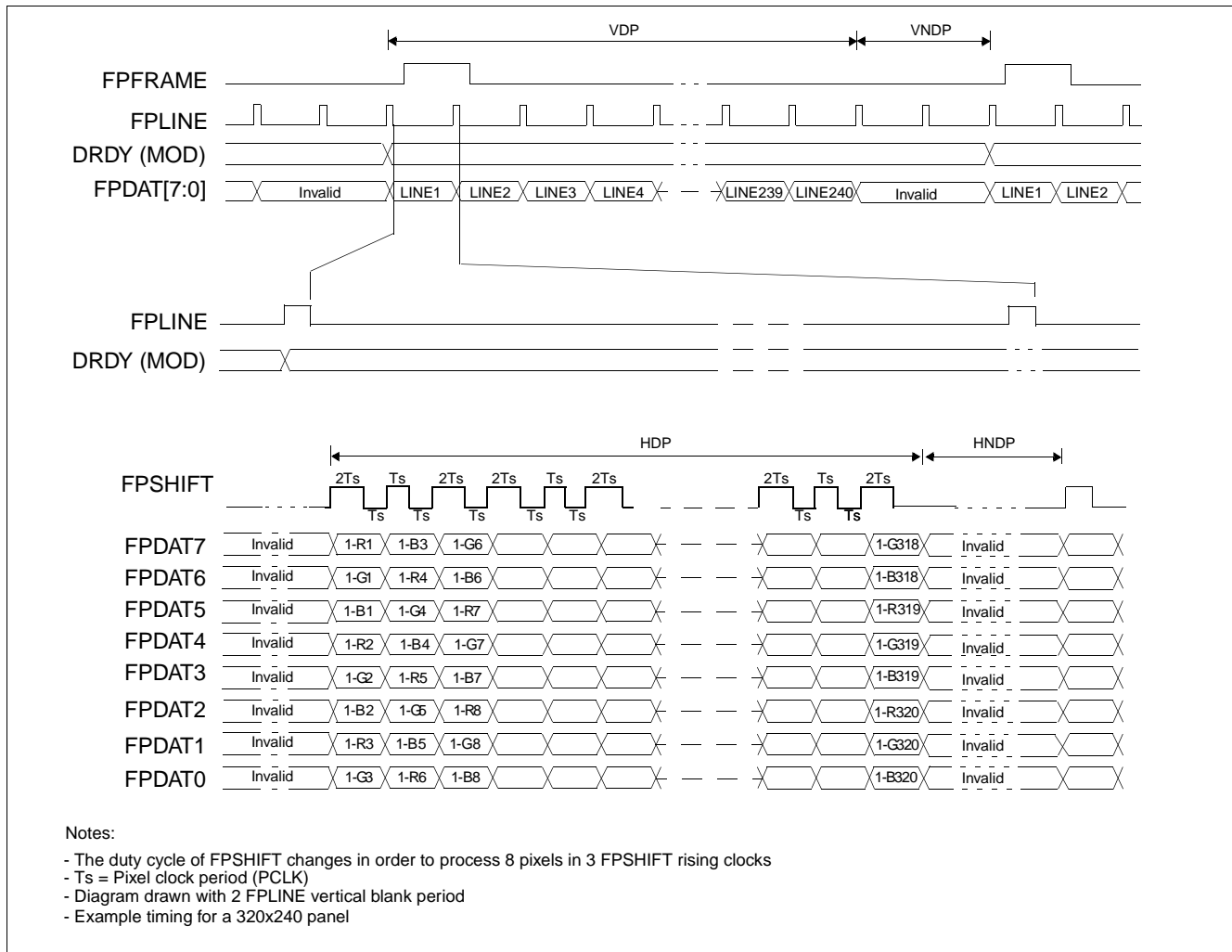


Figure 6-23: Single Color 8-Bit Panel Timing (Format 2)

VDP = Vertical Display Period
 = $(\text{REG}[34\text{h}] \text{ bits } 9:0) + 1$ Lines

VNDP = Vertical Non-Display Period
 = $\text{VT} - \text{VDP}$
 = $(\text{REG}[30\text{h}] \text{ bits } 9:0) - (\text{REG}[34\text{h}] \text{ bits } 9:0)$ Lines

HDP = Horizontal Display Period
 = $((\text{REG}[24\text{h}] \text{ bits } 6:0) + 1) \times 8T_s$

HNDP = Horizontal Non-Display Period
 = $\text{HT} - \text{HDP}$
 = $((\text{REG}[20\text{h}] \text{ bits } 6:0) + 1) \times 8T_s - ((\text{REG}[24\text{h}] \text{ bits } 6:0) + 1) \times 8T_s$

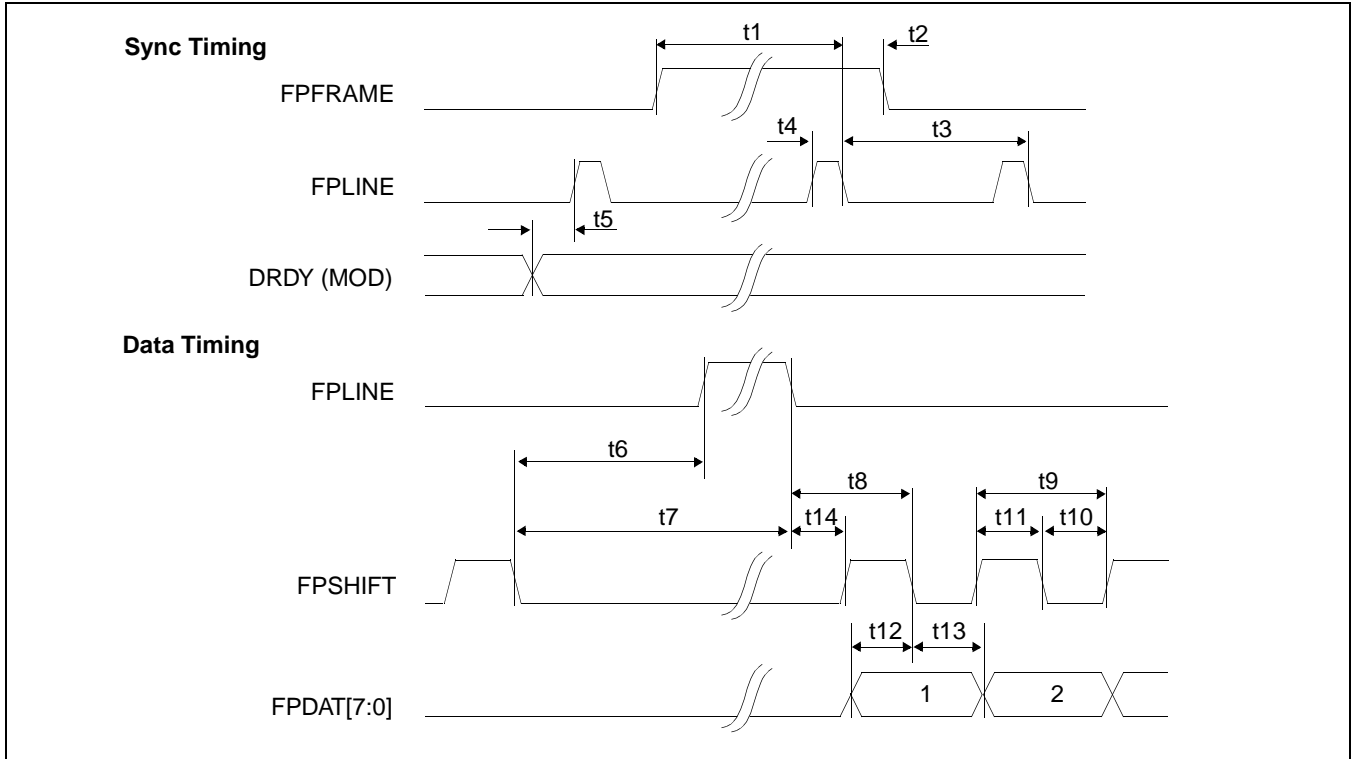


Figure 6-24: Single Color 8-Bit Panel A.C. Timing (Format 2)

Table 6-24: Single Color 8-Bit Panel A.C. Timing (Format 2)

Symbol	Parameter	Min	Typ	Max	Units
t1	FPFRAME setup to FPLINE falling edge	note 2			Ts (note 1)
t2	FPFRAME hold from FPLINE falling edge	note 3			Ts
t3	FPLINE period	note 4			Ts
t4	FPLINE pulse width	note 5			Ts
t5	MOD transition to FPLINE rising edge	note 6			Ts
t6	FPSHIFT falling edge to FPLINE rising edge	note 7			Ts
t7	FPSHIFT falling edge to FPLINE falling edge	t6 + t4			Ts
t8	FPLINE falling edge to FPSHIFT falling edge	t14 + 2			Ts
t9	FPSHIFT period	2			Ts
t10	FPSHIFT pulse width low	1			Ts
t11	FPSHIFT pulse width high	1			Ts
t12	FPDAT[7:0] setup to FPSHIFT falling edge	1			Ts
t13	FPDAT[7:0] hold to FPSHIFT falling edge	1			Ts
t14	FPLINE falling edge to FPSHIFT rising edge	note 8			Ts

1. Ts = pixel clock period
2. t1_{min} = HPS + t4_{min}
3. t2_{min} = t3_{min} - (HPS + t4_{min})
4. t3_{min} = HT
5. t4_{min} = HPW
6. t5_{min} = HPS - 1
7. t6_{min} = HPS - (HDP + HDPS) + 1, if negative add t3_{min}
8. t14_{min} = HDPS - (HPS + t4_{min}), if negative add t3_{min}

6.4.7 Single Color 16-Bit Panel Timing

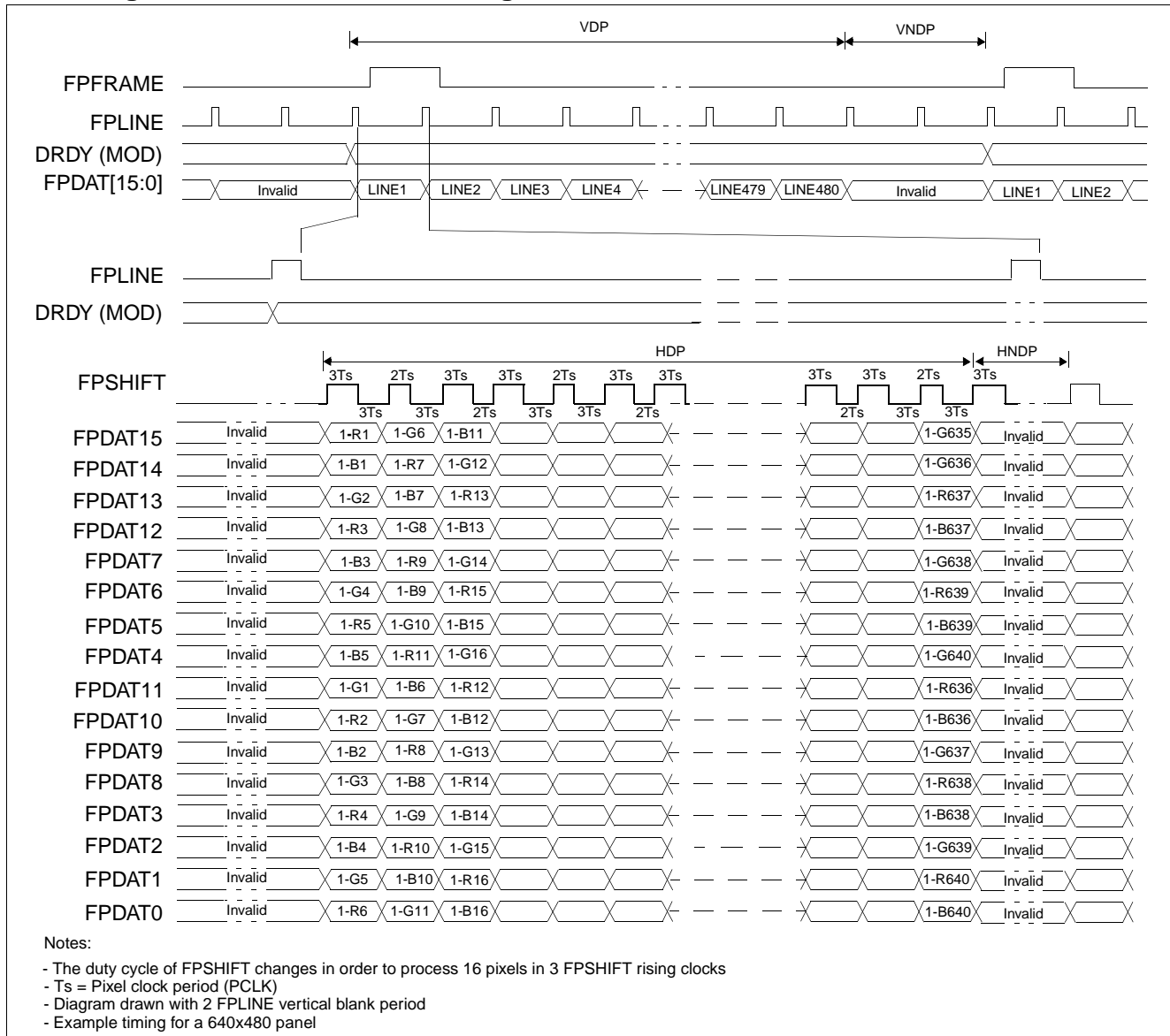


Figure 6-25: Single Color 16-Bit Panel Timing

- VDP = Vertical Display Period
= (REG[34h] bits 9:0) + 1 Lines
- VNDP = Vertical Non-Display Period
= VT - VDP
= (REG[30h] bits 9:0) - (REG[34h] bits 9:0) Lines
- HDP = Horizontal Display Period
= ((REG[24h] bits 6:0) + 1) x 8Ts
- HNDP = Horizontal Non-Display Period
= HT - HDP
= (((REG[20h] bits 6:0) + 1) x 8Ts) - (((REG[24h] bits 6:0) + 1) x 8Ts)

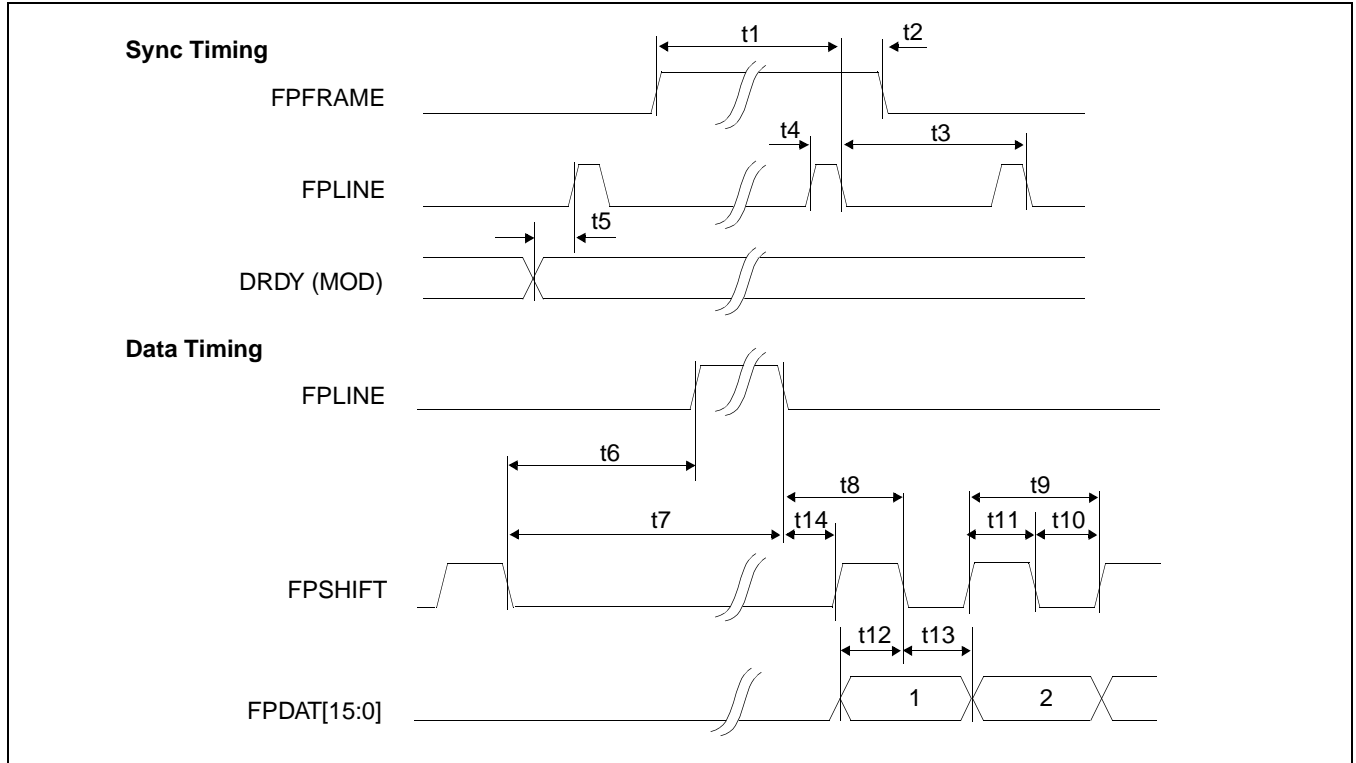


Figure 6-26: Single Color 16-Bit Panel A.C. Timing

Table 6-25: Single Color 16-Bit Panel A.C. Timing

Symbol	Parameter	Min	Typ	Max	Units
t1	FPFRAME setup to FPLINE falling edge	note 2			Ts (note 1)
t2	FPFRAME hold from FPLINE falling edge	note 3			Ts
t3	FPLINE period	note 4			Ts
t4	FPLINE pulse width	note 5			Ts
t5	MOD transition to FPLINE rising edge	note 6			Ts
t6	FPSHIFT falling edge to FPLINE rising edge	note 7			Ts
t7	FPSHIFT falling edge to FPLINE falling edge	t6 + t4			Ts
t8	FPLINE falling edge to FPSHIFT falling edge	t14 + 3			Ts
t9	FPSHIFT period	5			Ts
t10	FPSHIFT pulse width low	2			Ts
t11	FPSHIFT pulse width high	2			Ts
t12	FPDAT[15:0] setup to FPSHIFT rising edge	2			Ts
t13	FPDAT[15:0] hold to FPSHIFT rising edge	2			Ts
t14	FPLINE falling edge to FPSHIFT rising edge	note 8			Ts

1. Ts = pixel clock period
2. $t1_{min} = HPS + t4_{min}$
3. $t2_{min} = t3_{min} - (HPS + t4_{min})$
4. $t3_{min} = HT$
5. $t4_{min} = HPW$
6. $t5_{min} = HPS - 1$
7. $t6_{min} = HPS - (HDP + HDPS) + 2$, if negative add $t3_{min}$
8. $t14_{min} = HDPS - (HPS + t4_{min})$, if negative add $t3_{min}$

6.4.8 Generic TFT Panel Timing

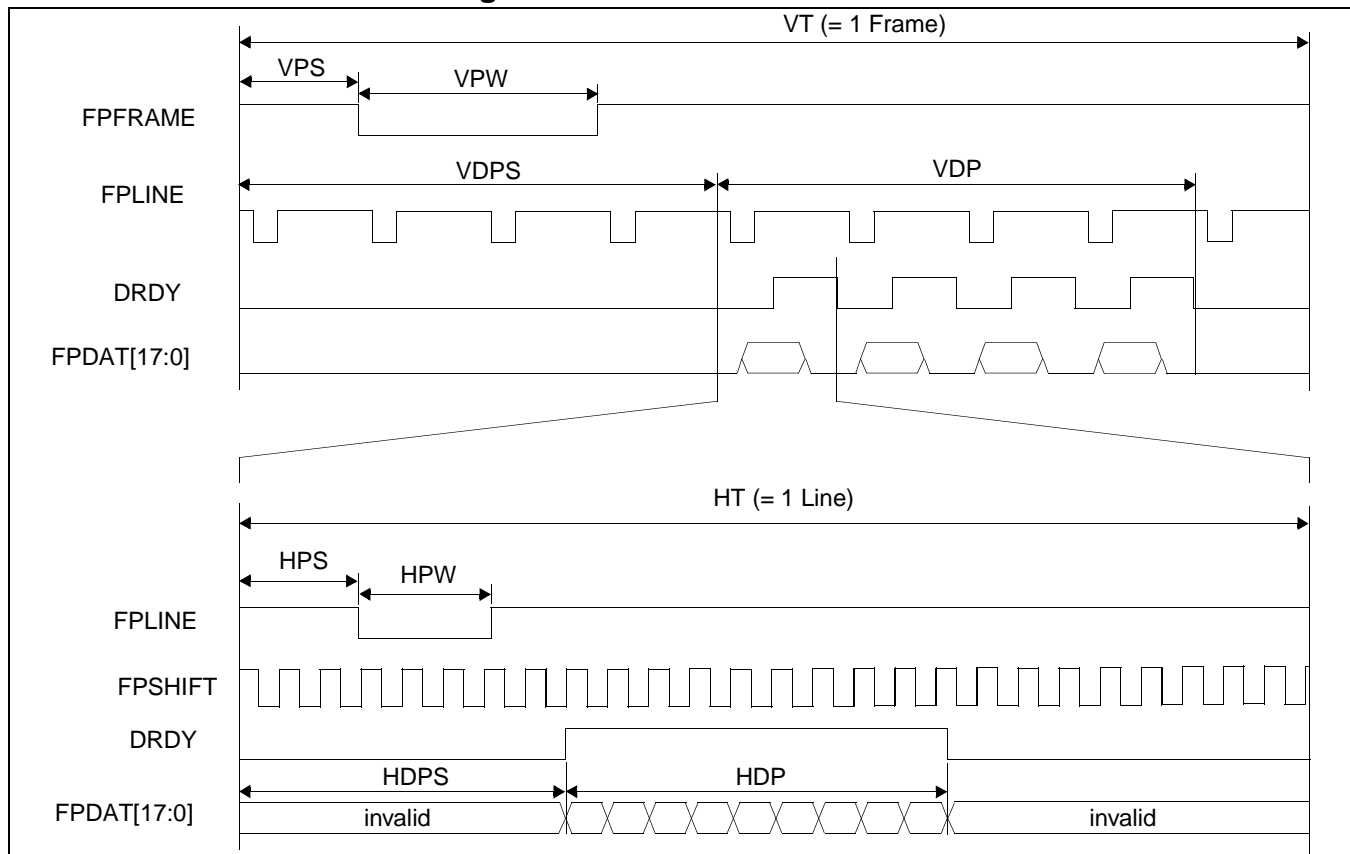


Figure 6-27: Generic TFT Panel Timing

VT	= Vertical Total	= [(REG[30h] bits 9-0) + 1] lines
VPS	= FPPFRAME Pulse Start Position	= (REG[3Ch] bits 9-0) lines
VPW	= FPPFRAME Pulse Width	= [(REG[3Ch] bits 18-16) + 1] lines
VDPS	= Vertical Display Period Start Position	= (REG[38h] bits 9-0) lines
VDP	= Vertical Display Period	= [(REG[34h] bits 9-0) + 1] lines
HT	= Horizontal Total	= [((REG[20h] bits 6-0) + 1) x 8] pixels
HPS	= FPLINE Pulse Start Position	= [(REG[2Ch] bits 9-0) + 1] pixels
HPW	= FPLINE Pulse Width	= [(REG[2Ch] bits 22-16) + 1] pixels
HDPS	= Horizontal Display Period Start Position	= [(REG[28h] bits 9-0) + 5] pixels
HDP	= Horizontal Display Period	= [((REG[24h] bits 6-0) + 1) x 8] pixels

*For TFT panels, the HDP must be a minimum of 8 pixels and must be increased by multiples of 8.

*Panel Type Bits (REG[0Ch] bits 1-0) = 01 (TFT)

*FPLINE Pulse Polarity Bit (REG[2Ch] bit 23) = 0 (active low)

*FPPFRAME Polarity Bit (REG[3Ch] bit 23) = 0 (active low)

6.4.9 9/12/18-Bit TFT Panel Timing

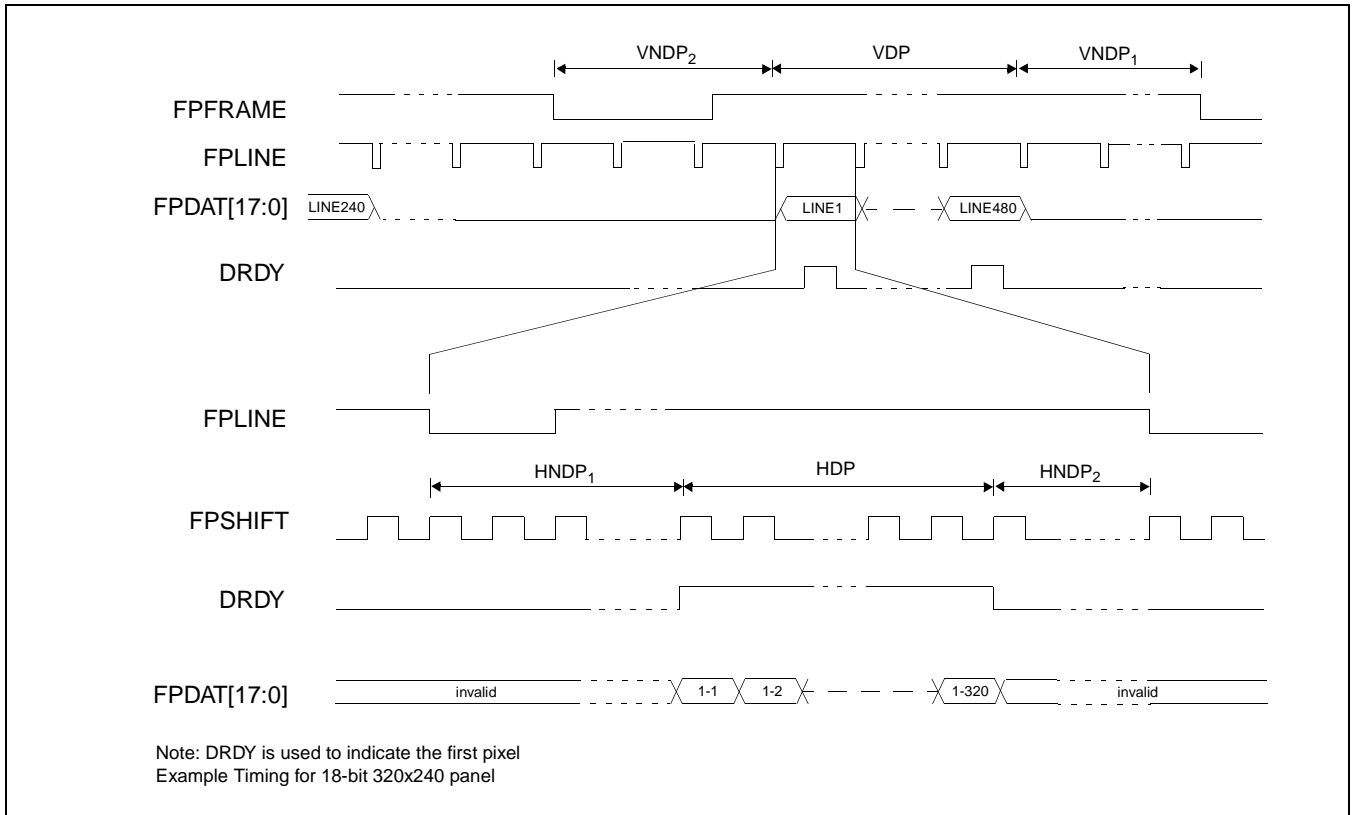


Figure 6-28: 18-Bit TFT Panel Timing

- VDP = Vertical Display Period
= VDP Lines
- VNDP = Vertical Non-Display Period
= VNDP1 + VNDP2
= VT - VDP Lines
- VNDP1 = Vertical Non-Display Period 1
= VNDP - VNDP2 Lines
- VNDP2 = Vertical Non-Display Period 2
= VDPS - VPS Lines if negative add VT
- HDP = Horizontal Display Period
= HDP Ts
- HNDP = Horizontal Non-Display Period
= HNDP1 + HNDP2
= HT - HDP Ts
- HNDP1 = Horizontal Non-Display Period 1
= HDPS - HPS Ts if negative add HT
- HNDP2 = Horizontal Non-Display Period 2
= HPS - (HDP + HDPS) Ts if negative add HT

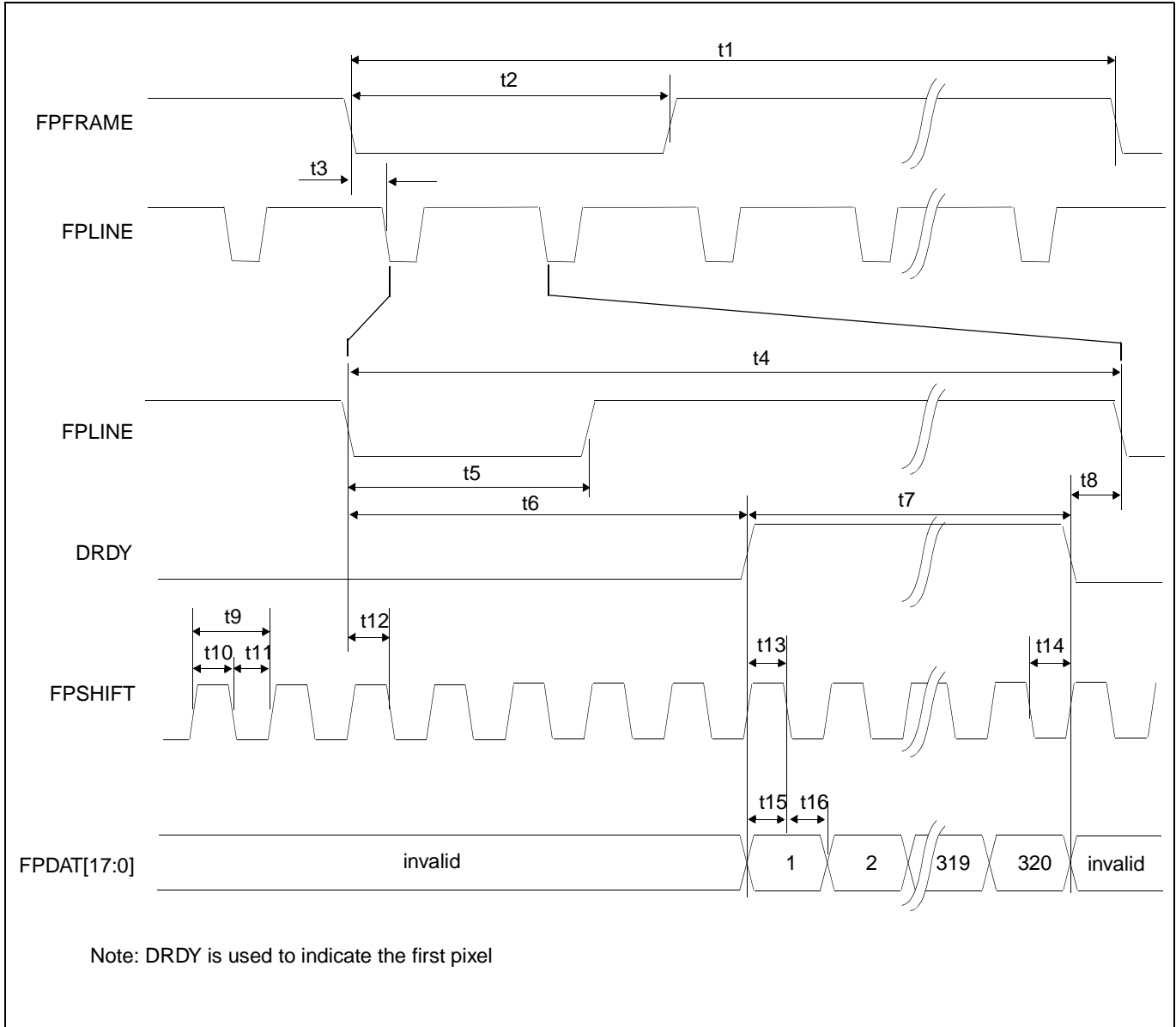


Figure 6-29: TFT A.C. Timing

Table 6-26: TFT A.C. Timing

Symbol	Parameter	Min	Typ	Max	Units
t1	FPFRAME cycle time	VT			Lines
t2	FPFRAME pulse width low	VPW			Lines
t3	FPFRAME falling edge to FPLINE falling edge phase difference	HPS			Ts (note 1)
t4	FPLINE cycle time	HT			Ts
t5	FPLINE pulse width low	HPW			Ts
t6	FPLINE Falling edge to DRDY active	note 2		250	Ts
t7	DRDY pulse width	HDP			Ts
t8	DRDY falling edge to FPLINE falling edge	note 3			Ts
t9	FPSHIFT period	1			Ts
t10	FPSHIFT pulse width high	0.5			Ts
t11	FPSHIFT pulse width low	0.5			Ts
t12	FPLINE setup to FPSHIFT falling edge	0.5			Ts
t13	DRDY to FPSHIFT falling edge setup time	0.5			Ts
t14	DRDY hold from FPSHIFT falling edge	0.5			Ts
t15	Data setup to FPSHIFT falling edge	0.5			Ts
t16	Data hold from FPSHIFT falling edge	0.5			Ts

1. Ts = pixel clock period
2. t6min = HDPS - HPS if negative add HT
3. t8min = HPS - (HDP + HDPS) if negative add HT

6.4.10 160x160 Sharp 'Direct' HR-TFT Panel Timing (e.g. LQ031B1DDxx)

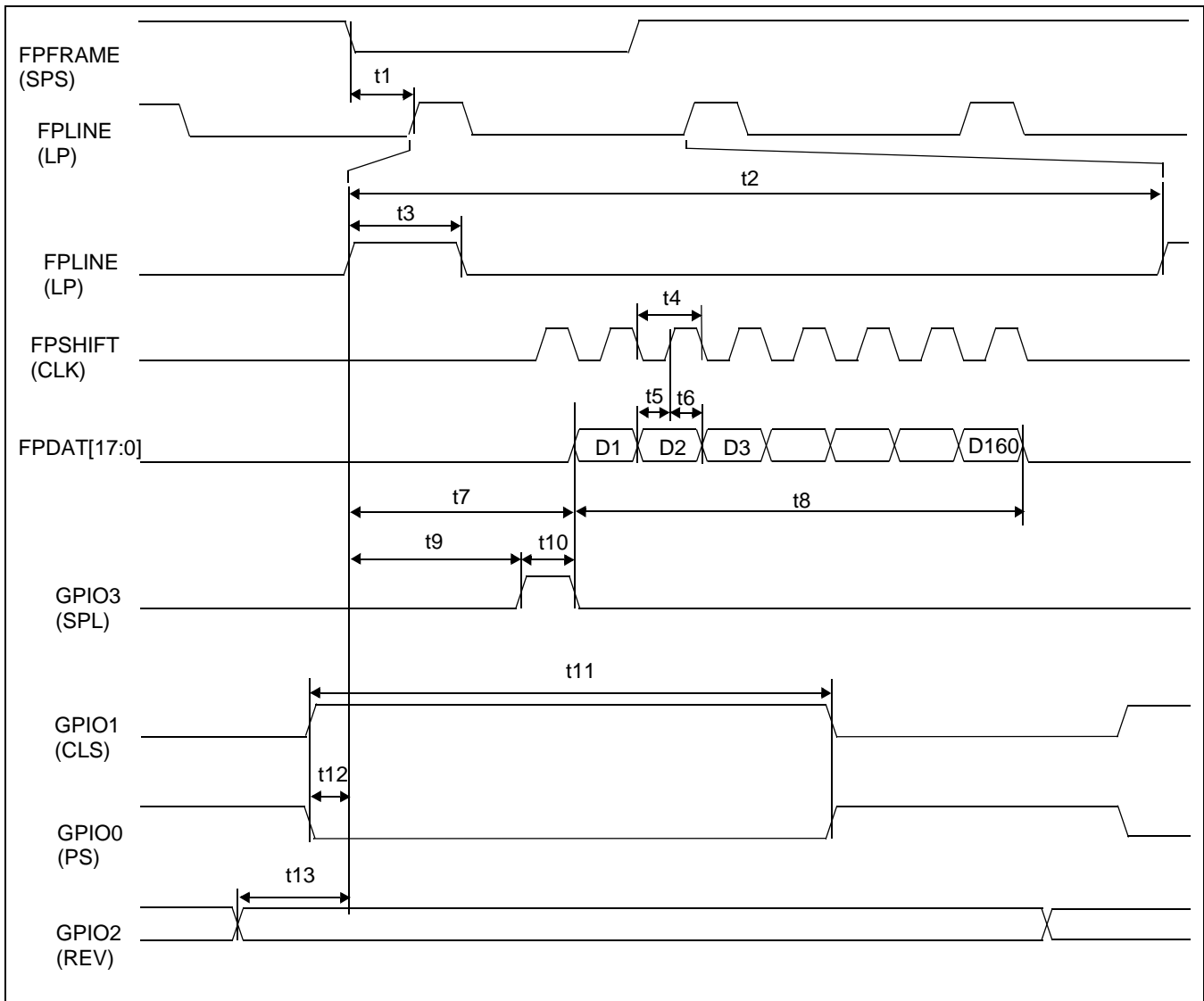


Figure 6-30: 160x160 Sharp 'Direct' HR-TFT Panel Horizontal Timing

Table 6-27: 160x160 Sharp 'Direct' HR-TFT Horizontal Timing

Symbol	Parameter	Min	Typ	Max	Units
t1	FPLINE start position		13		Ts (note 1)
t2	Horizontal total period	180		220	Ts
t3	FPLINE width		2		Ts
t4	FPSHIFT period		1		Ts
t5	Data setup to FPSHIFT rising edge	0.5			Ts
t6	Data hold from FPSHIFT rising edge	0.5			Ts
t7	Horizontal display start position		5		Ts
t8	Horizontal display period		160		Ts
t9	FPLINE rising edge to GPIO3 rising edge		4		Ts
t10	GPIO3 pulse width		1		Ts
t11	GPIO1(GPIO0) pulse width		136		Ts
t12	GPIO1 rising edge (GPIO0 falling edge) to FPLINE rise edge		4		Ts
t13	GPIO2 toggle edge to FPLINE rise edge		10		Ts

1. Ts = pixel clock period
2. t1typ = (REG[2Ch] bits 9-0) + 1
3. t2typ = ((REG[20h] bits 6-0) + 1) x 8
4. t3typ = (REG[2Ch] bits 22-16) + 1
5. t7typ = ((REG[28h] bits 9-0) + 5) - ((REG[2Ch] bits 9-0) + 1)
6. t8typ = ((REG[24h] bits 6-0) + 1) x 8

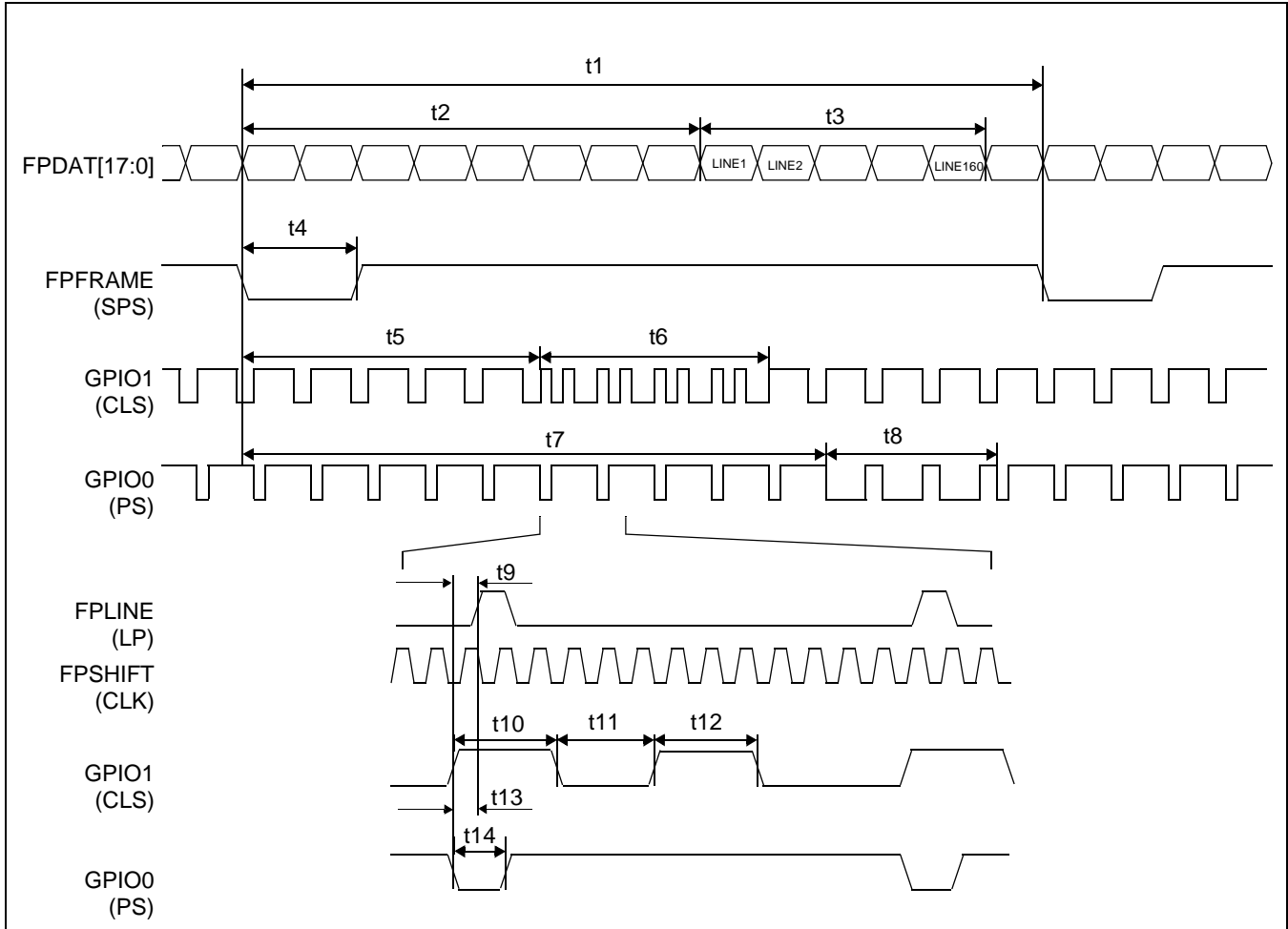


Figure 6-31: 160x160 Sharp 'Direct' HR-TFT Panel Vertical Timing

Table 6-28: 160x160 Sharp 'Direct' HR-TFT Panel Vertical Timing

Symbol	Parameter	Min	Typ	Max	Units
t1	Vertical total period	203		264	Lines
t2	Vertical display start position		40		Lines
t3	Vertical display period		160		Lines
t4	Vertical sync pulse width		2		Lines
t5	FPPFRAME falling edge to GPIO1 alternate timing start		5		Lines
t6	GPIO1 alternate timing period		4		Lines
t7	FPPFRAME falling edge to GPIO0 alternate timing start		40		Lines
t8	GPIO0 alternate timing period		162		Lines
t9	GPIO1 first pulse rising edge to FPLINE rising edge		4		Ts (note 1)
t10	GPIO1 first pulse width		48		Ts
t11	GPIO1 first pulse falling edge to second pulse rising edge		40		Ts
t12	GPIO1 second pulse width		48		Ts
t13	GPIO0 falling edge to FPLINE rising edge		4		Ts
t14	GPIO0 low pulse width		24		Ts

1. Ts = pixel clock period

6.4.11 320x240 Sharp 'Direct' HR-TFT Panel Timing (e.g. LQ039Q2DS01)

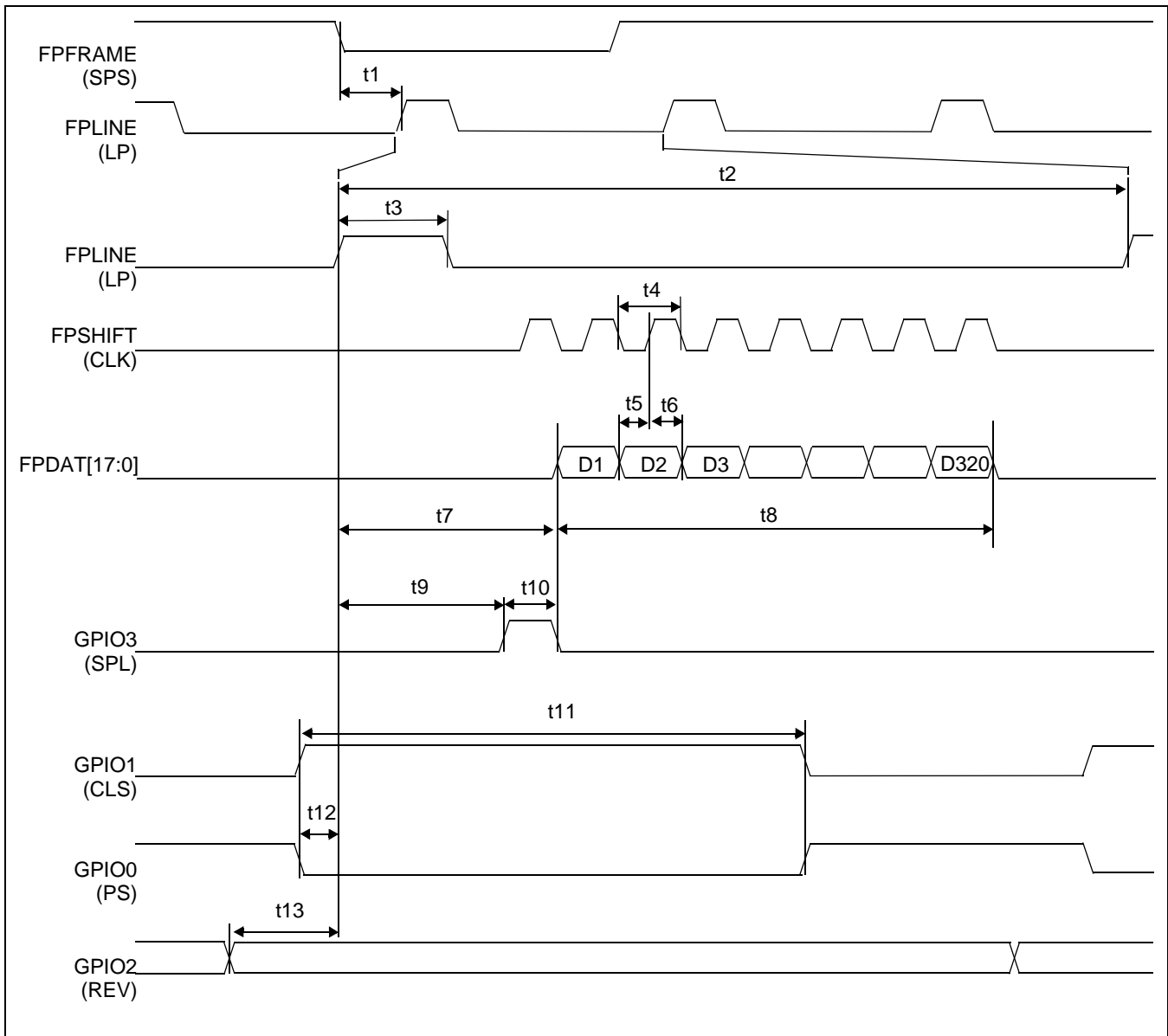


Figure 6-32: 320x240 Sharp 'Direct' HR-TFT Panel Horizontal Timing

Table 6-29: 320x240 Sharp 'Direct' HR-TFT Panel Horizontal Timing

Symbol	Parameter	Min	Typ	Max	Units
t1	FPLINE start position		14		Ts (note 1)
t2	Horizontal total period	400		440	Ts
t3	FPLINE width		1		Ts
t4	FPSHIFT period		1		Ts
t5	Data setup to FPSHIFT rising edge	0.5			Ts
t6	Data hold from FPSHIFT rising edge	0.5			Ts
t7	Horizontal display start position		60		Ts
t8	Horizontal display period		320		Ts
t9	FPLINE rising edge to GPIO3 rising edge		59		Ts
t10	GPIO3 pulse width		1		Ts
t11	GPIO1(GPIO0) pulse width		353		Ts
t12	GPIO1 rising edge (GPIO0 falling edge) to FPLINE rise edge		5		Ts
t13	GPIO2 toggle edge to FPLINE rise edge		11		Ts

1. Ts = pixel clock period
2. t1typ = (REG[2Ch] bits 9-0) + 1
3. t2typ = ((REG[20h] bits 6-0) + 1) x 8
4. t3typ = (REG[2Ch] bits 22-16) + 1
5. t7typ = ((REG[28h] bits 9-0) + 5) - ((REG[2Ch] bits 9-0) + 1)
6. t8typ = ((REG[24h] bits 6-0) + 1) x 8

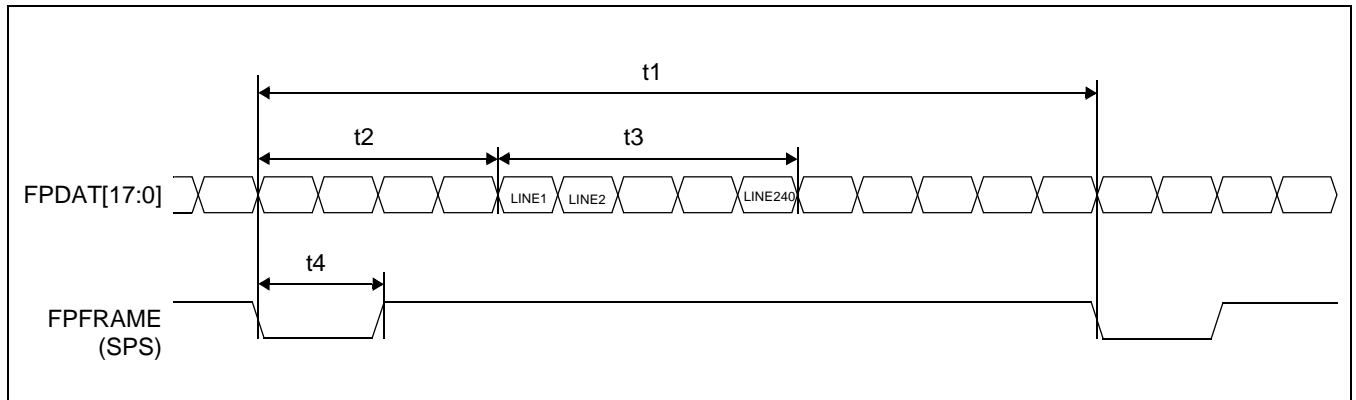


Figure 6-33: 320x240 Sharp 'Direct' HR-TFT Panel Vertical Timing

Table 6-30: 320x240 Sharp 'Direct' HR-TFT Panel Vertical Timing

Symbol	Parameter	Min	Typ	Max	Units
t1	Vertical total period	245		330	Lines
t2	Vertical display start position		4		Lines
t3	Vertical display period		240		Lines
t4	Vertical sync pulse width		2		Lines

6.5 USB Timing

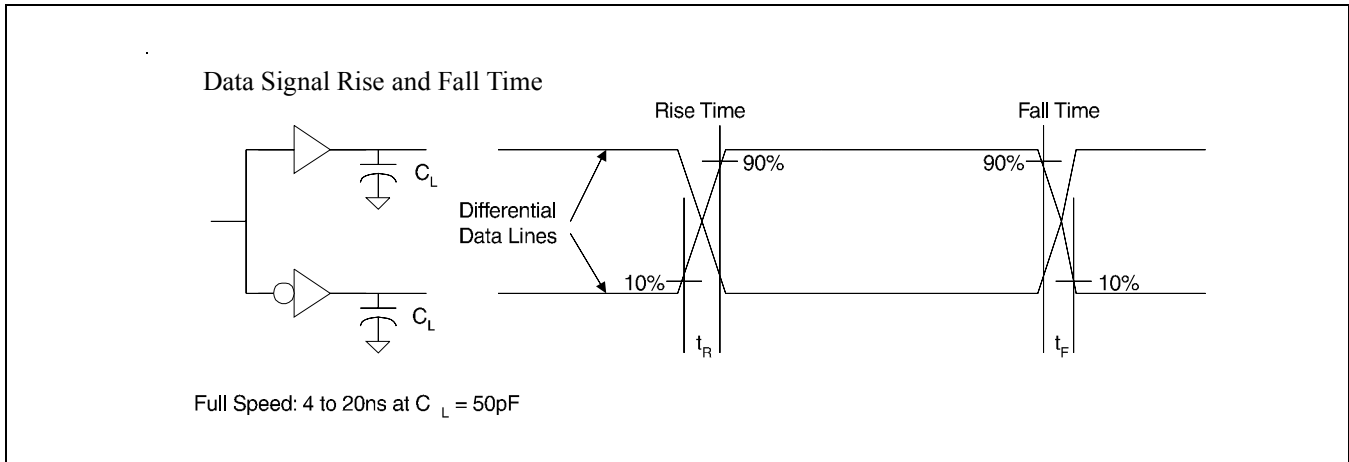


Figure 6-34 Data Signal Rise and Fall Time

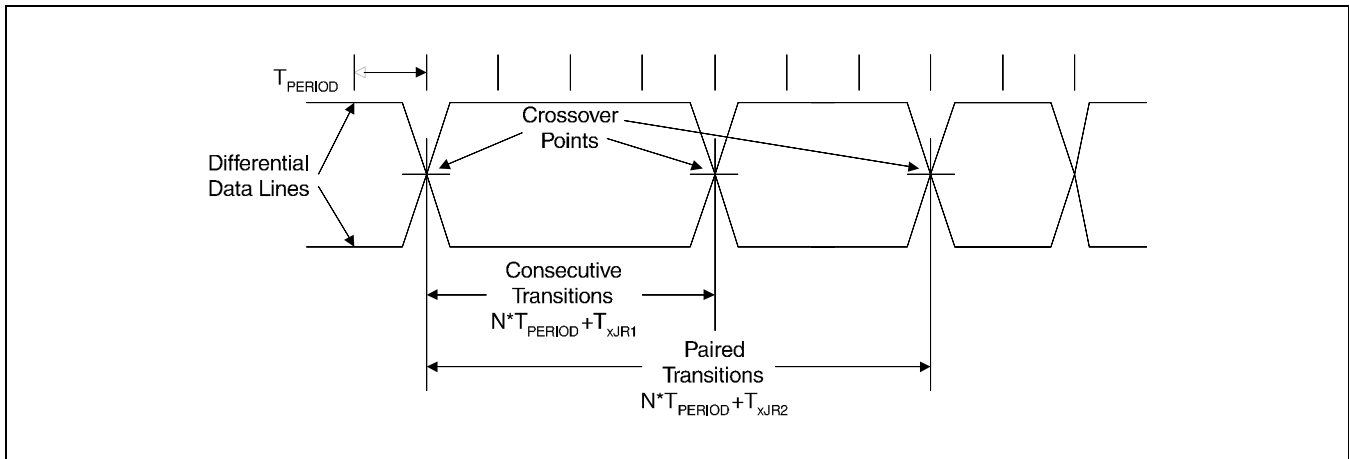


Figure 6-35 Differential Data Jitter

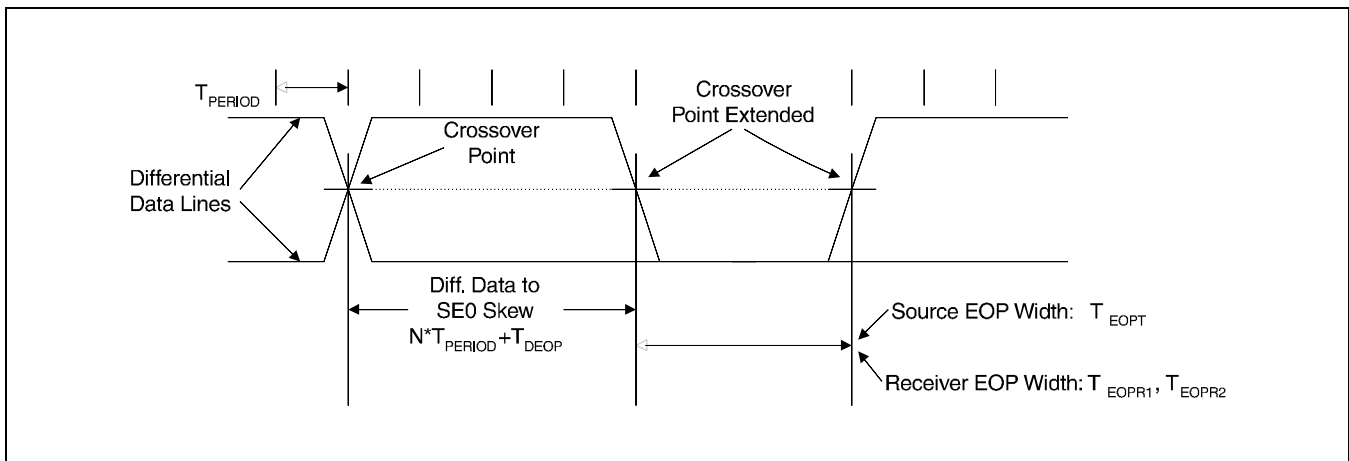


Figure 6-36 Differential to EOP Transition Skew and EOP Width

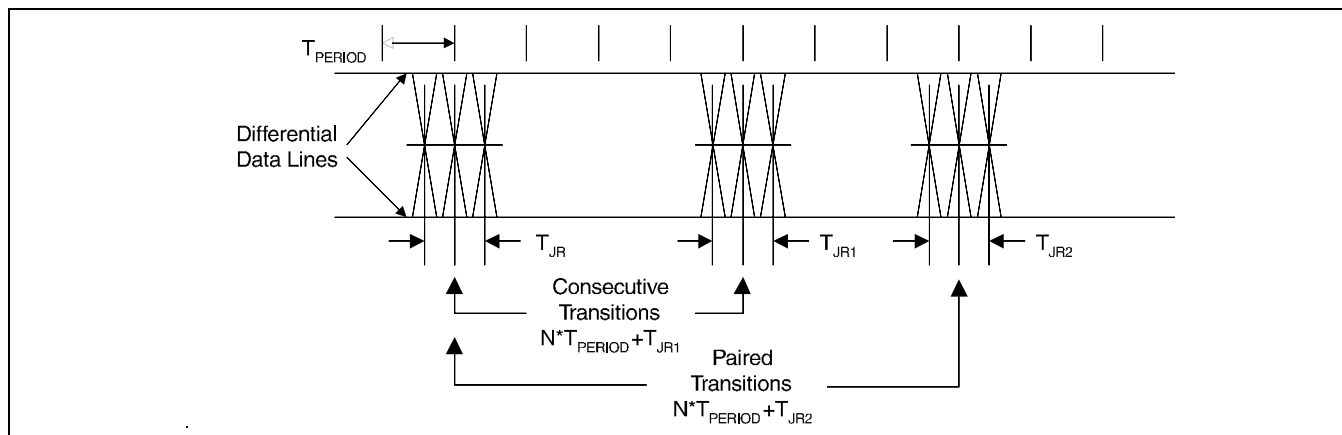


Figure 6-37 Receiver Jitter Tolerance

Table 6-31 USB Interface Timing

Symbol	Parameter	Conditions	Waveform	Min	Typ	Max	Unit
USB _{FREQ}	USB Clock Frequency				48		MHz
T _{PERIOD}	USB Clock Period		Figure 6-34		$\frac{1}{\text{USB}_{\text{FREQ}}}$		
T _R	Rise & Fall Times	C _L = 50 pF Notes 1,2	Figure 6-34	4		20	ns
T _F				4		20	
T _{RFM}	Rise/Fall time matching	(T _R / T _F)	Figure 6-34	90		110	%
V _{CRS}	Output Signal Crossover Voltage			1.3		2.0	V
Z _{DRV}	Driver Output Resistance	Steady State Drive		28 ^{Note 5}		44	Ω
T _{DRATE}	Data Rate			11.97	12	12.03	Mbs
T _{DDJ1}	Source Differential Driver Jitter to Next Transition	Notes 3,4.	Figure 6-35	-3.5	0	3.5	ns
T _{DDJ2}	Source Differential Driver Jitter for Paired Transitions	Notes 3,4	Figure 6-35	-4.0	0	4.0	ns
T _{DEOP}	Differential to EOP Transition Skew	Note 4	Figure 6-36	-2	0	5	ns
T _{EOP1}	Source EOP Width	Note 4	Figure 6-36	160	167	175	ns
T _{JR1}	Receiver Data Jitter Tolerance to Next Transition	Note 4	Figure 6-37	-18.5	0	18.5	ns
T _{JR2}	Receiver Data Jitter Tolerance for Paired Transitions	Note 4	Figure 6-37	-9	0	9	ns
T _{EOPR1}	EOP Width at Receiver; Must reject as EOP	Note 4	Figure 6-36	40			ns
T _{EOPR2}	EOP Width at Receiver; Must accept as EOP	Note 4	Figure 6-36	80			ns

- 1 Measured from 10% to 90% of the data signal.
- 2 The rising and falling edges should be smoothly transitioning (monotonic).
- 3 Timing difference between the differential data signals.
- 4 Measured at crossover point of differential data signals.
- 5 20 Ω is placed in series to meet this USB specification. The actual driver output impedance is 15 Ω.

7 Clocks

7.1 Clock Descriptions

7.1.1 BCLK

BCLK is an internal clock derived from CLKI. BCLK can be a divided version ($\div 1$, $\div 2$) of CLKI. CLKI is typically derived from the host CPU bus clock.

The source clock options for BCLK may be selected as in the following table.

Table 7-1: BCLK Clock Selection

Source Clock Options	BCLK Selection
CLKI	CNF6 = 0
CLKI $\div 2$	CNF6 = 1

Note

For synchronous bus interfaces, it is recommended that BCLK be set the same as the CPU bus clock (not a divided version of CLKI) e.g. SH-3, SH-4.

7.1.2 MCLK

MCLK provides the internal clock required to access the embedded SRAM. The S1D13A04 is designed with efficient power saving control for clocks (clocks are turned off when not used); reducing the frequency of MCLK does not necessarily save more power. Furthermore, reducing the MCLK frequency relative to the BCLK frequency increases the CPU cycle latency and so reduces screen update performance. For a balance of power saving and performance, the MCLK should be configured to have a high enough frequency setting to provide sufficient screen refresh as well as acceptable CPU cycle latency.

Note

The maximum frequency of MCLK is 50MHz (30MHz if running CORE V_{DD} at $2.0V \pm 10\%$). As MCLK is derived from BCLK, when BCLK is greater than 50MHz, MCLK must be divided using REG[04h] bits 5-4.

The Memory Controller Power Save Status bit (REG[14h] bit 6) must return a 1 before disabling the MCLK source.

The source clock options for MCLK may be selected as in the following table.

Table 7-2: MCLK Clock Selection

Source Clock Options	MCLK Selection
BCLK	REG[04h] bits 5-4 = 00
BCLK $\div 2$	REG[04h] bits 5-4 = 01
BCLK $\div 3$	REG[04h] bits 5-4 = 10
BCLK $\div 4$	REG[04h] bits 5-4 = 11

7.1.3 PCLK

PCLK is the internal clock used to control the panel. It should be chosen to match the optimum frame rate of the panel. See Section 10, “Frame Rate Calculation” on page 143 for details on the relationship between PCLK and frame rate.

Some flexibility is possible in the selection of PCLK. Firstly, panels typically have a range of permissible frame rates. Secondly, it may be possible to choose a higher PCLK frequency and tailor the horizontal non-display period to bring down the frame-rate to its optimal value.

The source clock options for PCLK may be selected as in the following table.

Table 7-3: PCLK Clock Selection

Source Clock Options	PCLK Selection
MCLK	REG[08h] bits 7-0 = 00h
MCLK ÷2	REG[08h] bits 7-0 = 10h
MCLK ÷3	REG[08h] bits 7-0 = 20h
MCLK ÷4	REG[08h] bits 7-0 = 30h
MCLK ÷8	REG[08h] bits 7-0 = 40h
BCLK	REG[08h] bits 7-0 = 01h
BCLK ÷2	REG[08h] bits 7-0 = 11h
BCLK ÷3	REG[08h] bits 7-0 = 21h
BCLK ÷4	REG[08h] bits 7-0 = 31h
BCLK ÷8	REG[08h] bits 7-0 = 41h
CLKI	REG[08h] bits 7-0 = 02h
CLKI ÷2	REG[08h] bits 7-0 = 12h
CLKI ÷3	REG[08h] bits 7-0 = 22h
CLKI ÷4	REG[08h] bits 7-0 = 32h
CLKI ÷8	REG[08h] bits 7-0 = 42h
CLKI2	REG[08h] bits 7-0 = 03h
CLKI2 ÷2	REG[08h] bits 7-0 = 13h
CLKI2 ÷3	REG[08h] bits 7-0 = 23h
CLKI2 ÷4	RREG[08h] bits 7-0 = 33h
CLKI2 ÷8	REG[08h] bits 7-0 = 43h

There is a relationship between the frequency of MCLK and PCLK that must be maintained.

Table 7-4: Relationship between MCLK and PCLK

SwivelView Orientation	Color Depth (bpp)	MCLK to PCLK Relationship
SwivelView 0° and 180°	16	$f_{MCLK} \geq f_{PCLK}$
	8	$f_{MCLK} \geq f_{PCLK} \div 2$
	4	$f_{MCLK} \geq f_{PCLK} \div 4$
	2	$f_{MCLK} \geq f_{PCLK} \div 8$
	1	$f_{MCLK} \geq f_{PCLK} \div 16$
SwivelView 90° and 270°	16/8/4/2/1	$f_{MCLK} \geq 1.25f_{PCLK}$

7.1.4 PWMCLK

PWMCLK is the internal clock used by the Pulse Width Modulator for output to the panel.

The source clock options for PWMCLK may be selected as in the following table.

Table 7-5: PWMCLK Clock Selection

Source Clock Options	PWMCLK Selection
CLKI	REG[70h] bits 2-1 = 00
CLKI2	REG[70h] bits 2-1 = 01
MCLK	REG[70h] bits 2-1 = 10
PCLK	REG[70h] bits 2-1 = 11

For further information on controlling PWMCLK, see “PWM Clock Configuration Register” on page 114..

7.1.5 USBCLK

CLKUSB is an internal clock derived from USBCLK and should be fixed at 48MHz. USBCLK must be active to access the USB Registers.

7.2 Clock Selection

The following diagram provides a logical representation of the S1D13A04 internal clocks.

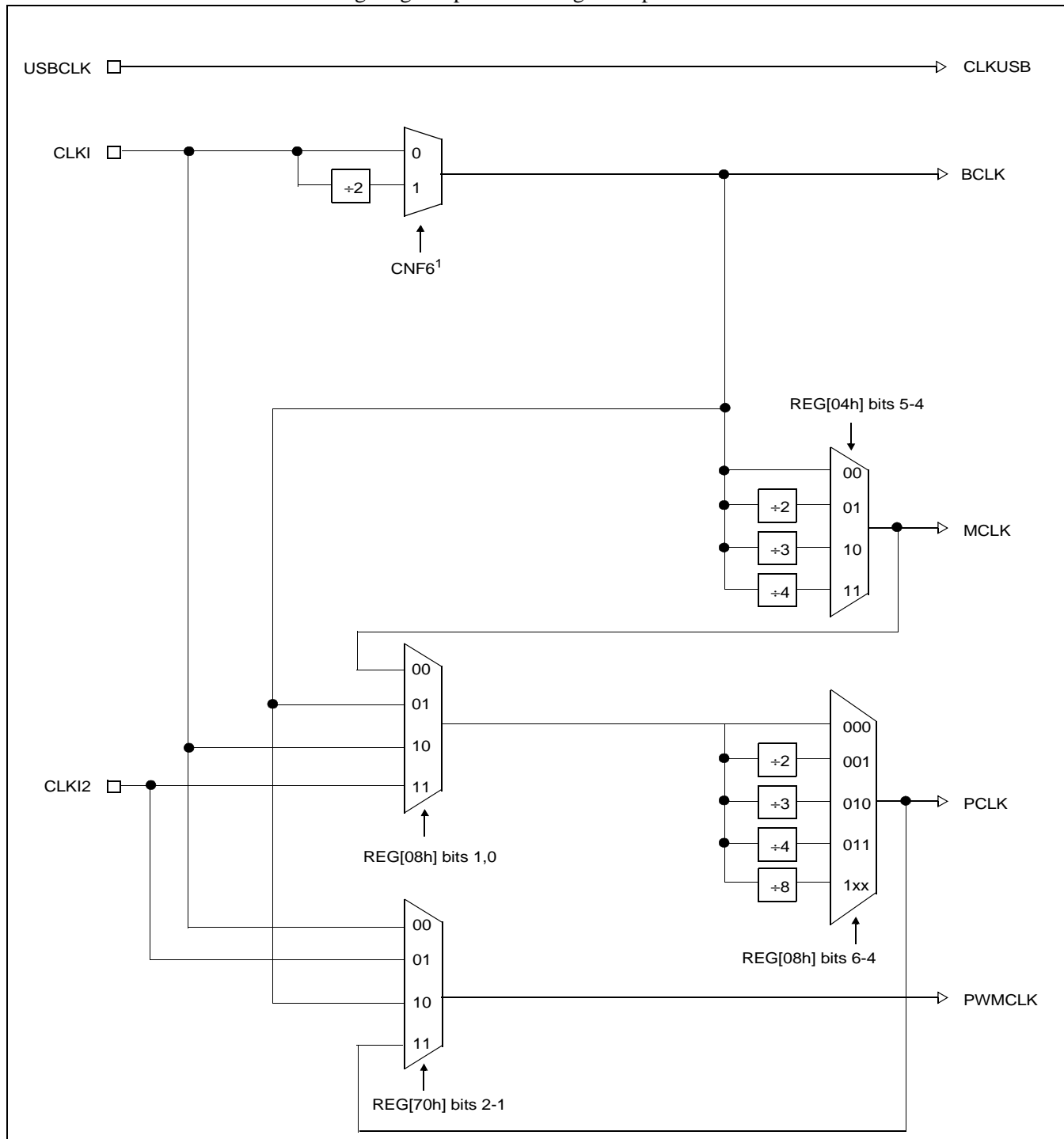


Figure 7-1: Clock Selection

Note
¹ CNF6 must be set at RESET#.

7.3 Clocks versus Functions

Table 7-6: “S1D13A04 Internal Clock Requirements”, lists the internal clocks required for the following S1D13A04 functions.

Table 7-6: S1D13A04 Internal Clock Requirements

Function	Bus Clock (BCLK)	Memory Clock (MCLK)	Pixel Clock (PCLK)	PWM Clock (PWMCLK)	USB Clock (USBCLK)
Register Read/Write	Required	Not Required	Not Required	Not Required ¹	Not Required
Memory Read/Write	Required	Required	Not Required	Not Required ¹	Not Required
Look-Up Table Register Read/Write	Required	Required	Not Required	Not Required ¹	Not Required
Software Power Save	Required	Not Required	Not Required	Not Required ¹	Not Required
LCD Output	Required	Required	Required	Not Required ¹	Not Required
USB Register Read/Write	Required	Not Required	Not Required	Not Required ¹	Required

Note

¹PWMCLK is an optional clock (see Section 7.1.4, “PWMCLK” on page 86).

8 Registers

This section discusses how and where to access the S1D13A04 registers. It also provides detailed information about the layout and usage of each register.

8.1 Register Mapping

The S1D13A04 registers are memory-mapped. When the system decodes the input pins as CS# = 0 and M/R# = 0, the registers may be accessed. The register space is decoded by AB[16:0] and is mapped as follows.

Table 8-1: S1D13A04 Register Mapping

M/R#	Address	Size	Function
1	00000h to 28000h	160K bytes	SRAM memory
0	0000h to 0088h	136 bytes	Configuration registers
0	4000h to 4054h	84 bytes	USB registers
0	8000h to 8019h	25 bytes	2D Acceleration Registers
0	10000h to 1FFFEh	65536 bytes (64K bytes)	2D Accelerator Data Port

8.2 Register Set

The S1D13A04 register set is as follows.

Table 8-2: S1D13A04 Register Set

Register	Pg	Register	Pg
LCD Register Descriptions (Offset = 0h)			
Read-Only Configuration Registers			
REG[00h] Product Information Register	91		
Clock Configuration Registers			
REG[04h] Memory Clock Configuration Register	92	REG[08h] Pixel Clock Configuration Register	92
Panel Configuration Registers			
REG[0Ch] Panel Type & MOD Rate Register	93	REG[10h] Display Settings Register	94
REG[14h] Power Save Configuration Register	96		
Look-Up Table Registers			
REG[18h] Look-Up Table Write Register	98	REG[1Ch] Look-Up Table Read Register	99
Display Mode Registers			
REG[20h] Horizontal Total Register	100	REG[24h] Horizontal Display Period Register	100
REG[28h] Horizontal Display Period Start Position Register	101	REG[2Ch] FPLINE Register	101
REG[30h] Vertical Total Register	102	REG[34h] Vertical Display Period Register	102
REG[38h] Vertical Display Period Start Position Register	103	REG[3Ch] FPFAME Register	103
REG[40h] Main Window Display Start Address Register	104	REG[44h] Main Window Line Address Offset Register	104

Table 8-2: SID13A04 Register Set

Register	Pg	Register	Pg
Picture-in-Picture Plus (PIP⁺) Registers			
REG[50h] PIP ⁺ Window Display Start Address Register	105	REG[54h] PIP ⁺ Window Line Address Offset Register	105
REG[58h] PIP ⁺ Window X Positions Register	106	REG[5Ch] PIP ⁺ Window Y Positions Register	108
Miscellaneous Registers			
REG[60h] Special Purpose Register	110	REG[64h] GPIO Status and Control Register	112
REG[70h] PWM Clock Configuration Register	114	REG[74h] PWMOUT Duty Cycle Register	115
REG[80h] Scratch Pad A Register	116	REG[84h] Scratch Pad B Register	116
REG[88h] Scratch Pad C Register	116		
USB Register Descriptions (Offset = 4000h)			
REG[4000h] Control Register	117	REG[4002h] Interrupt Enable Register 0	118
REG[4004h] Interrupt Status Register 0	119	REG[4006h] Interrupt Enable Register 1	120
REG[4008h] Interrupt Status Register 1	120	REG[4010h] Endpoint 1 Index Register	121
REG[4012h] Endpoint 1 Receive Mailbox Data Register	121	REG[4018h] Endpoint 2 Index Register	121
REG[401Ah] Endpoint 2 Transmit Mailbox Data Register	122	REG[401Ch] Endpoint 2 Interrupt Polling Interval Register	122
REG[4020h] Endpoint 3 Receive FIFO Data Register	122	REG[4022h] Endpoint 3 Receive FIFO Count Register	122
REG[4024h] Endpoint 3 Receive FIFO Status Register	123	REG[4026h] Endpoint 3 Maximum Packet Size Register	123
REG[4028h] Endpoint 4 Transmit FIFO Data Register	123	REG[402Ah] Endpoint 4 Transmit FIFO Count Register	124
REG[402Ch] Endpoint 4 Transmit FIFO Status Register	124	REG[402Eh] Endpoint 4 Maximum Packet Size Register	124
REG[4030h] Endpoint 4 Maximum Packet Size Register	124	REG[4032h] USB Status Register	125
REG[4034h] Frame Counter MSB Register	126	REG[4036h] Frame Counter LSB Register	126
REG[4038h] Extended Register Index	126	REG[403Ah] Extended Register Data	126
REG[403Ah], Index[00h] Vendor ID MSB	126	REG[403Ah], Index[01h] Vendor ID LSB	126
REG[403Ah], Index[02h] Product ID MSB	127	REG[403Ah], Index[03h] Product ID LSB	127
REG[403Ah], Index[04h] Release Number MSB	127	REG[403Ah], Index[05h] Release Number LSB	127
REG[403Ah], Index[06h] Receive FIFO Almost Full Threshold	127	REG[403Ah], Index[07h] Transmit FIFO Almost Empty Threshold	127
REG[403Ah], Index[08h] USB Control	128	REG[403Ah], Index[09h] Maximum Power Consumption	128
REG[403Ah], Index[0Ah] Packet Control	128	REG[403Ah], Index[0Bh] Reserved	129
REG[403Ah], Index[0Ch] FIFO Control	129	REG[4040h] USBFC Input Control Register	130
REG[4042h] Reserved	130	REG[4044h] Pin Input Status / Pin Output Data Register	131
REG[4046h] Interrupt Control Enable Register 0	131	REG[4048h] Interrupt Control Enable Register 1	131
REG[404Ah] Interrupt Control Status/Clear Register 0	132	REG[404Ch] Interrupt Control Status/Clear Register 1	133
REG[404Eh] Interrupt Control Masked Status Register 0	134	REG[4050h] Interrupt Control Masked Status Register 1	134
REG[4052h] USB Software Reset Register	134	REG[4054h] USB Wait State Register	134
2D Acceleration (BitBLT) Register Descriptions (Offset = 8000h)			
REG[8000h] BitBLT Control Register	135	REG[8004h] BitBLT Status Register	136
REG[8008h] BitBLT Command Register	137	REG[800Ch] BitBLT Source Start Address Register	139
REG[8010h] BitBLT Destination Start Address Register	139	REG[8014h] BitBLT Memory Address Offset Register	140
REG[8018h] BitBLT Width Register	140	REG[801Ch] BitBLT Height Register	140
REG[8020h] BitBLT Background Color Register	141	REG[8024h] BitBLT Foreground Color Register	141
2D Acceleration (BitBLT) Data Register Descriptions (Offset = 10000h)			
AB16-AB0 = 10000h-1FFFEh, 2D Accelerator (BitBLT) Data Memory Mapped Region Register			141

8.3 LCD Register Descriptions (Offset = 0h)

Unless specified otherwise, all register bits are set to 0 during power-on.

8.3.1 Read-Only Configuration Registers

Product Information Register															Read Only	
REG[00h]															Default = 2Cxx282Ch	
Product Code bits 5-0						Revision Code bits 1-0		n/a	CNF[6:0] Status							
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	
Display Buffer Size bits 7-0								Product Code bits 5-0						Revision Code bits 1-0		
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	

- bits 31-26 **Product Code**
These read-only bits indicate the product code. The product code is 001011 (0Bh).
- bits 25-24 **Revision Code**
These are read-only bits that indicates the revision code. The revision code is 00.
- bits 22-16 **CNF[6:0] Status**
These read-only status bits return the status of the configuration pins CNF[6:0]. CNF[6:0] are latched at the rising edge of RESET#. For a functional description of each configuration bit (CNF[6:0]), see Section 4.4, “Summary of Configuration Options” on page 30.

Note

CNF3 Status (bit 19) always reads back a 1. The CNF3 pin is reserved and must be set to 1.

- bits 15-8 **Display Buffer Size Bits [7:0]**
This is a read-only register that indicates the size of the SRAM display buffer measured in 4K byte increments. The S1D13A04 display buffer is 160K bytes and therefore this register returns a value of 40 (28h).

Value of this register = display buffer size ÷ 4K bytes
 = 160K bytes ÷ 4K bytes
 = 40 (28h)
- bits 7-2 **Product Code**
These read-only bits indicate the product code. The product code is 001011 (0Bh).
- bits 1-0 **Revision Code**
These are read-only bits that indicates the revision code. The revision code is 00.

8.3.2 Clock Configuration Registers

Memory Clock Configuration Register															
REG[04h] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a										MCLK Divide Select bits 1-0		Reserved			
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 5-4

MCLK Divide Select Bits [1:0]

These bits determine the divide used to generate the Memory Clock (MCLK) from the Bus Clock (BCLK).

Table 8-3: MCLK Divide Selection

MCLK Divide Select Bits	BCLK to MCLK Frequency Ratio
00	1:1
01	2:1
10	3:1
11	4:1

bit 0

Reserved.

This bit must be set to 0.

Pixel Clock Configuration Register																
REG[08h] Default = 00000000h																
Read/Write																
n/a																
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	
n/a										PCLK Divide Select bits 2-0			n/a		PCLK Source Select bits 1-0	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	

bits 6-4

PCLK Divide Select Bits [1:0]

These bits determine the divide used to generate the Pixel Clock (PCLK) from the Pixel Clock Source.

Table 8-4: PCLK Divide Selection

PCLK Divide Select Bits	PCLK Source to PCLK Frequency Ratio
000	1:1
001	2:1
010	3:1
011	4:1
100	8:1
101 - 111	Reserved

bits 1-0

PCLK Source Select Bits [1:0]

These bits determine the source of the Pixel Clock (PCLK).

Table 8-5: PCLK Source Selection

PCLK Source Select Bits	PCLK Source
00	MCLK
01	BCLK
10	CLKI
11	CLKI2

8.3.3 Panel Configuration Registers

Panel Type & MOD Rate Register														Read/Write			
REG[0Ch]														Default = 00000000h			
n/a										MOD Rate bits 5-0							
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
n/a										Panel Data Format Select	Color/Mono Panel Select	Panel Data Width bits 1-0		'Direct' HR-TFT Res Select	n/a	Panel Type bits 1-0	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		

bits 21-16

MOD Rate Bits [5:0]

These bits are for passive LCD panels only.

When these bits are all 0, the MOD output signal (DRDY) toggles every FPFrames.

For a non-zero value *n*, the MOD output signal (DRDY) toggles every *n* FPLines.

bit 7

Panel Data Format Select

When this bit = 0, 8-bit single color passive LCD panel data format 1 is selected. For AC timing see Section 6.4.5, "Single Color 8-Bit Panel Timing (Format 1)" on page 66.

When this bit = 1, 8-bit single color passive LCD panel data format 2 is selected. For AC timing see Section 6.4.6, "Single Color 8-Bit Panel Timing (Format 2)" on page 68.

bit 6

Color/Mono Panel Select

When this bit = 0, a monochrome LCD panel is selected.

When this bit = 1, a color LCD panel is selected.

bits 5-4

Panel Data Width Bits [1:0]

These bits select the data width size of the LCD panel.

Table 8-6: Panel Data Width Selection

Panel Data Width Bits [1:0]	Passive Panel Data Width Size	Active Panel Data Width Size
00	4-bit	9-bit
01	8-bit	12-bit
10	16-bit	18-bit
11	Reserved	Reserved

bit 3 'Direct' HR-TFT Resolution Select
This bit selects one of two panel resolutions when the 'Direct' HR-TFT interface is selected. This bit has no effect for other panel types.

Table 8-7: Active Panel Resolution Selection

'Direct' HR-TFT Resolution Select Bit	HR-TFT Resolution
0	160x160
1	320x240

bits 1-0 Panel Type Bits[1:0]
These bits select the panel type.

Table 8-8: LCD Panel Type Selection

Panel Type Bits [1:0]	Panel Type
00	STN
01	TFT
10	'Direct' HR-TFT
11	Reserved

Display Settings Register												Read/Write			
REG[10h]												Default = 00000000h			
n/a						Pixel Doubling Vertical	Pixel Doubling Horiz.	Display Blank	Dithering Disable	n/a	SW Video Invert	PIP* Window Enable	n/a	SwivelView Mode Select	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a											Bits-per-pixel Select (actual value: 1, 2, 4, 8 or 16 bpp)				
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bit 25 Pixel Doubling Vertical Enable
This bit controls the pixel doubling feature for the vertical dimension or height of the panel (i.e. 160 pixel high data doubled to 320 pixel high panel).
When this bit = 1, pixel doubling in the vertical dimension (height) is enabled.
When this bit = 0, there is no hardware effect.

Note

Pixel Doubling is not designed to support color depths of 1 bit-per-pixel or SwivelView 90° / 270° modes.

bit 24 Pixel Doubling Horizontal Enable
This bit controls the pixel doubling feature for the horizontal dimension or width of the panel (i.e. 160 pixel wide data doubled to 320 pixel wide panel)
When this bit = 1, pixel doubling in the horizontal dimension (width) is enabled.
When this bit = 0, there is no hardware effect.

Note

Pixel Doubling is not designed to support color depths of 1 bit-per-pixel or SwivelView 90° / 270° modes.

- bit 23 Display Blank
When this bit = 0, the LCD display pipeline is enabled.
When this bit = 1, the LCD display pipeline is disabled and all LCD data outputs are forced to zero (i.e., the screen is blanked).
- bit 22 Dithering Disable
When this bit = 0, dithering on the passive LCD panel is enabled, allowing a maximum of 64K colors (2^{18}) or 64 gray shades in 1/2/4/8 bpp mode. In 16bpp mode, only 64K colors (2^{16}) can also be achieved.
When this bit = 1, dithering on the passive LCD panel is disabled, allowing a maximum of 4096 colors (2^{12}) or 16 gray shades.
The dithering algorithm provides more shades of each primary color.

Note

For a summary of the results of dithering for each color depth, see Table 8-10: “LCD Bit-per-pixel Selection,” on page 96.

- bit 20 Software Video Invert
When this bit = 0, video data is normal.
When this bit = 1, video data is inverted.

Note

Video data is inverted after the Look-Up Table

- bit 19 PIP+ Window Enable
This bit enables a PIP+ window within the main window. The location of the PIP+ window within the landscape window is determined by the PIP+ X Position register (REG[58h]) and PIP+ Y Position register (REG[5Ch]). The PIP+ window has its own Display Start Address register (REG[50h]) and Memory Address Offset register (REG[54h]). The PIP+ window shares the same color depth and SwivelView™ orientation as the main window.
When this bit = 1, the PIP+ window is enabled.
When this bit = 0, the PIP+ window is disabled.

- bit 17-16 SwivelView Mode Select Bits [1:0]
These bits select different SwivelView™ orientations:

Table 8-9: SwivelView™ Mode Select Options

SwivelView Mode Select Bits	SwivelView Orientation
00	0° (Normal)
01	90°
10	180°
11	270°

bits 4-0

Bit-per-pixel Select Bits [4:0]

These bits select the color depth (bit-per-pixel) for the displayed data for both the main window and the PIP⁺ window (if active).

1, 2, 4 and 8 bpp modes use the 18-bit LUT, allowing maximum 64K colors. 16 bpp mode bypasses the LUT, allowing only 64K colors.

Table 8-10: LCD Bit-per-pixel Selection

Bit-per-pixel Select Bits [4:0]	Color Depth (bpp)	Maximum Number of Colors/Shades		Max. No. Of Simultaneously Displayed Colors/Shades
		Passive Panel (Dithering On)	TFT Panel	
00000	Reserved			
00001	1 bpp	64K/64	64K/64	2/2
00010	2 bpp	64K/64	64K/64	4/4
00011	Reserved			
00100	4 bpp	64K/64	64K/64	16/16
00101 - 00111	Reserved			
01000	8 bpp	64K/64	64K/64	256/64
10000	16 bpp	64K/64	64K/64	64K/64
10001 - 11111	Reserved			

Power Save Configuration Register														Read/Write				
REG[14h]														Default = 00000010h				
n/a																		
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16			
n/a								VNDP Status (RO)	Memory Power Save Status (RO)	n/a	Power Save Enable	n/a				'Direct' HR-TFT GPO Control		
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0			

bit 7

Vertical Non-Display Period Status (Read-only)

This is a read-only status bit.

When this bit = 0, the LCD panel output is in a Vertical Display Period.

When this bit = 1, the LCD panel output is in a Vertical Non-Display Period.

bit 6

Memory Controller Power Save Status (Read-only)

This read-only status bit indicates the power save state of the memory controller and must be checked before turning off the MCLK source clock.

When this bit = 0, the memory controller is powered up.

When this bit = 1, the memory controller is powered down and the MCLK source can be turned off.

Note

Memory writes are possible during power save mode because the S1D13A04 dynamically enables the memory controller for display buffer writes.

bit 4

Power Save Mode Enable

When this bit = 1, the software initiated power save mode is enabled.

When this bit = 0, the software initiated power save mode is disabled.

At reset, this bit is set to 1. For a summary of Power Save Mode, see Section 15, "Power Save Mode" on page 159.

Note

Memory writes are possible during power save mode because the S1D13A04 dynamically enables the memory controller for display buffer writes.

Power Considerations:

The S1D13A04 may experience higher than normal Quiescent Current immediately after applying power. To prevent this condition, the following start up sequence must be followed:

1. Power-up/Reset the S1D13A04.
2. Initialize all registers.
3. Disable power save mode (set REG[14h] bit 4 to 0)

Note

By default, Power Save Mode is enabled (equal to 1) after power-up/Reset. If it is desirable/necessary to remain in power save mode for any length of time after power-up/Reset, the above described condition can be prevented by performing a R/W access to the embedded memory.

bit 0

'Direct' HR-TFT LCD Interface GPO Control

This bit is for HR-TFT panels only. For all other panel types, this bit has no effect. When the 'direct' HR-TFT LCD interface is selected (REG[0Ch] bits 1-0 = 10), the DRDY pin becomes a general purpose output (GPO). This GPO can be used to control the HR-TFT MOD signal.

When this bit = 0, DRDY (GPO) is forced low.

When this bit = 1, DRDY (GPO) is forced high.

8.3.4 Look-Up Table Registers

Look-Up Table Write Register															
REG[18h] Default = 00000000h												Write Only			
LUT Write Address						LUT Red Write Data						n/a			
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
LUT Green Write Data						n/a		LUT Blue Write Data						n/a	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

Note

The S1D13A04 has three 256-position, 6-bit wide LUTs, one for each of red, green, and blue (see Section 12, “Look-Up Table Architecture” on page 145).

Note

This is a write-only register and returns 00h if read.

bits 31-24

LUT Write Address Bits [7:0]

These bits form a pointer into the Look-Up Table (LUT) which is used to write the LUT Red, Green, and Blue data. **When the S1D13A04 is set to a host bus interface using little endian (CNF4 = 0), the RGB data is updated to the LUT with the completion of a write to these bits.**

Note

When a value is written to the LUT Write Address Bits, the same value is automatically placed in the LUT Read Address Bits (REG[1Ch] bits 31-24).

bits 23-18

LUT Red Write Data Bits [5:0]

These bits contains the data to be written to the red component of the Look-Up Table. The LUT position is controlled by the LUT Write Address bits (bits 31-24).

bits 15-10

LUT Green Write Data Bits [5:0]

These bits contains the data to be written to the green component of the Look-Up Table. The LUT position is controlled by the LUT Write Address bits (bits 31-24).

bits 7-2

LUT Blue Write Data Bits [5:0]

These bits contains the data to be written to the blue component of the Look-Up Table. The LUT position is controlled by the LUT Write Address bits (bits 31-24). **When the S1D13A04 is set to a host bus interface using big endian (CNF4 = 1), the RGB data is updated to the LUT with the completion of a write to these bits.**

Look-Up Table Read Register																	
REG[1Ch] Default = 00000000h								Write Only (bits 31-24)/Read Only									
LUT Read Address (write only)								LUT Red Read Data						n/a			
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
LUT Green Read Data								n/a		LUT Blue Read Data						n/a	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		

Note

The S1D13A04 has three 256-position, 6-bit wide LUTs, one for each of red, green, and blue (see Section 12, “Look-Up Table Architecture” on page 145).

bits 31-24

LUT Read Address Bits [7:0] (Write Only)

This register forms a pointer into the Look-Up Table (LUT) which is used to read LUT data. Red data is read from bits 23-18, green data from bits 15-10, and blue data from bits 7-2.

Note

If a write to the LUT Write Address Bits (REG[18h] bits 31-24) is made, the LUT Read Address bits are automatically updated with the same value.

bits 23-18

LUT Red Read Data Bits [5:0] (Read Only)

These bits point to the data from the red component of the Look-Up Table. The LUT position is controlled by the LUT Read Address bits (bits 31-24). This is a read-only register.

bits 15-10

LUT Green Read Data Bits [5:0] (Read Only)

These bits point to the data from the green component of the Look-Up Table. The LUT position is controlled by the LUT Read Address bits (bits 31-24). This is a read-only register.

bits 7-2

LUT Blue Read Data Bits [5:0] (Read Only)

These bits point to the data from the blue component of the Look-Up Table. The LUT position is controlled by the LUT Read Address bits (bits 31-24). This is a read-only register.

8.3.5 Display Mode Registers

Horizontal Total Register															
REG[20h] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a															
Horizontal Total bits 6-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 6-0

Horizontal Total Bits [6:0]

These bits specify the LCD panel Horizontal Total period, in 8 pixel resolution. The Horizontal Total is the sum of the Horizontal Display period and the Horizontal Non-Display period. Since the maximum Horizontal Total is 1024 pixels, the maximum panel resolution supported is 800x600.

REG[20h] bits 6:0 = (Horizontal Total in number of pixels ÷ 8) - 1

Note

- ¹ For all panels this register must be programmed such that:
 $HDPS + HDP < HT$
 $HT - HDP \geq 8MCLK$
- ² For passive panels, this register must be programmed such that:
 $HPS + HPW < HT$
- ³ See Section 6.4, "Display Interface" on page 56.

Horizontal Display Period Register															
REG[24h] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a															
Horizontal Display Period bits 6-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 6-0

Horizontal Display Period Bits [6:0]

These bits specify the LCD panel Horizontal Display period, in 8 pixel resolution. The Horizontal Display period should be less than the Horizontal Total to allow for a sufficient Horizontal Non-Display period.

REG[24h] bits 6:0 = (Horizontal Display Period in number of pixels ÷ 8) - 1

Note

For passive panels, HDP must be a minimum of 32 pixels and must be increased by multiples of 16.

For TFT panels, HDP must be a minimum of 8 pixels and must be increased by multiples of 8.

Note

See Section 6.4, "Display Interface" on page 56.

Horizontal Display Period Start Position Register															
REG[28h] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						Horizontal Display Period Start Position bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 9-0 Horizontal Display Period Start Position Bits [9:0]
 These bits specify a value used in the calculation of the Horizontal Display Period Start Position (in 1 pixel resolution) for TFT and ‘direct’ HR-TFT panels.

For passive LCD panels these bits must be set to 00h which will result in HDPS = 22.
 $HDPS = (REG[28h] \text{ bits } 9-0) + 22$

For TFT panels, HDPS is calculated using the following formula.
 $HDPS = (REG[28h] \text{ bits } 9-0) + 5$

Note

This register must be programmed such that the following formula is valid.
 $HDPS + HDP < HT$

FPLINE Register															
REG[2Ch] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	FPLINE Polarity 23	22	21	20	19	18	17	16
n/a						FPLINE Pulse Start Position bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bit 23 FPLINE Pulse Polarity
 This bit selects the polarity of the horizontal sync signal. For passive panels, this bit must be set to 1. For TFT panels, this bit is set according to the horizontal sync signal of the panel (typically FPLINE or LP).
 When this bit = 0, the horizontal sync signal is active low.
 When this bit = 1, the horizontal sync signal is active high.

bits 22-16 FPLINE Pulse Width Bits [6:0]
 These bits specify the width of the panel horizontal sync signal, in 1 pixel resolution. The horizontal sync signal is typically FPLINE or LP, depending on the panel type.
 $REG[2Ch] \text{ bits } 22-16 = \text{FPLINE Pulse Width in number of pixels} - 1$

Note

For passive panels, these bits must be programmed such that the following formula is valid.
 $HPW + HPS < HT$

Note

See Section 6.4, “Display Interface” on page 56.

bits 9-0 FPLINE Pulse Start Position Bits [9:0]
These bits specify the start position of the horizontal sync signal, in 1 pixel resolution.
FPLINE Pulse Start Position in pixels = (REG[2Ch] bits 9-0) + 1

Note

For passive panels, these bits must be programmed such that the following formula is valid.

$$HPW + HPS < HT$$

Note

See Section 6.4, “Display Interface” on page 56.

Vertical Total Register															
REG[30h] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						Vertical Total bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 9-0 Vertical Total Bits [9:0]
These bits specify the LCD panel Vertical Total period, in 1 line resolution. The Vertical Total is the sum of the Vertical Display Period and the Vertical Non-Display Period. The maximum Vertical Total is 1024 lines.
REG[30h] bits 9:0 = Vertical Total in number of lines - 1

Note

¹ This register must be programmed such that the following formula is valid.

$$VDPS + VDP < VT$$

² See Section 6.4, “Display Interface” on page 56.

Vertical Display Period Register															
REG[34h] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						Vertical Display Period bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 9-0 Vertical Display Period Bits [9:0]
These bits specify the LCD panel Vertical Display period, in 1 line resolution. The Vertical Display period should be less than the Vertical Total to allow for a sufficient Vertical Non-Display period.
REG[34h] bits 9:0 = Vertical Display Period in number of lines - 1

Note

¹ This register must be programmed such that the following formula is valid.

$$VDPS + VDP < VT$$

² See Section 6.4, “Display Interface” on page 56.

Vertical Display Period Start Position Register															
REG[38h] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						Vertical Display Period Start Position bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 9-0 Vertical Display Period Start Position Bits [9:0]
 These bits specify the Vertical Display Period Start Position for panels in 1 line resolution.

For passive LCD panels these bits must be set to 00h.

For TFT panels, VDPS is calculated using the following formula.

$$VDPS = REG[38h] \text{ bits 9-0}$$

Note

¹This register must be programmed such that the following formula is valid.

$$VDPS + VDP < VT$$

²See Section 6.4, “Display Interface” on page 56.

FPFRAME Register																
REG[3Ch] Default = 00000000h																
Read/Write																
n/a								FPFRAME Polarity	n/a				FPFRAME Pulse Width bits 2-0			
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	
n/a						FPFRAME Pulse Start Position bits 9-0										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	

bit 23 FPFRAME Pulse Polarity
 This bit selects the polarity of the vertical sync signal. For passive panels, this bit must be set to 1. For TFT panels, this bit is set according to the horizontal sync signal of the panel (typically FPFRAME, SPS or DY).

When this bit = 0, the vertical sync signal is active low.

When this bit = 1, the vertical sync signal is active high.

bits 18-16 FPFRAME Pulse Width Bits [2:0]
 These bits specify the width of the panel vertical sync signal, in 1 line resolution. The vertical sync signal is typically FPFRAME, SPS or DY, depending on the panel type.

$$REG[3Ch] \text{ bits 18-16} = \text{FPFRAME Pulse Width in number of lines} - 1$$

Note

See Section 6.4, “Display Interface” on page 56.

bits 9-0 FPFRAME Pulse Start Position Bits [9:0]
These bits specify the start position of the vertical sync signal, in 1 line resolution.

For passive panels, these bits must be set to 00h.

For TFT panels, VDPS is calculated using the following formula.

$$\text{VPS} = \text{REG}[3\text{Ch}] \text{ bits } 9\text{-}0$$

Note

See Section 6.4, “Display Interface” on page 56.

Main Window Display Start Address Register															Read/Write
REG[40h]															Default = 00000000h
n/a															bit 16
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
Main Window Display Start Address bits 15-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 16-0 Main Window Display Start Address Bits [16:0]
This register specifies the starting address, in DWORDS, for the LCD image in the display buffer for the main window.

Note that this is a double-word (32-bit) address. An entry of 00000h into these registers represents the first double-word of display memory, an entry of 00001h represents the second double-word of the display memory, and so on. Calculate the Display Start Address as follows:

$$\text{REG}[40\text{h}] \text{ bits } 16\text{:}0 = \text{image address} \div 4 \text{ (valid only for SwivelView } 0^\circ\text{)}$$

Note

For information on setting this register for other SwivelView orientations, see Section 13, “SwivelView™” on page 151.

Main Window Line Address Offset Register															Read/Write
REG[44h]															Default = 00000000h
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						Main Window Line Address Offset bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 9-0 Main Window Line Address Offset Bits [9:0]
This register specifies the offset, in DWORDS, from the beginning of one display line to the beginning of the next display line in the main window. **Note that this is a 32-bit address increment.** Calculate the Line Address Offset as follows:

$$\text{REG}[44\text{h}] \text{ bits } 9\text{:}0 = \text{display width in pixels} \div (32 \div \text{bpp})$$

Note

A virtual display can be created by programming this register with a value greater than the formula requires. When a virtual display is created the image width is larger than the display width and the displayed image becomes a window into the larger virtual image.

8.3.6 Picture-in-Picture Plus (PIP⁺) Registers

PIP ⁺ Display Start Address Register															Read/Write
REG[50h]															Default = 00000000h
n/a															bit 16
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
PIP ⁺ Display Start Address bits 15-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 16-0

PIP⁺ Display Start Address Bits [16:0]

These bits form the 17-bit address for the starting double-word of the PIP⁺ window.

Note that this is a double-word (32-bit) address. An entry of 00000h into these registers represents the first double-word of display memory, an entry of 00001h represents the second double-word of the display memory, and so on.

Note

These bits have no effect unless the PIP⁺ Window Enable bit is set to 1 (REG[10h] bit 19).

PIP ⁺ Line Address Offset Register															Read/Write
REG[54h]															Default = 00000000h
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						PIP ⁺ Line Address Offset bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 9-0

PIP⁺ Window Line Address Offset Bits [9:0]

These bits are the LCD display's 10-bit address offset from the starting double-word of line "n" to the starting double-word of line "n + 1" for the PIP⁺ window. **Note that this is a 32-bit address increment.**

Note

These bits have no effect unless the PIP⁺ Window Enable bit is set to 1 (REG[10h] bit 19).

PIP ⁺ X Positions Register															
REG[58h] Default = 00000000h															
Read/Write															
n/a						PIP ⁺ X End Position bits 9-0									
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						PIP ⁺ X Start Position bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

Note

The effect of REG[58h] through REG[5Ch] takes place only after REG[5Ch] is written and at the next vertical non-display period.

bits 25-16

PIP⁺ Window X End Position Bits [9:0]

These bits determine the X end position of the PIP⁺ window in relation to the origin of the panel. Due to the S1D13A04 SwivelView feature, the X end position may not be a horizontal position value (only true in 0° and 180° SwivelView). For further information on defining the value of the X End Position register, see Section 14, “Picture-in-Picture Plus (PIP+)” on page 156.

The register is also incremented differently based on the SwivelView orientation. For 0° and 180° SwivelView the X end position is incremented by x pixels where x is relative to the current color depth.

Table 8-11: 32-bit Address Increments for Color Depth

Color Depth	Pixel Increment (x)
1 bpp	32
2 bpp	16
4 bpp	8
8 bpp	4
16 bpp	2

For 90° and 270° SwivelView the X end position is incremented in 1 line increments.

Depending on the color depth, some of the higher bits in this register are unused because the maximum horizontal display width is 1024 pixels.

Note

These bits have no effect unless the PIP⁺ Window Enable bit is set to 1 (REG[10h] bit 19).

bits 9-0

PIP⁺ Window X Start Position Bits [9:0]

These bits determine the X start position of the PIP⁺ window in relation to the origin of the panel. Due to the S1D13A04 SwivelView feature, the X start position may not be a horizontal position value (only true in 0° and 180° SwivelView). For further information on defining the value of the X Start Position register, see Section 14, “Picture-in-Picture Plus (PIP+)” on page 156.

The register is also incremented differently based on the SwivelView orientation. For 0° and 180° SwivelView the X start position is incremented by x pixels where x is relative to the current color depth.

Table 8-12: 32-bit Address Increments for Color Depth

Color Depth	Pixel Increment (x)
1 bpp	32
2 bpp	16
4 bpp	8
8 bpp	4
16 bpp	2

For 90° and 270° SwivelView the X start position is incremented in 1 line increments.

Depending on the color depth, some of the higher bits in this register are unused because the maximum horizontal display width is 1024 pixels.

Note

These bits have no effect unless the PIP⁺ Window Enable bit is set to 1 (REG[10h] bit 19).

PIP ⁺ Y Positions Register															
REG[5Ch] Default = 00000000h															
Read/Write															
n/a						PIP ⁺ Y End Position bits 9-0									
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						PIP ⁺ Y Start Position bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

Note

- ¹ The effect of REG[58h] through REG[5Ch] takes place only after REG[5Ch] is written and at the next vertical non-display period.
- ² For host bus interfaces using little endian (CNF4 = 0), a write to bits 31-24 causes the PIP⁺ Window Y End Position to take effect.
For host bus interfaces using big endian (CNF4 = 1), a write to bits 7-0 causes the PIP⁺ Window Y End Position to take effect.

bits 25-16

PIP⁺ Window Y End Position Bits [9:0]

These bits determine the Y end position of the PIP⁺ window in relation to the origin of the panel. Due to the S1D13A04 SwivelView feature, the Y end position may not be a vertical position value (only true in 0° and 180° SwivelView). For further information on defining the value of the Y End Position register, see Section 14, “Picture-in-Picture Plus (PIP+)” on page 156.

The register is also incremented differently based on the SwivelView orientation. For 0° and 180° SwivelView the Y end position is incremented in 1 line increments. For 90° and 270° SwivelView the Y end position is incremented by *y* pixels where *y* is relative to the current color depth.

Table 8-13: 32-bit Address Increments for Color Depth

Color Depth	Pixel Increment (y)
1 bpp	32
2 bpp	16
4 bpp	8
8 bpp	4
16 bpp	2

Depending on the color depth, some of the higher bits in this register are unused because the maximum vertical display height is 1024 pixels.

Note

These bits have no effect unless the PIP⁺ Window Enable bit is set to 1 (REG[10h] bit 19).

bits 9-0

PIP⁺ Window Y Start Position Bits [9:0]

These bits determine the Y start position of the PIP⁺ window in relation to the origin of the panel. Due to the S1D13A04 SwivelView feature, the Y start position may not be a vertical position value (only true in 0° and 180° SwivelView). For further information on defining the value of the Y Start Position register, see Section 14, “Picture-in-Picture Plus (PIP+)” on page 156.

The register is also incremented differently based on the SwivelView orientation. For 0° and 180° SwivelView the Y start position is incremented in 1 line increments. For 90° and 270° SwivelView the Y start position is incremented by *y* pixels where *y* is relative to the current color depth.

Table 8-14: 32-bit Address Increments for Color Depth

Color Depth	Pixel Increment (y)
1 bpp	32
2 bpp	16
4 bpp	8
8 bpp	4
16 bpp	2

Depending on the color depth, some of the higher bits in this register are unused because the maximum vertical display height is 1024 pixels.

Note

These bits have no effect unless the PIP⁺ Window Enable bit is set to 1 (REG[10h] bit 19).

8.3.7 Miscellaneous Registers

Special Purpose Register															
REG[60h] Default = 00000000h															
n/a								Reserved						Read/Write	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a								2D Byte Swap	Display Data Word Swap	Display Data Byte Swap	n/a		Latch Byte Select	n/a	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 23-16 Reserved.
These bits must be set to 0.

bit 7 2D Byte Swap
This bit enables the word data sent to/read from the 2D BitBLT port to be swapped (byte 0 and byte 1 are swapped).

Note

This bit is only used when the S1D13A04 is configured for Big Endian (CNF4 = 1 at RESET#). If configured for Little Endian (CNF4 = 0), this bit has no effect.

bit 6 Display Data Word Swap
The display pipe fetches 32-bits of data from the display buffer. This bit enables the lower 16-bit word and the upper 16-bit word to be swapped before sending them to the LCD display. If the Display Data Byte Swap bit is also enabled, then the byte order of the fetched 32-bit data is reversed.

Note

This bit is only used when the S1D13A04 is configured for Big Endian (CNF4 = 1 at RESET#). If configured for Little Endian (CNF4 = 0), this bit has no effect.

bit 5

Display Data Byte Swap

The display pipe fetches 32-bit of data from the display buffer. This bit enables byte 0 and byte 1 to be swapped, and byte 2 and byte 3 to be swapped, before sending them to the LCD display. If the Display Data Word Swap bit is also enabled, then the byte order of the fetched 32-bit data is reversed.

Note

This bit is only used when the S1D13A04 is configured for Big Endian (CNF4 = 1 at RESET#). If configured for Little Endian (CNF4 = 0), this bit has no effect.

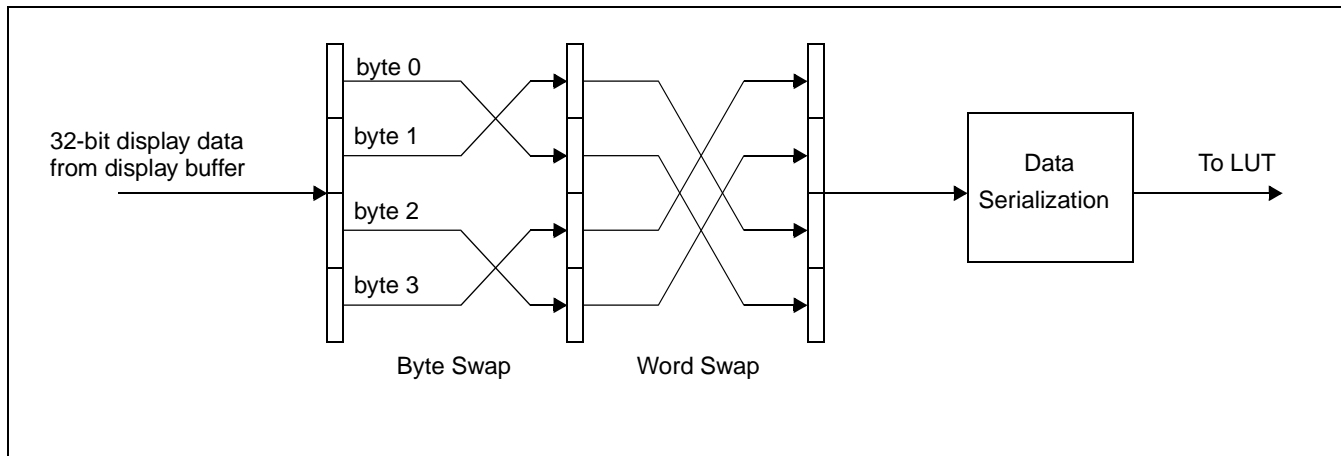


Figure 8-1: Display Data Byte/Word Swap

bit 2

Latch Byte Select

When this bit = 1, REG[5Ch] is latched in reverse order.
When this bit = 0, there is no hardware effect.

GPIO Status and Control Register																	
REG[64h]								Default = 20000000h								Read/Write	
GPIO7 Input Enable	GPIO6 Input Enable	GPIO5 Input Enable	GPIO4 Input Enable	GPIO3 Input Enable	GPIO2 Input Enable	GPIO1 Input Enable	GPIO0 Input Enable	GPIO7 Config	GPIO6 Config	GPIO5 Config	GPIO4 Config	GPIO3 Config	GPIO2 Config	GPIO1 Config	GPIO0 Config		
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
n/a								GPIO7 Status	GPIO6 Status	GPIO5 Status	GPIO4 Status	GPIO3 Status	GPIO2 Status	GPIO1 Status	GPIO0 Status		
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		

Note

The GPIO pins default as inputs after power-up and must be configured using the bits in this register.

Note

For information on GPIO pin mapping when the 'direct' HR-TFT LCD interface is selected, see Table 4-9: "LCD Interface Pin Mapping," on page 32.

bits 31-24

GPIO[7:0] Pin Input Enable bits

These bits are used to enable the input function of each GPIO pin. They must be changed to a 1 after power-on reset to enable the input function of the corresponding GPIO pin (default is 0).

Note

The default for GPIO5 Pin Input Enable is 1.

bits 23-16

GPIO[7:0] Pin IO Configuration

When this bit = 0 (default), the corresponding GPIO pin is configured as an input pin.
When this bit = 1, the corresponding GPIO pin is configured as an output pin.

Note

The input function of each GPIO pin must be enabled using the GPIO[7:0] Pin Input Enable bits (bits 31-24) before the input configuration takes effect.

bit 7

GPIO7 Pin IO Status

When GPIO7 is configured as an output, writing a 1 to this bit drives GPIO7 high and writing a 0 to this bit drives GPIO7 low.
When GPIO7 is configured as an input, a read from this bit returns the status of GPIO7.

bit 6

GPIO6 Pin IO Status

When GPIO6 is configured as an output, writing a 1 to this bit drives GPIO6 high and writing a 0 to this bit drives GPIO6 low.
When GPIO6 is configured as an input, a read from this bit returns the status of GPIO6.

bit 5

GPIO5 Pin IO Status

When GPIO5 is configured as an output, writing a 1 to this bit drives GPIO5 high and writing a 0 to this bit drives GPIO5 low.
When GPIO5 is configured as an input, a read from this bit returns the status of GPIO5.

- bit 4 GPIO4 Pin IO Status
When GPIO4 is configured as an output, writing a 1 to this bit drives GPIO4 high and writing a 0 to this bit drives GPIO4 low.
When GPIO4 is configured as an input, a read from this bit returns the status of GPIO4.
- bit 3 GPIO3 Pin IO Status
When the 'Direct' HR-TFT LCD interface is not selected (REG[0Ch] bits 1:0) and GPIO3 is configured as an output, writing a 1 to this bit drives GPIO3 high and writing a 0 to this bit drives GPIO3 low.
When the 'Direct' HR-TFT LCD interface is not selected (REG[0Ch] bits 1:0) and GPIO3 is configured as an input, a read from this bit returns the status of GPIO3.
- When the 'Direct' HR-TFT LCD interface is enabled (REG[0Ch] bits 1:0 = 10), GPIO0 outputs the SPL signal automatically and writing to this bit has no effect.
- bit 2 GPIO2 Pin IO Status
When the 'Direct' HR-TFT LCD interface is not selected (REG[0Ch] bits 1:0) and GPIO2 is configured as an output, writing a 1 to this bit drives GPIO2 high and writing a 0 to this bit drives GPIO2 low.
When the 'Direct' HR-TFT LCD interface is not selected (REG[0Ch] bits 1:0) and GPIO2 is configured as an input, a read from this bit returns the status of GPIO2.
- When the 'Direct' HR-TFT LCD interface is enabled (REG[0Ch] bits 1:0 = 10), GPIO0 outputs the REV signal automatically and writing to this bit has no effect.
- bit 1 GPIO1 Pin IO Status
When the 'Direct' HR-TFT LCD interface is not selected (REG[0Ch] bits 1:0) and GPIO1 is configured as an output, writing a 1 to this bit drives GPIO1 high and writing a 0 to this bit drives GPIO1 low.
When the 'Direct' HR-TFT LCD interface is not selected (REG[0Ch] bits 1:0) and GPIO1 is configured as an input, a read from this bit returns the status of GPIO1.
- When the 'Direct' HR-TFT LCD interface is enabled (REG[0Ch] bits 1:0 = 10), GPIO0 outputs the CLS signal automatically and writing to this bit has no effect.
- bit 0 GPIO0 Pin IO Status
When the 'Direct' HR-TFT LCD interface is not selected (REG[0Ch] bits 1:0) and GPIO0 is configured as an output, writing a 1 to this bit drives GPIO0 high and writing a 0 to this bit drives GPIO0 low.
When the 'Direct' HR-TFT LCD interface is not selected (REG[0Ch] bits 1:0) and GPIO0 is configured as an input, a read from this bit returns the status of GPIO0.
- When the 'Direct' HR-TFT LCD interface is enabled (REG[0Ch] bits 1:0 = 10), GPIO0 outputs the PS signal automatically and writing to this bit has no effect.

PWM Clock Configuration Register															
REG[70h] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a								PWM Clock Divide Select bits 3-0				PWM Clock Force High	PWMCLK Source Select bits 1-0		PWM Clock Enable
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

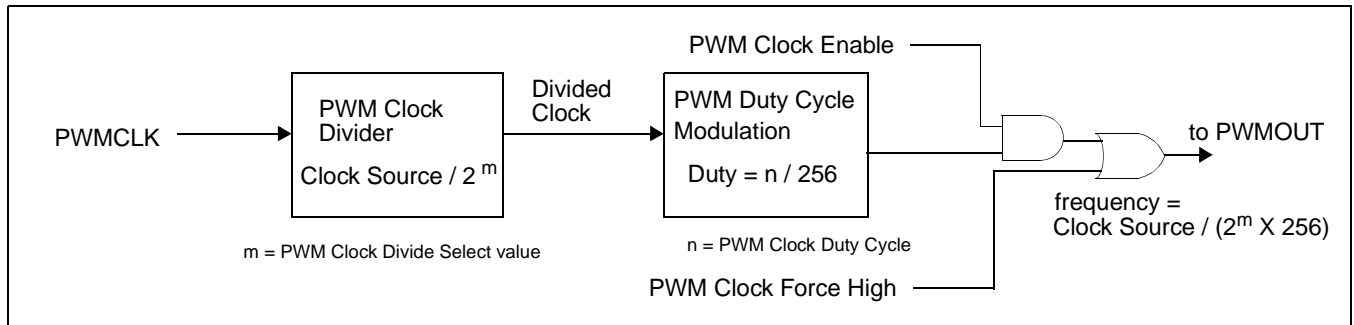


Figure 8-2: PWM Clock Block Diagram

Note

For further information on PWMCLK, see Section 7.1.4, “PWMCLK” on page 86.

bits 7-4

PWM Clock Divide Select Bits [3:0]

The value of these bits represents the power of 2 by which the selected PWM clock source is divided.

Table 8-15: PWM Clock Divide Select Options

PWM Clock Divide Select Bits [3:0]	PWM Clock Divide Amount
0h	1
1h	2
2h	4
3h	8
4h	16
5h	32
6h	64
7h	128
8h	256
9h	512
Ah	1024
Bh	2048
Ch	4096
Dh	8192
Eh	16384
Fh	32768

Note

This divided clock is further divided by 256 before it is output at PWMOUT.

- bit 3 PWM Clock Force High
When this bit = 0, the PWMOUT pin function is controlled by the PWM Clock enable bit.
When this bit = 1, the PWMOUT pin is forced to high.

- bit 1 PWMCLK Source Select Bits [1:0]
These bits determine the source of PWMCLK.

Table 8-16: PWMCLK Source Selection

PWMCLK Source Select Bits	PWMCLK Source
00	CLKI
01	CLKI2
10	BCLK
11	PCLK

Note

For further information on the PWMCLK source select, see Section 7.2, “Clock Selection” on page 87.

- bit 0 PWM Clock Enable
When this bit = 0, PWMOUT output acts as a general purpose output pin controllable by bit 3 of REG[70h].
When this bit = 1, the PWM Clock circuitry is enabled.

Note

The PWM Clock circuitry is disabled when Power Save Mode is enabled.

PWMOUT Duty Cycle Register																
REG[74h] Default = 00000000h																
Read/Write																
31	30	29	28	27	26	25	24	n/a	23	22	21	20	19	18	17	16
n/a								PWMOUT Duty Cycle bits 7-0								
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	

- bits 7-0 PWMOUT Duty Cycle Bits [7:0]
This register determines the duty cycle of the PWMOUT output.

Table 8-17: PWMOUT Duty Cycle Select Options

PWMOUT Duty Cycle [7:0]	PWMOUT Duty Cycle
00h	Always Low
01h	High for 1 out of 256 clock periods
02h	High for 2 out of 256 clock periods
...	...
FFh	High for 255 out of 256 clock periods

Scratch Pad A Register															
REG[80h]		Default = not applicable												Read/Write	
Scratch Pad A bits 31-24															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
Scratch Pad A bits 15-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 31-0

Scratch Pad A Bits [31:0]

This register contains general purpose read/write bits. These bits have no effect on hardware.

Note

The contents of the Scratch Pad A register defaults to an un-defined state after initial power-up. Any data written to this register remains intact when the S1D13A04 is reset, **as long as the chip is not powered off.**

Scratch Pad B Register															
REG[84h]		Default = not applicable												Read/Write	
Scratch Pad B bits 31-24															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
Scratch Pad B bits 15-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 31-0

Scratch Pad B Bits [31:0]

This register contains general purpose read/write bits. These bits have no effect on hardware.

Note

The contents of the Scratch Pad B register defaults to an un-defined state after initial power-up. Any data written to this register remains intact when the S1D13A04 is reset, **as long as the chip is not powered off.**

Scratch Pad C Register															
REG[88h]		Default = not applicable												Read/Write	
Scratch Pad C bits 31-24															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
Scratch Pad C bits 15-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 31-0

Scratch Pad C Bits [31:0]

This register contains general purpose read/write bits. These bits have no effect on hardware.

Note

The contents of the Scratch Pad C register defaults to an un-defined state after initial power-up. Any data written to this register remains intact when the S1D13A04 is reset, **as long as the chip is not powered off.**

8.4 USB Registers (Offset = 4000h)

The S1D13A04 USB device occupies a 48 byte local register space which can be accessed by the CPU on the local host interface.

To access the USB registers:

1. A valid USBCLK must be provided.
2. The USBCLK Enable bit (REG[4000h] bit 7) must be set to 1 and the USB Setup bit (REG[4000h] bit 2) must be set to 1. Both bits should be set together.

If any of the above conditions are not true, the USB registers must not be accessed.

Control Register REG[4000h]							Default = 00h		Read/Write
n/a									
15	14	13	12	11	10	9	8		
USBClk Enable	Software EOT	USB Enable	Endpoint 4 Stall	Endpoint 3 Stall	USB Setup	Reserved	Reserved		
7	6	5	4	3	2	1	0		

bit 7

USBClk Enable.

This bit allows the USBClk to be enabled/disabled allowing the S1D13A04 to save power when the USBClk is not required. This bit should be initially set with the USB Setup bit. However, it can be disabled/re-enabled individually.

When this bit = 1, the USBClk is enabled.

When this bit = 0, the USBClk is disabled.

Note

The USB Registers must not be accessed when this bit is 0.

bit 6

Software EOT

This bit determines the response to an IN request to Endpoint 4 when the transmit FIFO is empty. If this bit is asserted, the S1D13A04 responds to an IN request to Endpoint 4 with an ACK and a zero length packet if the FIFO is empty. If this bit is not asserted, the S1D13A04 responds to an IN request from Endpoint 4 with a NAK if the FIFO is empty, indicating that it expects to transmit more data. This bit is automatically cleared when the S1D13A04 responds to the host with a zero length packet when the FIFO is empty.

bit 5

USB Enable

Any device or configuration descriptor reads from the host will be acknowledged with a NAK until this bit is set. This allows time for the local CPU to set up the interrupt polling register, maximum packet size registers, and other configuration registers (e.g. Product ID and Vendor ID) before the host reads the descriptors.

Note

As the device and configuration descriptors cannot be read by the host until the USB Enable bit is set, the device enumeration process will not complete and the device will not be recognized on the USB.

- bit 4 Endpoint 4 Stall.
If this bit is set, host bulk reads from the transmit FIFO will result in a STALL acknowledge by the S1D13A04. No data will be returned to the USB host.
- bit 3 Endpoint 3 Stall.
If this bit is set, host bulk writes to the receive FIFO will result in a STALL acknowledge by the S1D13A04. Receive data will be discarded.
- bit 2 USB Setup
This bit is used by software to select between GPIO and USB functions for multifunction GPIO pins (GPIO[7:4]). This bit should be set at the same time as the USBClk Enable bit. When this bit = 1, the USB function is selected. When this bit = 0, the GPIO function is selected.
- Note**
The USB Registers must not be accessed when this bit is 0.
- bit 1 Reserved.
This bit must be set to 0.
- bit 0 Reserved.
This bit must be set to 0.

Interrupt Enable Register 0							Read/Write
REG[4002h]							Default = 00h
n/a							
15	14	13	12	11	10	9	8
Suspend Request Interrupt Enable	SOF Interrupt Enable	reserved	Endpoint 4 Interrupt Enable	Endpoint 3 Interrupt Enable	Endpoint 2 Interrupt Enable	Endpoint 1 Interrupt Enable	n/a
7	6	5	4	3	2	1	0

- bit 7 Suspend Request Interrupt Enable.
When set, this bit enables an interrupt to occur when the USB host is requesting the S1D13A04 USB device to enter suspend mode.
- bit 6 SOF Interrupt Enable.
When set, this bit enables an interrupt to occur when a start-of-frame packet is received by the S1D13A04.
- bit 5 Reserved.
This bit must be set to 0.
- bit 4 Endpoint 4 Interrupt Enable.
When set, this bit enables an interrupt to occur when a USB Endpoint 4 Data Packet has been sent by the S1D13A04.
- bit 3 Endpoint 3 Interrupt Enable.
When set, this bit enables an interrupt to occur when a USB Endpoint 3 Data Packet has been received by the S1D13A04.
- bit 2 Endpoint 2 Interrupt Enable.
When set, this bit enables an interrupt to occur when the USB Endpoint 2 Transmit Mailbox registers have been read by the USB host.

bit 1 Endpoint 1 Interrupt Enable.
When set, this bit enables an interrupt to occur when the USB Endpoint 1 Receive Mailbox registers have been written to by the USB host.

Interrupt Status Register 0							
REG[4004h]		Default = 00h				Read/Write	
n/a							
15	14	13	12	11	10	9	8
Suspend Request Interrupt Status	SOF Interrupt Status	Reserved	Endpoint 4 Interrupt Status	Endpoint 3 Interrupt Status	Endpoint 2 Interrupt Status	Endpoint 1 Interrupt Status	Upper Interrupt Active (read only)
7	6	5	4	3	2	1	0

bit 7 Suspend Request Interrupt Status.
This bit indicates when a suspend-request has been received by the S1D13A04. Writing a 1 clears this bit.

bit 6 SOF Interrupt Status.
This bit indicates when a start-of-frame packet has been received by the S1D13A04. Writing a 1 clears this bit.

bit 5 Reserved.
This bit must be set to 0.

bit 4 Endpoint 4 Interrupt Status.
This bit indicates when a USB Endpoint 4 Data packet has been sent by the S1D13A04. Writing a 1 clears this bit.

bit 3 Endpoint 3 Interrupt Status (Receive FIFO Valid).
This bit indicates when a USB Endpoint 3 Data packet has been received by the S1D13A04. No more packets to endpoint 3 will be accepted until this bit is cleared. Writing a 1 clears this bit.

bit 2 Endpoint 2 Interrupt Status.
This bit indicates when the USB Endpoint 2 Mailbox registers have been read by the USB host. Writing a 1 clears this bit.

bit 1 Endpoint 1 Interrupt Status (Receive Mailbox Valid).
This bit indicates when the USB Endpoint 1 Mailbox registers have been written to by the USB host. Writing a 1 clears this bit.

bit 0 Upper Interrupt Active (read only).
At least one interrupt status bit is set in register REG[4008h].

Interrupt Enable Register 1								
REG[4006h]		Default = 00h						Read/Write
n/a				n/a				
15	14	13	12	11	10	9	8	
n/a						Transmit FIFO Almost Empty Interrupt Enable	Receive FIFO Almost Full Interrupt Enable	
7	6	5	4	3	2	1	0	

bit 1 Transmit FIFO Almost Empty Interrupt Enable.
When set, this bit enables an interrupt to be generated when the Transmit FIFO Almost Empty status bit is set.

Note

The Transmit FIFO Almost Empty threshold must be set greater than zero, as the FIFO count must drop below the threshold to cause an interrupt.

bit 0 Receive FIFO Almost Full Interrupt Enable.
When set, this bit enables an interrupt to be generated when the Receive FIFO Almost Full status bit is set.

Note

The Receive FIFO Almost Full threshold must be set less than 64, as the FIFO count must rise above the threshold to cause an interrupt.

Interrupt Status Register 1								
REG[4008h]		Default = 00h						Read/Write
n/a				n/a				
15	14	13	12	11	10	9	8	
n/a						Transmit FIFO Almost Empty Status	Receive FIFO Almost Full Status	
7	6	5	4	3	2	1	0	

bit 1 Transmit FIFO Almost Empty Status.
This bit is set when the number of bytes in the Transmit FIFO is equal to the Transmit FIFO Almost Empty Threshold, and another byte is sent to the USB bus from the FIFO. Writing a 1 clears this bit.

bit 0 Receive FIFO Almost Full Status.
This bit is set when the number of bytes in the Receive FIFO is equal to the Receive FIFO Almost Full Threshold, and another byte is received from the USB bus into the FIFO. Writing a 1 clears this bit.

Endpoint 1 Index Register							
REG[4010h]		Default = 00h				Read/Write	
n/a							
15	14	13	12	11	10	9	8
n/a							
Endpoint 1 Index bits 2-0 (RO)				7	6	5	4
3	2	1	0				

bits 2-0 Endpoint 1 Index Register Bits [2:0].
 This register determines which Endpoint 1 Receive Mailbox is accessed when the Endpoint 1 Receive Mailbox Data register is read. This register is automatically incremented after the Endpoint 1 Receive Mailbox Data register is read. This index register wraps around to zero when it reaches the maximum count (7).

Endpoint 1 Receive Mailbox Data Register							
REG[4012h]		Default = 00h				Read Only	
n/a							
15	14	13	12	11	10	9	8
Endpoint 1 Receive Mailbox Data bits 7-0							
7	6	5	4	3	2	1	0

bits 7-0 Endpoint 1 Receive Mailbox Data Bits [7:0].
 This register is used to read data from one of the receive mailbox registers. Data is returned from the register selected by the Endpoint 1 Index Register. The eight receive mailbox registers are written by a USB bulk transfer to endpoint 1, and can be used to pass messages from the USB host to the local CPU. The format and content of the messages are user defined. If enabled, USB writes to this register can generate an interrupt.

Endpoint 2 Index Register							
REG[4018h]		Default = 00h				Read/Write	
n/a							
15	14	13	12	11	10	9	8
n/a							
Endpoint 2 Index bits 2-0				7	6	5	4
3	2	1	0				

bits 2-0 Endpoint 2 Index Register Bits [2:0].
 This register determines which Endpoint 2 Transmit Mailbox is accessed when the Endpoint 2 Transmit Mailbox Data register is read or written. This register is automatically incremented after the Endpoint 2 Transmit Mailbox Data port is read or written. This index register wraps around to zero when it reaches the maximum count (7).

Endpoint 2 Transmit Mailbox Data Register							
REG[401Ah]		Default = 00h				Read/Write	
n/a							
15	14	13	12	11	10	9	8
Endpoint 2 Transmit Mailbox Data bits 7-0							
7	6	5	4	3	2	1	0

bits 7-0

Endpoint 2 Transmit Mailbox Data Bits [7:0].

This register is used to read or write one of the transmit mailbox registers. The register being accessed is selected by the Endpoint 2 Index register. The eight Transmit Mailbox registers are written by the local CPU and are read by a USB transfer from endpoint 2. The format and content of the messages are user defined. If enabled, USB reads from this register can generate an interrupt.

Endpoint 2 Interrupt Polling Interval Register							
REG[401Ch]		Default = FFh				Read/Write	
n/a							
15	14	13	12	11	10	9	8
Interrupt Polling Interval bits 7-0							
7	6	5	4	3	2	1	0

bits 7-0

Interrupt Polling Interval Bits [7:0].

This register specifies the Endpoint 2 interrupt polling interval in milliseconds. It can be read by the host through the endpoint 2 descriptor.

Endpoint 3 Receive FIFO Data Register							
REG[4020h]		Default = 00h				Read Only	
n/a							
15	14	13	12	11	10	9	8
Endpoint 3 Receive FIFO Data bits 7-0							
7	6	5	4	3	2	1	0

bits7-0

Endpoint 3 Receive FIFO Data Bits [7:0].

This register is used by the local CPU to read USB receive FIFO data. The FIFO data is written by the USB host using bulk or isochronous transfers to endpoint 3.

Endpoint 3 Receive FIFO Count Register							
REG[4022h]		Default = 00h				Read Only	
n/a							
15	14	13	12	11	10	9	8
Receive FIFO Count bits 7-0							
7	6	5	4	3	2	1	0

bits 7-0

Receive FIFO Count Bits [7:0].

This register returns the number of receive FIFO entries containing valid entries. Values range from 0 (empty) to 64 (full).

Endpoint 3 Receive FIFO Status Register								
REG[4024h]		Default = 01h						Read/Write
n/a								
15	14	13	12	11	10	9	8	
n/a			Receive FIFO Flush	Receive FIFO Overflow	Receive FIFO Underflow	Receive FIFO Full (read only)	Receive FIFO Empty (read only)	
7	6	5	4	3	2	1	0	

- bit 4 Receive FIFO Flush
Writing to this bit causes the receive FIFO to be flushed. Reading this bit always returns a 0.
- bit 3 Receive FIFO Overflow
If set, this bit indicates that an attempt was made by the USB host to write to the receive FIFO when the receive FIFO was full. Writing a 1 clears this bit.
- bit 2 Receive FIFO Underflow
If set, this bit indicates that an attempt was made to read the receive FIFO when the receive FIFO was empty. Writing a 1 clears this bit.
- bit 1 Receive FIFO Full
If set, this bit indicates that the receive FIFO is full.
- bit 0 Receive FIFO Empty
If set, this bit indicates that the receive FIFO is empty.

Endpoint 3 Maximum Packet Size Register								
REG[4026h]		Default = 08h						Read/Write
n/a								
15	14	13	12	11	10	9	8	
Endpoint 3 Max Packet Size bits 7-0								
7	6	5	4	3	2	1	0	

- bits 7-0 Endpoint 3 Max Packet Size Bits [7:0].
This register specifies the maximum packet size for endpoint 3 in units of 8 bytes (default = 64 bytes). It can be read by the host through the endpoint 3 descriptor.

Endpoint 4 Transmit FIFO Data Register								
REG[4028h]		Default = 00h						Write Only
n/a								
15	14	13	12	11	10	9	8	
Transmit FIFO Data bits 7-0								
7	6	5	4	3	2	1	0	

- bits 7-0 Transmit FIFO Data Bits [7:0].
This register is used by the local CPU to write data to the transmit FIFO. The FIFO data is read by the USB host using bulk or isochronous transfers from endpoint 4.

Endpoint 4 Transmit FIFO Count Register							
REG[402Ah]		Default = 00h				Read Only	
n/a							
15	14	13	12	11	10	9	8
Transmit FIFO Count bits 7-0							
7	6	5	4	3	2	1	0

bits 7-0 Transmit FIFO Count Bits [7:0].
This register returns the number of transmit FIFO entries containing valid entries. Values range from 0 (empty) to 64 (full).

Endpoint 4 Transmit FIFO Status Register							
REG[402Ch]		Default = 01h				Read/Write	
n/a							
15	14	13	12	11	10	9	8
n/a		Transmit FIFO Valid	Transmit FIFO Flush	Transmit FIFO Overflow	Reserved	Transmit FIFO Full (read only)	Transmit FIFO Empty (read only)
7	6	5	4	3	2	1	0

bit 5 Transmit FIFO Valid.
If set, this bit allows the data in the Transmit FIFO to be read by the next read from the host. This bit is automatically cleared by a host read. This bit is only used if bit 0 in USB[403Ah] Index [0Ch] is set.

bit 4 Transmit FIFO Flush.
Writing to this bit causes the transmit FIFO to be flushed. Reading this bit always returns a 0.

bit 3 Transmit FIFO Overflow.
If set, this bit indicates that an attempt was made by the local CPU to write to the transmit FIFO when the transmit FIFO was full. Writing a 1 clears this bit.

bit 2 Reserved.

bit 1 Transmit FIFO Full (read only).
If set, this bit indicates that the transmit FIFO is full.

bit 0 Transmit FIFO Empty (read only).
If set, this bit indicates that the transmit FIFO is empty.

Endpoint 4 Maximum Packet Size Register							
REG[402Eh]		Default = 08h				Read/Write	
n/a							
15	14	13	12	11	10	9	8
Endpoint 4 Max Packet Size bits 7-0							
7	6	5	4	3	2	1	0

bits 7-0 Endpoint 4 Max Packet Size Bits [7:0].
This register specifies the maximum packet size for endpoint 4 in units of 8 bytes (default = 64 bytes). It can be read by the host through the endpoint 4 descriptor.

Revision Register							
REG[4030h] Default = 01h							Read Only
n/a							
15	14	13	12	11	10	9	8
Chip Revision bits 7-0							
7	6	5	4	3	2	1	0

bits 7-0 Chip Revision Bits [7:0].
This register returns current silicon revision number of the USB client.

USB Status Register							
REG[4032h] Default = 00h							Read/Write
n/a							
15	14	13	12	11	10	9	8
Suspend Control	USB Endpoint 4 STALL	USB Endpoint 4 NAK	USB Endpoint 4 ACK	USB Endpoint 3 STALL	USB Endpoint 3 NAK	USB Endpoint 3 ACK	Endpoint 2 Valid
7	6	5	4	3	2	1	0

- bit 7 Suspend Control.
If set, this bit indicates that there is a pending suspend request. Writing a 1 clears this bit and causes the S1D13A04 USB device to enter suspended mode.
- bit 6 USB Endpoint 4 STALL.
The last USB IN token could not be serviced because the endpoint was stalled (REG[4000h] bit 4 set), and was acknowledged with a STALL. Writing a 1 clears this bit.
- bit 5 USB Endpoint 4 NAK.
The last USB packet transmitted (IN packet) encountered a FIFO underrun condition, and was acknowledged with a NAK. Writing a 1 clears this bit.
- bit 4 USB Endpoint 4 ACK.
The last USB packet transmitted (IN packet) was successfully acknowledged with an ACK from the USB host. Writing a 1 clears this bit.
- bit 3 USB Endpoint 3 STALL.
The last USB packet received (OUT packet) could not be accepted because the endpoint was stalled (REG[4000h] bit 3 set), and was acknowledged with a STALL. Writing a 1 clears this bit.
- bit 2 USB Endpoint 3 NAK.
The last USB packet received (OUT packet) could not be accepted, and was acknowledged with a NAK. Writing a 1 clears this bit.
- bit 1 USB Endpoint 3 ACK.
The last USB packet received (OUT packet) was successfully acknowledged with an ACK. Writing a 1 clears this bit.
- bit 0 Endpoint 2 Valid.
When this bit is set, the 8-byte endpoint 2 mailbox registers have been written by the local CPU, but not yet read by the USB host. The local CPU should not write into these registers while this bit is set.

Frame Counter MSB Register							
REG[4034h]		Default = 00h				Read Only	
n/a							
15	14	13	12	11	10	9	8
n/a				Frame Counter bits 10-8			
7	6	5	4	3	2	1	0

Frame Counter LSB Register							
REG[4036h]		Default = 00h				Read Only	
n/a							
15	14	13	12	11	10	9	8
Frame Counter bits 7-0							
7	6	5	4	3	2	1	0

bits 10-0

Frame Counter Bits [10:0]

This register contains the frame counter from the most recent start-of-frame packet.

Extended Register Index							
REG[4038h]		Default = 00h				Read/Write	
n/a							
15	14	13	12	11	10	9	8
Extended Register Index bits 7-0							
7	6	5	4	3	2	1	0

bits 7-0

Extended Register Index Bits [7:0]

This register selects which extended data register is accessed when the REG[403Ah] is read or written.

Extended Register Data							
REG[403Ah]		Default = 04h				Read/Write	
n/a							
15	14	13	12	11	10	9	8
Extended Data bits 7-0							
7	6	5	4	3	2	1	0

bits 7-0

Extended Data Bits [7:0]

This port provides access to one of the extended data registers. The index of the current register is held in REG[4038h].

Vendor ID MSB							
REG[403Ah], Index[00h]		Default = 04h				Read/Write	
Vendor ID bits 15-8							
7	6	5	4	3	2	1	0

Vendor ID LSB							
REG[403Ah], Index[01h]		Default = B8h				Read/Write	
Vendor ID bits 7-0							
7	6	5	4	3	2	1	0

bits 15-0

Vendor ID Bits [15:0]

These registers determine the Vendor ID returned in a "Get Device Descriptor" request.

Product ID MSB							
REG[403Ah], Index[02h]				Default = 88h		Read/Write	
Product ID bits 15-8							
7	6	5	4	3	2	1	0

Product ID LSB							
REG[403Ah], Index[03h]				Default = 21h		Read/Write	
Product ID bits 7-0							
7	6	5	4	3	2	1	0

bits 15-0 Product ID Bits [15:0]
These registers determine the Product ID returned in a “Get Device Descriptor” request.

Release Number MSB							
REG[403Ah], Index[04h]				Default = 01h		Read/Write	
Release Number bits 15-8							
7	6	5	4	3	2	1	0

Release Number LSB							
REG[403Ah], Index[05h]				Default = 00h		Read/Write	
Release Number bits 7-0							
7	6	5	4	3	2	1	0

bits 15-0 Release Number Bits [15:0]
These registers determine the device release number returned in a “Get Device Descriptor” request.

Receive FIFO Almost Full Threshold							
REG[403Ah], Index[06h]				Default = 3Ch		Read/Write	
7	n/a	6	Receive FIFO Almost Full Threshold bits 5-0				
7	6	5	4	3	2	1	0

bits 5-0 Receive FIFO Almost Full Threshold Bits [5:0]
This register determines the threshold at which the receive FIFO almost full status bit is set.

Note

The Receive FIFO Almost Full threshold must be set less than 64, as the FIFO count must rise above the threshold to cause an interrupt.

Transmit FIFO Almost Empty Threshold							
REG[403Ah], Index[07h]				Default = 04h		Read/Write	
7	n/a	6	Transmit FIFO Almost Empty Threshold bits 5-0				
7	6	5	4	3	2	1	0

bits 5-0 Transmit FIFO Almost Empty Threshold Bits [5:0].
This register determines the threshold at which the transmit FIFO almost empty status bit is set.

Note

The Transmit FIFO Almost Empty threshold must be set greater than zero, as the FIFO count must drop below the threshold to cause an interrupt.

USB Control							Read/Write
REG[403Ah], Index[08h]							Default = 01h
n/a							USB String Enable
7	6	5	4	3	2	1	0

bit 0 USB String Enable.
When set, this bit allows the default Vendor and Product ID String Descriptors to be returned to the host. When this bit is cleared, the string index values in the Device Descriptor are set to zero.

Maximum Power Consumption								Read/Write
REG[403Ah], Index[09h]								Default = FAh
Maximum Current bits 7-0								
7	6	5	4	3	2	1	0	

bits 7-0 Maximum Current Bits [7:0].
The amount of current drawn by the peripheral from the USB port in increments of 2 mA. The S1D13A04 reports this value to the host controller in the configuration descriptor. The default and maximum value is 500 mA (FAh * 2 mA).

In order to comply with the USB specification the following formula must apply:
$$\text{REG}[403\text{Ah}] \text{ index}[09\text{h}] \leq \text{FAh}.$$

Packet Control								Read/Write
REG[403Ah], Index[0Ah]								Default = 00h
EP4 Data Toggle	EP3 Data Toggle	EP2 Data Toggle	EP1 Data Toggle	Reserved	Reserved	n/a	Reserved	
7	6	5	4	3	2	1	0	

bit 7 EP4 Data Toggle Bit.
Contains the value of the Data Toggle bit to be sent in response to the next IN token to endpoint 4 from the USB host.

Note

When a write is made to this bit, the value cannot be read back before a minimum of 12 USBCLK.

bit 6 EP3 Data Toggle Bit.
Contains the value of the Data Toggle bit expected in the next DATA packet to endpoint 3 from the USB host.

Note

When a write is made to this bit, the value cannot be read back before a minimum of 12 USBCLK.

bit 5 EP2 Data Toggle Bit.
Contains the value of the Data Toggle bit to be sent in response to the next IN token to endpoint 2 from the USB host.

Note

When a write is made to this bit, the value cannot be read back before a minimum of 12 USBCLK.

bit 4 EP1 Data Toggle Bit.
Contains the value of the Data Toggle bit expected in the next DATA packet to endpoint 1 from the USB host.

Note

When a write is made to this bit, the value cannot be read back before a minimum of 12 USBCLK.

bit 3 Reserved.
This bit must be set to 0.

bit 2 Reserved.
This bit must be set to 0.

bit 0 Reserved.
This bit must be set to 0.

Reserved							Read/Write
REG[403Ah], Index[0Bh]							Default = 00h
7	6	5	4	3	2	1	Reserved 0

bit 0 Reserved.
This bit must be set to 0.

FIFO Control							Read/Write
REG[403Ah], Index[0Ch]							Default = 00h
7	6	5	4	3	2	1	Transmit FIFO Valid Mode 0

bit 0 Transmit FIFO Valid Mode.
When set, this bit causes a NAK response to a host read request from the transmit FIFO (EP4) unless the FIFO Valid bit (in register EP4STAT) is set. When this bit is cleared, any data waiting in the transmit FIFO will be sent in response to a host read request, and the FIFO Valid bit is ignored.

USBFC Input Control Register							
REG[4040h]		Default = 0Dh				Read/Write	
n/a							
15	14	13	12	11	10	9	8
n/a	USCMPEN	Reserved	Reserved	ISO	WAKEUP	Reserved	Reserved
7	6	5	4	3	2	1	0

These bits control inputs to the USB module.

- bit 6 USCMPEN
This bit controls the USB differential input receiver.
0 = differential input receiver disabled
1 = differential input receiver enabled
- bits 5 Reserved.
This bit must be set to 0.
- bits 4 Reserved.
This bit must be set to 0.
- bit 3 ISO
This bits selects between isochronous and bulk transfer modes for the FIFOs (Endpoint 3 and Endpoint 4).
0 = Isochronous transfer mode
1 = Bulk transfer mode
- bit 2 WAKEUP
This active low bit initiates a USB remote wake-up.
0 = initiate USB remote wake-up
1 = no action
- bit 1 Reserved.
This bit must be set to 0.
- bit 0 Reserved.
This bit must be set to 0.

Reserved							
REG[4042h]							
n/a							
15	14	13	12	11	10	9	8
n/a							
7	6	5	4	3	2	1	0

Pin Input Status / Pin Output Data Register							
REG[4044h] Default = depends on USB input pin state							Read/Write
n/a							
15	14	13	12	11	10	9	8
n/a						USBDETECT Input Pin Status (read only)	USBPUP Output Pin Status
7	6	5	4	3	2	1	0

These bits can generate interrupts.

bit 1 USBDETECT Input Pin Status
This read-only bit indicates the status of the USBDETECT input pin after a steady-state period of 0.5 seconds.

bit 0 USBPUP Output Pin Status
This bit controls the state of the USBPUP output pin.

This bit must be set to 1 to enable the USB interface and USB registers. See the *S1D13A04 Programming Notes and Examples*, document number X37-A-G-003-xx for further information on this bit.

Interrupt Control Enable Register 0							
REG[4046h] Default = 00h							Read/Write
n/a							
15	14	13	12	11	10	9	8
n/a	USB Host Connected	Reserved	Reserved	Reserved	Reserved	USBRESET	Reserved
7	6	5	4	3	2	1	0

These bits enable interrupts from the corresponding bit of the Interrupt Control Status/Clear Register 0.

0 = corresponding interrupt bit disabled (masked).

1 = corresponding interrupt bit enabled.

Interrupt Control Enable Register 1							
REG[4048h] Default = 00h							Read/Write
n/a							
15	14	13	12	11	10	9	8
n/a	USB Host Disconnect	Reserved	Device Configured	Reserved	Reserved	Reserved	INT
7	6	5	4	3	2	1	0

These bits enable interrupts from the corresponding bit of the Interrupt Control Status/Clear Register 1.

0 = corresponding interrupt bit disabled (masked).

1 = corresponding interrupt bit enabled.

Interrupt Control Status/Clear Register 0							
REG[404Ah]		Default = 00h				Read/Write	
n/a							
15	14	13	12	11	10	9	8
n/a	USB Host Connected	Reserved	Reserved	Reserved	Reserved	USBRESET	Reserved
7	6	5	4	3	2	1	0

On reads, these bits represent the interrupt status for interrupts caused by low-to-high transitions on the corresponding signals.

0 (read) = no low-to-high event detected on the corresponding signal.

1 (read) = low-to-high event detected on the corresponding signal.

On writes, these bits clear the corresponding interrupt status bit.

0 (write) = corresponding interrupt status bit unchanged.

1 (write) = corresponding interrupt status bit cleared to zero.

These bits must always be cleared via a write to this register before first use. This will ensure that any changes on input pins during system initialization do not generate erroneous interrupts. The interrupt bits are used as follows.

- bit 6 USB Host Connected
Indicates the USB device is connected to a USB host.
- bit 5 Reserved.
Must be set to 0.
- bit 4 Reserved.
Must be set to 0.
- bit 3 Reserved.
Must be set to 0.
- bit 2 Reserved.
Must be set to 0.
- bit 1 USBRESET
Indicates the USB device is reset using the RESET# pin or using the USB port reset.
- bit 0 Reserved.
Must be set to 0.

Interrupt Control Status/Clear Register 1							
REG[404Ch]		Default = 00h				Read/Write	
n/a							
15	14	13	12	11	10	9	8
n/a	USB Host Disconnected	Reserved	Device Configured	Reserved	Reserved	Reserved	INT
7	6	5	4	3	2	1	0

On reads, these bits represent the interrupt status for interrupts caused by high-to-low transitions on the corresponding signals.

0 (read) = no high-to-low event detected on the corresponding signal.

1 (read) = high-to-low event detected on the corresponding signal.

On writes, these bits clear the corresponding interrupt status bit.

0 (write) = corresponding interrupt status bit unchanged.

1 (write) = corresponding interrupt status bit cleared to zero.

These bits must always be cleared via a write to this register before first use. This will ensure that any changes on input pins during system initialization do not generate erroneous interrupts. The interrupt bits are used as follows.

- bit 6 USB Host Disconnected
Indicates the USB device is disconnected from a USB host.
- bit 5 Reserved.
Must be set to 0.
- bit 4 Device Configured.
Indicates the USB device has been configured by the USB host.
- bit 3 Reserved.
Must be set to 0.
- bit 2 Reserved.
Must be set to 0.
- bit 1 Reserved.
Must be set to 0.
- bit 0 INT
Indicates an interrupt request originating from within the USB registers (REG[4000h] to REG[403Ah]).

Interrupt Control Masked Status Register 0							
REG[404Eh]		Default = 00h				Read Only	
n/a							
15	14	13	12	11	10	9	8
n/a	USB Host Connected	Reserved	Reserved	Reserved	Reserved	USBRESET	Reserved
7	6	5	4	3	2	1	0

These read-only bits represent the logical AND of the corresponding Interrupt Control Status/Clear Register 0 (REG[404Ah]) and the Interrupt Control Enable Register 0 (REG[4046h]).

Interrupt Control Masked Status Register 1							
REG[4050h]		Default = 00h				Read Only	
n/a							
15	14	13	12	11	10	9	8
n/a	USB Host Disconnected	Reserved	Device Configured	Reserved	Reserved	Reserved	INT
7	6	5	4	3	2	1	0

These read-only bits represent the logical AND of the corresponding Interrupt Control Status/Clear Register 1 (REG[404Ch]) and the Interrupt Control Enable Register 1 (REG[4048h]).

USB Software Reset Register							
REG[4052h]		Default = 00h				Write Only	
n/a							
15	14	13	12	11	10	9	8
USB Software Reset (Code = 10100100) bits 7-0							
7	6	5	4	3	2	1	0

bits 7-0

USB Software Reset Bits [7:0] (Write Only)

When the specific code of 10100100b is written to these bits the USB module of the S1D13A04 is reset. Use of the above code avoids the possibility of accidentally resetting the USB.

USB Wait State Register							
REG[4054h]		Default = 00h				Read/Write	
n/a							
15	14	13	12	11	10	9	8
n/a						USB Wait State bits 1-0	
7	6	5	4	3	2	1	0

bits 1-0

USB Wait State Bits [1:0]

This register controls the number of wait states the S1D13A04 uses for its internal USB support. For all bus interfaces supported by the S1D13A04 **these bits must be set to 01.**

8.5 2D Acceleration (BitBLT) Registers (Offset = 8000h)

These registers control the S1D13A04 2D Acceleration engine. For detailed BitBLT programming instructions, see the *S1D13A04 Programming Notes and Examples*, document number X37A-G-003-xx.

BitBLT Control Register														Read/Write		
REG[8000h]														Default = 00000000h		
n/a														Color Format Select	Dest Linear Select	Source Linear Select
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	
n/a														BitBLT Enable (WO)		
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	

- bit 18 BitBLT Color Format Select
This bit selects the color format that the 2D operation is applied to.
When this bit = 0, 8 bpp (256 color) format is selected.
When this bit = 1, 16 bpp (64K color) format is selected.
- bit 17 BitBLT Destination Linear Select
When this bit = 1, the Destination BitBLT is stored as a contiguous linear block of memory.
When this bit = 0, the Destination BitBLT is stored as a rectangular region of memory.
The BitBLT Memory Address Offset register (REG[8014h]) determines the address offset from the start of one line to the next line.
- bit 16 BitBLT Source Linear Select
When this bit = 1, the Source BitBLT is stored as a contiguous linear block of memory.
When this bit = 0, the Source BitBLT is stored as a rectangular region of memory.
The BitBLT Memory Address Offset register (REG[8014h]) determines the address offset from the start of one line to the next line.
- bit 0 BitBLT Enable
This bit is write only.
Setting this bit to 1 begins the 2D BitBLT operation. **This bit must not be set to 0 while a BitBLT operation is in progress.**

Note

To determine the status of a BitBLT operation use the BitBLT Busy Status bit (REG[8004h] bit 0).

BitBLT Status Register														Read Only		
REG[8004h]														Default = 00000000h		
n/a			Number of Used FIFO Entries						n/a			Number of Free FIFO Entries (0 means full)				
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	
n/a									FIFO Not Empty	FIFO Half Full	FIFO Full Status	n/a			BitBLT Busy Status	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	

bits 28-24

Number of Used FIFO Entries Bits [4:0]

These bits indicate the minimum number of FIFO entries currently in use (there may be more values in internal pipeline stages).

bits 20-16

Number of Free FIFO Entries Bits [4:0]

These bits indicate the number of empty FIFO entries available. If these bits return a 0, the FIFO is full.

bit 6

BitBLT FIFO Not-Empty Status

This is a read-only status bit.

When this bit = 0, the BitBLT FIFO is empty.

When this bit = 1, the BitBLT FIFO has at least one data.

To reduce system memory read latency, software can monitor this bit prior to a BitBLT read burst operation.

The following table shows the number of words available in BitBLT FIFO under different status conditions.

Table 8-18: BitBLT FIFO Words Available

BitBLT FIFO Full Status (REG[8004h] Bit 4)	BitBLT FIFO Half Full Status (REG[8004h] Bit 5)	BitBLT FIFO Not Empty Status (REG[8004h] Bit 6)	Number of Words available in BitBLT FIFO
0	0	0	0
0	0	1	1 to 6
0	1	1	7 to 14
1	1	1	15 to 16

bit 5

BitBLT FIFO Half Full Status

This is a read-only status bit.

When this bit = 1, the BitBLT FIFO is half full or greater than half full.

When this bit = 0, the BitBLT FIFO is less than half full.

bit 4

BitBLT FIFO Full Status

This is a read-only status bit.

When this bit = 1, the BitBLT FIFO is full.

When this bit = 0, the BitBLT FIFO is not full.

bit 0 BitBLT Busy Status
This bit is a read-only status bit.
When this bit = 1, the BitBLT operation is in progress.
When this bit = 0, the BitBLT operation is complete.

Note

During a BitBLT Read operation, the BitBLT engine does not attempt to keep the FIFO full. If the FIFO becomes full, the BitBLT operation stops temporarily as data is read out of the FIFO. The BitBLT will restart only when less than 14 values remain in the FIFO.

BitBLT Command Register												Read/Write			
REG[8008h]												Default = 00000000h			
n/a												BitBLT ROP Code bits 3-0			
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a												BitBLT Operation bits 3-0			
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 19-16 BitBLT Raster Operation Code/Color Expansion Bits [3:0]
ROP Code for Write BitBLT and Move BitBLT. Bits 2-0 also specify the start bit position for Color Expansion.

Table 8-19 : BitBLT ROP Code/Color Expansion Function Selection

BitBLT ROP Code Bits [3:0]	Boolean Function for Write BitBLT and Move BitBLT	Boolean Function for Pattern Fill	Start Bit Position for Color Expansion
0000	0 (Blackness)	0 (Blackness)	bit 0
0001	$\sim S \cdot \sim D$ or $\sim(S + D)$	$\sim P \cdot \sim D$ or $\sim(P + D)$	bit 1
0010	$\sim S \cdot D$	$\sim P \cdot D$	bit 2
0011	$\sim S$	$\sim P$	bit 3
0100	$S \cdot \sim D$	$P \cdot \sim D$	bit 4
0101	$\sim D$	$\sim D$	bit 5
0110	$S \wedge D$	$P \wedge D$	bit 6
0111	$\sim S + \sim D$ or $\sim(S \cdot D)$	$\sim P + \sim D$ or $\sim(P \cdot D)$	bit 7
1000	$S \cdot D$	$P \cdot D$	bit 0
1001	$\sim(S \wedge D)$	$\sim(P \wedge D)$	bit 1
1010	D	D	bit 2
1011	$\sim S + D$	$\sim P + D$	bit 3
1100	S	P	bit 4
1101	$S + \sim D$	$P + \sim D$	bit 5
1110	$S + D$	$P + D$	bit 6
1111	1 (Whiteness)	1 (Whiteness)	bit 7

Note

S = Source, D = Destination, P = Pattern.
 \sim = NOT, \cdot = Logical AND, $+$ = Logical OR, \wedge = Logical XOR

bits 3-0

BitBLT Operation Bits [3:0]

Specifies the 2D Operation to be carried out based on the following table.

Table 8-20 : BitBLT Operation Selection

BitBLT Operation Bits [3:0]	BitBLT Operation
0000	Write BitBLT with ROP.
0001	Read BitBLT.
0010	Move BitBLT in positive direction with ROP.
0011	Move BitBLT in negative direction with ROP.
0100	Transparent Write BitBLT.
0101	Transparent Move BitBLT in positive direction.
0110	Pattern Fill with ROP.
0111	Pattern Fill with transparency.
1000	Color Expansion.
1001	Color Expansion with transparency.
1010	Move BitBLT with Color Expansion.
1011	Move BitBLT with Color Expansion and transparency.
1100	Solid Fill.
Other combinations	Reserved

BitBLT Source Start Address Register															Read/Write	
REG[800Ch]															Default = 00000000h	
n/a											BitBLT Source Start Address bits 20-16					
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	
BitBLT Source Start Address bits 15-0																
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	

bits 20-0

BitBLT Source Start Address Bits [20:0]

A 21-bit register that specifies the source start address for the BitBLT operation.

If data is sourced from the CPU, then bit 0 is used for byte alignment within a 16-bit word and the other address bits are ignored. In pattern fill operation, the BitBLT Source Start Address is defined by the following equation.

$$\text{Value programmed to the Source Start Address Register} = \text{Pattern Base Address} + \text{Pattern Line Offset} + \text{Pixel Offset.}$$

The following table shows how Source Start Address Register is defined for 8 and 16 bpp color depths.

Table 8-21 : BitBLT Source Start Address Selection

Color Format	Pattern Base Address[20:0]	Pattern Line Offset[2:0]	Pixel Offset[3:0]
8 bpp	BitBLT Source Start Address[20:6]	BitBLT Source Start Address[5:3]	BitBLT Source Start Address[2:0]
16 bpp	BitBLT Source Start Address[20:7]	BitBLT Source Start Address[6:4]	BitBLT Source Start Address[3:0]

Note

For further information on the BitBLT Source Start Address register, see the *S1D13A04 Programming Notes and Examples*, document number X37A-G-003-xx.

BitBLT Destination Start Address Register															Read/Write	
REG[8010h]															Default = 00000000h	
n/a											BitBLT Destination Start Address bits 20-16					
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	
BitBLT Destination Start Address bits 15-0																
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	

bits 20-0

BitBLT Destination Start Address Bits [20:0]

A 21-bit register that specifies the destination start address for the BitBLT operation.

BitBLT Memory Address Offset Register															
REG[8014h] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						BitBLT Memory Address Offset bits 10-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 10-0

BitBLT Memory Address Offset Bits [10:0]

These bits are the display's 11-bit address offset from the starting word of line n to the starting word of line $n + 1$. They are used only for address calculation when the BitBLT is configured as a rectangular region of memory. They are not used for the displays.

BitBLT Width Register															
REG[8018h] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						BitBLT Width bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 9-0

BitBLT Width Bits [9:0]

A 10-bit register that specifies the BitBLT width in pixels - 1.

$$\text{BitBLT width in pixels} = (\text{ContentsOfThisRegister}) + 1$$

BitBLT Height Register															
REG[801Ch] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						BitBLT Height bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 9-0

BitBLT Height Bits [9:0]

A 10-bit register that specifies the BitBLT height in lines - 1.

$$\text{BitBLT height in lines} = (\text{ContentsOfThisRegister}) + 1$$

BitBLT Background Color Register																	
REG[8020h]														Default = 00000000h		Read/Write	
n/a																	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
BitBLT Background Color bits 15-0																	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		

bits 15-0 BitBLT Background Color Bits [15:0]
This register specifies the BitBLT background color for Color Expansion or key color for Transparent BitBLT. For 16 bpp color depths (REG[8000h] bit 18 = 1), bits 15-0 are used. For 8 bpp color depths (REG[8000h] bit 18 = 0), bits 7-0 are used.

Note

For Big Endian implementations, see the *S1D13A04 Programming Notes and Examples*, document number X37A-G-003-xx.

BitBLT Foreground Color Register																	
REG[8024h]														Default = 00000000h		Read/Write	
n/a																	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
BitBLT Foreground Color bits 15-0																	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		

bits 15-0 BitBLT Foreground Color Bits [15:0]
This register specifies the BitBLT foreground color for Color Expansion or Solid Fill. For 16 bpp color depths (REG[8000h] bit 18 = 1), bits 15-0 are used. For 8 bpp color depths (REG[8000h] bit 18 = 0), bits 7-0 are used.

Note

For Big Endian implementations, see the *S1D13A04 Programming Notes and Examples*, document number X37A-G-003-xx.

8.6 2D Accelerator (BitBLT) Data Register Descriptions

The 2D Accelerator (BitBLT) data registers decode AB15-AB0 and require AB16 = 1. The BitBLT data registers are 32-bit wide. Byte access to the BitBLT data registers is not allowed.

2D Accelerator (BitBLT) Data Memory Mapped Region Register															
AB16-AB0 = 10000h-1FFFEh, even addresses														Read/Write	
BitBLT Data bits 31-16															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
BitBLT Data bits 15-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

bits 15-0 BitBLT Data Bits [15:0]
This register specifies the BitBLT data. This register is loosely decoded from 10000h to 1FFFEh.

9 2D Accelerator (BitBLT) Engine

9.1 Overview

The S1D13A04 is designed with a built-in 2D BitBLT engine which increases the performance of Bit Block Transfers (BitBLT). It supports 8 and 16 bit-per-pixel color depths.

The BitBLT engine supports rectangular and linear addressing modes for source and destination in a positive direction for all BitBLT operations except the move BitBLT which also supports in a negative direction.

The BitBLT operations support byte alignment of all types. The BitBLT engine has a dedicated BitBLT IO access space. This allows the BitBLT engine to support simultaneous BitBLT and host side operations.

9.2 BitBLT Operations

The S1D13A04 2D BitBLT engine supports the following BitBLTs. For detailed information on using the individual BitBLT operations, refer to the S1D13A04 Programming Notes and Examples, document number X37A-G-003-xx.

- Write BitBLT.
- Move BitBLT.
- Solid Fill BitBLT.
- Pattern Fill BitBLT.
- Transparent Write BitBLT.
- Transparent Move BitBLT.
- Read BitBLT.
- Color Expansion BitBLT.
- Move BitBLT with Color Expansion.

Note

For details on the BitBLT registers, see Section 8.5, “2D Acceleration (BitBLT) Registers (Offset = 8000h)” on page 135.

10 Frame Rate Calculation

The following formula is used to calculate the display frame rate.

$$\text{FrameRate} = \frac{f_{\text{PCLK}}}{(\text{HT}) \times (\text{VT})}$$

Where:

f_{PCLK} = PCLK frequency (Hz)

HT = Horizontal Total
= ((REG[20h] bits 6-0) + 1) x 8 Pixels

VT = Vertical Total
= ((REG[30h] bits 9-0) + 1) Lines

11 Display Data Formats

The following diagrams show the display mode data formats for a little-endian system.

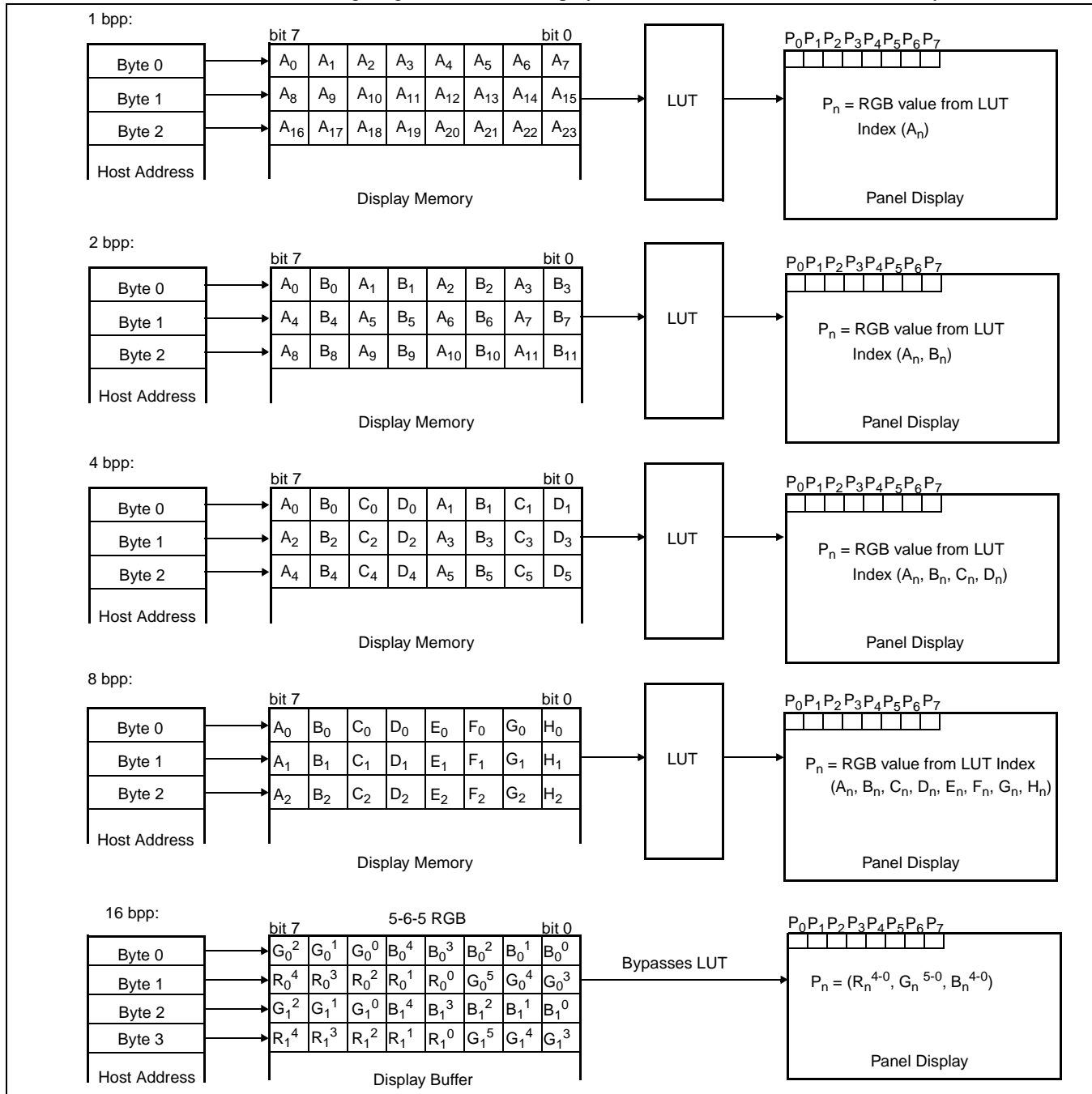


Figure 11-1: 4/8/16 Bit-Per-Pixel Display Data Memory Organization

Note

1. The Host-to-Display mapping shown here is for a little endian system.
2. For 16 bpp format, R_n, G_n, B_n represent the red, green, and blue color components.

12 Look-Up Table Architecture

The following figures are intended to show the display data output path only.

Note

When Video Data Invert is enabled the video data is inverted after the Look-Up Table.

12.1 Monochrome Modes

The green Look-Up Table (LUT) is used for all monochrome modes.

1 Bit-per-pixel Monochrome Mode

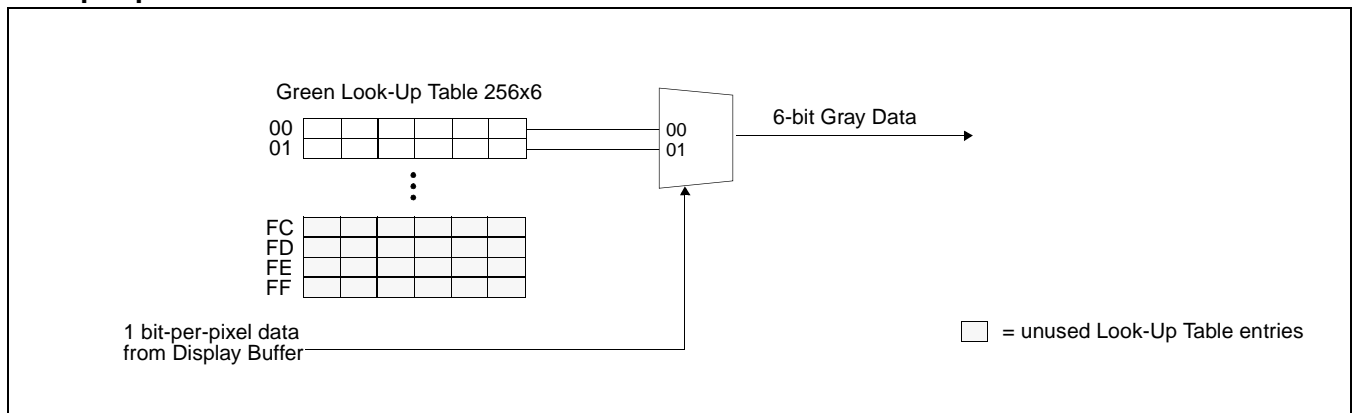


Figure 12-1: 1 Bit-per-pixel Monochrome Mode Data Output Path

2 Bit-per-pixel Monochrome Mode

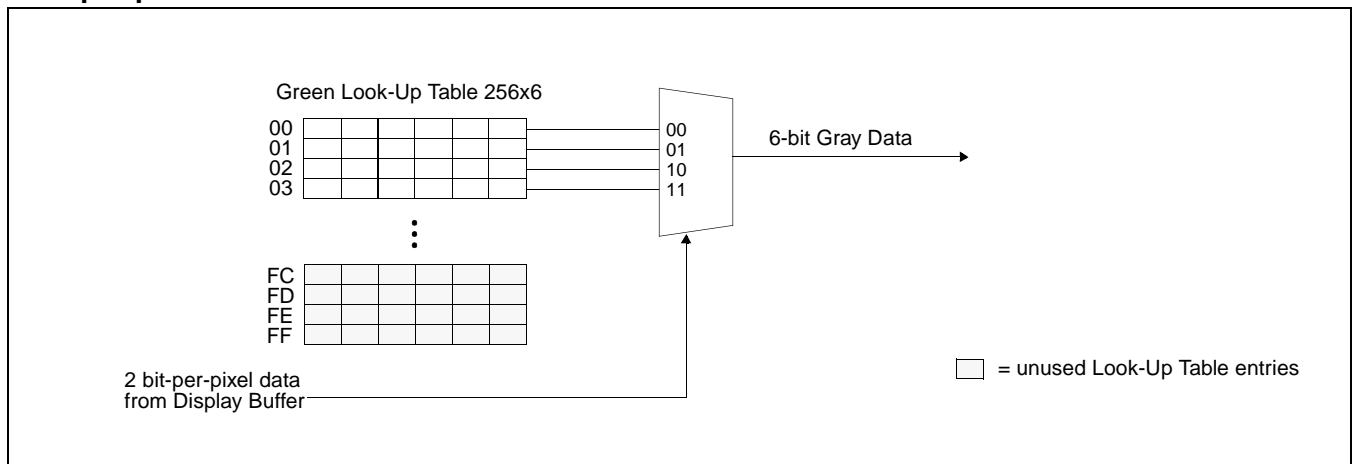


Figure 12-2: 2 Bit-per-pixel Monochrome Mode Data Output Path

4 Bit-per-pixel Monochrome Mode

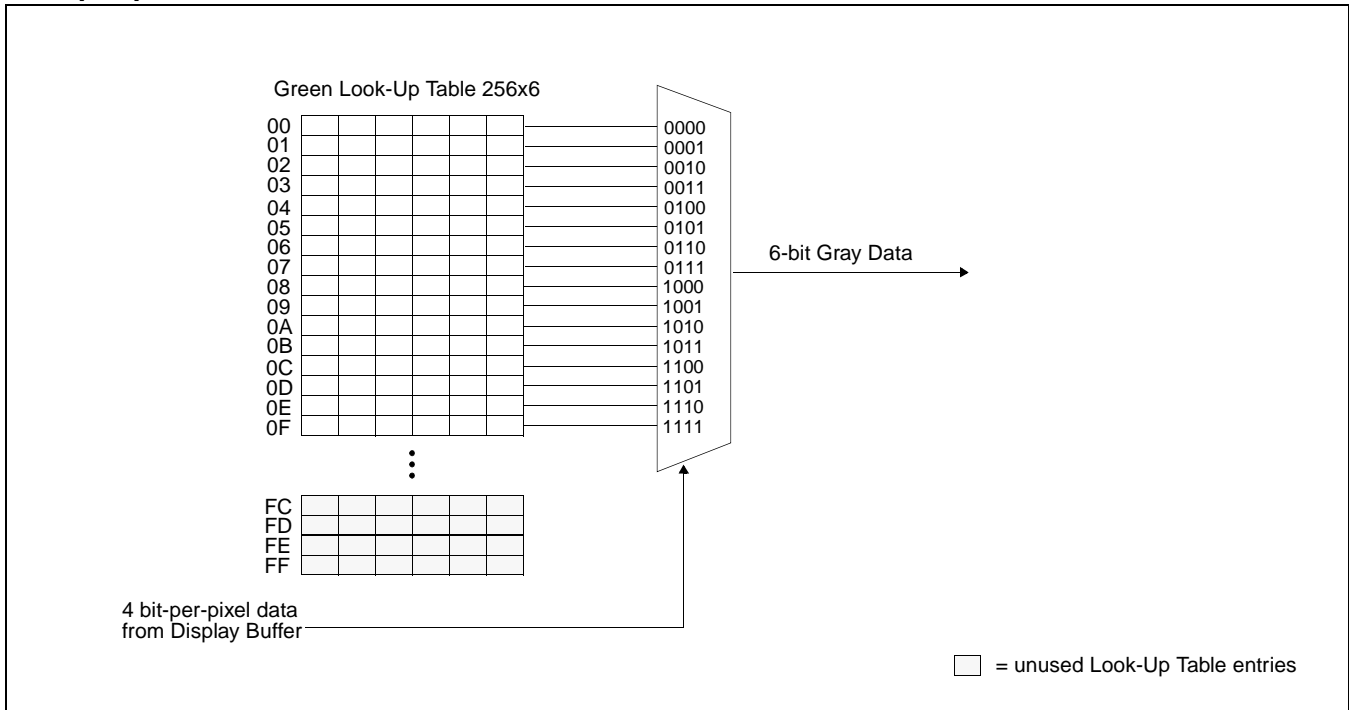


Figure 12-3: 4 Bit-per-pixel Monochrome Mode Data Output Path

8 Bit-per-pixel Monochrome Mode

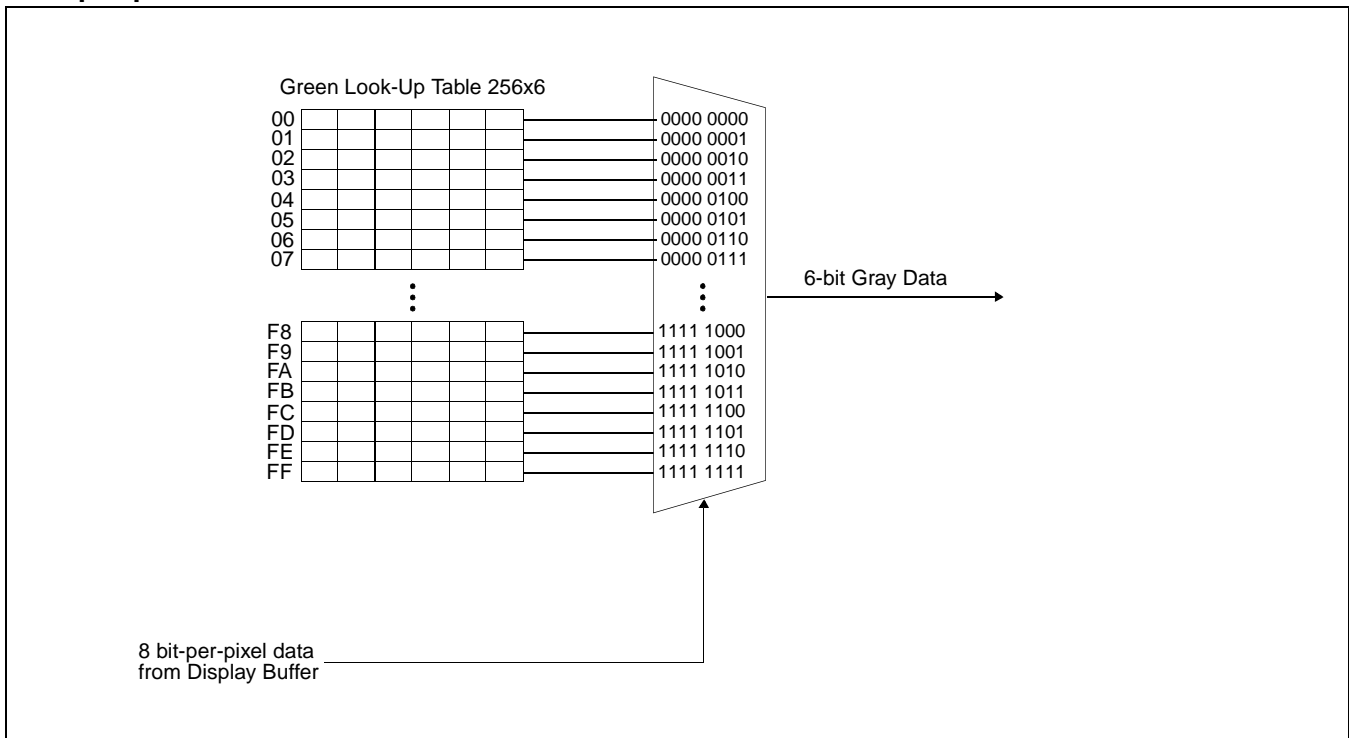


Figure 12-4: 8 Bit-per-pixel Monochrome Mode Data Output Path

16 Bit-Per-Pixel Monochrome Mode

The LUT is bypassed and the green data is directly mapped for this color depth– “Display Data Formats” on page 144..

12.2 Color Modes

1 Bit-Per-Pixel Color

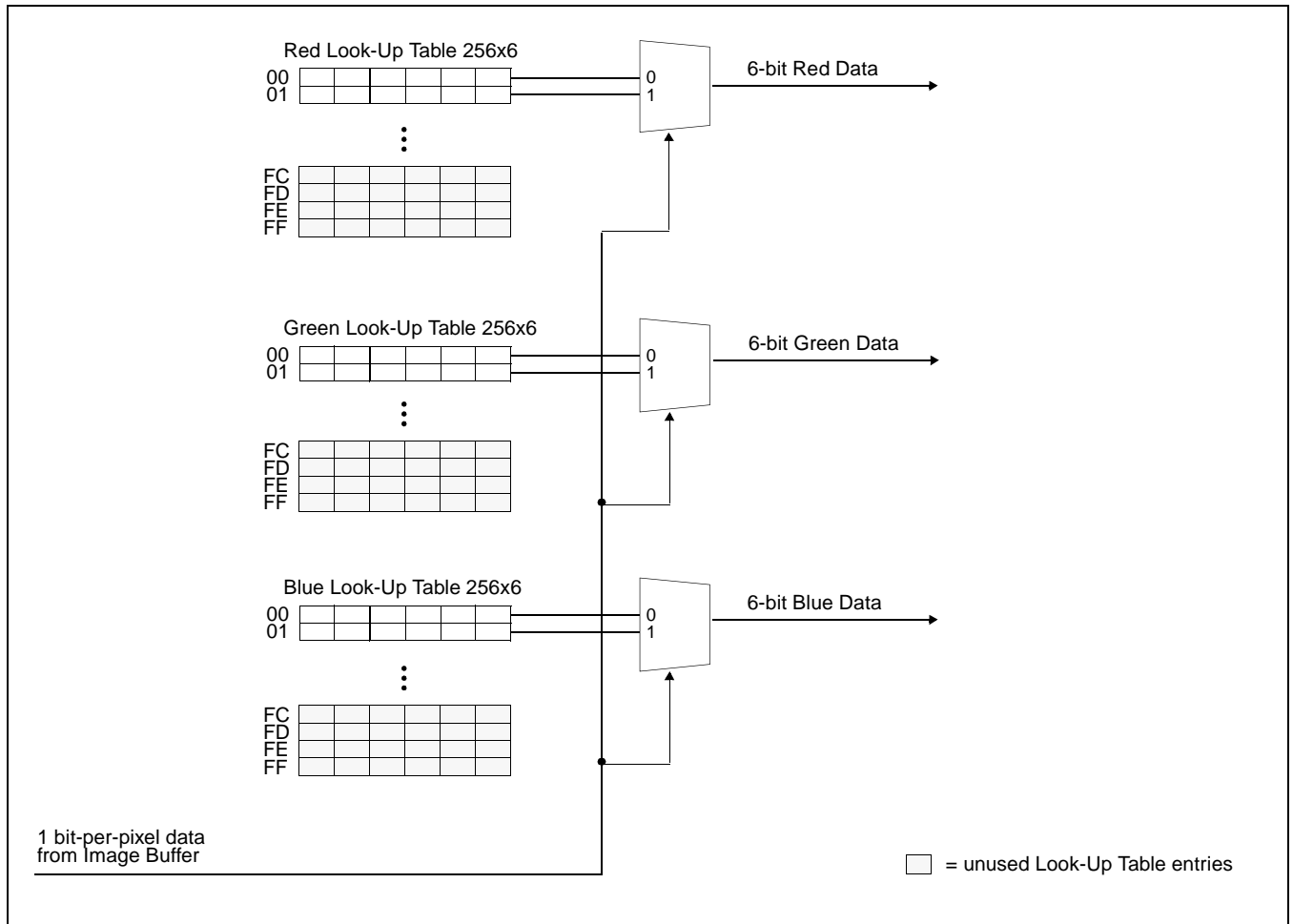


Figure 12-5: 1 Bit-Per-Pixel Color Mode Data Output Path

2 Bit-Per-Pixel Color

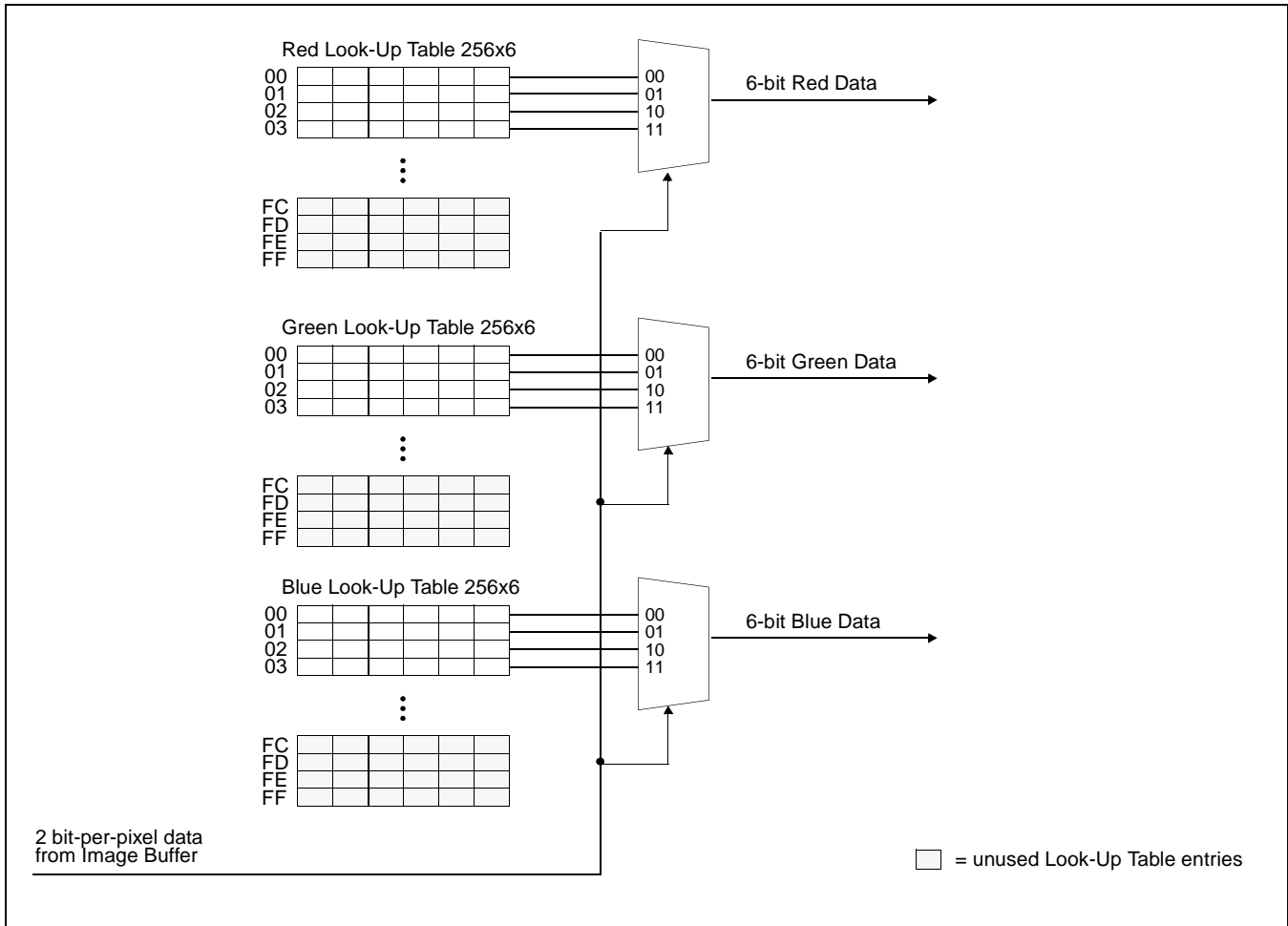


Figure 12-6: 2 Bit-Per-Pixel Color Mode Data Output Path

4 Bit-Per-Pixel Color

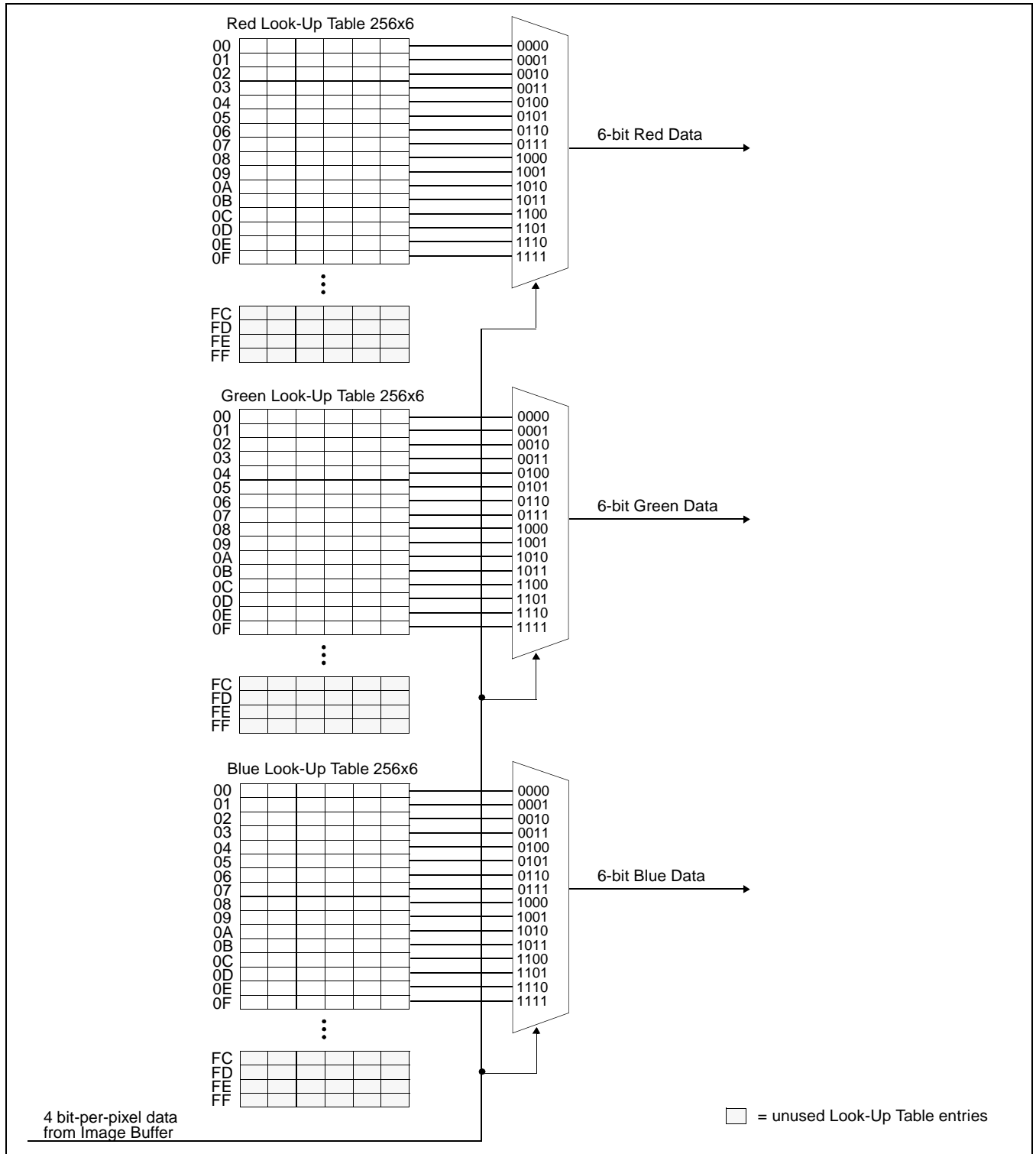


Figure 12-7: 4 Bit-Per-Pixel Color Mode Data Output Path

8 Bit-per-pixel Color Mode

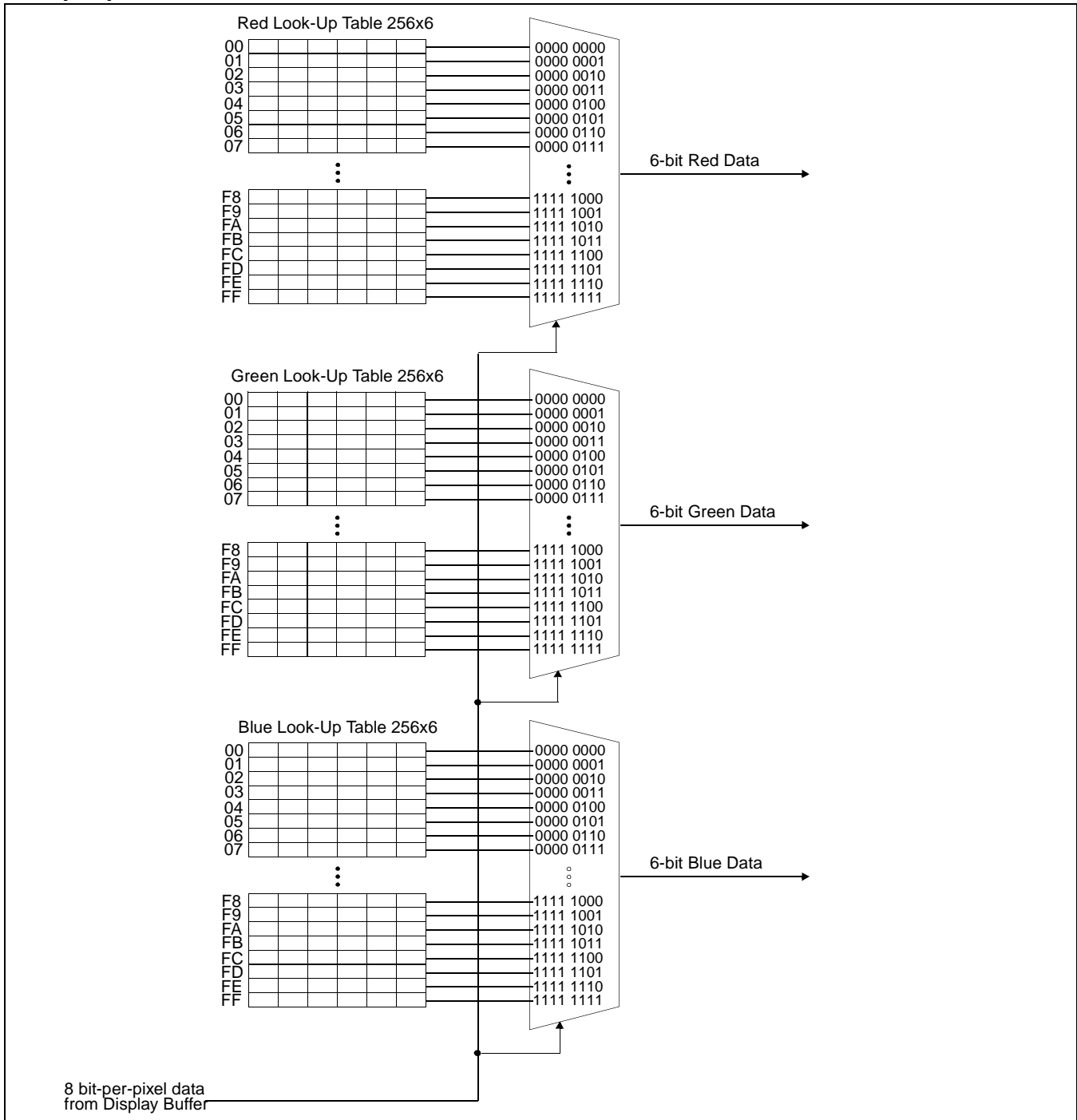


Figure 12-8: 8 Bit-per-pixel Color Mode Data Output Path

16 Bit-Per-Pixel Color Mode

The LUT is bypassed and the color data is directly mapped for this color depth— “Display Data Formats” on page 144.

13 SwivelView™

13.1 Concept

Most computer displays are refreshed in landscape orientation – from left to right and top to bottom. Computer images are stored in the same manner. SwivelView™ is designed to rotate the displayed image on an LCD by 90°, 180°, or 270° in a counter-clockwise direction. The rotation is done in hardware and is transparent to the user for all display buffer reads and writes. By processing the rotation in hardware, SwivelView™ offers a performance advantage over software rotation of the displayed image.

The image is not actually rotated in the display buffer since there is no address translation during CPU read/write. The image is rotated during display refresh.

Note

The Pixel Doubling feature of the S1D13A04 is not available in 90° and 270° Swivel-View rotations.

13.2 90° SwivelView™

90° SwivelView™ requires the Memory Clock (MCLK) to be at least 1.25 times the frequency of the Pixel Clock (PCLK), i.e. $MCLK \geq 1.25PCLK$.

The following figure shows how the programmer sees a 320x480 portrait image and how the image is being displayed. The application image is written to the S1D13A04 in the following sense: A–B–C–D. The display is refreshed by the S1D13A04 in the following sense: B-D-A-C.

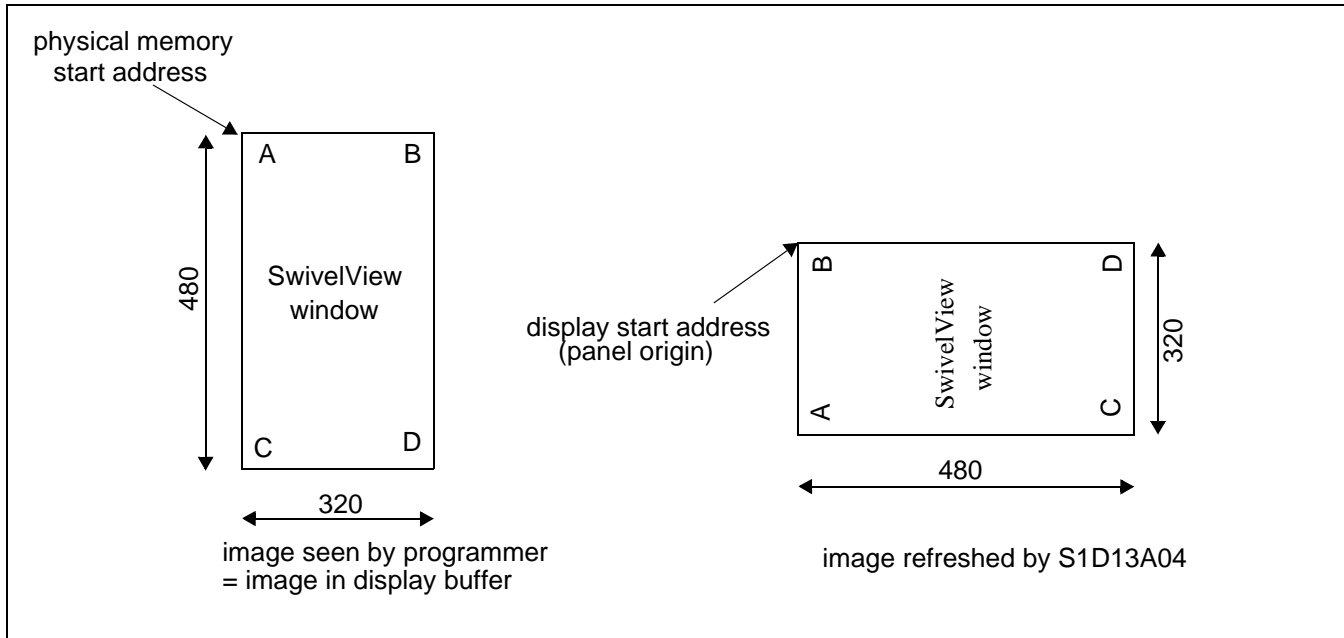


Figure 13-1: Relationship Between The Screen Image and the Image Refreshed in 90° SwivelView.

13.2.1 Register Programming

Enable 90° SwivelView™ Mode

Set SwivelView™ Mode Select bits (REG[10h] bits 17:16) to 01.

Display Start Address

The display refresh circuitry starts at pixel “B”, therefore the Main Window Display Start Address register (REG[40h]) must be programmed with the address of pixel “B”. To calculate the value of the address of pixel “B” use the following formula (assumes 8 bpp color depth).

$$\begin{aligned} \text{REG}[40\text{h}] \text{ bits } 16:0 &= ((\text{image address} + (\text{panel height} \times \text{bpp} \div 8)) \div 4) - 1 \\ &= ((0 + (320 \text{ pixels} \times 8 \text{ bpp} \div 8)) \div 4) - 1 \\ &= 79 (4\text{Fh}) \end{aligned}$$

Line Address Offset

The Main Window Line Address Offset register (REG[44h]) is based on the display width and programmed using the following formula.

$$\begin{aligned} \text{REG}[44\text{h}] \text{ bits } 9:0 &= \text{display width in pixels} \div (32 \div \text{bpp}) \\ &= 320 \text{ pixels} \div (32 \div 8 \text{ bpp}) \\ &= 80 (50\text{h}) \end{aligned}$$

13.3 180° SwivelView™

The following figure shows how the programmer sees a 480x320 landscape image and how the image is being displayed. The application image is written to the S1D13A04 in the following sense: A–B–C–D. The display is refreshed by the S1D13A04 in the following sense: D–C–B–A.

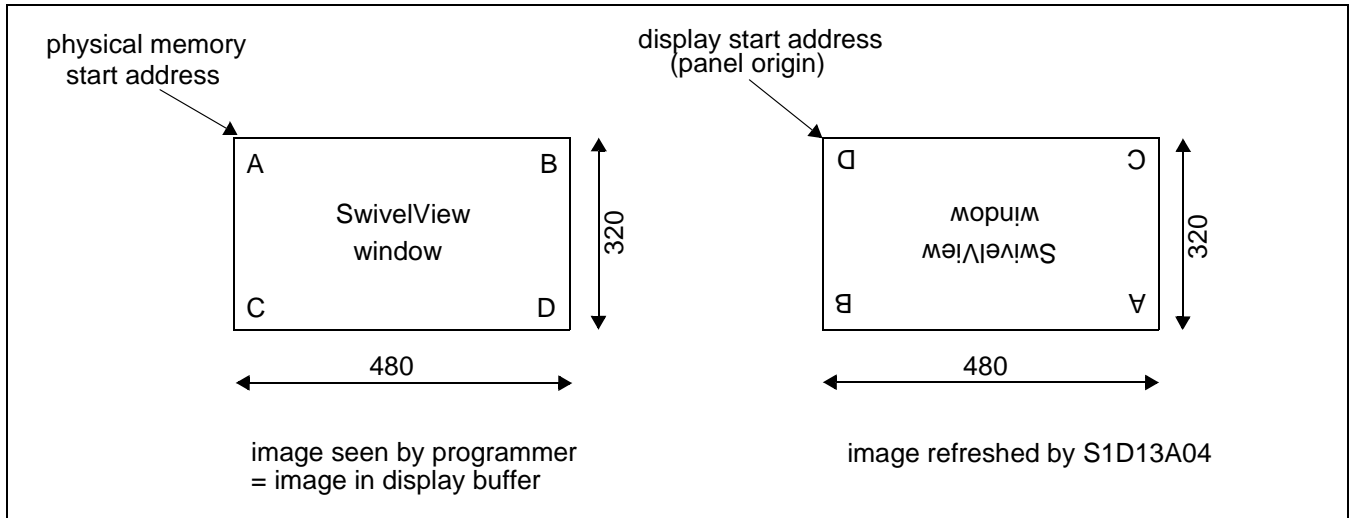


Figure 13-2: Relationship Between The Screen Image and the Image Refreshed in 180° SwivelView.

13.3.1 Register Programming

Enable 180° SwivelView™ Mode

Set SwivelView™ Mode Select bits (REG[10h] bits 17:16) to 10.

Display Start Address

The display refresh circuitry starts at pixel “D”, therefore the Main Window Display Start Address register (REG[40h]) must be programmed with the address of pixel “D”. To calculate the value of the address of pixel “D” use the following formula (assumes 8 bpp color depth).

$$\begin{aligned}
 &\text{REG}[40\text{h}] \text{ bits } 16:0 \\
 &= ((\text{image address} + (\text{offset} \times (\text{panel height} - 1) + \text{panel width}) \times \text{bpp} \div 8) \div 4) - 1 \\
 &= ((0 + (480 \text{ pixels} \times 319 \text{ pixels} + 480 \text{ pixels}) \times 8 \text{ bpp} \div 8) \div 4) - 1 \\
 &= 38399 \text{ (95FFh)}
 \end{aligned}$$

Line Address Offset

The Main Window Line Address Offset register (REG[44h]) is based on the display width and programmed using the following formula.

$$\begin{aligned} \text{REG}[44\text{h}] \text{ bits } 9:0 &= \text{display width in pixels} \div (32 \div \text{bpp}) \\ &= 480 \text{ pixels} \div (32 \div 8 \text{ bpp}) \\ &= 120 \text{ (78h)} \end{aligned}$$

13.4 270° SwivelView™

270° SwivelView™ requires the Memory Clock (MCLK) to be at least 1.25 times the frequency of the Pixel Clock (PCLK), i.e. $\text{MCLK} \geq 1.25\text{PCLK}$.

The following figure shows how the programmer sees a 320x480 portrait image and how the image is being displayed. The application image is written to the S1D13A04 in the following sense: A-B-C-D. The display is refreshed by the S1D13A04 in the following sense: C-A-D-B.

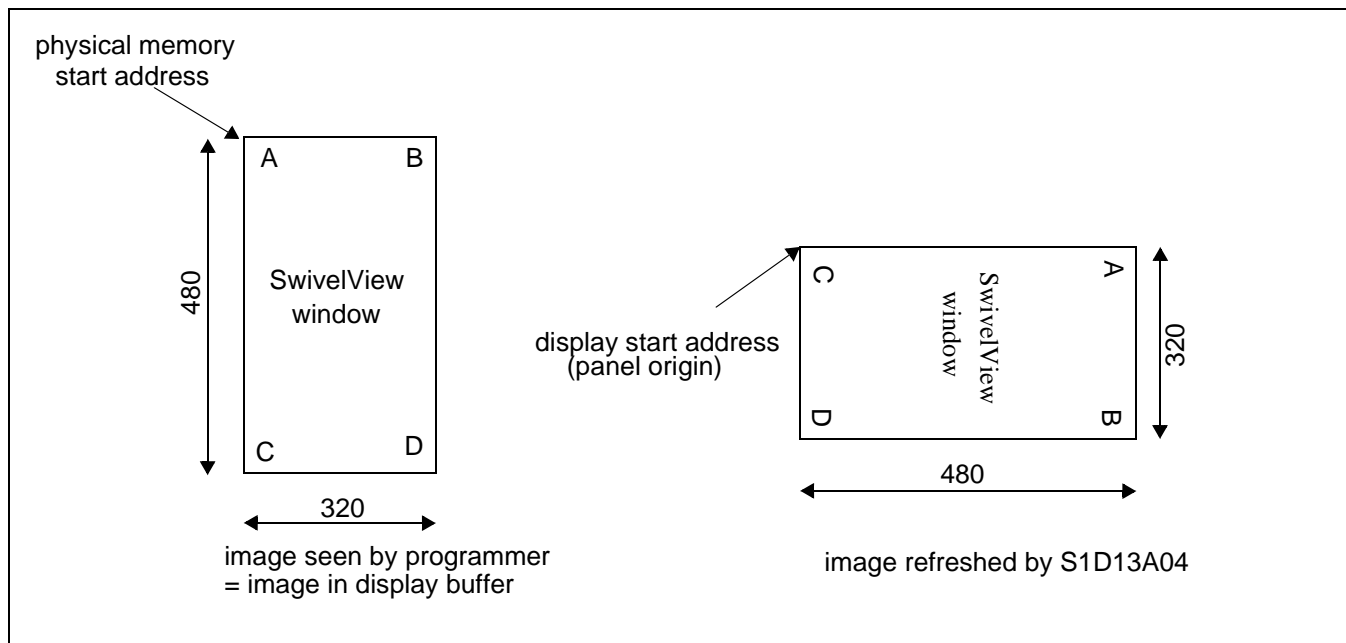


Figure 13-3: Relationship Between The Screen Image and the Image Refreshed in 270° SwivelView.

13.4.1 Register Programming

Enable 270° SwivelView™ Mode

Set SwivelView™ Mode Select bits (REG[10h] bits 17:16) to 11.

Display Start Address

The display refresh circuitry starts at pixel “C”, therefore the Main Window Display Start Address register (REG[40h]) must be programmed with the address of pixel “C”. To calculate the value of the address of pixel “C” use the following formula (assumes 8 bpp color depth).

$$\begin{aligned} \text{REG}[40\text{h}] \text{ bits } 16:0 &= (\text{image address} + ((\text{panel width} - 1) \times \text{offset} \times \text{bpp} \div 8) \div 4) \\ &= (0 + ((480 \text{ pixels} - 1) \times 320 \text{ pixels} \times 8 \text{ bpp} \div 8) \div 4) \\ &= 38320 \text{ (95B0h)} \end{aligned}$$

Line Address Offset

The Main Window Line Address Offset register (REG[44h]) is based on the display width and programmed using the following formula.

$$\begin{aligned} \text{REG}[44\text{h}] \text{ bits } 9:0 &= \text{display width in pixels} \div (32 \div \text{bpp}) \\ &= 320 \text{ pixels} \div (32 \div 8 \text{ bpp}) \\ &= 80 \text{ (50h)} \end{aligned}$$

14 Picture-in-Picture Plus (PIP⁺)

14.1 Concept

Picture-in-Picture Plus (PIP⁺) enables a secondary window (or PIP⁺ window) within the main display window. The PIP⁺ window may be positioned anywhere within the virtual display and is controlled through the PIP⁺ Window control registers (REG[50h] through REG[5Ch]). The PIP⁺ window retains the same color depth and SwivelView orientation as the main window.

The following diagram shows an example of a PIP⁺ window within a main window and the registers used to position it.

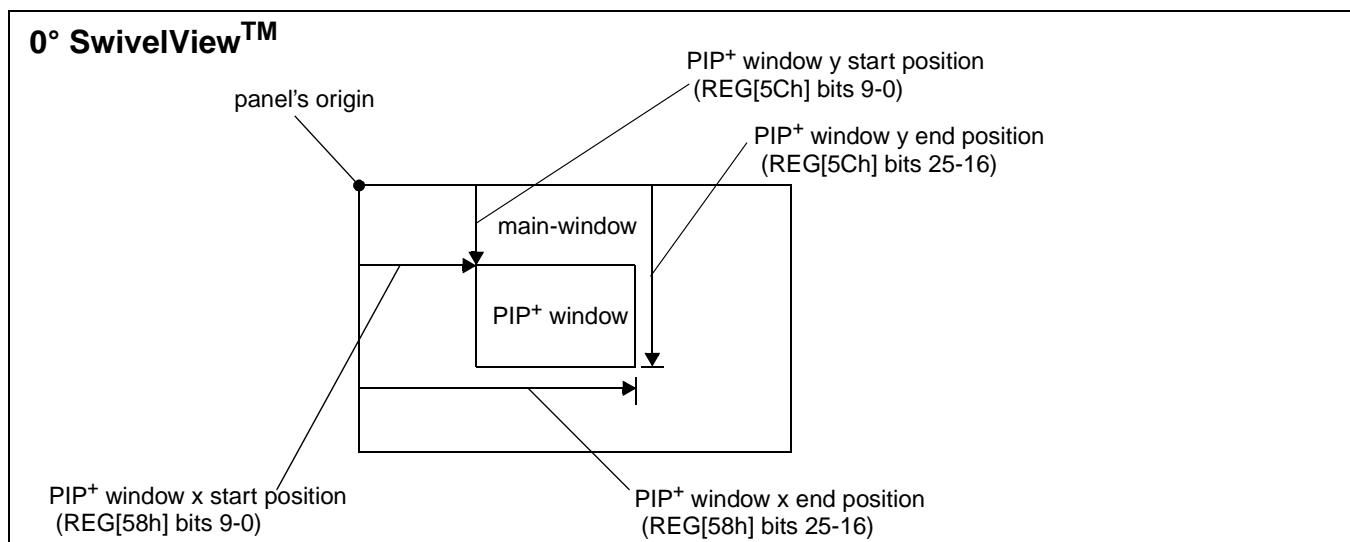


Figure 14-1: Picture-in-Picture Plus with SwivelView disabled

14.2 With SwivelView Enabled

14.2.1 SwivelView 90°

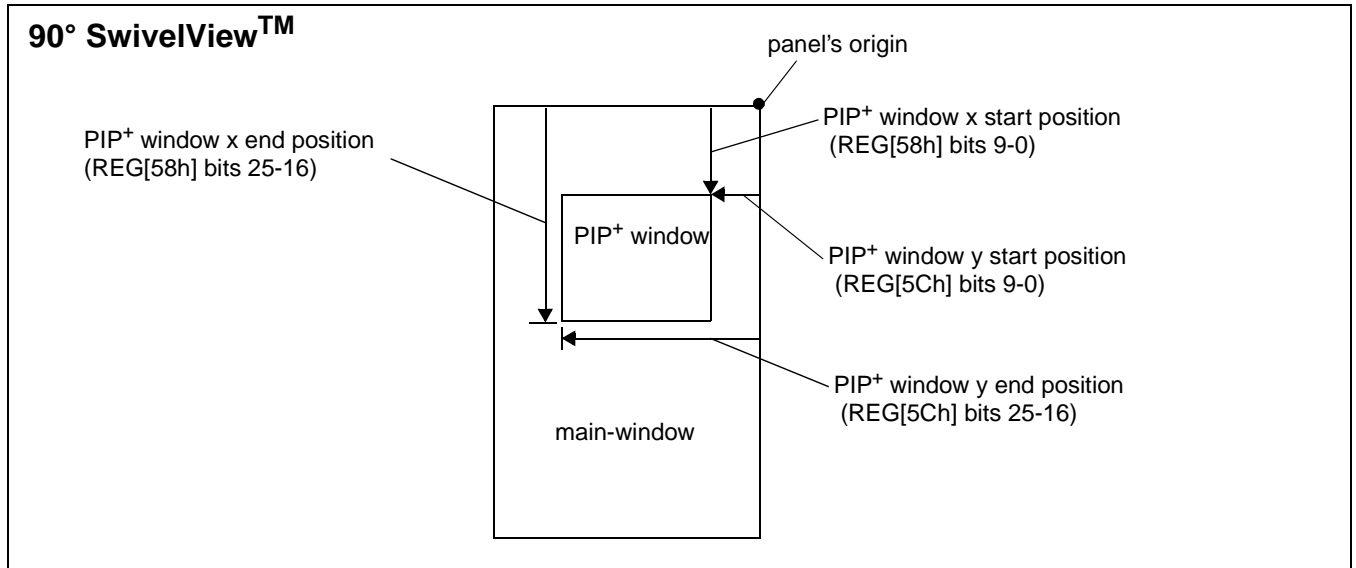


Figure 14-2: Picture-in-Picture Plus with SwivelView 90° enabled

14.2.2 SwivelView 180°

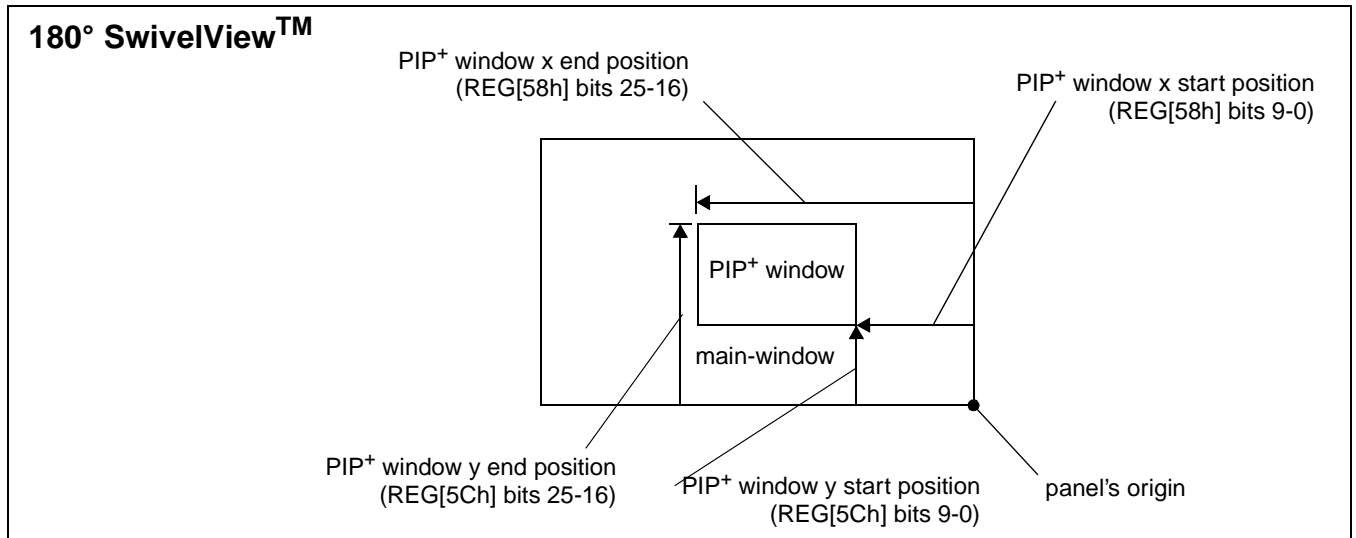


Figure 14-3: Picture-in-Picture Plus with SwivelView 180° enabled

14.2.3 SwivelView 270°

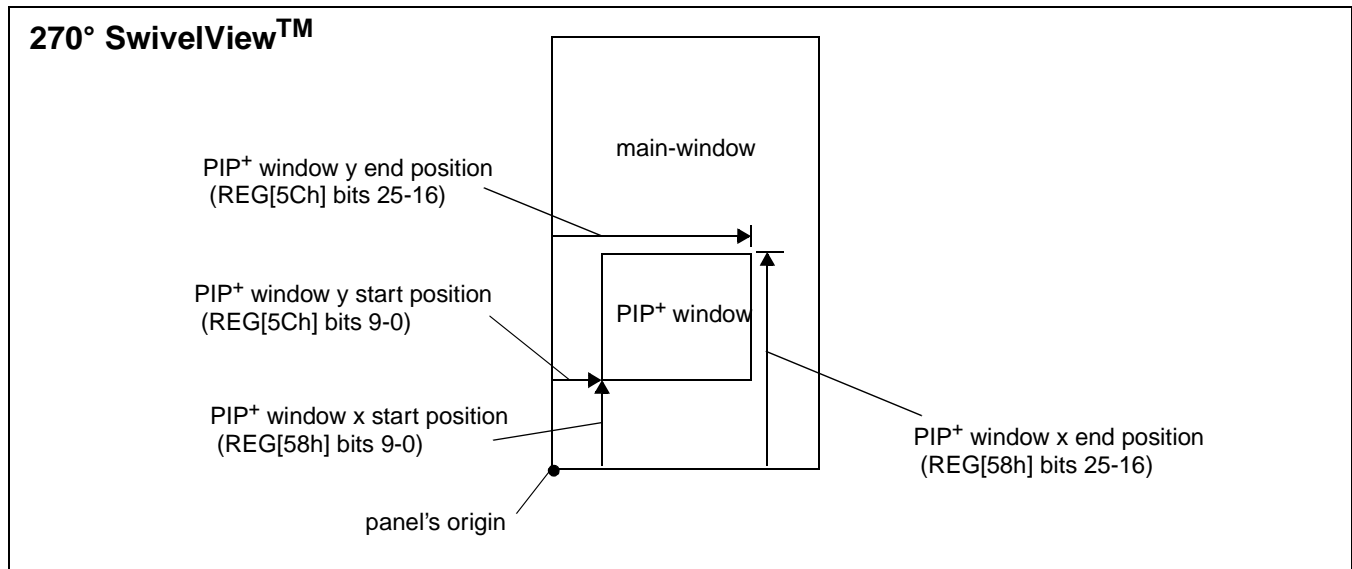


Figure 14-4: Picture-in-Picture Plus with SwivelView 270° enabled

15 Power Save Mode

A software initiated Power Save Mode is incorporated into the S1D13A04 to accommodate the need for power reduction in the hand-held devices market. This mode is enable via the Power Save Mode Enable bit (REG[14h] bit 4).

Software Power Save Mode saves power by powering down the control signals and stopping display refresh accesses to the display buffer. For programming information on disabling the clocks, see the *S1D13A04 Programming Notes and Examples*, document number X37A-G-003-xx.

Table 15-1: Power Save Mode Function Summary

	Software Power Save	Normal
IO Access Possible?	Yes	Yes
Memory Writes Possible?	Yes ¹	Yes
Memory Reads Possible?	No ¹	Yes
Look-Up Table Registers Access Possible?	Yes	Yes
USB Registers Access Possible?	No	Yes
Display Active?	No	Yes
LCD I/F Outputs	Forced Low	Active
PWMCLK	Stopped	Active
Access Possible for GPIO pins configured for HR-TFT?	Forced Low	Active
Access Possible for GPIO Pins configured as GPIOs?	Yes ²	Yes
USB Running?	No	Yes

Note

¹ When power save mode is enabled, the memory controller is powered down and the status of the memory controller is indicated by the Memory Controller Power Save Status bit (REG[14h] bit 6). However, memory writes are possible during power save mode because the S1D13A04 dynamically enables the memory controller for display buffer writes. **This ability does not increase power consumption.**

²GPIOs can be accessed, and if configured as outputs can be changed.

After reset, the S1D13A04 is always in Power Save Mode. Software must initialize the chip (i.e. programs all registers) and then clear the Power Save Mode Enable bit. For further details, see the register description for REG[14h] bit 4.

16 Mechanical Data

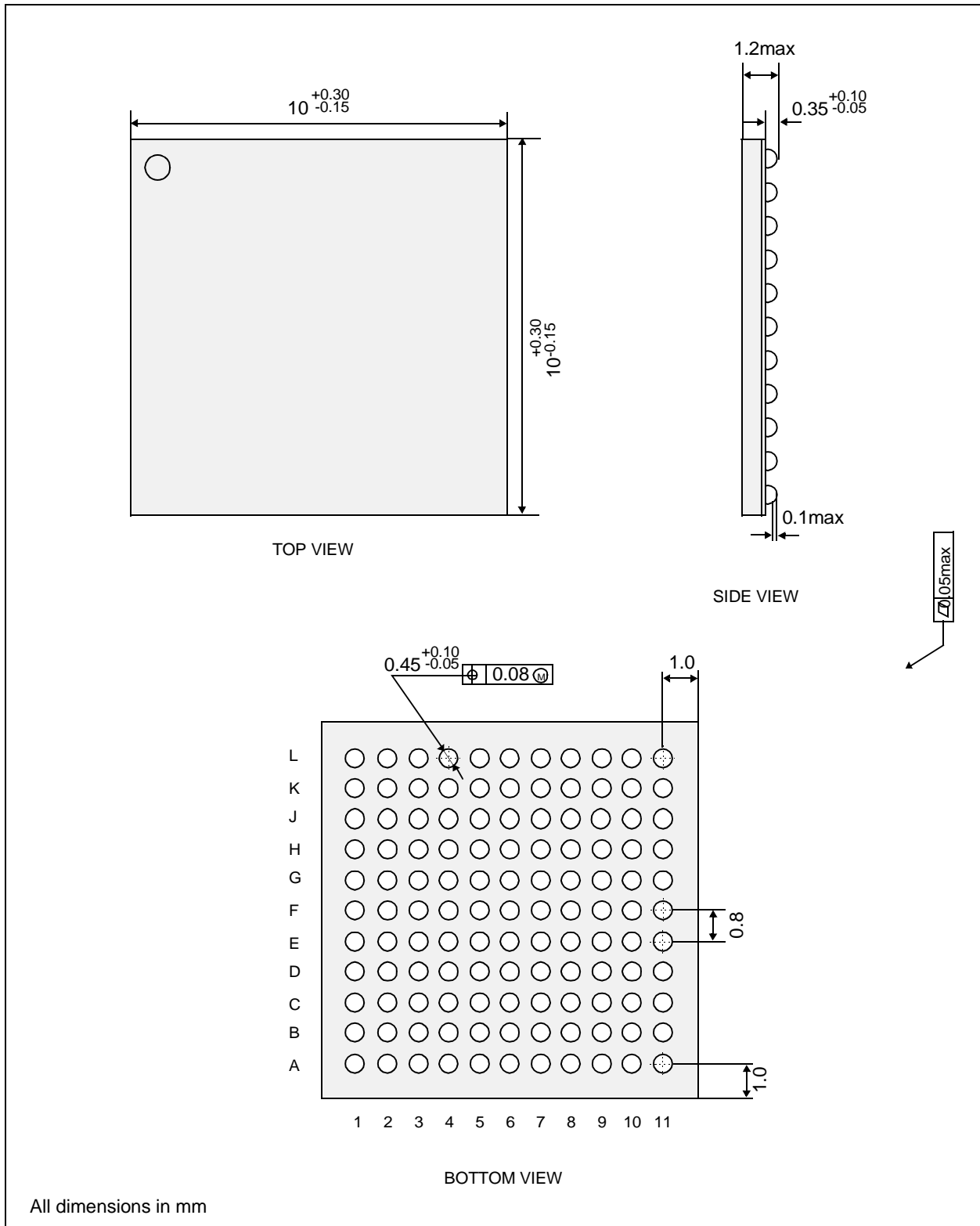
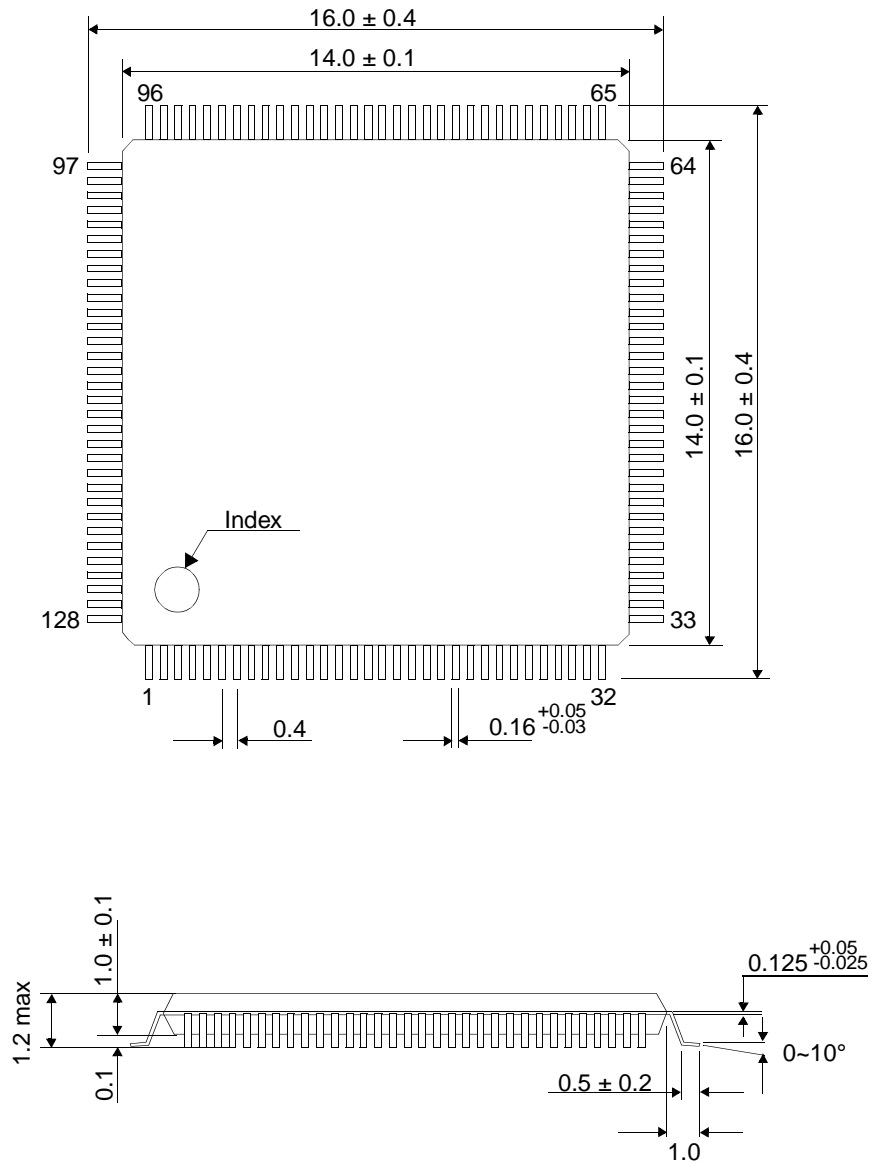


Figure 16-1: Mechanical Data PFBGA 121-pin Package



All dimensions in mm

Figure 16-2: Mechanical Data TQFP15 128-pin Package

17 References

The following documents contain additional information related to the S1D13A04. Document numbers are listed in parenthesis after the document name. All documents can be found at the Epson Research and Development Website at www.erd.epson.com.

- 13A04CFG Configuration Utility Users Manual (X37A-B-001-xx)
- 13A04PLAY Diagnostic Utility Users Manual (X37A-B-002-xx)
- 13A04BMP Demonstration Program User Manual (X37A-B-003-xx)
- S1D13A04 Product Brief (X37A-C-001-xx)
- S1D13A04 Wind River WindML v2.0 Display Drivers (X37A-E-002-xx)
- S1D13A04 Linux Console Driver (X37A-E-004-xx)
- S1D13A04 QNX Photon v2.0 Display Drivers (X37A-E-005-xx)
- S1D13A04 Windows CE v3.x Display Drivers (X37A-E-006-xx)
- Interfacing to the Toshiba TMPR3905/3912 Microprocessor (X37A-G-002-xx)
- S1D13A04 Programming Notes And Examples (X37A-G-003-xx)
- S5U13A04B00C Rev. 1.0 Evaluation Board User Manual (X37A-G-004-xx)
- Interfacing to the PC Card Bus (X37A-G-005-xx)
- S1D13A04 Power Consumption (X37A-G-006-xx)
- Interfacing to the NEC VR4102/VR4111 Microprocessors (X37A-G-007-xx)
- Interfacing to the NEC VR4181 Microprocessor (X37A-G-008-xx)
- Interfacing to the Motorola MPC821 Microprocessor (X37A-G-009-xx)
- Interfacing to the Motorola MCF5307 "Coldfire" Microprocessors (X37A-G-010-xx)
- Connecting to the Sharp HR-TFT Panels (X37A-G-011-xx)
- Interfacing to the Motorola MC68VZ328 Dragonball Microprocessor (X37A-G-012-xx)
- Interfacing to the Intel StrongARM SA-1110 Microprocessor (X37A-G-013-xx)
- S1D13A04 Register Summary (X37A-R-001-xx)

18 Sales and Technical Support

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Errata No. X00Z-P-001-01

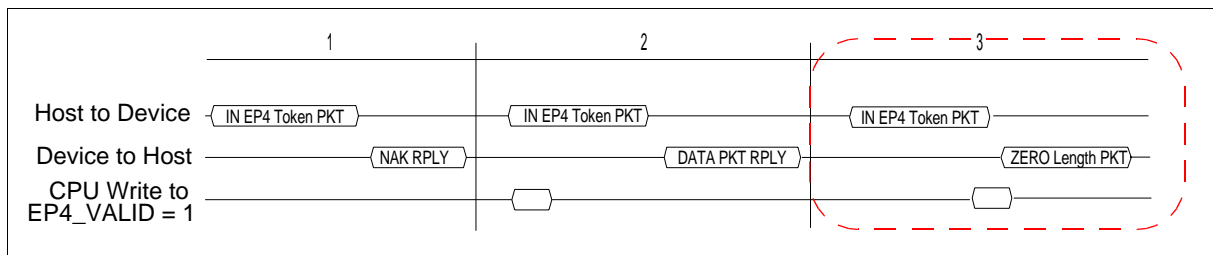
Device: S1D13A03, S1D13A04, S1D13A05.

Description: Setting EP4 FIFO Valid bit while NAKing an IN token.

Bit 5 of REG[402Ch] indicates to the S1D13A0x controller when data in the endpoint 4 FIFO is ready to be transferred to the host computer. Changing the state of this bit at certain times may generate an error.

When the S1D13A0x USB controller receives an endpoint 4 IN request and endpoint 4 is not ready to transmit data (REG[402Ch] bit 5 = 0), the response is a NAK packet. If endpoint 4 is toggled to a ready to transmit state just before a NAK response packet is sent, the controller may erroneously send a zero length packet instead. When this happens, the data toggle state will be incorrectly set for the next endpoint 4 data transmit.

The following timing diagram shows the error occurring in section 3.



This unexpected occurrence of a zero length packet may cause file system handling errors for some operating systems.

Corrective Action:

There are two software solutions for this occurrence.

Disable USB Receiver before setting the EP4 FIFO Valid bit

The first solution involves disabling the USB receiver to avoid responding to an EP4 IN packet. During the time the USB receiver is disabled the EP4 FIFO Valid bit is set.

When the local CPU is ready to send data on endpoint 4 the steps to follow are:

1. Disable the USB differential input receiver (REG[4040h] bit 6 = 0)
2. Wait a minimum of 1 μ s. If needed, delays may be added
3. Enable the EP4 FIFO Valid bit (REG[402Ch] bit 5 = 1)
4. Clear the EP4 Interrupt status bit (REG[4004h] bit 4 = 1)
5. Enable the USB differential input receiver (REG[4040h] bit 6 = 1)

Note

Steps 1 through 5 are time critical and must be performed in less than 6 μ s.

Note

To comply with “EP4 NAK Status not set correctly in USB Status register”, steps 3 and 4 must be completed within 5 μ s of each other. For further information on “EP4 NAK Status not set correctly in USB Status register”, see the S1D13A0x Programming Notes and Examples, document numbers X36A-G-003-xx, X37A-G-003-xx, and X40A-G-003-xx.

EP4 FIFO Valid bit set after NAK and before the next IN token.

The second solution is to wait until immediately after the USB has responded to an IN request with a NAK packet before setting the transmit FIFO valid bit. This solution is recommended only for fast processors.

When the local CPU is ready to send data on endpoint 4, it must first detect that a NAK packet has been sent. This is done by reading the EP4 Interrupt Status bit (REG[4004h] bit 4). If the EP4 FIFO Valid bit was not set, the EP4 Interrupt Status bit is set only if a NAK packet has been sent. When the local CPU detects the NAK it must immediately set the EP4 FIFO Valid bit (before responding to the next IN token).

After filling the EP4 FIFO the steps to follow before setting the EP4 FIFO Valid bit are:

1. Clear the EP4 Interrupt Status bit (REG[4004h] bit 4)
2. Read the EP4 Interrupt Status bit (REG[4004h] bit 4) until it is set
3. Set the EP4 FIFO Valid bit (REG[402Ch] bit 5 = 1)

Note

The setting of the EP4 FIFO Valid bit is time critical. The EP4 FIFO Valid bit must be set within 3 μ s after the EP4 Interrupt Status has been set internally by the S1D13A0x.

EPSON®



S1D13A04 LCD/USB Companion Chip

Programming Notes and Examples

Document Number: X37A-G-003-05

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1 Introduction

This guide discusses programming issues and provides examples for the main features of the S1D13A04, such as SwivelView, Picture-in-Picture Plus, and the BitBLT engine. The example source code referenced in this guide is available on the web at www.erd.epson.com.

This guide also introduces the Hardware Abstraction Layer (HAL), which is designed to simplify the programming of the S1D13A04. Most S1D13xxx products have HAL support, thus allowing OEMs to do multiple designs with a common code base.

This document is updated as appropriate. Please check the Epson Research and Development website at www.erd.epson.com for the latest revision of this document and source before beginning any development.

We appreciate your comments on our documentation. Please contact us via email at documentation@erd.epson.com.

2 Identifying the S1D13A04

The S1D13A04 can be identified by reading the value contained in the Product Information Register (REG[00h]). To identify the S1D13A04 follow the steps below.

1. Read REG[00h].
2. The production version of the S1D13A04 returns a value of 2Cxx282Ch (where xx depends on the configuration of the CNF[6:0] pins). This value can be broken down into the following.
 1. The product code for the S1D13A04 is 0Bh (001011 binary) and can be found in bits 7-2 and also in bits 31-26.
 2. The revision code is 0h (00 binary) and can be found in bits 1-0 and again in bits 25-24.
 3. The display buffer size is 28h (00101000 binary) and is contained in bits 15-8.

Note

The display buffer size is the distinguishing value between the S1D13A03 and the S1D13A04 as they share the same product code and revision code. For the correct display buffer size for the S1D13A03, see the *S1D13A03 Hardware Functional Specification*, document number X36A-A-001-xx.

3 Initialization

This section describes how to initialize the S1D13A04. Sample code for performing initialization of the S1D13A04 is provided in the file **init13A04.c** which is available on the internet at www.erd.epson.com.

S1D13A04 initialization can be broken into the following steps.

1. Set all registers to initial values. The values are obtained by using the `s1d13A04.h` file that is exported by the **13A04CFG.EXE** configuration utility. For more information on 13A04CFG, see the *13A04CFG User Manual*, document number X37A-B-001-xx.
2. Program the Look-Up Table (LUT) with color values. For details on programming the LUT, see Section 5, “Look-Up Table (LUT)” on page 17.
3. Clear the display buffer.

If the system implementation uses a clock chip instead of a fixed oscillator, refer to the HAL (Hardware Abstraction Layer) sample code available on the internet at www.erd.epson.com. For example, the Epson S5U13A04B00C evaluation board uses a Cypress clock chip.

4 Memory Models

The S1D13A04 contains a display buffer of 160K bytes and supports color depths of 1, 2, 4, 8, and 16 bit-per-pixel. For each color depth, the data format is packed pixel.

Packed pixel data may be envisioned as a stream of pixels. In this stream, pixels are packed adjacent to each other. If a pixel requires four bits, then it is located in the four most significant bits of a byte. The pixel to the immediate right on the display occupies the lower four bits of the same byte. The next two pixels to the immediate right are located in the following byte, etc.

4.1 Display Buffer Location

The S1D13A04 display buffer is 160K bytes of embedded SRAM. The display buffer is memory mapped and is accessible directly by software. The memory block location assigned to the S1D13A04 display buffer varies with each individual hardware platform.

For further information on the display buffer, see the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

For further information on the S1D13A04 Evaluation Board, see the *S5U13A04B00C Evaluation Board Rev. 1.0 User Manual*, document number X37A-G-004-xx.

4.2 Memory Organization for One Bit-per-pixel (2 Colors/Gray Shades)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Pixel 0	Pixel 1	Pixel 2	Pixel 3	Pixel 4	Pixel 5	Pixel 6	Pixel 7

Figure 4-1: Pixel Storage for 1 Bpp in One Byte of Display Buffer

At a color depth of 1 bpp, each byte of display buffer contains eight adjacent pixels. Setting or resetting any pixel requires reading the entire byte, masking out the unchanged bits and setting the appropriate bits to 1.

One bit pixels provide 2 gray shades/color possibilities. For monochrome panels the gray shades are generated by indexing into the first two elements of the green component of the Look-Up Table (LUT). For color panels the 2 colors are derived by indexing into the first 2 positions of the LUT.

4.3 Memory Organization for Two Bit-per-pixel (4 Colors/Gray Shades)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Pixel 0 bits 1-0		Pixel 1 bits 1-0		Pixel 2 bits 1-0		Pixel 3 bits 1-0	

Figure 4-2: Pixel Storage for 2 Bpp in One Byte of Display Buffer

At a color depth of 2 bpp, each byte of display buffer contains four adjacent pixels. Setting or resetting any pixel requires reading the entire byte, masking out the unchanged bits and setting the appropriate bits to 1.

Two bit pixels provide 4 gray shades/color possibilities. For monochrome panels the gray shades are generated by indexing into the first 4 elements of the green component of the Look-Up Table (LUT). For color panels the 4 colors are derived by indexing into the first 4 positions of the LUT.

4.4 Memory Organization for Four Bit-per-pixel (16 Colors/Gray Shades)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Pixel 0 bits 3-0				Pixel 1 bits 3-0			

Figure 4-3: Pixel Storage for 4 Bpp in One Byte of Display Buffer

At a color depth of 4 bpp, each byte of display buffer contains two adjacent pixels. Setting or resetting any pixel requires reading the entire byte, masking out the upper or lower nibble (4 bits) and setting the appropriate bits to 1.

Four bit pixels provide 16 gray shades/color possibilities. For monochrome panels the gray shades are generated by indexing into the first 16 elements of the green component of the Look-Up Table (LUT). For color panels the 16 colors are derived by indexing into the first 16 positions of the LUT.

4.5 Memory Organization for 8 Bpp (256 Colors/64 Gray Shades)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Pixel 0 bits 7-0							

Figure 4-4: Pixel Storage for 8 Bpp in One Byte of Display Buffer

At a color depth of 8 bpp, each byte of display buffer represents one pixel on the display. At this color depth the read-modify-write cycles are eliminated making the update of each pixel faster.

Each byte indexes into one of the 256 positions of the LUT. The S1D13A04 LUT supports six bits per primary color. This translates into 256K possible colors when color mode is selected. Therefore the display has 256 colors available out of a possible 256K colors.

When a monochrome panel is selected, the green component of the LUT is used to determine the intensity. The green indices, with six bits, can resolve 64 gray shades. Display memory values > 64 are truncated. Thus a display memory value of 65 (1000 0001) displays the same intensity as a display memory value of 1.

4.6 Memory Organization for 16 Bpp (65536 Colors/64 Gray Shades)

Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8
Red Component bits 4-0				Green Component bits 5-3			
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Green Component bits 2-0			Blue Component bits 4-0				

Figure 4-5: Pixel Storage for 16 Bpp in Two Bytes of Display Buffer

At a color depth of 16 bpp the S1D13A04 is capable of displaying 64K (65536) colors. The 64K color pixel is divided into three parts: five bits for red, six bits for green, and five bits for blue. In this mode the LUT is bypassed and output goes directly into the Frame Rate Modulator.

Should monochrome mode be chosen at this color depth, the output sends the six bits of the green LUT component to the modulator for a total of 64 possible gray shades.

Note

This operation is similar to 8 bpp, but it requires twice as much memory for the display.

5 Look-Up Table (LUT)

This section discusses programming the S1D13A04 Look-Up Table (LUT). Included is a summary of the LUT registers, recommendations for color/gray shade LUT values, and additional programming considerations. For a discussion of the LUT architecture, refer to the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

The S1D13A04 is designed with a LUT consisting of 256 indexed red/green/blue entries. Each LUT entry is six bits wide. The color depth (bpp) determines how many indices are used. For example, 1 bpp uses the first 2 indices, 2 bpp uses the first 4 indices, 4 bpp uses the first 16 indices and 8 bpp uses all 256 indices. 16 bpp bypasses the LUT.

In color modes, the pixel values stored in the display buffer index directly to an RGB value stored in the LUT. In monochrome modes, the pixel value indexes into the green component of the LUT and the amount of green at that index controls the intensity.

5.1 Registers

5.1.1 Look-Up Table Write Register

Look-Up Table Write Register														Write Only	
REG[18h]														Default = 00000000h	
LUT Write Address								LUT Red Write Data						n/a	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
LUT Green Write Data						n/a		LUT Blue Write Data						n/a	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

This register receives the data to be written to the red (bits 23-18), green (bits 15-10), and blue (bits 7-2) components of the Look-Up Table (LUT). Also contained in this register is the LUT Write Address (bits 31-24) which forms a pointer to the location in the LUT where the RGB components will be written.

This register should be accessed using a 32-bit write cycle to ensure proper operation. If the Look-Up Table Write Register is accessed with 8 or 16-bit write, it is important to understand that the LUT data is latched into the LUT only after the completion of the write to the LUT Write Address bits. On little endian systems, this means a write to bits 31-24. On big endian systems, this means a write to bits 7-2.

This is a write-only register and returns 00h if read.

Note

For further information on the S1D13A04 LUT architecture, see the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

5.1.2 Look-Up Table Read Registers

Look-Up Table Read Register																	
REG[1Ch] Default = 00000000h								Write Only (bits 31-24)/Read Only									
LUT Read Address (write only)								LUT Red Read Data						n/a			
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
LUT Green Read Data								n/a		LUT Blue Read Data						n/a	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		

This register contains the data returned from the red (bits 23-18), green (bits 15-10), and blue (bits 7-2) components of the Look-Up Table (LUT). Also contained in this register is the LUT Read Address (bits 31-24) which forms a pointer to the location in the LUT where the RGB components are read from.

Reading the LUT is a two step process. First the desired index must be set by writing the LUT Read Address bits with the desired index. Second, the LUT values are retrieved by reading from the Look-Up Table Read Register.

Bits 31-24 are write only and will return 00h if read.

Note

For further information on the S1D13A04 LUT architecture, see the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

5.2 Look-Up Table Organization

- The Look-Up Table treats the value of a pixel as an index into an array. For example, a pixel value of zero would point to the first LUT entry, whereas a pixel value of seven would point to the eighth LUT entry.
- The value contained in each LUT entry represents the intensity of the given color or gray shade. This intensity can range in value from 0 to 3Fh.
- The S1D13A04 Look-Up Table is linear. This means increasing the LUT entry number results in a brighter color or gray shade. For example, a LUT entry of FCh in the red bank results in bright red output while a LUT entry of 1Ch results in dull red.

Table 5-1: Look-Up Table Configurations

Color Depth	Look-Up Table Indices Used			Effective Gray Shades/Colors
	RED	GREEN	BLUE	
1 bpp gray		2		2 gray shades
2 bpp gray		4		4 gray shades
4 bpp gray		16		16 gray shades
8 bpp gray		64		64 gray shades
16 bpp gray				64 gray shades
1 bpp color	2	2	2	2 colors
2 bpp color	4	4	4	4 colors
4 bpp color	16	16	16	16 colors
8 bpp color	256	256	256	256 colors
16 bpp color				65536 colors

= Indicates the Look-Up Table is not used for that display mode

5.2.1 Gray Shade Modes

Gray shade (monochrome) modes are defined by the Color/Mono Panel Select bit (REG[0Ch] bit 6). When this bit is set to 0, the value output to the panel is derived solely from the green component of the LUT.

For each gray shade a table of sample LUT values is provided. These LUT values are a standardized set of intensities used by the Epson S1D13A04 utility programs.

Note

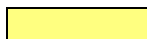
These LUT values carry eight bits of significance. The S1D13A04 LUT uses only the six MSB. The 2 LSB are ignored.

1 bpp gray shade

The 1 bpp gray shade mode uses the green component of the first 2 LUT entries. The remaining indices of the LUT are unused.

Table 5-2: Suggested LUT Values for 1 Bpp Gray Shade

Index	Red	Green	Blue
00	00	00	00
01	00	FF	00
02	00	00	00
...	00	00	00
FF	00	00	00



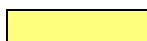
Unused entries

2 bpp gray shade

The 2 bpp gray shade mode uses the green component of the first 4 LUT entries. The remaining indices of the LUT are unused.

Table 5-3: Suggested LUT Values for 4 Bpp Gray Shade

Index	Red	Green	Blue
00	00	00	00
01	00	55	00
02	00	AA	00
03	00	FF	00
04	00	00	00
...	00	00	00
FF	00	00	00



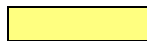
Unused entries

4 bpp gray shade

The 4 bpp gray shade mode uses the green component of the first 16 LUT entries. The remaining indices of the LUT are unused.

Table 5-4: Suggested LUT Values for 4 Bpp Gray Shade

Index	Red	Green	Blue
00	00	00	00
01	00	10	00
02	00	20	00
03	00	34	00
04	00	44	00
05	00	54	00
06	00	68	00
07	00	78	00
08	00	88	00
09	00	9C	00
0A	00	AC	00
0B	00	BC	00
0C	00	CC	00
0D	00	DC	00
0E	00	EC	00
0F	00	FC	00
10	00	00	00
...	00	00	00
FF	00	00	00



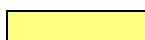
Unused entries

8 bpp gray shade

When configured for 8 bpp gray shade mode, the green component of all 256 LUT entries may be used. However, this results in redundant values where each of the 256 pixel values can only be mapped into 1 of 64 gray shades.

Table 5-5: Suggested LUT Values for 8 Bpp Gray Shade

Index	Red	Green	Blue	Index	Red	Green	Blue
00	00	00	00	20	00	80	00
01	00	04	00	21	00	84	00
02	00	08	00	22	00	88	00
03	00	0C	00	23	00	8C	00
04	00	10	00	24	00	90	00
05	00	14	00	25	00	94	00
06	00	18	00	26	00	98	00
07	00	1C	00	27	00	9C	00
08	00	20	00	28	00	A0	00
09	00	24	00	29	00	A4	00
0A	00	28	00	2A	00	A8	00
0B	00	2C	00	2B	00	AC	00
0C	00	30	00	2C	00	B0	00
0D	00	34	00	2D	00	B4	00
0E	00	38	00	2E	00	B8	00
0F	00	3C	00	2F	00	BC	00
10	00	40	00	30	00	C0	00
11	00	44	00	31	00	C4	00
12	00	48	00	32	00	C8	00
13	00	4C	00	33	00	CC	00
14	00	50	00	34	00	D0	00
15	00	54	00	35	00	D4	00
16	00	58	00	36	00	D8	00
17	00	5C	00	37	00	DC	00
18	00	60	00	38	00	E0	00
19	00	64	00	39	00	E4	00
1A	00	68	00	3A	00	E8	00
1B	00	6C	00	3B	00	EC	00
1C	00	70	00	3C	00	F0	00
1D	00	74	00	3D	00	F4	00
1E	00	78	00	3E	00	F8	00
1F	00	7C	00	3F	00	FC	00
				40	00	00	00
				...	00	00	00
				FF	00	00	00

 Unused entries

16 bpp gray shade

The Look-Up Table is bypassed at this color depth, therefore programming the LUT is not required.

As with 8 bpp there are limitations to the colors which can be displayed. In this mode the six bits of green are used to set the absolute intensity of the image. This results in 64 gray shades.

5.2.2 Color Modes

In color display modes, the number of LUT entries used is determined by the color depth. For each color depth a table of sample LUT values is provided. These LUT values are a standardized set of colors used by the Epson S1D13A04 utility programs.

Note

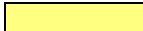
These LUT values carry eight bits of significance. The S1D13A04 LUT uses only the six MSB. The 2 LSB are ignored.

1 bpp color

When the S1D13A04 is configured for 1 bpp color mode the first 2 entries in the LUT are used. The remaining indices of the LUT are unused.

Table 5-6: Suggested LUT Values for 1 bpp Color

Index	Red	Green	Blue
00	00	00	00
01	FF	FF	FF
02	00	00	00
...	00	00	00
FF	00	00	00

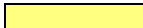
 = Indicates unused entries in the LUT

2 bpp color

When the S1D13A04 is configured for 2 bpp color mode the first 4 entries in the LUT are used. The remaining indices of the LUT are unused.

Table 5-7: Suggested LUT Values for 2 bpp Color

Index	Red	Green	Blue
00	00	00	00
01	00	00	FF
02	FF	00	00
03	FF	FF	FF
04	00	00	00
...	00	00	00
FF	00	00	00

 = Indicates unused entries in the LUT

4 bpp color

When the S1D13A04 is configured for 4 bpp color mode the first 16 entries in the LUT are used. The remaining indices of the LUT are unused.

The following table shows LUT values that simulate those of a VGA operating in 16 color mode.

Table 5-8: Suggested LUT Values for 4 bpp Color

Index	Red	Green	Blue
00	00	00	00
01	00	00	AA
02	00	AA	00
03	00	AA	AA
04	AA	00	00
05	AA	00	AA
06	AA	AA	00
07	AA	AA	AA
08	00	00	00
09	00	00	FF
0A	00	FF	00
0B	00	FF	FF
0C	FF	00	00
0D	FF	00	FF
0E	FF	FF	00
0F	FF	FF	FF
10	00	00	00
...	00	00	00
FF	00	00	00

= Indicates unused entries in the LUT

8 bpp color

When the S1D13A04 is configured for 8 bpp color mode all 256 entries in the LUT are used.

The S1D13A04 LUT has six bits (64 intensities) of intensity control per primary color which is the same as a standard VGA RAMDAC.

The following table shows LUT values that simulate the VGA default color palette.

Table 5-9: Suggested LUT Values 8 bpp Color

Index	R	G	B	Index	R	G	B	Index	R	G	B	Index	R	G	B
00	00	00	00	40	00	00	00	80	FF	FF	00	C0	00	00	00
01	00	00	AA	41	00	00	11	81	FF	EF	00	C1	00	11	11
02	00	AA	00	42	00	00	22	82	FF	DE	00	C2	00	22	22
03	00	AA	AA	43	00	00	33	83	FF	CD	00	C3	00	33	33
04	AA	00	00	44	00	00	44	84	FF	BC	00	C4	00	44	44
05	AA	00	AA	45	00	00	55	85	FF	AB	00	C5	00	55	55
06	AA	AA	00	46	00	00	66	86	FF	9A	00	C6	00	66	66
07	AA	AA	AA	47	00	00	77	87	FF	89	00	C7	00	77	77
08	55	55	55	48	00	00	89	88	FF	77	00	C8	00	89	89
09	00	00	FF	49	00	00	9A	89	FF	66	00	C9	00	9A	9A
0A	00	FF	00	4A	00	00	AB	8A	FF	55	00	CA	00	AB	AB
0B	00	FF	FF	4B	00	00	BC	8B	FF	44	00	CB	00	BC	BC
0C	FF	00	00	4C	00	00	CD	8C	FF	33	00	CC	00	CD	CD
0D	FF	00	FF	4D	00	00	DE	8D	FF	22	00	CD	00	DE	DE
0E	FF	FF	00	4E	00	00	EF	8E	FF	11	00	CE	00	EF	EF
0F	FF	FF	FF	4F	00	00	FF	8F	FF	00	00	CF	00	FF	FF
10	00	00	00	50	00	00	FF	90	FF	00	00	D0	FF	00	00
11	11	11	11	51	00	11	FF	91	FF	00	11	D1	FF	11	11
12	22	22	22	52	00	22	FF	92	FF	00	22	D2	FF	22	22
13	33	33	33	53	00	33	FF	93	FF	00	33	D3	FF	33	33
14	44	44	44	54	00	44	FF	94	FF	00	44	D4	FF	44	44
15	55	55	55	55	00	55	FF	95	FF	00	55	D5	FF	55	55
16	66	66	66	56	00	66	FF	96	FF	00	66	D6	FF	66	66
17	77	77	77	57	00	77	FF	97	FF	00	77	D7	FF	77	77
18	89	89	89	58	00	89	FF	98	FF	00	89	D8	FF	89	89
19	9A	9A	9A	59	00	9A	FF	99	FF	00	9A	D9	FF	9A	9A
1A	AB	AB	AB	5A	00	AB	FF	9A	FF	00	AB	DA	FF	AB	AB
1B	BC	BC	BC	5B	00	BC	FF	9B	FF	00	BC	DB	FF	BC	BC
1C	CD	CD	CD	5C	00	CD	FF	9C	FF	00	CD	DC	FF	CD	CD
1D	DE	DE	DE	5D	00	DE	FF	9D	FF	00	DE	DD	FF	DE	DE
1E	EF	EF	EF	5E	00	EF	FF	9E	FF	00	EF	DE	FF	EF	EF
1F	FF	FF	FF	5F	00	FF	FF	9F	FF	00	FF	DF	FF	FF	FF
20	00	00	00	60	00	FF	FF	A0	FF	00	FF	E0	00	FF	00
21	11	00	00	61	00	FF	EF	A1	EF	00	FF	E1	11	FF	11
22	22	00	00	62	00	FF	DE	A2	DE	00	FF	E2	22	FF	22
23	33	00	00	63	00	FF	CD	A3	CD	00	FF	E3	33	FF	33

Table 5-9: Suggested LUT Values 8 bpp Color (Continued)

Index	R	G	B	Index	R	G	B	Index	R	G	B	Index	R	G	B
24	44	00	00	64	00	FF	BC	A4	BC	00	FF	E4	44	FF	44
25	55	00	00	65	00	FF	AB	A5	AB	00	FF	E5	55	FF	55
26	66	00	00	66	00	FF	9A	A6	9A	00	FF	E6	66	FF	66
27	77	00	00	67	00	FF	89	A7	89	00	FF	E7	77	FF	77
28	89	00	00	68	00	FF	77	A8	77	00	FF	E8	89	FF	89
29	9A	00	00	69	00	FF	66	A9	66	00	FF	E9	9A	FF	9A
2A	AB	00	00	6A	00	FF	55	AA	55	00	FF	EA	AB	FF	AB
2B	BC	00	00	6B	00	FF	44	AB	44	00	FF	EB	BC	FF	BC
2C	CD	00	00	6C	00	FF	33	AC	33	00	FF	EC	CD	FF	CD
2D	DE	00	00	6D	00	FF	22	AD	22	00	FF	ED	DE	FF	DE
2E	EF	00	00	6E	00	FF	11	AE	11	00	FF	EE	EF	FF	EF
2F	FF	00	00	6F	00	FF	00	AF	00	00	FF	EF	FF	FF	FF
30	00	00	00	70	00	FF	00	B0	00	00	00	F0	00	00	FF
31	00	11	00	71	11	FF	00	B1	11	00	11	F1	11	11	FF
32	00	22	00	72	22	FF	00	B2	22	00	22	F2	22	22	FF
33	00	33	00	73	33	FF	00	B3	33	00	33	F3	33	33	FF
34	00	44	00	74	44	FF	00	B4	44	00	44	F4	44	44	FF
35	00	55	00	75	55	FF	00	B5	55	00	55	F5	55	55	FF
36	00	66	00	76	66	FF	00	B6	66	00	66	F6	66	66	FF
37	00	77	00	77	77	FF	00	B7	77	00	77	F7	77	77	FF
38	00	89	00	78	89	FF	00	B8	89	00	89	F8	89	89	FF
39	00	9A	00	79	9A	FF	00	B9	9A	00	9A	F9	9A	9A	FF
3A	00	AB	00	7A	AB	FF	00	BA	AB	00	AB	FA	AB	AB	FF
3B	00	BC	00	7B	BC	FF	00	BB	BC	00	BC	FB	BC	BC	FF
3C	00	CD	00	7C	CD	FF	00	BC	CD	00	CD	FC	CD	CD	FF
3D	00	DE	00	7D	DE	FF	00	BD	DE	00	DE	FD	DE	DE	FF
3E	00	EF	00	7E	EF	FF	00	BE	EF	00	EF	FE	EF	EF	FF
3F	00	FF	00	7F	FF	FF	00	BF	FF	00	FF	FF	FF	FF	FF

16 bpp color

The Look-Up Table is bypassed at this color depth, therefore programming the LUT is not required.

6 Power Save Mode

The S1D13A04 is designed for very low-power applications. During normal operation, the internal clocks are dynamically disabled when not required. The S1D13A04 design also includes a Power Save Mode to further save power. When Power Save Mode is initiated, LCD power sequencing is required to ensure the LCD bias power supply is disabled properly. For further information on LCD power sequencing, see Section 6.3, “LCD Power Sequencing” on page 28.

For Power Save Mode AC Timing, see the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

6.1 Overview

The S1D13A04 includes a software initiated Power Save Mode. Enabling/disabling Power Save Mode is controlled using the Power Save Mode Enable bit (REG[14h] bit 4).

While Power Save Mode is enabled the following conditions apply.

- Registers are accessible (USB registers are not accessible)
- Memory writes are possible¹
- Memory reads are not possible
- LCD display is inactive.
- LCD interface outputs are forced low.

Note

¹ Memory writes are possible during power save mode because the S1D13A04 dynamically enables the memory controller for display buffer writes.

6.2 Registers

6.2.1 Power Save Mode Enable

Power Save Configuration Register															
REG[14h] Default = 00000010h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a								VNDP Status (RO)	Memory Power Save Status (RO)	n/a	Power Save Mode Enable	n/a			'Direct' HR-TFT GPO Control
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

The Power Save Mode Enable bit initiates Power Save Mode when set to 1. Setting the bit to 0 disables Power Save Mode and returns the S1D13A04 to normal mode. At reset this bit is set to 1.

Note

Enabling/disabling Power Save Mode requires proper LCD Power Sequencing. See Section 6.3, “LCD Power Sequencing” on page 28.

6.2.2 Memory Controller Power Save Status

Power Save Configuration Register															
REG[14h] Default = 00000010h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a								VNDP Status (RO)	Memory Power Save Status (RO)	n/a	Power Save Mode Enable	n/a			'Direct' HR-TFT GPO Control
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

The Memory Controller Power Save Status bit is a read-only status bit which indicates the power save state of the S1D13A04 SRAM interface. When this bit returns a 1, the SRAM interface is powered down and the memory clock source may be disabled. When this bit returns a 0, the SRAM interface is active. This bit returns a 0 after a chip reset.

Note

Memory writes are possible during power save mode because the S1D13A04 dynamically enables the memory controller for display buffer writes.

6.3 LCD Power Sequencing

The S1D13A04 requires LCD power sequencing (the process of powering-on and powering-off the LCD panel). LCD power sequencing allows the LCD bias voltage to discharge prior to shutting down the LCD signals, preventing long term damage to the panel and avoiding unsightly “lines” at power-on/power-off.

Proper LCD power sequencing for power-off requires a delay from the time the LCD power is disabled to the time the LCD signals are shut down. Power-on requires the LCD signals to be active prior to applying power to the LCD. This time interval depends on the LCD bias power supply design. For example, the LCD bias power supply on the S5U13A04B00C Evaluation Board requires 0.5 seconds to fully discharge. Other power supply designs may vary.

This section assumes the LCD bias power is controlled through GPIO0. The S1D13A04 GPIO pins are multi-use pins and may not be available in all system designs. For further information on the availability of GPIO pins, see the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

Note

This section discusses LCD power sequencing for passive and TFT (non-HR-TFT) panels only. For further information on LCD power sequencing the HR-TFT, see *Connecting to the Sharp HR-TFT Panels*, document number X37A-G-011-xx.

6.4 Enabling Power Save Mode

Power Save Mode must be enabled using the following steps.

1. Turn off the LCD bias power.

Note

The S5U13A04B00C uses GPIO0 to control the LCD bias power supplies. Your system design may vary.

2. Wait for the LCD bias power supply to discharge. The discharge time is based on the discharge rate of the power supply.
3. Enable Power Save Mode - set REG[14h] bit 4 to 1.

The S1D13A04 is now in Power Save Mode. To further increase power savings PCLK and MCLK can be switched off (see steps 4 and 5).

4. At this time, the LCD pixel clock source may be disabled.
5. After the Memory Controller Power Save Status bit (REG[14h] bit 6) returns a 1, the Memory Clock source may be shut down.

6.5 Disabling Power Save Mode

Bring the S1D13A04 out of Power Save Mode using the following steps.

1. If the Memory Clock source is shut down, it must be started.
2. If the pixel clock is disabled, it must be started.
3. Disable Power Save Mode - set REG[14h] bit 4 to 0.
4. Wait for the LCD bias power supply to charge. The charge is based on the time required for the LCD power supply to reach operating voltage.
5. Enable the LCD bias power.

Note

The S5U13A04B00C uses GPIO0 to control the LCD bias power supplies. Your system design may vary.

7 SwivelView™

Most computer displays operate in landscape mode. In landscape mode the display is typically wider than it is high. For example, a display size of 320x240 is 320 pixels wide and 240 lines high.

SwivelView rotates the display image counter-clockwise in ninety degree increments. Rotating the image on a 320x240 display by 90 or 270 degrees yields a display that is now 240 pixels wide and 320 lines high.

The S1D13A04 provides hardware support for SwivelView in all color depths (1, 2, 4, 8 and 16 bpp).

For further details on the SwivelView feature, see the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

7.1 SwivelView Registers

These are the registers which control the SwivelView feature.

Display Settings Register											Read/Write				
REG[10h]											Default = 00000000h				
n/a						Pixel Doubling Vertical	Pixel Doubling Horiz.	Display Blank	Dithering Disable	n/a	SW Video Invert	PIP+ Window Enable	n/a	SwivelView Mode Select	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a											Bits-per-pixel Select (actual value: 1, 2, 4, 8 or 16 bpp)				
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

SwivelView Mode Select

The SwivelView modes are selected using the SwivelView Mode Select Bits[1:0] (bits 17-16). The combinations of these bits provide the following rotations.

Table 7-1: SwivelView Mode Select Bits

SwivelView Mode Select Bit 1	SwivelView Mode Select Bit 0	SwivelView Orientation
0	0	0° (normal)
0	1	90°
1	0	180°
1	1	270°

Main Window Display Start Address Register															Read/Write
REG[40h]															Default = 00000000h
n/a															bit 16
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
Main Window Display Start Address bits 15-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

Main Window Display Start Address

The Main Window Display Start Address register represents a DWORD address which points to the start of the main window image in the display buffer. An address of 0 is the start of the display buffer. For the following SwivelView mode descriptions, the *desired byte address* is the starting display address for the main window image.

In SwivelView 0°, program the start address
= desired byte address ÷ 4

In SwivelView 90°, program the start address
= ((desired byte address + (panel height × bpp ÷ 8)
+ ((4 - (panel height × bpp ÷ 8)) & 03h)) ÷ 4) - 1

In SwivelView 180°, program the start address
= ((desired byte address + (Main Window Stride × (panel height - 1))
+ (panel width × bpp ÷ 8) + ((4 - (panel width × bpp ÷ 8)) & 03h)) ÷ 4) - 1

In SwivelView 270°, program the start address
= (desired byte address + ((panel width - 1) × Main Window Stride)) ÷ 4

Note

Truncate all fractional values before writing to the address registers.

Note

SwivelView 0° and 180° require the panel width to be a multiple of 32 ÷ bits-per-pixel. SwivelView 90° and 270° require the panel height to be a multiple of 32 ÷ bits-per-pixel. If this is not possible, refer to Section 7.3, “Limitations”.

Main Window Line Address Offset Register															
REG[44h] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						Main Window Line Address Offset bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

Main Window Line Address Offset

The Main Window Line Address Offset register indicates the number of dwords per line in the main window image.

For SwivelView 0° and 180°, the image width must be at least the panel width. For SwivelView 90° and 270°, the image width must be at least the panel height. In addition, the image width must be a multiple of 32 ÷ bpp. If the image width is not such a multiple, a slightly larger width must be chosen (see Section 7.3, “Limitations”).

Panel width and *panel height* refer to the physical panel dimensions in pixels. *Stride* is the number of bytes required for one line of the image; the offset register represents the stride in DWORD steps.

Main Window Stride = image width × bpp ÷ 8

Note

Image width can be larger than panel width (or panel height, for SwivelView 90° or 270°).

number of dwords per line = image width ÷ (32 ÷ bpp)

7.2 Examples

Example 1: In SwivelView 0° (normal) mode, program the main window registers for a 320x240 panel at a color depth of 4 bpp.

- Determine the main window display start address.
The main window is typically placed at the start of display memory which is at display address 0.

main window display start address register
= desired byte address ÷ 4
= 0

Program the Main Window Display Start Address register. REG[40h] is set to 00000000h.

2. Determine the main window line address offset.

$$\begin{aligned} \text{number of dwords per line} &= \text{image width} \div (32 \div \text{bpp}) \\ &= 320 \div (32 \div 4) \\ &= 40 \\ &= 28\text{h} \end{aligned}$$

Program the Main Window Line Address Offset register. REG[44h] is set to 00000028h.

Example 2: In SwivelView 90° mode, program the main window registers for a 320x240 panel at a color depth of 4 bpp.

1. Determine the main window display start address.
The main window is typically placed at the start of display memory, which is at display address 0.

$$\begin{aligned} \text{main window display start address register} &= ((\text{desired byte address} + (\text{panel height} \times \text{bpp} \div 8) \\ &\quad + ((4 - (\text{panel height} \times \text{bpp} \div 8)) \& 03\text{h})) \div 4) - 1 \\ &= ((0 + (240 \times 4 \div 8) + ((4 - (240 \times 4 \div 8)) \& 03\text{h})) \div 4) - 1 \\ &= 29 \\ &= 1\text{Dh} \end{aligned}$$

Program the Main Window Display Start Address register. REG[40h] is set to 0000001Dh.

2. Determine the main window line address offset.

$$\begin{aligned} \text{number of dwords per line} &= \text{image width} \div (32 \div \text{bpp}) \\ &= 240 \div (32 \div 4) \\ &= 30 \\ &= 1\text{Eh} \end{aligned}$$

Program the Main Window Line Address Offset register. REG[44h] is set to 0000001Eh.

Example 3: In SwivelView 180° mode, program the main window registers for a 320x240 panel at a color depth of 4 bpp.

1. Determine the main window display start address.
The main window is typically placed at the start of display memory which is at display address 0.

Main Window Stride

$$\begin{aligned} &= \text{image width} \times \text{bpp} \div 8 \\ &= 320 \times 4 \div 8 \\ &= 160 \\ &= \text{A0h} \end{aligned}$$

main window display start address register

$$\begin{aligned} &= ((\text{desired byte address} + (\text{Main Window Stride} \times (\text{panel height} - 1)) \\ &\quad + (\text{panel width} \times \text{bpp} \div 8) + ((4 - (\text{panel width} \times \text{bpp} \div 8)) \& 03\text{h})) \div 4) - 1 \\ &= ((0 + (160 \times (240 - 1)) + (320 \times 4 \div 8) + ((4 - (320 \times 4 \div 8)) \& 03\text{h})) \div 4) - 1 \\ &= 9599 \\ &= 257\text{Fh} \end{aligned}$$

Program the Main Window Display Start Address register. REG[40h] is set to 0000257Fh.

2. Determine the main window line address offset.

number of dwords per line

$$\begin{aligned} &= \text{image width} \div (32 \div \text{bpp}) \\ &= 320 \div (32 \div 4) \\ &= 40 \\ &= 28\text{h} \end{aligned}$$

Program the Main Window Line Address Offset register. REG[44h] is set to 00000028h.

Example 4: In SwivelView 270° mode, program the main window registers for a 320x240 panel at a color depth of 4 bpp.

1. Determine the main window display start address.
The main window is typically placed at the start of display memory, which is at display address 0.

Main Window Stride

$$\begin{aligned} &= \text{image width} \times \text{bpp} \div 8 \\ &= 240 \times 4 \div 8 \\ &= 120 \\ &= 78\text{h} \end{aligned}$$

main window display start address register

$$\begin{aligned} &= (\text{desired byte address} + ((\text{panel width} - 1) \times \text{Main Window Stride})) \div 4 \\ &= (0 + ((320 - 1) \times 120)) \div 4 \\ &= 9570 \\ &= 2562\text{h} \end{aligned}$$

Program the Main Window Display Start Address register. REG[40h] is set to 00002562h.

2. Determine the main window line address offset.

number of dwords per line

$$\begin{aligned} &= \text{image width} \div (32 \div \text{bpp}) \\ &= 240 \div (32 \div 4) \\ &= 30 \\ &= 1\text{Eh} \end{aligned}$$

Program the Main Window Line Address Offset register. REG[44h] is set to 0000001Eh.

7.3 Limitations

7.3.1 SwivelView 0° and 180°

In SwivelView 0° and 180°, the Main Window Line Address Offset register (REG[44h]) requires the *panel width* to be a multiple of $32 \div \text{bits-per-pixel}$. If this is not the case, then the Main Window Line Address Offset register must be programmed to a longer line which meets this requirement. This longer line creates a virtual image where the width is *main window line address offset register* $\times 32 \div \text{bits-per-pixel}$. In SwivelView 0°, this virtual image should be drawn in display memory as left justified, and in SwivelView 180°, this virtual image should be drawn in display memory as right justified. A left-justified image is one drawn in display memory such that each of the image's lines only use the left most portion of the line width defined by the line address offset register (i.e. starting at horizontal position 0). A right-justified image is one drawn in display memory such that each of the image's lines only use the right most portion of the line width defined by the line address offset register (i.e. starting at a non-zero horizontal position which is the virtual width - image width).

7.3.2 SwivelView 90° and 270°

In SwivelView 90° and 270°, the Main Window Line Address Offset register (REG[44h]) requires the *panel height* to be a multiple of $32 \div \text{bits-per-pixel}$. If this is not the case, then the Main Window Line Address Offset register must be programmed to a longer line which meets this requirement. This longer line creates a virtual image whose width is *main window line address offset register* $\times 32 \div \text{bits-per-pixel}$. In SwivelView 270°, this virtual image should be drawn in display memory as left justified, and in SwivelView 90°, this virtual image should be drawn in display memory as right justified. A left-justified image is one drawn in display memory such that each of the image's lines only use the left most portion of the line width defined by the line address offset register (i.e. starting at horizontal position 0). A right-justified image is one drawn in display memory such that each of the image's lines only use the right most portion of the line width defined by the line address offset register (i.e. starting at a non-zero horizontal position which is the virtual width - image width).

8 Picture-in-Picture Plus

Picture-in-Picture Plus (PIP⁺) enables a secondary window (or PIP⁺ window) within the main display window. The PIP⁺ window may be positioned anywhere within the virtual display and is controlled through the PIP⁺ Window control registers (REG[50h] through REG[5Ch]). The PIP⁺ window retains the same color depth and SwivelView orientation as the main window.

A PIP⁺ window can be used to display temporary items such as a dialog box or to “float” the display item so that the system doesn’t have to exclude the area during screen repaints.

The following diagram shows an example of a PIP⁺ window within a main window.

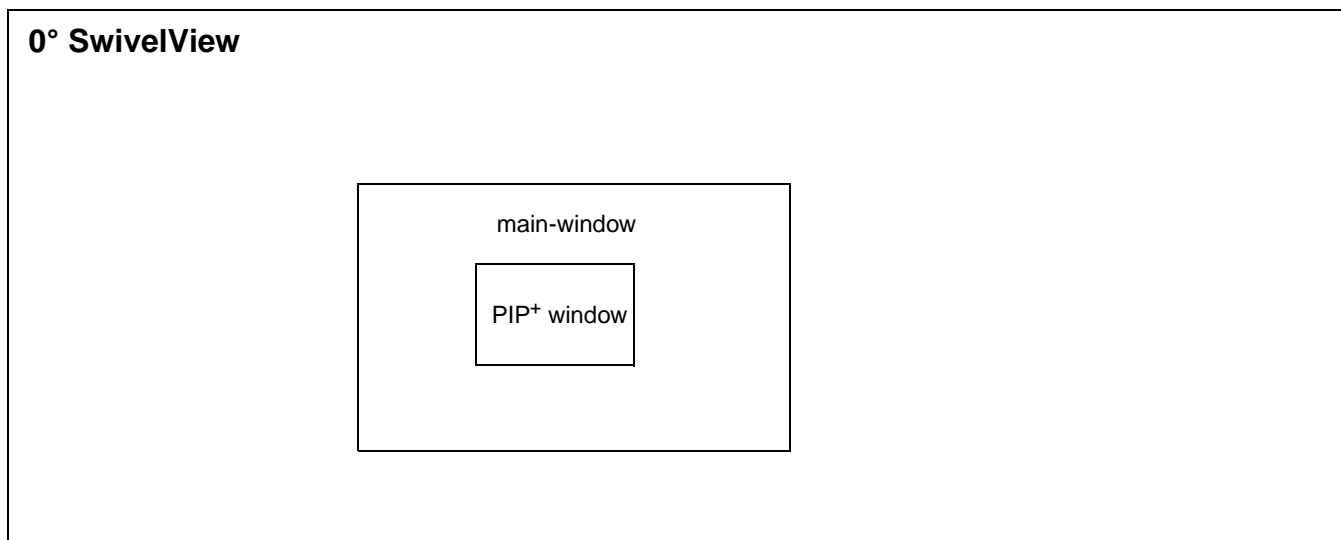


Figure 8-1: Picture-in-Picture Plus with SwivelView disabled

8.1 Registers

The following registers control the Picture-In-Picture Plus feature.

Display Settings Register														Read/Write	
REG[10h]														Default = 00000000h	
n/a						Pixel Doubling Vertical	Pixel Doubling Horiz.	Display Blank	Dithering Disable	n/a	SW Video Invert	PIP ⁺ Window Enable	n/a	SwivelView Mode Select	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a											Bits-per-pixel Select (actual value: 1, 2, 4, 8 or 16 bpp)				
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

PIP⁺ Window Enable

The PIP⁺ Window Enable bit enables a PIP⁺ window within the main window. The location of the PIP⁺ window within the landscape window is determined by the PIP⁺ X Position register (REG[58h]) and PIP⁺ Y Position register (REG[5Ch]). The PIP⁺ window has its own Display Start Address register (REG[50h]) and Line Address Offset register (REG[54h]). The PIP⁺ window shares the same color depth and SwivelView™ orientation as the main window.

PIP ⁺ Display Start Address Register															Read/Write
REG[50h]															Default = 00000000h
n/a														bit 16	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
PIP ⁺ Display Start Address bits 15-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

PIP⁺ Display Start Address

The PIP⁺ Display Start Address register is a DWORD which represents an address that points to the start of the PIP⁺ window image in the display buffer. An address of 0 is the start of the display buffer. For the following PIP⁺ descriptions, the *desired byte address* is the starting display address for the PIP⁺ window image.

In SwivelView 0°, program the start address
= desired byte address ÷ 4

In SwivelView 90°, program the start address
= ((desired byte address + (PIP⁺ width × bpp ÷ 8)
+ ((4 - (PIP⁺ width × bpp ÷ 8)) & 03h)) ÷ 4) - 1

In SwivelView 180°, program the start address
= ((desired byte address + (PIP⁺ Stride × (PIP⁺ height - 1))
+ (PIP⁺ width × bpp ÷ 8) + ((4 - (PIP⁺ width × bpp ÷ 8)) & 03h)) ÷ 4) - 1

In SwivelView 270°, program the start address
= (desired byte address + ((PIP⁺ height - 1) × PIP⁺ Stride)) ÷ 4

Note

Truncate all fractional values before writing to the address registers.

Note

SwivelView 0° and 180° require the PIP⁺ width to be a multiple of 32 ÷ bits-per-pixel. SwivelView 90° and 270° require the PIP⁺ height to be a multiple of 32 ÷ bits-per-pixel. If this is not possible, refer to Section 8.3, “Limitations” .

PIP ⁺ Line Address Offset Register																	
REG[54h]														Default = 00000000h		Read/Write	
n/a																	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
n/a						PIP ⁺ Line Address Offset bits 9-0											
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		

PIP⁺ Line Address Offset

The PIP⁺ Line Address Offset register indicates the number of dwords per line in the PIP⁺ window image.

The image width must be a multiple of 32 ÷ bpp. If the image width is not such a multiple, a slightly larger width must be chosen (see Section 8.3, “Limitations”).

PIP⁺ width and *PIP⁺ height* refer to the PIP⁺ dimensions as seen in SwivelView 0° (landscape mode). *Stride* is the number of bytes required for one line of the image; the offset register represents the stride in DWORD steps.

$$\text{PIP}^+ \text{ Stride} = \text{image width} \times \text{bpp} \div 8$$

For SwivelView 0° and 180°,

$$\text{PIP}^+ \text{ Width} = ((\text{REG}[58\text{h}] \text{ bits } 25:16) - (\text{REG}[58\text{h}] \text{ bits } 9:0) + 1) \times (32 \div \text{bpp})$$

$$\text{PIP}^+ \text{ Height} = (\text{REG}[5\text{Ch}] \text{ bits } 25:16) - (\text{REG}[5\text{Ch}] \text{ bits } 9:0) + 1$$

For SwivelView 90° and 270°,

$$\text{PIP}^+ \text{ Width} = ((\text{REG}[5\text{Ch}] \text{ bits } 25:16) - (\text{REG}[5\text{Ch}] \text{ bits } 9:0) + 1) \times (32 \div \text{bpp})$$

$$\text{PIP}^+ \text{ Height} = (\text{REG}[58\text{h}] \text{ bits } 25:16) - (\text{REG}[58\text{h}] \text{ bits } 9:0) + 1$$

Note

Image width can be larger than PIP⁺ width (or PIP⁺ height, for SwivelView 90° or 270°).

$$\text{number of dwords per line} = \text{image width} \div (32 \div \text{bpp})$$

PIP ⁺ X Positions Register															
REG[58h] Default = 00000000h															
Read/Write															
n/a						PIP ⁺ X End Position bits 9-0									
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						PIP ⁺ X Start Position bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

PIP⁺ X End Position

The PIP⁺ X End Position bits determine the horizontal end of the PIP⁺ window in 0° and 180° SwivelView orientations. These bits determine the vertical end position in 90° and 270° SwivelView. For further information on defining the value of the X End Position, see Section 8.2, “Picture-In-Picture-Plus Examples” on page 45.

This register also increments differently based on the SwivelView orientation. For 0° and 180° SwivelView the X End Position is incremented by *X* pixels where *X* is relative to the current color depth. For 90° and 270° SwivelView the X End Position is incremented in 1 line increments.

Table 8-1: 32-bit Address Increments for PIP⁺ X Position in SwivelView 0° and 180°

Bits-Per-Pixel (Color Depth)	Pixel Increment (X)
1 bpp	32
2 bpp	16
4 bpp	8
8 bpp	4
16 bpp	2

In SwivelView 0°, these bits set the horizontal coordinates (x) of the PIP⁺ window’s right edge. Increasing x moves the right edge towards the right in steps of $32 \div \text{bits-per-pixel}$ (see Table 8-1:). The horizontal coordinates start at pixel 0.

Program the PIP⁺ Window X End Position so that

$$\text{PIP}^+ \text{ Window X End Position} = x \div (32 \div \text{bits-per-pixel})$$

Note

Truncate the fractional part of the above equation.

In SwivelView 90°, these bits set the vertical coordinates (y) of the PIP⁺ window’s bottom edge. Increasing y moves the bottom edge downward in 1 line steps. The vertical coordinates start at line 0.

Program the PIP⁺ Window X End Position so that

$$\text{PIP}^+ \text{ Window X End Position} = y$$

In SwivelView 180°, these bits set the horizontal coordinates (x) of the PIP⁺ window's left edge. Increasing x moves the left edge towards the right in steps of $32 \div \text{bits-per-pixel}$ (see Table 8-1:). The horizontal coordinates start at pixel 0.

Program the PIP⁺ Window X End Position so that

$$\text{PIP}^+ \text{ Window X End Position} = (\text{panel width} - x - 1) \div (32 \div \text{bits-per-pixel})$$

Note

Truncate the fractional part of the above equation.

In SwivelView 270°, these bits set the vertical coordinates (y) of the PIP⁺ window's top edge. Increasing y moves the top edge downwards in 1 line steps. The vertical coordinates start at line 0.

Program the PIP⁺ Window X End Position so that

$$\text{PIP}^+ \text{ Window X End Position} = \text{panel width} - y - 1$$

PIP⁺ X Start Position

The PIP⁺ X Start Position bits determine the horizontal position of the start of the PIP⁺ window in 0° and 180° SwivelView orientations. These bits determine the vertical start position in 90° and 270° SwivelView. For further information on defining the value of the X Start Position, see Section 8.2, "Picture-In-Picture-Plus Examples" on page 45.

The register also increments differently based on the SwivelView orientation. For 0° and 180° SwivelView the X Start Position is incremented by X pixels where X is relative to the current color depth. For 90° and 270° SwivelView the X Start Position is incremented in 1 line increments.

Table 8-2: 32-bit Address Increments for Color Depth

Bits-per-pixel (Color Depth)	Pixel Increment (X)
1 bpp	32
2 bpp	16
4 bpp	8
8 bpp	4
16 bpp	2

In SwivelView 0°, these bits set the horizontal coordinates (x) of the PIP⁺ windows's left edge. Increasing x moves the left edge towards the right in steps of $(32 \div \text{bits-per-pixel})$ (see Table 8-2:). The horizontal coordinates start at pixel 0.

Program the PIP⁺ Window X Start Position so that

$$\text{PIP}^+ \text{ Window X Start Position} = x \div (32 \div \text{bits-per-pixel})$$

Note

Truncate the fractional part of the above equation.

In SwivelView 90°, these bits set the vertical coordinates (y) of the PIP⁺ window's top edge. Increasing y moves the top edge downward in 1 line steps. The vertical coordinates start at line 0.

Program the PIP⁺ Window X Start Position so that

$$\text{PIP}^+ \text{ Window X Start Position} = y$$

In SwivelView 180°, these bits set the horizontal coordinates (x) of the PIP⁺ window's right edge. Increasing x moves the right edge towards the right in steps of (32 ÷ bits-per-pixel) (see Table 8-2:). The horizontal coordinates start at pixel 0.

Program the PIP⁺ Window X Start Position so that

$$\text{PIP}^+ \text{ Window X Start Position} = (\text{panel width} - x - 1) \div (32 \div \text{bits-per-pixel})$$

Note

Truncate the fractional part of the above equation.

In SwivelView 270°, these bits set the vertical coordinates (y) of the PIP⁺ window's bottom edge. Increasing y moves the bottom edge downwards in 1 line steps. The vertical coordinates start at line 0.

Program the PIP⁺ Window X Start Position so that

$$\text{PIP}^+ \text{ Window X Start Position} = \text{panel width} - y - 1$$

PIP ⁺ Y Positions Register															
REG[5Ch] Default = 00000000h Read/Write															
n/a						PIP ⁺ Y End Position bits 9-0									
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						PIP ⁺ Y Start Position bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

PIP⁺ Y End Position

The PIP⁺ Y End Position bits determine the vertical end position of the PIP⁺ window in 0° and 180° SwivelView orientations. These bits determine the horizontal end position in 90° and 270° SwivelView. For further information on defining the value of the Y End Position, see Section 8.2, "Picture-In-Picture-Plus Examples" on page 45.

The register also increments differently based on the SwivelView orientation. For 0° and 180° SwivelView the Y End Position is incremented in 1 line increments. For 90° and 270° SwivelView the Y End Position is incremented by *Y* pixels where *Y* is relative to the current color depth.

Table 8-3: 32-bit Address Increments for Color Depth

Bits-Per-Pixel (Color Depth)	Pixel Increment (Y)
1 bpp	32
2 bpp	16
4 bpp	8
8 bpp	4
16 bpp	2

In SwivelView 0°, these bits set the vertical coordinates (y) of the PIP⁺ windows's bottom edge. Increasing y moves the bottom edge downwards in 1 line steps. The vertical coordinates start at line 0.

Program the PIP⁺ Window Y End Position so that
$$\text{PIP}^+ \text{ Window Y End Position} = y$$

In SwivelView 90°, these bits set the horizontal coordinates (x) of the PIP⁺ window's left edge. Increasing x moves the left edge towards the right in steps of (32 ÷ bits-per-pixel) (see Table 8-3:). The horizontal coordinates start at pixel 0.

Program the PIP⁺ Window Y End Position so that
$$\text{PIP}^+ \text{ Window Y End Position} = (\text{panel height} - x - 1) \div (32 \div \text{bits-per-pixel})$$

Note

Truncate the fractional part of the above equation.

In SwivelView 180°, these bits set the vertical coordinates (y) of the PIP⁺ window's top edge. Increasing y moves the top edge downwards in 1 line steps. The vertical coordinates start at line 0.

Program the PIP⁺ Window Y End Position so that
$$\text{PIP}^+ \text{ Window Y End Position} = \text{panel height} - y - 1$$

In SwivelView 270°, these bits set the horizontal coordinates (x) of the PIP⁺ window's right edge. Increasing x moves the right edge towards the right in steps of (32 ÷ bits-per-pixel) (see Table 8-3:). The horizontal coordinates start at pixel 0.

Program the PIP⁺ Window Y End Position so that
$$\text{PIP}^+ \text{ Window Y End Position} = x \div (32 \div \text{bits-per-pixel})$$

Note

Truncate the fractional part of the above equation.

PIP⁺ Y Start Position

The PIP⁺ Y Start Position bits determine the vertical start position of the PIP⁺ window in 0° and 180° SwivelView orientations. These bits determine the horizontal start position in 90° and 270° SwivelView. For further information on defining the value of the Y Start Position, see Section 8.2, “Picture-In-Picture-Plus Examples” on page 45.

The register also increments differently based on the SwivelView orientation. For 0° and 180° SwivelView the Y Start Position is incremented in 1 line increments. For 90° and 270° SwivelView the Y Start Position is incremented by *Y* pixels where *Y* is relative to the current color depth.

Table 8-4: 32-bit Address Increments for Color Depth

Bits-Per-Pixel (Color Depth)	Pixel Increment (Y)
1 bpp	32
2 bpp	16
4 bpp	8
8 bpp	4
16 bpp	2

In SwivelView 0°, these bits set the vertical coordinates (*y*) of the PIP⁺ windows’s top edge. Increasing *y* moves the top edge downwards in 1 line steps. The vertical coordinates start at line 0.

Program the PIP⁺ Window Y Start Position so that

$$\text{PIP}^+ \text{ Window Y Start Position} = y$$

In SwivelView 90°, these bits set the horizontal coordinates (*x*) of the PIP⁺ window’s right edge. Increasing *x* moves the right edge towards the right in steps of $(32 \div \text{bits-per-pixel})$ (see Table 8-4:). The horizontal coordinates start at pixel 0.

Program the PIP⁺ Window Y Start Position so that

$$\text{PIP}^+ \text{ Window Y Start Position} = (\text{panel height} - x - 1) \div (32 \div \text{bits-per-pixel})$$

Note

Truncate the fractional part of the above equation.

In SwivelView 180°, these bits set the vertical coordinates (*y*) of the PIP⁺ window’s bottom edge. Increasing *y* moves the bottom edge downwards in 1 line steps. The vertical coordinates start at line 0.

Program the PIP⁺ Window Y Start Position so that

$$\text{PIP}^+ \text{ Window Y Start Position} = \text{panel height} - y - 1$$

In **SwivelView 270°**, these bits set the horizontal coordinates (x) of the PIP⁺ window's left edge. Increasing x moves the left edge towards the right in steps of (32 ÷ bits-per-pixel) (see Table 8-4:). The horizontal coordinates start at pixel 0.

Program the PIP⁺ Window Y Start Position so that

$$\text{PIP}^+ \text{ Window Y Start Position} = x \div (32 \div \text{bits-per-pixel})$$

Note

Truncate the fractional part of the above equation.

8.2 Picture-In-Picture-Plus Examples

8.2.1 SwivelView 0° (Landscape Mode)

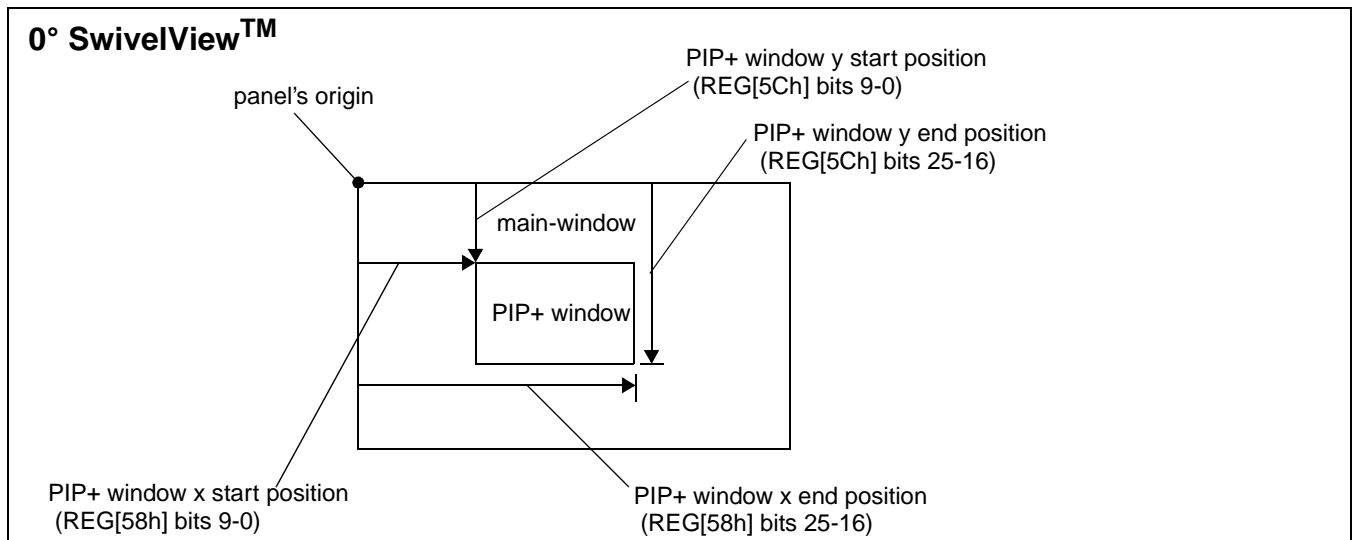


Figure 8-2: Picture-in-Picture Plus with SwivelView disabled

SwivelView 0° (or landscape) is a mode in which both the main and PIP⁺ window are non-rotated. The images for each window are typically placed consecutively, with the main window image starting at address 0 and followed by the PIP⁺ window image. In addition, both images must start at addresses which are dword-aligned (the last two bits of the starting address must be 0).

Note

It is possible to use the same image for both the main window and PIP⁺ window. To do so, set the PIP⁺ Line Address Offset register (REG[54h]) to the same value as the Main Window Line Address Offset register (REG[44h]).

Example 5: Program the PIP⁺ window registers for a 320x240 panel at 4 bpp, with the PIP⁺ window positioned at (80, 60) with a width of 160 and a height of 120.

1. Determine the value for the PIP⁺ Window X Positions and PIP⁺ Window Y Positions registers. Let the top left corner of the PIP⁺ window be (x1, y1), and let the bottom right corner be (x2, y2), where $x2 = x1 + \text{width} - 1$ and $y2 = y1 + \text{height} - 1$. The PIP⁺ Window X Positions register sets the horizontal coordinates of the PIP⁺ window's top left and bottom right corners. The PIP⁺ Window Y Positions register sets the vertical coordinates of the PIP⁺ window's top left and bottom right corners.

The required values are calculated as follows:

X Start Position

$$\begin{aligned} &= x1 \div (32 \div \text{bpp}) \\ &= 80 \div (32 \div 4) \\ &= 10 \\ &= 0Ah \end{aligned}$$

Y Start Position

$$\begin{aligned} &= y1 \\ &= 60 \\ &= 3Ch \end{aligned}$$

X End Position

$$\begin{aligned} &= x2 \div (32 \div \text{bpp}) \\ &= (80 + 160 - 1) \div (32 \div 4) \\ &= 29.875 \\ &= 1Dh \text{ (truncated fractional part)} \end{aligned}$$

Y End Position

$$\begin{aligned} &= y2 \\ &= 60 + 120 - 1 \\ &= 179 \\ &= B3h \end{aligned}$$

2. Program the PIP⁺ Window X Positions register with the X Start Position in bits 9-0 and the X End Position in bits 25-16. REG[58h] is set to 001D000Ah.
Program the PIP⁺ Window Y Positions register with the Y Start Position in bits 9-0 and the Y End Position in bits 25-16. REG[5Ch] is set to 00B3003Ch.

Due to truncation, the dimensions of the PIP⁺ window may have changed. Recalculate the PIP⁺ window width and height below:

PIP⁺ Width

$$\begin{aligned} &= ((\text{REG}[58\text{h}] \text{ bits } 25:16) - (\text{REG}[58\text{h}] \text{ bits } 9:0) + 1) \times (32 \div \text{bpp}) \\ &= (1\text{Dh} - 0\text{Ah} + 1) \times (32 \div 4) \\ &= 160 \text{ pixels} \end{aligned}$$

PIP Height

$$\begin{aligned} &= (\text{REG}[5\text{Ch}] \text{ bits } 25:16) - (\text{REG}[5\text{Ch}] \text{ bits } 9:0) + 1 \\ &= \text{B3h} - 3\text{Ch} + 1 \\ &= 120 \text{ lines} \end{aligned}$$

3. Determine the PIP⁺ display start address.
The main window image must take up $320 \times 240 \text{ pixels} \times \text{bpp} \div 8 = 9600\text{h}$ bytes. If the main window starts at address 0h, the PIP⁺ window can start at 9600h.

PIP⁺ display start address

$$\begin{aligned} &= \text{desired byte address} \div 4 \\ &= 9600\text{h} \div 4 \\ &= 2580\text{h}. \end{aligned}$$

Program the PIP⁺ Display Start Address register. REG[50h] is set to 00002580h.

4. Determine the PIP⁺ line address offset.

number of dwords per line

$$\begin{aligned} &= \text{image width} \div (32 \div \text{bpp}) \\ &= 160 \div (32 \div 4) \\ &= 20 \\ &= 14\text{h} \end{aligned}$$

Program the PIP⁺ Line Address Offset register. REG[54h] is set to 00000014h.

5. Enable the PIP⁺ window.

Program the PIP⁺ Window Enable bit. REG[10h] bit 19 is set to 1.

8.2.2 SwivelView 90°

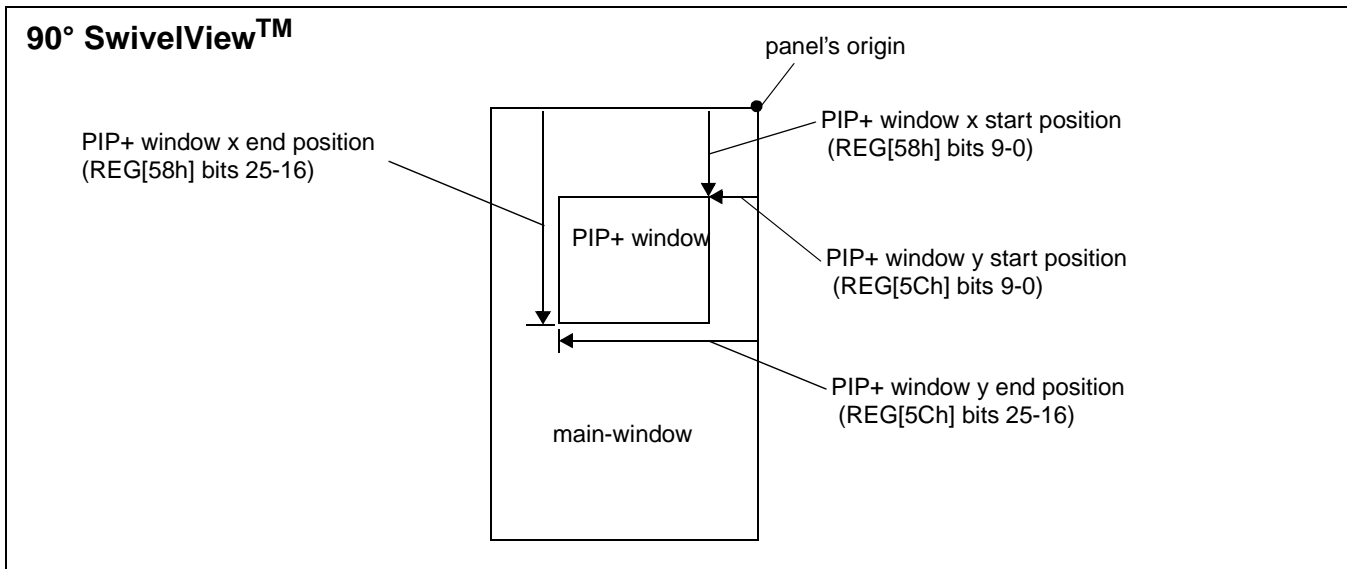


Figure 8-3: Picture-in-Picture Plus with SwivelView 90° enabled

SwivelView 90° is a mode in which both the main and PIP+ windows are rotated 90° counter-clockwise when shown on the panel. The images for each window are typically placed consecutively, with the main window image starting at address 0 and followed by the PIP+ window image. In addition, both images must start at addresses which are dword-aligned (the last two bits of the starting address must be 0).

Note

It is possible to use the same image for both the main window and PIP+ window. To do so, set the PIP+ Line Address Offset register (REG[54h]) to the same value as the Main Window Line Address Offset register (REG[44h]).

Example 6: In SwivelView 90°, program the PIP+ window registers for a 320x240 panel at 4 bpp, with the PIP+ window positioned at SwivelView 90° coordinates (60, 80) with a width of 120 and a height of 160.

1. Determine the value for the PIP+ Window X Positions and PIP+ Window Y Positions registers. Let the top left corner of the PIP+ window be (x1, y1), and let the bottom right corner be (x2, y2), where $x2 = x1 + \text{width} - 1$ and $y2 = y1 + \text{height} - 1$. The PIP+ Window X Positions register sets the vertical coordinates of the PIP+ window's top right and bottom left corners. The PIP+ Window Y Positions register sets the horizontal coordinates of the PIP+ window's top right and bottom left corners.

The required values are calculated as follows:

X Start Position

$$\begin{aligned} &= y1 \\ &= 80 \\ &= 50h \end{aligned}$$

Y Start Position

$$\begin{aligned} &= (\text{panel height} - x2 - 1) \div (32 \div \text{bpp}) \\ &= (240 - (60 + 120 - 1) - 1) \div (32 \div 4) \\ &= 7.5 \\ &= 07h \text{ (truncated fractional part)} \end{aligned}$$

X End Position

$$\begin{aligned} &= y2 \\ &= 80 + 160 - 1 \\ &= 239 \\ &= EFh \end{aligned}$$

Y End Position

$$\begin{aligned} &= (\text{panel height} - x1 - 1) \div (32 \div \text{bpp}) \\ &= (240 - 60 - 1) \div (32 \div 4) \\ &= 22.375 \\ &= 16h \text{ (truncated fractional part)} \end{aligned}$$

2. Program the PIP⁺ Window X Positions register with the X Start Position in bits 9-0 and the X End Position in bits 25-16. REG[58h] is set to 00EF0050h.
Program the PIP⁺ Window Y Positions register with the Y Start Position in bits 9-0 and the Y End Position in bits 25-16. REG[5Ch] is set to 00160007h.

Due to truncation, the dimensions of the PIP⁺ window may have changed. Recalculate the PIP⁺ window width and height below:

PIP⁺ Width

$$\begin{aligned}
 &= ((\text{REG}[5\text{Ch}] \text{ bits } 25:16) - (\text{REG}[5\text{Ch}] \text{ bits } 9:0) + 1) \times (32 \div \text{bpp}) \\
 &= (16\text{h} - 07\text{h} + 1) \times (32 \div 4) \\
 &= 128 \text{ pixels (note that this is different from the desired width)}
 \end{aligned}$$

PIP Height

$$\begin{aligned}
 &= (\text{REG}[58\text{h}] \text{ bits } 25:16) - (\text{REG}[58\text{h}] \text{ bits } 9:0) + 1 \\
 &= \text{EFh} - 50\text{h} + 1 \\
 &= 160 \text{ lines}
 \end{aligned}$$

3. Determine the PIP⁺ display start address.
The main window image must take up $320 \times 240 \text{ pixels} \times \text{bpp} \div 8 = 9600\text{h}$ bytes. If the main window starts at address 0h, then the PIP⁺ window can start at 9600h.

PIP⁺ display start address

$$\begin{aligned}
 &= ((\text{desired byte address} + (\text{PIP}^+ \text{ width} \times \text{bpp} \div 8) \\
 &\quad + ((4 - (\text{PIP}^+ \text{ width} \times \text{bpp} \div 8)) \& 03\text{h}) \div 4) - 1 \\
 &= ((9600\text{h} + (128 \times 4 \div 8) + ((4 - (128 \times 4 \div 8)) \& 03\text{h})) \div 4) - 1 \\
 &= 9615 \\
 &= 258\text{Fh}
 \end{aligned}$$

Program the PIP⁺ Display Start Address register. REG[50h] is set to 0000258Fh.

4. Determine the PIP⁺ line address offset.

number of dwords per line

$$\begin{aligned}
 &= \text{image width} \div (32 \div \text{bpp}) \\
 &= 128 \div (32 \div 4) \\
 &= 16 \\
 &= 10\text{h}
 \end{aligned}$$

Program the PIP⁺ Line Address Offset register. REG[54h] is set to 00000010h.

5. Enable the PIP⁺ window.

Program the PIP⁺ Window Enable bit. REG[10h] bit 19 is set to 1.

8.2.3 SwivelView 180°

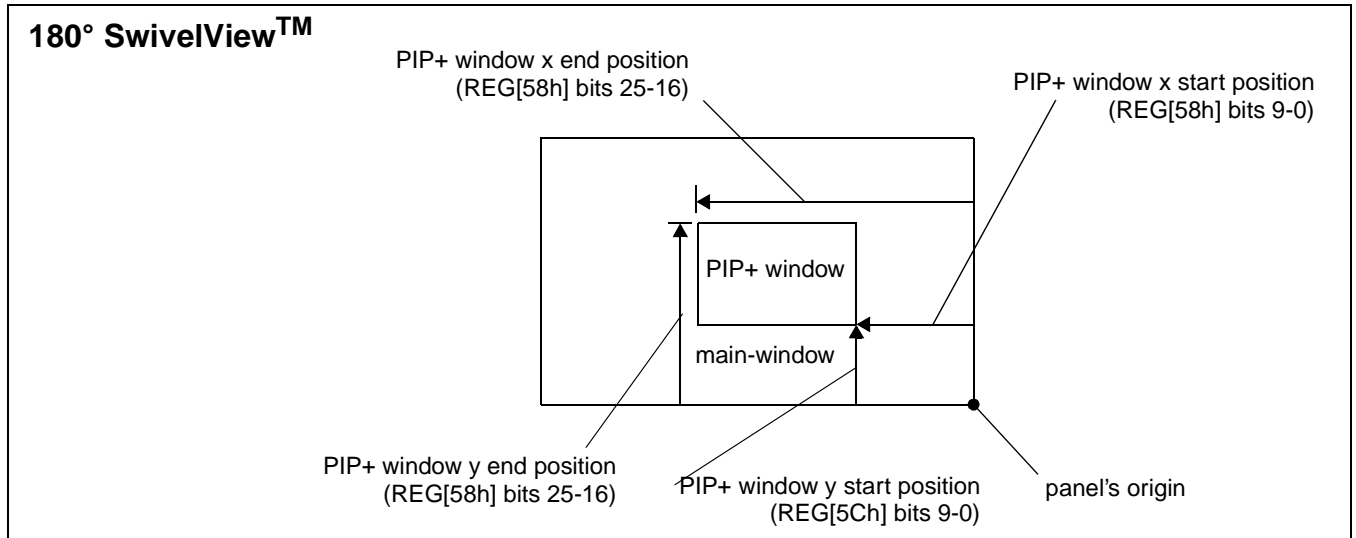


Figure 8-4: Picture-in-Picture Plus with SwivelView 180° enabled

SwivelView 180° is a mode in which both the main and PIP+ windows are rotated 180° counter-clockwise when shown on the panel. The images for each window are typically placed consecutively, with the main window image starting at address 0 and followed by the PIP+ window image. In addition, both images must start at addresses which are dword-aligned (the last two bits of the starting address must be 0).

Note

It is possible to use the same image for both the main window and PIP+ window. To do so, set the PIP+ Line Address Offset register (REG[54h]) to the same value as the Main Window Line Address Offset register (REG[44h]).

Example 7: In SwivelView 180°, program the PIP+ window registers for a 320x240 panel at 4 bpp, with the PIP+ window positioned at SwivelView 180° coordinates (80, 60) with a width of 160 and a height of 120.

1. Determine the value for the PIP+ Window X Positions and PIP+ Window Y Positions registers. Let the top left corner of the PIP+ window be (x1, y1), and let the bottom right corner be (x2, y2), where $x2 = x1 + \text{width} - 1$ and $y2 = y1 + \text{height} - 1$. The PIP+ Window X Positions register sets the horizontal coordinates of the PIP+ window's bottom right and top left corner. The PIP+ Window Y Positions register sets the vertical coordinates of the PIP+ window's bottom right and top left corner.

The required values are calculated as follows:

X Start Position

$$\begin{aligned} &= (\text{panel width} - x2 - 1) \div (32 \div \text{bpp}) \\ &= (320 - (80 + 160 - 1) - 1) \div (32 \div 4) \\ &= 10 \\ &= 0Ah \end{aligned}$$

Y Start Position

$$\begin{aligned} &= \text{panel height} - y2 - 1 \\ &= 240 - (60 + 120 - 1) - 1 \\ &= 60 \\ &= 3Ch \end{aligned}$$

X End Position

$$\begin{aligned} &= (\text{panel width} - x1 - 1) \div (32 \div \text{bpp}) \\ &= (320 - 80 - 1) \div (32 \div 4) \\ &= 29.875 \\ &= 1Dh \text{ (truncated fractional part)} \end{aligned}$$

Y End Position

$$\begin{aligned} &= \text{panel height} - y1 - 1 \\ &= 240 - 60 - 1 \\ &= 179 \\ &= B3h \end{aligned}$$

Program the PIP⁺ Window X Positions register with the X Start Position in bits 9-0 and the X End Position in bits 25-16. REG[58h] is set to 001D000Ah.

Program the PIP⁺ Window Y Positions register with the Y Start Position in bits 9-0 and the Y End Position in bits 25-16. REG[5Ch] is set to 00B3003Ch.

Due to truncation, the dimensions of the PIP⁺ window may have changed. Recalculate the PIP⁺ window width and height below:

$$\begin{aligned}\text{PIP}^+ \text{ Width} &= ((\text{REG}[58\text{h}] \text{ bits } 25:16) - (\text{REG}[58\text{h}] \text{ bits } 9:0) + 1) \times (32 \div \text{bpp}) \\ &= (1\text{Dh} - 0\text{Ah} + 1) \times (32 \div 4) \\ &= 160 \text{ pixels}\end{aligned}$$

$$\begin{aligned}\text{PIP}^+ \text{ Height} &= (\text{REG}[5\text{Ch}] \text{ bits } 25:16) - (\text{REG}[5\text{Ch}] \text{ bits } 9:0) + 1 \\ &= \text{B3h} - 3\text{Ch} + 1 \\ &= 120 \text{ lines}\end{aligned}$$

- Determine the PIP⁺ display start address.
The main window image must take up $320 \times 240 \text{ pixels} \times \text{bpp} \div 8 = 9600\text{h}$ bytes. If the main window starts at address 0h, then the PIP⁺ window can start at 9600h.

$$\begin{aligned}\text{PIP}^+ \text{ Stride} &= \text{image width} \times \text{bpp} \div 8 \\ &= 160 \times 4 \div 8 \\ &= 80 \\ &= 50\text{h}\end{aligned}$$

$$\begin{aligned}\text{PIP}^+ \text{ display start address} &= ((\text{desired byte address} + (\text{PIP}^+ \text{ Stride} \times (\text{PIP}^+ \text{ height} - 1)) \\ &\quad + (\text{PIP}^+ \text{ width} \times \text{bpp} \div 8) + ((4 - (\text{PIP}^+ \text{ width} \times \text{bpp} \div 8)) \& 03\text{h})) \div 4) - 1 \\ &= ((9600\text{h} + (80 \times (120 - 1)) + (160 \times 4 \div 8) + ((4 - (160 \times 4 \div 8)) \& 03\text{h})) \div 4) - 1 \\ &= 11999 \\ &= 2\text{EDFh}\end{aligned}$$

Program the PIP⁺ Display Start Address register. REG[50h] is set to 00002EDFh.

- Determine the PIP⁺ line address offset.

$$\begin{aligned}\text{number of dwords per line} &= \text{image width} \div (32 \div \text{bpp}) \\ &= 160 \div (32 \div 4) \\ &= 20 \\ &= 14\text{h}\end{aligned}$$

Program the PIP⁺ Line Address Offset register. REG[54h] is set to 00000014h.

- Enable the PIP⁺ window.

Program the PIP⁺ Window Enable bit. REG[10h] bit 19 is set to 1.

8.2.4 SwivelView 270°

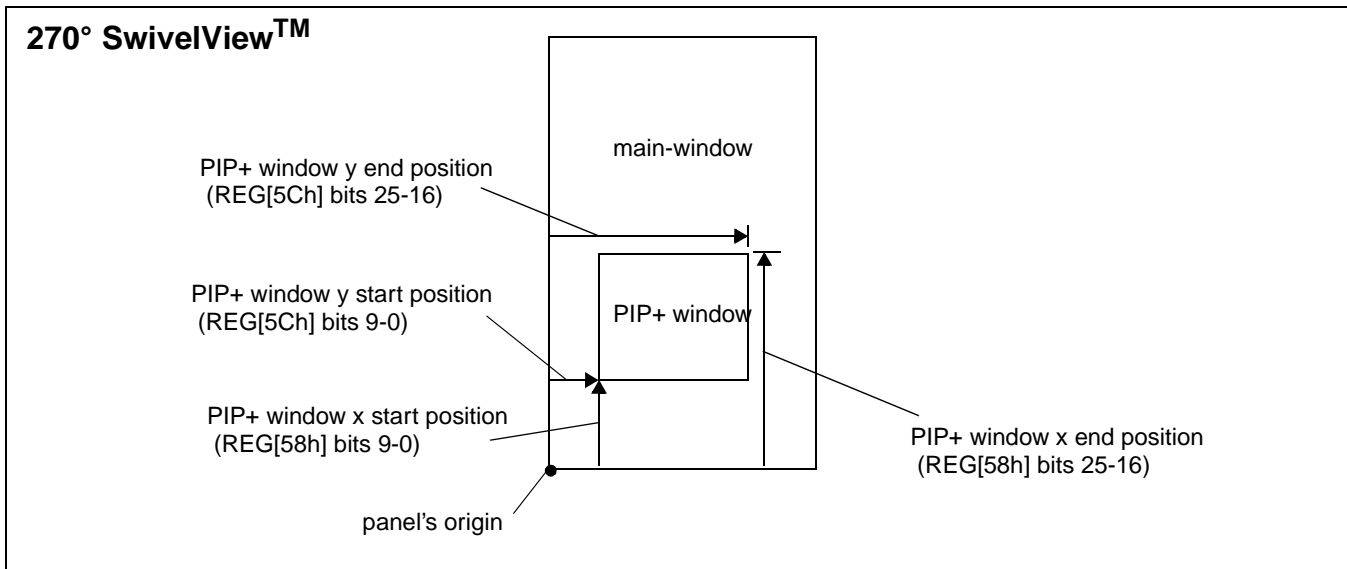


Figure 8-5: Picture-in-Picture Plus with SwivelView 270° enabled

SwivelView 270° is a mode in which both the main and PIP⁺ windows are rotated 270° counter-clockwise when shown on the panel. The images for each window are typically placed consecutively, with the main window image starting at address 0 and followed by the PIP⁺ window image. In addition, both images must start at addresses which are dword-aligned (the last two bits of the starting address must be 0).

Note

It is possible to use the same image for both the main window and PIP⁺ window. To do so, set the PIP⁺ Line Address Offset register (REG[54h]) to the same value as the Main Window Line Address Offset register (REG[44h]).

Example 8: In SwivelView 270°, program the PIP⁺ window registers for a 320x240 panel at 4 bpp, with the PIP⁺ window positioned at SwivelView 270° coordinates (60, 80) with a width of 120 and a height of 160.

1. Determine the value for the PIP⁺ Window X Positions and PIP⁺ Window Y Positions registers. Let the top left corner of the PIP⁺ window be (x1, y1), and let the bottom right corner be (x2, y2), where $x2 = x1 + \text{width} - 1$ and $y2 = y1 + \text{height} - 1$. The PIP⁺ Window X Positions register sets the vertical coordinates of the PIP⁺ window's top right and bottom left corner. The PIP⁺ Window Y Positions register sets the horizontal coordinates of the PIP⁺ window's top right and bottom left corner.

The required values are calculated as follows:

X Start Position

$$\begin{aligned} &= \text{panel width} - y2 - 1 \\ &= 320 - (80 + 160 - 1) - 1 \\ &= 80 \\ &= 50\text{h} \end{aligned}$$

Y Start Position

$$\begin{aligned} &= x1 \div (32 \div \text{bpp}) \\ &= 60 \div (32 \div 4) \\ &= 7.5 \\ &= 07\text{h (truncated fractional part)} \end{aligned}$$

X End Position

$$\begin{aligned} &= \text{panel width} - y1 - 1 \\ &= 320 - 80 - 1 \\ &= 239 \\ &= \text{EFh} \end{aligned}$$

Y End Position

$$\begin{aligned} &= x2 \div (32 \div \text{bpp}) \\ &= (60 + 120 - 1) \div (32 \div 4) \\ &= 22.375 \\ &= 16\text{h (truncated fractional part)} \end{aligned}$$

2. Program the PIP⁺ Window X Positions register with the X Start Position in bits 9-0 and the X End Position in bits 25-16. REG[58h] is set to 00EF0050h.
Program the PIP⁺ Window Y Positions register with the Y Start Position in bits 9-0 and the Y End Position in bits 25-16. REG[5Ch] is set to 00160007h.

Due to truncation, the dimensions of the PIP⁺ window may have changed. Recalculate the PIP⁺ window width and height below:

PIP⁺ Width

$$\begin{aligned} &= ((\text{REG}[5\text{Ch}] \text{ bits } 25:16) - (\text{REG}[5\text{Ch}] \text{ bits } 9:0) + 1) \times (32 \div \text{bpp}) \\ &= (16\text{h} - 07\text{h} + 1) \times (32 \div 4) \\ &= 128 \text{ pixels (note that this is different from the desired width)} \end{aligned}$$

PIP Height

$$\begin{aligned} &= (\text{REG}[58\text{h}] \text{ bits } 25:16) - (\text{REG}[58\text{h}] \text{ bits } 9:0) + 1 \\ &= \text{EFh} - 50\text{h} + 1 \\ &= 160 \text{ lines} \end{aligned}$$

3. Determine the PIP⁺ display start address.
The main window image must take up $320 \times 240 \text{ pixels} \times \text{bpp} \div 8 = 9600\text{h}$ bytes. If the main window starts at address 0h, then the PIP⁺ window can start at 9600h.

PIP⁺ Stride

$$\begin{aligned} &= \text{image width} \times \text{bpp} \div 8 \\ &= 128 \times 4 \div 8 \\ &= 64 \\ &= 40\text{h} \end{aligned}$$

PIP⁺ display start address

$$\begin{aligned} &= (\text{desired byte address} + ((\text{PIP}^+ \text{ height} - 1) \times \text{PIP}^+ \text{ Stride})) \div 4 \\ &= (9600\text{h} + ((160 - 1) \times 64)) \div 4 \\ &= 12144 \\ &= 2\text{F}70\text{h} \end{aligned}$$

Program the PIP⁺ Display Start Address register. REG[50h] is set to 00002F70h.

4. Determine the PIP⁺ line address offset.

number of dwords per line

$$\begin{aligned} &= \text{image width} \div (32 \div \text{bpp}) \\ &= 128 \div (32 \div 4) \\ &= 16 \\ &= 10\text{h} \end{aligned}$$

Program the PIP⁺ Line Address Offset register. REG[54h] is set to 00000010h.

5. Enable the PIP⁺ window.

Program the PIP⁺ Window Enable bit. REG[10h] bit 19 is set to 1.

8.3 Limitations

8.3.1 SwivelView 0° and 180°

The PIP⁺ Line Address Offset register (REG[54h]) requires the PIP⁺ window image *width* to be a multiple of $32 \div \text{bits-per-pixel}$. If this formula is not satisfied, then the PIP⁺ Line Address Offset register must be programmed to the next larger value that satisfies the formula.

8.3.2 SwivelView 90° and 270°

The PIP⁺ Line Address Offset register (REG[54h]) requires the PIP⁺ window image *width* to be a multiple of $32 \div \text{bits-per-pixel}$. If this formula is not satisfied, then the PIP⁺ Line Address Offset register must be programmed to the next larger value that satisfies the formula.

9 2D BitBLT Engine

BitBLT is an acronym for Bit Block Transfer. The 2D BitBLT Engine in the S1D13A04 is designed to increase the speed of the most common GUI operations by off-loading work from the CPU, reducing traffic on the system bus and freeing the CPU sooner for other tasks.

BitBLTs require a destination - a place to write the display data. Most BitBLTs have a source of data for the BitBLT and many also incorporate a pattern. The pattern, source, and destination operands are combined using logical AND, OR, XOR and NOT operations. The combining process is called a Raster Operation (ROP) and results in the final pixel data to be written to the destination address.

The S1D13A04 2D BitBLT engine supports a total of sixteen ROPs and works at 8 bpp and 16 bpp color depths. This section describes the BitBLT registers and provides some sample BitBLT operations.

9.1 Registers

The S1D13A04 BitBLT registers are located 8000h bytes from the start of S1D13A04 address space. The registers are labelled, according to their byte offset, as REG[8000h] through REG[8024h]. The following is a description of all BitBLT registers.

BitBLT Control Register														Read/Write					
REG[8000h]														Default = 00000000h					
n/a														Color Format Select	Dest Linear Select	Source Linear Select			
31	30	29	28	27	26	25	24	23	22	21	20	19	18				17	16	
n/a														BitBLT Enable (WO)					
15	14	13	12	11	10	9	8	7	6	5	4	3	2		1	0			

Color Format Select

The Color Format Select bit indicates to the BitBLT engine what color depth to assume for the BitBLT operation. The BitBLT engine uses this information to set the step size for internal counters.

When this bit = 0, 8 bpp is selected and when this bit = 1, 16 bpp is selected.

Destination Linear Select

The Destination Linear Select bit determines how the BitBLT destination address pointer is updated when the BitBLT reaches the end of a row.

When the end of a row is reached and rectangular is selected the destination address is updated to point to the beginning of the next row of a rectangular area. The offset to the start of the next row is contained in the BitBLT Memory Address Offset register (REG[8014h]).

When the end of a row is reached and destination linear is selected the destination address is updated to the next available memory offset. The result is data which is jammed together with one row immediately following the next in display memory. This is useful when it is desired to compactly save a rectangular area into off screen memory.

When this bit = 0, the BitBLT destination is stored as a rectangular region of memory. When this bit = 1, the BitBLT destination is stored as a contiguous linear block of memory.

Source Linear Select

The Source Linear Select bit determines how the source address pointer is updated when the BitBLT reaches the end of a row.

When the end of a row is reached and rectangular is selected the source address is updated to point to the beginning of the next row of a rectangular area. The offset to the start of the next row is contained in the BitBLT Memory Address Offset register (REG[8014h]).

When the end of a row is reached and source linear is selected the source address is updated to the next available memory offset. The result is data, which was jammed together with one row immediately following the next in display memory, can now be expanded back to a rectangular area.

When this bit = 0, the BitBLT source is stored as a rectangular region of memory. When this bit = 1, the BitBLT source is stored as a contiguous linear block of memory.

BitBLT Enable

This bit is write only.

Setting this bit to 1 begins the 2D BitBLT operation. **This bit must not be set to 0 while a BitBLT operation is in progress.**

Note

To determine the status of a BitBLT operation use the BitBLT Busy Status bit (REG[8004h] bit 0).

BitBLT Status Register																
REG[8004h]											Default = 00000000h				Read Only	
n/a			Number of Used FIFO Entries						n/a			Number of Free FIFO Entries (0 means full)				
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	
n/a									FIFO Not Empty	FIFO Half Full	FIFO Full Status	n/a			BitBLT Busy Status	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	

Number of Used FIFO Entries

This is a read-only status.

This field indicates the minimum number of FIFO entries currently in use (there may be more in the internal pipeline). If these bits return a 0, the FIFO is empty.

Number of Free FIFO Entries

This is a read-only status bit

This field indicates the number of empty FIFO entries available. If these bits return a 0, the FIFO is full.

FIFO Not-Empty

This is a read-only status bit.

When this bit = 0, the BitBLT FIFO is empty. When this bit = 1, the BitBLT FIFO has at least one data. To reduce system latency, software can monitor this bit prior to a BitBLT read burst operation.

The following table shows the number of words available in the BitBLT FIFO under different status conditions.

Table 9-1: BitBLT FIFO Words Available

BitBLT FIFO Full Status (REG[8004h] Bit 4)	BitBLT FIFO Half Full Status (REG[8004h] Bit 5)	BitBLT FIFO Not Empty Status (REG[8004h] Bit 6)	Number of Words available in BitBLT FIFO
0	0	0	0
0	0	1	1 to 6
0	1	1	7 to 14
1	1	1	15 to 16

BitBLT FIFO Half Full Status

This is a read-only status bit.

When this bit = 1, the BitBLT FIFO is half full or greater than half full. When this bit = 0, the BitBLT FIFO is less than half full.

BitBLT FIFO Full Status

This is a read-only status bit.

When this bit = 1, the BitBLT FIFO is full. When this bit = 0, the BitBLT FIFO is not full.

BitBLT Busy Status

This bit is a read-only status bit.

When this bit = 1, the BitBLT operation is in progress. When this bit = 0, the BitBLT operation is complete.

Note

During a BitBLT Read operation, the BitBLT engine does not attempt to keep the FIFO full. If the FIFO becomes full, the BitBLT operation stops temporarily as data is read out of the FIFO. The BitBLT will restart only when less than 14 values remain in the FIFO.

BitBLT Command Register												Read/Write			
REG[8008h]												Default = 00000000h			
n/a												BitBLT ROP Code bits 3-0			
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a												BitBLT Operation bits 3-0			
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

BitBLT ROP Code

The BitBLT ROP Code specifies the Raster Operation to be used for Write and Move BitBLTs. In addition, for Color Expansion, the BitBLT ROP Code bits 2-0 specify the start bit position for Color Expansion BitBLTs.

Table 9-2 : BitBLT ROP Code/Color Expansion Function Selection

BitBLT ROP Code Bits [3:0]	Boolean Function for Write BitBLT and Move BitBLT	Boolean Function for Pattern Fill	Start Bit Position for Color Expansion
0000	0 (Blackness)	0 (Blackness)	bit 0
0001	$\sim S \cdot \sim D$ or $\sim(S + D)$	$\sim P \cdot \sim D$ or $\sim(P + D)$	bit 1
0010	$\sim S \cdot D$	$\sim P \cdot D$	bit 2
0011	$\sim S$	$\sim P$	bit 3
0100	$S \cdot \sim D$	$P \cdot \sim D$	bit 4
0101	$\sim D$	$\sim D$	bit 5
0110	$S \wedge D$	$P \wedge D$	bit 6
0111	$\sim S + \sim D$ or $\sim(S \cdot D)$	$\sim P + \sim D$ or $\sim(P \cdot D)$	bit 7
1000	$S \cdot D$	$P \cdot D$	bit 0
1001	$\sim(S \wedge D)$	$\sim(P \wedge D)$	bit 1
1010	D	D	bit 2
1011	$\sim S + D$	$\sim P + D$	bit 3
1100	S	P	bit 4
1101	$S + \sim D$	$P + \sim D$	bit 5
1110	$S + D$	$P + D$	bit 6
1111	1 (Whiteness)	1 (Whiteness)	bit 7

Note

S = Source, D = Destination, P = Pattern.
 \sim = NOT, \cdot = Logical AND, $+$ = Logical OR, \wedge = Logical XOR

BitBLT Operation

The BitBLT Operation selects which BitBLT operation performed. The following table lists the available BitBLT operations.

Table 9-3 : BitBLT Operation Selection

BitBLT Operation Bits [3:0]	BitBLT Operation
0000	Write BitBLT with ROP This operation refers to BitBLTs where data is to be transferred from system memory to display memory
0001	Read BitBLT This operation refers to BitBLTs where data is to be transferred from display memory to system memory
0010	Move BitBLT in positive direction with ROP This operation is used to transfer data from display memory to display memory
0011	Move BitBLT in negative direction with ROP This operation is used to transfer data from display memory to display memory
0100	Transparent Write BitBLT Like the Write BitBLT this operation is used when transferring data from system memory to display memory, the difference is that destination pixels will be left "as is" when source pixels of a specified color are encountered.
0101	Transparent Move BitBLT in positive direction As with the Move BitBLTs this operation is used to transfer data from display memory to display memory. The difference is that destination pixels will be left "as is" when source pixels of a specified color are encountered.
0110	Pattern Fill with ROP Fills the specified area of display memory with a repeating pattern stored in display memory.
0111	Pattern Fill with transparency As with the Pattern Fill, this BitBLT fills a specified area of display memory with a repeating pattern, destination pixels will be left "as is" when source pixels of a specified color are encountered.
1000	Color Expansion This BitBLT expands the bits of the source data into full pixels at the destination. If a source bit is 0 the destination pixel will be background color and if the source bit is 1 the destination pixel will be of foreground color. The source data for Color Expansion BitBLTs is always system memory.
1001	Color Expansion with transparency Like the Color Expansion BitBLT, this operations expands each bit of the source data to occupy a full destination pixel. The difference, is that destination pixels corresponding to source bits of 0 will be left "as is". The data source is system memory
1010	Move BitBLT with Color Expansion This BitBLT works the same as the Color Expansion BitBLT however the source of the BitBLT is display memory.
1011	Move BitBLT with Color Expansion and transparency This BitBLT works the same as the Color Expansion with Transparency BitBLT however the source of the BitBLT is display memory.
1100	Solid Fill BitBLT Use this BitBLT to fill a given area with one solid color.
Other combinations	Reserved

BitBLT Source Start Address Register											Read/Write					
REG[800Ch]											Default = 00000000h					
n/a											BitBLT Source Start Address bits 20-16					
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	
BitBLT Source Start Address bits 15-0																
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	

BitBLT Source Start Address

This register has multiple meanings depending on the BitBLT operation it specifies. It can be either:

- the start address in display memory of the source data for BitBLTs where the source is display memory (i.e. Move BitBLTs).
- in pattern fill operations, the BitBLT Source Start Address determines where in the pattern to begin the BitBLT operation and is defined by the following equation:

$$\text{Value programmed to the Source Start Address Register} = \text{Pattern Base Address} + \text{Pattern Line Offset} + \text{Pixel Offset}.$$
- the data alignment for 16 bpp BitBLTs where the source of BitBLT data is the CPU (i.e. Write BitBLTs).

The following table shows how Source Start Address Register is defined for 8 and 16 bpp color depths.

Table 9-4 : BitBLT Source Start Address Selection

Color Format	Pattern Base Address[20:0]	Pattern Line Offset[2:0]	Pixel Offset[3:0]
8 bpp	BitBLT Source Start Address[20:6]	BitBLT Source Start Address[5:3]	BitBLT Source Start Address[2:0]
16 bpp	BitBLT Source Start Address[20:7]	BitBLT Source Start Address[6:4]	BitBLT Source Start Address[3:0]

BitBLT Destination Start Address Register											Read/Write					
REG[8010h]											Default = 00000000h					
n/a											BitBLT Destination Start Address bits 20-16					
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	
BitBLT Destination Start Address bits 15-0																
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	

BitBLT Destination Start Address

This register specifies the initial destination address for BitBLT operations. For rectangular destinations this address represents the upper left corner of the BitBLT rectangle. If the operation is a Move BitBLT in a Negative Direction, these bits define the address of the lower right corner of the rectangle.

BitBLT Memory Address Offset Register															
REG[8014h] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						BitBLT Memory Address Offset bits 10-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

BitBLT Memory Address Offset

This register specifies the 11-bit address offset from the starting word of line n to the starting word of line $n + 1$. The offset value is only used for address calculation when the BitBLT is configured as rectangular.

BitBLT Width Register															
REG[8018h] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						BitBLT Width bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

BitBLT Width

This register specifies the width of a BitBLT in pixels - 1.

$$\text{BitBLT width (in pixels)} = \text{REG}[8018\text{h}] + 1$$

BitBLT Height Register															
REG[801Ch] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						BitBLT Height bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

BitBLT Height

This register specifies the height of the BitBLT in lines - 1.

$$\text{BitBLT height (in lines)} = \text{REG}[801\text{Ch}] + 1$$

BitBLT Background Color Register															
REG[8020h] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
BitBLT Background Color bits 15-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

BitBLT Background Color

This register specifies either:

- the BitBLT background color for Color Expansion

or

- the key color for Transparent BitBLT. For 8 bpp BitBLTs, bits 7-0 are used to specify the key color and for 16 bpp BitBLTs, bits 15-0 are used.

BitBLT Foreground Color Register															
REG[8024h] Default = 00000000h															
Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
BitBLT Foreground Color bits 15-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

BitBLT Foreground Color

This register specifies the foreground color for Color Expansion or Solid Fill BitBLTs. For 8 bpp BitBLTs, bits 7-0 are used to specify the color and for 16 bpp BitBLTs, bits 15-0 are used.

2D Accelerator (BitBLT) Data Memory Mapped Region Register															
AB16-AB0 = 10000h-1FFFEh, even addresses															
Read/Write															
BitBLT Data bits 31-16															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
BitBLT Data bits 15-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

BitBLT Data bits

This register is used by the local CPU to send data to the BitBLT engine for Write and Color Expansion BitBLTs and is used to read data from the BitBLT engine for Read BitBLTs. The register should be treated as any other register it is however loosely decoded from 10000h to 1FFFEh.

Note

The BitBLT data registers are 32 bits wide but are accessed on WORD boundaries using 16 bit accesses. Byte access to the BitBLT data registers is not allowed.

Note

Accesses to this register, other than for purposes of a BitBLT operation may cause the 13A04 to stop responding and the system to hang.

9.2 BitBLT Descriptions

The S1D13A04 supports 13 fundamental BitBLT operations:

- Write BitBLT with ROP
- Read BitBLT
- Move BitBLT in positive direction with ROP
- Move BitBLT in negative direction with ROP
- Transparent Write BitBLT
- Transparent Move BitBLT in positive direction
- Pattern Fill with ROP
- Pattern Fill with Transparency
- Color Expansion
- Color Expansion with Transparency
- Move BitBLT with Color Expansion
- Move BitBLT with Color Expansion and Transparency
- Solid Fill

Most of the 13 operations are self completing. This means once the BitBLT operation begins it completes without further assistance from the local CPU. No data transfers are required to or from the local CPU. Five BitBLT operations (Write BitBLT with ROP, Transparent Write BitBLT, Color Expansion, Color Expansion with Transparency, Read BitBLT) require data to be written to/read from the BitBLT engine. This data must be transferred one word (16-bits) at a time. This does not imply only 16-bit CPU instructions are acceptable. If a system is able to separate one DWORD write into two WORD writes and the CPU writes the low word before the high word, then 32-bit CPU instructions are acceptable. Otherwise, 16-bit CPU instructions are required.

The data is not directly written to/read from the display buffer. It is written to/read from the BitBLT FIFO through the 64K byte BitBLT aperture specified at the address of REG[10000h]. The 16 word FIFO can be written to only when not full and can be read from only when not empty. Failing to monitor the FIFO status can result in a BitBLT FIFO overflow or underflow.

While the FIFO is being written to by the CPU, it is also being emptied by the S1D13A04. If the S1D13A04 empties the FIFO faster than the CPU can fill it, it may not be possible to cause an overflow/underflow. In these cases, performance can be improved by not monitoring the FIFO status. However, this is very much platform dependent and must be determined for each system.

9.2.1 Write BitBLT with ROP

Write BitBLTs increase the speed of transferring data from system memory to the display buffer. The Write BitBLT with ROP accepts data from the CPU and *writes* it into display memory. This BitBLT is typically used to copy a bitmap image from system memory to the display buffer.

Write BitBLTs support 16 ROPs, the most frequently used being ROP 0Ch (Copy Source to Destination). Write BitBLTs support both rectangular and linear destinations. Using a linear destination it is possible to move an image to off screen memory in a compact format for later restoration using a Move BitBLT.

During a Write BitBLT operation the BitBLT engine expects to receive a particular number of WORDs and it is the responsibility of the CPU to provide the required amount of data.

When performing BitBLT at 16 bpp color depth the number of WORDS to be sent is the same as the number of pixels to be transferred as each pixel is one WORD wide. The number of WORD writes the BitBLT engine expects is calculated using the following formula.

$$\begin{aligned}\text{WORDS} &= \text{Pixels} \\ &= \text{BitBLTWidth} \times \text{BitBLTHeight}\end{aligned}$$

When the color depth is 8 bpp the formula must take into consideration that the BitBLT engine accepts only WORD accesses and each pixel is one BYTE. This may lead to a different number of WORD transfers than there are pixels to transfer.

The number of WORD accesses is dependant on the position of the first pixel within the first WORD of each row. Is the pixel stored in the low byte or the high byte of the WORD? This aspect of the BitBLT is called phase and is determined as follows:

Source phase is 0 when the first pixel is in the low byte and the second pixel is in the high byte of the WORD. When the source phase is 0, bit 0 of the Source Start Address Register is 0. The Source Phase is 1 if the first pixel of each row is contained in the high byte of the WORD, the contents of the low byte are ignored. When the source phase is 1, bit 0 of the Source Start Address Register is set.

Depending on the Source Phase and the BitBLT Width, the last WORD may contain only one pixel. In this case it is always in the low byte. The number of WORD writes the BitBLT engine expects for 8 bpp color depths is shown in the following formula.

$$\text{WORDS} = ((\text{BitBLTWidth} + 1 + \text{SourcePhase}) \div 2) \times \text{BitBLTHeight}$$

The BitBLT engine requires this number of WORDS to be sent from the local CPU before it will end the Write BitBLT operation.

Note

The BitBLT engine counts WORD writes made to the BitBLT register space. This does not imply only 16-bit CPU instructions are acceptable. If a system is able to separate one DWORD write into two WORD writes **and** the CPU writes the low word before the high word, then 32-bit CPU instructions are acceptable. Otherwise, 16-bit CPU instructions are required.

Example 9: Write a 100 x 20 rectangle at the screen coordinates $x = 25$, $y = 38$ using a 320x240 display at a color depth of 8 bpp.

1. Calculate the destination address (upper left corner of the screen BitBLT rectangle) using the following formula.

$$\begin{aligned} \text{DestinationAddress} &= (y \times \text{ScreenStride}) + (x \times \text{BytesPerPixel}) \\ &= (38 \times 320) + (25 \times 1) \\ &= 12185 \\ &= 2F99\text{h} \end{aligned}$$

where:

BytesPerPixel = 1 for 8 bpp

BytesPerPixel = 2 for 16 bpp

ScreenStride = DisplayWidthInPixels \times BytesPerPixel = 320 for 8 bpp

Program the BitBLT Destination Start Address Register. REG[8010h] is set to 2F99h.

2. Program the BitBLT Width Register to 100 - 1. REG[8018h] is set to 63h (99 decimal).
3. Program the BitBLT Height Register to 20 - 1. REG[801Ch] is set to 13h (19 decimal).
4. Program the Source Phase in the BitBLT Source Start Address Register. In this example the data is WORD aligned, so the source phase is 0. REG[800Ch] is set to 00h.
5. Program the BitBLT Operation Register to select the Write BitBLT with ROP. REG[8008h] bits 3-0 are set to 0h.
6. Program the BitBLT ROP Code Register to select Destination = Source. REG[8008h] bits 19-16 are set to 0Ch.
7. Program the BitBLT Color Format Select bit for 8 bpp operations. REG[8000h] bit 18 is set to 0.
8. Program the BitBLT Memory Offset Register to the ScreenStride in WORDS:

$$\begin{aligned} \text{BLTMemoryOffset} &= \text{DisplayWidthInPixels} \div \text{BytesPerPixel} \\ &= 320 \div 2 \\ &= \text{A0h} \end{aligned}$$

REG[8014h] is set to A0h.

9. Calculate the number of WORDS the BitBLT engine expects to receive.

$$\begin{aligned}
 \text{WORDS} &= ((\text{BLTWidth} + 1 + \text{SourcePhase}) \div 2) \times \text{BLTHeight} \\
 &= (100 + 1) \div 2 \times 20 \\
 &= 1000 \\
 &= 3E8h
 \end{aligned}$$

10. Program the BitBLT Destination/Source Linear Select bits for a rectangular BitBLT (BitBLT Destination Linear Select = 0, BitBLT Source Linear Select = 0).

Start the BitBLT operation **and wait for the BitBLT engine to start**. REG[8000h] bit 0 is set to 1, then wait until REG[8004h] bit 0 returns a 1.

11. Prior to writing any data to the BitBLT FIFO, confirm the BitBLT FIFO is not full (REG[8004h] bit 4 returns a 0).

If the BitBLT FIFO Not Empty Status (REG[8004h] bit 6) returns a 0, the FIFO is empty. Write up to 16 WORDS to the BitBLT data register area.

If the BitBLT FIFO Not Empty Status (REG[8004h] bit 6) returns a 1 and the BitBLT FIFO Half Full Status (REG[8004h] bit 5) returns a 0 then you can write up to 8 WORDS.

If the BitBLT FIFO Full Status returns a 1, do not write to the BitBLT FIFO until it returns a 0.

The following table summarizes how many words can be written to the BitBLT FIFO.

Table 9-5: Possible BitBLT FIFO Writes

BitBLT Status Register (REG[8004h])			Word Writes Available
FIFO Not Empty Status	FIFO Half Full Status	FIFO Full Status	
0	0	0	16
1	0	0	8
1	1	0	up to 8
1	1	1	0 (do not write)

Note

The sequence of register initialization is irrelevant as long as all required registers are programmed before the BitBLT is started.

9.2.2 Color Expansion BitBLT

Similar to the Write BitBLT, the Color Expansion BitBLT requires the CPU to feed data to the BitBLT data register. It differs in that bits set to one in the source data become a complete pixel of foreground color. Source bits set to zero are converted to a pixel of background color. The intended use of this BitBLT operation is to increase the speed of writing text to display memory. As with the Write BitBLT, all data sent to the BitBLT engine must be WORD (16-bit) writes.

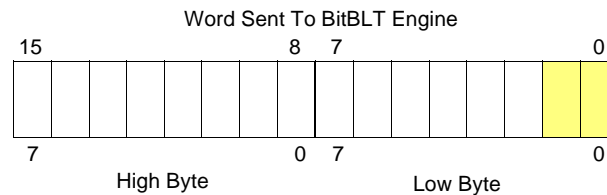
The BitBLT engine expands first the low byte, then the high byte starting at bit 7 of each byte. The start byte of the first WORD to be expanded and the start bit position within this byte must be specified. The start byte position is selected by setting source address bit 0 to 0 to start expanding the low byte or 1 to start expanding the high byte.

Partially “masked” color expansion BitBLTs can be used when drawing a portion of a pattern (i.e. a portion of a character) on the screen. The following examples illustrate how one WORD is expanded using the Color Expansion BitBLT.

1. To expand bits 0-1 of the word:

Source Address = 0
Start Bit Position = 1
BitBLT Width = 2

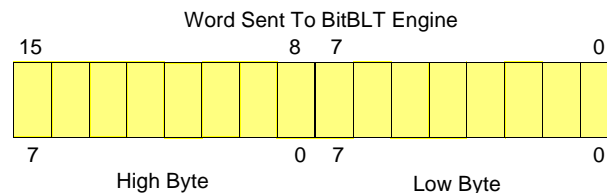
The following bits are expanded.



2. To expand bits 0-15 of the word (entire word)

Source Address = 0
Start Bit Position = 7 (bit seven of the low byte)
BitBLT Width = 16

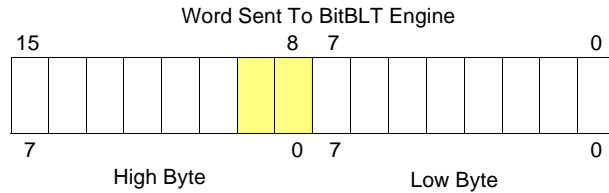
The following bits are expanded.



3. To expand bits 8-9 of the word

Source Address = 1
Start Bit Position = 1
BitBLT Width = 2

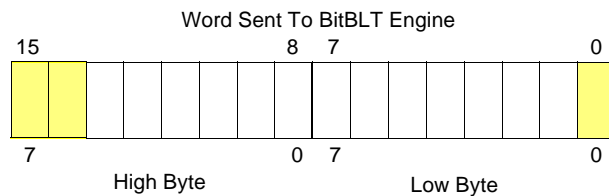
The following bits are expanded.



4. To expand bits 0,15-14 of the word

Source Address = 0
Start Bit Position = 0
BitBLT Width = 3

The following bits are expanded.

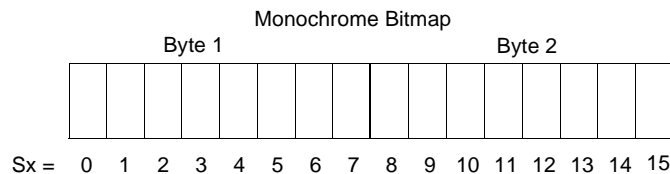


All subsequent WORDS in one BitBLT line are then serially expanded starting at bit 7 of the low byte until the end of the BitBLT line. All unused bits in the last WORD are discarded. It is extremely important that the exact number of WORDS is provided to the BitBLT engine. The number of WORDS is calculated from the following formula. This formula is valid for all color depths (8/16 bpp).

$$\text{WORDS} = ((S_x \text{ MOD } 16 + \text{BitBLTWidth} + 15) \div 16) \times \text{BitBLTHeight}$$

where:

S_x is the X coordinate of the starting pixel in a word aligned monochrome bitmap.



**Example 10: Color expand a rectangle of 12 x 18 starting at the coordinates
Sx = 125, Sy = 17 using a 320x240 display at a color depth of 8 bpp.**

This example assumes a monochrome, WORD aligned bitmap of dimensions 300 x 600 with the origin at an address A. The color expanded rectangle will be displayed at the screen coordinates X = 20, Y = 30. The foreground color corresponds to the LUT entry at index 134, the background color to index 124.

1. First we need to calculate the address of the WORD within the monochrome bitmap containing the pixel x = 125, y = 17.

$$\begin{aligned}
 \text{SourceAddress} &= \text{BitmapOrigin} + (y \times \text{SourceStride}) + (x \div 8) \\
 &= A + (Sy \times \text{SourceStride}) + (Sx \div 8) \\
 &= A + (17 \times 38) + (125 \div 8) \\
 &= A + 646 + 15 \\
 &= A + 661
 \end{aligned}$$

where:

$$\begin{aligned}
 \text{SourceStride} &= (\text{BitmapWidth} + 15) \div 16 \\
 &= (300 + 15) \div 16 \\
 &= 19 \text{ WORDS per line} \\
 &= 38 \text{ BYTES per line}
 \end{aligned}$$

2. Calculate the destination address (upper left corner of the screen BitBLT rectangle) using the following formula.

$$\begin{aligned}
 \text{DestinationAddress} &= (Y \times \text{ScreenStride}) + (X \times \text{BytesPerPixel}) \\
 &= (30 \times 320) + (20 \times 1) \\
 &= 9620 \\
 &= 2594h
 \end{aligned}$$

where:

BytesPerPixel = 1 for 8 bpp

BytesPerPixel = 2 for 16 bpp

ScreenStride = DisplayWidthInPixels \times BytesPerPixel = 320 for 8 bpp

Program the BitBLT Destination Start Address Register. REG[8010h] is set to 2594h.

3. Program the BitBLT Width Register to 12 - 1. REG[8018h] is set to 0Bh (11 decimal).
4. Program the BitBLT Height Register to 18 - 1. REG[801Ch] is set to 11h (17 decimal).
5. Program the Source Phase in the BitBLT Source Start Address Register. In this example the source address equals A + 661 (odd), so REG[800Ch] is set to 1.

Since only bit 0 flags the source phase, more efficient code would simply write the low byte of the SourceAddress into REG[800Ch] directly -- not needing to test for an odd/even address. Note that in 16 bpp color depths the Source address is guaranteed to be even.

6. Program the BitBLT Operation Register to select the Color Expand BitBLT. REG[8008h] bits 3-0 are set to 8h.

7. Program the Color Expansion Register. The formula for this example is as follows.

$$\begin{aligned} \text{Color Expansion} &= 7 - (Sx \text{ MOD } 8) \\ &= 7 - (125 \text{ MOD } 8) \\ &= 7 - 5 \\ &= 2 \end{aligned}$$

REG[8008h] is set to 2h.

8. Program the Background Color Register to the background color. REG[8020h] is set to 7Ch (124 decimal).
9. Program the Foreground Color Register to the foreground color. REG[8024h] is set to 86h (134 decimal).
10. Program the BitBLT Color Format Register for 8 bpp operation. REG[8000h] bit 18 is set to 0.
11. Program the BitBLT Memory Offset Register to the ScreenStride in WORDS.

$$\begin{aligned} \text{BltMemoryOffset} &= \text{ScreenStride} \div 2 \\ &= 320 \div 2 \\ &= \text{A0h} \end{aligned}$$

REG[8014h] is set to A0h.

12. Calculate the number of WORDS the BitBLT engine expects to receive.

First, the number of WORDS in one BitBLT line must be calculated as follows.

$$\begin{aligned} \text{WordsOneLine} &= ((125 \text{ MOD } 16) + 12 + 15) \div 16 \\ &= (13 + 12 + 15) \div 16 \\ &= 40 \div 16 \\ &= 2 \end{aligned}$$

Therefore, the total WORDS the BitBLT engine expects to receive is calculated as follows.

$$\begin{aligned} \text{WORDS} &= \text{WordsOneLine} \times 18 \\ &= 2 \times 18 \\ &= 36 \end{aligned}$$

13. Program the BitBLT Destination/Source Linear Select bits for a rectangular BitBLT (BitBLT Destination Linear Select = 0, BitBLT Source Linear Select = 0).

Start the BitBLT operation and **wait for the BitBLT Engine to start**. REG[8000h] bit 0 is set to 1, then wait until REG[8004h] bit 0 returns a 1.

14. Prior to writing all WORDS to the BitBLT FIFO, confirm the BitBLT FIFO is not full (REG[8004h] bit 4 returns a 0). One WORD expands into 16 pixels which fills all 16 FIFO words in 16 bpp or 8 FIFO words in 8 bpp.

The following table summarizes how many words can be written to the BitBLT FIFO.

Table 9-6: Possible BitBLT FIFO Writes

BitBLT Status Register (REG[8004h])			8 bpp Word Writes Available	16 bpp Word Writes Available
FIFO Not Empty Status	FIFO Half Full Status	FIFO Full Status		
0	0	0	2	1
1	0	0	1	0 (do not write)
1	1	0	0 (do not write)	
1	1	1		

Note

The sequence of register setup is irrelevant as long as all required registers are programmed before the BitBLT is started.

9.2.3 Color Expansion BitBLT With Transparency

This BitBLT operation is virtually identical to the Color Expand BitBLT, the difference is in how background bits are handled. Bits in the source bitmap which are set to zero result in the destination pixel being left untouched. Bits set to one are expanded to the foreground color.

Use this BitBLT operation to overlay text onto any background while leaving the background intact.

Refer to the Color Expansion BitBLT for sample calculations and keep the following points in mind:

- Program the BitBLT operation bits, REG[8008h] bits 3-0, to 09h instead of 08h.
- Setting a background color, REG[8020h], is not required.

9.2.4 Solid Fill BitBLT

The Solid Fill BitBLT fills a rectangular area of the display buffer with a solid color. This operation is used to paint large screen areas or to set areas of the display buffer to a given value.

This BitBLT operation is self completing. After setting the width, height, destination start position and (foreground) color the BitBLT engine is started. When the region of display memory is filled with the given color the BitBLT engine will automatically stop.

Example 11: Fill a red 9 x 301 rectangle at the screen coordinates x = 100, y = 10 using a 320x240 display at a color depth of 16 bpp.

1. Calculate the destination address (upper left corner of the destination rectangle) using the following formula.

$$\begin{aligned} \text{DestinationAddress} &= (y \times \text{ScreenStride}) + (x \times \text{BytesPerPixel}) \\ &= (10 \times (320 \times 2)) + (100 \times 2) \\ &= 6600 \\ &= 19C8h \end{aligned}$$

where:

BytesPerPixel = 1 for 8 bpp

BytesPerPixel = 2 for 16 bpp

ScreenStride = DisplayWidthInPixels \times BytesPerPixel = 640 for 16 bpp.

Program the BitBLT Destination Start Address Register. REG[8010h] is set to 19C8h.

2. Program the BitBLT Width Register to 9 - 1. REG[8018h] is set to 08h.
3. Program the BitBLT Height Register to 301 - 1. REG[801Ch] is set to 12Ch (300 decimal).
4. Program the BitBLT Foreground Color Register. REG[8024h] is set to F800h (Full intensity red in 16 bpp is F800h).
5. Program the BitBLT Operation Register to select Solid Fill. REG[8008h] bits 3-0 are set to 0Ch.
6. Program the BitBLT Color Format Register for 16 bpp operations. REG[8000h] bit 18 is set to 1.
7. Program the BitBLT Memory Offset Register to the ScreenStride in WORDS.

$$\begin{aligned} \text{BltMemoryOffset} &= \text{ScreenStride} \div 2 \\ &= 320 \\ &= 140h \end{aligned}$$

REG[8014h] is set to 0140h.

8. Program the BitBLT Destination/Source Linear Select bits for a rectangular BitBLT (BitBLT Destination Linear Select = 0, BitBLT Source Linear Select = 0).

Start the BitBLT operation. REG[8000h] bit 0 is set to 1.

Note

The sequence of register setup is irrelevant as long as all required registers are programmed before the BitBLT is started.

9.2.5 Move BitBLT in a Positive Direction with ROP

The Move BitBLT is used to copy one area of display memory to another area in display memory.

The source and the destination areas of the BitBLT may be either rectangular or linear. Performing a rectangular to rectangular Move BitBLT creates an exact copy of one portion of video memory at the second location. Selecting a rectangular source to linear destination would be used to compactly store an area of displayed video memory into non-displayed video memory. Later, the area could be restored by performing a linear source to rectangular destination Move BitBLT.

The Move BitBLT in a Positive Direction with ROP is a self completing operation. Once the width, height and the source and destination start addresses are setup and the BitBLT is started the BitBLT engine will complete the operation automatically.

Should the source and destination areas overlap a decision has to be made as to whether to use a Positive or Negative direction so that source data is not overwritten by the move before it is used. Refer to Figure 9-1: to see when to make the decision to switch to the Move BitBLT in a Negative direction. When the destination area overlaps the original source area and the destination address is greater then the source address, use the Move BitBLT in Negative Direction with ROP.

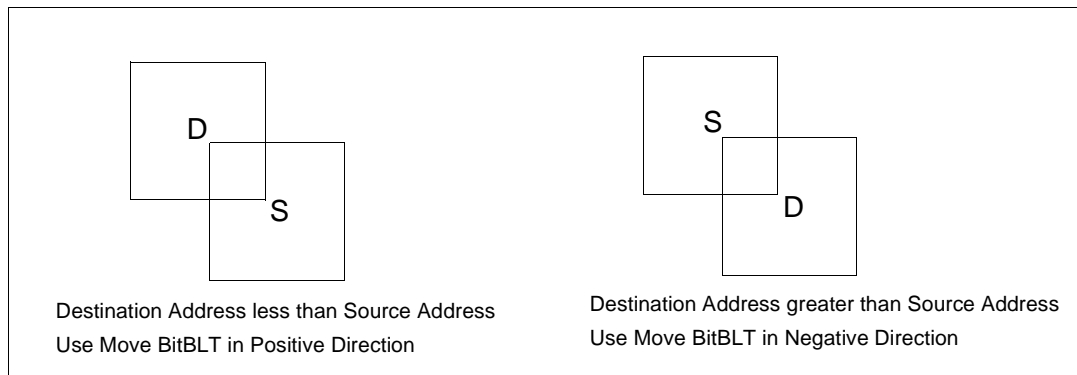


Figure 9-1: Move BitBLT Usage

Example 12: Copy a 9 x 101 rectangle at the screen coordinates $x = 100$, $y = 10$ to screen coordinates $x = 200$, $y = 20$ using a 320x240 display at a color depth of 16 bpp.

1. Calculate the source and destination addresses (upper left corners of the source and destination rectangles), using the following formula.

$$\begin{aligned} \text{SourceAddress} &= (y \times \text{ScreenStride}) + (x \times \text{BytesPerPixel}) \\ &= (10 \times (320 \times 2)) + (100 \times 2) \\ &= 6600 \\ &= 19C8h \end{aligned}$$

$$\begin{aligned} \text{DestinationAddress} &= (y \times \text{ScreenStride}) + (x \times \text{BytesPerPixel}) \\ &= (20 \times (320 \times 2)) + (200 \times 2) \\ &= 13200 \\ &= 3390h \end{aligned}$$

where:

BytesPerPixel = 1 for 8 bpp

BytesPerPixel = 2 for 16 bpp

ScreenStride = DisplayWidthInPixels \times BytesPerPixel = 640 for 16 bpp

Program the BitBLT Source Start Address Register. REG[800Ch] is set to 19C8h.

Program the BitBLT Destination Start Address Register. REG[8010h] is set to 3390h.

2. Program the BitBLT Width Register to 9 - 1. REG[8018h] is set to 08h.
3. Program the BitBLT Height Register to 101 - 1. REG[801Ch] is set to 64h (100 decimal).
4. Program the BitBLT Operation Register to select the Move BitBLT in Positive Direction with ROP. REG[8008h] bits 3-0 are set to 2h.
5. Program the BitBLT ROP Code Register to select Destination = Source. REG[8008h] bits 19-16 are set to 0Ch.
6. Program the BitBLT Color Format Select bit for 16 bpp operations. REG[8000h] bit 18 is set to 1.
7. Program the BitBLT Memory Offset Register to the ScreenStride in WORDS.

$$\begin{aligned} \text{BltMemoryOffset} &= \text{ScreenStride} \div 2 \\ &= 320 \\ &= 140h \end{aligned}$$

REG[8014h] is set to 0140h.

8. Program the BitBLT Destination/Source Linear Select bits for a rectangular BitBLT (BitBLT Destination Linear Select = 0, BitBLT Source Linear Select = 0).

Start the BitBLT operation. REG[8000h] bit 0 is set to 1.

Note

The sequence of register setup is irrelevant as long as all required registers are programmed before the BitBLT is started.

9.2.6 Move BitBLT in Negative Direction with ROP

The Move BitBLT in Negative Direction with ROP is similar to the Move BitBLT in Positive direction. Use this BitBLT operation when the source and destination BitBLT areas overlap and the destination address is greater than the source address. Refer to Figure 9-1: on page 76 to see when to make the decision to switch to the Move BitBLT in a Positive direction.

When using the Move BitBLT in Negative Direction it is necessary to calculate the addresses of the last pixels as opposed to the first pixels. This means calculating the addresses of the lower right corners as opposed to the upper left corners.

Example 13: Copy a 9 x 101 rectangle at the screen coordinates x = 100, y = 10 to screen coordinates X = 105, Y = 20 using a 320x240 display at a color depth of 16 bpp.

In the following example, the coordinates of the source and destination rectangles intentionally overlap.

1. Calculate the source and destination addresses (**lower right** corners of the source and destination rectangles) using the following formula.

$$\begin{aligned}
 \text{SourceAddress} &= ((y + \text{Height} - 1) \times \text{ScreenStride}) + ((x + \text{Width} - 1) \times \text{BytesPerPixel}) \\
 &= ((10 + 101 - 1) \times (320 \times 2)) + ((100 + 9 - 1) \times 2) \\
 &= 70616 \\
 &= 113D8h
 \end{aligned}$$

$$\begin{aligned}
 \text{DestinationAddress} &= ((Y + \text{Height} - 1) \times \text{ScreenStride}) + ((X + \text{Width} - 1) \times \text{BytesPerPixel}) \\
 &= ((20 + 101 - 1) \times (320 \times 2)) + ((105 + 9 - 1) \times 2) \\
 &= 77026 \\
 &= 12CE2h
 \end{aligned}$$

where:

BytesPerPixel = 1 for 8 bpp

BytesPerPixel = 2 for 16 bpp

ScreenStride = DisplayWidthInPixels \times BytesPerPixel = 640 for 16 bpp

Program the BitBLT Source Start Address Register. REG[800Ch] is set to 113D8h.

Program the BitBLT Destination Start Address Register. REG[8010h] is set to 12CE2h.

2. Program the BitBLT Width Register to 9 - 1. REG[8018h] is set to 08h.
3. Program the BitBLT Height Register to 101 - 1. REG[801Ch] is set to 64h (100 decimal).
4. Program the BitBLT Operation Register to select the Move BitBLT in Negative Direction with ROP. REG[8008] bits 3-0 are set to 3h.
5. Program the BitBLT ROP Code Register to select Destination = Source. REG[8008h] bits 19-16 are set to 0Ch.
6. Program the BitBLT Color Format Select bit for 16 bpp operations. REG[8000h] bit 18 is set to 1.
7. Program the BitBLT Memory Offset Register to the ScreenStride in WORDS.

$$\begin{aligned}\text{BltMemoryOffset} &= \text{ScreenStride} \div 2 \\ &= 320 \\ &= 140\text{h}\end{aligned}$$

REG[8014h] is set to 0140h.

8. Program the BitBLT Destination/Source Linear Select bits for a rectangular BitBLT (BitBLT Destination Linear Select = 0, BitBLT Source Linear Select = 0).

Start the BitBLT operation. REG[8000h] bit 0 is set to 1.

Note

The sequence of register setup is irrelevant as long as all required registers are programmed before the BitBLT is started.

9.2.7 Transparent Write BitBLT

Transparent Write BitBLTs are similar to the Write BitBLT with ROP with two differences; first, a specified color in the source data leaves the destination pixel untouched and second ROPs are not supported.

This operation is used to copy a bitmap image from system memory to the display buffer with one color marked as transparent. Pixels of the transparent color are not transferred. This allows fast display of non-rectangular or masked images. For example, consider a source bitmap having a red circle on a blue background. By selecting the blue as the transparent color and using the Transparent Write BitBLT on the whole rectangle, the effect is a BitBLT of the red circle only.

During a Transparent Write BitBLT operation the BitBLT engine expects to receive a particular number of WORDs and it is the responsibility of the CPU to provide the required amount of data.

When performing BitBLTs at 16 bpp color depth the number of WORDS to be sent is the same as the number of pixels as each pixel is one WORD wide. The number of WORD writes the BitBLT engine expects is calculated using the following formula.

$$\begin{aligned}\text{WORDS} &= \text{Pixels} \\ &= \text{BitBLTWidth} \times \text{BitBLTHeight}\end{aligned}$$

When the color depth is 8 bpp the formula must take into consideration that the BitBLT engine accepts only WORD accesses and each pixel is one BYTE. This may lead to a different number of WORD transfers than there are pixels to transfer.

The number of WORD accesses is dependant on the position of the first pixel within the first WORD of each row. Is the pixel stored in the low byte or the high byte of the WORD? This aspect of the BitBLT is called phase and is determined as follows:

Source phase is 0 when the first pixel is in the low byte and the second pixel is in the high byte of the WORD. When the source phase is 0, bit 0 of the Source Start Address Register is 0. The Source Phase is 1 if the first pixel of each row is contained in the high byte of the WORD, the contents of the low byte are ignored. When the source phase is 1, bit 0 of the Source Start Address Register is set.

Depending on the Source Phase and the BitBLT Width, the last WORD may contain only one pixel. In this case it is always in the low byte. The number of WORD writes the BitBLT engine expects for 8 bpp color depths is shown in the following formula.

$$\text{WORDS} = ((\text{BitBLTWidth} + 1 + \text{SourcePhase}) \div 2) \times \text{BitBLTHeight}$$

Once the Transparent Write BitBLT begins, the BitBLT engine remains active until all pixels have been written. The BitBLT engine requires this number of WORDS to be sent from the local CPU before it ends the Transparent Write BitBLT operation.

Note

The BitBLT engine counts WORD writes made to the BitBLT register. This does not imply only 16-bit CPU instructions are acceptable. If a system is able to separate one DWORD write into two WORD writes **and** the CPU writes the low word before the high word, then 32-bit CPU instructions are acceptable. Otherwise, 16-bit CPU instructions are required.

Example 14: Write 100 x 20 pixels at the screen coordinates x = 25, y = 38 using a 320x240 display at a color depth of 8 bpp. Transparent color is high intensity blue (assume LUT Index 124).

1. Calculate the destination address (upper left corner of the screen BitBLT rectangle), using the formula:

$$\begin{aligned}\text{DestinationAddress} &= (y \times \text{ScreenStride}) + (x \times \text{BytesPerPixel}) \\ &= (38 \times 320) + (25 \times 1) \\ &= 12185 \\ &= 2F99\text{h}\end{aligned}$$

where:

BytesPerPixel = 1 for 8 bpp

BytesPerPixel = 2 for 16 bpp

ScreenStride = DisplayWidthInPixels × BytesPerPixel = 320 for 8 bpp

Program the BitBLT Destination Start Address Register. REG[8010h] is set to 2F99h.

2. Program the BitBLT Width Register to 100 - 1. REG[8018h] is set to 63h (99 decimal).
3. Program the BitBLT Height Register to 20 - 1. REG[801Ch] is set to 13h (19 decimal).
4. Program the Source Phase in the BitBLT Source Start Address Register. In this example, the data is WORD aligned, so the source phase is 0. REG[800Ch] is set to 00h.
5. Program the BitBLT Operation Register to select Transparent Write BitBLT. REG[8008h] bits 3-0 are set to 4h.
6. Program the BitBLT Background Color Register to select transparent color. REG[8020h] is set to 7Ch (124 decimal).
7. Program the BitBLT Color Format Select bit for 8 bpp operations. REG[8000h] bit 18 is set to 0.
8. Program the BitBLT Memory Offset Register to the ScreenStride in WORDS.

$$\begin{aligned}\text{BltMemoryOffset} &= \text{ScreenStride} \div 2 \\ &= 320 \div 2 \\ &= 160 \\ &= \text{A0h}\end{aligned}$$

REG[8014h] is set to 0A0h.

9. Calculate the number of WORDS the BitBLT engine expects to receive.

$$\begin{aligned}\text{WORDS} &= ((\text{BLTWidth} + 1 + \text{SourcePhase}) \div 2) \times \text{BLTHeight} \\ &= (100 + 1 + 0) \div 2 \times 20 \\ &= 1000 \\ &= \text{3E8h}\end{aligned}$$

10. Program the BitBLT Destination/Source Linear Select bits for a rectangular BitBLT (BitBLT Destination Linear Select = 0, BitBLT Source Linear Select = 0).

Start the BitBLT operation **and wait for the BitBLT engine to start**. REG[8000h] bit 0 is set to 1, then wait until REG[8004h] bit 0 returns a 1.

11. Prior to writing any data to the BitBLT FIFO, confirm the BitBLT FIFO is not full (REG[8004h] bit 4 returns a 0).
 If the BitBLT FIFO Not Empty Status (REG[8004h] bit 6) returns a 0, the FIFO is empty. Write up to 16 WORDS to the BitBLT data register area.
 If the BitBLT FIFO Not Empty Status (REG[8004h] bit 6) returns a 1 and the BitBLT FIFO Half Full Status (REG[8004h] bit 5) returns a 0 then you can write up to 8 WORDS.
 If the BitBLT FIFO Full Status returns a 1, do not write to the BitBLT FIFO until it returns a 0.

The following table summarizes how many words can be written to the BitBLT FIFO.

Table 9-7: Possible BitBLT FIFO Writes

BitBLT Status Register (REG[8004h])			Word Writes Available
FIFO Not Empty Status	FIFO Half Full Status	FIFO Full Status	
0	0	0	16
1	0	0	8
1	1	0	less than 8
1	1	1	0 (do not write)

Note

The sequence of register setup is irrelevant as long as all required registers are programmed before the BitBLT is started.

9.2.8 Transparent Move BitBLT in Positive Direction

The Transparent Move BitBLT in Positive Direction combines the capabilities of the Move BitBLT with the ability to define a transparent color. Use this operation to copy a masked area of display memory to another area in display memory.

The source and the destination areas of the BitBLT may be either rectangular or linear. Performing a rectangular to rectangular Move BitBLT creates an exact copy of one portion of video memory at the second location. Selecting a rectangular source to linear destination would be used to compactly store an area of displayed video memory into non-displayed video memory. Later, the area could be restored by performing a linear source to rectangular destination Move BitBLT.

The transparent color is not copied during this operation, whatever pixel color existed in the destination will be there when the BitBLT completes. This allows fast display of non-rectangular images. For example, consider a source bitmap having a red circle on a blue background. By selecting the blue color as the transparent color and using the Transparent Move BitBLT on the whole rectangle, the effect is a BitBLT of the red circle only.

Note

The Transparent Move BitBLT is supported **only** in a positive direction.

Example 15: Copy a 9 x 101 rectangle at the screen coordinates $x = 100$, $y = 10$ to screen coordinates $X = 200$, $Y = 20$ using a 320x240 display at a color depth of 16 bpp. Transparent color is blue.

1. Calculate the source and destination addresses (upper left corners of the source and destination rectangles), using the formula:

$$\begin{aligned}\text{SourceAddress} &= (y \times \text{ScreenStride}) + (x \times \text{BytesPerPixel}) \\ &= (10 \times (320 \times 2)) + (100 \times 2) \\ &= 6600 \\ &= 19C8\text{h}\end{aligned}$$

$$\begin{aligned}\text{DestinationAddress} &= (Y \times \text{ScreenStride}) + (X \times \text{BytesPerPixel}) \\ &= (20 \times (320 \times 2)) + (200 \times 2) \\ &= 13200 \\ &= 3390\text{h}\end{aligned}$$

where:

BytesPerPixel = 1 for 8 bpp

BytesPerPixel = 2 for 16 bpp

ScreenStride = DisplayWidthInPixels \times BytesPerPixel = 640 for 16 bpp

Program the BitBLT Source Start Address Register. REG[800Ch] is set to 19C8h.

Program the BitBLT Destination Start Address Register. REG[8010h] is set to 3390h.

2. Program the BitBLT Width Register to 9 - 1. REG[8018h] is set to 08h.
3. Program the BitBLT Height Register to 101 - 1. REG[801Ch] is set to 64h (100 decimal).
4. Program the BitBLT Operation Register to select the Transparent Move BitBLT in Positive Direction. REG[8008h] bits 3-0 are set to 05h.
5. Program the BitBLT Background Color Register to select blue as the transparent color. REG[8020h] is set to 001Fh (Full intensity blue in 16 bpp is 001Fh).
6. Program the BitBLT Color Format Register to select 16 bpp operations. REG[8000h] bit 18 is set to 1.
7. Program the BitBLT Memory Offset Register to the ScreenStride in WORDS.

$$\begin{aligned}\text{BlMemoryOffset} &= \text{ScreenStride} \div 2 \\ &= 320 \\ &= 140\text{h}\end{aligned}$$

REG[8014h] is set to 0140h.

8. Program the BitBLT Destination/Source Linear Select bits for a rectangular BitBLT (BitBLT Destination Linear Select = 0, BitBLT Source Linear Select = 0).

Start the BitBLT operation. REG[8000h] bit 0 is set to 1.

Note

The order of register setup is irrelevant as long as all relevant registers are programmed before the BitBLT is initiated.

9.2.9 Pattern Fill BitBLT with ROP

The Pattern Fill BitBLT with ROP fills a specified area of display memory with a pattern. The pattern is repeated until the fill area is completely filled. The fill pattern is limited to an eight by eight pixel array and must be loaded to off-screen video memory before starting the BitBLT. The pattern can be logically combined with the destination using any of the 16 ROP codes, but typically the copy pattern ROP is used (ROP code 0Ch).

A pattern is defined to be an array of 8x8 pixels and the pattern data must be stored in consecutive bytes of display memory (64 consecutive bytes for 8 bpp color depths and 128 bytes for 16 bpp color depths). For 8 bpp color depths the pattern must begin on a 64 byte boundary, for 16 bpp color depths the pattern must begin on a 128 byte boundary.

This operation is self completing. Once the parameters have been entered and the BitBLT started the BitBLT engine will fill all of the specified memory with the pattern.

To fill an area using the pattern BitBLT, the BitBLT engine requires the location of the pattern, the destination rectangle position and size, and the ROP code. The BitBLT engine also needs to know which pixel from the pattern is the first pixel in the destination rectangle (the pattern start phase). This allows seamless redrawing of any part of the screen using the pattern fill.

Example 16: Fill a 100 x 150 rectangle at the screen coordinates $x = 10$, $y = 20$ with the pattern in off-screen memory at offset 27000h using a 320x240 display at a color depth of 8 bpp. The first pixel (upper left corner) of the rectangle is the pattern pixel at $x = 3$, $y = 4$.

1. Calculate the destination address (upper left corner of the destination rectangle), using the formula:

$$\begin{aligned} \text{DestinationAddress} &= (y \times \text{ScreenStride}) + (x \times \text{BytesPerPixel}) \\ &= (20 \times 320) + (10 \times 1) \\ &= 6410 \\ &= 190Ah \end{aligned}$$

where:

BytesPerPixel = 1 for 8 bpp

BytesPerPixel = 2 for 16 bpp

ScreenStride = DisplayWidthInPixels \times BytesPerPixel = 320 for 8 bpp

Program the BitBLT Destination Start Address Register. REG[8010h] is set to 190Ah.

2. Calculate the source address. This is the address of the pixel in the pattern that is the origin of the destination fill area. The pattern begins at offset 156K but the first pattern pixel is at $x = 3$, $y = 4$. Therefore, an offset within the pattern itself must be calculated.

SourceAddress

$$\begin{aligned}
 &= \text{PatternOffset} + \text{StartPatternY} \times 8 \times \text{BytesPerPixel} + \text{StartPatternX} \times \text{BytesPerPixel} \\
 &= 156\text{K} + (4 \times 8 \times 1) + (3 \times 1) \\
 &= 156\text{K} + 35 \\
 &= 159779 \\
 &= 27023\text{h}
 \end{aligned}$$

where:

BytesPerPixel = 1 for 8 bpp

BytesPerPixel = 2 for 16 bpp

Program the BitBLT Source Start Address Register. REG[800Ch] is set to 27023h.

3. Program the BitBLT Width Register to 100 - 1. REG[8018h] is set to 63h (99 decimal).
4. Program the BitBLT Height Register to 150-1. REG[801Ch] is set to 95h (149 decimal).
5. Program the BitBLT Operation Register to select the Pattern Fill with ROP. REG[8008h] bits 3-0 are set to 6h.
6. Program the BitBLT ROP Code Register to select Destination = Source. REG[8008h] bits 19-16 are set to 0Ch.
7. Program the BitBLT Color Format Select bit for 8 bpp operations. REG[8000h] bit 18 is set to 0.
8. Program the BitBLT Memory Offset Register to the ScreenStride in WORDS.

$$\begin{aligned}
 \text{BltMemoryOffset} &= \text{ScreenStride} \div 2 \\
 &= 320 \div 2 \\
 &= 160 \\
 &= \text{A0h}
 \end{aligned}$$

REG[8014h] is set to 00A0h.

9. Program the BitBLT Destination/Source Linear Select bits for a rectangular BitBLT (BitBLT Destination Linear Select = 0, BitBLT Source Linear Select = 0).

Start the BitBLT operation. REG[8000h] bit 0 is set to 1.

Note

The sequence of register setup is irrelevant as long as all required registers are programmed before the BitBLT is started.

9.2.10 Pattern Fill BitBLT with Transparency

This operation is very similar to the Pattern Fill BitBLT with the difference being that one color can be specified to be transparent. Whenever the Transparent color is encountered in the pattern data the destination is left as is. This operation is useful to create hatched or striped patterns where the original image shows through the hatching.

The requirements for this BitBLT are the same as for the Pattern Fill BitBLT the only change in programming is that the BitBLT Operation field of REG[8008h] must be set to 07h and the BitBLT Background color register, REG[8020h] must be set to the desired color.

Example 17: Fill a 100 x 150 rectangle at the screen coordinates $x = 10$, $y = 20$ with the pattern in off-screen memory at offset 27000h using a 320x240 display at a color depth of 8 bpp. The first pixel (upper left corner) of the rectangle is the pattern pixel at $x = 3$, $y = 4$. Transparent color is blue (assumes LUT index 1).

1. Calculate the destination address (upper left corner of destination rectangle), using the formula:

$$\begin{aligned} \text{DestinationAddress} &= (y \times \text{ScreenStride}) + (x \times \text{BytesPerPixel}) \\ &= (20 \times 320) + (10 \times 1) \\ &= 6410 \\ &= 190Ah \end{aligned}$$

where:

BytesPerPixel = 1 for 8 bpp

BytesPerPixel = 2 for 16 bpp

ScreenStride = DisplayWidthInPixels \times BytesPerPixel = 320 for 8 bpp

Program the BitBLT Destination Start Address Register. REG[8010h] is set to 190Ah.

2. Calculate the source address. This is the address of the pixel in the pattern that is the origin of the destination fill area. The pattern begins at offset 1M, but the first pattern pixel is at $x = 3$, $y = 4$. Therefore, an offset within the pattern itself must be calculated.

SourceAddress

$$\begin{aligned} &= \text{PatternOffset} + \text{StartPatternY} \times 8 \times \text{BytesPerPixel} + \text{StartPatternX} \times \text{BytesPerPixel} \\ &= 156K + (4 \times 8 \times 1) + (3 \times 1) \\ &= 156K + 35 \\ &= 159779 \\ &= 27023h \end{aligned}$$

where:

BytesPerPixel = 1 for 8 bpp

BytesPerPixel = 2 for 16 bpp

Program the BitBLT Source Start Address Register. REG[800Ch] is set to 27023h.

3. Program the BitBLT Width Register to 100 - 1. REG[8018h] is set to 63h (99 decimal).

4. Program the BitBLT Height Register to 150-1. REG[801Ch] is set to 95h (149 decimal).
5. Program the BitBLT Operation Register to select the Pattern Fill BitBLT with Transparency. REG[8008h] bits 3-0 are set to 7h.
6. Program the BitBLT Background Color Register to select transparent color. This example uses blue (LUT index 1) as the transparent color. REG[8020h] is set to 01h.
7. Program the BitBLT Color Format Select bit for 8 bpp operations. REG[8000h] bit 18 is set to 0.
8. Program the BitBLT Memory Offset Register to the ScreenStride in WORDS.

$$\begin{aligned}\text{BltMemoryOffset} &= \text{ScreenStride} \div 2 \\ &= 320 \div 2 \\ &= 160 \\ &= \text{A0h}\end{aligned}$$

REG[8014h] is set to A0h.

9. Program the BitBLT Destination/Source Linear Select bits for a rectangular BitBLT (BitBLT Destination Linear Select = 0, BitBLT Source Linear Select = 0).

Start the BitBLT operation. REG[8000h] bit 0 is set to 1.

Note

The sequence of register setup is irrelevant as long as all required registers are programmed before the BitBLT is started.

9.2.11 Move BitBLT with Color Expansion

The Move BitBLT with Color Expansion takes a monochrome bitmap as the source and color expands it into the destination. All bits set to one in the source are expanded to destination pixels of the selected foreground color. All bits set to zero in the source are expanded to pixels of the selected background color.

The Move BitBLT with Color Expansion is used to accelerate text drawing. A monochrome bitmap of a font, in off-screen video memory, occupies very little space and takes advantage of the hardware acceleration. Since the foreground and background colors are programmable, text of any color can be created.

The Move BitBLT with Color Expansion can move data from one rectangular area to another, or either the source or destination may be specified to be linear. Storing rectangular display data in linear format in off screen memory results in a tremendous space saving.

Example 18: Color expand a 9 x 16 rectangle using the pattern in off-screen memory at 27000h and move it to the screen coordinates x = 200, y = 20. Assume a 320x240 display at a color depth of 16 bpp, Foreground color of black, and background color of white.

1. Calculate the destination and source addresses (upper left corner of the destination and source rectangles), using the formula.

$$\begin{aligned} \text{DestinationAddress} &= (y \times \text{ScreenStride}) + (x \times \text{BytesPerPixel}) \\ &= (20 \times (320 \times 2)) + (200 \times 2) \\ &= 13200 \\ &= 3390\text{h} \end{aligned}$$

where:

BytesPerPixel = 1 for 8 bpp

BytesPerPixel = 2 for 16 bpp

ScreenStride = DisplayWidthInPixels \times BytesPerPixel = 640 for 16 bpp

$$\begin{aligned} \text{SourceAddress} &= 156\text{K} \\ &= 27000\text{h} \end{aligned}$$

Program the BitBLT Destination Start Address Register. REG[8010h] is set to 3390h.

Program the BitBLT Source Start Address Register. REG[800Ch] is set to 27000h.

2. Program the BitBLT Width Register to 9 - 1. REG[8018h] is set to 08h.
3. Program the BitBLT Height Register to 16 - 1. REG[801Ch] is set to 0Fh.
4. Program the BitBLT ROP Code/Color Expansion Register. REG[8008h] bits 19-16 are set to 7h.
5. Program the BitBLT Operation Register to select the Move BitBLT with Color Expansion. REG[8008h] bits 3-0 are set to 0Bh.
6. Program the BitBLT Foreground Color Register to select black (in 16 bpp black = 0000h). REG[8024h] is set to 0000h.

7. Program the BitBLT Background Color Register to select white (in 16 bpp white = FFFFh). REG[8024h] is set to FFFFh.
8. Program the BitBLT Color Format Select bit for 16 bpp operations. REG[8000h] bit 18 is set to 1.
9. Program the BitBLT Memory Offset Register to the ScreenStride in WORDS.

$$\begin{aligned}\text{BltMemoryOffset} &= \text{ScreenStride} \div 2 \\ &= 320 \\ &= 140\text{h}\end{aligned}$$

REG[8014h] is set to 0140h.

10. Program the BitBLT Destination/Source Linear Select bits for a rectangular BitBLT (BitBLT Destination Linear Select = 0, BitBLT Source Linear Select = 0).

Start the BitBLT operation. REG[8000h] bit 0 is set to 1.

Note

The sequence of register setup is irrelevant as long as all required registers are programmed before the BitBLT is started.

9.2.12 Transparent Move BitBLT with Color Expansion

The Transparent Move BitBLT with Color Expansion is virtually identical to the Move BitBLT with Color Expansion. This operation expands bits set to one in the source bitmap to the foreground color. Bits set to zero in the source bitmap leave the corresponding destination pixel as is.

Setup and use this operation is exactly as with the Move BitBLT with Color Expansion.

9.2.13 Read BitBLT

This Read BitBLT increases the speed of transferring data from the video memory to system memory. This BitBLT complements the Write BitBLT and is typically used to save a part of the display buffer to the system memory. Once the Read BitBLT begins, the BitBLT engine remains active until all the pixels have been read.

During a Read BitBLT operation the BitBLT engine expects to send a particular number of WORDs to the CPU and it is the responsibility of the CPU to read the required amount of data.

When performing BitBLT at 16 bpp color depth the number of WORDS to be sent is the same as the number of pixels to be transferred as each pixel is one WORD wide. The number of WORD writes the BitBLT engine expects is calculated using the following formula.

$$\begin{aligned}\text{WORDS} &= \text{Pixels} \\ &= \text{BitBLTWidth} \times \text{BitBLTHeight}\end{aligned}$$

When the color depth is 8 bpp the formula must take into consideration that the BitBLT engine accepts only WORD accesses and pixels are only one BYTE. This may lead to a different number of WORD transfers than there are pixels to transfer.

The number of WORD accesses is dependant on the position of the first pixel within the first WORD of each destination row. Is the pixel stored in the low byte or the high byte of the WORD? Read BitBLT phase is determined as follows:

Destination phase is 0 when the first pixel is in the low byte and the second pixel is in the high byte of the WORD. When the destination phase is 0, bit 0 of the Destination Start Address Register is 0. The destination phase is 1 if the first pixel of each destination row is contained in the high byte of the WORD, the contents of the low byte are ignored. When the destination phase is 1, bit 0 of the Destination Start Address Register is set.

Depending on the destination phase and the BitBLT width, the last WORD may contain only one pixel. In this case it is always in the low byte. The number of WORD writes the BitBLT engine expects for 8 bpp color depths is shown in the following formula.

$$\text{WORDS} = ((\text{BitBLTWidth} + 1 + \text{DestinationPhase}) \div 2) \times \text{BitBLTHeight}$$

The BitBLT engine requires this number of WORDS to be sent from the local CPU before it will end the Write BitBLT operation.

Example 19: Read 100 x 20 pixels at the screen coordinates x = 25, y = 38 and save to system memory. Assume a display of 320x240 at a color depth of 8 bpp.

1. Calculate the source address (upper left corner of the screen BitBLT rectangle), using the formula.

$$\begin{aligned}\text{SourceAddress} &= (y \times \text{ScreenStride}) + (x \times \text{BytesPerPixel}) \\ &= (38 \times 320) + (25 \times 1) \\ &= 12185 \\ &= 2F99\text{h}\end{aligned}$$

where:

BytesPerPixel = 1 for 8 bpp

BytesPerPixel = 2 for 16 bpp

ScreenStride = DisplayWidthInPixels \times BytesPerPixels = 320 for 8 bpp

Program the BitBLT Source Start Address Register. REG[800Ch] is set to 2F99h.

2. Program the BitBLT Width Register to 100 - 1. REG[8018h] is set to 63h (99 decimal).
3. Program the BitBLT Height Register to 20 - 1. REG[801Ch] is set to 13h (19 decimal).

4. Program the Destination Phase in the BitBLT Destination Start Address Register. In this example, the data is WORD aligned, so the destination phase is 0. REG[8010h] is set to 00h.
5. Program the BitBLT Operation to select the Read BitBLT. REG[8008h] bits 3-0 are set to 1h.
6. Program the BitBLT Color Format Select bit for 8 bpp operations. REG[8000h] bit 18 is set to 0.
7. Program the BitBLT Memory Offset Register to the ScreenStride in WORDS.

$$\begin{aligned} \text{BltMemoryOffset} &= \text{ScreenStride} \div 2 \\ &= 320 \div 2 \\ &= 160 \\ &= \text{A0h} \end{aligned}$$

REG[8014h] is set to 0A0h.

8. Calculate the number of WORDS the BitBLT engine expects to receive.

$$\begin{aligned} \text{WORDS} &= ((\text{BLTWidth} + 1 + \text{DestinationPhase}) \div 2) \times \text{BLTHeight} \\ &= (100 + 1 + 0) \div 2 \times 20 \\ &= 1000 \\ &= \text{3E8h} \end{aligned}$$

9. Program the BitBLT Destination/Source Linear Select bits for a rectangular BitBLT (BitBLT Destination Linear Select = 0, BitBLT Source Linear Select = 0).

Start the BitBLT operation **and wait for the BitBLT engine to start**. REG[8000h] bit 0 is set to 1, then wait until REG[8004h] bit 0 returns a 1.

10. Prior to reading from the BitBLT FIFO, confirm the BitBLT FIFO is not empty (REG[8004h] bit 4 returns a 1). If the BitBLT FIFO Not Empty Status (REG[8004h] bit 6) returns a 1 and the BitBLT FIFO Half Full Status (REG[8004h] bit 5) returns a 0 then you can read up to 8 WORDS. If the BitBLT FIFO Full Status returns a 1, read up to 16 WORDS. If the BitBLT FIFO Not Empty Status returns a 0 (the FIFO is empty), do not read from the BitBLT FIFO until it returns a 1.

The following table summarizes how many words can be read from the BitBLT FIFO.

Table 9-8: Possible BitBLT FIFO Reads

BitBLT Status Register (REG[8004h])			Word Reads Available
FIFO Not Empty Status	FIFO Half Full Status	FIFO Full Status	
0	0	0	0 (do not read)
1	0	0	up to 8
1	1	0	8
1	1	1	16

Note

The sequence of register initialization is irrelevant as long as all required registers are programmed before the BitBLT is started.

9.3 S1D13A04 BitBLT Synchronization

A BitBLT operation can only be started if the BitBLT engine is not busy servicing another BitBLT. Before a new operation is started, software must confirm the BitBLT Busy Status bit (REG[8004h] bit 0) is set to zero. The status of this bit can either be tested **after** each BitBLT operation, or **before** each BitBLT operation.

Testing the BitBLT Status After

Testing the BitBLT Active Status after starting a new BitBLT is simpler and less prone to errors.

To test after each BitBLT operation, perform the following.

1. Program and start the BitBLT engine.
2. Wait for the current BitBLT operation to finish -- Poll the BitBLT Busy Status bit (REG[8004h] bit 0) until it returns a 0.
3. Continue with program execution.

Testing the BitBLT Status Before

Testing the BitBLT Active Status before starting a new BitBLT results in better performance, as both CPU and BitBLT engine can be running at the same time. This is most useful for BitBLTs that are self completing (once started they don't require any CPU assistance). While the BitBLT engine is busy, the CPU can do other tasks. To test before each BitBLT operation, perform the following.

1. Wait for the current BitBLT operation to finish -- Poll the BitBLT Busy Status bit (REG[8004h] bit 0) until it returns a 0.
2. Program and start the new BitBLT operation.
3. Continue with program execution (CPU and BitBLT engine work independently).

This approach can pose problems when mixing CPU and BitBLT access to the display buffer. For example, if the CPU writes a pixel while the BitBLT engine is running and the CPU writes a pixel before the BitBLT finishes, the pixel may be overwritten by the BitBLT. To avoid this scenario, always assure no BitBLT is in progress before accessing the display buffer with the CPU, or don't use the CPU to access the display buffer at all.

9.4 S1D13A04 BitBLT Known Limitations

The S1D13A04 BitBLT engine has the following limitations.

- The 2D Accelerator Data Memory Mapped register must not be accessed except during BitBLT operations. Read from the register only during Read BitBLT operations and write to the register only during Write and Color Expand BitBLTs. Accessing the register at any other time may result in S1D13A04 stopping to respond and the system to freeze.
- The Read and Write BitBLT operations are not available when the S1D13A04 is configured for the Redcap or Dragonball without DTACK host bus interfaces.
- A BitBLT operation cannot be terminated once it has been started.

9.5 Sample Code

Sample code demonstrating how to program the S1D13A04 BitBLT engine is provided in the file **A04sample.zip**. This file is available on the internet at www.erd.epson.com.

10 Programming the USB Controller

USB (Universal Serial Bus) is an external bus designed to ease the connection and use of peripheral devices. USB incorporates a host/client architecture in which the host initiates all data transactions and the client either receives or supplies data to the host.

USB offers the following features to the end user:

- Single plug type for all peripheral devices.
- Support for up to 127 simultaneous devices.
- Speeds up to 12 Megabits per second.
- “hot-plugging” peripherals.

The S1D13A04 USB controller supports revision 1.1 of the USB specification. The S1D13A04 USB controller handles many common USB tasks without requiring local processor intervention. For example, setup and data transfers are handled automatically by the S1D13A04 controller. The controller notifies the local CPU, through an interrupt, when data is ready to be read from the FIFO or when data has been transmitted to the host.

This section demonstrates how to program and use the S1D13A04 USB controller. Topics covered include:

- Basic concepts such as registers and interrupts
- Initialization and data transfers
- S1D13A04 USB known issues.

10.1 Registers and Interrupts

10.1.1 Registers

Configuration, interrupt notification, and data transfers are all done using the S1D13A04 USB registers. The USB registers are located 4000h bytes past the beginning of S1D13A04 address space and should be written/read using 16 bit accesses.

On most systems the start of S1D13A04 address space, is fixed by the system design. The S1D13A04 evaluation board uses a PCI interface, thus the start of S1D13A04 address space may vary from one session to the next. Example code is written using a pointer to the USB registers (pUSB). The USB examples do not show how to obtain the register address. For a description of how to get the register address when using the S5U13A04B00C evaluation board, refer to the function `halAcquireController()` in Section 11, “Hardware Abstraction Layer” on page 112.

10.1.2 Interrupts

The S1D13A04 uses an interrupt to notify the local CPU when a USB event, which requires servicing, occurs. Events, such as USB reset and data transfer notifications generate interrupts.

It is beyond the scope of this document to explain how to setup and configure the interrupt system for the variety of platforms the S1D13A04 supports. The examples and flowcharts assume there is one interrupt handling routine which will determine the cause of the interrupt and call the appropriate handler function. It is assumed the user understands the mechanics and architecture of their system well enough setup a routine which will receive an interrupt notification and determine the cause of the interrupt.

10.2 Initialization

Initialization describes the process of setting the registers state to enable the USB controller for use. There are two cases where the USB registers need to be initialized. When the system is powered up and the registers need to be prepared for first use. The second time the registers need to be initialized is after receiving a RESET request from the host controller.

Refer to Section 10.2.2, “USB Registers” on page 96 for an example of the register initialization sequence.

10.2.1 GPIO Setup

The S1D13A04 shares four lines between GPIO and USB use. Before **any** accesses are made to the USB section the GPIO lines **must** be configured. To set the GPIO lines write the binary value 0010xxxx-1101xxxx-00000000-xxxxxxxx (2xDx00xxh) to REG[64h], the GPIO Status and Control register.

Note

X's represent a don't care state. Depending on other system configuration (i.e. panel technology) certain don't care bits may have to be set also. See the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001, for more information regarding the bits in the GPIO Status and Control register.

10.2.2 USB Registers

The steps described below are typical of the startup of the S1D13A04 USB controller.

- registers are set to an initial value
- the S1D13A04 is connected to a USB host controller
- the host controller issues a RESET command
- the USB registers are re-initialized

As initialization for both steps are similar it is recommended that one routine perform the sequence. The following table depicts a typical register initialization sequence.

Table 10-1: USB Controller Initialization Sequence

Register		Value (hex)	Notes
REG[4040h]	USBFC INPUT CONTROL	40	Enable the USB differential input receiver and indicate we are a bulk transfer self powered device. (for ISOchronous mode, use 43h)
REG[4044]	PIN IO STATUS DATA	01	USBPUP must be set to enable the USB interface and registers. REG[4000h] to REG[403Ah] cannot be written until this bit is set.
REG[4000]	CONTROL	84	Enable the clocks and USB GPIO pins.
REG[4024]	EP3 RECEIVE FIFO STATUS	1C	Clear EP3 status.
REG[402C]	USB EP4 TX FIFO STATUS	1C	Clear EP4 status
REG[4032]	USB STATUS	7E	Clear EP2 valid bit
REG[4004]	INTERRUPT STATUS 0	FF	Clear any pending USB interrupts
REG[4010]	EP1 INDEX	00	Set EP1 index to zero
REG[4018]	EP2 INDEX	00	Set EP2 index to zero
ext REG[00]	VENDOR ID MSB	??	Provide appropriate vendor ID
ext REG[01]	VENDOR ID LSB	??	
ext REG[02]	PRODUCT ID MSB	??	Provide appropriate product ID
ext Reg[03]	PRODUCT ID LSB	??	
ext REG[0C]	FIFO CONTROL	01	Enable EP4 (FIFO) valid transfer mode.
REG[4002]	INT ENABLE 0	0A	Enable interrupts for EP1 and EP3
REG[4004]	INT STATUS 0	0A	Make sure any pending interrupts are cleared.
REG[4046]	INTERRUPT CONTROL ENABLE 0	02	Enable RESET and endpoints notifications
REG[4048]	INTERRUPT CONTROL ENABLE 1	01	
REG[404A]	INTERRUPT CONTROL STATUS/CLEAR 0	7F	Clear ALL interrupt status...
REG[404C]	INTERRUPT CONTROL STATUS/CLEAR 1	7F	
REG[4000]	CONTROL	A4	Enable the USB port for use

The USB controller is ready for operation with the following configuration:

- Endpoint 1 (mailbox receive) is configured for bulk OUT and Endpoint 2 (mailbox transmit) is configured for interrupt IN. The functionality of these endpoints cannot be altered.
- Endpoint 3 (FIFO receive) is configured for bulk in and Endpoint 4 (FIFO transmit) is configured for bulk out. Endpoints 3 and 4 may also be configured for isochronous operation.

When the S1D13A04 is connected to a host controller, the host will issue a RESET command to the S1D13A04. In response to the RESET the S1D13A04 clears all USB registers in the range REG[4000h] to REG[403Ah]. The client software must respond to the reset and reprogram the USB registers. A host controller may issue a RESET at any time during operation.

After the S1D13A04 receives the RESET and re-initializes the registers, the host controller starts the USB SETUP phase. The SETUP sequence is handled entirely by the S1D13A04 USB controller. After the setup is complete the S1D13A3 is ready to begin transferring data.

Note

Prior to initializing the registers, host controller accesses are responded to with NAKs. After being configured, host controller accesses will be handled in the normal way.

Note

A Vendor ID can be obtained through the USB Implementers Forum at <http://www.usb.org>.

10.3 Data Transfers

The S1D13A04 USB requires very little local CPU assistance during data transfers. For the most part data transfers from the host involve reading a FIFO data register when notified of that the transfer is complete or writing a FIFO register and setting a 'ready' bit to send data to the host.

The following sections expand on the data transfer mechanism.

10.3.1 Receiving Data from the Host - the OUT command

Data transferred from the host to the S1D13A04 is directed to either EndPoint 1 (the mailbox) or EndPoint 3 (the FIFO). When the data packet has been successfully received the S1D13A04 generates an interrupt.

On receipt of the interrupt the local CPU examines the masked interrupt status registers REG[404Eh] and REG[4050h] to determine the source of the interrupt. If the interrupt came from bit 0 of the Negative Interrupt Masked Status register, REG[4050h], the next step is to examine REG[4004] to determine the exact cause of the interrupt.

Endpoint 1 - Mailbox Receive

If the cause of the interrupt is determined to be EndPoint 1 (REG[4004h] bit 1 = 1), then the data is read from the EndPoint 1 data register (REG[4012h]). The following figure shows the procedure for the CPU to read the mailbox register.

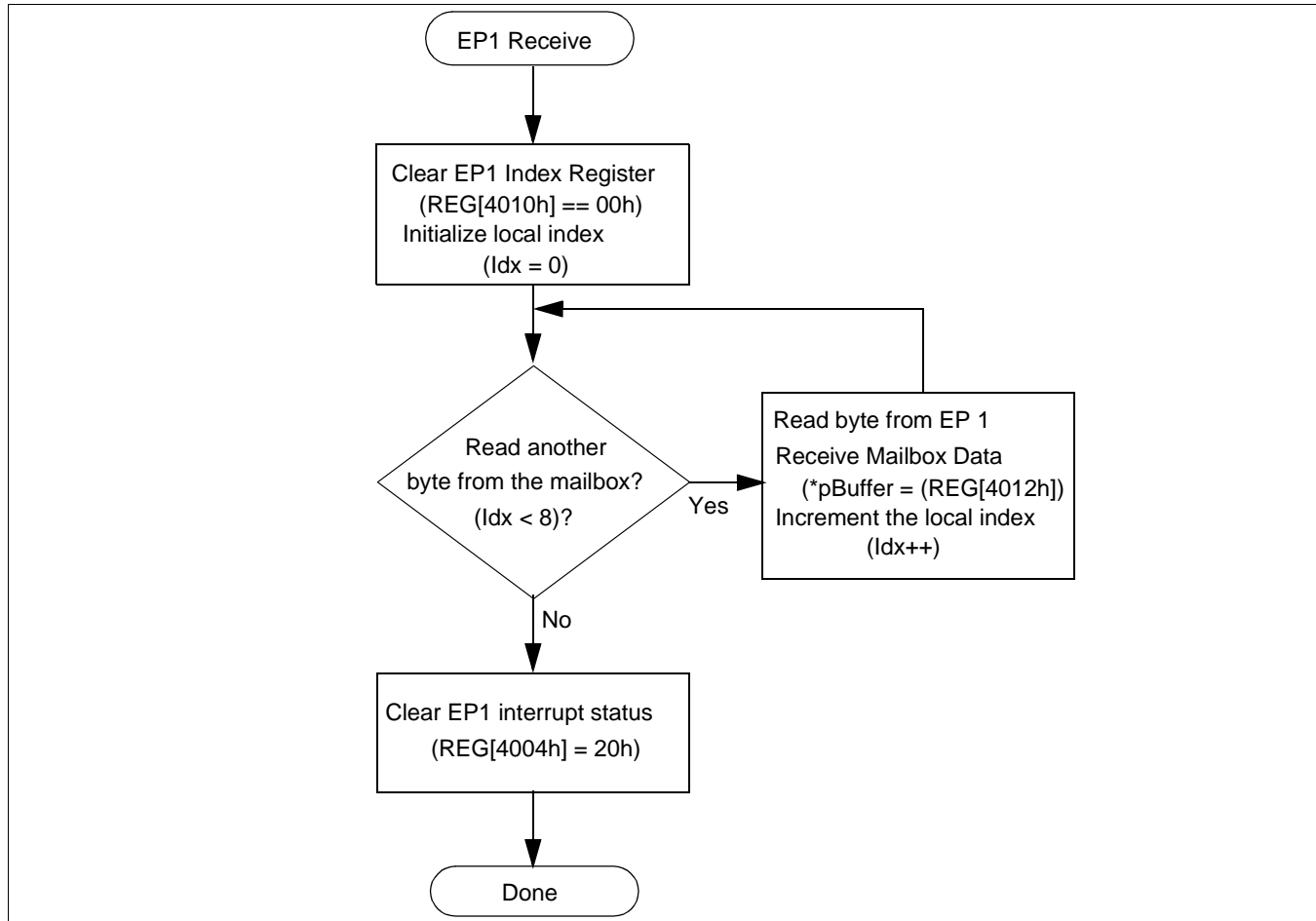


Figure 10-1: Endpoint 1 Data Reception

Note

In this diagram reference is made to two pseudo-variables:

Idx is an integer used as a loop counter

pBuffer is a pointer to eight bytes of memory to store the EP1 data

Endpoint 3 - FIFO Receive

If the cause of the interrupt is determined to be EndPoint 3, REG[4004h] bit 3 = 1b, then the host controller has sent data to EndPoint 3. Figure 10-2: shows the procedure for reading data from EndPoint 3.

An EndPoint 3 interrupt is generated when the number of bytes in the receive FIFO equal the value in the Receive FIFO Almost Full Threshold register (REG[403Ah], Index[06h]). The default value is sixty bytes. On systems where bulk transfers are used, the default value for the receive FIFO threshold should be satisfactory.

Systems with slow processors, high interrupt service latency, or configured for isochronous operation may have to decrease this value to allow the CPU time to begin reading data before the data transfer overflows the FIFO.

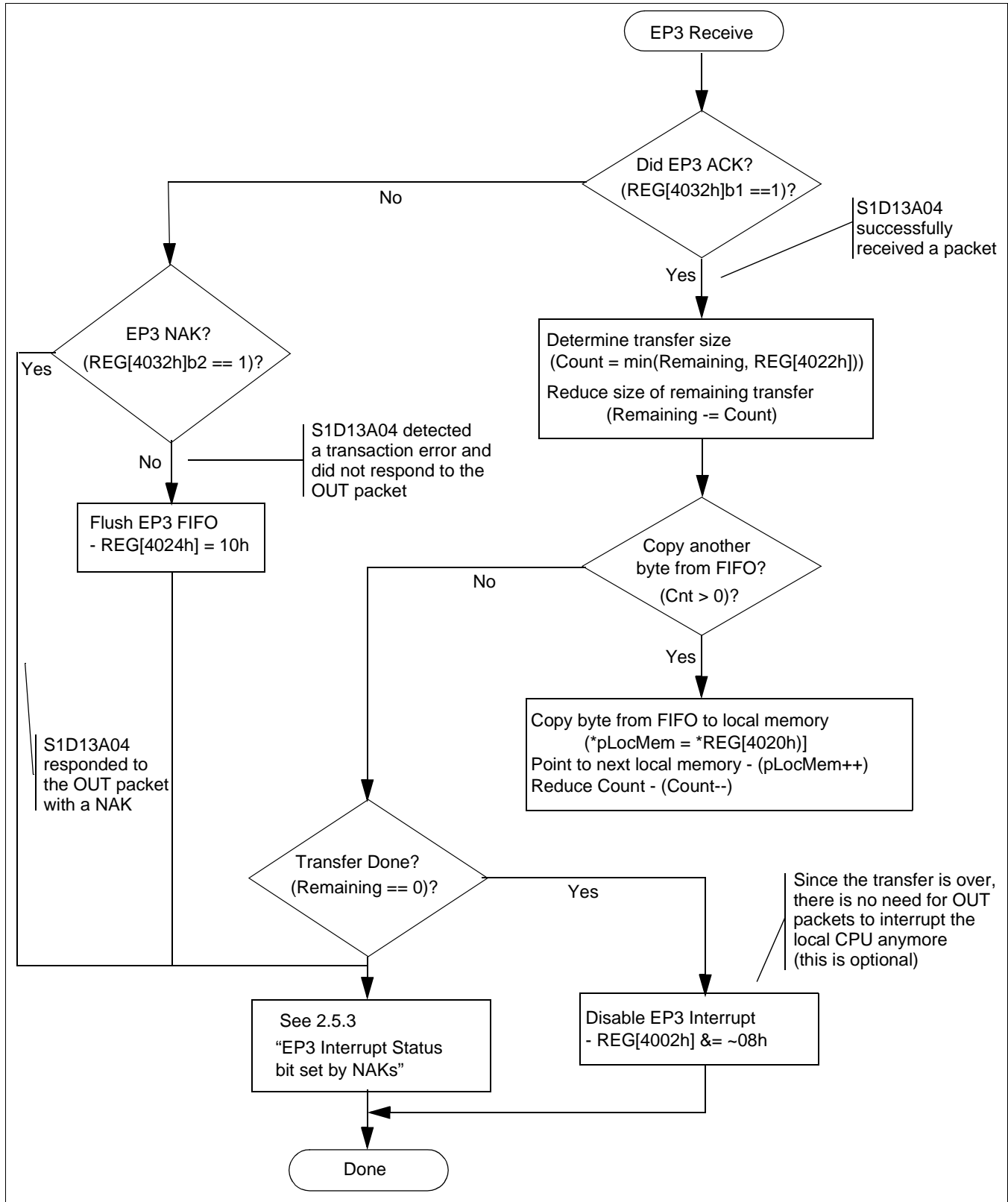


Figure 10-2: Endpoint 3 Data Reception

10.3.2 Sending Data to the Host - the IN command

Data transfers to the host controller occur when the host issues an IN command. The data comes from EndPoint 2 (the mailbox) or EndPoint 4 (the FIFO). The data transfer is handled automatically by the S1D13A04 and requires no CPU assistance.

Data transfers, from the S1D13A04 to the host controller, are performed by writing the data into either EndPoint 2 (mailbox) or EndPoint 4 (FIFO) data registers. After writing the data to the registers a control bit indicating that mailbox or FIFO data is valid is set.

Endpoint 2 - Mailbox Transmit

Figure 10-3: shows the logical flow for sending data to the host controller using EndPoint 2, the mailbox.

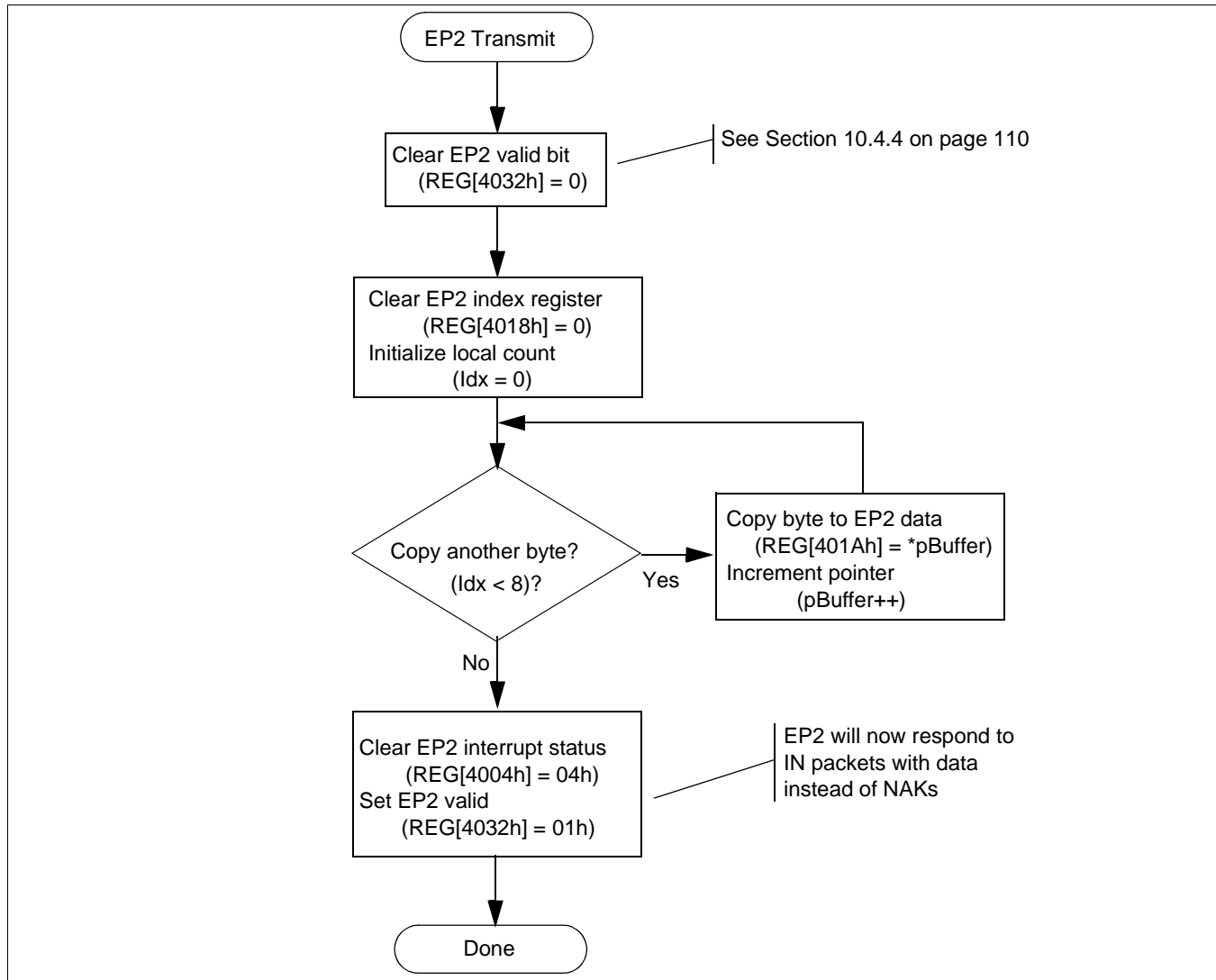


Figure 10-3: EndPoint 2 Data Transmission

Note

In this diagram reference is made to two pseudo-variables:

Idx is an integer used as a loop counter

pBuffer is a pointer to eight bytes of memory to send to the host

Endpoint 4 - Data Transmit

Transferring data to the host controller using the FIFO controller has additional overhead as this routine must run tests to ensure error free data transmission.

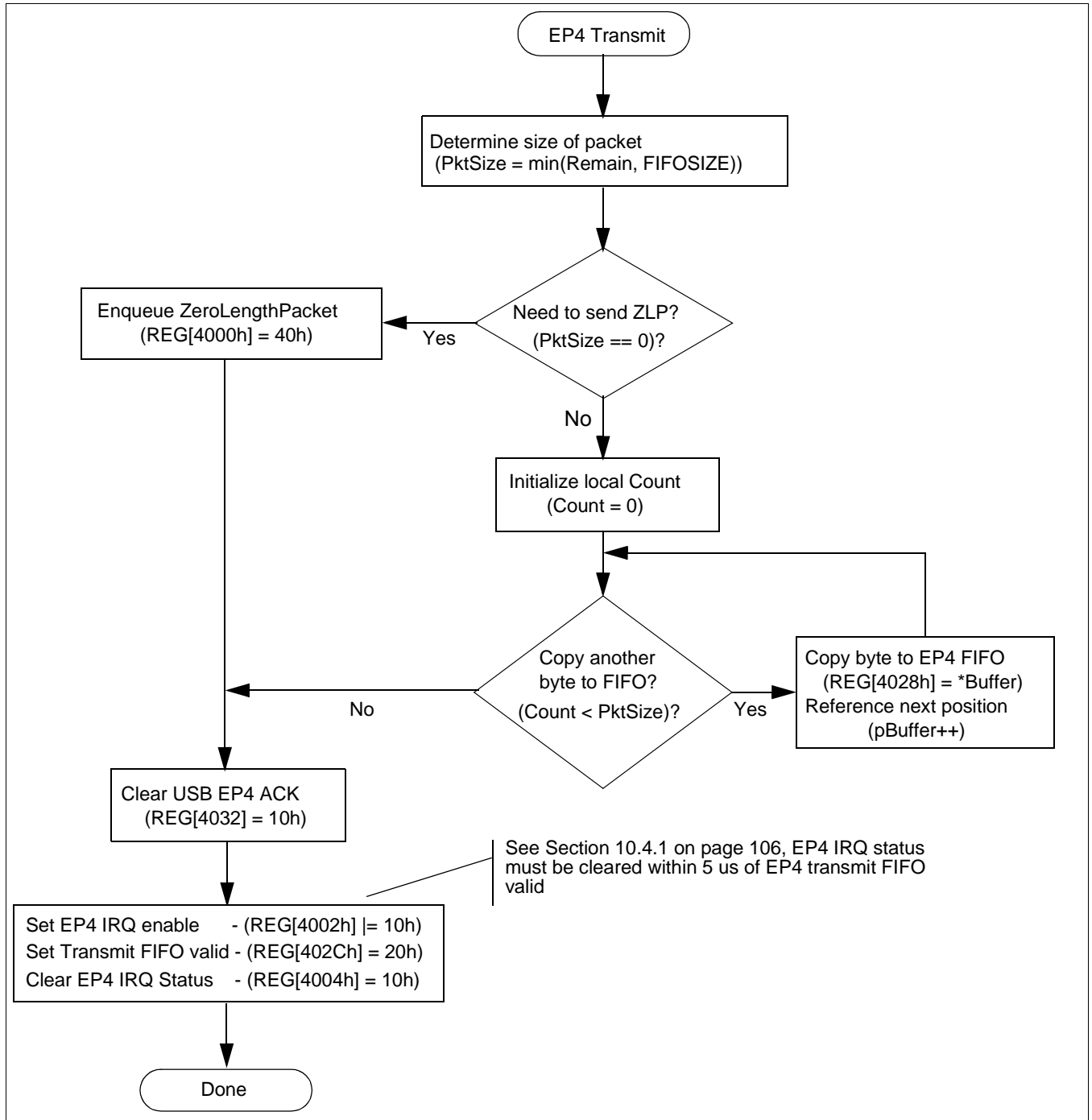


Figure 10-4: Endpoint 4 Data Transmission

Note

In this example there are three variables:
 PktSize is an integer containing the number of bytes to transfer in this packet
 Count is an integer used for local loop control
 pBuffer is a pointer to an array of at least FIFOSIZE bytes.

To ensure the host controller receives the packet error free, an interrupt handler for EndPoint 4 must be configured and the flow control as shown in the following diagram must be implemented.

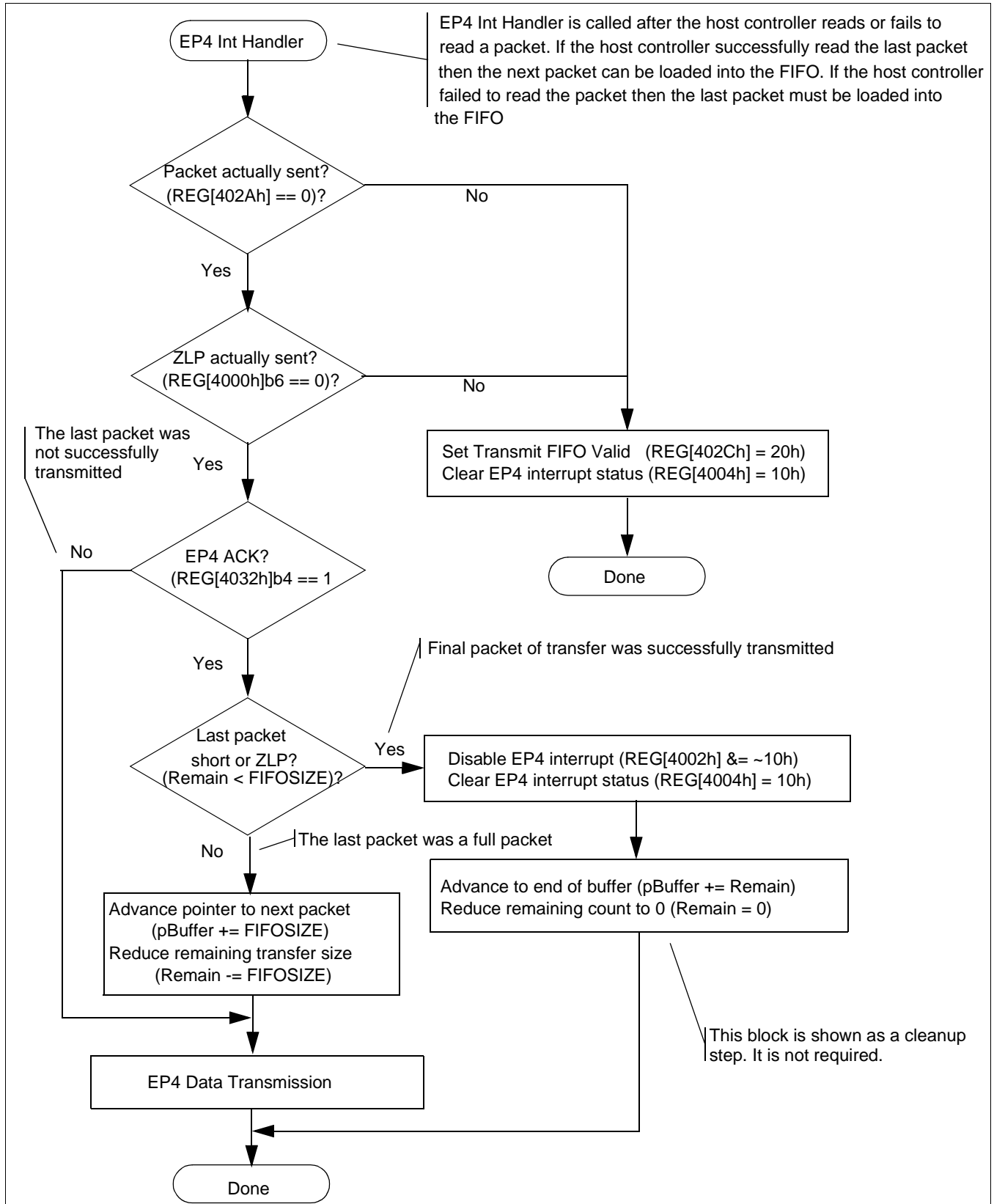


Figure 10-5: Endpoint 4 Interrupt Handling

Note

In the diagram the variables:

pBuffer is a pointer to the local memory buffer containing the data to be transferred to the host controller

Remain is an integer tracking the number of bytes still to be sent.

10.4 Known Issues

This section presents known issues with USB transfers when using the S1D13A04 USB controller.

10.4.1 EP4 NAK Status not set correctly in USB Status Register

The EP4 NAK status bit is not set in the USB Status Register (REG4032h) when the S1D13A04 responds to an IN request on EP4 with a NAK. As a result, a local CPU receiving an “EP4 Packet Transmitted” interrupt may mistakenly believe a bus error occurred in the most recently transmitted packet.

Work Around

Disable the EP4 Packet Transmitted interrupt when no data is queued for transmission to the local CPU. The basic flow is:

In Chip Initialization Code

Do not enable ‘EP4 Packet Transmitted’ bit in Interrupt Enable Register 0 (REG[4002h]).

When Local Side Wishes to Send Data

1. Put data to transmit in FIFO.
2. Enable ‘EP4 Packet Transmitted’ bit in Interrupt Enable Register 0.
3. Set FIFO Valid (if using FIFO Valid Mode == TRUE). See Section 10.4.2 on page 107 for more information on setting the FIFO Valid.
4. Clear ‘EP4 Packet Transmitted’ status bit in Interrupt Status Register 0 (REG[4004]).

Note

Step 4 is time-critical. It must be performed within 5 μ s after Step 3.

In Packet Transmitted Interrupt Routine

Disable ‘EP4 Packet Transmitted’ bit in Interrupt Enable Register 0.

10.4.2 Write to EP4 FIFO Valid bit cleared by NAK

After the local CPU sets EP4 FIFO Valid (in Endpoint 4 FIFO Status Register, REG[402Ch]), the S1D13A04 will erroneously clear the EP4 valid bit if the S1D13A04 is concurrently sending a NAK handshake in response to a previous IN token to EP4.

Work Around

The work-around is in the 'EP4 Packet Transmitted' interrupt routine. It requires the interrupt routine to know whether the recently queued packet was a zero-length packet or not, so that must be stored as a bit when the packet was loaded into the FIFO. On entry to the 'EP4 Packet Transmitted' interrupt routine:

For a non-zero-length Packet

Check the FIFO count. If it is non-zero, this error occurred. In that case, set FIFO Valid again, clear the interrupt status bit, and exit the interrupt routine.

For a zero-length Packet

Check the Software EOT bit (in Control Register, REG[4000h]). If it is set, the FIFO Valid write failed. In that case, set FIFO Valid again, clear the interrupt status bit, and exit the interrupt routine

10.4.3 EP3 Interrupt Status bit set by NAKs

When receiving Bulk OUT packets from a Host PC, the S1D13A04 "Endpoint 3 Interrupt Status" interrupt typically is used to notify the peripheral firmware that a packet has been received. This bit also serves as the "Receive FIFO Valid" bit, so additional packets addressed to Endpoint 3 are NAKed until this status bit is cleared. Once cleared, however, it may become set by another packet which is NAKed by the S1D13A04, causing the Receive FIFO to become "Valid" again. The Host PC may immediately attempt to re-transmit the NAKed packet. The firmware should be written to prevent a cycle in which the FIFO is "Valid" each time that the Host PC sends an OUT packet.

The following rules govern the S1D13A04's behavior regarding packets received on Endpoint 3:

Rule A. At the end of a received OUT token to EP3 (and before the data is received), the S1D13A04 decides to NAK the packet if the "EP3 Interrupt Status" bit is set, and will therefore throw away data received.

Rule B. At the end of a received packet (including one which is NAKed), the S1D13A04 sets the "EP3 Interrupt Status" bit.

Rule C. Local firmware should clear the "EP3 Interrupt Status" bit after reading all bytes out of the EP3 Receive FIFO.

The following figure shows how a repeating cycle of NAKed OUT packets may occur.

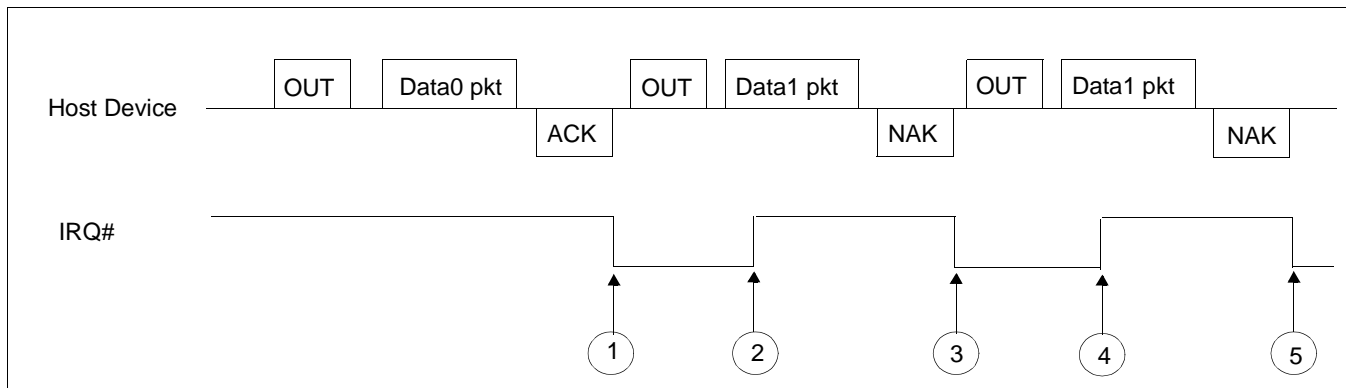


Figure 10-6: Firmware Looping Continuously on Received OUT packets

At Point 1, the EP3 Interrupt activates because a packet has been received. In response, the firmware reads the bytes out of the packet and clears the interrupt at Point 2. A second packet is already being received at Point 2, and the S1D13A04 has already decided to NAK this packet due to Rule A. At point 3, the S1D13A04 has NAKed the packet and asserts the Interrupt status bit.

Again, the local firmware responds to the interrupt, and seeing it is only a “NAK” interrupt, clears the interrupt condition at Point 4. However, the Host PC has begun to retry the second packet already, so the packet will again get NAKed due to Rule B. This cycle could continue until something changes the flow of OUT packets – for instance, an SOF at the beginning of the next frame, or packet traffic directed at another device or endpoint.

Work Around

The normal program flow for a packet which the S1D13A04 NAKs is as follows:

1. S1D13A04 asserts IRQ# after NAKing a received packet on EP3.
2. Local CPU is interrupted, enters interrupt routine.
3. Local CPU reads Interrupt Status Register 0 (REG[4004h]) and sees “EP3 Packet Received” interrupt bit.
4. Local CPU reads USB Status Register (REG[4032h]) and sees “NAK” bit set.
5. Local CPU clears Interrupt Status Register 0 (REG[4004h]) “EP3 Packet Status” interrupt bit.
6. Local CPU clears USB Status Register (REG[4032]) “NAK” bit.

The technique for avoiding this potential pitfall depends on the speed of the peripheral CPU. The critical timing parameter is the time from the S1D13A04 asserting IRQ# to the firmware clearing the “EP3 Packet Received” bit in Interrupt Status Register 0.

For a Fast CPU

A CPU which can clear the Interrupt Status Register 0 bit within 10 msec after the S1D13A04 asserts the IRQ# signal requires no extra code to prevent the potential cycling. In this case, the CPU is fast enough to clear a NAKed packet's Interrupt Status Register 0 bit before another packet can be received.

For a Slow CPU

A CPU which can't meet the timing requirements for a fast CPU above will require some additional firmware to eliminate the potential for this cycle. After successfully receiving a packet on Endpoint 3 and emptying received data out of the FIFO, the firmware should follow the flow in the following figure.

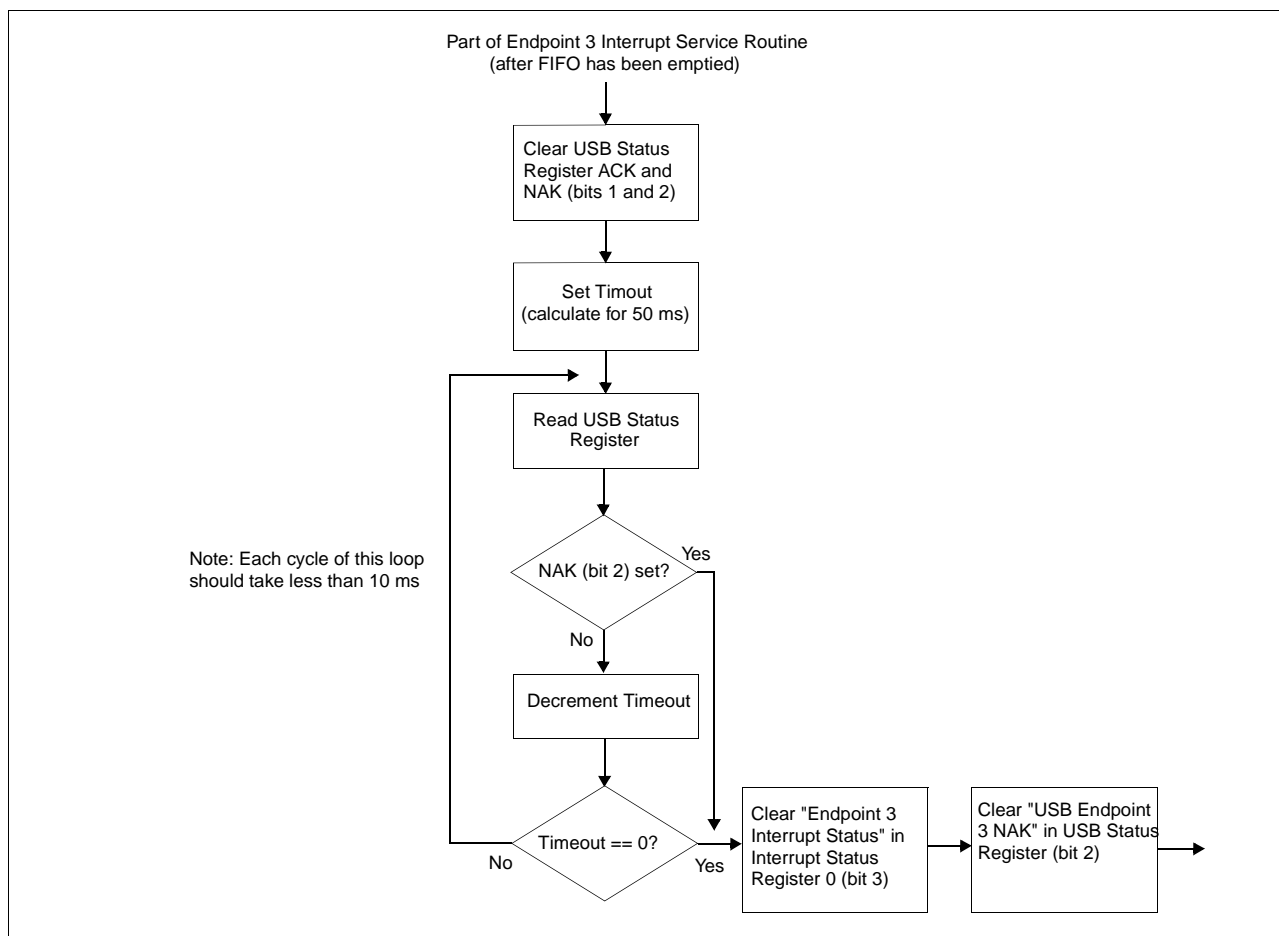


Figure 10-7: Endpoint 3 Program Flow for Slow CPU

10.4.4 “EP2 Valid Bit” in USB Status can be erroneously set by firmware

“Endpoint 2 Valid” is the only bit in USB Status which is not written as a “Yes/CLR” bit. Therefore, the firmware must do a read-modify-write sequence when clearing other bits in Interrupt Status Register 0 (REG[4004h]), to preserve the state of “Endpoint 2 Valid”. However, this read-modify-write could lead to erroneously setting the EP2 Valid bit if the following sequence occurs with “EP2 Valid” set True:

1. Firmware reads Interrupt Status Register 0 to do a read-modify-write
2. Data from EP2 is sent to Host PC, causing S1D13A04 to clear EP2 Valid
3. Firmware writes modified value to Interrupt Status Register 0

In this case, the firmware has set EP2 Valid in Step 3 after it was cleared by the Host PC, erroneously validating EP2 for the next IN token from the Host.

Work Around

First, the firmware should do the read-modify-write operation as described above anytime it is modifying bits in “USB Status”.

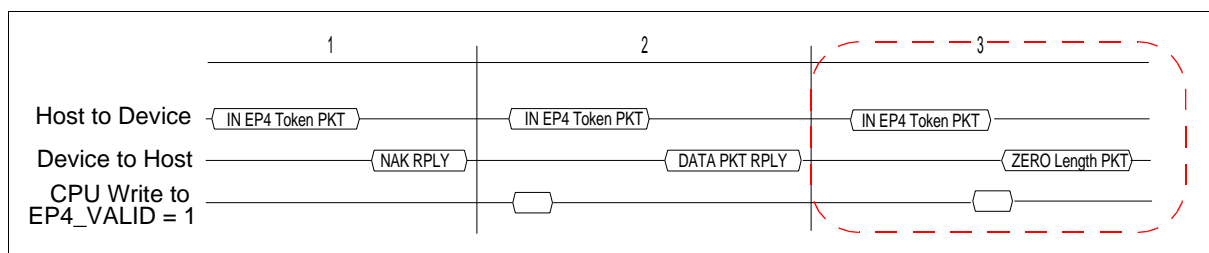
Second, when the firmware recognizes an interrupt for “EP2 Packet Transmitted”, it should immediately write a ‘0’ to USB Status Register. This will clear the EP2 Valid bit in the unlikely event that it was erroneously set during a read-modify-write operation.

10.4.5 Setting EP4 FIFO Valid bit while NAKing IN token

Bit 5 of REG[402Ch] indicates to the S1D13A04 controller when data in the endpoint 4 FIFO is ready to be transferred to the host computer. Changing the state of this bit at certain times may generate an error.

When the S1D13A04 USB controller receives an endpoint 4 IN request and endpoint 4 is not ready to transmit data (REG[402Ch] bit 5 = 0), the response is a NAK packet. If endpoint 4 is toggled to a ready to transmit state just before a NAK response packet is sent, the controller may erroneously send a zero length packet instead. When this happens, the data toggle state will be incorrectly set for the next endpoint 4 data transmit.

The following timing diagram shows the error occurring in section 3.



This unexpected occurrence of a zero length packet may cause file system handling errors for some operating systems.

Work Around

There are two software solutions for this occurrence.

Disable USB Receiver before setting the EP4 FIFO Valid bit

The first solution involves disabling the USB receiver to avoid responding to an EP4 IN packet. During the time the USB receiver is disabled the EP4 FIFO Valid bit is set.

When the local CPU is ready to send data on endpoint 4 the steps to follow are:

2. Disable the USB differential input receiver (REG[4040h] bit 6 = 0)
3. Wait a minimum of 1 μ s. If needed, delays may be added
4. Enable the EP4 FIFO Valid bit (REG[402Ch] bit 5 = 1)
5. Clear the EP4 Interrupt status bit (REG[4004h] bit 4 = 1)
6. Enable the USB differential input receiver (REG[4040h] bit 6 = 1)

Note

Steps 1 through 5 are time critical and must be performed in less than 6 μ s.

Note

To comply with “EP4 NAK Status not set correctly in USB Status register”, steps 3 and 4 must be completed within 5 μ s of each other. For further information on “EP4 NAK Status not set correctly in USB Status register”, see Section 10.4.1, “EP4 NAK Status not set correctly in USB Status Register” .

EP4 FIFO Valid bit set after NAK and before the next IN token

The second solution is to wait until immediately after the USB has responded to an IN request with a NAK packet before setting the transmit FIFO valid bit. This solution is recommended only for fast processors.

When the local CPU is ready to send data on endpoint 4, it must first detect that a NAK packet has been sent. This is done by reading the EP4 Interrupt Status bit (REG[4004h] bit 4). If the EP4 FIFO Valid bit was not set, the EP4 Interrupt Status bit is set only if a NAK packet has been sent. When the local CPU detects the NAK it must immediately set the EP4 FIFO Valid bit (before responding to the next IN token).

After filling the EP4 FIFO the steps to follow before setting the EP4 FIFO Valid bit are:

1. Clear the EP4 Interrupt Status bit (REG[4004h] bit 4)
2. Read the EP4 Interrupt Status bit (REG[4004h] bit 4) until it is set
3. Set the EP4 FIFO Valid bit (REG[402Ch] bit 5 = 1)

The setting of the EP4 FIFO Valid bit is time critical. The EP4 FIFO Valid bit must be set within 3 μ s after the EP4 Interrupt Status has been set internally by the S1D13A04.

11 Hardware Abstraction Layer

11.1 Introduction

The S1D13A04 Hardware Abstraction Layer (HAL) is a collection of routines intended to simplify the programming for the S5U13A04B00C evaluation board. Programmers can use the HAL to assist in rapid software prototyping for the S5U13A04B00C evaluation board.

The HAL routines are divided into discrete functional blocks. The functions for startup and clock control offer specific support for the S5U13A04B00C evaluation board, while other routines demonstrate memory and register access techniques. For a complete list, see Table 11-1; “HAL Library API” .

11.2 API for the HAL Library

The following table lists the functions provided by the S1D13A04 HAL library.

Table 11-1: HAL Library API

Function	Description
Startup	
halAcquireController	This routine loads the driver required to access the S1D13A04, locates the and returns the address of the controller.
halInitController	Initializes the controller for use. This includes setting the programmable clock and initializing registers as well as setting the lookup table and clearing video memory.
Memory Access	
halReadDisplay8	Reads one byte from display memory
halReadDisplay16	Reads one word from display memory
halReadDisplay32	Reads one double word from display memory
halWriteDisplay8	Writes one byte to display memory
halWriteDisplay16	Writes on word to display memory
halWriteDisplay32	Writes on double word to display memory
Register Access	
halReadReg8	Reads one byte from a control register
halReadReg16	Reads one word from a control register
halReadReg32	Reads one dword from a control register
halWriteReg8	Writes one byte to a control register
halWriteReg16	Writes one word to a control register
halWriteReg32	Writes one dword to a control registers
Clock Support	
halSetClock	Programs the ICD2061A Programmable Clock Generator.
halGetClock	Returns the frequency of the requested ICD2061A clock
Miscellaneous	
halGetVersionInfo	Returns a standardized startup banner message
halGetLastError	Returns the numerical value of the last error and optionally an ASCII string describing the error
halInitLUT	This routine sets the LUT to uniform values for color/mono panels at all color depths

11.2.1 Startup Routines

There are two routines dedicated to startup and initializing the S1D13A04. Typically these two functions are the first two HAL routines a program will call.

The startup routines locate the S1D13A04 controller and initialize HAL data structures. As the name suggests, the initialization routine prepares the S1D13A04 for use. Splitting the startup functionality allows programs to start and locate the S1D13A04 but delay or possibly never initialize the controller.

Boolean `halAcquireController(UInt32 * pMem, UInt32 * pReg)`

Description: This routine initializes data structures and initiates the link between the application software and the hardware. When the S1D13A04 HAL is used this routine must be the first HAL function called.

On PCI platforms, the routine attempts to load the S1D13xxx driver. If the driver loads successfully then a check is made for the existence of an S1D13A04.

Parameters: `pMem` Pointer to an unsigned 32-bit integer which will receive the offset to the first byte of display memory. The offset may be cast to a pointer to access display memory.

`pReg` Pointer to an unsigned 32-bit integer which will receive the offset to the first byte of register space. The offset may be cast to a pointer and to access S1D13A04 registers.

On Win32 systems the returned offsets correspond to a linear addresses within the callers address space.

Return Value: `TRUE` (non-zero) if the routine is able to locate an S1D13A04.
`pMem` will contain the offset to the first byte of display memory.
`pRegs` will contain the address of the first 13A04 control register.

`FALSE` (zero) if an S1D13A04 is not located.
`pMem` and `pRegs` will be undefined.
If additional error information is required call `halGetLastError()`.

Note

1. This routine **must** be called before any other HAL routine is called.
2. For programs written for the S1D13A04 evaluation board, an application may call this routine to obtain pointers to the registers and display memory and then perform all S1D13A04 accesses directly.
3. This routine does not modify S1D13A04 registers or memory.

Boolean halInitController(UInt32 Flags)

Description:

This routine performs the initialization portion of the startup sequence. Initialization of the S1D13A04 evaluation board consists of several steps:

- Program the ICD2061A clock generator
- Set the initial state of the control
- Set the LUT to its default value
- Clear video memory

All display memory and nearly every control register can or will be affected by the initialization.

Any, or all, of the initialization steps may be bypassed according to values contained in the Flags parameter. This allows for conditional run-time changes to the initialization.

Parameters:

Flags contains initialization specific information. The default action of the HAL is to perform all initialization steps. Flags contain specific instructions for bypassing certain initialization steps. The values for Flags are:

fDONT_RESET

The first step of the initialization process is to perform a software reset. Setting this flag bypasses the software reset.

fDONT_SET_CLOCKS

Setting this flag causes initialization to skip programming the ICD2061A clock generator. Normally the clock on the S5U13A04B00C is programmed to configured values during initialization.

fDONT_INIT_REGS

Bypass register initialization. Normally the init process sets the register values to a known state. Setting this flag bypasses this step.

fDONT_INIT_LUT

Bypass look-up table initialization.

fDONT_CLEAR_MEM

The final step of the initialization process is to clear video display memory. Setting this flag will bypass this step.

Return Value:

TRUE (non-zero) if the initialization was successful.
FALSE (zero) if the HAL was unable to initialize the S1D13A04
If additional error information is required call halGetLastError()

11.2.2 Memory Access

The S1D13A04 HAL includes six memory access functions. The primary purpose of the memory access functions is to demonstrate how to access display memory using the C programming language. Most programs that need to access memory will bypass the HAL and access memory directly.

UInt8 halReadDisplay8(UInt32 Offset)

Description: Reads and returns the value of one byte of display memory.
Parameters: Offset A 32 bit offset to the byte to be read from display memory
Return Value: The value of the byte at the requested offset.

UInt16 halReadDisplay16(UInt32 Offset)

Description: Reads and returns the value of one word of display memory.
Parameters: Offset A 32 bit byte offset to the word to be read from display memory
To prevent system slowdowns and possibly memory faults, Offset should be a word multiple.
Return Value: The value of the word at the requested offset.

UInt32 halReadDisplay32(UInt32 Offset)

Description: Reads and returns the value of one dword of display memory.
Parameters: Offset A 32 bit byte offset to the dword to be read from display memory.
To prevent system slowdowns and possibly memory faults, Offset should be a dword multiple.
Return Value: The value of the dword at the requested offset.

void halWriteDisplay8(UInt32 Offset, UInt8 Value, UInt32 Count)

Description: Writes a byte into display memory at the requested address.
Parameters: Offset A 32 bit byte offset to the byte to be written to display memory.
Value The byte value to be written to display memory.
Count The number of times to repeat Value in memory. By including a count (or loop) value this function can efficiently fill display memory.
Return Value: Nothing.

void halWriteDisplay16(UInt32 Offset, UInt16 Value, UInt32 Count)**Description:** Writes a word into display memory at the requested offset.

Parameters:

Offset	a 32 bit byte offset to the byte to be written to display memory. To prevent system slowdowns and possibly memory faults, Offset should be a word multiple.
Value	the word value to be written to display memory.
Count	the number of times to repeat the Value in memory. By including a count (or loop) value this function can efficiently fill display memory.

Return Value: Nothing.**void halWriteDisplay32(UInt32 Offset, UInt32 Value, UInt32 Count)****Description:** Writes a dword into display memory at the requested offset.

Parameters:

Offset	A 32 bit byte offset to the byte to be written to display memory. To prevent system slowdowns and possibly memory faults, Offset should be a dword multiple.
Value	The dword value to be written to display memory.
Count	The number of times to repeat the Value in memory. By including a count (or loop) value this function can efficiently fill display memory.

Return Value: Nothing.**11.2.3 Register Access**

The S1D13A04 HAL includes six register access functions. The primary purpose of the register access functions is to demonstrate how to access the S1D13A04 control registers using the C programming language. Most programs that need to access the registers will bypass the HAL and access the registers directly.

UInt8 halReadReg8(UInt32 Index)**Description:** Reads and returns the contents of one byte of an S1D13A04 register at the requested offset. No S1D13A04 registers are changed

Parameters:

Index	32 bit offset to the register to read. Index is zero based from the beginning of register address space. (e.g. if Index == 04h then the Memory Clock Configuration register will be read and if Index == 8000h then the BitBLT Control Register will be read)
-------	---

Return Value: The value read from the register.

Use caution in selecting the index and when interpreting values returned from halReadReg8() to ensure the correct meaning is given to the values. Changing between big endian and little endian will move relative register offsets.

UInt16 halReadReg16(UInt32 Index)

Description: Reads and returns the contents of one word of an S1D13A04 register at the requested offset. No S1D13A04 register are changed.

Parameters: Index 32 bit offset to the register to read. Index is zero based from the beginning of register address space. (e.g. if Index == 04h then the Memory Clock Configuration register will be read and if Index == 8000h then the BitBLT Control Register will be read)

Return Value: The word value read from the register.

Use caution in determining the index and interpreting the values returned from halReadReg16() to ensure the correct meaning is given to the values. Changing between big and little endian will move relative register offsets resulting in different values.

UInt16 halReadReg32(UInt32 Index)

Description: Reads and returns the dword value of an S1D13A04 register at the requested offset. No S1D13A04 register are changed.

Parameters: Index 32 bit offset to the register to read. Index is zero based from the beginning of register address space. (e.g. if Index == 04h then the Memory Clock Configuration register will be read and if Index == 8000h then the BitBLT Control Register will be read)

Return Value: The dword value read from the register.

void halWriteReg8(UInt32 Index, UInt8 Value)

Description: Writes an 8 bit value to the register at the requested offset.

Parameters: Index 32 bit offset to the register to write. Index is zero based from the beginning of register address space. (e.g. if Index == 04h then the Memory Clock Configuration register will be written to and if Index == 8000h then the BitBLT Control Register will be written to)

Value The byte value to write to the register. Changing between big and little endian will move relative register offsets. Use caution in interpreting the index and values to write to registers using the halWriteReg8() function to ensure that register are programmed correctly.

Return Value: Nothing.

void halWriteReg16(UInt32 Index, UInt16 Value)

Description: Writes a 16 bit value to the S1D13A04 register at the requested offset.

Parameters:

Index	32 bit byte offset to the register to write. Index is zero based from the beginning of register address space. (e.g. if Index == 04h then the Memory Clock Configuration register will be written to and if Index == 8000h then the BitBLT Control Register will be written to)
Value	The word value to write to the register.

Return Value: Nothing.

Changing between big and little endian will move relative register offsets. Use caution in interpreting the index and values to write to registers using the halWriteReg8() function to ensure that register are programmed correctly.

void halWriteReg32(UInt32 Index, UInt32 Value)

Description: Writes a 32 bit value (dword) to the register at the requested offset.

Parameters:

Index	32 bit byte offset to the register to write. Index is zero based from the beginning of register address space. (e.g. if Index == 04h then the Memory Clock Configuration register will be written to and if Index == 8000h then the BitBLT Control Register will be written to)
Value	The dword value to write to the register.

Return Value: Nothing.

11.2.4 Clock Support

To maximize flexibility, S1D13A04 evaluation boards include a programmable clock. The following HAL routines provide support for the programmable clock.

Boolean halSetClock(UInt32 ClkiFreq, UInt32 Clki2Freq)

Description: This routine program the ICD2061A programmable clock generator to the specified frequency.

Parameters:

ClkiFreq	The desired frequency, in Hz, for CLKI.
Clki2Freq	The desired frequency, in Hz, for CLKI2.
dwFrequency	The desired frequency (in Hz).

Return Value: TRUE (non-zero) if the function was successful in setting the clock.

FALSE (zero) if there was an error detected while trying to set the clock.
If additional error information is required call halGetLastError().

UInt32 halGetClock(CLOCKSELECT Clock)

- Description:** Returns the frequency of the clock input identified by 'Clock'.
- Parameters:** Clock Indicates which clock to read. This value can be CLKI or CLKI2.
- Return Value:** The frequency, in Hz, of the requested clock.

11.2.5 Miscellaneous

The miscellaneous function are an assortment of routines, determined to be beneficial to a number of programs and hence warranted being included in the HAL.

void halGetVersionInfo(const char * szProgName, const char * szDesc, const char * szVersion, char * szRetStr, int nLength)

Description: This routine creates a standardized startup banner by merging program and HAL specific information. The newly formulated string is returned to the calling program for display.

The final formatted string will resemble:

```
13A04PROGRAM - Internal test and diagnostic program - Build: 1234 [HAL: 1234]  
Copyright (c) 2000,2001 Epson Research and Development, Inc.  
All Rights Reserved.
```

- Parameters:**
- szProgName Pointer to an ASCIIZ string containing the name of the program. (e.g. "PROGRAM")
 - szDesc Pointer to an ASCIIZ string containing a description of what this program is intended to do. (e.g. "Internal test and diagnostic program")
 - szVersion Pointer to an ASCIIZ string containing the build info for this program. This should be the revision info string as updated by VSS. (e.g. "\$Revision: 30 \$")
 - szRetStr Pointer to a buffer into which the product and version information will be formatted into.
 - nLength Total number of bytes in the string pointed to by szRetStr. This function will write nLength or fewer bytes to the buffer pointed to by szRetStr.
- Return Value:** Nothing.

int halGetLastError(char * ErrMsg, int MaxSize)

Description: This routine retrieves the last error detected by the HAL.

Parameters: ErrMsg When halGetLastError() returns ErrMsg will point to the textual error message. If ErrMsg is NULL then only the error code will be returned.

MaxSize Maximum number of bytes, including the final '\0' that can be placed in the string pointed to by ErrMsg.

Return Value: The numerical value of the internal error number.

HALEXTERN void hallnitLUT(void)

Description: To standardize the appearance of test and validation programs, it was decided the HAL would have the ability to set the lookup table to uniform values.

The routine cracks the color depth and display type to determine which LUT values to use and proceeds to write the LUT entries.

Parameters: None

Return Value: Nothing.

12 Sample Code

Example source code demonstrating programming the S1D13A04 using the HAL library is available on the internet at www.erd.epson.com.

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READ-ONLY CONFIGURATION REGISTERS

Product Information Register REG[00h] Default = 2Cx282Ch Read Only															
Product Code						Revision Code		n/a	CNF[6:4] Status				Reserved	CNF[2:0] Status	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
Display Buffer Size								Product Code						Revision Code	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

CLOCK CONFIGURATION REGISTERS

Memory Clock Configuration Register REG[04h] Default = 00000000h Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a										MCLK Divide Select		Reserved			
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Pixel Clock Configuration Register REG[08h] Default = 00000000h Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a								PCLK Divide Select				n/a		PCLK Source Select	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

PANEL CONFIGURATION REGISTERS

Panel Type & MOD Rate Register REG[0Ch] Default = 00000000h Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	MOD Rate		19	18	17	16
n/a								Panel Data Format Select	Color/Mono Panel Select	Panel Data Width		'Direct' HR-TFT Res Select	n/a	Panel Type	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Display Settings Register REG[10h] Default = 00000000h Read/Write															
n/a															
31	30	29	28	27	26	Pixel Doubling Vertical	Pixel Doubling Horiz.	Display Blank	Dithering Disable	n/a	SW Video Invert	PIP* Window Enable	n/a	Swivel/View Mode Select	
25	24	23	22	21	20	19	18	17	16	Bits-per-pixel Select (actual value: 1, 2, 4, 8, 16 bpp)					
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Power Save Configuration Register REG[14h] Default = 00000010h Read/Write															
n/a															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a								VNDP Status (RO)	Memory Power Save Status (RO)	n/a	Power Save Enable	n/a			'Direct' HR-TFT GPO Control
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

LOOK-UP TABLE REGISTERS

Look-Up Table Write Register REG[18h] Default = 00000000h Write Only															
LUT Write Address															
31	30	29	28	27	26	25	24	23	22	LUT Red Write Data				n/a	
LUT Green Write Data								n/a		LUT Blue Write Data				n/a	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Look-Up Table Read Register REG[1Ch] Default = 00000000h Write Only (bits 31-24)/Read Only															
LUT Read Address (write only)															
31	30	29	28	27	26	25	24	23	22	LUT Red Read Data				n/a	
LUT Green Read Data								n/a		LUT Blue Read Data				n/a	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

DISPLAY MODE REGISTERS

Horizontal Total Register REG[20h] Default = 00000000h Read/Write																	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
n/a									Horizontal Total bits 6-0								
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		
Horizontal Display Period Register REG[24h] Default = 00000000h Read/Write																	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
n/a									Horizontal Display Period bits 6-0								
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		
Horizontal Display Period Start Position Register REG[28h] Default = 00000000h Read/Write																	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
n/a						Horizontal Display Period Start Position bits 9-0											
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		
FPLINE Register REG[2Ch] Default = 00000000h Read/Write																	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
n/a								FPLINE Polarity	FPLINE Pulse Width bits 6-0								
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		
n/a									FPLINE Pulse Start Position bits 9-0								
Vertical Total Register REG[30h] Default = 00000000h Read/Write																	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
n/a						Vertical Total bits 9-0											
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		
Vertical Display Period Register REG[34h] Default = 00000000h Read/Write																	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
n/a						Vertical Display Period bits 9-0											
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		
Vertical Display Period Start Position Register REG[38h] Default = 00000000h Read/Write																	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
n/a						Vertical Display Period Start Position bits 9-0											
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		
FPFRAME Register REG[3Ch] Default = 00000000h Read/Write																	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
n/a								FPFRAME Polarity	n/a				FPFRAME Pulse Width bits 2-0				
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		
n/a									FPFRAME Pulse Start Position bits 9-0								
Main Window Display Start Address Register REG[40h] Default = 00000000h Read/Write																	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
Main Window Display Start Address bits 15-0															bit 16		
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		
Main Window Line Address Offset Register REG[44h] Default = 00000000h Read/Write																	
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16		
n/a						Main Window Line Address Offset bits 9-0											
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0		

PICTURE-IN-PICTURE PLUS (PIP+) REGISTERS

PIP+ Display Start Address Register REG[50h] Default = 00000000h Read/Write															
n/a															bit 16
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
PIP+ Window Display Start Address bits 15-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
PIP+ Window Line Address Offset Register REG[54h] Default = 00000000h Read/Write															
n/a															bit 16
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						PIP+ Window Line Address Offset bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
PIP+ Window X Positions Register REG[58h] Default = 00000000h Read/Write															
n/a															bit 16
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						PIP+ Window X End Position bits 9-0									
n/a						PIP+ Window X Start Position bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
PIP+ Window Y Positions Register REG[5Ch] Default = 00000000h Read/Write															
n/a															bit 16
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a						PIP+ Window Y End Position bits 9-0									
n/a						PIP+ Window Y Start Position bits 9-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

MISCELLANEOUS REGISTERS

Special Purpose Register REG[60h] Default = 00000000h Read/Write															
n/a								Reserved							
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a								2D Byte Swap	Display Data Word Swap	Display Data Byte Swap	n/a		Latch Byte Select	n/a	
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
GPIO Status and Control Register REG[64h] Default = 00000000h Read/Write															
GPIO7 Input Enable	GPIO6 Input Enable	GPIO5 Input Enable	GPIO4 Input Enable	GPIO3 Input Enable	GPIO2 Input Enable	GPIO1 Input Enable	GPIO0 Input Enable	GPIO7 Config	GPIO6 Config	GPIO5 Config	GPIO4 Config	GPIO3 Config	GPIO2 Config	GPIO1 Config	GPIO0 Config
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a								GPIO7 Status	GPIO6 Status	GPIO5 Status	GPIO4 Status	GPIO3 Status	GPIO2 Status	GPIO1 Status	GPIO0 Status
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Brightness (PWM) Configuration Register REG[70h] Default = 00000000h Read/Write															
n/a															bit 16
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a								PWM Clock Divide Select bits 3-0				PWM Clock Force High	PWMCLK Source Select bits 1-0		PWM Clock Enable
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Brightness (PWM) Duty Cycle Register REG[74h] Default = 00000000h Read/Write															
n/a															bit 16
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
n/a								PWMOUT Duty Cycle bits 7-0							
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Scratch Pad A Register REG[80h] Default = not applicable Read/Write															
Scratch Pad A bits 31-24															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
Scratch Pad A bits 15-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Scratch Pad B Register REG[84h] Default = not applicable Read/Write															
Scratch Pad B bits 31-24															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
Scratch Pad B bits 15-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Scratch Pad C Register REG[88h] Default = not applicable Read/Write															
Scratch Pad C bits 31-24															
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
Scratch Pad C bits 15-0															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0

USB REGISTERS

Control Register REG[4000h] Default = 00h																Read/Write										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	USBClk Enable	Software EOT	USB Enable	Endpoint 4 Stall	Endpoint 3 Stall	USB Setup	Reserved	Reserved			
Interrupt Enable Register 0 REG[4002h] Default = 00h																Read/Write										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	Suspend Request Interrupt Enable	SOF Interrupt Enable	Reserved	Endpoint 4 Interrupt Enable	Endpoint 3 Interrupt Enable	Endpoint 2 Interrupt Enable	Endpoint 1 Interrupt Enable	n/a			
Interrupt Status Register 0 REG[4004h] Default = 00h																Read/Write										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	Suspend Request Interrupt Status	SOF Interrupt Status	Reserved	Endpoint 4 Interrupt Status	Endpoint 3 Interrupt Status	Endpoint 2 Interrupt Status	Endpoint 1 Interrupt Status	Upper Interrupt Active (read only)			
Interrupt Enable Register 1 REG[4006h] Default = 00h																Read/Write										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	n/a						Transmit FIFO Almost Empty Interrupt Enable	Receive FIFO Almost Full Interrupt Enable			
Interrupt Status Register 1 REG[4008h] Default = 00h																Read/Write										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	n/a						Transmit FIFO Almost Empty Status	Receive FIFO Almost Full Status			
Endpoint 1 Index Register REG[4010h] Default = 00h																Read/Write										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	n/a						Endpoint 1 Index bits 2-0				
Endpoint 1 Receive Mailbox Data Register REG[4012h] Default = 00h																Read Only										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	n/a						Endpoint 1 Receive Mailbox Data bits 7-0				
Endpoint 2 Index Register REG[4018h] Default = 00h																Read/Write										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	n/a						Endpoint 2 Index bits 2-0				
Endpoint 2 Transmit Mailbox Data Register REG[401Ah] Default = 00h																Read/Write										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	n/a						Endpoint 2 Transmit Mailbox Data bits 7-0				
Endpoint 2 Interrupt Polling Interval Register REG[401Ch] Default = FFh																Read/Write										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	n/a						Endpoint 2 Interrupt Polling Interval bits 7-0				
Endpoint 3 Receive FIFO Data Register REG[4020h] Default = 00h																Read Only										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	n/a						Endpoint 3 Receive FIFO Data bits 7-0				
Endpoint 3 Receive FIFO Count Register REG[4022h] Default = 00h																Read Only										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	n/a						Endpoint 3 Receive FIFO Count bits 7-0				
Endpoint 3 Receive FIFO Status Register REG[4024h] Default = 01h																Read/Write										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	n/a						Receive FIFO Flush	Reserved	Receive FIFO Underflow	Receive FIFO Full (read only)	Receive FIFO Empty (read only)
Endpoint 3 Maximum Packet Size Register REG[4026h] Default = 08h																Read/Write										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	n/a						Endpoint 3 Max Packet Size bits 7-0				
Endpoint 4 Transmit FIFO Data Register REG[4028h] Default = 00h																Write Only										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	n/a						Endpoint 4 Transmit FIFO Data bits 7-0				
Endpoint 4 Transmit FIFO Count Register REG[402Ah] Default = 00h																Read Only										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	n/a						Endpoint 4 Transmit FIFO Count bits 7-0				

Endpoint 4 Transmit FIFO Status Register REG[402Ch] Default = 01h																Read/Write			
n/a										Transmit FIFO Valid	Transmit FIFO Flush	Transmit FIFO Overflow	Reserved	Transmit FIFO Full (read only)	Transmit FIFO Empty (read only)				
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
Endpoint 4 Maximum Packet Size Register REG[402Eh] Default = 08h																Read/Write			
n/a										Endpoint 4 Max Packet Size bits 7-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
Revision Register REG[4030h] Default = 01h																Read Only			
n/a										Chip Revision bits 7-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
USB Status Register REG[4032h] Default = 00h																Read/Write			
n/a										Suspend Control	USB Endpoint 4 STALL	USB Endpoint 4 NAK	USB Endpoint 4 ACK	USB Endpoint 3 STALL	USB Endpoint 3 NAK	USB Endpoint 3 ACK	Endpoint 2 Valid		
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
Frame Counter MSB Register REG[4034h] Default = 00h																Read Only			
n/a										Frame Counter bits 10-8									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
Frame Counter LSB Register REG[4036h] Default = 00h																Read Only			
n/a										Frame Counter bits 7-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
Extended Register Index REG[4038h] Default = 00h																Read/Write			
n/a										Extended Register Index bits 7-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
Extended Register Data REG[403Ah] Default = 04h																Read/Write			
n/a										Extended Register Data bits 7-0									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
Vendor ID MSB REG[403Ah], Index[00h] Default = 04h								Read/Write		Vendor ID LSB REG[403Ah], Index[01h] Default = B8h								Read/Write	
Vendor ID bits 15-8								Vendor ID bits 7-0											
7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0				
Product ID MSB REG[403Ah], Index[02h] Default = 88h								Read/Write		Product ID LSB REG[403Ah], Index[03h] Default = 21h								Read/Write	
Product ID bits 15-8								Product ID bits 7-0											
7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0				
Release Number MSB REG[403Ah], Index[04h] Default = 01h								Read/Write		Release Number LSB REG[403Ah], Index[05h] Default = 00h								Read/Write	
Release Number bits 15-8								Release Number bits 7-0											
7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0				
Receive FIFO Almost Full Threshold REG[403Ah], Index[06h] Default = 3Ch								Read/Write		Transmit FIFO Almost Empty Threshold REG[403Ah], Index[07h] Default = 04h								Read/Write	
n/a		Receive FIFO Almost Full Threshold bits 5-0						n/a		Transmit FIFO Almost Empty Threshold bits 5-0									
7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0				
USB Control REG[403Ah], Index[08h] Default = 01h								Read/Write		Maximum Power Consumption REG[403Ah], Index[09h] Default = FAh								Read/Write	
n/a								USB String Enable		Maximum Current bits 7-0									
7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0				
Packet Control REG[403Ah], Index[0Ah] Default = 00h								Read/Write		Reserved REG[403Ah], Index[0Bh] Default = 00h								Read/Write	
EP4 Data Toggle	EP3 Data Toggle	EP2 Data Toggle	EP1 Data Toggle	Reserved	Reserved	n/a	Reserved	n/a								Reserved			
7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0				
FIFO Control REG[403Ah], Index[0Ch] Default = 00h								Read/Write											
n/a								Transmit FIFO Valid Mode											
7	6	5	4	3	2	1	0												
USBFC Input Control Register REG[4040h] Default = 0Dh																Read/Write			
n/a										USCMPEM	Reserved	Reserved	ISO	WAKEUP	Reserved	Reserved			
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
Reserved REG[4042h] Default = 1Dh																Read Only			
n/a										Reserved									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				

Pin Input Status / Pin Output Data Register REG[4044h] Default = depends on USB input pin state															Read/Write				
n/a													USBDETECT Input Pin Status (read only)	USBPUP Output Pin Status					
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
Positive Interrupt Enable (Mask) Register REG[4046h] Default = 00h															Read/Write				
n/a													USBDETECT	USBPUP	DEVCFG	SUSP	OSCRUN	USBRESET	INT
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
Negative Interrupt Enable (Mask) Register REG[4048h] Default = 00h															Read/Write				
n/a													USBDETECT	USBPUP	DEVCFG	SUSP	OSCRUN	USBRESET	INT
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
Positive Interrupt Status/Clear Register REG[404Ah] Default = 00h															Read/Write				
n/a													USBDETECT	USBPUP	DEVCFG	SUSP	OSCRUN	USBRESET	INT
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
Negative Interrupt Status/Clear Register REG[404Ch] Default = 00h															Read/Write				
n/a													USBDETECT	USBPUP	DEVCFG	SUSP	OSCRUN	USBRESET	INT
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
Positive Interrupt Masked Status Register REG[404Eh] Default = 00h															Read Only				
n/a													USBDETECT	USBPUP	DEVCFG	SUSP	OSCRUN	USBRESET	INT
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
Negative Interrupt Masked Status Register REG[4050h] Default = 00h															Read Only				
n/a													USBDETECT	USBPUP	DEVCFG	SUSP	OSCRUN	USBRESET	INT
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
USB Software Reset Register REG[4052h] Default = 00h															Write Only				
n/a								USB Software Reset (Code = 10100100) bits 7-0											
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				
USB Wait State Register REG[4054h] Default = 00h															Read/Write				
n/a								n/a						USB Wait State bits 1-0					
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0				

2D ACCELERATION (BitBLT) REGISTERS

BitBLT Control Register REG[8000h] Default = 00000000h																Read/Write			
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	Color Format Select	Dest Linear Select	Source Linear Select	
n/a																			
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	BitBLT Enable (WO)			
n/a																			
BitBLT Status Register REG[8004h] Default = 00000000h																Read Only			
31	30	29	Number of Used FIFO Entries				23	22	Number of Free FIFO Entries (0 means full)										
n/a							n/a												
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	FIFO Not Empty	FIFO Half Full	FIFO Full Status	BitBLT Busy Status
n/a															n/a				
BitBLT Command Register REG[8008h] Default = 00000000h																Read/Write			
31	30	29	28	27	26	25	24	23	22	21	20	BitBLT ROP Code bits 3-0							
n/a																			
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	BitBLT Operation bits 3-0			
n/a																			
BitBLT Source Start Address Register REG[800Ch] Default = 00000000h																Read/Write			
31	30	29	28	27	26	25	24	23	22	21	BitBLT Source Start Address bits 20-16								
n/a																			
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	BitBLT Source Start Address bits 15-0			
n/a																			
BitBLT Destination Start Address Register REG[8010h] Default = 00000000h																Read/Write			
31	30	29	28	27	26	25	24	23	22	21	BitBLT Destination Start Address bits 20-16								
n/a																			
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	BitBLT Destination Start Address bits 15-0			
n/a																			
BitBLT Memory Address Offset Register REG[8014h] Default = 00000000h																Read/Write			
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	n/a			
n/a																			
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	BitBLT Memory Address Offset bits 10-0			
n/a																			
BitBLT Width Register REG[8018h] Default = 00000000h																Read/Write			
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	n/a			
n/a																			
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	BitBLT Width bits 9-0			
n/a																			
BitBLT Height Register REG[801Ch] Default = 00000000h																Read/Write			
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	n/a			
n/a																			
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	BitBLT Height bits 9-0			
n/a																			
BitBLT Background Color Register REG[8020h] Default = 00000000h																Read/Write			
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	n/a			
n/a																			
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	BitBLT Background Color bits 15-0			
n/a																			
BitBLT Foreground Color Register REG[8024h] Default = 00000000h																Read/Write			
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	n/a			
n/a																			
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	BitBLT Foreground Color bits 15-0			
n/a																			
2D Accelerator (BitBLT) Data Memory Mapped Region Register AB16-AB0 = 10000h-1FFFh, even addresses Read/Write																			
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	BitBLT Data bits 31-16			
n/a																			
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	BitBLT Data bits 15-0			
n/a																			

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EPSON®



S1D13A04 LCD/USB Companion Chip

13A04CFG Configuration Program

Document Number: X37A-B-001-01

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13A04CFG

13A04CFG is an interactive Windows® program that calculates register values for a user-defined S1D13A04 configuration. The configuration information can be used to directly alter the operating characteristics of the S1D13A04 utilities or any program built with the Hardware Abstraction Layer (HAL) library. Alternatively, the configuration information can be saved in a variety of text file formats for use in other applications.

S1D13A04 Supported Evaluation Platforms

13A04CFG runs on PC system running Windows 9x/ME/NT/2000 and can modify Win32 .exe files and .s9 format files.

Installation

Create a directory for **13A04cfg.exe** and copy the files **13A04cfg.exe** and **panels.def** to that directory. **Panels.def** contains configuration information for a number of panels and must reside in the same directory as **13A04cfg.exe**.

Usage

To start 13A04CFG from the Windows desktop, click on the My Computer icon and run the program 13a04cfg.exe from the installed directory.

To start 13A04CFG from a Windows command prompt, change to the directory **13A04cfg.exe** was installed to and type the command **13A04cfg**.

The basic procedure for using 13A04CFG is:

1. Start 13A04CFG as described above.
2. Open an existing file to serve as a starting reference point (this step is optional).
3. Modify the configuration. For specific information on editing the configuration, see “13A04CFG Configuration Tabs” on page 6.
4. Save the new configuration. The configuration information can be saved in two ways; as an ASCII text file or by modifying an executable image on disk.

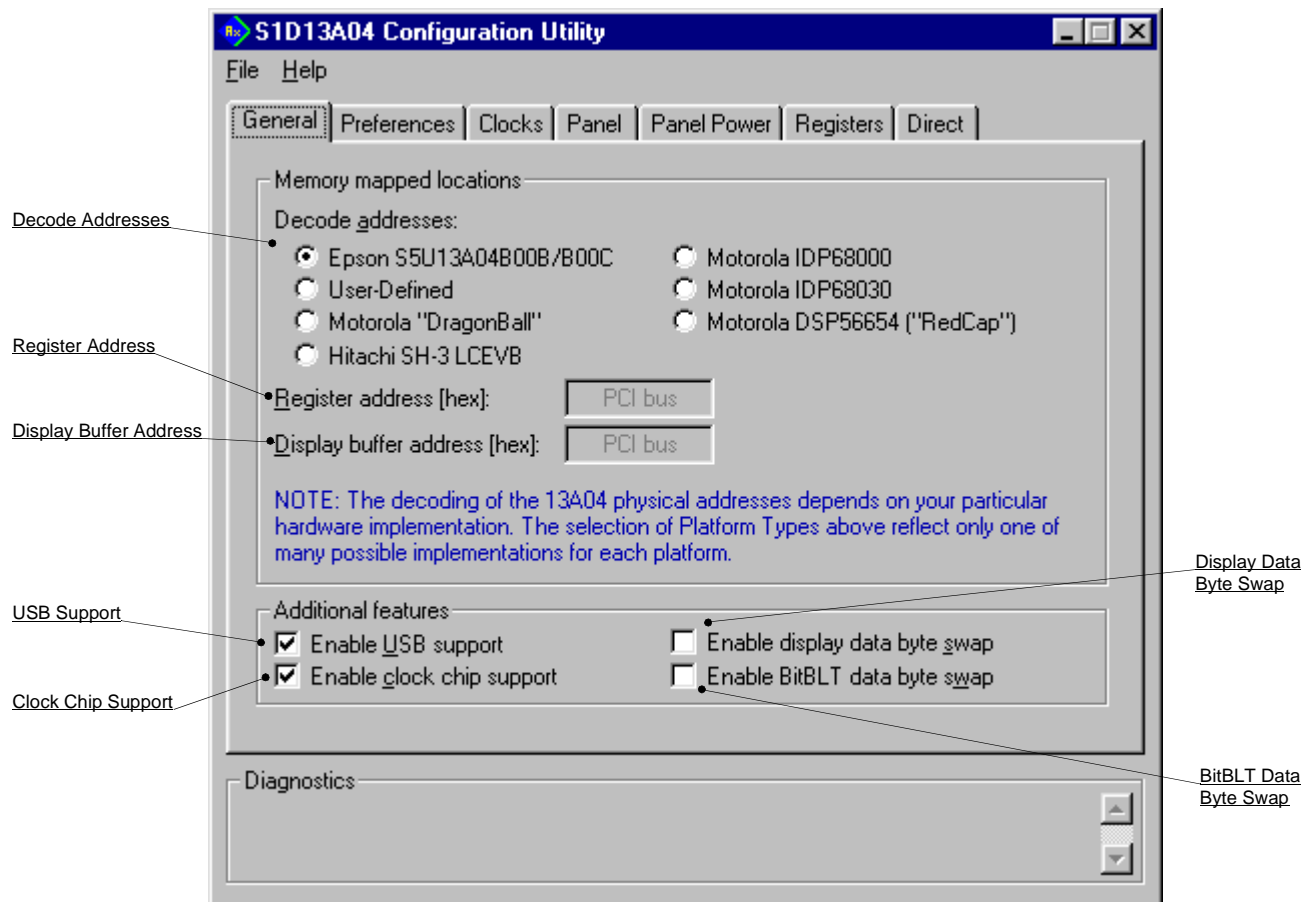
Several ASCII text file formats are supported. Most are formatted C header files used to build display drivers or standalone applications.

Utility files based on the Hardware Abstraction Layer (HAL) can be modified directly by 13A04CFG.

13A04CFG Configuration Tabs

13A04CFG displays a series of tabs which can be selected at the top of the main window. Each tab allows the configuration of a specific aspect of S1D13A04 operation. The following sections describe the purpose and use of each of the tabs.

General Tab



The General tab contains settings that define the S1D13A04 operating environment.

Decode Addresses

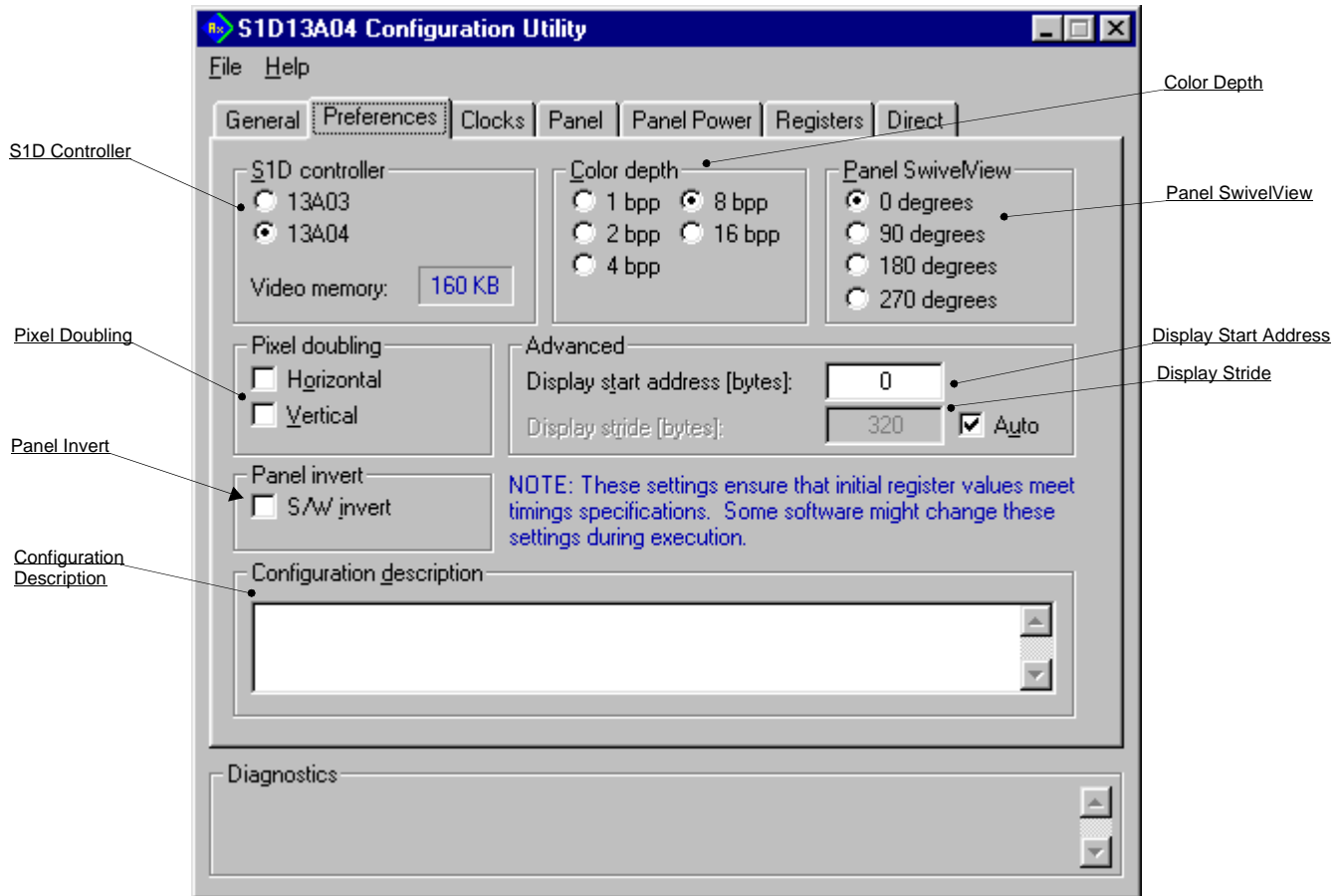
Selecting one of the listed evaluation platforms changes the values for the “Register address” and “Display buffer address” fields. The values used for each evaluation platform are examples of possible implementations as used by the Epson S5U13A04B00C evaluation board. If your hardware implementation differs from the addresses used, select the User-Defined option and enter the correct values for “Register address” and “Display buffer address”.

Note

When “Epson S5U13A04B00B/B00C Evaluation Board” is selected, the register and display buffer addresses are blanked because the evaluation board uses the PCI interface and the decode addresses are determined by the system BIOS during boot-up.

Register Address	<p>The physical address of the start of register decode space (in hexadecimal).</p> <p>This field is automatically set according to the Decode Address unless the “User-Defined” decode address is selected.</p>
Display Buffer Address	<p>The physical address of the start of display buffer decode space (in hexadecimal).</p> <p>This field is automatically set according to the Decode Address unless the “User-Defined” decode address is selected.</p>
USB Support	<p>The S1D13A04 contains a USB client controller. If this box is checked, chip initialization will configure GPIO[7:4] for use by the USB controller. For further information on the S1D13A04 USB implementation, see the <i>S1D13A04 Hardware Functional Specification</i>, document number X37A-A-001-xx.</p>
Clock Chip Support	<p>The S5U13A04B00C evaluation board implements a Cypress ICD2061A Clock Synthesizer which can be used to generate CLKI and CLKI2. When this box is checked, GPIO[3:1] are reserved for Clock Synthesizer support. Selecting a HR-TFT panel will disable this feature as the HR-TFT requires GPIO[3:0].</p> <p>Note that this feature is only available when using the S5U13A04B00C.</p>
Display Data Byte Swap	<p>When this box is checked, the word data from the display buffer is byte swapped. This box should be checked when the display color depth is 16 bpp on a Big Endian platform.</p>
BitBLT Data Byte Swap	<p>When this box is checked, word data sent to/read from the 2D BitBLT memory is byte swapped. This box should be checked when using the 2D BitBLT functions on a Big Endian platform.</p>

Preferences Tab



The Preference tab contains settings pertaining to the initial display state. During runtime these settings may be changed.

S1D Controller

13A04CFG is designed to support the S1D13A03 and the S1D13A04. When a controller is selected, 13A04CFG determines the maximum panel dimensions based on the video memory (display buffer) size. Some S1D13A04 configurations may not be possible for the S1D13A03 due to memory limitations.

Pixel Doubling

These settings allow the Pixel Doubling feature to be configured independently in the horizontal and vertical dimensions. Pixel doubling causes each pixel of display data to be extended to two pixels. This feature can be useful for using existing software on larger panels.

Horizontal

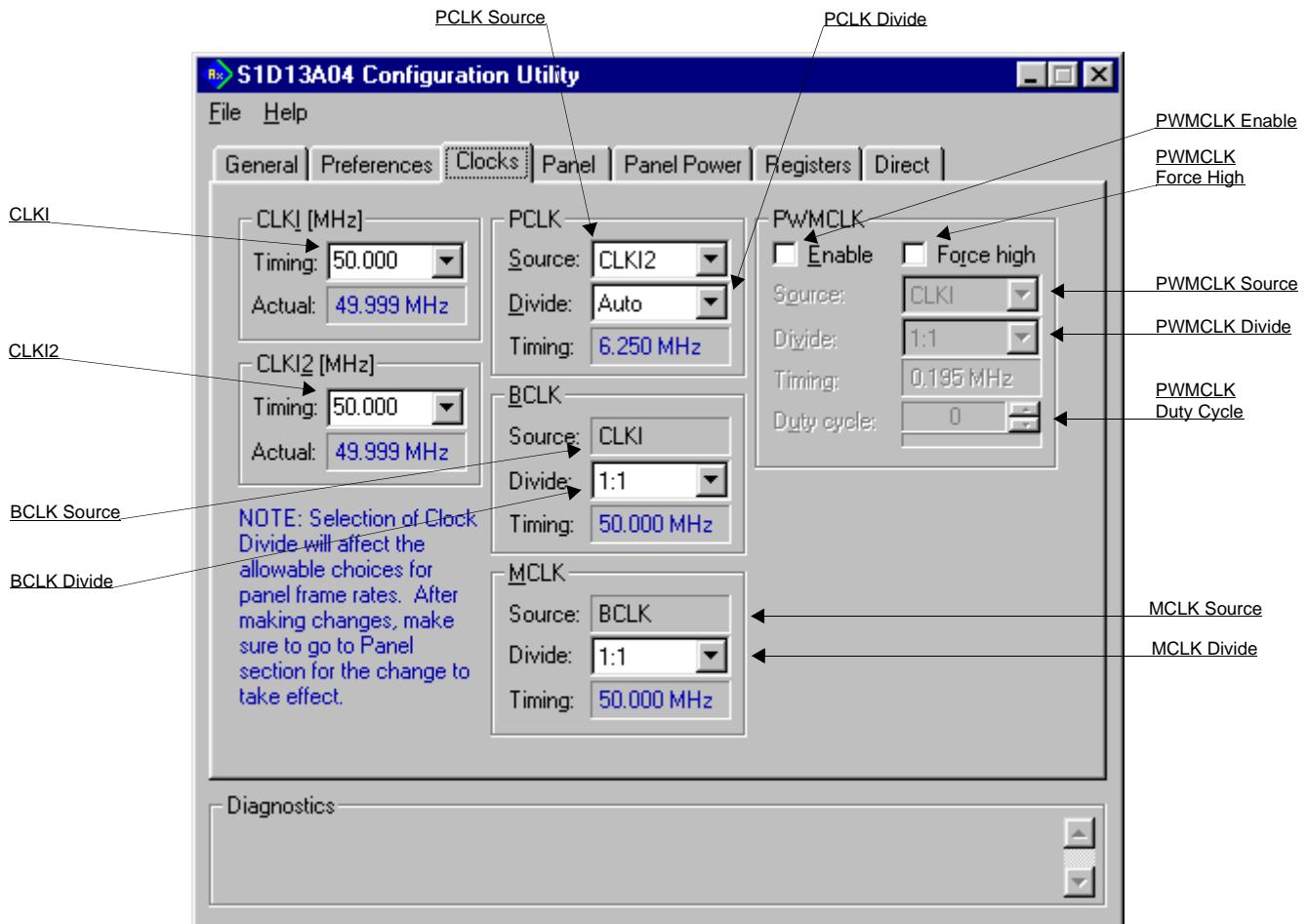
When this box is checked, pixel doubling in the horizontal direction is enabled. **Note that the S1D13A04 does not support horizontal pixel doubling at a color depth of 1 bpp.**

Vertical	When this box is checked, pixel doubling in the vertical direction is enabled. Note that the S1D13A04 does not support vertical pixel doubling at a color depth of 1 bpp.
Panel Invert	The S1D13A04 can invert the display data going to the LCD panel. When this box is checked, video data is inverted. Display data is inverted after the Look-Up Table, which means colors are truly inverted.
Configuration Description	This field allows the user to enter a description for a particular configuration. This field is saved in the HAL information and is displayed when a HAL-based utility is run.
Color Depth	Sets the initial color depth on the LCD panel. If there is insufficient display buffer for the selected width and height then a warning is displayed in the diagnostic area.
Panel SwivelView	The S1D13A04 SwivelView feature is capable of rotating the image displayed on an LCD panel 90°, 180°, or 270° in a counter-clockwise direction. This setting determines the initial orientation of the panel.
Advanced	These settings allow fine tuning of the start address and stride. The start address defines the offset into the display buffer (video memory) of the pixel which will be displayed in the top left corner of the panel. Stride defines the number of bytes required to step from the first pixel on one row to the first pixel on the next row (i.e. 160 pixel wide display at 16 bpp requires 320 bytes per horizontal row).
Display Start Address	This option sets the start address for the main window of the panel. Typically the start address is set to zero.
Display Stride	This option sets the stride for the main window of the panel. To set the stride equal to the size of the display, select the “auto” box. To increase the stride, uncheck the “auto” box and enter the desired stride.

Note

The stride value must be greater than or equal to the number of bytes required by one line of display memory.

Clocks Tab



The Clocks tab simplifies the selection of input clock frequencies and the sources of internal clocking signals. For further information regarding clocking and clock sources, refer to the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

The CLKI and CLKI2 frequencies represent clock values the system provides to the S1D13A04. It is the responsibility of the system designer to ensure that the correct clock frequencies are supplied to the S1D13A04.

Note

Changing clock values may modify or invalidate Panel settings. Confirm all settings on the Panel tab after modifying any clock settings.

The S1D13A04 may use one or two clock sources. Two clock sources allow greater flexibility in the selection of display type and memory speed.

CLKI	This setting determines the frequency of CLKI.
Timing	Set this value by selecting a preset frequency from the drop down list or entering the desired frequency (MHz) in the edit box.
Actual	This field displays the actual value of the CLKI frequency. If “Enable clock chip support” is selected on the General Tab, then this value may differ slightly from the value entered in the timing control.
CLKI2	This setting determines the frequency of CLKI2.
Timing	Set this value by selecting a preset frequency from the drop down list or entering the desired frequency (MHz) in the edit box.
Actual	This field displays the actual value of the CLKI2 frequency. If “Enable clock chip support” is selected on the General Tab, then this value may differ slightly from the value entered in the timing control.
PCLK	These settings select the clock source and divisor for the internal pixel clock (PCLK).
Source	Selects the PCLK source. Possible sources include CLKI, CLKI2, BCLK or MCLK. Note that BCLK and MCLK may be previously divided from CLKI or CLKI2.
Divide	Specifies the divide ratio for the clock source. The divide ratio is applied to the PCLK source to derive PCLK. Selecting “Auto” for the divisor allows the configuration program to calculate the best clock divisor. Unless a very specific clocking is being specified, it is best to leave this setting on “Auto”.
Timing	This field shows the actual PCLK used by the configuration process.

BCLK	These settings select the clock source and divisor for the internal bus interface clock (BCLK).
Source	The BCLK source is always CLKI.
Divide	Specifies the divide ratio for the clock source. The divide ratio is applied to the BCLK source to derive BCLK.
Timing	This field shows the actual BCLK frequency used by the configuration process.
MCLK	These settings select the clock source and input clock divisor for the internal memory clock (MCLK). For the best performance, MCLK should be set as close to the maximum (50 MHz) as possible.
Source	The MCLK source is always BCLK.
Divide	Specifies the divide ratio for the clock source. The divide ratio is applied to the MCLK source to derive MCLK.
	This divide ratio should be left at 1:1 unless the resultant MCLK is greater than 50MHz.
Timing	This field shows the actual MCLK frequency used by the configuration process.

PWMCLK

These controls configure various PWMCLK settings. The PWMCLK is the internal clock used by the Pulse Width Modulator for output to the panel.

Enable

When this box is checked, the PWMCLK circuitry is enabled.

Force High

The signal PWMOUT is forced high when this box is checked. When not checked, PWMOUT will be low if PWM is not enabled or will change state according to the configured values when PWM is enabled.

Source

Selects the PWMCLK source. Possible sources include CLKI, CLKI2, MCLK, and PCLK.

Divide

Specifies the divide ratio for the clock source. The divide ratio is applied to the PWMCLK source to derive PWMCLK.

Note

After this divide is applied, PWMCLK is further divided by 256 to achieve the final PWMCLK frequency.

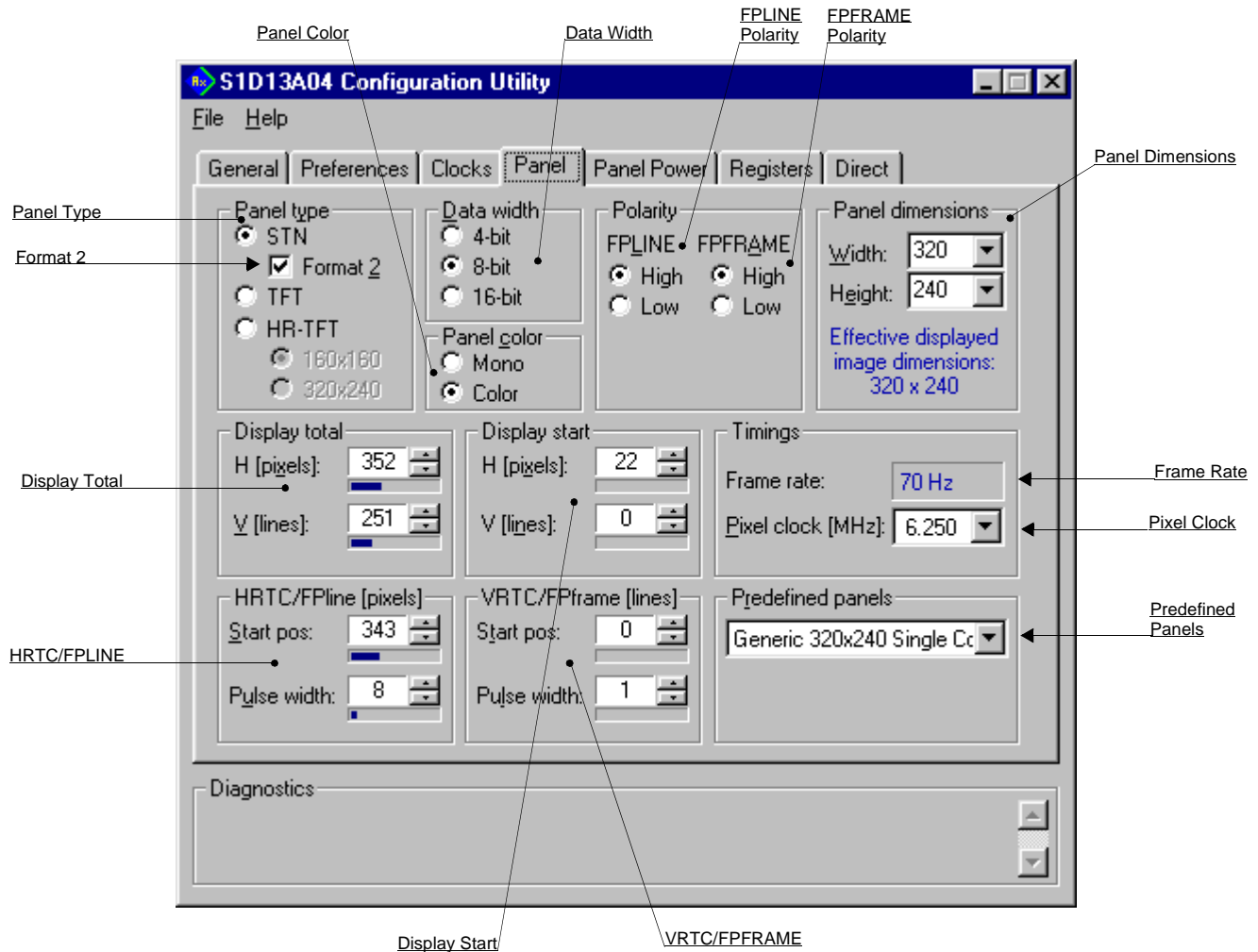
Timing

This field shows the actual PWMCLK frequency used by the configuration process.

Duty Cycle

Selects the number of cycles that PWMOUT is high out of 256 clock periods.

Panel Tab



The S1D13A04 supports many panel types. This tab allows configuration of most panel related settings such as dimensions, type and timings.

Panel Type

Selects between passive (STN), active (TFT), and reflective (HR-TFT) panel types.

Some options may change or become unavailable when the STN/TFT/HR-TFT setting is changed.

Re-confirm all settings on this tab after the Panel Type is changed.

Format 2

Selects color STN panel data format 2. This option only applies to 8-bit color STN panels.

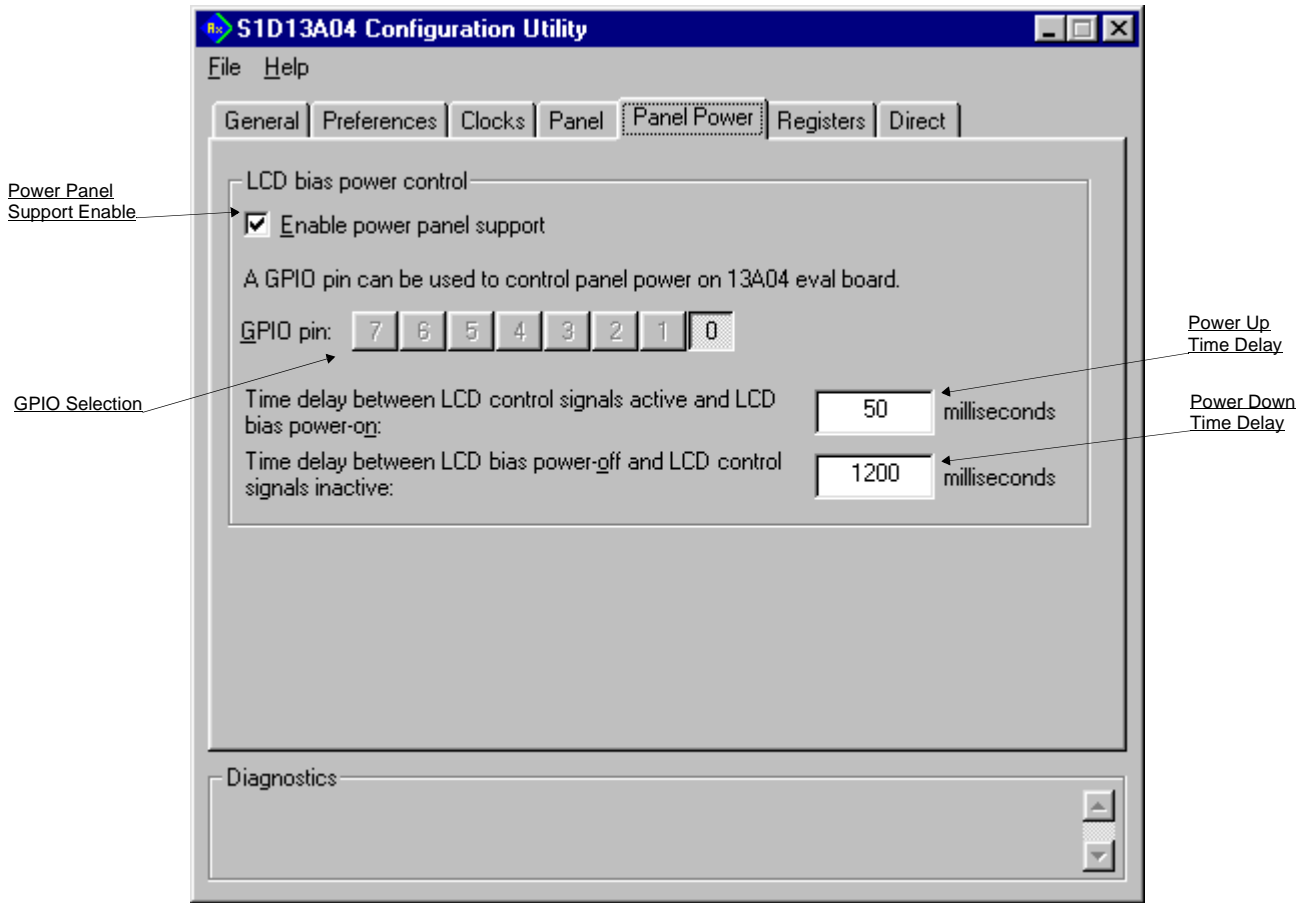
See the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx, for description of format 1 / format 2 data formats. Most new panels use the format 2 data format.

Data Width	<p>Selects the panel data width. Panel data width is the number of bits of data transferred to the LCD panel on each clock cycle and shouldn't be confused with color depth which determines the number of displayed colors.</p> <p>When a passive panel type is selected, the available options are 4, 8, and 16 bit. When an active panel type (TFT/HR-TFT) is selected, the available options are 9, 12, and 18 bit.</p>
Panel Color	Selects between a monochrome or color panel.
Polarity	<p>Allows selection of the polarity for the FPLINE and FPRAME pulses.</p> <p>Note Selecting the wrong pulse polarity may damage the panel.</p>
FPLINE Polarity	<p>Selects the polarity of the FPLINE pulse.</p> <p>Refer to the panel specification for the correct polarity of the FPLINE pulse.</p>
FPFRAME Polarity	<p>Selects the polarity of the FPFRAME pulse.</p> <p>Refer to the panel specification for the correct polarity of the FPFRAME pulse.</p>
Panel Dimensions	<p>These fields specify the panel width and height. A number of common widths and height are available in the selection boxes. If the width/height of your panel is not listed, enter the actual panel dimensions into the edit field.</p> <p>For passive panels, manually entered pixel widths must be a minimum of 32 pixels and can be increased by multiples of 16. For TFT panels, manually entered pixel widths must be a minimum of 8 pixels and can be increased by multiples of 8. If a value is entered that does not meet these requirements, 13A04CFG rounds up the value to the next allowable width. The changed value is reported in the diagnostics portion of the window.</p>

Display Total	<p>It is recommended that the automatically generated Display Total values be used. However, manual adjustment may be used to improve the quality of the displayed image by fine tuning the horizontal and vertical display totals.</p> <p>The display total equals the display period plus the non-display period. Refer to <i>S1D13A04 Hardware Functional Specification</i>, document number X37A-A-001-xx, for a complete description of the Display Total settings.</p> <p>Note</p> <p>If the horizontal or vertical display totals are set too small, 13A04CFG will display a yellow warning message in the diagnostics portion of the window.</p>
Display Start	<p>It is recommended that the automatically generated Display Start values be used. However, manual adjustment may be used to improve the quality of the displayed image by fine tuning the horizontal and vertical display start positions.</p> <p>Refer to <i>S1D13A04 Hardware Functional Specification</i>, document number X37A-A-001-xx, for a complete description of the Display Start settings.</p> <p>Note</p> <p>If the horizontal or vertical display start values are set to values that violate the S1D13A04 Hardware Functional Specification, 13A04CFG will display a yellow warning message in the diagnostics portion of the window.</p>
Frame Rate	<p>The Frame Rate (in Hz) is calculated and displayed based on the current settings as selected on the various tabs. If the resulting Frame Rate is not acceptable, adjust the settings to change the frame rate.</p> <p>Panel dimensions are fixed therefore frame rate can only be adjusted by changing either the PCLK frequency or display total values.</p>
Pixel Clock	<p>Select the desired Pixel Clock (in MHz) from the drop-down list. The range of frequencies displayed is dependent on the PCLK source and divide settings as selected on the Clocks tab.</p>

HRTC/FPLINE (pixels)	These settings allow fine tuning of the TFT line pulse parameters. Refer to <i>S1D13A04 Hardware Functional Specification</i> , document number X37A-A-001-xx for a complete description of the FPLINE pulse settings.
Start pos	Specifies the delay (in pixels) from the start of the horizontal non-display period to the leading edge of the FPLINE pulse.
Pulse Width	Specifies the width (in pixels) of the horizontal sync signal (FPLINE).
VRTC/FPFRAME (lines)	These settings allow fine tuning of the frame pulse parameters. Refer to <i>S1D13A04 Hardware Functional Specification</i> , document number X37A-A-001-xx, for a complete description of the FPFRAME pulse settings.
Start pos	Specifies the delay (in lines) from the start of the vertical non-display period to the leading edge of the FPFRAME pulse.
Pulse width	Specifies the pulse width (in lines) of the vertical sync signal (FPFRAME).
Predefined Panels	13A04CFG uses a file (panels.def) which contains predefined settings for a number of LCD panels. If the file panels.def is present in the same directory as 13A04cfg.exe , the predefined panels are available in the drop-down list. If a panel is selected from the list, 13A04CFG preconfigures its settings to nominal panel values.

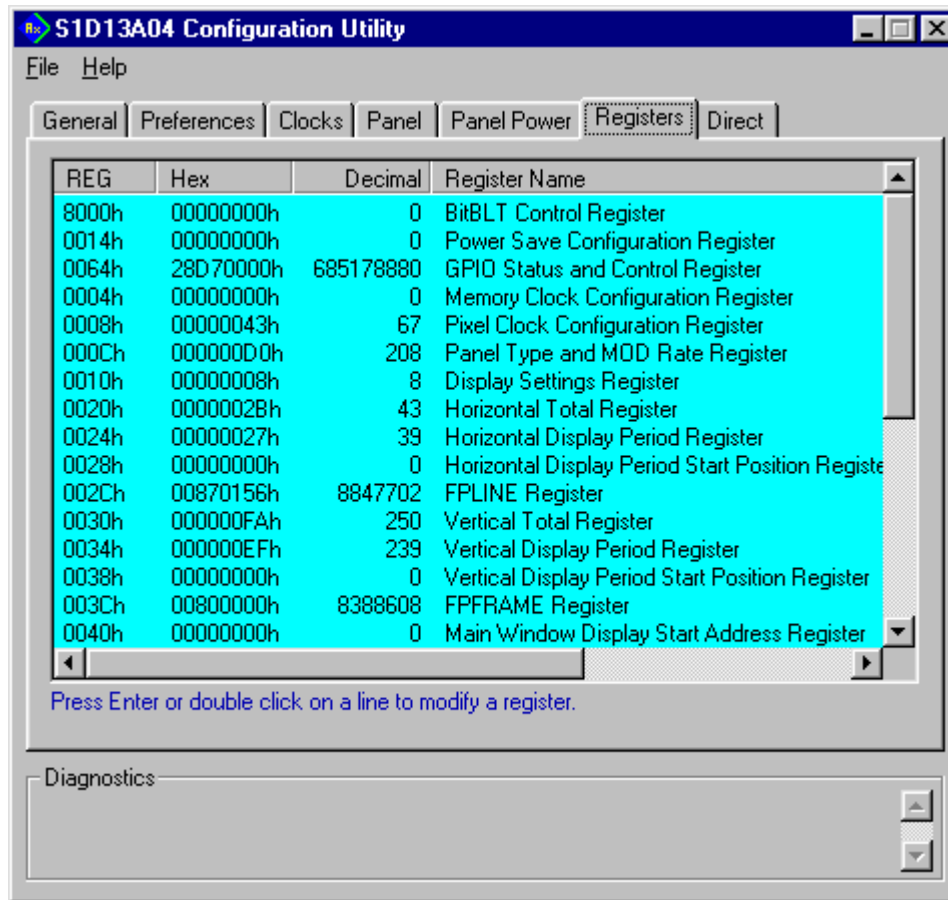
Panel Power Tab



The S5U13A04B00C evaluation board is designed to use a GPIO signal to control the LCD bias power. The following settings configure panel power support.

Power Panel Support Enable	When this box is checked, the LCD bias power to the panel is controlled by the selected GPIO pin. When this box is unchecked, the LCD bias power must be controlled by the CPU or some other means.
GPIO Selection	This setting selects the GPIO pin used to control the LCD bias power. The default is GPIO0 (GPIO0 is used on the S1D13A04 evaluation board).
Power Up Time Delay	<p>This setting controls the maximum time delay between when the S1D13A04 control signals are turned on and the LCD panel is powered-on. This setting must be configured according to the specification for the panel being used.</p> <p>This value is used by Epson evaluation software designed for the S5U13A04B00C evaluation board.</p>
Power Down Time Delay	<p>This setting controls the minimum time delay between when the LCD panel is powered-off and when the S1D13A04 control signals are turned off. This setting must be configured according to the specification for the panel being used.</p> <p>This value is used by Epson evaluation software designed for the S5U13A04B00C evaluation board.</p>

Registers Tab



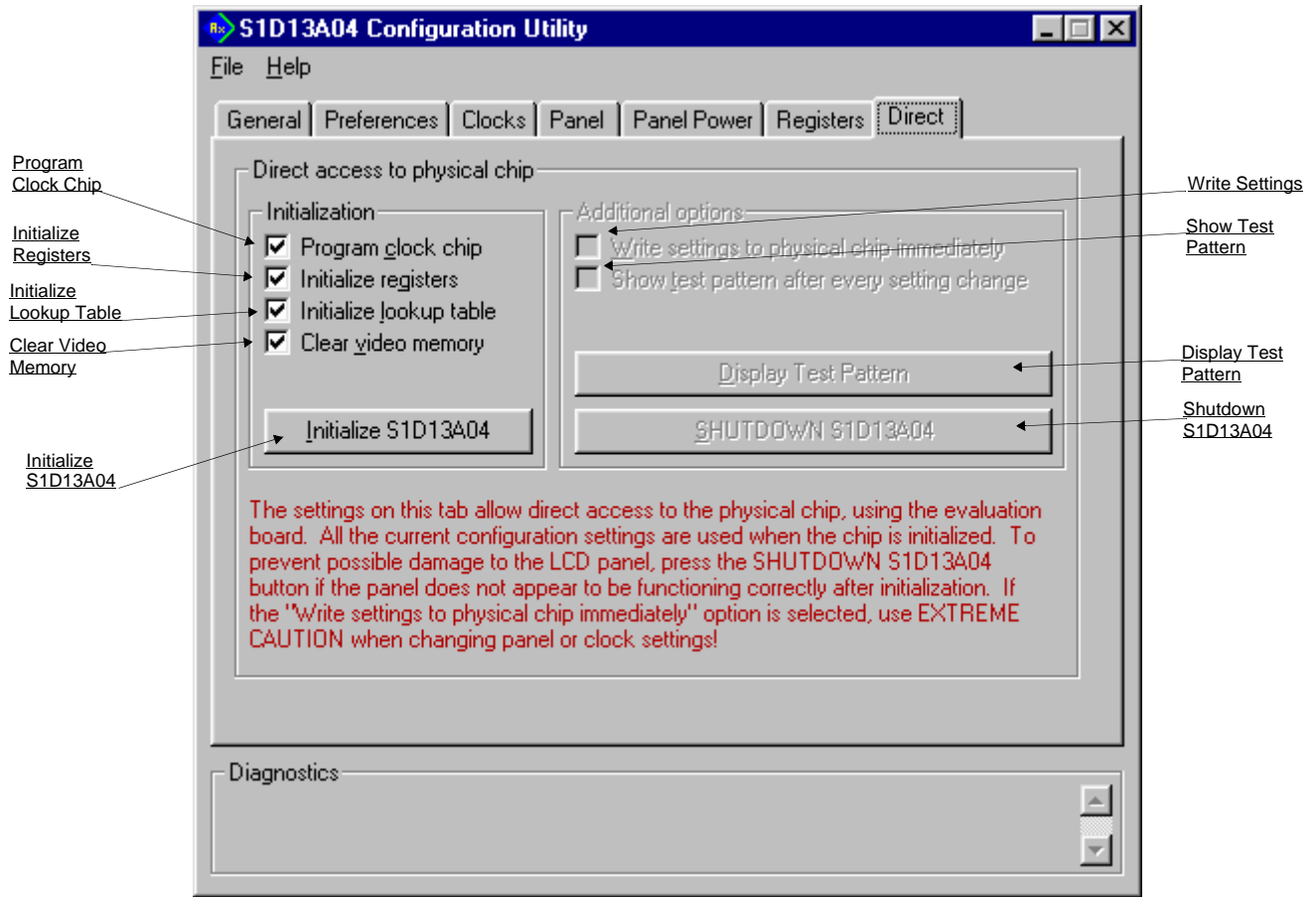
The Registers tab allows viewing and editing of the S1D13A04 register values.

Scroll up and down the list to view register values which are determined from the configuration settings of the previous tabs. Hovering over a register displays a pop-up help box which describes the functionality of the bits in that register. Register settings may be changed by double-clicking on the register in the listing. **Manual changes to the registers are not checked for errors, so caution is warranted when directly editing these values.** It is strongly recommended that the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx be referred to before making any manual register settings.

Note

Manually entered values may be changed by 13A04CFG if further configuration changes are made on other tabs. In this case, the user is notified.

Direct Tab



This tab allows the user to directly interact with the S1D13A04 configuration process. The effect of register changes on the displayed image can be observed before writing any configuration files. Fine tuning adjustments may be made to achieve the best possible image on the panel.

Using this tab requires that a S5U13A04B00C evaluation board is installed in the computer and a panel is attached.

Initialization

These settings define which actions will be carried out when the Initialize S1D13A04 button is clicked.

Program Clock Chip

The S5U13A04B00C evaluation board design includes a clock chip which can provide the signals for CLKI and CLKI2. Checking this box will include programming the clock chip as part of the S1D13A04 initialization.

Initialize Registers

When this box is checked the S1D13A04 registers will be programmed to their configured values as part of the initialization.

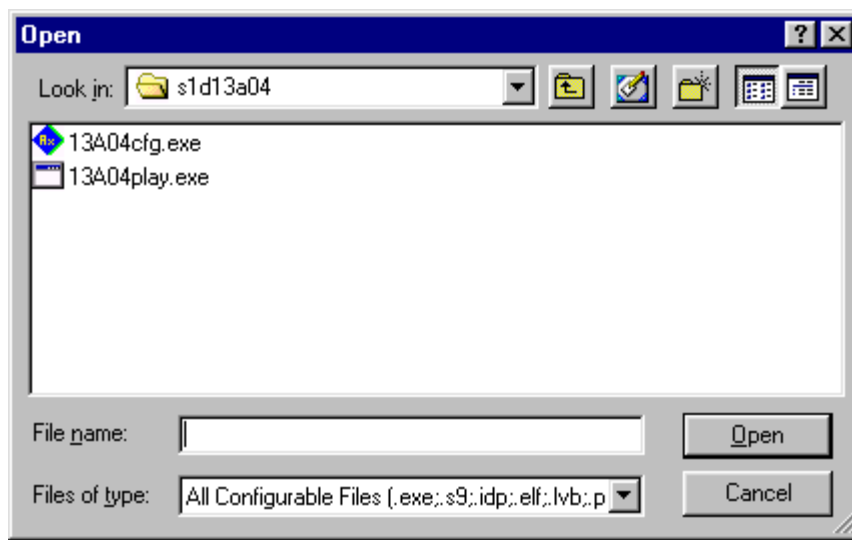
Initialize Lookup Table	When this box is checked the S1D13A04 Lookup Table is programmed as part of the initialization.
Clear Video Memory	When this box is checked the S1D13A04 display buffer will be cleared (set to zeros) as part of the initialization.
Initialize S1D13A04	Clicking this button initializes the S1D13A04 according to the options selected.
Additional Options	After initializing the S1D13A04, these further options become available.
Write Settings	This setting should be used with caution. Checking this box will cause setting changes on any tab to immediately update the associated register(s).
Show Test Pattern	Checking this box will update the test pattern in the display buffer after every setting change. This is useful when fine tuning panel settings as the results of the change are immediately visible.
Display Test Pattern	Clicking this button causes 13A04CFG to draw a test pattern into display memory. The pattern is based on the configured width, height, rotation and color depth.
Shutdown S1D13A04	Clicking this button shuts down the S1D13A04. This feature is necessary should a setting change appear to be damaging or harmful to the attached panel.

13A04CFG Menus

The following sections describe each of the options in the File and Help menus.

Open...

From the Menu Bar, select “File”, then “Open...” to display the Open File Dialog Box.



The Open option allows 13A04CFG to read the configuration information from programs based on the HAL library. When 13A04CFG opens a file it scans the file for an identification string, and if found, reads the configuration information. This feature may be used to quickly arrive at a starting point for register configuration. The only requirement is that the file being opened must contain a valid S1D13A04 HAL library information block.

13A04CFG supports a variety of executable file formats. Select the file type(s) 13A04CFG should display in the Files of Type drop-down list and then select the filename from the list and click on the Open button.

Note

13A04CFG is designed to work with utilities programmed using a given version of the HAL. If the configuration structure contained in the executable file differs from the version 13A04CFG expects the Open will fail and an error message is displayed. This may happen if the version of 13A04CFG is substantially older, or newer, than the file being opened.

Save

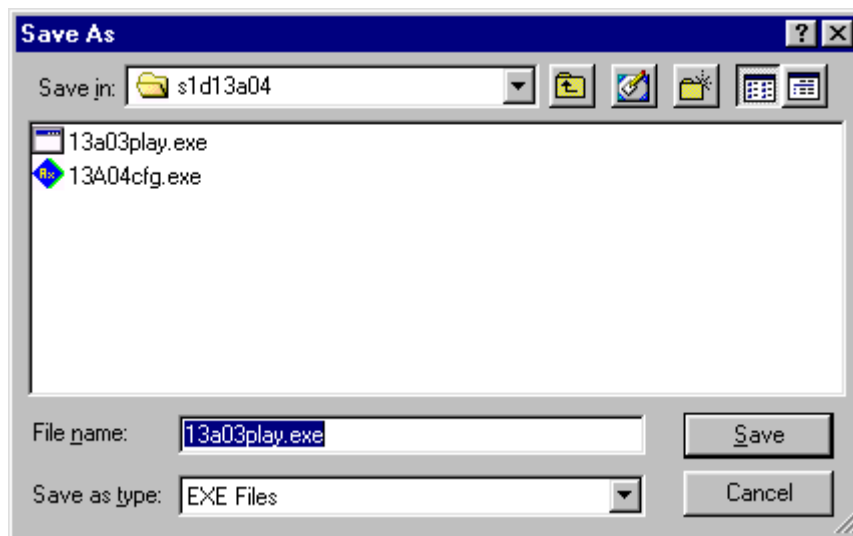
From the Menu Bar, select “File”, then “Save” to initiate the save action. The Save menu option allows a fast save of the configuration information to a file that was opened with the Open menu option.

Note

This option is only available once a file has been opened.

Save As...

From the Menu Bar, select “File”, then “Save As...” to display the Save As Dialog Box.



“Save as” is very similar to Save except a dialog box is displayed allowing the user to name the file before saving.

Using this technique a tester can configure a number of files differing only in configuration information and name (e.g. BMP60Hz.EXE, BMP72Hz.EXE, BMP75Hz.EXE where only the frame rate changes in each of these files).

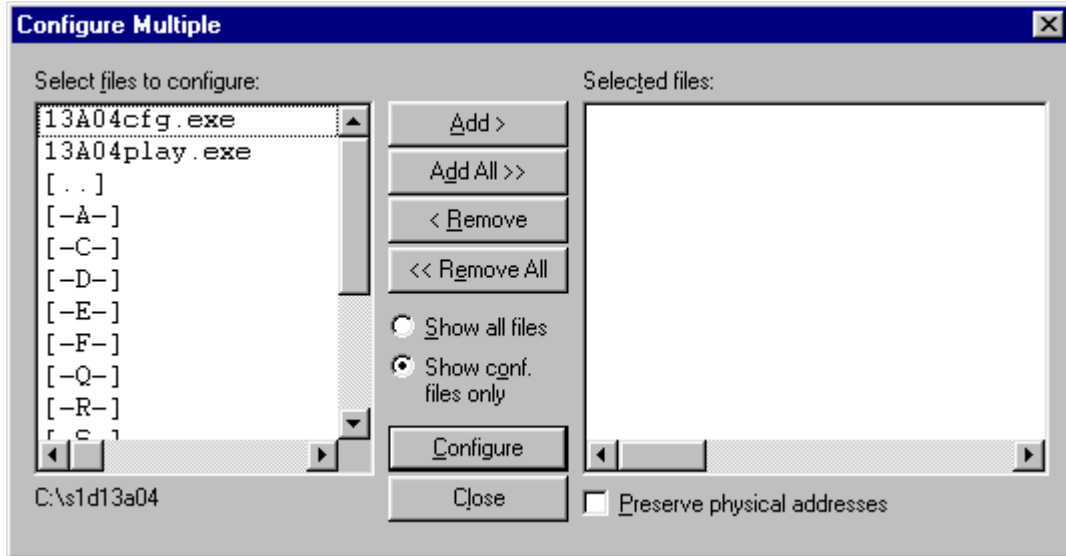
Note

When “Save As” is selected then an exact duplicate of the file as opened by the “Open” option is created containing the new configuration information.

Configure Multiple

After determining the desired configuration, “Configure Multiple” allows the information to be saved into one or more executable files built with the HAL library.

From the Menu Bar, select “File”, then “Configure Multiple” to display the Configure Multiple Dialog Box. This dialog box is also displayed when a file(s) is dragged onto the 13A04CFG window.



The left pane lists files available for configuration; the right pane lists files that have been selected for configuration. Files can be selected by clicking the “Add” or “Add All” buttons, double clicking any file in the left pane, or by dragging the file(s) from Windows Explorer.

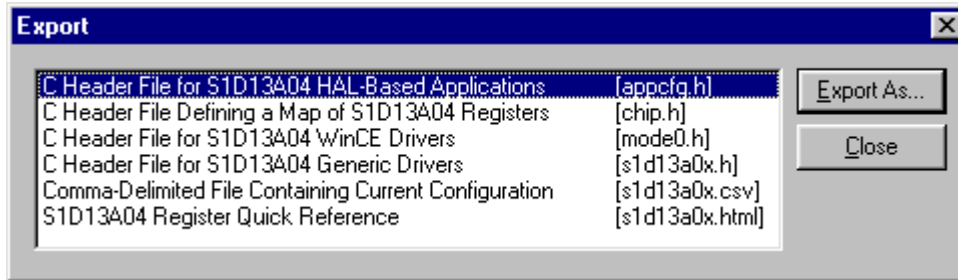
Selecting “Show all files” displays all files in the selected directory, whereas selecting “Show conf. files only” will display only files that can be configured using 13A04CFG (i.e. .exe, .s9).

Checking “Preserve Physical Addresses” instructs 13A04CFG to use the register and display buffer address values the files were previously configured with. Addresses specified in the General Tab are discarded. This is useful when configuring several programs for various hardware platforms at the same time. For example, if configuring PCI, MPC and IDP based programs at the same time for a new panel type, the physical addresses for each are retained. This feature is primarily intended for the test lab where multiple hardware configurations exist and are being tested.

Export

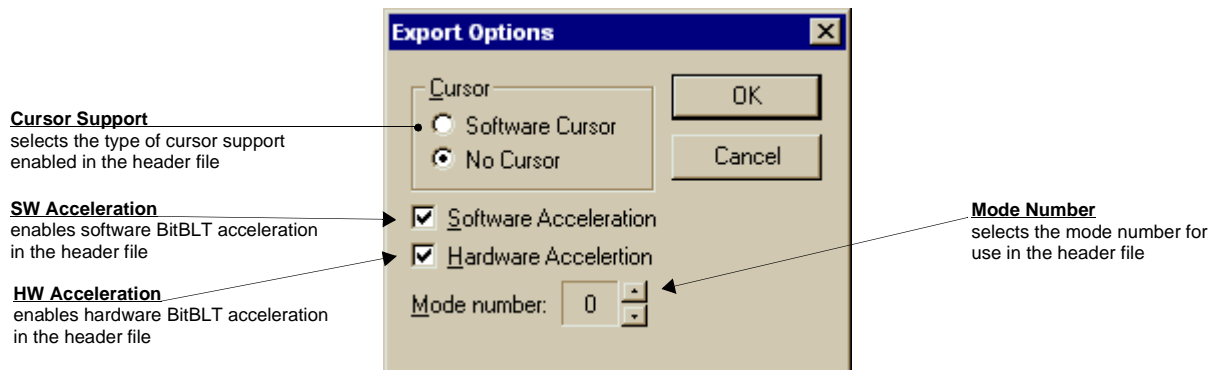
After determining the desired configuration, “Export” permits the user to save the register information as a variety of ASCII text file formats. The following is a list and description of the currently supported output formats:

- a C header file for use in writing HAL library based applications.
- a C header file which lists each register and the value it should be set to.
- a C header file for use in developing Window CE display drivers.
- a C header file for use in developing display drivers for other operating systems such as Linux, QNX, and VxWorks WindML.
- a comma delimited text file containing an offset, a value, and a description for each S1D13A04 register.
- a .html based reference guide to the S1D13A04 registers.



After selecting the file format, click the “Export As...” button to display the file dialog box which allows the user to enter a filename before saving. Clicking the “Preview” button uses Notepad or the web browser to display a copy of the file to be saved.

When the **C Header File for S1D13A04 WinCE Drivers** option is selected as the export type, additional options are available and can be selected by clicking on the Options button. The options dialog appears as:



Enable Tooltips

Tooltips provide useful information about many of the items on the configuration tabs. Placing the mouse pointer over nearly any item on any tab generates a popup window containing helpful advice and hints.

To enable/disable tooltips check/uncheck the “Tooltips” option from the “Help” menu.

Note

Tooltips are enabled by default.

Tooltip Delay

This option sets the length of time the cursor must be left over an item before its associated tooltip is displayed.

ERD on the Web

This “Help” menu item is actually a hotlink to the Epson Research and Development website. Selecting “Help” then “ERD on the Web” starts the default web browser and points it to the ERD product web site.

The latest software, drivers, and documentation for the S1D13A04 is available at this website.

Update Common Controls

13A04CFG uses some of the latest common control DLLs. On systems using earlier versions of the common controls, certain controls may not appear correctly. This option updates the Common Controls required for proper operation of 13A04CFG.

About 13A04CFG

Selecting the “About 13A04CFG” option from the “Help” menu displays the about dialog box for 13A04CFG. The about dialog box contains version information and the copyright notice for 13A04CFG.

Comments

- On any tab particular options may be grayed out if selecting them would violate the operational specification of the S1D13A04 (i.e. Selecting TFT or STN on the Panel tab enables/disables options specific to the panel type).
- The file **panels.def** is a text file containing operational specifications for several supported, and tested, panels. This file can be edited with any text editor.
- 13A04CFG allows manually altering register values. The manual changes may violate memory and LCD timings as specified in the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx. If this is done, unpredictable results may occur. Epson Research and Development, Inc. does not assume liability for any damage done to the display device as a result of configuration errors.
- Yellow diagnostic warnings are designed to draw attention to important errors in the configuration and should be corrected before saving the configuration.
- **13A04CFG** can be configured by making a copy of the file 13A04cfg.exe and configuring the copy. It is not possible to configure the original while it is running. The newly saved information becomes the default configuration for that copy of 13A04cfg.exe.

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S1D13A04 LCD/USB Companion Chip

13A04PLAY Diagnostic Utility

Document Number: X37A-B-002-01

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13A04PLAY

13A04PLAY is a diagnostic utility which allows a user to read/write all the registers and display buffer of the S1D13A04. Commands are received from the standard input device, and messages are sent to the standard output device. On Intel platforms the console provides standard input/output. For most embedded systems the serial device provides these functions.

Commands can be entered interactively by a user, or be executed from a script file. Scripting is a powerful feature which allows command sequences to be used repeatedly without re-entry.

S1D13A04 Supported Evaluation Platforms

13A04PLAY is available as an executable for PCs running Windows® 9x/ME/NT/2000 and as C source code which can be modified and recompiled to allow 13A04PLAY to run on other platforms.

Installation

PC platform

Copy the file **13A04play.exe** to a directory in the path (e.g. PATH=C:\S1D13A04).

Embedded platform

Download the program **13A04play** to the system.

Usage

PC platform

At the prompt, type:

13A04play [/?]

Where:

/? displays copyright and program version information.

Embedded platform

Execute **13A04play** and at the prompt, type the command line argument **/?**.

Where:

/? displays copyright and program version information.

Commands

The following commands are intended to be used from within the 13A04PLAY program. However, simple commands can also be executed from the command line (e.g. **13A04play f 0 14000 AB q**).

Note

If the host platform is big endian, reading/writing words and dwords to/from the registers and display buffer may be incorrect. It may be necessary for the user to manually swap the bytes in order to perform the IO correctly. For further information on little/big endian and the S1D13A04 byte/word swapping capabilities, see the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

CLKI freq

Sets the frequency of CLKI.

Where:

freq The desired frequency for CLKI (in MHz).

CLKI2 freq

Sets the frequency of CLKI2.

Where:

freq The desired frequency for CLKI2 (in MHz).

D[8|16|32] [startaddr [endaddr|len]]

Displays a memory dump from the specified display buffer address range.

Where:

8 16 32	The unit size: 8-bit (bytes), 16-bit (words), 32-bit (dwords). If a unit size is not specified, this command uses the unit size from the last Dump command performed. If no previous Dump command has been issued, the unit size defaults to 8-bit.
startaddr	The starting address to read data from. Specifying a period (.) uses the same starting address as the last Dump command performed. Specifying a startaddr of two periods (..) will back the start address by the size of <i>len</i> .
endaddr len	Determines how many units to continue dumping the contents of the display buffer. A number without a prefix represents a physical ending address. If an "L" prefix is used, the number that follows represents <i>len</i> , which is the number of bytes/words/dwords to be dumped. <i>Len</i> is based on the unit size. For example, 'L8' when the unit size is 16-bit would cause the Dump command to dump 8 words from the starting address.

F[8|16|32] startaddr endaddr|len data1 [data2 data3 ...]

Fills a specified address range in the display buffer.

Where:

8 16 32	The unit size: 8-bit (bytes), 16-bit (words), 32-bit (dwords). If a unit size is not specified, this command uses the unit size from the last Fill command performed. If no previous Fill command has been issued, the unit size defaults to 8-bit.
startaddr	The starting address to begin filling at. Specifying a period (.) uses the same starting address as the last Fill command performed.
endaddr len	Determines how many units to fill the display buffer with. A number without a prefix represents a physical ending address. If a "L" prefix is used, the number that follows represents <i>len</i> , or the number of bytes/words/dwords to be filled. <i>Len</i> is based on the unit size. For example, 'L8' when the unit size is 16-bit would cause the Fill command to fill 8 words from the starting address.
data	The value(s) used to fill the display buffer. If multiple values are given, the pattern repeats through memory. Values can be combinations of 'text' or numbers. Numbers are assumed to be hexadecimal values unless otherwise specified with the correct suffix (binary = i, octal = o, decimal = t, hexadecimal = h). For example, 101i = 101 binary.

H [lines]

Sets the number of lines of data that are displayed at a time. The display is halted after the specified number of lines. Setting the number of lines to 0 disables the halt function and allows the data to continue displaying until all data has been shown.

This command is useful when large blocks of the display buffer or the contents of the LUT are being viewed.

Where:

lines	Number of lines that are shown before halting the displayed data (decimal value).
-------	---

I

Initializes the S1D13A04 registers with the default register settings as configured by the utility 13A04CFG. To initialize the S1D13A04 with different register values, reconfigure 13A04PLAY using 13A04CFG. For further information on 13A04CFG, see the *13A04CFG User Manual*, document number X37A-B-001-xx.

Note

If the "I" command is used before 13A04PLAY is configured, an error message is displayed and no further action is taken.

L index [red green blue]

Reads/writes the red, green, and blue Look-Up Table (LUT) components. If the red, green, and blue components are not specified, the LUT for the given index is read and the RGB values are displayed.

Where:

index	Index into the LUT (hex).
red	Red component of the LUT (hex).
green	Green component of the LUT (hex).
blue	Blue component of the LUT (hex).

Note

Only bits 7-2 of each color are used in the LUT. The least significant two bits of the colors are discarded. For example, the command **L 0 1 2 3** will set each RGB component of LUT index 0 to 0, as the values 1, 2, and 3 use only the least significant bits.

LA

Reads and displays all LUT values.

M [bpp]

Sets the color depth (bpp). If no color depth is provided, information about the current settings are displayed.

Where:

bpp	Color depth to be set (1/2/4/8/16 bpp).
-----	---

Note

This command reads and interprets the S1D13A04 control registers. To function correctly the registers must have been initialized using the 'I' command.

Q

Quits the program.

R[8|16|32] [addr1 addr2 addr3 ...]

Reads the display buffer at the address locations given.

Where:

8 16 32	The unit size: 8-bit (bytes), 16-bit (words), 32-bit (dwords). If a unit size is not specified, this command uses the unit size from the last Read command performed. If no previous Read command has been issued, the unit size defaults to 8-bit.
addr	The address to read data from. Multiple addresses can be given.

Run scriptfile

This command opens the file scriptfile and executes each line as if it were typed from the command prompt. For more information on scriptfiles, see **Section , “Script Files” on page 12.**

Where:

scriptfile The file containing 13A04PLAY commands

S[8|16|32] startaddr endaddr|len data1 [data2 data3 data4 ...]

Search the display buffer for the given data.

Where:

8 16 32	The unit size: 8-bit (bytes), 16-bit (words), 32-bit (dwords). If a unit size is not specified, this command uses the unit size from the last Search command performed. If no previous Search command has been issued, the unit size defaults to 8-bit.
startaddr	The starting address to begin the search from. Specifying a period (.) uses the same starting address as the last Search command performed.
endaddr len	Determines how many units of the display buffer will be searched through. A number without a prefix represents a physical ending address. If a “L” prefix is used, the number that follows represents <i>len</i> , or the number of bytes/words/dwords to be searched through. <i>Len</i> is based on the unit size. For example, 'L8' when the unit size is 16-bit would cause the Search command to search 8 words from the starting address.
data	The value(s) to search the display buffer for. Values can be combinations of 'text' or numbers. Numbers are assumed to be hexadecimal values unless otherwise specified with the correct suffix (binary = i, octal = o, decimal = t, hexadecimal = h). For example, 101i = 101 binary.

Show

Shows a test pattern on the display. The test pattern is based on current register settings and may not display correctly if the registers are not configured properly.

Use this command to display an image during testing. After adjusting a register value, use the show command to view the effect on the display.

U index [data]

Reads/writes data to the USB register at index. If no data is specified, the command reads and displays the contents from the USB register at index.

Where:

index

Index into the USB registers (hex).

data

The value to be written to the register. Numbers are assumed to be hexadecimal values unless otherwise specified with the correct suffix (binary = i, octal = o, decimal = t, hexadecimal = h).

For example, 101i = 101 binary.

UA

Reads and displays the contents of all the USB registers.

UX index [data]

This command automates the writes/reads into the indexed USB registers located at REG[4038h] and REG[403Ah]. Index represents the value that would be written into REG[4038h] if separate operations were done to access the index and data registers. If no data is specified, the command reads and displays the contents of the specified extended USB register.

Where:

index

Index into USB registers at REG[403Ah] (hex).

data

The value to be written to the indexed data register. Numbers are assumed to be hexadecimal values unless otherwise specified with the correct suffix (binary = i, octal = o, decimal = t, hexadecimal = h).

For example, 101i = 101 binary.

W[8|16|32] [startaddr [data1 data2 data3 data4 ...]]

Writes the given data sequence to the display buffer starting at startaddr location.

Where:

8 16 32	The unit size: 8-bit (bytes), 16-bit (words), 32-bit (dwords). If a unit size is not specified, this command uses the unit size from the last Write command performed. If no previous Write command has been issued, the unit size defaults to 8-bit.
startaddr	The starting address to write data to. Specifying a period (.) uses the same starting address as the last Write command performed.
data	Values to write to the display buffer. If no data is given, then this function enters MODIFY mode. This mode prompts the user with the address and it's current data. While in this mode, the user can type any of the following. <ul style="list-style-type: none"> - new values (in hex) - ENTER or SPACE - moves to the next memory location (if data is specified the previous memory location is updated, if no data is specified no change is made) - "-" - moves to the previous memory location - "Q" or "." - exits MODIFY mode

X[8|16|32] [index [data]]

Reads/writes data to the register at index. If no data is specified, the register is read and the contents are displayed.

Where:

8 16 32	The unit size: 8-bit (bytes), 16-bit (words), 32-bit (dwords). If a unit size is not specified, this command uses the unit size from the last X command performed. If no previous X command has been issued, the unit size defaults to 8-bit.
index	Index into the registers (hex).
data	The value to be written to the register. Numbers are assumed to be hexadecimal values unless otherwise specified with the correct suffix (binary = i, octal = o, decimal = t, hexadecimal = h). For example, 101i = 101 binary.

XA

Reads and displays the contents of all the S1D13A04 registers.

?

Displays the help screen. The help screen contains a summary of all available commands.

13A04PLAY Example

1. Configure **13A04PLAY** using the utility **13A04CFG**. For further information on 13A04CFG, see the *13A04CFG User Manual*, document number X37A-B-001-xx.
2. Type **13A04PLAY** to start the program.
3. Type **?** for help.
4. Type **i** to initialize the registers.
5. Type **xa** to display the contents of the registers.
6. Type **x 80** to read register 80h.
7. Type **x 80 10** to write 10h to register 80h.
8. Type **f 0 ffff aa** to fill the first 10000h bytes of the display buffer with AAh.
9. Type **d 0 ff** to read the first 100h bytes of the display buffer.
10. Type **show** to display a test pattern.
11. Type **m** to display current mode information.
12. Type **m 2** to set the color depth to 2 bpp.
13. Type **show** to display a test pattern.
14. Type **q** to exit the program.

Script Files

13A04PLAY can be controlled by a script file. This is useful when:

- there is no display to monitor command keystroke accuracy.
- various registers must be quickly changed to view results.

A script file is an ASCII text file with one 13A04PLAY command per line. Script files can be executed from within 13A04PLAY using the Run command (e.g. = run dumpregs.scr). Alternately, the script file can be executed from the OS command prompt. On a PC platform, a typical script command line might be:

```
13A04PLAY run dumpregs.scr > results
```

This causes the file **dumpregs.scr** to be interpreted as commands by 13A04PLAY and the results to be redirected to the file **results**.

Example 1: Create a script file that reads all registers.

```
; This file initializes the S1D13A04 and reads the registers.  
; Note: after a semicolon (;), all characters on a line are ignored.
```

```
; Initialize the S1D13A04  
i
```

```
; Read all registers  
xa
```

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S1D13A04 LCD/USB Companion Chip

13A04BMP Demonstration Program

Document Number: X37A-B-003-01

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13A04BMP

13A04BMP is a demonstration utility used to show the S1D13A04 display capabilities by rendering bitmap images on the display device. The program displays a bitmap stored in Windows BMP file format and then exits. 13A04BMP supports SviweView™ (90°, 180°, and 270° hardware rotation of the display image).

13A04BMP is designed to operate on a personal computer (PC) within a 32-bit environment only (Windows® 9x/NT/ME/2000). Other embedded platforms are not supported due to the possible lack of system memory or structured file system.

The 13A04BMP demonstration utility must be configured to work with your hardware configuration. The program 13A04CFG can be used to configure 13A04BMP. For further information on 13A04CFG, refer to the *13A04CFG Users Manual*, document number X37A-B-001-xx.

S1D13A04 Supported Evaluation Platforms

13A04BMP supports the following S1D13A04 evaluation platforms:

- PC with an Intel 80x86 processor running Windows 9x/NT/ME/2000.

Note

The 13A04BMP source code may be modified by the OEM to support other evaluation platforms.

Installation

Copy the file **13A04bmp.exe** to a directory in the path (e.g. PATH=C:\S1D13A04).

Usage

At the prompt, type:

13A04bmp bmpfile [/?]

Where:

bmpfile Specifies filename of the windows format bmp image to be displayed.

/? Displays the help message.

Note

13A04BMP displays the bmpfile image and returns to the prompt.

13A04BMP Examples

To display a .bmp image in the main window on an LCD, type the following:

13A04bmp bmpfile.bmp

Note

To display a bmpfile using SwivelView, configure 13A04bmp.exe for the selected SwivelView mode using the configuration program 13A04CFG. For further information on 13A04CFG, see the *13A04CFG User Manual*, document number X37A-B-001-xx.

Comments

- 13A04BMP displays only Windows BMP format images.
- A 24-bit true color bitmap is displayed at a color depth of 16 bit-per-pixel.
- Only the green component of the image is seen on a monochrome panel.
- 13A04BMP does not perform any image translations. The image to display must be the desired dimensions and color depth.

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S1D13A04 LCD/USB Companion Chip

Wind River WindML v2.0 Display Drivers

Document Number: X37A-E-002-01

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Wind River WindML v2.0 DISPLAY DRIVERS

The Wind River WindML v2.0 display drivers for the S1D13A04 LCD/USB Companion Chip are intended as “reference” source code for OEMs developing for Wind River’s WindML v2.0. The driver package provides support for both 8 and 16 bit-per-pixel color depths. The source code is written for portability and contains functionality for most features of the S1D13A04. Source code modification is required to produce a smaller, more efficient driver for mass production.

The WindML display drivers are designed around a common configuration include file called **mode0.h** which is generated by the configuration utility 13A04CFG. This design allows for easy customization of display type, clocks, decode addresses, rotation, etc. by OEMs. For further information on 13A04CFG, see the *13A04CFG Configuration Program User Manual*, document number X37A-B-001-xx.

Note

The WindML display drivers are provided as “reference” source code only. They are intended to provide a basis for OEMs to develop their own drivers for WindML v2.0.

These drivers are not backwards compatible with UGL v1.2. For information on UGL v1.2 display drivers, contact us via email at erdvdcsw_info@erd.epson.com.

This document and the source code for the WindML display drivers is updated as appropriate. Please check the Epson Research and Development website at www.erd.epson.com for the latest revisions before beginning any development.

We appreciate your comments on our documentation. Please contact us via email at documentation@erd.epson.com.

Building a WindML v2.0 Display Driver

The following instructions produce a bootable disk that automatically starts the UGL demo program. These instructions assume that Wind River's Tornado platform is already installed.

Note

For the example steps where the drive letter is given as "x:". Substitute "x" with the drive letter that your development environment is on.

1. Create a working directory and unzip the WindML display driver into it.

From a command prompt or GUI interface create a new directory (e.g. x:\13A04).

Unzip the file **13A04windml.zip** to the newly created working directory. The files will be unzipped to the directories "x:\13A04\8bpp" and "x:\13A04\16bpp".

2. Configure for the target execution model.

This example build creates a VxWorks image that fits onto and boots from a single floppy diskette. In order for the VxWorks image to fit on the disk certain modifications are required.

Replace the file "x:\Tornado\target\config\pcPentium\config.h" with the file "x:\13A04\8bpp\File\config.h" (or "x:\13A04\16bpp\File\config.h"). The new **config.h** file removes networking components and configures the build image for booting from a floppy disk.

Note

Rather than simply replacing the original **config.h** file, rename it so the file can be kept for reference purposes.

3. Build a boot ROM image.

From the Tornado tool bar, select Build -> Build Boot ROM. Select "pcPentium" as the BSP and "bootrom_uncmp" as the image.

4. Create a bootable disk (in drive A:).

From a command prompt change to the directory "x:\Tornado\host\x86-win32\bin" and run the batch file **torvars.bat**. Next, change to the directory "x:\Tornado\target\config\pcPentium" and type:

```
mkboot a: bootrom_uncmp
```

5. If necessary, generate a new **mode0.h** configuration file.

The file **mode0.h** contains the register values required to set the screen resolution, color depth (bpp), display type, rotation, etc. The **mode0.h** file included with the drivers, may not contain applicable values and must be regenerated. The configuration program 13A04CFG can be used to build a new **mode0.h** file. If building for 8 bpp, place the new **mode0.h** file in the directory “x:\13A04\8bpp\File”. If building for 16 bpp, place the new **mode0.h** file in “x:\13A04\16bpp\File”.

Note

Mode0.h should be created using the configuration utility 13A04CFG. For more information on 13A04CFG, see the *13A04CFG Configuration Program User Manual*, document number X37A-B-001-xx available at www.erd.epson.com.

6. Build the WindML v2.0 library.

From a command prompt change to the directory “x:\Tornado\host\x86-win32\bin” and run the batch file **torvars.bat**. Next, change to the directory “x:\Tornado\target\src\ugl” and type the command:

```
make CPU=PENTIUM ugl
```

7. Open the S1D13A04 workspace.

From the Tornado tool bar, select File->Open Workspace...->Existing->Browse... and select the file “x:\13A04\8bpp\13A03.wsp” (or “x:\13A04\16bpp\13A04.wsp”).

8. Add support for single line comments.

The WindML v2.0 display driver source code uses single line comment notation, “//”, rather than the ANSI conventional comments, “/*...*/”.

To add support for single line comments follow these steps:

- a. In the Tornado “Workspace Views” window, click on the “Builds” tab.
- b. Expand the “8bpp Builds” (or “16bpp Builds”) view by clicking on the “+” next to it. The expanded view will contain the item “default”. Right-click on “default” and select “Properties...”. A “Properties:” window will appear.
- c. Select the “C/C++ compiler” tab to display the command switches used in the build. Remove the “-ansi” switch from the line that contains “-g -mpentium -ansi -nostdinc -DRW_MULTI_THREAD”.
(Refer to GNU ToolKit user's guide for details)

9. Compile the VxWorks image.

Select the “Builds” tab in the Tornado “Workspace Views” window.

Right-click on “8bpp files” (or “16bpp files”) and select “Dependencies...”. Click on “OK” to regenerate project file dependencies for “All Project files”.

Right-click on “8bpp files” (or “16bpp files”) and select “ReBuild All(vxWorks)” to build VxWorks.

10. Copy the VxWorks file to the diskette.

From a command prompt or through the Windows interface, copy the file “x:\13A04\8bpp\default\vxWorks” (or “x:\13A04\16bpp\default\vxWorks”) to the bootable disk created in step 4.

11. Start the VxWorks demo.

Boot the target PC with the VxWorks bootable diskette to run the UGL demo program automatically.

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S1D13A04 LCD/USB Companion Chip

Linux Console Driver

Document Number: X37A-E-004-01

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Linux Console Driver

The Linux console driver for the S1D13A04 LCD/USB Companion Chip is intended as “reference” source code for OEMs developing for Linux, and supports 4, 8, and 16 bit-per-pixel color depths. The source code is written for portability and contains functionality for most features of the S1D13A04. Source code modification is required to provide a smaller driver for mass production.

A Graphical User Interface (GUI) such as Gnome can obtain the frame buffer address from this driver allowing the Linux GUI the ability to update the display.

The console driver is designed around a common configuration include file called **s1d13A04.h**, which is generated by the configuration utility 13A04CFG. This design allows for easy customization of display type, clocks, decode addresses, rotation, etc. by OEMs. For further information on 13A04CFG, see the *13A04CFG Configuration Program User Manual*, document number X37A-B-001-xx.

Note

The Linux console driver is provided as “reference” source code only. The driver is intended to provide a basis for OEMs to develop their own drivers for Linux.

This document and the source code for the Linux console drivers are updated as appropriate. Please check the Epson Research and Development website at www.erd.epson.com for the latest revisions or before beginning any development.

We appreciate your comments on our documentation. Please contact us via e-mail at documentation@erd.epson.com.

Building the Console Driver for Linux Kernel 2.2.x

Follow the steps below to construct a copy of the Linux operating system using the S1D13A04 as the console display device. These instructions assume that the GNU development environment is installed and the user is familiar with GNU and the Linux operating system.

1. Acquire the Linux kernel source code.

You can obtain the Linux kernel source code from your Linux supplier or download the source from: <ftp://ftp.kernel.org>.

The S1D13A04 reference driver requires Linux kernel 2.2.x. The example S1D13A04 reference driver available on www.erd.epson.com was built using Red Hat Linux 6.1, kernel version 2.2.17.

For information on building the kernel refer to the readme file at:
<ftp://ftp.linuxberg.com/pub/linux/kernel/README>

Note

Before continuing with modifications for the S1D13A04, you should ensure that you can build and start the Linux operating system.

2. Unzip the console driver files.

Using a zip file utility, unzip the S1D13A04 archive to a temporary directory. (e.g. /tmp)

When completed the following files should be located in the temporary directory.

s1d13xxfb.c
s1d13A04.h
Config.in
fbmem.c
fbcon-cfb4.c, and
Makefile

3. Copy the console driver files to the build directory.

Copy the following files to the directory `/usr/src/linux/drivers/video`.

`/tmp/s1d13xxfb.c` and
`/tmp/s1d13A04.h`

Copy the remaining source files into the directory `/usr/src/linux/drivers/video` replacing the files of the same name.

`/tmp/Config.in`
`/tmp/fbmem.c`
`/tmp/fbcon-cfb4.c`, and
`/tmp/Makefile`

If your kernel version is not 2.2.17, or you want to retain greater control of the build process, use a text editor to cut and paste the sections dealing with the Epson driver in the corresponding files of the same names.

4. Modify s1d13A04.h

The file s1d13A04.h contains the register values required to set the screen resolution, color depth (bpp), display type, active display (LCD), display rotation, etc.

Before building the console driver, refer to the descriptions in the file s1d13A04.h for the default settings of the console driver. If the default does not match the configuration you are building for then s1d13A04.h must be regenerated with the correct information.

Use the program 13A04CFG to generate the required header file. For information on how to use 13A04CFG, refer to the *13A04CFG Configuration Program User Manual*, document number X37A-B-001-xx, available at www.erd.epson.com

After selecting the desired configuration, choose “File->Export” and select the “C Header File for S1D13A04 Generic Drivers” option. Save the new configuration as s1d13A04.h in the /usr/src/linux/drivers/video, replacing the original configuration file.

5. Configure the video options.

From the command prompt in the directory /usr/src/linux run the command:

make menuconfig

This command will start a text based interface which allows the selection of build time parameters. From the text interface under “Console drivers” options, select:

- “Support for frame buffer devices”
- “Epson LCD/CRT controllers support”
- “S1D13A04 support”
- “Advanced low level driver options”
- “xBpp packed pixels support” *

* where x is the color depth being compile for.

If you are using the Epson S5U13A04B00C evaluation board then you must also select:

- “Epson PCI Bridge adapter support”

Once you have configured the kernel options, save and exit the configuration utility.

6. Compile and install the kernel.

Build the kernel with the following sequence of commands.

- make dep
- make clean
- make bzImage
- /sbin/lilo (if running lilo)

7. Boot to the Linux operating system.

If you are using lilo (Linux Loader), modify the lilo configuration file as discussed in the kernel build README file. If there were no errors during the build, from the command prompt run:

```
lilo
```

and reboot your system.

Note

In order to use the S1D13A04 console driver with X server, you need to configure the X server to use the FBDEV device. The information on the necessary files and instructions for this process is available on the Internet at www.xfree86.org

Building the Console Driver for Linux Kernel 2.4.x

Follow the steps below to construct a copy of the Linux operating system using the S1D13A04 as the console display device. These instructions assume that the GNU development environment is installed and the user is familiar with GNU and the Linux operating system.

1. Acquire the Linux kernel source code.

You can obtain the Linux kernel source code from your Linux supplier or download the source from: <ftp://ftp.kernel.org>.

The S1D13A04 reference driver requires Linux kernel 2.4.x or greater. The example S1D13A04 reference driver available on www.erd.epson.com was built using Red Hat Linux 6.1, kernel version 2.4.5.

For information on building the kernel refer to the readme file at:
<ftp://ftp.linuxberg.com/pub/linux/kernel/README>

Note

Before continuing with modifications for the S1D13A04, you should ensure that you can build and start the Linux operating system.

2. Unzip the console driver files.

Using a zip file utility, unzip the S1D13A04 archive to a temporary directory. (e.g. /tmp)

When completed the files:

Config.in
fbmem.c
fbcon-cfb4.c
Makefile

should be located in the temporary directory (/tmp), and the files:

Makefile
s1d13xxfb.c
s1d13A04.h

should be located in a sub-directory called epson within the temporary directory (/tmp/epson).

3. Copy the console driver files to the build directory. Make the directory `/usr/src/linux/drivers/video/epson`.

Copy the files

`/tmp/epson/s1d13xxfb.c`

`/tmp/epson/s1d13A04.h`

`/tmp/epson/Makefile`

to the directory `/usr/src/linux/drivers/video/epson`.

Copy the remaining source files

`/tmp/Config.in`

`/tmp/fbmem.c`

`/tmp/fbcon-cfb4.c`

`/tmp/Makefile`

into the directory `/usr/src/linux/drivers/video` replacing the files of the same name.

If your kernel version is not 2.4.5 or you want to retain greater control of the build process then use a text editor and cut and paste the sections dealing with the Epson driver in the corresponding files of the same names.

4. Modify `s1d13A04.h`

The file `s1d13A04.h` contains the register values required to set the screen resolution, color depth (bpp), display type, active display (LCD/CRT), display rotation, etc.

Before building the console driver, refer to the descriptions in the file `s1d13A04.h` for the default settings of the console driver. If the default does not match the configuration you are building for then `s1d13A04.h` will have to be regenerated with the correct information.

Use the program `13A04CFG` to generate the required header file. For information on how to use `13A04CFG`, refer to the *13A04CFG Configuration Program User Manual*, document number `X37A-B-001-xx`, available at www.erd.epson.com

After selecting the desired configuration, choose “File->Export” and select the “C Header File for S1D13A04 Generic Drivers” option. Save the new configuration as `s1d13A04.h` in the `/usr/src/linux/drivers/video`, replacing the original configuration file.

5. Configure the video options.

From the command prompt in the directory `/usr/src/linux` run the command:
`make menuconfig`

This command will start a text based interface which allows the selection of build time parameters. From the options presented select:

- “Code maturity level” options
 - “Prompt for development and/or incomplete drivers”
- “Console drivers” options
 - “Frame-buffer support”
 - “Support for frame buffer devices (EXPERIMENTAL)”
 - “EPSON LCD/CRT/TV controller support”
 - “EPSON S1D13A04 Support”
 - “Advanced low-level driver options”
 - “xbpp packed pixels support” *

* where x is the color depth being compiled for.

If you are using the Epson S5U13A04B00C evaluation board then you must also select:

- “Epson PCI Bridge adapter support”

Once you have configured the kernel options, save and exit the configuration utility.

6. Compile and install the kernel

Build the kernel with the following sequence of commands:

- `make dep`
- `make clean`
- `make bzImage`
- `/sbin/lilo` (if running lilo)

7. Boot to the Linux operating system

If you are using lilo (Linux Loader), modify the lilo configuration file as discussed in the kernel build README file. If there were no errors during the build, from the command prompt run:

- `lilo`

and reboot your system.

Note

In order to use the S1D13A04 console driver with X server, you need to configure the X server to use the FBDEV device. A good place to look for the necessary files and instructions on this process is on the Internet at www.xfree86.org

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S1D13A04 LCD/USB Companion Chip

QNX Photon v2.0 Display Driver

Document Number: X37A-E-005-01

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QNX Photon v2.0 Display Driver

The Photon v2.0 display driver for the S1D13A04 LCD/USB Companion Chip is intended as “reference” source code for OEMs developing for QNX platforms. The driver package provides support for 8 and 16 bit-per-pixel color depths. The source code is written for portability and contains functionality for most features of the S1D13A04. Source code modification is required to provide a smaller driver for mass production.

The current revision of the driver is designed for use with either QNX RTP or QNX4 from the latest product CD (Dec. 99).

The Photon v2.0 display driver is designed around a common configuration include file called **S1D13A04.h**, which is generated by the configuration utility 13A04CFG. This design allows for easy customization of display type, clocks, decode addresses, rotation, etc. by OEMs. For further information on 13A04CFG, see the *13A04CFG Configuration Program User Manual*, document number X37A-B-001-xx.

Note

The QNX display drivers are provided as “reference” source code only. They are intended to provide a basis for OEMs to develop their own drivers for QNX Photon v2.0.

This document and the source code for the QNX display drivers are updated as appropriate. Please check the Epson Research and Development website at www.erd.epson.com for the latest revisions before beginning any development.

We appreciate your comments on our documentation. Please contact us via e-mail at documentation@erd.epson.com.

Building the Photon v2.0 Display Driver

The following steps build the Photon v2.0 display driver and integrate it into the QNX operating system. These instructions assume the QNX developer environment is correctly installed and the developer is familiar with building for the QNX operating system.

Unpack the Graphics Driver Development Kit Archive

1. Install the QNX ddk package using the Package Manager utility.

For information about the Drivers Development Kit contact QNX directly.

2. Once the ddk package is installed, copy the directory tree /usr/src/gddk_v1.0 into the Project directory.
3. Change directory to Project/gddk_1.0/devg.
4. Unpack the display driver files using the commands:

```
#gunzip S1D13A04.tar.gz
```

```
#tar -xvf S1D13A04.tar
```

This unpacks the files into the directory Project/gddk_1.0/devg/S1D13A04.

Configure the Driver

The files **s1d13A04_16.h** and **s1d13A04_8.h** contain register values required to set the screen resolution, color depth (bpp), display type, rotation, etc. The **s1d13A04.h** file included with the drivers may not contain applicable values and must be regenerated. The configuration program 13A04CFG can be used to build new **s1d13A04_16.h** and **s1d13A04_8.h** files.

Note

S1d13A04.h should be created using the configuration utility 13A04CFG. For more information on 13A04CFG, see the *13A04CFG Configuration Program User Manual*, document number X37A-B-001-xx available at www.erd.epson.com.

Build the Driver

The first time the driver is built, the following command ensures that all drivers and required libraries are built. At the root of the Project source tree, type **make**.

Note

To build drivers for X86 NTO type 'OSLIST=nto CPULIST=x86 make'.

Further builds do not require all libraries to be re-built. To build only the S1D13A04 display driver, change to the directory gddk_1.0/devg/s1d13A04 and type **make**.

Installing the Driver

The build step produces two library images:

- lib/disputil/nto/x86/libdisputil.so
- lib/disputil/nto/x86/libffb.so

For the loader to locate them, the files need to be renamed and copied to the lib directory.

1. Rename libdisputil.so to libdisputil.so.1 and libffb.so to libffb.so.1.
2. Copy the files new files libdisputil.so.1 and libffb.so.1 to the directory /usr/lib.
3. Copy the file devg-S1D13A04.so to the /lib/dll directory.

Note

To locate the file devg-S1D13A04.so, watch the output of the ‘true’ command during the makefile build.

Modify the trap file graphics-modes in the /etc/system/config directory by inserting the following lines at the top of the file.

```
io-graphics -dldevg-S1D13A04.so -gWxHxC -IO -d0x0,0x0;#640,480,8 Epson
```

```
io-graphics -dldevg-S1D13A04.so -gWxHxC -IO -d0x0,0x0;#640,480,16 Epson
```

Where:

- W is the configured width of the display
- H is the configured height of the display
- C is the color depth in bpp (either 8 or 16)

Run the Driver

Note

For the remaining steps the S5U13A04B00C evaluation board must be installed on the test platform.

It is recommended that the driver be verified **before starting QNX with the S1D13A04 as the primary display**. To verify the driver, type the following command at the root of the Project source tree (gddk_1.0 directory).

```
util/bench/nto/x86/o/devg-bench -dldevg/S1D13A04/nto/x86/dll/devg-S1D13A04.so  
-mW,H,C,F -d0x0,0x0
```

Where:

W is the configured width of the display
H is the configured height of the display
C is the color depth in bpp (either 8 or 16)
F is the configured frame rate

This command starts the bench utility which will initialize the driver as the secondary display and exercise the drivers main functions. If the display appears satisfactory, restart QNX Photon and the restart will result in the S1D13A04 display driver becoming the primary display device.

Comments

- To restore the display driver to the default, comment out changes made to the trap file crt.\$NODE.

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S1D13A04 LCD/USB Companion Chip

Windows® CE 3.x Display Driver

Document Number: X37A-E-006-01

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WINDOWS® CE 3.0 DISPLAY DRIVER

Windows CE v3.0 display driver for the S1D13A04 LCD/USB Companion Chip is intended as “reference” source code for OEMs developing for the Microsoft Window CE platform. The driver supports 4, 8 and 16 bit-per-pixel color depths in landscape mode (no rotation), and 8 and 16 bit-per-pixel color depths in SwivelView™ 90 degree, 180 degree and 270 degree modes. The source code is written for portability and contains functionality for most features of the S1D13A04. Source code modification is required to provide a smaller driver for mass production.

The Windows CE v3.0 display driver is designed around a common configuration include file called **mode0.h**, which is generated by the configuration utility 13A04CFG. This design allows for easy customization of display type, clocks, decode addresses, rotation, etc. by OEMs. For further information on 13A04CFG, see the *13A04CFG Configuration Program User Manual*, document number X37A-B-001-xx.

Note

The Windows CE display driver is provided as “reference” source code only. They are intended to provide a basis for OEMs to develop their own drivers for Microsoft Windows CE v3.0.

This document and the source code for the Windows CE v3.0 driver is updated as appropriate. Before beginning any development, please check the Epson Research and Development Website at www.erd.epson.com for the latest revisions.

We appreciate your comments on our documentation. Please contact us via email at documentation@erd.epson.com.

Example Driver Builds

The following section describes how to build the Windows CE display driver for Windows CE Platform Builder 3.00 using the GUI interface.

Build for CEPC (X86) on Windows CE Platform Builder 3.00 using the GUI Interface

1. Install Microsoft Windows 2000 Professional, or Windows NT Workstation version 4.0 with Service Pack 5 or later.
2. Install Platform Builder 3.00.
3. Start Platform Builder by double-clicking on the Microsoft Windows CE Platform Builder icon.
4. Create a new project.
 - a. Select File | New.
 - b. In the dialog box, select the Platforms tab.
 - c. In the platforms dialog box, select “WCE Platform”, set a location for the project (such as x:\myproject), set the platform name (such as myplatform), and set the processor to “Win32 (WCE x86)”.
 - d. Click the OK button.
 - e. In the dialog box “WCE Platform - Step 1 of 2”, select CEPC.
 - f. Click the Next button.
 - g. In the dialog box “WCE Platform - Step 2 of 2”, select Maximum OS (Maxall).
 - h. Click the Finish button.
 - i. In the dialog box “New Platform Information”, click the OK button.
5. Set the active configuration to “Win32 (WCE x86) Release”.
 - a. From the Build menu, select “Set Active Configuration”.
 - b. Select “MYPLATFORM - Win32 (WCE x86) Release”.
 - c. Click the OK button.

6. Add the environment variable DDI_S1D13A04.
 - a. From the Platform menu, select “Settings”.
 - b. Select the “Environment” tab.
 - c. In the Variable box, type “DDI_S1D13A04”.
 - d. In the Value box, type “1”.
 - e. Click the Set button.
 - f. Click the OK button.
7. Create a new directory S1D13A04, under x:\wince300\platform\cepc\drivers\display, and copy the S1D13A04 driver source code into this new directory.
8. Add the S1D13A04 driver component.
 - a. From the Platform menu, select “Insert | User Component”.
 - b. Set “Files of type:” to “All Files (*.*)”.
 - c. Select the file **x:\wince300\platform\cepc\drivers\display\S1D13A04\sources**.
 - d. Click the OK button.
 - e. In the “User Component Target File” dialog box, select browse and then select the path and the file name of “sources” as in step c.
 - f. Click the OK button.
9. Delete the component “ddi_flat”.
 - a. In the Platform window, select the ComponentView tab.
 - b. Show the tree for MYPLATFORM components by clicking on the ‘+’ sign at the root of the tree.
 - c. Right-click on the ddi_flat component.
 - d. Select “Delete”.
 - e. From the File menu, select “Save Workspace”.

10. From the Platform window, click the ParameterView Tab. Show the tree for MY-PLATFORM Parameters by clicking the '+' sign at the root of the tree. Expand the the WINCE300 tree and then click on "Hardware Specific Files" and then double click "PLATFORM.BIB". Edit the file the PLATFORM.BIB file and make the following two changes:

- a. Insert the following text after the line "IF ODO_NODISPLAY !":

```
IF CEPC_DDI_S1D13A04
    ddi.dll $(_FLATRELEASEDIR)\S1D13A04.dll NK SH
ENDIF
```

- b. Find the section shown below, and insert the lines as marked:

```
IF CEPC_DDI_FLAT !
IF DDI_S1D13A04!           ;Insert this line
IF CEPC_DDI_S3VIRGE !
IF CEPC_DDI_CT655X !
IF CEPC_DDI_VGA8BPP !
IF CEPC_DDI_S3TRIO64 !
IF CEPC_DDI_ATI !
    ddi.dll $(_FLATRELEASEDIR)\ddi_flat.dll NK SH
ENDIF
ENDIF
ENDIF
ENDIF
ENDIF
ENDIF
ENDIF           ;Insert this line
ENDIF
```

11. Modify MODE0.H.

The file MODE0.H (located in x:\wince300\platform\cepc\drivers\display\S1D13A04) contains the register values required to set the screen resolution, color depth (bpp), display type, display rotation, etc.

Before building the display driver, refer to the descriptions in the file MODE0.H for the default settings of the console driver. If the default does not match the configuration you are building for then MODE0.H will have to be regenerated with the correct information.

Use the program 13A04CFG to generate the header file. For information on how to use 13A04CFG, refer to the *13A04CFG Configuration Program User Manual*, document number X37A-B-001-xx, available at www.erd.epson.com

After selecting the desired configuration, choose “File->Export” and select the “C Header File for S1D13A04 WinCE Driver” option. Save the new configuration as MODE0.H in the \wince300\platform\cepc\drivers\display, replacing the original configuration file.

12. From the Platform window, click the ParameterView Tab. Show the tree for MY-PLATFORM Parameters by clicking the ‘+’ sign at the root of the tree. Expand the the WINCE300 tree and click on “Hardware Specific Files”, then double click “PLATFORM.REG”. Edit the file PLATFORM.REG to match the screen resolution, color depth, and rotation information in MODE0.H.

For example, the display driver section of PLATFORM.REG should be as follows when using a 320x240 LCD panel with a color depth of 8 bpp and a SwivelView mode of 0° (landscape):

```
; Default for EPSON Display Driver
; 320x240 at 8 bits/pixel, LCD display, no rotation
; Useful Hex Values
; 640=0x280 480=0x1E0 320=140 240=0xF0
[HKEY_LOCAL_MACHINE\Drivers\Display\S1D13A04]
“Width”=dword:140
“Height”=dword:F0
“Bpp”=dword:8
“Rotation”=dword:0
```

13. From the Build menu, select “Rebuild Platform” to generate a Windows CE image file (NK.BIN) in the project directory x:\myproject\myplatform\reldir\x86_release\nk.bin.

Installation for CEPC Environment

Once the NK.BIN file is built, the CEPC environment can be started by booting either from a floppy or hard drive configured with a Windows 9x operating system. The two methods are described below.

1. To start CEPC from a floppy drive:

- a. Create a bootable floppy disk.
- b. Copy HIMEM.SYS to the floppy disk and edit CONFIG.SYS on the floppy disk to contain only the following line:

```
device=a:\himem.sys
```

- c. Edit AUTOEXEC.BAT on the floppy disk to contain the following lines:

```
loadcepc /B:38400 /C:1 c:\nk.bin
```

- d. Search for **loadcepc.exe** in the Windows CE directories and copy **loadcepc.exe** to the bootable floppy disk.
- e. Copy NK.BIN to c:\.
- f. Boot the system from the bootable floppy disk.

2. To start CEPC from a hard drive:

- a. Search for **loadcepc.exe** in the Windows CE directories and copy **loadcepc.exe** to C:\.
- b. Edit CONFIG.SYS on the hard drive to contain only the following line:

```
device=c:\himem.sys
```

- c. Edit AUTOEXEC.BAT on the hard drive to contain the following lines:

```
loadcepc /B:38400 /C:1 c:\nk.bin
```

- d. Copy NK.BIN to c:\.
- e. Boot the system.

Configuration

There are several issues to consider when configuring the display driver. The issues cover debugging support, register initialization values and memory allocation. Each of these issues is discussed in the following sections.

Compile Switches

There are several switches, specific to the S1D13A04 display driver, which affect the display driver.

The switches are added or removed from the compile switches in the file SOURCES.

WINCEVER

This option is automatically set to the numerical version of WinCE for version 2.12 or later. If the environment variable, `_WINCEOSVER` is not defined, then `WINCEVER` will default to 2.11. The S1D13A04 display driver may test against this option to support different WinCE version-specific features.

EnablePreferVmem

This option enables the use of off-screen video memory. When this option is enabled, WinCE can optimize some BLT operations by using off-screen video memory to store images. You may need to disable this option if your off-screen video memory is limited.

ENABLE_CLOCK_CHIP

This option is used to enable support for the ICD2061A clock generator. This clock chip is used on the S5U13A04B00C evaluation board. The S1D13A04 display drivers can program the clock chip to support the frequencies required in the MODE tables.

If you are not using the S5U13A04B00C evaluation board, you should disable this option.

EpsonMessages

This debugging option enables the display of EPSON-specific debug messages. These debug messages are sent to the serial debugging port. This option should be disabled unless you are debugging the display driver, as they will significantly impact the performance of the display driver.

DEBUG_MONITOR

This option enables the use of the debug monitor. The debug monitor can be invoked when the display driver is first loaded and can be used to view registers, and perform a few debugging tasks. The debug monitor is still under development and is **UNTESTED**.

This option should remain disabled unless you are performing specific debugging tasks that require the debug monitor.

MonoPanel

This option is intended for the support of monochrome panels only.

The option causes palette colors to be grayscaled for correct display on a mono panel. For use with color panels this option should not be enabled.

DEBUG_BLT

This option enables special BLT debugging messages on the debugging serial port. This option, when enabled, will drastically impact display driver performance, and should only be used to track down failures in the BLT operations.

This option should be disabled unless doing BLT debugging.

Mode File

A second variable which will affect the finished display driver is the register configurations contained in the mode file.

The MODE tables (contained in files MODE0.H, MODE1.H, MODE2.H . . .) contain register information to control the desired display mode. The MODE tables must be generated by the configuration program 13A04CFG.EXE. The display driver comes with one example MODE table:

- MODE0.H - LCD 8-bit STN color, 320x240, 8bpp, 70Hz

By default, only MODE0.H is used by the display driver. New mode tables can be created using the 13A04CFG program. Edit the #include section of MODE.H to add the new mode table.

If you only support a single mode table, you do not need to add any information to the WinCE registry. If, however, you support more than one display mode, you should create registry values (see below) that will establish the initial display mode. If your display driver contains multiple mode tables, and if you do not add any registry values, the display driver will default to the first mode table in your list.

To select which display mode the display driver should use upon boot, add the following lines to your PLATFORM.REG file:

```
[HKEY_LOCAL_MACHINE\Drivers\Display\S1D13A04]
```

“Width”=dword:140
“Height”=dword:F0
“Bpp”=dword:8
“Rotation”=dword:0

Note that all dword values are in hexadecimal, therefore 140h = 320 and F0h = 240. When the display driver starts, it will read these values in the registry and attempt to match a mode table against them. All values must be present and valid for a match to occur, otherwise the display driver will default to the first mode table in your list.

A WinCE desktop application (or control panel applet) can change these registry values, and the display driver will select a different mode upon warmboot. This allows the display driver to support different display configurations and/or orientations. An example application that controls these registry values will be made available upon the next release of the display driver; preliminary alpha code is available by special request.

Resource Management Issues

The Windows CE 3.0 OEM must deal with certain display driver issues relevant to Windows CE 3.0. These issues require the OEM balance factors such as: system vs. display memory utilization, video performance, and power off capabilities.

The section “Simple Display Driver Configuration” on page 13 provides a configuration which should work with most Windows CE platforms. This section is only intended as a means of getting started. Once the developer has a functional system, it is recommended to optimize the display driver configuration as described below in “Description of Windows CE Display Driver Issues”.

Description of Windows CE Display Driver Issues

The following are some issues to consider when configuring the display driver to work with Windows CE:

1. When Windows CE enters the Suspend state (power-off), the LCD controller and display memory may lose power, depending on how the OEM sets up the system. If display memory loses power, all images stored in display memory are lost.

If power-off/power-on features are required, the OEM has several options:

- If display memory power is turned off, add code to the display driver to save any images in display memory to system memory before power-off, and add code to restore these images after power-on.
- If display memory power is turned off, instruct Windows CE to redraw all images upon power-on. Unfortunately it is not possible to instruct Windows CE to redraw any off-screen images, such as icons, slider bars, etc., so in this case the OEM must also configure the display driver to never use off-screen memory.
- Ensure that display memory never loses power.

2. Using off-screen display memory significantly improves display performance. For example, slider bars appear more smooth when using off-screen memory. To enable or disable the use of off-screen memory, edit the file: `x:\wince300\platform\cepc\drivers\display\S1D13A04\sources`. In `SOURCES`, there is a line which, when uncommented, will instruct Windows CE to use off-screen display memory (if sufficient display memory is available):

```
CDEFINES=$(CDEFINES) -DEnablePreferVmem
```

3. In the file `PROJECT.REG` under CE 3.0, there is a key called `PORepaint` (search the Windows CE directories for `PROJECT.REG`). `PORepaint` is relevant when the Suspend state is entered or exited. `PORepaint` can be set to 0, 1, or 2 as described below:

- a. `PORepaint=0`

- This mode tells Windows CE not to save or restore display memory on suspend or resume.
- Since display data is not saved and not repainted, this is the FASTEST mode.
- Main display data in display memory must NOT be corrupted or lost on suspend. The memory clock must remain running.
- Off-screen data in display memory must NOT be corrupted or lost on suspend. The memory clock must remain running.
- This mode cannot be used if power to the display memory is turned off.

- b. `PORepaint=1`

- This is the default mode for Windows CE.
- This mode tells Windows CE to save the main display data to the system memory on suspend.
- This mode is used if display memory power is going to be turned off when the system is suspended, and there is enough system memory to save the image.
- Any off-screen data in display memory is LOST when suspended. Therefore off-screen memory usage must either be disabled in the display driver (i.e: `EnablePreferVmem` not defined in `SOURCES` file), or new OEM-specific code must be added to the display driver to save off-screen data to system memory when the system is suspended, and restored when resumed.
- If off-screen data is used (provided that the OEM has provided code to save off-screen data when the system suspends), additional code must be added to the display driver's surface allocation routine to prevent the display driver from allocating the "main memory save region" in display memory. When WinCE OS attempts to allocate a buffer to save the main display data, WinCE OS marks the allocation request as preferring display memory. We believe this is incorrect. Code must be added to prevent this specific allocation from being allocated in display memory - it MUST be allocated from system memory.
- Since the main display data is copied to system memory on suspend, and then simply copied back on resume, this mode is FAST, but not as fast as mode 0.

- c. `PORepaint=2`
 - This mode tells WinCE to not save the main display data on suspend, and causes WinCE to REPAINT the main display on resume.
 - This mode is used if display memory power is going to be turned off when the system is suspended, and there is not enough system memory to save the image.
 - Any off-screen data in display memory is LOST, and since there is insufficient system memory to save display data, off-screen memory usage MUST be disabled.
 - When the system is resumed, WinCE instructs all running applications to repaint themselves. This is the SLOWEST of the three modes.

Simple Display Driver Configuration

The following display driver configuration should work with most platforms running Windows CE. This configuration disables the use of off-screen display memory and forces the system to redraw the main display upon power-on.

1. This step disables the use of off-screen display memory.
Edit the file `x:\wince300\platform\cepc\drivers\display\S1D13A04\sources` and change the line

```
CDEFINES=$(CDEFINES) -DEnablePreferVmem
```

to

```
#CDEFINES=$(CDEFINES) -DEnablePreferVmem
```

2. This step causes the system to redraw the main display upon power-on. This step is only required if display memory loses power when Windows CE is shut down. If display memory is kept powered up (set the S1D13A04 in powersave mode), then the display data will be maintained and this step can be skipped.

Search for the file `PROJECT.REG` in your Windows CE directories, and inside `PROJECT.REG` find the key `PORepaint`. Change `PORepaint` as follows:

```
"PORepaint"=dword:2
```

Comments

- The display driver is CPU independent, allowing use of the driver for several Windows CE Platform Builder supported platforms. The file **s1dflat.cpp** will require editing for the correct values of PhysicalPortAddr, PhysicalVmemAddr, etc.
- The sample code defaults to a 320x240 8-bit color passive LCD panel in SwivelView 0° mode (landscape) with a color depth of 8 bpp. To support other settings, use 13A04CFG.EXE to generate a proper MODE0.H file. For further information, refer to the *13A04CFG Configuration Program User Manual*, document number X37A-B-001-xx.
- By default, the 13A04CFG program assumes PCI addressing for the S5U13A04B00C evaluation board. This means that the display driver will automatically locate the S1D13A04 by scanning the PCI bus (currently only supported for the CEPC platform). If you select the address option “Other” and fill in your own custom addresses for the registers and video memory, then the display driver will not scan the PCI bus and will use the specific addresses you have chosen.
- If you are running the display driver on hardware other than the S5U13A04B00C evaluation board, you must ensure that your hardware provides the correct clock frequencies for CLKI and CLKI2. 13A04CFG defaults to 50MHz for both CLKI and CLKI2.

On the evaluation board, the display driver will correctly program the clock chip to support the CLKI and CLKI2 frequencies. On customer hardware, you must ensure that the clocks you provide to all clock inputs match the settings you chose in the Clocks tab of the 13A04CFG program. For more information on setting the clocks, see the *13A04CFG User Manual*, document number X37A-B-001-xx.

If you run the S1D13A04 with a single clock source, make sure your clock sources for PCLK, BCLK, and MCLK are correctly set to use the correct clock input source. Also ensure that you enable the clock dividers as required for different display hardware.

- If you are using 13A04CFG.EXE to produce multiple MODE tables, make sure you change the Mode Number setting for each mode table you generate. The display driver supports multiple mode tables, but only if each mode table has a unique mode number. For more information on setting the mode number, see the *13A04CFG User Manual*, document number X37A-B-001-xx.
- The 13A04CFG program assumes you are using the S5U13A04B00C evaluation board, and defaults the Panel Power control to GPIO0. 13A04CFG allows you to change the GPIO pin used to control panel power, or to disable the use of GPIO pins altogether. If this is changed from the default, your driver will no longer be able to control panel power on the S5U13A04B00C evaluation board, and your panel may not be powered up correctly.
- At this time, the driver has been tested on the x86 CPUs and have been run with Platform Builder v3.00.

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S1D13A04 LCD/USB Companion Chip

Windows® CE 3.x USB Driver

Document Number: X37A-E-007-01

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WINDOWS® CE 3.0 USB DRIVER

The Windows CE v3.0 USB driver for the S1D13A04 LCD/USB Companion Chip is a client driver, which supports Microsoft ActiveSync 3.1. This driver is intended as “reference” source code for OEMs developing for the Microsoft Window CE platform and provide a basis for OEMs to develop their own drivers.

This document and the source code for the Windows CE v3.0 USB driver is updated as appropriate. Before beginning any development, please check the Epson Research and Development Website at www.erd.epson.com for the latest revisions.

We appreciate your comments on our documentation. Please contact us via email at documentation@erd.epson.com.

Example Driver Builds

Build for CEPC (X86) on Windows CE Platform Builder 3.00 using the GUI Interface

1. Install Microsoft Windows 2000 Professional or Windows 98.
2. Install Platform Builder 3.00.
3. Start Platform Builder by double-clicking on the Microsoft Windows CE Platform Builder icon.
4. Create a new project.
 - a. Select File | New.
 - b. In the dialog box, select the Platforms tab.
 - c. In the Platforms dialog box:
 - select “WCE Platform”
 - set a location for the project (such as **x:\myproject**)
 - set the platform name (such as myplatform)
 - set the processor to “Win32 (WCE x86)”
 - d. Click the OK button.
 - e. In the WCE Platform - Step 1 of 2 dialog box, select “CEPC”.
 - f. Click the Next button.
 - g. In the WCE Platform - Step 2 of 2 dialog box, select “Maximum OS (Maxall)”.
 - h. Click the Finish button.
 - i. In the New Platform Information dialog box, click the OK button.
5. Set the active configuration to “Win32 (WCE x86) Release”.
 - a. From the Build menu, select “Set Active Configuration”.
 - b. Select “MYPLATFORM - Win32 (WCE x86) Release”.
 - c. Click the OK button.
6. Add the environment variable USB_S1D13A04.
 - a. From the Platform menu, select “Settings”.
 - b. Select the “Environment” tab.
 - c. In the Variable box, type “USB_S1D13A04”.
 - d. In the Value box, type “1”.
 - e. Click the Set button.
 - f. Click the OK button.

7. Create a new directory 13A04USB, under **x:\wince300\platform\cepc\drivers**, and copy the 13A04USB driver source code into this new directory.
8. Add the 13A04USB driver component.
 - a. From the Platform menu, select “Insert | User Component”.
 - b. Set Files of type: to “All Files (*.*)”.
 - c. Select the file **x:\wince300\platform\cepc\drivers\13A04USB\sources**.
 - d. Click the OK button.
 - e. In the User Component Target File dialog box, select browse and then select the path/filename of the file **sources**.
 - f. Click the OK button.
9. From the Platform window, click the Parameter View Tab. Show the tree for MY-PLATFORM Parameters by clicking on the ‘+’ sign at the root of the tree. Expand the the WINCE300 tree and then click the “Hardware Specific Files” and then double click the “PLATFORM.BIB”. Edit the file **platform.bib** and make the following two changes:

Find the section shown below, and insert the line as marked:

IF IMGUSB

IF CEPC_UHCI

uhci.dll \$(_FLATRELEASEDIR)\uhci.dll NK SH

ENDIF

IF CEPC_OHCI

ohci.dll \$(_FLATRELEASEDIR)\ohci.dll NK SH

ENDIF

usbd.dll \$(_FLATRELEASEDIR)\usbd.dll NK SH

usbhid.dll \$(_FLATRELEASEDIR)\usbhid.dll NK SH

ENDIF

IF USB_S1D13A04 ;Insert this line

13a04usb.dll \$(_FLATRELEASEDIR)\13a04usb.dll NK SH ;Insert this line

ENDIF ;Insert this line

ENDIF

- From the Platform window, click the Parameter View Tab. Show the tree for MY-PLATFORM Parameters by clicking on the '+' sign at the root of the tree. Expand the WINCE300 tree and click the "Hardware Specific Files", then double click the "PLATFORM.REG". Insert the following section in the file **platform.reg** to include the settings for 13A04USB driver.

```
[HKEY_LOCAL_MACHINE\Drivers\BuiltIn\13A0XUSB]
"Dll"="13A04USB.dll"
"Prefix"="COM"
"Tsp"="Unimodem.dll"
"DeviceArrayIndex"=dword:1
"Order"=dword:2
"DeviceType"=dword:0
"FriendlyName"="S1D13A04 USB"
"DevConfig"=hex: 10,00, 00,00, 05,00,00,00, 10,01,00,00, 00,4B,00,00, 00,00, 08,
00, 00, 00,00,00,00
; "PhysicalAddress"=dword:0x08000000 ; for non-cepc environment only
; "IRQ"=dword:05 ; for non-cepc environment only
```

- From the Build menu, select "Rebuild Platform" to generate a Windows CE image file (**nk.bin**) in the project directory **x:\myproject\myplatform\reldir\x86_release\nk.bin**.

Installation and Execution from CEPC Environment

Once the **nk.bin** file is built, the CEPC environment can be started by booting either from a floppy (step 1) or a hard drive (step 2) configured with a Windows 9x operating system. Both methods are described below.

1. To start CEPC by booting from a floppy drive:
 - a. Create a bootable floppy disk.
 - b. Copy **himem.sys** to the floppy disk and edit **config.sys** on the floppy disk to contain only the following line:

device=a:\himem.sys
 - c. Edit **autoexec.bat** on the floppy disk to contain the following line:

loadcepc /B:38400 /C:1 c:\nk.bin
 - d. Search for **loadcepc.exe** in your Windows CE directories, and copy the file to the bootable floppy disk.
 - e. Copy **nk.bin** to c:\.
 - f. Boot the system from the bootable floppy disk.
2. To start CEPC by booting from a hard drive:
 - a. Search for **loadcepc.exe** in the Windows CE directories, and copy the file to C:\.
 - b. Edit **config.sys** on the hard drive to contain only the following line:

device=c:\himem.sys
 - c. Edit **autoexec.bat** on the hard drive to contain the following line:

loadcepc /B:38400 /C:1 c:\nk.bin
 - d. Copy **nk.bin** to C:\.
 - e. Boot the system
3. Install Windows 2000 Professional on a host machine.
4. Install ActiveSync 3.1 on the host machine.
5. Install the included **wceusbsh.sys** on the host machine, by following the procedures below:
 - a. Unzip the file **wceusbsh.zip** to a directory on your hard drive.
 - b. Find the file **wceusbsh.inf**.
 - c. Right click the “WCEUSBSH.INF” file icon.
 - d. Select Install.

6. Connect a USB cable from the USB device (S5U13A04B00C board) to the USB host machine.
7. Boot the Windows CE machine from a floppy (created in step 1) or from the hard drive (created in step 2).
8. From the Windows CE desktop:
 - click the Start button
 - click Run
 - click Browse.
9. Find the file **repllog.exe** (by default it resides in \windows) and select it.
10. Click the OK button. The ActiveSync window on the host desktop is automatically invoked, and the New Partnership window is opened automatically. This window prompts: "Would you like to set up a partnership?"
11. Click the No button.
12. Click the Next button.
13. The Microsoft ActiveSync Window is opened automatically and should display "Guest connected".
14. Click the "Explore" button from the Microsoft ActiveSync window. File transfers are now possible through the USB cable.

Compile Switches

There are switches specific to the S1D13A04 USB driver which affect the USB driver. These switches are added or removed from the compile switches in the file **sources**.

CEPC

This option must be set for the CEPC platform and removed for all other platforms.

EPSONMESSAGES

This debugging option enables the display of EPSON-specific debug messages. These debug messages are sent to the serial debugging port. This option should be disabled unless you are debugging the USB driver, as they will significantly impact the performance of the USB driver.

Address and IRQ Modifications

- The USB driver is CPU independent, and it can be used on other platforms that support USB under Windows CE Platform Builder 3.0. If this driver is to support non-cepc platforms, the file **project.reg** requires editing to set the correct values of “PhysicalAddress” and “IRQ”.
- The variables `DEFAULT_PHYSICAL_ADDRESS` and `DEFAULT_IRQ` in the file **13a0xhw.h** must be changed to reflect the values required by each implementation.
- If the entries of “PhysicalAddress” and “IRQ” are removed from the **project.reg** file, the USB driver uses the values of `DEFAULT_PHYSICAL_ADDRESS` and `DEFAULT_IRQ` contained in the file **13a0xhw.h**.

Comments

- S5U13A04B00C Evaluation Board must be configured to enable USB support. This includes configuration changes to the dip switch and confirming that the proper USBChk is available on U13. See the *S5U13A04B00C Rev. 1.0 Evaluation Board User Manual*, document number X37A-G-004-xx.
- This USB driver is independent of the S1D13A04 Windows CE v3.x display driver, but may be run together with S1D13A04 display driver. For information on the S1D13A04 CE Display Driver, see the *Windows CE 3.x Display Driver*, document number X37A-E-006-xx.
- At this time, the driver has been tested on the x86 CPUs and has been run with Platform Builder v3.00.

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S1D13XXX 32-Bit Windows Device Driver Installation Guide

Document No. X00A-E-003-04

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S1D13XXX 32-Bit Windows Device Driver Installation Guide

This manual describes the installation of the Windows 9x/ME/NT 4.0/2000 device drivers for the S5U13xxxB00x series of Epson Evaluation Boards.

The file S1D13XXX.VXD is required for using the Epson supplied Intel32 evaluation and test programs for the S1D13xxx family of LCD controllers with Windows 9x/ME.

The file S1D13XXX.SYS is required for using the Epson supplied Intel32 evaluation and test programs for the S1D13xxx family of LCD controllers with Windows NT 4.0/2000.

The file S1D13XXX.INF is the install script.

For updated drivers, ask your Sales Representative or visit Epson Electronics America on the World Wide Web at www.eea.epson.com.

Driver Requirements

Video Controller	: S1D13xxx
Display Type	: N/A
BIOS	: N/A
DOS Program	: No
Dos Version	: N/A
Windows Program	: Yes, Windows 9x/ME/NT 4.0/2000 device driver
Windows DOS Box	: N/A
Windows Full Screen	: N/A
OS/2	: N/A

Installation

Windows NT Version 4.0

All evaluation boards require the driver to be installed as follows.

1. Install the evaluation board in the computer and boot the computer.
2. Copy the files S1D13XXX.INF and S1D13XXX.SYS to a directory on a local hard drive.
3. Right click your mouse on the file S1D13XXX.INF and select INSTALL from the menu.
4. Windows will install the device driver and ask you to restart.

Windows 2000

All PCI Bus Evaluation Cards

1. Install the evaluation board in the computer and boot the computer.
2. Windows will detect the new hardware as a new PCI Device and bring up the FOUND NEW HARDWARE dialog box.
3. Click NEXT.
4. The New Hardware Wizard will bring up the dialog box to search for a suitable driver.
5. Click NEXT.
6. When Windows does not find the driver it will allow you to specify the location of it. Type the driver location or select BROWSE to find it.
7. Click NEXT.
8. Windows 2000 will open the installation file and show the option EPSON PCI Bridge Card. Select this file and click OPEN.
9. Windows then shows the path to the file. Click OK.
10. Click NEXT.
11. Click FINISH.

All ISA Bus Evaluation Cards

1. Install the evaluation board in the computer and boot the computer.
2. Go to the CONTROL PANEL and select ADD/REMOVE HARDWARE, click NEXT.
3. Select ADD/TROUBLESHOOT A DEVICE, and click NEXT. Windows 2000 will attempt to detect any new plug and play device and fail.
4. The CHOOSE HARDWARE DEVICE dialog box appears. Select ADD NEW HARDWARE and click NEXT.
5. Select NO I WANT TO SELECT FROM A LIST and click NEXT.
6. Select OTHER DEVICE from the list and click NEXT.
7. Click HAVE DISK.
8. Specify the location of the driver files, select the S1D13XXX INF file and click OPEN.
9. Click OK.

Windows 98/ME

All PCI Bus Evaluation Cards

1. Install the evaluation board in the computer and boot the computer.
2. Windows will detect the new hardware as a new PCI Device and bring up the ADD NEW HARDWARE dialog box.
3. Click NEXT.
4. Windows will look for the driver. When Windows does not find the driver it will allow you to specify the location of it. Type the driver location or select BROWSE to find it.
5. Click NEXT.
6. Windows will open the installation file and show the option EPSON PCI Bridge Card.
7. Click FINISH.

All ISA Bus Evaluation Cards

1. Install the evaluation board in the computer and boot the computer.
2. Go to the CONTROL PANEL and double-click on ADD NEW HARDWARE to launch the ADD NEW HARDWARE WIZARD. Click NEXT.
3. Windows will attempt to detect any new plug and play device and fail. Click NEXT.
4. Windows will ask you to let it detect the hardware, or allow you to select from a list. Select NO, I WANT TO SELECT THE HARDWARE FROM A LIST and click NEXT.
5. From the list select OTHER DEVICES and click NEXT.
6. Click HAVE DISK and type the path to the driver files, or select browse to find the driver.
7. Click OK.
8. The driver will be identified as EPSON PCI Bridge Card. Click NEXT.
9. Click FINISH.

Windows 95 OSR2

All PCI Bus Evaluation Cards

1. Install the evaluation board in the computer and boot the computer.
2. Windows will detect the card as a new PCI Device and launch the UPDATE DEVICE DRIVER wizard.

If The Driver is on Floppy Disk

3. Place the disk into drive A: and click NEXT.
4. Windows will find the EPSON PCI Bridge Card.
5. Click FINISH to install the driver.
6. Windows will ask you to restart the system.

If The Driver is not on Floppy Disk

3. Click NEXT, Windows will search the floppy drive and fail.
4. Windows will attempt to load the new hardware as a Standard VGA Card.
5. Click CANCEL. The Driver must be loaded from the CONTROL PANEL under ADD/NEW HARDWARE.
6. Select NO for Windows to DETECT NEW HARDWARE.
7. Click NEXT.
8. Select OTHER DEVICES from HARDWARE TYPE and Click NEXT.
9. Click HAVE DISK.
10. Specify the location of the driver and click OK.
11. Click OK.
12. EPSON PCI Bridge Card will appear in the list.
13. Click NEXT.
14. Windows will install the driver.
15. Click FINISH.
16. Windows will ask you to restart the system.
17. Windows will re-detect the card and ask you to restart the system.

All ISA Bus Evaluation Cards

1. Install the evaluation board in the computer and boot the computer.
2. Go to the CONTROL PANEL and select ADD NEW HARDWARE.
3. Click NEXT.
4. Select NO and click NEXT.
5. Select OTHER DEVICES and click NEXT.
6. Click Have Disk.
7. Specify the location of the driver files and click OK.
8. Click Next.
9. Click Finish.

Previous Versions of Windows 95

All PCI Bus Evaluation Cards

1. Install the evaluation board in the computer and boot the computer.
2. Windows will detect the card.
3. Select DRIVER FROM DISK PROVIDED BY MANUFACTURER.
4. Click OK.
5. Specify a path to the location of the driver files.
6. Click OK.
7. Windows will find the S1D13XXX.INF file.
8. Click OK.
9. Click OK and Windows will install the driver.

All ISA Bus Evaluation Cards

1. Install the evaluation board in the computer and boot the computer.
2. Go to the CONTROL PANEL and select ADD NEW HARDWARE.
3. Click NEXT.
4. Select NO and click NEXT.
5. Select OTHER DEVICES from the HARDWARE TYPES list.
6. Click HAVE DISK.
7. Specify the location of the driver files and click OK.
8. Select the file S1D13XXX.INF and click OK.
9. Click OK.
10. The EPSON PCI Bridge Card should be selected in the list window.
11. Click NEXT.
12. Click NEXT.
13. Click Finish.

EPSON®



S1D13A04 LCD/USB Companion Chip

S5U13A04B00C Rev. 1.0 Evaluation Board User Manual

Document Number: X37A-G-004-02

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1 Introduction

This manual describes the setup and operation of the S5U13A04B00C Rev. 1.0 Evaluation Board. The board is designed as an evaluation platform for the S1D13A04 LCD/USB Companion Chip.

This user manual is updated as appropriate. Please check the Epson Research and Development Website at www.erd.epson.com for the latest revision of this document before beginning any development.

We appreciate your comments on our documentation. Please contact us via email at documentation@erd.epson.com.

2 Features

Following are some features of the S5U13A04B00C Rev. 1.0 Evaluation Board:

- 121-pin PFBGA S1D13A04 Embedded Memory LCD Controller with 160K bytes of embedded SRAM.
- PCI bus operation through onboard PCI bridge.
- CPU/Bus interface header strips for non-PCI bus operation.
- Configuration options.
- On-board adjustable positive LCD bias power supply from +23V to +40V.
- On-board adjustable negative LCD bias power supply from -14V to -24V.
- Software adjustable backlight intensity support using PWMOUT.
- 4/8-bit 3.3V or 5V single monochrome passive LCD panel support.
- 4/8/16-bit 3.3V or 5V single color passive LCD panel support.
- 9/12/18-bit 3.3V or 5V active matrix TFT LCD panel support.
- Direct interface for 18-bit Epson D-TFD LCD panel support.
- Direct interface for 18-bit Sharp HR-TFT LCD panel support.
- Programmable clock synthesizer to CLKI and CLKI2 for maximum clock flexibility.
- Connector for USB client support.
- Software initiated power save mode.
- Selectable clock source for CLKI and CLKI2.
- External oscillator support for CLKI and CLKI2.

3 Installation and Configuration

The S5U13A04B00C is designed to support as many platforms as possible. The S5U13A04B00C incorporates a DIP switch and five jumpers which allow both the evaluation board and the S1D13A04 LCD controller to be configured for a specified evaluation platform.

3.1 Configuration DIP Switches

The S1D13A04 has configuration inputs (CNF[6:0]) which are read on the rising edge of RESET#. In order to configure the S1D13A04 for multiple Host Bus Interfaces an eight-position DIP switch (SW1) is required. The following figure shows the location of DIP switch SW1 on the S5U13A04B00C.

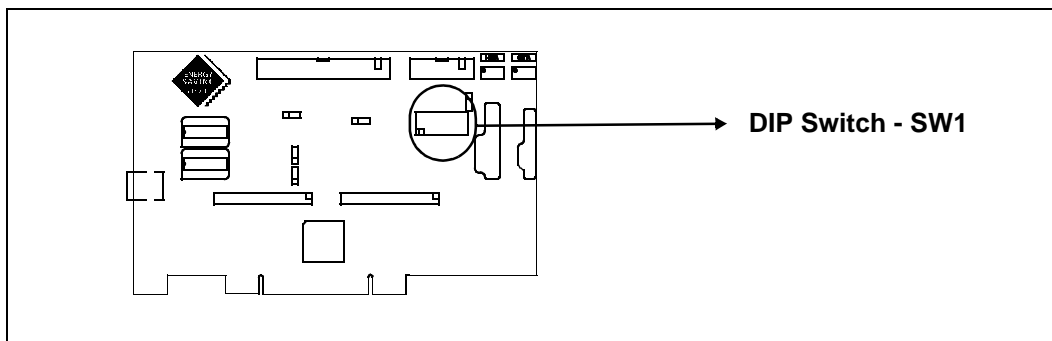


Figure 3-1: Configuration DIP Switch (SW1) Location

The S1D13A04 has seven configuration inputs (CONF[6:0]) which are read on the rising edge of RESET#. All S1D13A04 configuration inputs are fully configurable using an eight position DIP switch as described below.

Table 3-1: Configuration DIP Switch Settings

Switch (SW1)	S1D13A04 Signal	Value on this pin at rising edge of RESET# is used to configure:				
		Closed (On/1)		Open (Off/0)		
SW1-5, SW1-[3:1]	CNF4, CNF[2:0]	Select host bus interface as follows:				Host Bus Interface SH-4/SH-3 interface, Big Endian SH-4/SH-3 interface, Little Endian MC68K #1, Big Endian Reserved MC68K #2, Big Endian Reserved Generic #1, Big Endian Generic #1, Little Endian Reserved Generic #2, Little Endian RedCap 2, Big Endian Reserved DragonBall, Big Endian Reserved Reserved
		CNF4	CNF2	CNF1	CNF0	
		1	0	0	0	
		0	0	0	0	
		1	0	0	1	
		0	0	0	1	
		1	0	1	0	
		0	0	1	0	
		1	0	1	1	
		0	0	1	1	
		1	1	0	0	
		0	1	0	0	
		1	1	0	1	
		0	1	0	1	
1	1	1	0			
0	1	1	0			
X	1	1	1			
SW1-4	CNF3	Configure GPIO pins as inputs at power-on		Configure GPIO[7:6] and GPIO[4:0] as outputs and GPIO5 as an input at power-on (for use when USB is selected)		
SW1-6	CNF5	WAIT# is active high		WAIT# is active low		
SW1-7	CNF6	CLKI to BCLK Divide ratio 2:1		CLKI to BCLK divide ratio 1:1		
SW1-8	-	Disable PCI bridge for non-PCI host		Enable PCI bridge for PCI host		

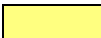
= Required settings when used with PCI Bridge FPGA

3.2 Configuration Jumpers

The S5U13A04B00C has five jumper blocks which configure various setting on the board. The jumper positions for each function are shown below.

Table 3-2: Jumper Summary

Jumper	Function	Position 1-2	Position 2-3	No Jumper
JP1	CLKI Source	VCLKOUT from clock synthesizer	External oscillator (U7)	BUSCLK from Header H4
JP2	CLKI2 Source	MCLKOUT from clock synthesizer	External oscillator (U8)	—
JP3	LCD Panel Voltage	+5V LCDVCC	+3.3V LCDVCC	—
JP4	GP0I0 Polarity on H1	Normal (Active High)	Inverted (Active Low)	GPIO0 not sent to H1
JP5	GPIO0 function select	GPIO0	HR-TFT PS signal	—

 = recommended settings

JP1 - CLKI Source

JP1 selects the source for the CLKI input pin.

Position 1-2 sets the CLKI source to VCLKOUT from the Cypress clock synthesizer (default setting).

Position 2-3 sets the CLKI source to the external oscillator at U7.

When no jumper is installed, the CLKI source is set to the BUSCLK signal from Header H4.

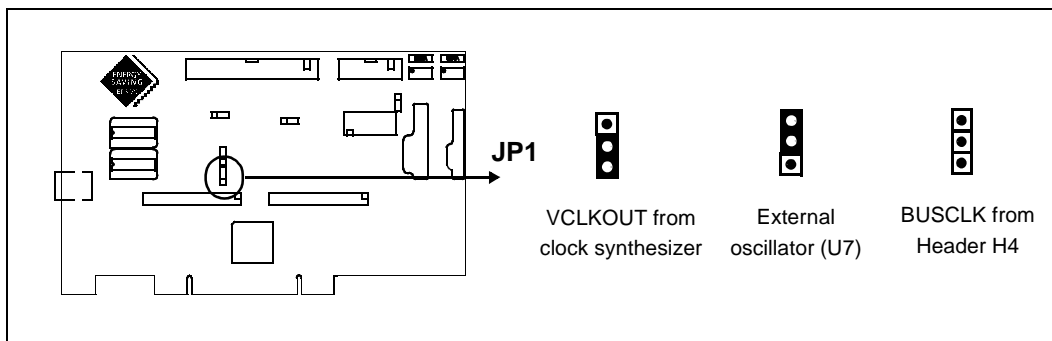


Figure 3-2: Configuration Jumper (JP1) Location

JP2 - CLKI2 Source

JP2 selects the source for the CLKI2 input pin.

Position 1-2 sets the CLKI2 source to MCLKOUT from the Cypress clock synthesizer (default setting).

Position 2-3 sets the CLKI2 source to the external oscillator at U8.

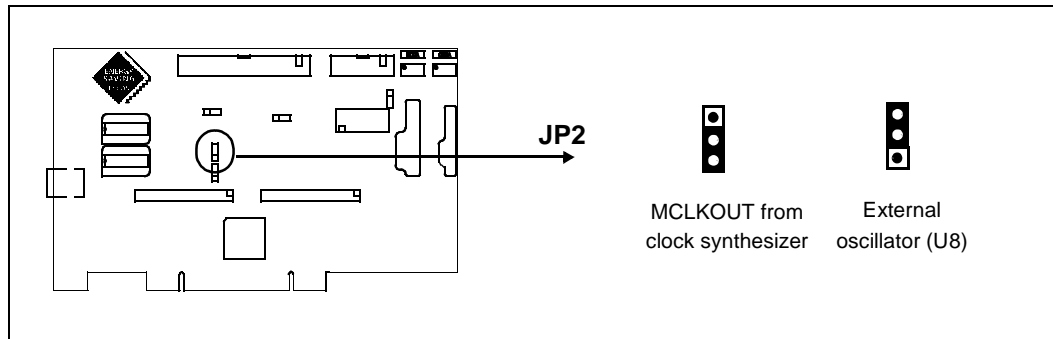


Figure 3-3: Configuration Jumper (JP2) Location

JP3 - LCD Panel Voltage

JP3 selects the voltage level to the LCD panel.

Position 1-2 sets the voltage level to 5.0V (default setting).

Position 2-3 sets the voltage level to 3.3V.

Note

When configured for Sharp HR-TFT or Epson D-TFD panels, JP3 and JP5 must be set to position 2-3.

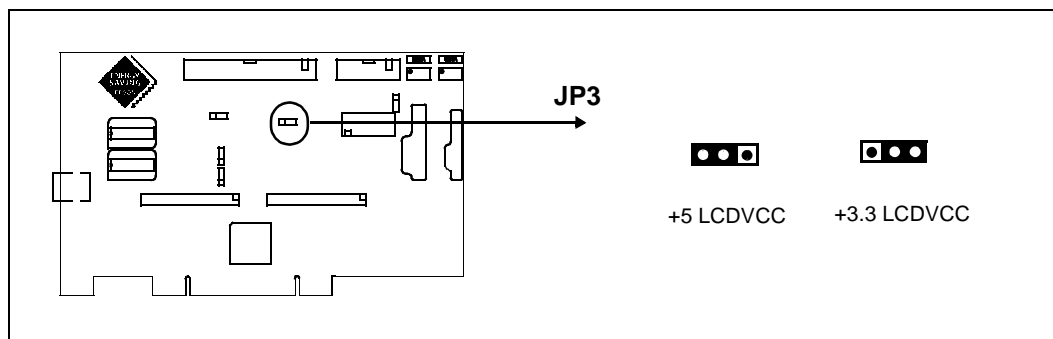


Figure 3-4: Configuration Jumper (JP3) Location

JP4 - GPIO0 Polarity on H1

JP4 selects the polarity of the GPIO0 signal available on the LCD Connector H1.
Position 1-2 sends the GPIO0 signal directly to H1 (default setting).
Position 2-3 inverts the GPIO0 signal before sending it to H1.
When no jumper is installed, GPIO0 is not sent to H1.

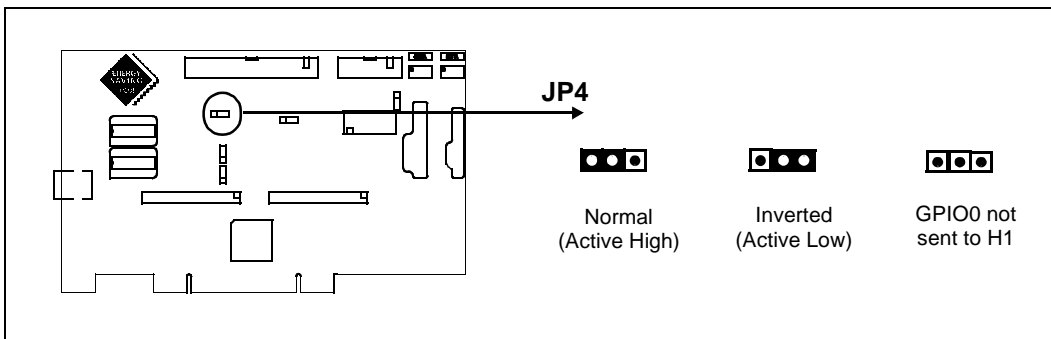


Figure 3-5: Configuration Jumper (JP4) Location

JP5 - GPIO0 Selection

JP5 selects the function of the GPIO0 signal.
Position 1-2 GPIO0 used to control the LCD bias power supplies for STN panels.
Position 2-3 GPIO0 used as the PS signal when the S1D13A04 is configured for HR-TFT panel type.

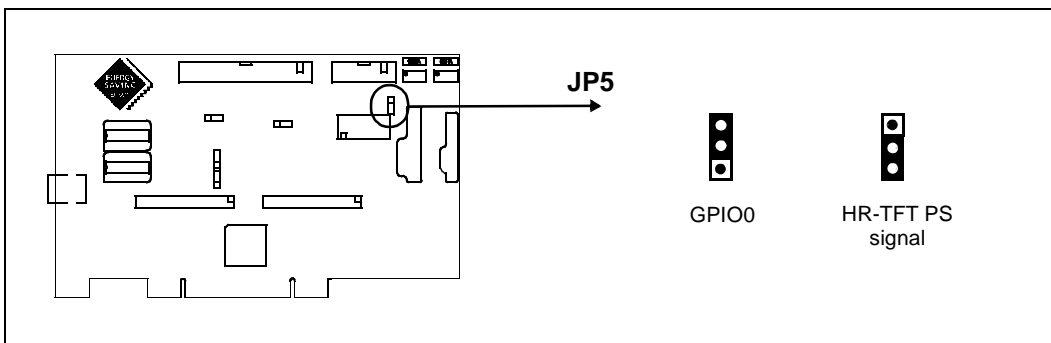


Figure 3-6: Configuration Jumper (JP5) Location

4 CPU Interface

4.1 CPU Interface Pin Mapping

Table 4-1: CPU Interface Pin Mapping

S1D13A04 Pin Name	Generic #1	Generic #2	Hitachi SH-3 /SH-4	Motorola MC68K #1	Motorola MC68K #2	Motorola REDCAP2	Motorola MC68EZ328/ MC68VZ328 DragonBall
AB[17:1]	A[17:1]	A[17:1]	A[17:1]	A[17:1]	A[17:1]	A[17:1]	A[17:1]
AB0	A0 ¹	A0	A0 ¹	LDS#	A0	A0 ¹	A0 ¹
DB[15:0]	D[15:0]	D[15:0]	D[15:0]	D[15:0]	D[15:0] ²	D[15:0]	D[15:0]
CS#	External Decode		CSn#	External Decode		\overline{CSn}	\overline{CSX}
M/R#	External Decode						
CLKI	BUSCLK	BUSCLK	CKIO	CLK	CLK	CLK	CLK
BS#	Connected to HIOV _{DD} ³		BS#	AS#	AS#	Connected to HIOV _{DD} ³	
RD/WR#	RD1#	Connected to HIOV _{DD} ³	RD/WR#	R/W#	R/W#	R/\overline{W}	Connected to HIOV _{DD} ³
RD#	RD0#	RD#	RD#	Connected to HIOV _{DD} ³	SIZ1	\overline{OE}	\overline{OE}
WE0#	WE0#	WE#	WE0#	Connected to HIOV _{DD} ³	SIZ0	$\overline{EB1}$	\overline{LWE}
WE1#	WE1#	BHE#	WE1#	UDS#	DS#	$\overline{EB0}$	\overline{UWE}
WAIT#	WAIT#	WAIT#	WAIT#/RDY#	DTACK#	DSACK1#	N/A	\overline{DTACK}
RESET#	RESET#	RESET#	RESET#	RESET#	RESET#	$\overline{RESET_OUT}$	\overline{RESET}

Note

¹ A0 for these busses is not used internally by the S1D13A04 and should be connected to V_{SS}.

² If the target MC68K bus is 32-bit, then these signals should be connected to D[31:16].

³ These pins are not used in their corresponding host interface mode. Systems are responsible for externally connecting them to Host Interface IO V_{DD}.

4.2 CPU Bus Connector Pin Mapping

Table 4-2: CPU Bus Connector (H3) Pinout

Connector Pin No.	Comments
1	Connected to DB0 of the S1D13A04
2	Connected to DB1 of the S1D13A04
3	Connected to DB2 of the S1D13A04
4	Connected to DB3 of the S1D13A04
5	Ground
6	Ground
7	Connected to DB4 of the S1D13A04
8	Connected to DB5 of the S1D13A04
9	Connected to DB6 of the S1D13A04
10	Connected to DB7 of the S1D13A04
11	Ground
12	Ground
13	Connected to DB8 of the S1D13A04
14	Connected to DB9 of the S1D13A04
15	Connected to DB10 of the S1D13A04
16	Connected to DB11 of the S1D13A04
17	Ground
18	Ground
19	Connected to DB12 of the S1D13A04
20	Connected to DB13 of the S1D13A04
21	Connected to DB14 of the S1D13A04
22	Connected to DB15 of the S1D13A04
23	Connected to RESET# of the S1D13A04
24	Ground
25	Ground
26	Ground
27	+12 volt supply
28	+12 volt supply
29	Connected to WE0# of the S1D13A04
30	Connected to WAIT# of the S1D13A04
31	Connected to CS# of the S1D13A04
32	Connected to MR# of the S1D13A04
33	Connected to WE1# of the S1D13A04
34	Connected to +3.3V

Table 4-3: CPU Bus Connector (H4) Pinout

Connector Pin No.	Comments
1	Connected to AB0 of the S1D13A04
2	Connected to AB1 of the S1D13A04
3	Connected to AB2 of the S1D13A04
4	Connected to AB3 of the S1D13A04
5	Connected to AB4 of the S1D13A04
6	Connected to AB5 of the S1D13A04
7	Connected to AB6 of the S1D13A04
8	Connected to AB7 of the S1D13A04
9	Ground
10	Ground
11	Connected to AB8 of the S1D13A04
12	Connected to AB9 of the S1D13A04
13	Connected to AB10 of the S1D13A04
14	Connected to AB11 of the S1D13A04
15	Connected to AB12 of the S1D13A04
16	Connected to AB13 of the S1D13A04
17	Ground
18	Ground
19	Connected to AB14 of the S1D13A04
20	Connected to AB15 of the S1D13A04
21	Connected to AB16 of the S1D13A04
22	Connected to AB17 of the S1D13A04
23	Not connected
24	Not connected
25	Ground
26	Ground
27	+5 volt supply
28	+5 volt supply
29	Connected to RD/WR# of the S1D13A04
30	Connected to BS# of the S1D13A04
31	Connected to BUSCLK of the S1D13A04
32	Connected to RD# of the S1D13A04
33	Not connected
34	Not connected

5 LCD Interface Pin Mapping

Table 5-1: LCD Signal Connector (H1)

Pin Name	Connector Pin No.	Monochrome Passive Panel		Color Passive Panel				Color TFT Panel			
		Single		Single				Others			Sharp HR-TFT
		4-bit	8-bit	4-bit	Format 1	Format 2	16-Bit	9-bit	12-bit	18-bit	18-bit
FPDAT0	1	driven 0	D0	driven 0	D0 (B5) ¹	D0 (G3) ¹	D0 (R6) ¹	R2	R3	R5	R5
FPDAT1	3	driven 0	D1	driven 0	D1 (R5) ¹	D1 (R3) ¹	D1 (G5) ¹	R1	R2	R4	R4
FPDAT2	5	driven 0	D2	driven 0	D2 (G4) ¹	D2 (B2) ¹	D2 (B4) ¹	R0	R1	R3	R3
FPDAT3	7	driven 0	D3	driven 0	D3 (B3) ¹	D3 (G2) ¹	D3 (R4) ¹	G2	G3	G5	G5
FPDAT4	9	D0	D4	D0 (R2) ¹	D4 (R3) ¹	D4 (R2) ¹	D8 (B5) ¹	G1	G2	G4	G4
FPDAT5	11	D1	D5	D1 (B1) ¹	D5 (G2) ¹	D5 (B1) ¹	D9 (R5) ¹	G0	G1	G3	G3
FPDAT6	13	D2	D6	D2 (G1) ¹	D6 (B1) ¹	D6 (G1) ¹	D10 (G4) ¹	B2	B3	B5	B5
FPDAT7	15	D3	D7	D3 (R1) ¹	D7 (R1) ¹	D7 (R1) ¹	D11 (B3) ¹	B1	B2	B4	B4
FPDAT8	17	driven 0	driven 0	driven 0	driven 0	driven 0	D4 (G3) ¹	B0	B1	B3	B3
FPDAT9	19	driven 0	driven 0	driven 0	driven 0	driven 0	D5 (B2) ¹	driven 0	R0	R2	R2
FPDAT10	21	driven 0	driven 0	driven 0	driven 0	driven 0	D6 (R2) ¹	driven 0	driven 0	R1	R1
FPDAT11	23	driven 0	driven 0	driven 0	driven 0	driven 0	D7 (G1) ¹	driven 0	driven 0	R0	R0
FPDAT12	25	driven 0	driven 0	driven 0	driven 0	driven 0	D12 (R3) ¹	driven 0	G0	G2	G2
FPDAT13	27	driven 0	driven 0	driven 0	driven 0	driven 0	D13 (G2) ¹	driven 0	driven 0	G1	G1
FPDAT14	29	driven 0	driven 0	driven 0	driven 0	driven 0	D14 (B1) ¹	driven 0	driven 0	G0	G0
FPDAT15	31	driven 0	driven 0	driven 0	driven 0	driven 0	D15 (R1) ¹	driven 0	B0	B2	B2
FPDAT16	4	driven 0	driven 0	driven 0	driven 0	driven 0	driven 0	driven 0	driven 0	B1	B1
FPDAT17	6	driven 0	driven 0	driven 0	driven 0	driven 0	driven 0	driven 0	driven 0	B0	B0
FPSHIFT	33	FPSHIFT									CLK
DRDY	35 & 38	MOD			FPSHIFT2	MOD			DRDY		GPO ³
FPLINE	37	FPLINE									LP
FPFRAME	39	FPFRAME									SPS
GND	2, 8, 14, 20, 26	GND									
PWMOUT	28	PWMOUT									
VLCD	30	Adjustable -24V to -14V negative LCD bias									
VCC	32	LCDVCC (3.3V or 5V)									
+12V	34	+12V									
VDDH	36	Adjustable +23V to +40V positive LCD bias									
DISPLAY ²	40	GPIO0 (for controlling on-board LCD bias power supply on/off)									PS

Note

- ¹ These pin mappings use signal names commonly used for each panel type, however signal names may differ between panel manufacturers. The values shown in brackets represent the color components as mapped to the corresponding FPDATxx signals at the first valid edge of FPSHIFT. For further FPDATxx to LCD interface mapping, see *SID13A04 Hardware Functional Specification*, document number X37A-A-001-xx.
- ² DISPLAY can be disconnected from GPIO0 using JP5 (2-3 position) and can be inverted on H1 setting JP4 to 2-3.
- ³ When the 'Direct' HR-TFT interface is selected, DRDY becomes a general purpose output (GPO) controllable using the 'Direct' HR-TFT LCD Interface GPO Control bit (REG[14h] bit 0). This GPO can be used to control the HR-TFT MOD signal if required. For further information, see the *SID13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

Table 5-2: Extended LCD Signal Connector (H2)

Pin Name	Connector Pin No.	Monochrome Passive Panel		Color Passive Panel				Color TFT Panel				USB ³	
		Single		Single				Others			HR-TFT ²		
		4-bit	8-bit	4-bit	8-bit	8-bit	16-Bit	9-bit	12-bit	18-bit	18-bit		
GPIO0 ¹	1	GPIO0										PS	GPIO0
GPIO1 ¹	3	GPIO1										CLS	GPIO1
GPIO2 ¹	5	GPIO2										REV	GPIO2
GPIO3 ¹	7	GPIO3										SPL	GPIO3
GPIO4 ¹	9	GPIO4											USBPUP
GPIO5 ¹	11	GPIO5											USBDETECT
GPIO6 ¹	13	GPIO6											USBDM
GPIO7 ¹	15	GPIO7											USB DP
GND	2, 4, 6, 8, 10, 12, 14, 16	GND											

Note

- ¹ When Switch SW1-4 is open (CNF3 = 0 at RESET#), GPIO[7:6] and GPIO[4:0] are set as outputs at 0 (low state) and GPIO5 is set as an input at power-on/RESET# (for use when USB is selected). If SW1-4 is closed then GPIO[7:0] are set as inputs upon power-on/RESET#.
- ² If the 'Direct' HR-TFT interface is selected (REG[0Ch] bits 1:0 = 10), GPIO[3:0] are used for the 'Direct' HR-TFT interface. GPIO[7:4] remain available for USB support or as GPIOs.
- ³ If USB support is enabled (REG[4000h] bit 7 = 1), GPIO[7:4] are used by the USB interface. GPIO[3:0] remain available for 'Direct' HR-TFT interface support or as GPIOs.

6 Technical Description

6.1 PCI Bus Support

The S1D13A04 **does not** have on-chip PCI bus interface support. The S1D13A04B00C uses the PCI Bridge FPGA to support the PCI bus.

6.2 Direct Host Bus Interface Support

The S5U13A04B00C is specifically designed to work using the PCI Bridge FPGA in a standard PCI bus environment. However, the S1D13A04 directly supports many other host bus interfaces. Connectors H3 and H4 provide the necessary IO pins to interface to these host buses. For further information on the host bus interfaces supported, see “CPU Interface” on page 14.

Note

The PCI Bridge FPGA must be disabled using SW1-8 in order for direct host bus interface to operate properly.

6.3 S1D13A04 Embedded Memory

The S1D13A04 has 160K bytes of embedded SRAM. The 160K byte display buffer address space is directly and contiguously available through the 18-bit address bus.

6.4 Adjustable LCD Panel Negative Power Supply

Most monochrome passive LCD panels require a negative power supply to provide between -14V and -24V ($I_{out}=45mA$). Such a power supply (VLCD) has been provided on the S5U13A04B00C board. VLCD can be adjusted using potentiometer R39 to provide an output voltage from -14V to -24V, and is enabled/disabled using the S1D13A04 general purpose signal, GPIO0 (active high).

Note

When manually adjusting the voltage, set the potentiometer according to the panel’s specific power requirements **before connecting the panel**.

6.5 Adjustable LCD Panel Positive Power Supply

Most color passive LCD panels and most single monochrome 640x480 passive LCD panels require a positive power supply to provide between +23V and +40V ($I_{out}=45mA$). Such a power supply (VDDH) has been provided on the S5U13A04B00C board. VDDH can be adjusted using R22 to provide an output voltage from +23V to +40V, and is enabled/disabled using the S1D13A04 general purpose signal, GPIO0 (active high).

Note

When manually adjusting the voltage, set the potentiometer according to the panel's specific power requirements **before connecting the panel**.

6.6 Software Adjustable LCD Backlight Intensity Support Using PWM

The S1D13A04 provides Pulse Width Modulation output on PWMOUT. PWMOUT can be used to control LCD panels which support PWM control of the backlight inverter. The PWMOUT signal is provided on the LCD connector, H1.

6.7 LCD Panel Support

The S1D13A04 directly supports:

- Single-panel, single drive passive displays.
 - 4/8-bit monochrome interface.
 - 4/8/16-bit color interface.
- Active Matrix TFT interface.
 - 9/12/18-bit interface.
- 'Direct' support for 18-bit Sharp HR-TFT LCD or compatible interface.

All the necessary signals are provided on the 40-pin LCD connector H1 and the 16-pin LCD connector H2. For connection information, refer to Table 5-1: "LCD Signal Connector (H1)" on page 17 and Table 5-2: "Extended LCD Signal Connector (H2)" on page 18.

6.7.1 Direct LCD Connector

The direct LCD Connector (H1) provides all LCD panel signals required for Active Matrix TFT and Passive LCD panels. These signals are buffered to either a 3.3V level or a 5.0V level depending on the setting of JP3. See Table 3-2: "Jumper Summary" on page 11.

6.7.2 Extended LCD Connector

The S1D13A04 directly supports Sharp 18-bit HR-TFT and compatible panels. The extended LCD connector (H2) provides the extra signals required to support these panels. The signals on this connector are provided directly from the S1D13A04 without any buffering and are 3.3V signals.

6.8 USB Support

The S1D13A04 USB controller provides a Revision 1.1 compliant USB client. The S1D13A04 acts as a USB device and connects to an upstream hub or USB host through connector J1 on the S5U13A04B00C evaluation board. Clamping diodes have been added to protect the USB bus from ESD and shorting.

6.8.1 USB IRQ Support

The S1D13A04 supports interrupts through output pin, IRQ. This interrupt can be used to support interrupts from the USB client of the S1D13A04. The S5U13A04B00C evaluation board supports this capability by connecting IRQ to the PCI interrupt INTA# of the PCI slot that the S5U13A04B00C evaluation board is connected to.

7 Clock Synthesizer and Clock Options

For maximum flexibility, the S5U13A04B00C implements a Cypress ICD2061A Clock Synthesizer. MCLKOUT from the clock synthesizer is connected to CLKI2 of the S1D13A04 and VCLKOUT from the clock synthesizer is connected to CLKI of the S1D13A04. A 14.31818MHz crystal (Y1) is connected to XTALIN and XTALOUT of the clock synthesizer and provides the reference clock to the clock synthesizer.

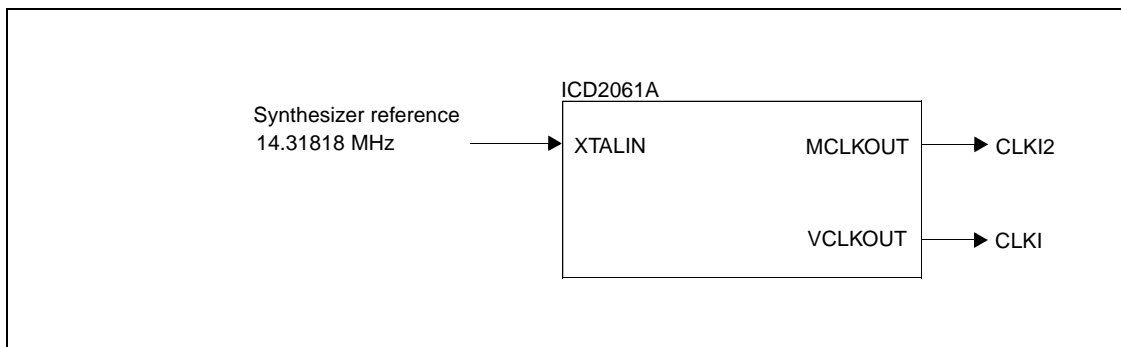


Figure 7-1: Symbolic Clock Synthesizer Connections

At power-on, CLKI2 (MCLKOUT) is configured to be 40MHz and CLKI (VCLKOUT) is configured at 25.175MHz.

Note

If a Sharp HR-TFT panel is selected, the clock synthesizer cannot be programmed, and external oscillators must provide the clock signals to CLKI and CLKI2. Jumpers JP1 and JP2 allow selection of external oscillators U7 and U8 as the clock source for both CLKI and CLKI2. For further information, see Table 3-2: “Jumper Summary” on page 11.

7.1 Clock Programming

The S1D13A04 utilities automatically program the clock generator. If manual programming of the clock generator is required, refer to the source code for the S1D13A04 utilities available on the internet at www.erd.epson.com.

For further information on programming the clock generator, refer to the *Cypress ICD2061A specification*.

Note

When CLKI and CLKI2 are programmed to multiples of each other (e.g. CLKI = 20MHz, CLKI2 = 40MHz), the clock output signals from the Cypress clock generator may jitter. Refer to the Cypress ICD2061A specification for details.

To avoid this problem, set CLKI and CLKI2 to different frequencies and use the S1D13A04 internal clock divides to obtain the lower frequencies.

8 References

8.1 Documents

- Epson Research and Development, Inc., *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.
- Epson Research and Development, Inc., *S1D13A04 Programming Notes and Examples*, document number X37A-G-003-xx.
- Cypress Semiconductor Corporation, *ICD2061A Data Sheet*.

8.2 Document Sources

- Epson Research and Development: <http://www.erd.epson.com>.
- Cypress Semiconductor Corporation Website: <http://www.cypress.com>.

9 Parts List

Table 9-1: Parts List

Item	Qty	Designation	Part Value	Description	Manufacturer / Part No. / Assembly Instructions
1	25	C1-C11,C15-C21,C28,C30,C32-C35,C48	0.1u	Ceramic Chip, 50V X7R +/-5%, 1206 pckg.	Kemet C1206C104J5RAC
2	1	C12	10u,10V	Tantalum C-Size, +/-10%	Kemet T491C106K010AS
3	2	C13,C14	n/p	1206 pckg.	Do not populate
4	7	C22,C2,C29,C31,C45,C46,C49	68u, 10V	Tantalum D-Size, +/-10%	Kemet T491D86K010AS
5	3	C23-C25	10uF, 63V	Electrolytic, Radial Lead, +/-20%	NIPPON/UNITED CHEMI-CON KMF63VB10RM5X11LL
6	1	C27	56uF, 35V	Electrolytic, Radial Lead, +/-20%	NIPPON/UNITED CHEMI-CON KMF35VB56RM5X11LL
7	9	C36-C44	0.22uF	Ceramic Chip, 50V X7R +/-5%, 1206 pckg.	Kemet C1206C224J5RAC
8	2	C47,C50	33u, 20V	Tantalum D-Size, +/-10%	Kemet T491D33K020AS
9	2	D1,D2	BAV99	Ultra high-speed switching diode	Rohm BAV99
10	1	H1	HEADER 20X2	20x2, .025" sq. shrouded header, keyed	Thomas&Betts P/N:636-4027 or equivalent
11	1	H2	HEADER 8X2	8x2, .025" sq. shrouded header, keyed	Thomas&Betts P/N:636-1627 or equivalent
12	2	H3,H4	HEADER 17X2	17x2, .025" sq. unshrouded header, right angle	Thomas&Betts P/N:609-3407 or equivalent
13	5	JP1-JP5	HEADER 3	3x1 .1" pitch unshrouded header	
14	1	J1	USB Type B	Right Angle, Type B USB Connector	AMP787780-1
15	1	L1	1uH	Inductor	RCD MCI-1812 1uH MT
16	1	L2	Ferrite	Ferrite Bead	Phillips BDS3/3/8.9-4S2
17	1	Q1	MMBT3906	PNP signal transistor, SOT23 package	Motorola MMBT3906LT1
18	1	Q2	MMBT2222A	NPN transistor, SOT-23 pckg.	Motorola MMBT2222A
19	13	R1-R7,R23,R31,R34-R37	15K, 5%	1206 Resistor	
20	8	R8-R14,R16	330K, 5%	1206 Resistor	
21	6	R15,R17,R18,R32,R33,R38	1K, 5%	1206 Resistor	
22	3	R19,R20,R30	100K, 5%	1206 Resistor	
23	1	R21	470 Ohm, 5%	1206 Resistor	
24	1	R22	200K Pot	200K Trim Pot	Bourns 3386W-1-204
25	2	R24,R25	20 Ohm, 1%	1206 Resistor	
26	2	R26,R29	301K, 1%	1206 Resistor	
27	1	R27	1.5K, 1%	1206 Resistor	

Table 9-1: Parts List

Item	Qty	Designation	Part Value	Description	Manufacturer / Part No. / Assembly Instructions
28	1	R28	150K, 1%	1206 Resistor	
29	1	R39	100K Pot	100K Trim Pot	Bourns 3386W-1-104
30	1	SW1	SW DIP-8	Dip Switch, 8-Position	
31	1	SW2	SW DIP-4	DIP switch, 4-position	Do Not Populate
32	1	U1	S1D13A04F0A	121-pin PFBGA 13A04 LCDC	Supplied by Epson R&D
33	1	U2	LT1117CST-5	5V fixed voltage regulator, SOT-223	Linear Technology LT1117CST-5
34	1	U3	ICD2061A	Clock Chip, Wide SO-16 pckg	Cypress ICD2061A
35	1	U4	74AHC04	Inverter, SO-14 package	TI74AHC04
36	2	U5,U6	NC7SZ04	TinyLogic UHS inverter, SOT23-5 package	Fairchild NC7SZ04
37	2	U7,U8	Test Socket	14 pin narrow DIP, screw machine socket	Sockets for oscillator input
38	1	U9	RD-0412	Positive LCD Bias Power Supply	Taiyo Yuden/Xentek Positive Power Supply, RD-0412
39	1	U10	EPN001	Negative LCD Bias Power Supply	Taiyo Yuden/Xentek Negative Power Supply, EPN001
40	1	U11	NC7ST04	TinyLogic HST inverter, SOT23-5 package	Fairchild NC7ST04
41	1	U12	LT1118CST-2.5	2.5V fixed volt reg / SOT-223	Linear Technology LT1118CST-2.5
42	1	U13	LT1117CM-3.3	3.3V fixed volt reg / 3 Lead PlasticDD	Linear Technology LT1117CM-3.3
43	3	U14-U16	74HCT244	Buffer, SO-20 package	TI74HCT244
44	1	U17	EPF6016TC144-2	144-pin TQFP FLEX6000 FPGA	Altera EPF6016TC144-2
45	1	U18	EPC1441PC8	8-pin DIP pckg, OTP EPROM	Altera EPC1441PC8, socketed
46	1	U19	74HCT125	Buffer, SO-14 package	TI74HCT125
47	1	Y1	14.31818MHz	14.31818MHz crystal, Fox HC-49	FOX FoxS/143-20 14.31818MHz
48	1	Y2	48MHz Osc	SMD 48MHz oscillator	Epson SG-615PH-48.000MHz
49	1	U18	Socket	4-pin narrow DIP, screw machine socket	Socket for U18
50	5	(JP1-JP5)	Shunts	Jumper Shunts	
51	1	Z1	PCI Bracket	PCI bracket with slot for USB Type B connector	Hansen Industries
52	2	Z2	Screw	Screw, pan head, #4-40 x 1/4"	Use to Assemble PCI bracket onto PCB board

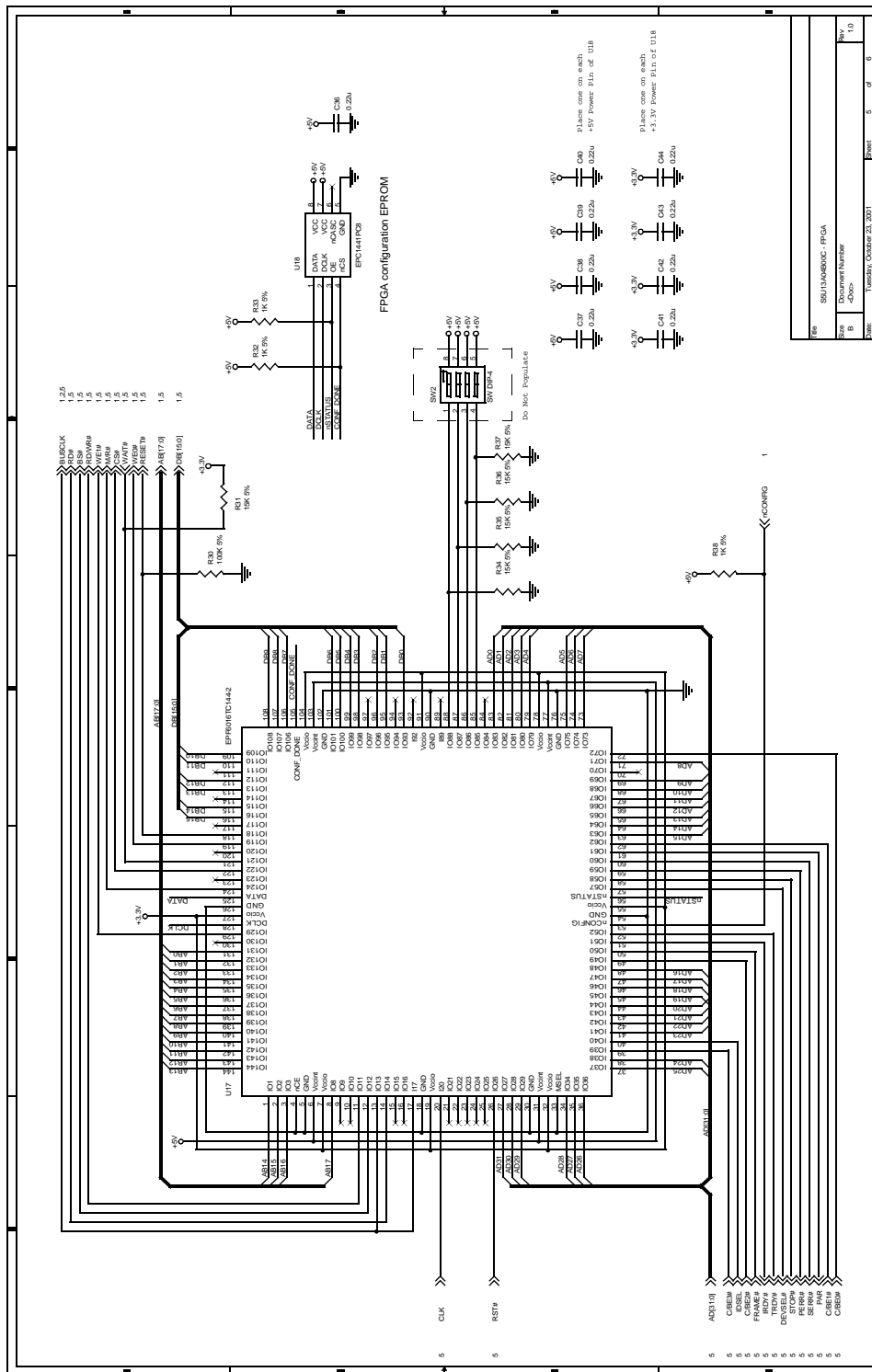


Figure 10-5: S1D13A04B00C Schematics (5 of 6)

11 Board Layout

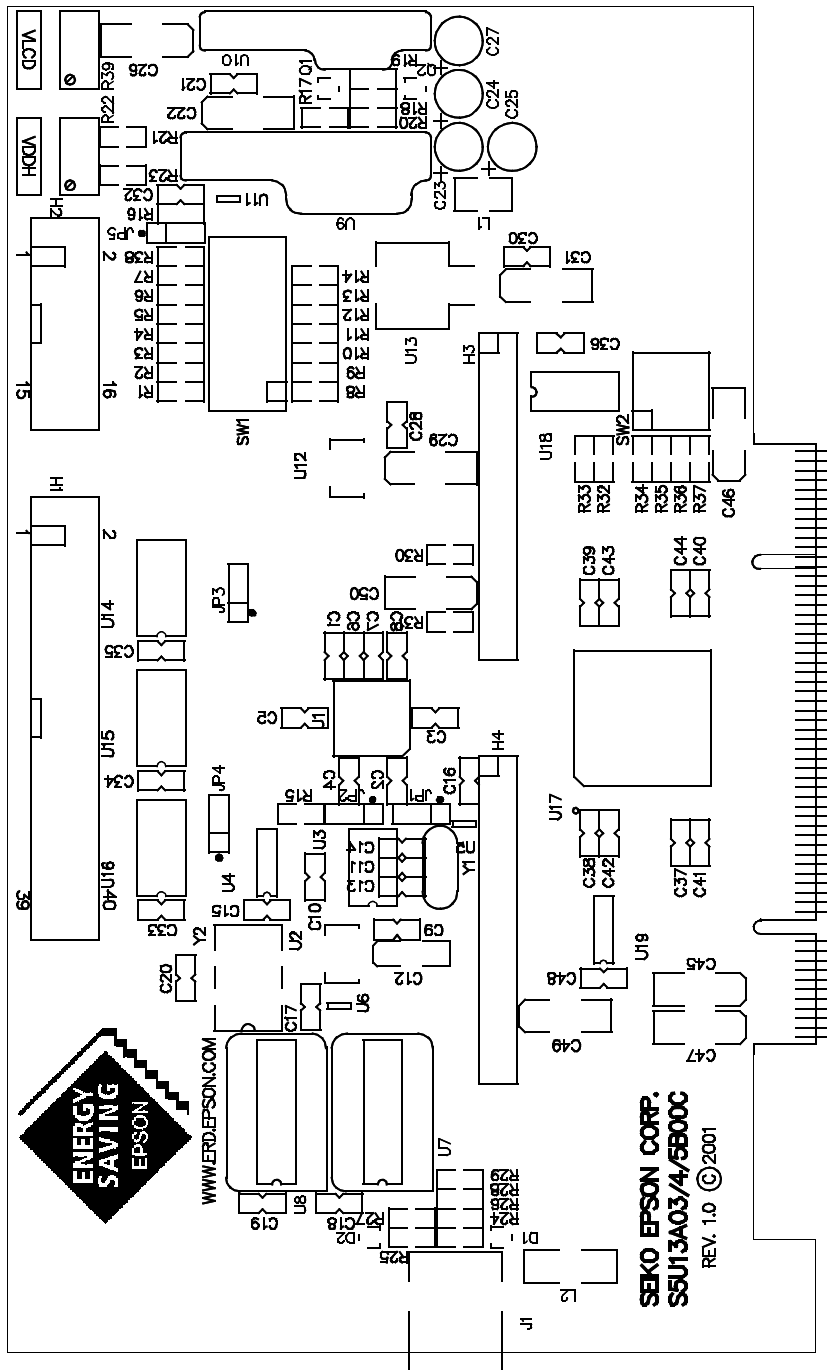


Figure 11-1: S5U13A04B00C Board Layout

12 Sales and Technical Support

12.1 Epson Companion Chips (S1D13A04)

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EPSON®



S1D13A04 LCD/USB Companion Chip

Interfacing to the Toshiba MIPS TMPR3905/3912 Microprocessors

Document Number: X37A-G-002-01

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1 Introduction

This application note describes the hardware and software environment necessary to provide an interface between the S1D13A04 USB/LCD Companion Chip and the Toshiba MIPS TMPR3905/3912 processors.

The designs described in this document are presented only as examples of how such interfaces might be implemented. This application note is updated as appropriate. Please check the Epson Research and Development website at www.erd.epson.com for the latest revision of this document before beginning any development.

We appreciate your comments on our documentation. Please contact us via email at documentation@erd.epson.com.

2 Interfacing to the TMPR3905/12

2.1 The Toshiba TMPR3905/12 System Bus

The TMPR39XX family of processors features a high-speed system bus typical of modern MIPS RISC microprocessors. This section provides an overview of the operation of the CPU bus in order to establish interface requirements.

2.1.1 Overview

The TMPR3905/12 is a highly integrated controller developed for handheld products. The microprocessor is based on the R3900 MIPS RISC processor core. The TMPR3905/12 implements an external 26-bit address bus and a 32-bit data bus allowing it to communicate with its many peripheral units. The address bus is multiplexed (A[12:0]) using an address latch signal (ALE) which controls the driving of the address onto the address bus. The full 26-bit address bus (A[25:0]) is generated to devices not capable of receiving a multiplexed address, using external latches (controlled by ALE).

The TMPR3905/12 provides two, revision 2.01 compliant, PC Card slots. The 16-bit PC Card slots provide a 26-bit multiplexed address and additional control signals which allow access to three 64M byte address ranges: IO, memory, and attribute space. The signal CARDREG* selects memory space when high and attribute or IO space when low. Memory and attribute space are accessed using the write and read enable signals (WE* and RD*). When CARDREG* is low, card IO space is accessed using separate write (CARDIOWR*) and read (CARDIORD*) control signals.

2.1.2 Card Access Cycles

A data transfer is initiated when the address is placed on the PC Card bus and one, or both, of the card enable signals (CARD1CSL* and CARD1CSH*) are driven low. CARDREG* is inactive for memory and IO cycles. If only CARD1CSL* is driven low, 8-bit data transfers are enabled and A0 specifies whether the even or odd data byte appears on the PC Card data bus lines D[7:0]. If only CARD1CSH* is driven low, an odd byte transfer occurs on PC Card data lines D[15:8]. If both CARD1CSL* and CARD1CSH* are driven low, a 16-bit word transfer takes place on D[15:0].

During a read cycle, either RD* or CARDIORD* is driven low depending on whether a memory or IO cycle is specified. A write cycle is specified by driving WE* (memory cycle) or CARDIOWR* (IO cycle) low. The cycle can be lengthened by driving CARD1WAIT* low for the time required to complete the cycle.

Figure 2-1: "Toshiba 3905/12 PC Card Memory/Attribute Cycle," illustrates a typical memory/attribute cycle on the Toshiba 3905/12 PC Card bus.

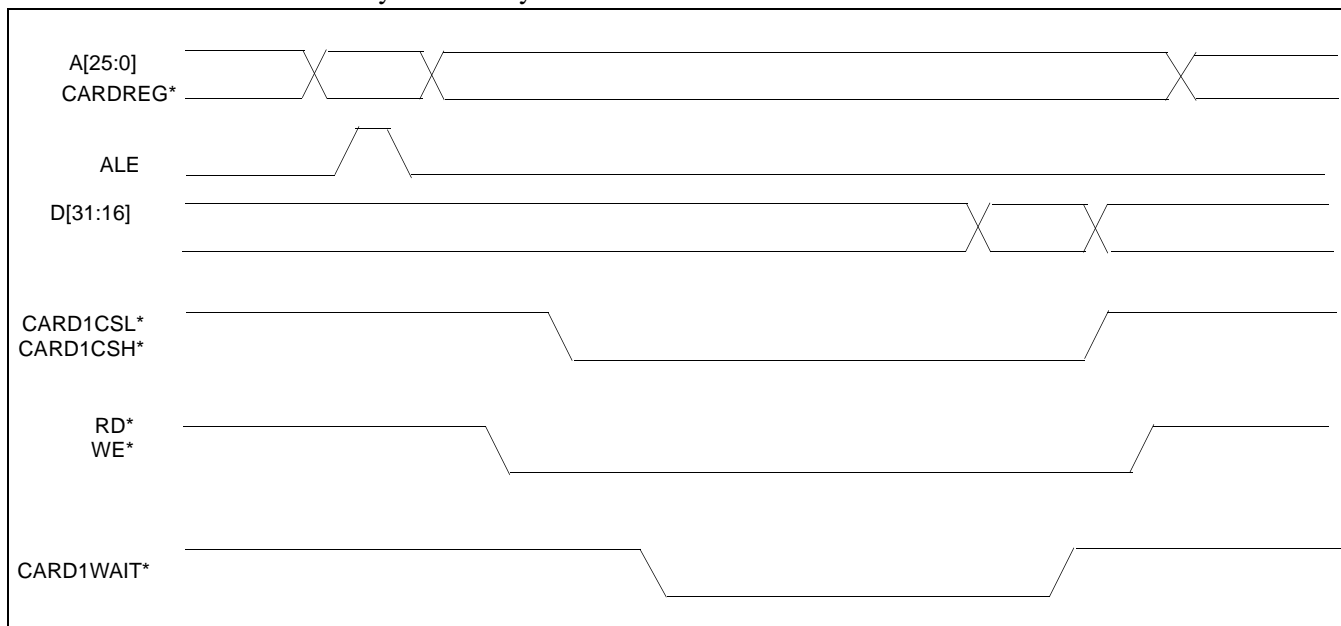


Figure 2-1: Toshiba 3905/12 PC Card Memory/Attribute Cycle

Figure 2-2: "Toshiba 3905/12 PC Card IO Cycle," illustrates a typical IO cycle on the Toshiba 3905/12 PC Card bus.

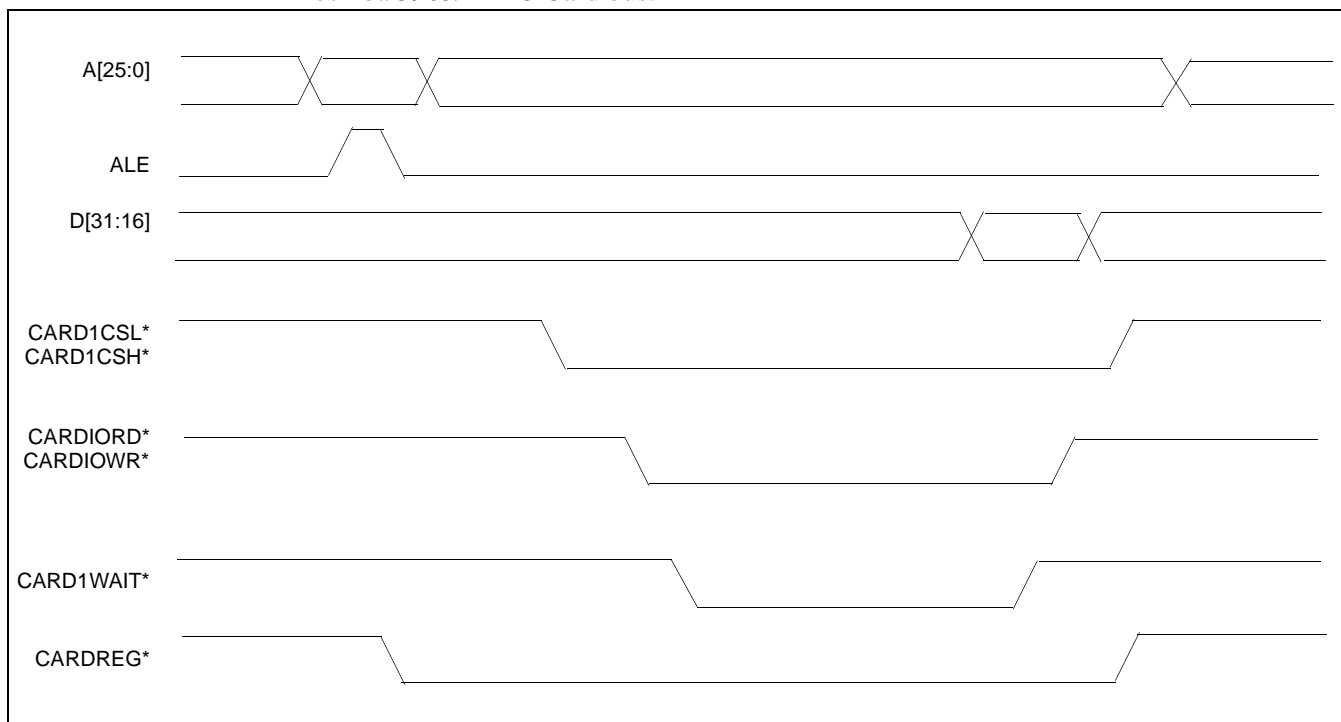


Figure 2-2: Toshiba 3905/12 PC Card IO Cycle

3 S1D13A04 Host Bus Interface

The S1D13A04 directly supports multiple processors. The S1D13A04 implements a 16-bit Generic #2 Host Bus Interface which is most suitable for connection to the Toshiba TMPR3905/12 microprocessor.

The Generic #2 Host Bus Interface is selected by the S1D13A04 on the rising edge of RESET#. After releasing reset the bus interface signals assume their selected configuration. For details on the S1D13A04 configuration, see Section 4.2, “S1D13A04 Hardware Configuration” on page 14.

3.1 Host Bus Interface Pin Mapping

The following table shows the functions of each Host Bus Interface signal.

Table 3-1: Host Bus Interface Pin Mapping

S1D13A04 Pin Names	Toshiba TMPR3905/12
AB[17:0]	External Decode
DB[15:8]	D[23:16]
DB[7:0]	D[31:24]
WE1#	External Decode
CS#	External Decode
M/R#	External Decode
CLKI	DCLKOUT
BS#	Connect to IO _{VDD} from the S1D13A04
RD/WR#	Connect to IO _{VDD} from the S1D13A04
RD#	CARDIORD*
WE0#	CARDIOWR*
WAIT#	CARD1WAIT*
RESET#	system $\overline{\text{RESET}}$

3.2 Host Bus Interface Signals

The Host Bus Interface requires the following signals.

- CLKI is a clock input required by the S1D13A04 Host Bus Interface as a source for its internal bus and memory clocks. This clock is typically driven by the host CPU system clock. For example, DCLKOUT from the Toshiba TMPR3905/12.
- The address inputs AB[12:0] are connected directly to the TMPR3905/12 address bus. Since the TMPR3905/12 has a multiplexed address bus, the other address inputs A[17:13] must be generated using an external latch controlled by the address latch enable signal (ALE). The low data byte on the TMPR3905/12 data bus for 16-bit ports is D[31:24] and connects to the S1D13A04 low data byte, D[7:0]. The high data byte on the TMPR3905/12 data bus for 16-bit ports is D[23:16] and connects to the S1D13A04 high data byte, D[15:0]. The hardware engineer must ensure that CNF4 selects the proper endian mode upon reset.
- Chip Select (CS#) is driven by external decoding circuitry to select the S1D13A04.
- M/R# (memory/register) selects between memory or register accesses. This signal is generated by the external address decode circuitry. For this example, M/R# is connected to address line A18, allowing system address A18 to select between memory or register accesses.
- WE1# is connected to CARD1CSH* and is the high byte enable for both read and write cycles.
- WE0# is connected to CARDIOWR* (the write enable signal) and must be driven low when the Toshiba TMPR3905/12 is writing data to the S1D13A04.
- RD# is connected to CARDIORD* (the read enable signal) and must be driven low when the Toshiba TMPR3905/12 is reading data from the S1D13A04.
- WAIT# connects to CARD1WAIT* and is a signal which is output from the S1D13A04 to the TMPR3905/12 that indicates when data is ready (read cycle) or accepted (write cycle) on the host bus. Since host CPU accesses to the S1D13A04 may occur asynchronously to the display update, it is possible that contention may occur in accessing the S1D13A04 internal registers and/or display buffer. The WAIT# line resolves these contentions by forcing the host to wait until the resource arbitration is complete.
- The Bus Status (BS#) and Read/Write (RD#/WR#) signals are not used in this implementation of the Toshiba TMPR3905/12 using the Generic #2 Host Bus Interface. These pins must be tied high (connected to IO V_{DD}).

4 Toshiba TMPR3905/12 to S1D13A04 Interface

4.1 Hardware Description

In this implementation, the S1D13A04 occupies the TMPR3905/12 PC Card slot #1 IO address space. IO address space closely matches the timing parameters for the S1D13A04 Generic #2 Host Bus Interface.

The address bus of the TMPR3905/12 PC Card interface is multiplexed and must be demultiplexed using an advanced CMOS latch (e.g., 74AHC373).

BS# (bus start) and RD/WR# are not used in this implementation and should be tied high (connected to IO V_{DD}).

A pull-up resistor is attached to WAIT# to speed up its rise time when terminating a cycle.

The following diagram demonstrates a typical implementation of the TMPR3905/12 to S1D13A04 interface.

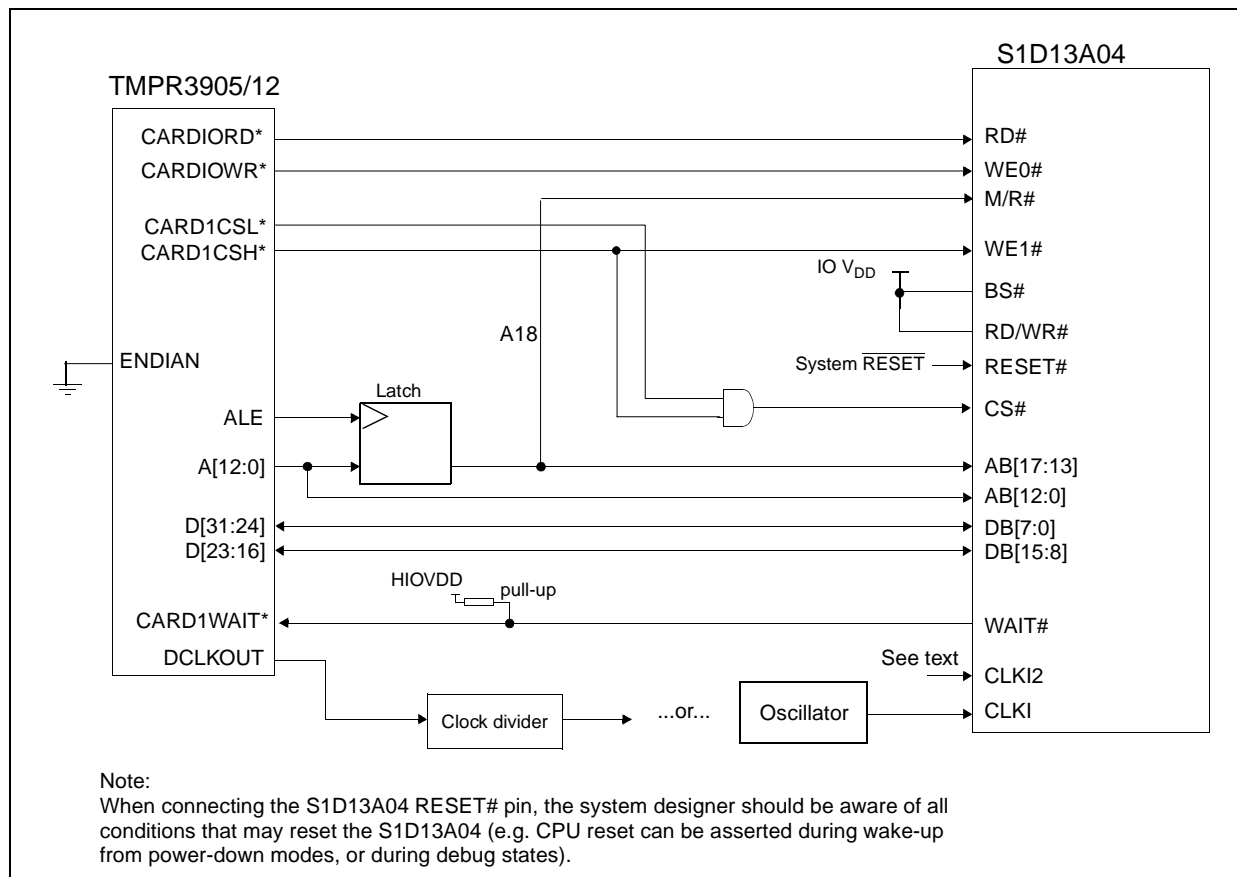


Figure 4-1: S1D13A04 to TMPR3905/12 Direct Connection

The Generic #2 Host Bus Interface control signals of the S1D13A04 are asynchronous with respect to the S1D13A04 bus clock. This gives the system designer full flexibility to choose the appropriate source (or sources) for CLKI and CLKI2. The choice of whether both clocks should be the same, and whether to use DCLKOUT (divided) as clock source, should be based on the desired:

- pixel and frame rates.
- power budget.
- part count.
- maximum S1D13A04 clock frequencies.

The S1D13A04 also has internal clock dividers providing additional flexibility.

4.2 S1D13A04 Hardware Configuration

The S1D13A04 latches CNF6 through CNF0 to allow selection of the bus mode and other configuration data on the rising edge of RESET#. For details on configuration, refer to the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

The table below shows the configuration settings important to the Generic #2 host bus interface used by the Toshiba TMPR3905/12.

Table 4-1: Summary of Power-On/Reset Options

S1D13A04 Configuration Input	Power-On/Reset State											
	1 (connected to IO V _{DD})	0 (connected to V _{SS})										
CNF4, CNF[2:0]	Select host bus interface as follows:											
	<table border="1"> <tr> <td>CNF4</td> <td>CNF2</td> <td>CNF1</td> <td>CNF0</td> <td>Host Bus</td> </tr> <tr> <td>0</td> <td>1</td> <td>0</td> <td>0</td> <td>Generic #2, Little Endian</td> </tr> </table>	CNF4	CNF2	CNF1	CNF0	Host Bus	0	1	0	0	Generic #2, Little Endian	
CNF4	CNF2	CNF1	CNF0	Host Bus								
0	1	0	0	Generic #2, Little Endian								
CNF3	Reserved. Must be set to 1.											
CNF5	WAIT# is active high	WAIT# is active low										
CNF6	CLKI to BCLK divide ratio 2:1	CLKI to BCLK divide ratio 1:1										

configuration for Toshiba TMPR3905/3912 microprocessor

4.3 Memory Mapping and Aliasing

In this implementation the TMPR3905/12 control signal CARDREG* is ignored. This means that the S1D13A04 takes up the entire PC Card slot 1.

The S1D13A04 is a memory mapped device and uses two 256K byte blocks which are selected using A18 from the TMPR3905/12 (A18 is connected to the S1D13A04 M/R# pin). The internal registers occupy the first 256K byte block and the 160K byte display buffer occupies the second 256K byte block.

The registers occupy the range 0h through 3FFFFh while the on-chip display memory occupies the range 40000h through 68000h. Demultiplexed address lines A[25:19] are ignored. Therefore, the S1D13A04 is aliased at 256K byte intervals over the 64M byte PC Card slot #1 memory space.

Note

If aliasing is undesirable, additional decoding circuitry must be added.

5 Software

Test utilities and display drivers are available for the S1D13A04. Full source code is available for both the test utilities and the drivers.

The test utilities are configurable for different panel types using a program called 13A04CFG (see document number X37A-B-001-xx), or by directly modifying the source. The display drivers can be customized by the OEM for different panel types, resolutions and color depths only by modifying the source.

The S1D13A04 test utilities and display drivers are available from your sales support contact (see Section 7, “*Sales and Technical Support*”) or www.erd.epson.com.

6 References

6.1 Documents

- Toshiba America Electrical Components, Inc., *TMPR3905/12 Specification*.
- Epson Research and Development, Inc., *S1D13A04 Hardware Functional Specification*, Document Number X37A-A-001-xx.
- Epson Research and Development, Inc., *S5U13A04B00C Rev. 1.0 ISA Bus Evaluation Board User Manual*, Document Number X37A-G-004-xx.
- Epson Research and Development, Inc., *S1D13A04 Programming Notes and Examples*, Document Number X37A-G-003-xx.

6.2 Document Sources

- Toshiba America Electrical Components Website: www.toshiba.com/taec.
- Epson Research and Development Website: www.erd.epson.com.

7 Sales and Technical Support

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7.2 Toshiba MIPS Tmpr3905/12 Processor

<http://www.toshiba.com/taec/nonflash/indexproducts.html>

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S1D13A04 LCD/USB Companion Chip

Interfacing to the PC Card Bus

Document Number: X37A-G-005-01

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1 Introduction

This application note describes the hardware and software environment required to interface the S1D13A04 LCD/USB Companion Chip and the PC Card (PCMCIA) bus.

The designs described in this document are presented only as examples of how such interfaces might be implemented. This application note is updated as appropriate. Please check the Epson Research and Development website at www.erd.epson.com for the latest revision of this document before beginning any development.

We appreciate your comments on our documentation. Please contact us via email at documentation@erd.epson.com.

2 Interfacing to the PC Card Bus

2.1 The PC Card System Bus

PC Card technology has gained wide acceptance in the mobile computing field as well as in other markets due to its portability and ruggedness. This section is an overview of the operation of the 16-bit PC Card interface conforming to the PCMCIA 2.0/JEIDA 4.1 Standard (or later).

2.1.1 PC Card Overview

The 16-bit PC Card provides a 26-bit address bus and additional control lines which allow access to three 64M byte address ranges. These ranges are used for common memory space, IO space, and attribute memory space. Common memory may be accessed by a host system for memory read and write operations. Attribute memory is used for defining card specific information such as configuration registers, card capabilities, and card use. IO space maintains software and hardware compatibility with hosts such as the Intel x86 architecture, which address peripherals independently from memory space.

Bit notation follows the convention used by most microprocessors, the high bit is the most significant. Therefore, signals A25 and D15 are the most significant bits for the address and data bus respectively.

Support is provided for on-chip DMA controllers. To find further information on these topics, refer to Section 6, “References” on page 15.

PC Card bus signals are asynchronous to the host CPU bus signals. Bus cycles are started with the assertion of either the -CE1 and/or the -CE2 card enable signals. The cycle ends once these signals are de-asserted. Bus cycles can be lengthened using the -WAIT signal.

Note

The PCMCIA 2.0/JEIDA 4.1 (and later) PC Card Standard support the two signals -WAIT and RESET which are not supported in earlier versions of the standard. The -WAIT signal allows for asynchronous data transfers for memory, attribute, and IO access cycles. The RESET signal allows resetting of the card configuration by the reset line of the host CPU.

2.1.2 Memory Access Cycles

A data transfer is initiated when the memory address is placed on the PC Card bus and one, or both, of the card enable signals (-CE1 and -CE2) are driven low. -REG must be kept inactive. If only -CE1 is driven low, 8-bit data transfers are enabled and A0 specifies whether the even or odd data byte appears on data bus lines D[7:0]. If both -CE1 and -CE2 are driven low, a 16-bit word transfer takes place. If only -CE2 is driven low, an odd byte transfer occurs on data lines D[15:8].

During a read cycle, -OE (output enable) is driven low. A write cycle is specified by driving -OE high and driving the write enable signal (-WE) low. The cycle can be lengthened by driving -WAIT low for the time needed to complete the cycle.

Figure 2-1: illustrates a typical memory access read cycle on the PC Card bus.

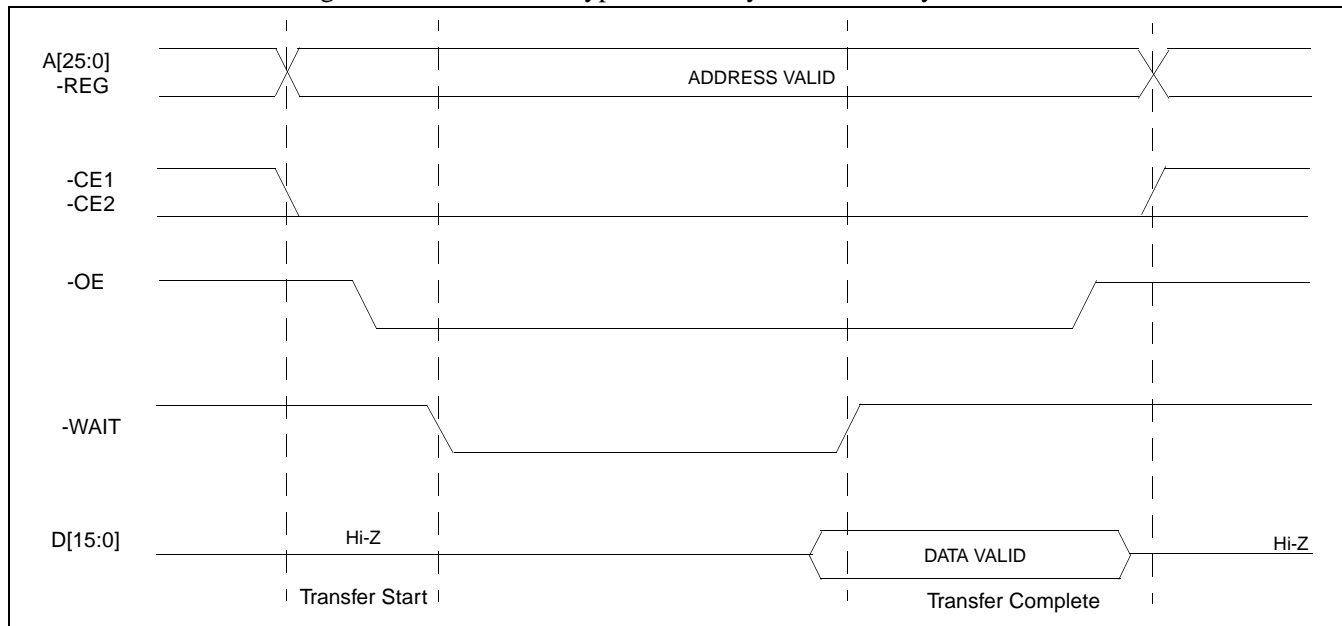


Figure 2-1: PC Card Read Cycle

Figure 2-2: illustrates a typical memory access write cycle on the PC Card bus.

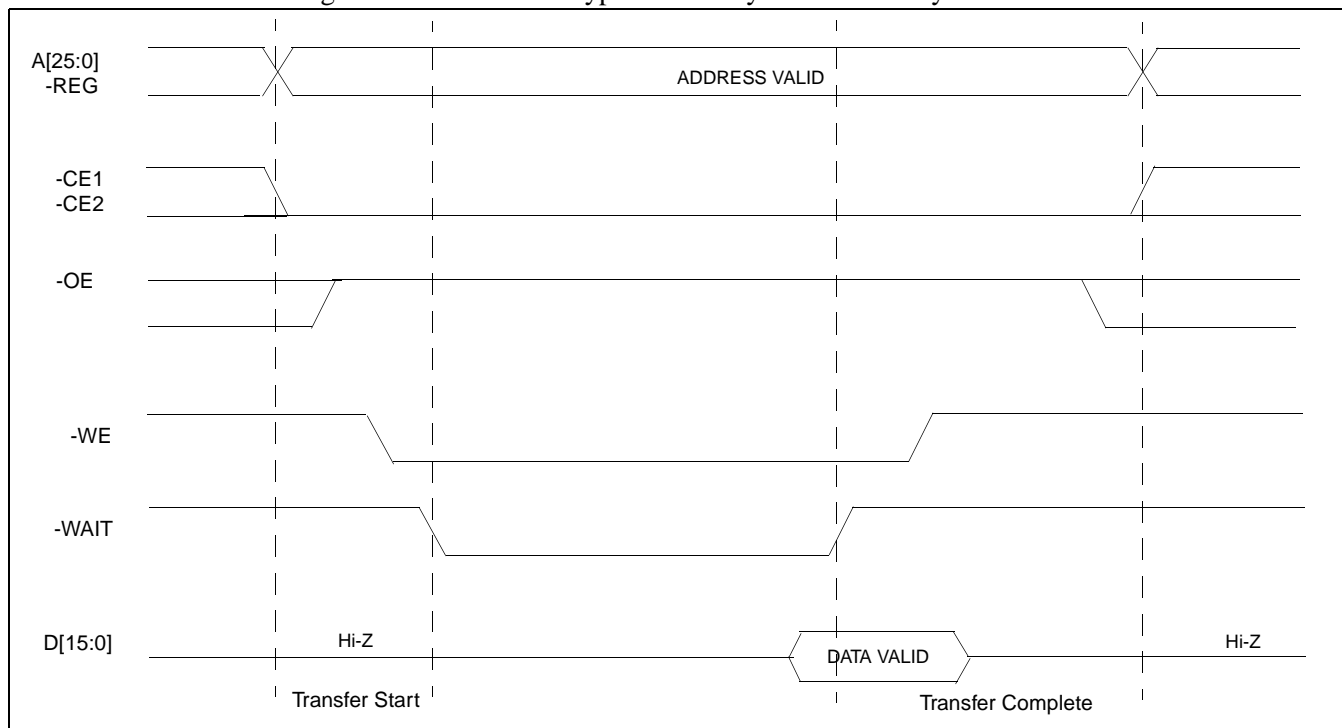


Figure 2-2: PC Card Write Cycle

3 S1D13A04 Host Bus Interface

The S1D13A04 directly supports multiple processors. The S1D13A04 implements a 16-bit Generic #2 Host Bus Interface which is most suitable for direct connection to the PC Card bus. Generic #2 supports an external Chip Select, shared Read/Write Enable for high byte, and individual Read/Write Enable for low byte.

The Generic #2 Host Bus Interface is selected by the S1D13A04 on the rising edge of RESET#. After RESET# is released, the bus interface signals assume their selected configuration. For details on the S1D13A04 configuration, see Section 4.2, “S1D13A04 Hardware Configuration” on page 13.

3.1 Host Bus Interface Pin Mapping

The following table shows the functions of each Host Bus Interface signal.

Table 3-1: Host Bus Interface Pin Mapping

S1D13A04 Pin Names	PC Card (PCMCIA)
AB[17:0]	A[17:0]
DB[15:0]	D[15:0]
WE1#	-CE2
CS#	External Decode
M/R#	A18
CLKI	see note
BS#	Connect to IO _{VDD} from the S1D13A04
RD/WR#	Connect to IO _{VDD} from the S1D13A04
RD#	-OE
WE0#	-WE
WAIT#	-WAIT
RESET#	Inverted RESET

Note

Although a clock is not directly supplied by the PC Card interface, one is required by the S1D13A04 Generic #2 Host Bus Interface. For an example of how this can be accomplished see the discussion on CLKI in Section 3.2, “Host Bus Interface Signals” on page 11.

3.2 Host Bus Interface Signals

The S1D13A04 Generic #2 Host Bus Interface requires the following signals from the PC Card bus.

- CLKI is a clock input which is required by the S1D13A04 Host Bus Interface as a source for its internal bus and memory clocks. This clock is typically driven by the host CPU system clock. Since the PC Card signalling is independent of any clock, CLKI can come from any oscillator already implemented. For example, the source for the CLKI2 input of the S1D13A04 may be used.
- The address inputs AB[17:0], and the data bus DB[15:0], connect directly to the PC Card address (A[17:0]) and data bus (D[15:0]), respectively. CNF4 must be set to select little endian mode.
- Chip Select (CS#) is driven by decoding the high-order address lines to select the proper register and memory address space.
- M/R# (memory/register) selects between memory or register accesses. This signal is generated by the external address decode circuitry. For this example, M/R# is connected to address line A18, allowing system address A18 to select between memory or register accesses.
- WE1# connects to -CE2 (the high byte chip select signal from the PC Card interface) which in conjunction with address bit 0 allows byte steering of read and write operations.
- WE0# connects to -WE (the write enable signal from the PC Card bus) and must be driven low when the PC Card bus is writing data to the S1D13A04.
- RD# connects to -OE (the read enable signal from the PC Card bus) and must be driven low when the PC Card bus is reading data from the S1D13A04.
- WAIT# is a signal output from the S1D13A04 that indicates the PC Card bus must wait until data is ready (read cycle) or accepted (write cycle) on the host bus. Since PC Card bus accesses to the S1D13A04 may occur asynchronously to the display update, it is possible that contention may occur in accessing the S1D13A04 internal registers and/or display buffer. The WAIT# line resolves these contentions by forcing the host to wait until the resource arbitration is complete.
- The Bus Status (BS#) and Read/Write (RD/WR#) signals are not used in this implementation of the PC Card bus using the Generic #2 Host Bus Interface. These pins must be tied high (connected to IO V_{DD}).
- The RESET# (active low) input of the S1D13A04 may be connected to the PC Card RESET (active high) using an inverter.

4 PC Card to S1D13A04 Interface

4.1 Hardware Connections

The S1D13A04 is interfaced to the PC Card bus with a minimal amount of glue logic. In this implementation, the address inputs (AB[17:0]) and data bus (DB[15:0]) connect directly to the CPU address (A[17:0]) and data bus (D[15:0]).

The PC Card interface does not provide a bus clock, so one must be supplied for the S1D13A04. Since the bus clock frequency is not critical, nor does it have to be synchronous to the bus signals, it may be the same as CLKI2.

BS# (bus start) and RD/WR# are not used by the Generic #2 Host Bus Interface and should be tied high (connected to IO V_{DD}).

The following diagram shows a typical implementation of the PC Card to S1D13A04 interface.

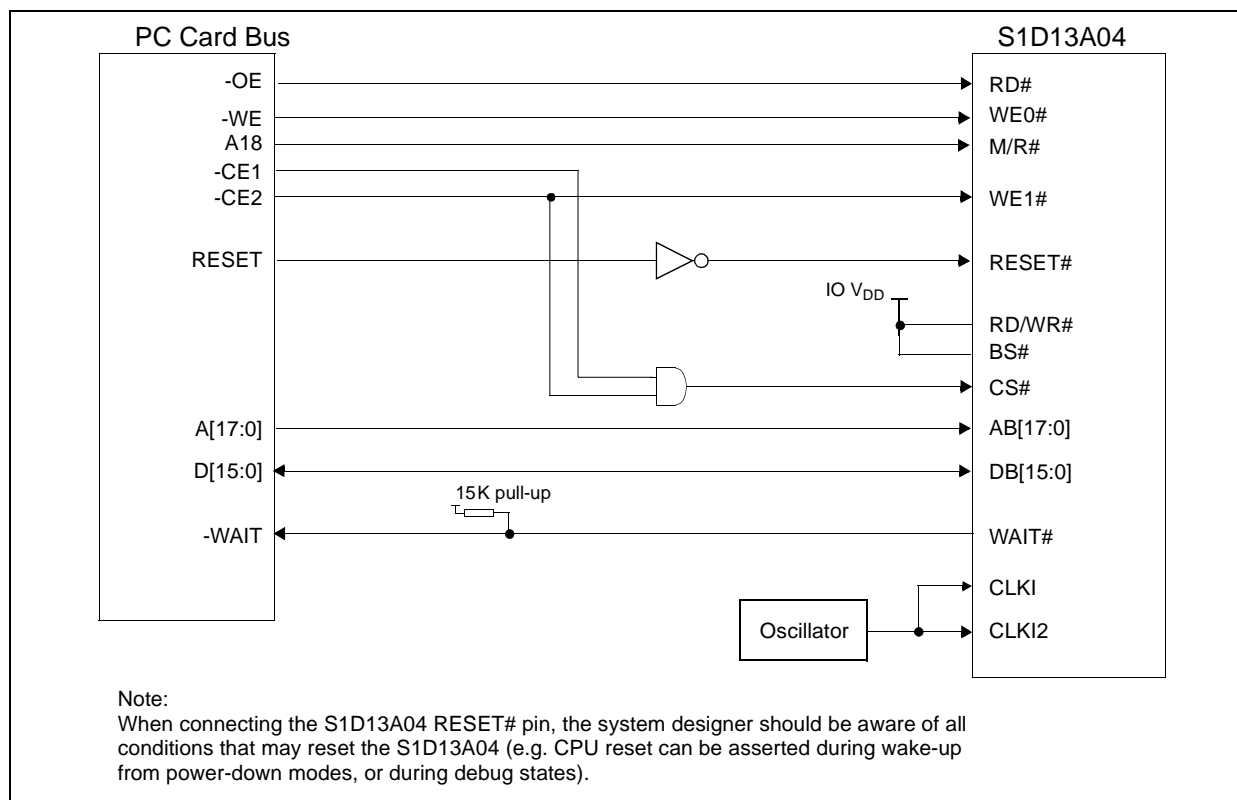


Figure 4-1: Typical Implementation of PC Card to S1D13A04 Interface

4.2 S1D13A04 Hardware Configuration

The S1D13A04 uses CNF6 through CNF0 to allow selection of the bus mode and other configuration data on the rising edge of RESET#. For details on configuration, refer to the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

The following table shows the configuration required for this implementation of a S1D13A04 to PC Card bus interface.

Table 4-1: Summary of Power-On/Reset Options

S1D13A04 Configuration Input	Power-On/Reset State											
	1 (connected to IO V _{DD})	0 (connected to V _{SS})										
CNF4, CNF[2:0]	Select host bus interface as follows: <table border="0"> <tr> <td>CNF4</td> <td>CNF2</td> <td>CNF1</td> <td>CNF0</td> <td>Host Bus</td> </tr> <tr> <td>0</td> <td>1</td> <td>0</td> <td>0</td> <td>Generic #2, Little Endian</td> </tr> </table>		CNF4	CNF2	CNF1	CNF0	Host Bus	0	1	0	0	Generic #2, Little Endian
CNF4	CNF2	CNF1	CNF0	Host Bus								
0	1	0	0	Generic #2, Little Endian								
CNF3	Reserved. Must be set to 1.											
CNF5	WAIT# is active high	WAIT# is active low										
CNF6	CLKI to BCLK divide ratio 2:1	CLKI to BCLK divide ratio 1:1										
	configuration for PC Card bus											

4.3 Register/Memory Mapping

The S1D13A04 is a memory mapped device. The S1D13A04 uses two 256K byte blocks which are selected using A18 from the PC Card bus (A18 is connected to the S1D13A04 M/R# pin). The internal registers occupy the first 256K byte block and the 160K byte display buffer occupies the second 256K byte block.

The PC Card socket provides 64M bytes of memory address space. However, the S1D13A04 only needs a 512K byte block of memory to accommodate its 160K byte display buffer and register set. For this reason, only address bits A[18:0] are used while A[25:19] are ignored. The S1D13A04's memory and registers are aliased every 512K bytes in the 64M byte PC Card memory address space.

Note

If aliasing is not desirable, the upper addresses must be fully decoded.

5 Software

Test utilities and display drivers are available for the S1D13A04. Full source code is available for both the test utilities and the drivers.

The test utilities are configurable for different panel types using a program called 13A04CFG (see document number X37A-B-001-xx), or by directly modifying the source. The display drivers can be customized by the OEM for different panel types, resolutions and color depths only by modifying the source.

The S1D13A04 test utilities and display drivers are available from your sales support contact (see Section 7, “*Sales and Technical Support*”) or www.erd.epson.com.

6 References

6.1 Documents

- PC Card (PCMCIA) Standard March 1997.
- Epson Research and Development, Inc., *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.
- Epson Research and Development, Inc., *S5U13A04B00C Rev. 1.0 Evaluation Board User Manual*, document number X37A-G-004-xx.
- Epson Research and Development, Inc., *S1D13A04 Programming Notes and Examples*, Document Number X37A-G-003-xx.

6.2 Document Sources

- PC Card Website: www.pc-card.com.
- Epson Research and Development Website: www.erd.epson.com.

7 Sales and Technical Support

7.1 EPSON LCD/USB Companion Chips (S1D13A04)

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7.2 PC Card Standard

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S1D13A04 LCD/USB Companion Chip

Power Consumption

Document Number: X37A-G-006-01

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1 S1D13A04 Power Consumption

S1D13A04 power consumption is affected by many system design variables.

- Input clock frequency (CLKI/CLKI2): the CLKI/CLKI2 frequency determines the LCD frame-rate, CPU performance to memory, and other functions – the higher the input clock frequency, the higher the frame-rate, performance and power consumption.
- CPU interface: the S1D13A04 current consumption depends on the BCLK frequency, data width, number of toggling pins, and other factors – the higher the BCLK, the higher the CPU performance and power consumption.
- V_{DD} voltage level: the voltage level affects power consumption – the higher the voltage, the higher the consumption.
- Display mode: the resolution and color depth affect power consumption – the higher the resolution/color depth, the higher the consumption.
- Internal CLK divide: internal registers allow the input clock to be divided before going to the internal logic blocks – the higher the divide, the lower the power consumption.

There is a power save mode in the S1D13A04. The power consumption is affected by various system design variables.

- Clock states during the power save mode: disabling the clocks during power save mode has substantial power savings.

1.1 Conditions

The following table provides an example of some 320x240 panels and the effects on power consumption of specific environments. The following conditions apply.

- All tests had an appropriate LCD panel connected to the LCD outputs of the S1D13A04.
- All tests were run with a static full color palette display.
- All tests were done using the Generic #1 host bus interface (BCLK = 33MHz).

Table 1-1: S1D13A04 Total Power Consumption for 320x240 panels

Test Condition <i>All COREV_{DD} = 2.0V and IOV_{DD} = 3.3V</i>					Power Consumption (mA)				
					S1D13A04 Active		Power Save Mode		
Resolution	Panel Type	Frame Rate	Clocks (MHz)	Color Depth	CORE	IO	CORE ¹	IO ¹	Clocks Removed ²
320x240	Color 8-bit Format 2	67	CLKI = 6 = BCLK = MCLK	4	1.7	0.7	0.7	0.1	0.1
		67	CLKI2 = 6 = PCLK	8	2.1	0.7	0.7	0.1	0.1
		67	USBCLK = 48	16	2.4	0.6	0.7	0.1	0.1
		67	CLKI = 6 = BCLK CLKI2 = 6 = PCLK USBCLK = grounded	8	1.8	0.6	0.4	0.0	0.1
	Color 4-bit	94	CLKI = 33.3 = MCLK = BCLK CLKI2 = grounded USBCLK = 48, PCLK = MCLK / 4	8	4.3	1.3	2.3	0.2	0.1
		94	CLKI = 33.3 = MCLK = BCLK CLKI2 = grounded USBCLK = grounded, PCLK = MCLK / 4	8	4.0	1.2	2.1	0.1	0.1
	18-bit TFT	79	CLKI = 33.3 = MCLK = BCLK CLKI2 = grounded USBCLK = 48, PCLK = MCLK / 4	8	3.7	2.9	2.3	0.2	0.1
		79	CLKI = 33.3 = MCLK = BCLK CLKI2 = grounded USBCLK = grounded, PCLK = MCLK / 4	8	3.5	2.8	2.1	0.1	0.1

1. The S1D13A04 has Power Save Mode enabled, but the clocks (CLKI, CLKI2 and USBCLK) remain active unless specified otherwise.
2. CLKI, CLKI2, and USBCLK are grounded for the Clocks Removed condition.

The following table provides an example of some 160x160 panels and the effects on power consumption of specific environments. The following conditions apply.

- All tests had an appropriate LCD panel connected to the LCD outputs of the S1D13A04.
- All tests were run with a static full color palette display, **except the test where the 2D BitBLT engine was running.**
- All tests were done using the Generic #1 host bus interface (BCLK = 33MHz).

Table 1-2: S1D13A04 Total Power Consumption for 160x160 panels

Test Condition <i>All COREV_{DD} = 2.0V and IOV_{DD} = 3.3V</i>					Power Consumption (mA)				
					S1D13A04 Active		Power Save Mode		
Resolution	Panel Type	Frame Rate	Clocks (MHz)	Color Depth	CORE	IO	CORE ¹	IO ¹	Clocks Removed ²
160x160	18-bit HR-TFT	67	CLKI = 33.3 = MCLK = BCLK	4	2.5	0.9	2.2	0.2	0.1
		67	CLKI2 = 3 = PCLK	8	2.6	1.1	2.2	0.2	0.1
		67	USBCLK = 48	16	2.8	1.1	2.2	0.2	0.1
	67	CLKI = 33.3 = MCLK = BCLK CLKI2 = 3 = PCLK USBCLK = grounded	16	2.6	1.0	4.2	0.1	0.1	
	18-bit HR-TFT	67	CLKI = 33.3 = MCLK = BCLK CLKI2 = 3 = PCLK USBCLK = 48, 2D BitBLT engine running ³	16	12.3	2.4	— ⁵	— ⁵	0.1
	18-bit HR-TFT	67	CLKI = 33.3 = MCLK = BCLK CLKI2 = 3 = PCLK USBCLK = 48, USB is active/running ⁴	16	12.4	1.6	— ⁵	— ⁵	0.1

1. The S1D13A04 has Power Save Mode enabled, but the clocks (CLKI, CLKI2 and USBCLK) remain active unless specified otherwise.
2. CLKI, CLKI2, and USBCLK are grounded for the Clocks Removed condition.
3. This test has the 2D BitBLT engine performing a Move BitBLT which requires a high-level of CPU activity and a rapidly updating display.
4. This test has the S1D13A04 USB module running a loop-back test.
5. This result is not applicable. See the 16 bpp color depth results for power save mode.

2 Summary

The system design variables in Section 1, “S1D13A04 Power Consumption” and in Table 1-1: “S1D13A04 Total Power Consumption for 320x240 panels” show that S1D13A04 power consumption depends on the specific implementation. Active Mode power consumption depends on the desired CPU performance and LCD frame-rate, whereas power save mode consumption depends on the CPU Interface and Input Clock state.

In a typical design environment, the S1D13A04 can be configured to be an extremely power-efficient LCD Controller with high performance and flexibility.

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S1D13A04 LCD/USB Companion Chip

Interfacing to the NEC VR4102 / VR4111 Microprocessors

Document Number: X37A-G-007-01

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1 Introduction

This application note describes the hardware and software environment required to interface the S1D13A04 LCD/USB Companion Chip and the NEC VR4102/4111 microprocessor. The NEC VR4102 and VR4111 microprocessors are specifically designed to support an external LCD controller.

The designs described in this document are presented only as examples of how such interfaces might be implemented. This application note is updated as appropriate. Please check the Epson Research and Development website at www.erd.epson.com for the latest revision of this document before beginning any development.

We appreciate your comments on our documentation. Please contact us via email at documentation@erd.epson.com.

2 Interfacing to the NEC VR4102/VR4111

2.1 The NEC VR41XX System Bus

The VR-Series family of microprocessors features a high-speed synchronous system bus typical of modern microprocessors. Designed with external LCD controller support and Windows® CE based embedded consumer applications in mind, the VR4102/VR4111 offers a highly integrated solution for portable systems. This section is an overview of the operation of the CPU bus to establish interface requirements.

2.1.1 Overview

The NEC VR series microprocessor is designed around the RISC architecture developed by MIPS. The VR4102 microprocessor is designed around the 66MHz VR4100 CPU core and the VR4111 is designed around the 80/100MHz VR4110 core. These microprocessors support 64-bit processing. The CPU communicates with the Bus Control Unit (BCU) through its internal SysAD bus. The BCU in turn communicates with external devices with its ADD and DATA busses which can be dynamically sized for 16 or 32-bit operation.

The NEC VR4102/VR4111 can directly support an external LCD controller through a dedicated bus interface. Specific control signals are assigned for an external LCD controller in order to provide an easy interface to the CPU. A 16M byte block of memory is assigned for the LCD controller with its own chip select and ready signals available. Word or byte accesses are controlled by the system high byte signal (SHB#).

2.1.2 LCD Memory Access Cycles

Once an address in the LCD block of memory is placed on the external address bus (ADD[25:0]) the LCD chip select (LCDCS#) is driven low. The read enable (RD#) or write enable (WR#) signals are driven low for the appropriate cycle. LCDRDY is driven low by the S1D13A04 to insert wait states into the cycle. The system high byte enable is driven low for 16-bit transfers and high for 8-bit transfers.

Figure 2-1: “NEC VR4102/VR4111 Read/Write Cycles,” shows the read and write cycles to the LCD Controller Interface.

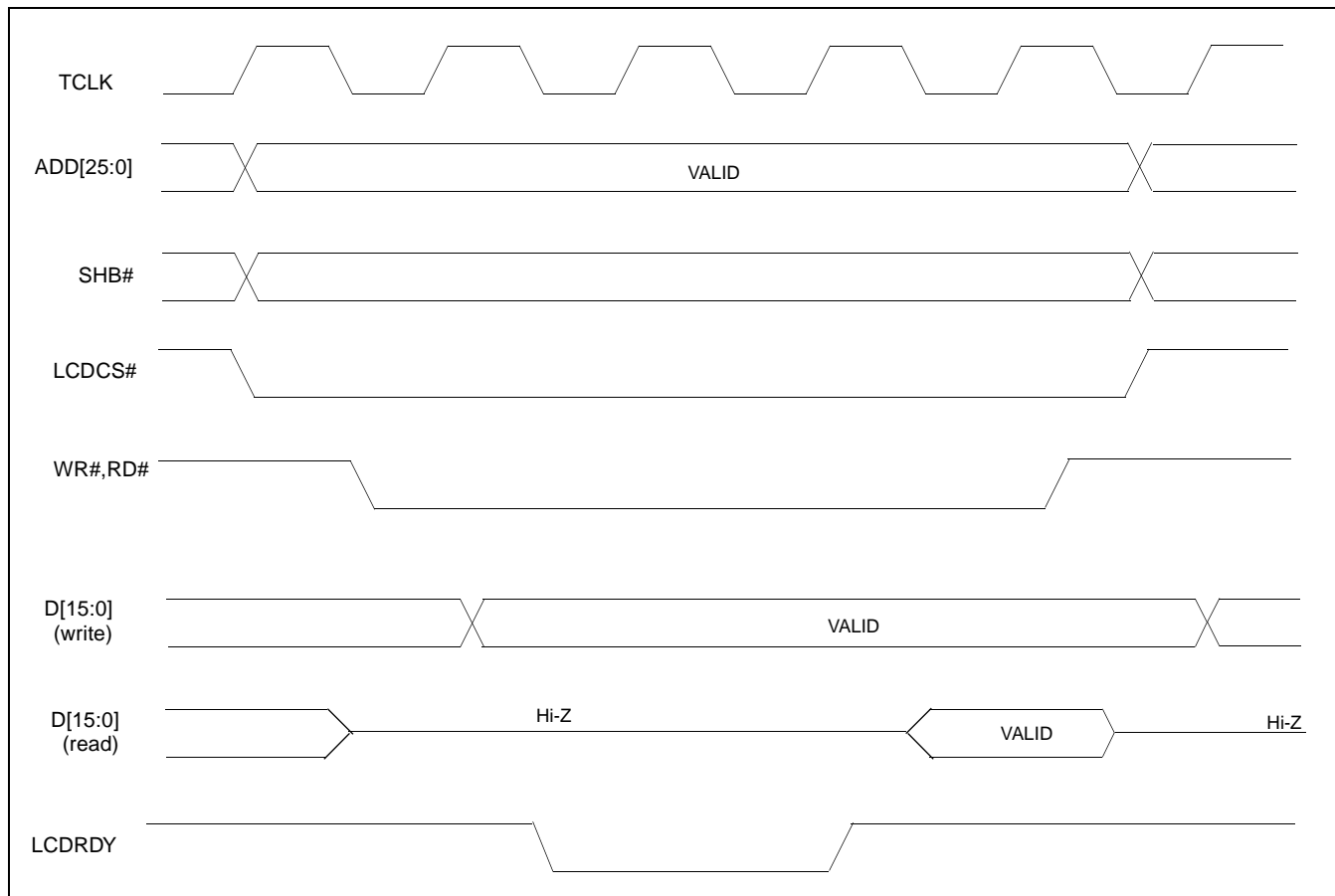


Figure 2-1: NEC VR4102/VR4111 Read/Write Cycles

3 S1D13A04 Host Bus Interface

The S1D13A04 directly supports multiple processors. The S1D13A04 implements a 16-bit Generic #2 Host Bus Interface which is most suitable for direct connection to the NEC VR4102/4111 microprocessor. Generic #2 supports an external Chip Select, shared Read/Write Enable for high byte, and individual Read/Write Enable for low byte.

The Generic #2 Host Bus Interface is selected by the S1D13A04 on the rising edge of RESET#. After RESET# is released, the bus interface signals assume their selected configuration. For details on the S1D13A04 configuration, see Section 4.2, “S1D13A04 Hardware Configuration” on page 13.

3.1 Host Bus Interface Pin Mapping

The following table shows the functions of each Host Bus Interface signal.

Table 3-1: Host Bus Interface Pin Mapping

S1D13A04 Pin Names	NEC VR4102/4111
AB[17:0]	ADD[17:0]
DB[15:0]	DAT[15:0]
WE1#	SHB#
CS#	LCDCS#
M/R#	ADD18
CLKI	BUSCLK
BS#	Connect to IO _{VDD} from the S1D13A04
RD/WR#	Connect to IO _{VDD} from the S1D13A04
RD#	RD#
WE0#	WR#
LCDRDY	WAIT#
RESET#	system $\overline{\text{RESET}}$

3.2 Host Bus Interface Signals

The Host Bus Interface requires the following signals:

- CLKI is a clock input which is required by the S1D13A04 Host Bus Interface as a source for its internal bus and memory clocks. This clock is typically driven by the host CPU system clock. For this example, BUSCLK from the NEC VR4102/4111 is used for CLKI.
- The address inputs AB[17:0], and the data bus DB[15:0], connect directly to the NEC VR4102/4111 address bus (ADD[17:0]) and data bus (DAT[15:0]), respectively. CNF4 must be set to select little endian mode.
- Chip Select (CS#) must be driven low by LCDCS# whenever the S1D13A04 is accessed by the VR4102/4111.
- M/R# (memory/register) selects between memory or register accesses. This signal is generated by the external address decode circuitry. For this example, M/R# is connected to address line ADD18, allowing system address ADD18 to select between memory or register accesses.
- WE1# connects to SHB# (the high byte enable signal from the NEC VR4102/4111) which in conjunction with address bit 0 allows byte steering of read and write operations.
- WE0# connects to WR# (the write enable signal from the NEC VR4102/4111) and must be driven low when the VR4102/4111 is writing data to the S1D13A04.
- RD# connects to RD# (the read enable signal from the NEC VR4102/4111) and must be driven low when the VR4102/4111 is reading data from the S1D13A04.
- WAIT# connects to LCDRDY and is a signal output from the S1D13A04 that indicates the VR4102/VR4111 must wait until data is ready (read cycle) or accepted (write cycle) on the host bus. Since VR4102/VR4111 accesses to the S1D13A04 may occur asynchronously to the display update, it is possible that contention may occur in accessing the S1D13A04 internal registers and/or display buffer. The WAIT# line resolves these contentions by forcing the host to wait until the resource arbitration is complete.
- The Bus Status (BS#) and Read/Write (RD/WR#) signals are not used in this implementation of the NEC VR4102/4111 interface using the Generic #2 Host Bus Interface. These pins must be tied high (connected to IO V_{DD}).

4 VR4102/VR4111 to S1D13A04 Interface

4.1 Hardware Description

The NEC VR4102/VR4111 microprocessor is specifically designed to support an external LCD controller by providing the internal address decoding and control signals necessary. By using the Generic # 2 Host Bus Interface, no glue logic is required to interface the S1D13A04 and the NEC VR4102/VR4111.

A pull-up resistor is attached to WAIT# to speed up its rise time when terminating a cycle.

BS# (bus start) and RD/WR# are not used by the Generic #2 Host Bus Interface and should be tied high (connected to IO V_{DD}).

The following diagram shows a typical implementation of the VR4102/VR4111 to S1D13A04 interface.

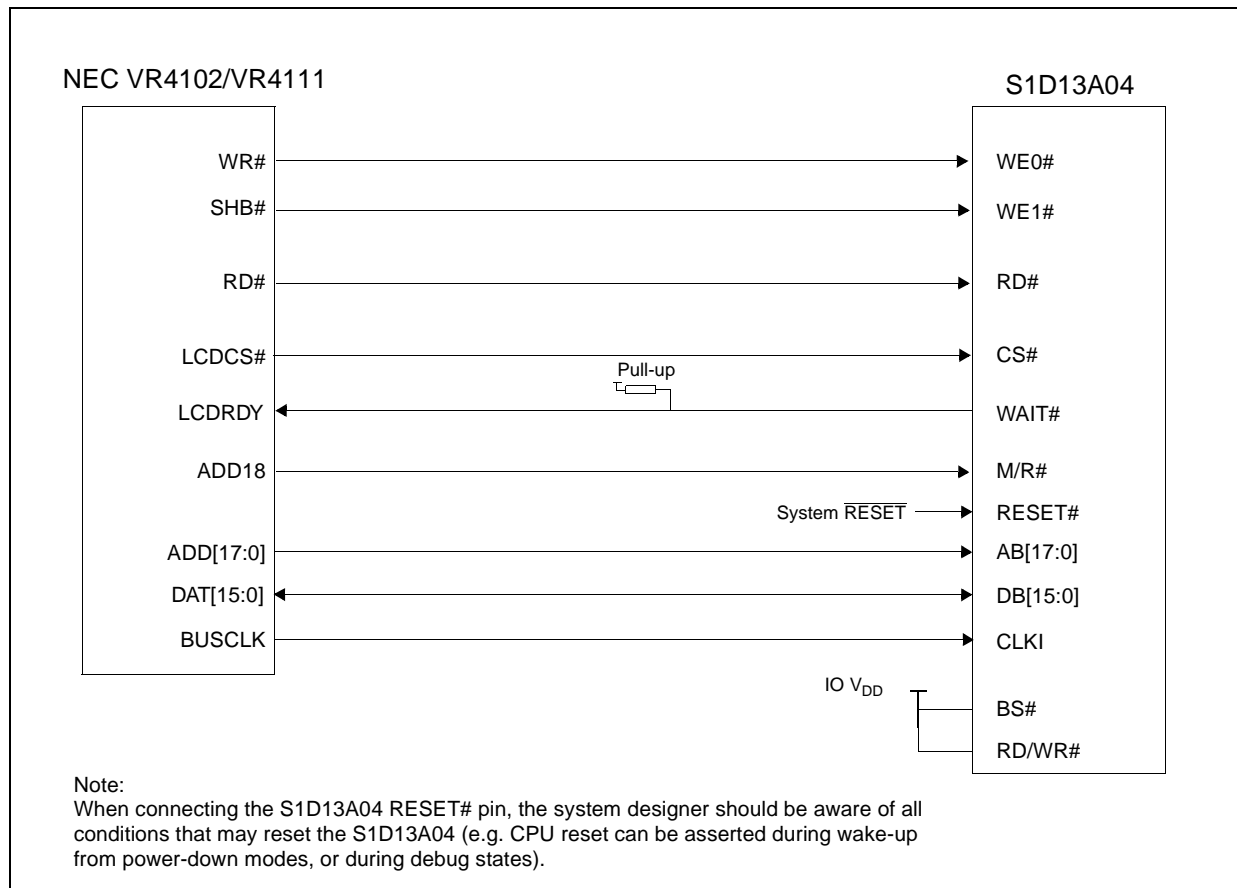


Figure 4-1: Typical Implementation of VR4102/VR4111 to S1D13A04 Interface

4.2 S1D13A04 Hardware Configuration

The S1D13A04 uses CNF6 through CNF0 to allow selection of the bus mode and other configuration data on the rising edge of RESET#. For details on configuration, refer to the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

The following table shows the configuration required for this implementation of a S1D13A04 to NEC VR4102/4111 interface.

Table 4-1: Summary of Power-On/Reset Options

S1D13A04 Configuration Input	Power-On/Reset State											
	1 (connected to IO V _{DD})	0 (connected to V _{SS})										
CNF4, CNF[2:0]	Select host bus interface as follows:											
	<table border="1"> <tr> <td>CNF4</td> <td>CNF2</td> <td>CNF1</td> <td>CNF0</td> <td>Host Bus</td> </tr> <tr> <td>0</td> <td>1</td> <td>0</td> <td>0</td> <td>Generic #2, Little Endian</td> </tr> </table>	CNF4	CNF2	CNF1	CNF0	Host Bus	0	1	0	0	Generic #2, Little Endian	
CNF4	CNF2	CNF1	CNF0	Host Bus								
0	1	0	0	Generic #2, Little Endian								
CNF3	Reserved. Must be set to 1.											
CNF5	WAIT# is active high	WAIT# is active low										
CNF6	CLKI to BCLK divide ratio 2:1	CLKI to BCLK divide ratio 1:1										
	configuration for NEC VR4102/VR4111 microprocessor											

4.3 NEC VR4102/VR4111 Configuration

The NEC VR4102/4111 provides the internal address decoding necessary to map an external LCD controller. Physical address 0A00_0000h to 0AFF_FFFFh (16M bytes) is reserved for an external LCD controller by the NEC VR4102/4111.

The S1D13A04 is a memory mapped device. The S1D13A04 uses two 256K byte blocks which are selected using ADD18 from the NEC VR4102/4111 (ADD18 is connected to the S1D13A04 M/R# pin). The internal registers occupy the first 256K byte block and the 160K byte display buffer occupies the second 256K byte block.

The starting address of the S1D13A04 internal registers is located at 0A00_0000h and the starting address of the display buffer is located at 0A04_0000h. These blocks are aliased over the entire 16M byte address space.

Note

If aliasing is not desirable, the upper addresses must be fully decoded.

The NEC VR4102/VR4111 has a 16-bit internal register named BCUCNTREG2 located at 0B00_0002h. It must be set to the value of 0001h which indicates that LCD controller accesses use a non-inverting data bus.

The 16-bit internal register named BCUCNTREG1 (located at 0B00_0000h) must have bit D[13] (ISA/LCD bit) set to 0. This reserves 16M bytes (from 0A00_0000h to 0AFF_FFFFh) for use by the LCD controller and not as ISA bus memory space.

5 Software

Test utilities and display drivers are available for the S1D13A04. Full source code is available for both the test utilities and the drivers.

The test utilities are configurable for different panel types using a program called 13A04CFG (see document number X37A-B-001-xx), or by directly modifying the source. The display drivers can be customized by the OEM for different panel types, resolutions and color depths only by modifying the source.

The S1D13A04 test utilities and display drivers are available from your sales support contact (see Section 7, “*Sales and Technical Support*”) or www.erd.epson.com.

6 References

6.1 Documents

- NEC Electronics Inc., *VR4102/VR4111 64/32-bit Microprocessor Preliminary User's Manual*.
- Epson Research and Development, Inc., *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.
- Epson Research and Development, Inc., *S5U13A04B00C Rev. 1.0 Evaluation Board User Manual*, document number X37A-G-004-xx.
- Epson Research and Development, Inc., *S1D13A04 Programming Notes and Examples*, document number X37A-G-003-xx.

6.2 Document Sources

- NEC Electronics Inc. Website: www.necel.com.
- Epson Research and Development Website: www.erd.epson.com.

7 Sales and Technical Support

7.1 EPSON LCD/USB Companion Chips (S1D13A04)

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S1D13A04 LCD/USB Companion Chip

Interfacing to the NEC VR4181A™ Microprocessor

Document Number: X37A-G-008-01

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1 Introduction

This application note describes the hardware and software environment required to interface the S1D13A04 LCD/USB Companion Chip and the NEC VR4181A microprocessor. The NEC VR4181A microprocessor is specifically designed to support an external LCD controller.

The designs described in this document are presented only as examples of how such interfaces might be implemented. This application note is updated as appropriate. Please check the Epson Research and Development website at www.erd.epson.com for the latest revision of this document before beginning any development.

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2 Interfacing to the NEC VR4181A

2.1 The NEC VR4181A System Bus

The VR-Series family of microprocessors features a high-speed synchronous system bus typical of modern microprocessors. Designed with external LCD controller support and Windows® CE based embedded consumer applications in mind, the VR4181A offers a highly integrated solution for portable systems. This section is an overview of the operation of the CPU bus to establish interface requirements.

2.1.1 Overview

The NEC VR4181A is designed around the RISC architecture developed by MIPS. This microprocessor is designed around the 100MHz VR4110 CPU core which supports the MIPS III and MIPS16 instruction sets. The CPU communicates with external devices via an ISA interface.

While the VR4181A has an embedded LCD controller, this internal controller can be disabled to provide direct support for an external LCD controller through its external ISA bus. A 64 to 512K byte block of memory is assigned to the external LCD controller with a dedicated chip select signal (LCDCS#). Word or byte accesses are controlled by the system high byte signal (#UBE).

2.1.2 LCD Memory Access Signals

The S1D13A04 requires an addressing range of 256K bytes. When the VR4181A external LCD controller chip select signal is programmed to a window of that size, the S1D13A04 resides in the VR4181A physical address range of 133C 0000h to 133F FFFFh. This range is part of the external ISA memory space.

The following signals are required to access an external LCD controller. All signals obey ISA signalling rules.

- A[16:0] is the address bus.
- #UBE is the high byte enable (active low).
- #LCDCS is the chip select for the S1D13A04 (active low).
- D[15:0] is the data bus.
- #MEMRD is the read command (active low).
- #MEMWR is the write command (active low).
- #MEMCS16 is the acknowledge for 16-bit peripheral capability (active low).
- IORDY is the ready signal from S1D13A04.
- SYCLK is the prescalable bus clock (optional).

Once an address in the LCD block of memory is accessed, the LCD chip select (#LCDCS) is driven low. The read or write enable signals (#MEMRD or #MEMWR) are driven low for the appropriate cycle and IORDY is driven low by the S1D13A04 to insert wait states into the cycle. The high byte enable (UBE#) is driven low for 16-bit transfers and high for 8-bit transfers.

3 S1D13A04 Host Bus Interface

The S1D13A04 directly supports multiple processors. The S1D13A04 implements a 16-bit Generic #2 Host Bus Interface which is most suitable for direct connection to the NEC VR4181A microprocessor. Generic #2 supports an external Chip Select, shared Read/Write Enable for high byte, and individual Read/Write Enable for low byte.

The Generic #2 Host Bus Interface is selected by the S1D13A04 on the rising edge of RESET#. After RESET# is released, the bus interface signals assume their selected configuration. For details on the S1D13A04 configuration, see Section 4.2, “S1D13A04 Hardware Configuration” on page 13.

3.1 Host Bus Interface Pin Mapping

The following table shows the functions of each Host Bus Interface signal.

Table 3-1: Host Bus Interface Pin Mapping

S1D13A04 Pin Names	NEC VR4181A
AB[17:0]	A[17:0]
DB[15:0]	D[15:0]
WE1#	#UBE
CS#	#LCDCS
M/R#	A18
CLKI	SYSCLK
BS#	Connect to IO _{VDD} from the S1D13A04
RD/WR#	Connect to IO _{VDD} from the S1D13A04
RD#	#MEMRD
WE0#	#MEMWR
WAIT#	IORDY
RESET#	RESET#

3.2 Host Bus Interface Signals

The interface requires the following signals.

- CLKI is a clock input which is required by the S1D13A04 Host Bus Interface as a source for its internal bus and memory clocks. This clock is typically driven by the host CPU system clock. For this example, SYSCLK from the NEC VR4181A is used for CLKI.
- The address inputs AB[17:0], and the data bus DB[15:0], connect directly to the NEC VR4181A address (A[17:0]) and data bus (D[15:0]), respectively. CNF4 must be set to select little endian mode.
- Chip Select (CS#) must be driven low by #LCDCS whenever the S1D13A04 is accessed by the VR4181A.
- M/R# (memory/register) selects between memory or register accesses. This signal is generated by the external address decode circuitry. For this example, M/R# is connected to address line A18, allowing system address A18 to select between memory or register accesses.
- WE1# connects to #UBE (the high byte enable signal from the NEC VR4181A) which in conjunction with address bit 0 allows byte steering of read and write operations.
- WE0# connects to #MEMWR (the write enable signal from the NEC VR4181A) and must be driven low when the NEC VR4181A is writing data to the S1D13A04.
- RD# connects to #MEMRD (the read enable signal from the NEC VR4181A) and must be driven low when the NEC VR4181A is reading data from the S1D13A04.
- WAIT# connects to IORDY and is a signal which is output from the S1D13A04 which indicates the NEC VR4181A must wait until data is ready (read cycle) or accepted (write cycle) on the host bus. Since VR4181A accesses to the S1D13A04 may occur asynchronously to the display update, it is possible that contention may occur in accessing the S1D13A04 internal registers and/or display buffer. The WAIT# line resolves these contentions by forcing the host to wait until the resource arbitration is complete.
- The Bus Status (BS#) and Read/Write (RD/WR#) signals are not used in this implementation of the NEC VR4181A interface using the Generic #2 Host Bus Interface. These pins must be tied high (connected to IO V_{DD}).

4 VR4181A to S1D13A04 Interface

4.1 Hardware Description

The NEC VR4181A microprocessor is specifically designed to support an external LCD controller by providing the internal address decoding and control signals necessary. By using the Generic # 2 Host Bus Interface, no glue logic is required to interface the S1D13A04 to the NEC VR4181A.

A pull-up resistor is attached to WAIT# to speed up its rise time when terminating a cycle.

#MEMCS16 of the NEC VR4181A is connected to #LCDCS to signal that the S1D13A04 is capable of 16-bit transfers.

BS# (bus start) and RD/WR# are not used by the Generic #2 Host Bus Interface and should be tied high (connected to IO V_{DD}).

The diagram below shows a typical implementation of the VR4181A to S1D13A04 interface.

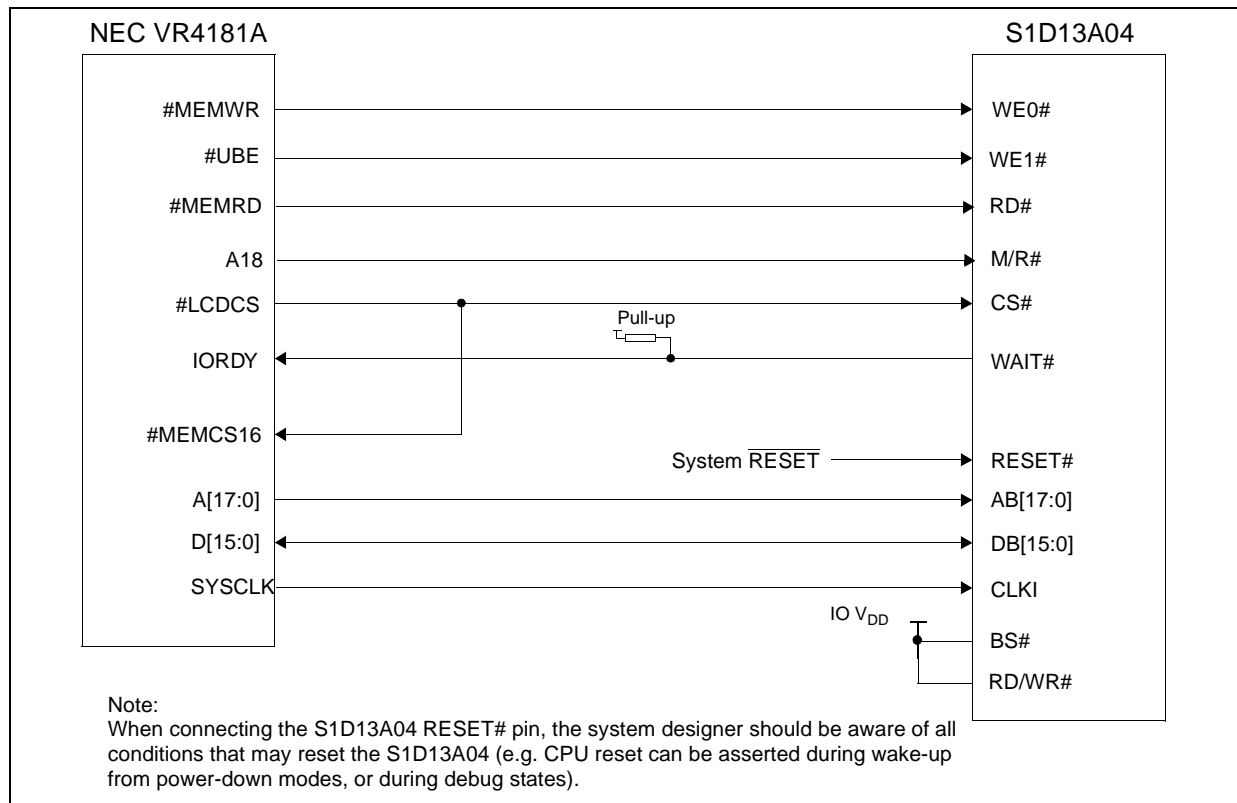


Figure 4-1: Typical Implementation of VR4181A to S1D13A04 Interface

4.2 S1D13A04 Hardware Configuration

The S1D13A04 uses CNF6 through CNF0 to allow selection of the bus mode and other configuration data on the rising edge of RESET#. For details on configuration, refer to the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

The following table shows the configuration required for this implementation of a S1D13A04 to NEC VR4181A interface.

Table 4-1: Summary of Power-On/Reset Options

S1D13A04 Configuration Input	Power-On/Reset State											
	1 (connected to IO V _{DD})	0 (connected to V _{SS})										
CNF4, CNF[2:0]	Select host bus interface as follows: <table border="0" style="width: 100%;"> <tr> <td style="text-align: center;">CNF4</td> <td style="text-align: center;">CNF2</td> <td style="text-align: center;">CNF1</td> <td style="text-align: center;">CNF0</td> <td style="text-align: center;">Host Bus</td> </tr> <tr> <td style="text-align: center;">0</td> <td style="text-align: center;">1</td> <td style="text-align: center;">0</td> <td style="text-align: center;">0</td> <td style="text-align: center;">Generic #2, Little Endian</td> </tr> </table>		CNF4	CNF2	CNF1	CNF0	Host Bus	0	1	0	0	Generic #2, Little Endian
CNF4	CNF2	CNF1	CNF0	Host Bus								
0	1	0	0	Generic #2, Little Endian								
CNF3	Reserved. Must be set to 1.											
CNF5	WAIT# is active high	WAIT# is active low										
CNF6	CLKI to BCLK divide ratio 2:1	CLKI to BCLK divide ratio 1:1										
	configuration for NEC VR4181A microprocessor											

4.3 NEC VR4181A Configuration

The S1D13A04 is a memory mapped device. The S1D13A04 uses two 256K byte blocks which are selected using A18 from the NEC VR4181A (A18 is connected to the S1D13A04 M/R# pin). The internal registers occupy the first 256K byte block and the 160K byte display buffer occupies the second 256K byte block.

When the VR4181A embedded LCD controller is disabled, the external LCD controller chip select signal (#LCDCS) decodes either a 64K byte, 128K byte, 256K byte, or 512K byte memory block in the VR4181A external ISA memory. The S1D13A04 requires this block of memory to be set to 512K bytes. With this configuration, the S1D13A04 internal registers starting address is located at physical memory location 133C_0000h and the display buffer is located at memory location 1340_0000h.

The NEC VR4181A must be configured through its internal registers to map the S1D13A04 to the external LCD controller space. The following register values must be set.

- Register LCDGPMD at address 0B00_032Eh must be set as follows.
 - Bit 7 must be set to 1 to disable the internal LCD controller and enable the external LCD controller interface. Disabling the internal LCD controller also maps pin SHCLK to #LCDCS and pin LOCLK to #MEMCS16.
 - Bits [1:0] must be set to 11b to reserve 512Kbytes of memory address range, 133C_0000h to 1343_FFFFh for the external LCD controller.
- Register GPMD2REG at address 0B00_0304h must be set as follows.
 - Bits [9:8] (GP20MD[1:0]) must be set to 11'b to map pin GPIO20 to #UBE.
 - Bits [5:4] (GP18MD[1:0]) must be set to 01'b to map pin GPIO18 to IORDY.

5 Software

Test utilities and display drivers are available for the S1D13A04. Full source code is available for both the test utilities and the drivers.

The test utilities are configurable for different panel types using a program called 13A04CFG (see document number X37A-B-001-xx), or by directly modifying the source. The display drivers can be customized by the OEM for different panel types, resolutions and color depths only by modifying the source.

The S1D13A04 test utilities and display drivers are available from your sales support contact (see Section 7, “*Sales and Technical Support*”) or www.erd.epson.com.

6 References

6.1 Documents

- NEC Electronics Inc., *NEC VR4181A Target Specification*, Revision 0.5, 9/11/98
- Epson Research and Development, Inc., *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.
- Epson Research and Development, Inc., *S5U13A04B00C Rev. 1.0 Evaluation Board User Manual*, document number X37A-G-004-xx.
- Epson Research and Development, Inc., *S1D13A04 Programming Notes and Examples*, document number X37A-G-003-xx.

6.2 Document Sources

- NEC Electronics Inc. Website: www.necel.com.
- Epson Research and Development Website: www.erd.epson.com.

7 Sales and Technical Support

7.1 EPSON LCD/USB Companion Chips (S1D13A04)

Japan

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EPSON®



S1D13A04 LCD/USB Companion Chip

Interfacing to the Motorola MPC82x Microprocessor

Document Number: X37A-G-009-01

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1 Introduction

This application note describes the hardware and software environment required to interface the S1D13A04 LCD/USB Companion Chip and the Motorola MPC82x microprocessor.

The designs described in this document are presented only as examples of how such interfaces might be implemented. This application note is updated as appropriate. Please check the Epson Research and Development website at www.erd.epson.com for the latest revision of this document before beginning any development.

We appreciate your comments on our documentation. Please contact us via email at documentation@erd.epson.com.

2 Interfacing to the MPC82x

2.1 The MPC8xx System Bus

The MPC8xx family of processors feature a high-speed synchronous system bus typical of modern RISC microprocessors. This section provides an overview of the operation of the CPU bus in order to establish interface requirements.

2.2 MPC8xx Bus Overview

The MPC8xx microprocessor family uses a synchronous address and data bus. All IO is synchronous to a square-wave reference clock called MCLK (Master Clock). This clock runs at the machine cycle speed of the CPU core (typically 25 to 50 MHz). Most outputs from the processor change state on the rising edge of this clock. Similarly, most inputs to the processor are sampled on the rising edge.

Note

The external bus can run at one-half the CPU core speed using the clock control register. This is typically used when the CPU core is operated above 50 MHz.

The MPC821 can generate up to eight independent chip select outputs, each of which may be controlled by one of two types of timing generators: the General Purpose Chip Select Module (GPCM) or the User-Programmable Machine (UPM). Examples are given using the GPCM.

It should be noted that all Power PC microprocessors, including the MPC8xx family, use bit notation opposite from the convention used by most other microprocessor systems. Bit numbering for the MPC8xx always starts with zero as the most significant bit, and increments in value to the least-significant bit. For example, the most significant bits of the address bus and data bus are A0 and D0, while the least significant bits are A31 and D31.

The MPC8xx uses both a 32-bit address and data bus. A parity bit is supported for each of the four byte lanes on the data bus. Parity checking is done when data is read from external memory or peripherals, and generated by the MPC8xx bus controller on write cycles. All IO accesses are memory-mapped meaning there is no separate IO space in the Power PC architecture.

Support is provided for both on-chip (DMA controllers) and off-chip (other processors and peripheral controllers) bus masters. For further information on this topic, refer to Section 6, "References" on page 22.

The bus can support both normal and burst cycles. Burst memory cycles are used to fill on-chip cache memory, and for certain on-chip DMA operations. Normal cycles are used for all other data transfers.

2.2.1 Normal (Non-Burst) Bus Transactions

A data transfer is initiated by the bus master by placing the memory address on address lines A0 through A31 and driving \overline{TS} (Transfer Start) low for one clock cycle. Several control signals are also provided with the memory address:

- $TSIZ[0:1]$ (Transfer Size) — indicates whether the bus cycle is 8, 16, or 32-bit.
- RD/\overline{WR} — set high for read cycles and low for write cycles.
- $AT[0:3]$ (Address Type Signals) — provides more detail on the type of transfer being attempted.

When the peripheral device being accessed has completed the bus transfer, it asserts \overline{TA} (Transfer Acknowledge) for one clock cycle to complete the bus transaction. Once \overline{TA} has been asserted, the MPC821 will not start another bus cycle until \overline{TA} has been de-asserted. The minimum length of a bus transaction is two bus clocks.

Figure 2-1: “Power PC Memory Read Cycle” illustrates a typical memory read cycle on the Power PC system bus.

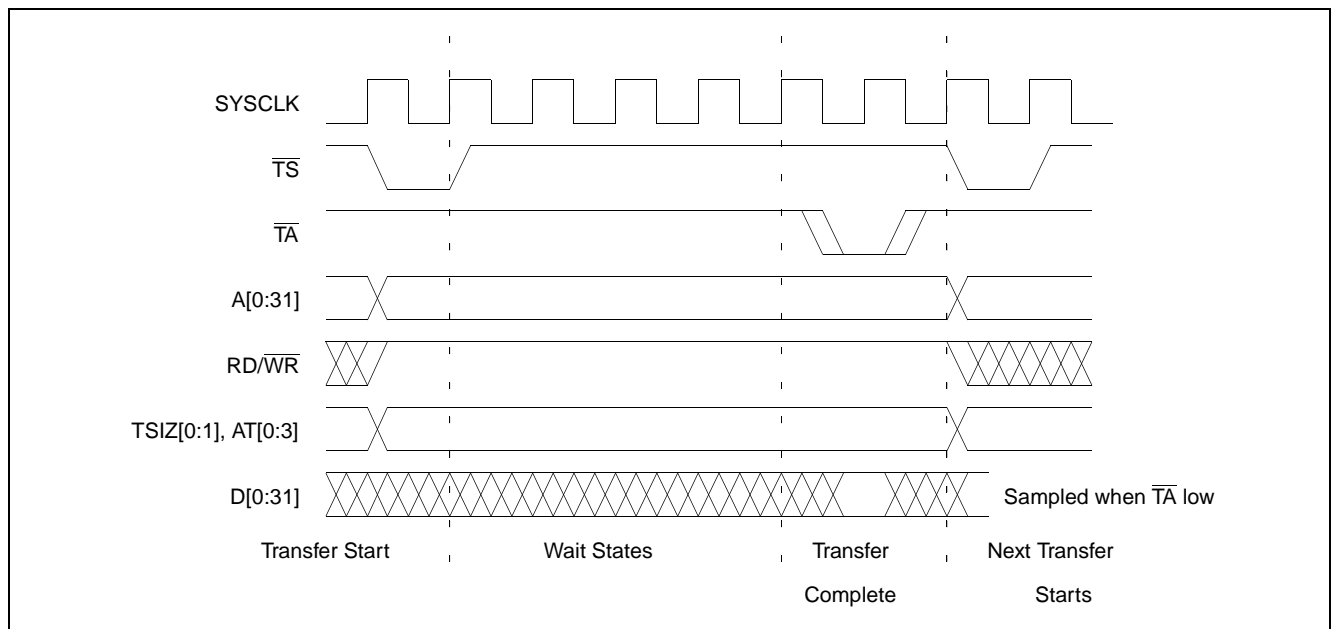


Figure 2-1: Power PC Memory Read Cycle

Figure 2-2: “Power PC Memory Write Cycle” illustrates a typical memory write cycle on the Power PC system bus.

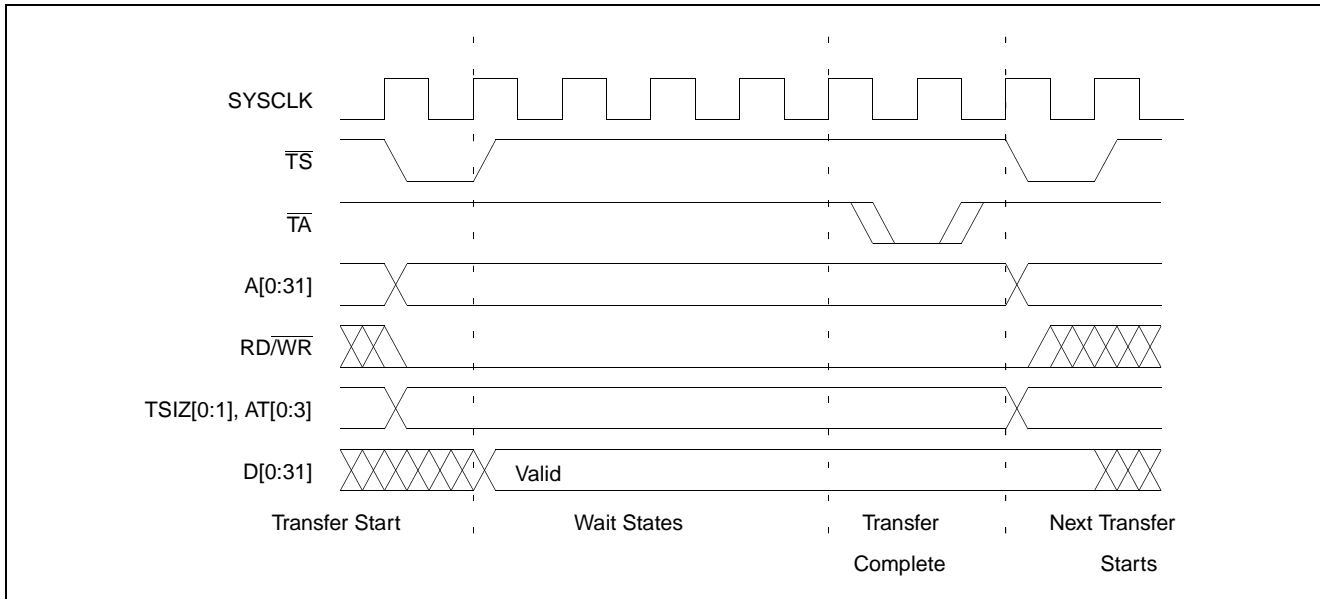


Figure 2-2: Power PC Memory Write Cycle

If an error occurs, \overline{TEA} (Transfer Error Acknowledge) is asserted and the bus cycle is aborted. For example, a peripheral device may assert \overline{TEA} if a parity error is detected, or the MPC821 bus controller may assert \overline{TEA} if no peripheral device responds at the addressed memory location within a bus time-out period.

For 32-bit transfers, all data lines (D[0:31]) are used and the two low-order address lines A30 and A31 are ignored. For 16-bit transfers, data lines D0 through D15 are used and address line A31 is ignored. For 8-bit transfers, data lines D0 through D7 are used and all address lines (A[0:31]) are used.

Note

This assumes that the Power PC core is operating in big endian mode (typically the case for embedded systems).

2.2.2 Burst Cycles

Burst memory cycles are used to fill on-chip cache memory and to carry out certain on-chip DMA operations. They are very similar to normal bus cycles with the following exceptions:

- Always 32-bit.
- Always attempt to transfer four 32-bit words sequentially.
- Always address longword-aligned memory (i.e. A30 and A31 are always 0:0).
- Do not increment address bits A28 and A29 between successive transfers; the addressed device must increment these address bits internally.

If a peripheral is not capable of supporting burst cycles, it can assert Burst Inhibit ($\overline{\text{BI}}$) simultaneously with $\overline{\text{TA}}$, and the processor reverts to normal bus cycles for the remaining data transfers.

Burst cycles are mainly intended to facilitate cache line fills from program or data memory. They are normally not used for transfers to/from IO peripheral devices such as the S1D13A04, therefore the interfaces described in this document do not attempt to support burst cycles.

2.3 Memory Controller Module

2.3.1 General-Purpose Chip Select Module (GPCM)

The General-Purpose Chip Select Module (GPCM) is used to control memory and peripheral devices which do not require special timing or address multiplexing. In addition to the chip select output, it can generate active-low Output Enable ($\overline{\text{OE}}$) and Write Enable ($\overline{\text{WE}}$) signals compatible with most memory and x86-style peripherals. The MPC821 bus controller also provides a Read/Write ($\text{RD}/\overline{\text{WR}}$) signal which is compatible with most 68K peripherals.

The GPCM is controlled by the values programmed into the Base Register (BR) and Option Register (OR) of the respective chip select. The Option Register sets the base address, the block size of the chip select, and controls the following timing parameters:

- The ACS bit field allows the chip select assertion to be delayed with respect to the address bus valid, by 0, 1/4, or 1/2 clock cycle.
- The CSNT bit causes chip select and $\overline{\text{WE}}$ to be negated 1/2 clock cycle earlier than normal.
- The TRLX (relaxed timing) bit inserts an additional one clock delay between assertion of the address bus and chip select. This accommodates memory and peripherals with long setup times.
- The EHTR (Extended hold time) bit inserts an additional 1-clock delay on the first access to a chip select.
- Up to 15 wait states may be inserted, or the peripheral can terminate the bus cycle itself by asserting $\overline{\text{TA}}$ (Transfer Acknowledge).
- Any chip select may be programmed to assert $\overline{\text{BI}}$ (Burst Inhibit) automatically when its memory space is addressed by the processor core.

Figure 2-3: “GPCM Memory Devices Timing” illustrates a typical cycle for a memory mapped device using the GPCM of the Power PC.

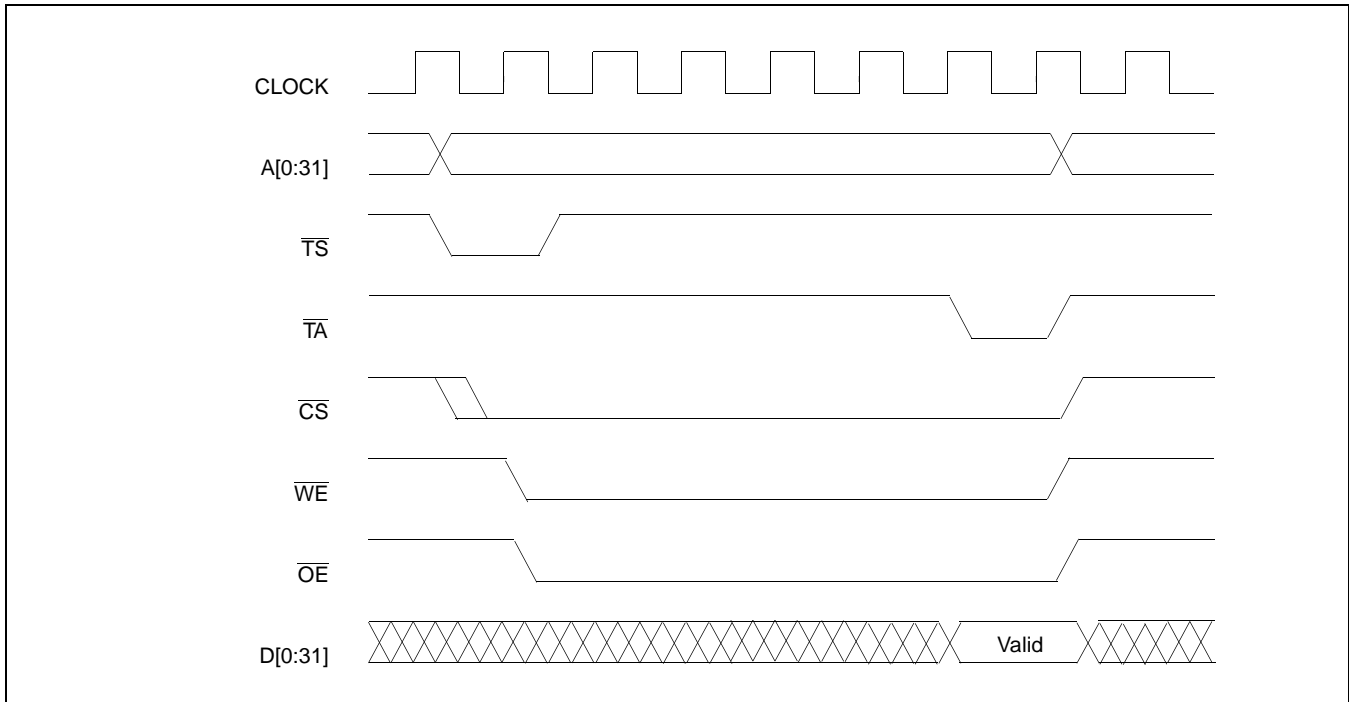


Figure 2-3: GPCM Memory Devices Timing

2.3.2 User-Programmable Machine (UPM)

The UPM is typically used to control memory types, such as Dynamic RAMs, which have complex control or address multiplexing requirements. The UPM is a general purpose RAM-based pattern generator which can control address multiplexing, wait state generation, and five general-purpose output lines on the MPC821. Up to 64 pattern locations are available, each 32 bits wide. Separate patterns may be programmed for normal accesses, burst accesses, refresh (timer) events, and exception conditions. This flexibility allows almost any type of memory or peripheral device to be accommodated by the MPC821.

In this application note, the GPCM is used instead of the UPM, since the GPCM has enough flexibility to accommodate the S1D13A04 and it is desirable to leave the UPM free to handle other interfacing duties, such as EDO DRAM.

3 S1D13A04 Host Bus Interface

The S1D13A04 directly supports multiple processors. The S1D13A04 implements a 16-bit Generic #1 Host Bus Interface which is most suitable for direct connection to the Motorola MPC82x microprocessor. Generic #1 supports a Chip Select and an individual Read Enable/Write Enable for each byte.

The Generic #1 Host Bus Interface is selected by the S1D13A04 on the rising edge of RESET#. After RESET# is released, the bus interface signals assume their selected configuration. For details on the S1D13A04 configuration, see Section 4.3, “S1D13A04 Hardware Configuration” on page 18.

3.1 Host Bus Interface Pin Mapping

The following table shows the functions of each Host Bus Interface signal.

Table 3-1: Host Bus Interface Pin Mapping

S1D13A04 Pin Names	Motorola MPC82x
AB[17:0]	A[14:31]
DB[15:0]	D[0:15]
WE1#	$\overline{WE0}$
CS#	$\overline{CS4}$
M/R#	A13
CLKI	SYSCLK
BS#	Connect to IO _{VDD} from the S1D13A04
RD/WR#	\overline{OE} (see note)
RD#	\overline{OE} (see note)
WE0#	$\overline{WE1}$
WAIT#	\overline{TA}
RESET#	System \overline{RESET}

Note

The Motorola MPC82x chip select module only handles 16-bit read cycles. As the S1D13A04 uses the chip select module to generate CS#, only 16-bit read cycles are possible and both the high and low byte enables can be driven by the MPC82x signal \overline{OE} .

3.2 Host Bus Interface Signals

The Host Bus Interface requires the following signals.

- CLKI is a clock input which is required by the S1D13A04 Host Bus Interface as a source for its internal bus and memory clocks. This clock is typically driven by the host CPU system clock. For this example, SYSCLK from the Motorola MPC82x is used for CLKI.
- The address inputs AB[17:0], and the data bus DB[15:0], connect directly to the MPC82x address (A[14:31]) and data bus (D[0:15]), respectively. CNF4 must be set to select big endian mode.
- Chip Select (CS#) must be driven low by $\overline{\text{CS4}}$ whenever the S1D13A04 is accessed by the Motorola MPC82x.
- M/R# (memory/register) selects between memory or register accesses. This signal is generated by the external address decode circuitry. For this example, M/R# is connected to address line A13, allowing system address A13 to select between memory or register accesses.
- WE0# connects to $\overline{\text{WE1}}$ (the low byte enable signal from the MPC82x) and must be driven low when the MPC82x is writing the low byte to the S1D13A04.
- WE1# connects to $\overline{\text{WE0}}$ (the high byte enable signal from the MPC82x) and must be driven low when the MPC82x is writing the high byte to the S1D13A04.
- RD# and RD/WR# are read enables for the low-order and high-order bytes, respectively. Both signals are driven low by $\overline{\text{OE}}$ when the Motorola MPC82x is reading data from the S1D13A04.
- WAIT# connects to $\overline{\text{TA}}$ and is a signal which is output from the S1D13A04 which indicates the MPC82x must wait until data is ready (read cycle) or accepted (write cycle) on the host bus. Since MPC82x accesses to the S1D13A04 may occur asynchronously to the display update, it is possible that contention may occur in accessing the S1D13A04 internal registers and/or display buffer. The WAIT# line resolves these contentions by forcing the host to wait until the resource arbitration is complete.
- The Bus Status (BS#) signal is not used in this implementation of the MPC82x interface using the Generic #1 Host Bus Interface. This pin must be tied high (connected to IO V_{DD}).

4 MPC82x to S1D13A04 Interface

4.1 Hardware Description

The interface between the S1D13A04 and the MPC82x requires no external glue logic. The polarity of the WAIT# signal must be selected as active high by connecting CNF5 to IO V_{DD} (see Table 4-2: “Summary of Power-On/Reset Options,” on page 18).

BS# (bus start) is not used in this implementation and should be tied high (connected to IO V_{DD}).

The following diagram shows a typical implementation of the MPC82x to S1D13A04 interface.

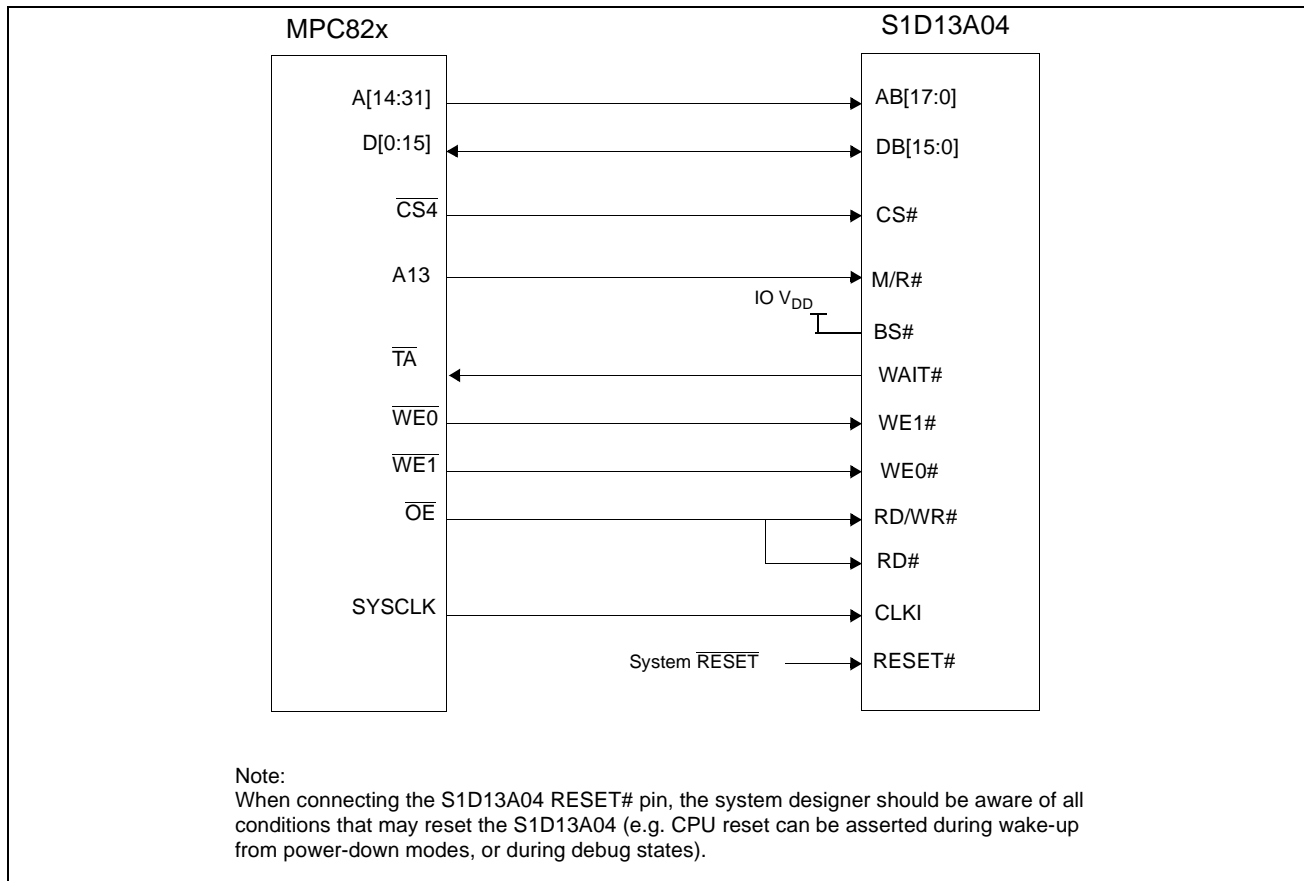


Figure 4-1: Typical Implementation of MPC82x to S1D13A04 Interface

Table 4-1: “List of Connections from MPC821ADS to S1D13A04” on page 16 shows the connections between the pins and signals of the MPC82x and the S1D13A04.

Note

The interface was designed using a Motorola MPC821 Application Development System (ADS). The ADS board has 5 volt logic connected to the data bus, so the interface included two 74F245 octal buffers on D[0:15] between the ADS and the S1D13A04. In a true 3 volt system, no buffering is necessary.

4.2 MPC821ADS Evaluation Board Hardware Connections

The following table details the connections between the pins and signals of the MPC821 and the S1D13A04.

Table 4-1: List of Connections from MPC821ADS to S1D13A04

MPC821 Signal Name	MPC821ADS Connector and Pin Name	S1D13A04 Signal Name
2.0V (see note 1)	P9-D24	COREVDD
3.3V	P9-A22	IOVDD
A14 (see note 2)	P6-C20	AB17
A15 (see note 2)	P6-D20	AB16
A16 (see note 2)	P6-B24	AB15
A17 (see note 2)	P6-C24	AB14
A18 (see note 2)	P6-D23	AB13
A19 (see note 2)	P6-D22	AB12
A20 (see note 2)	P6-D19	AB11
A21 (see note 2)	P6-A19	AB10
A22 (see note 2)	P6-D28	AB9
A23 (see note 2)	P6-A28	AB8
A24 (see note 2)	P6-C27	AB7
A25 (see note 2)	P6-A26	AB6
A26 (see note 2)	P6-C26	AB5
A27 (see note 2)	P6-A25	AB4
A28 (see note 2)	P6-D26	AB3
A29 (see note 2)	P6-B25	AB2
A30 (see note 2)	P6-B19	AB1
A31 (see note 2)	P6-D17	AB0
D0 (see note 3)	P12-A9	D15
D1 (see note 3)	P12-C9	D14
D2 (see note 3)	P12-D9	D13
D3 (see note 3)	P12-A8	D12
D4 (see note 3)	P12-B8	D11
D5 (see note 3)	P12-D8	D10
D6 (see note 3)	P12-B7	D9
D7 (see note 3)	P12-C7	D8
D8 (see note 3)	P12-A15	D7
D9 (see note 3)	P12-C15	D6

Table 4-1: List of Connections from MPC821ADS to S1D13A04 (Continued)

MPC821 Signal Name	MPC821ADS Connector and Pin Name	S1D13A04 Signal Name
D10 (see note 3)	P12-D15	D5
D11 (see note 3)	P12-A14	D4
D12 (see note 3)	P12-B14	D3
D13 (see note 3)	P12-D14	D2
D14 (see note 3)	P12-B13	D1
D15 (see note 3)	P12-C13	D0
SRESET	P9-D15	RESET#
SYSCLK	P9-C2	CLKI
CS4	P6-D13	CS#
TA	P6-B6 to inverter enabled by CS#	WAIT#
WE0	P6-B15	WE1#
WE1	P6-A14	WE0#
OE	P6-B16	RD/WR#, RD#
A13	P6-C21	M/R#
GND	P12-A1, P12-B1, P12-A2, P12-B2, P12-A3, P12-B3, P12-A4, P12-B4, P12-A5, P12-B5, P12-A6, P12-B6, P12-A7	Vss

Note

1. The PCMCIA connector (P9) provides 2.0V on D[23:25] and can be used as the source for COREVDD. However, at 2.0V the S1D13A04 MCLK is has a maximum frequency of 30MHz. To increase memory performance (MCLK maximum = 50MHz) an external 2.5V power supply must be connected to COREVDD.
2. The bit numbering of the Motorola MPC821 bus signals is reversed from the normal convention, e.g.: the most significant address bit is A0, the next is A1, A2, etc.
3. The bit numbering of the Motorola MPC821 data signals is reversed from the normal convention, e.g.: the most significant address bit is D0, the next is D1, D2, etc.

4.3 S1D13A04 Hardware Configuration

The S1D13A04 uses CNF6 through CNF0 to allow selection of the bus mode and other configuration data on the rising edge of RESET#. For details on configuration, refer to the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

The following table shows the configuration required for this implementation of a S1D13A04 to Motorola MPC82x microprocessor.

Table 4-2: Summary of Power-On/Reset Options

S1D13A04 Configuration Input	Power-On/Reset State											
	1 (connected to IO V _{DD})	0 (connected to V _{SS})										
CNF4, CNF[2:0]	Select host bus interface as follows:											
	<table border="1"> <thead> <tr> <th>CNF4</th> <th>CNF2</th> <th>CNF1</th> <th>CNF0</th> <th>Host Bus</th> </tr> </thead> <tbody> <tr> <td>1</td> <td>0</td> <td>1</td> <td>1</td> <td>Generic #1, Big Endian</td> </tr> </tbody> </table>	CNF4	CNF2	CNF1	CNF0	Host Bus	1	0	1	1	Generic #1, Big Endian	
CNF4	CNF2	CNF1	CNF0	Host Bus								
1	0	1	1	Generic #1, Big Endian								
CNF3	Reserved. Must be set to 1.											
CNF5	WAIT# is active high	WAIT# is active low										
CNF6	CLKI to BCLK divide ratio 2:1	CLKI to BCLK divide ratio 1:1										

configuration for Motorola MPC82x microprocessor

4.4 Register/Memory Mapping

The DRAM on the MPC821 ADS board extends from address 0 through 3F FFFFh, so the S1D13A04 is addressed starting at 40 0000h. The S1D13A04 uses two 256K byte blocks which are selected using A13 from the MPC821 (A13 is connected to the S1D13A04 M/R# pin). The internal registers occupy the first 256K byte block and the 160K byte display buffer occupies the second 256K byte block.

4.5 MPC82x Chip Select Configuration

Chip select 4 is used to control the S1D13A04. The following options are selected in the base address register (BR4).

- BA (0:16) = 0000 0000 0100 0000 0 – set starting address of S1D13A04 to 40 0000h
- AT (0:2) = 0 – ignore address type bits.
- PS (0:1) = 1:0 – memory port size is 16 bits
- PARE = 0 – disable parity checking
- WP = 0 – disable write protect
- MS (0:1) = 0:0 – select General Purpose Chip Select module to control this chip select
- V = 1 – set valid bit to enable chip select

The following options were selected in the option register (OR4).

- AM (0:16) = 1111 1111 1100 0000 0 – mask all but upper 10 address bits; S1D13A04 consumes 4M byte of address space
- ATM (0:2) = 0 – ignore address type bits
- CSNT = 0 – normal $\overline{\text{CS}}/\overline{\text{WE}}$ negation
- ACS (0:1) = 1:1 – delay $\overline{\text{CS}}$ assertion by $\frac{1}{2}$ clock cycle from address lines
- BI = 1 – assert Burst Inhibit
- SCY (0:3) = 0 – wait state selection; this field is ignored since external transfer acknowledge is used; see SETA below
- SETA = 1 – the S1D13A04 generates an external transfer acknowledge using the WAIT# line
- TRLX = 0 – normal timing
- EHTR = 0 – normal timing

4.6 Test Software

The test software to exercise this interface is very simple. It configures chip select 4 (CS4) on the MPC82x to map the S1D13A04 to an unused 512K byte block of address space and loads the appropriate values into the option register for CS4. Then the software runs a tight loop reading the S1D13A04 Revision Code Register REG[00h]. This allows monitoring of the bus timing on a logic analyzer.

The following source code was entered into the memory of the MPC821ADS using the line-by-line assembler in MPC8BUG (the debugger provided with the ADS board). Once the program was executed on the ADS, a logic analyzer was used to verify operation of the interface hardware.

It is important to note that when the MPC821 comes out of reset, its on-chip caches and MMU are disabled. If the data cache is enabled, then the MMU must be set up so that the S1D13A04 memory block is tagged as non-cacheable, to ensure that accesses to the S1D13A04 occurs in proper order, and also to ensure that the MPC821 does not attempt to cache any data read from or written to the S1D13A04 or its display buffer.

The source code for this test routine is as follows:

```
BR4      equ      $120          ; CS4 base register
OR4      equ      $124          ; CS4 option register
MemStart equ      $44 0000      ; address of S1D13A04 display buffer
RevCodeReg equ     $40 0000     ; address of Revision Code Register

Start    mfspr      r1,IMMR      ; get base address of internal registers
         andis.    r1,r1,$ffff   ; clear lower 16 bits to 0
         andis.    r2,r0,0       ; clear r2
         oris     r2,r2,MemStart  ; write base address
         ori      r2,r2,$0801    ; port size 16 bits; select GPCM; enable
         stw     r2,BR4(r1)      ; write value to base register
         andis.   r2,r0,0        ; clear r2
         oris     r2,r2,$ffc0    ; address mask - use upper 10 bits
         ori      r2,r2,$0708    ; normal CS negation; delay CS 1/2 clock;
                                   ; inhibit burst
         stw     r2,OR4(r1)      ; write to option register
         andis.   r1,r0,0        ; clear r1
         oris     r1,r1,MemStart  ; point r1 to start of S1D13A04 mem space
Loop     lbz     r0,RevCodeReg(r1) ; read revision code into r1
         b       Loop           ; branch forever

end
```

Note

MPC8BUG does not support comments or symbolic equates. These have been added for clarity only.

5 Software

Test utilities and display drivers are available for the S1D13A04. Full source code is available for both the test utilities and the drivers.

The test utilities are configurable for different panel types using a program called 13A04CFG (see document number X37A-B-001-xx), or by directly modifying the source. The display drivers can be customized by the OEM for different panel types, resolutions and color depths only by modifying the source.

The S1D13A04 test utilities and display drivers are available from your sales support contact (see Section 7, “*Sales and Technical Support*”) or www.erd.epson.com.

6 References

6.1 Documents

- Motorola Inc., *Power PC MPC821 Portable Systems Microprocessor User's Manual*, Motorola Publication no. MPC821UM/; available on the Internet at http://www.mot.com/SPS/ADC/pps/_subpgs/_documentation/821/821UM.html.
- Epson Research and Development, Inc., *S1D13A04 Hardware Functional Specification*, Document Number X37A-A-001-xx.
- Epson Research and Development, Inc., *S5U13A04B00C Rev. 1.0 Evaluation Board User Manual*, Document Number X37A-G-004-xx.
- Epson Research and Development, Inc., *S1D13A04 Programming Notes and Examples*, Document Number X37A-G-003-xx.

6.2 Document Sources

- Motorola Inc. Literature Distribution Center: (800) 441-2447.
- Motorola Inc. Website: www.mot.com.
- Epson Research and Development Website: www.erd.epson.com.

7 Sales and Technical Support

7.1 EPSON LCD/USB Companion Chips (S1D13A04)

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7.2 Motorola MPC821 Processor

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S1D13A04 LCD/USB Companion Chip

Interfacing to the Motorola MCF5307 "ColdFire" Microprocessor

Document Number: X37A-G-010-01

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1 Introduction

This application note describes the hardware and software environment required to interface the S1D13A04 LCD/USB Companion Chip and the Motorola MCF5307 “Coldfire” Processor.

The designs described in this document are presented only as examples of how such interfaces might be implemented. This application note is updated as appropriate. Please check the Epson Research and Development website at www.erd.epson.com for the latest revision of this document before beginning any development.

We appreciate your comments on our documentation. Please contact us via email at documentation@erd.epson.com.

2 Interfacing to the MCF5307

2.1 The MCF5307 System Bus

The MCF5200/5300 family of processors feature a high-speed synchronous system bus typical of modern microprocessors. This section is an overview of the operation of the CPU bus in order to establish interface requirements.

2.1.1 Overview

The MCF5307 microprocessor family uses a synchronous address and data bus, very similar in architecture to the MC68040 and MPC8xx. All outputs and inputs are timed with respect to a square-wave reference clock called BCLK0 (Master Clock). This clock runs at a software-selectable divisor rate from the machine cycle speed of the CPU core (typically 20 to 33 MHz). Both the address and the data bus are 32 bits in width. All IO accesses are memory-mapped; there is no separate IO space in the Coldfire architecture.

The bus can support two types of cycles, normal and burst. Burst memory cycles are used to fill on-chip cache memories, and for certain on-chip DMA operations. Normal cycles are used for all other data transfers.

2.1.2 Normal (Non-Burst) Bus Transactions

A data transfer is initiated by the bus master by placing the memory address on address lines A31 through A0 and driving \overline{TS} (Transfer Start) low for one clock cycle. Several control signals are also provided with the memory address:

- $SIZ[1:0]$ (Transfer Size) — indicates whether the bus cycle is 8, 16, or 32-bit.
- R/\overline{W} — set high for read cycles and low for write cycles.
- $TT[1:0]$ (Transfer Type Signals) — provides more detail on the type of transfer being attempted.
- \overline{TIP} (Transfer In Progress) — asserts whenever a bus cycle is active.

When the peripheral device being accessed has completed the bus transfer, it asserts \overline{TA} (Transfer Acknowledge) for one clock cycle to complete the bus transaction. Once \overline{TA} has been asserted, the MCF5307 will not start another bus cycle until \overline{TA} has been de-asserted. The minimum length of a bus transaction is two bus clocks.

Figure 2-1: “MCF5307 Memory Read Cycle,” illustrates a typical memory read cycle on the MCF5307 system bus.

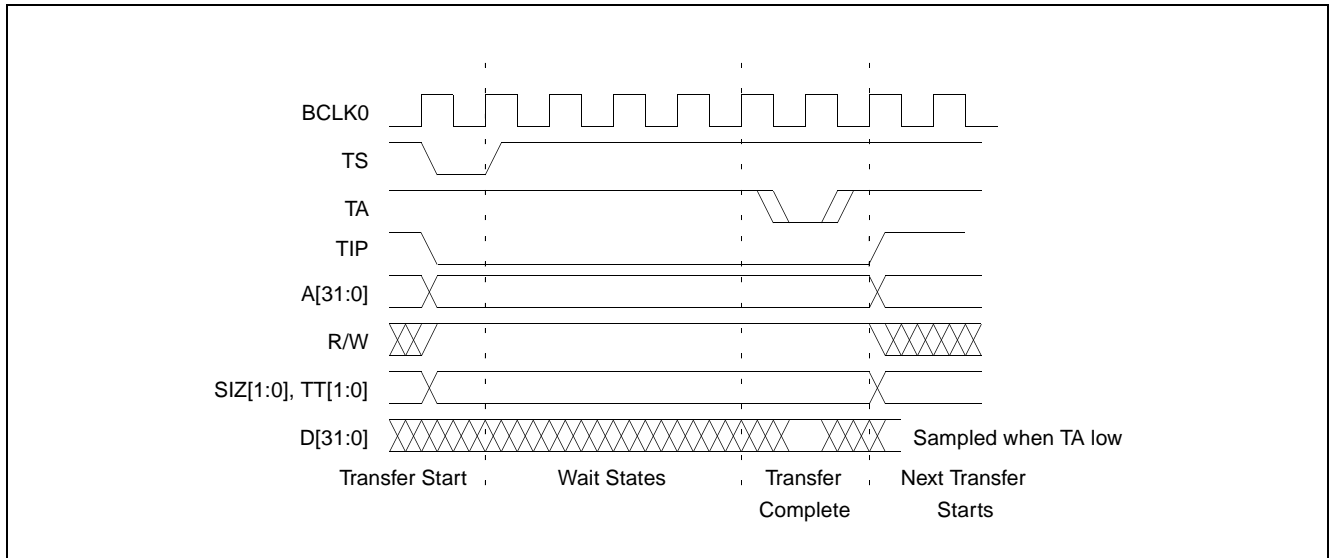


Figure 2-1: MCF5307 Memory Read Cycle

Figure 2-2: “MCF5307 Memory Write Cycle,” illustrates a typical memory write cycle on the MCF5307 system bus.

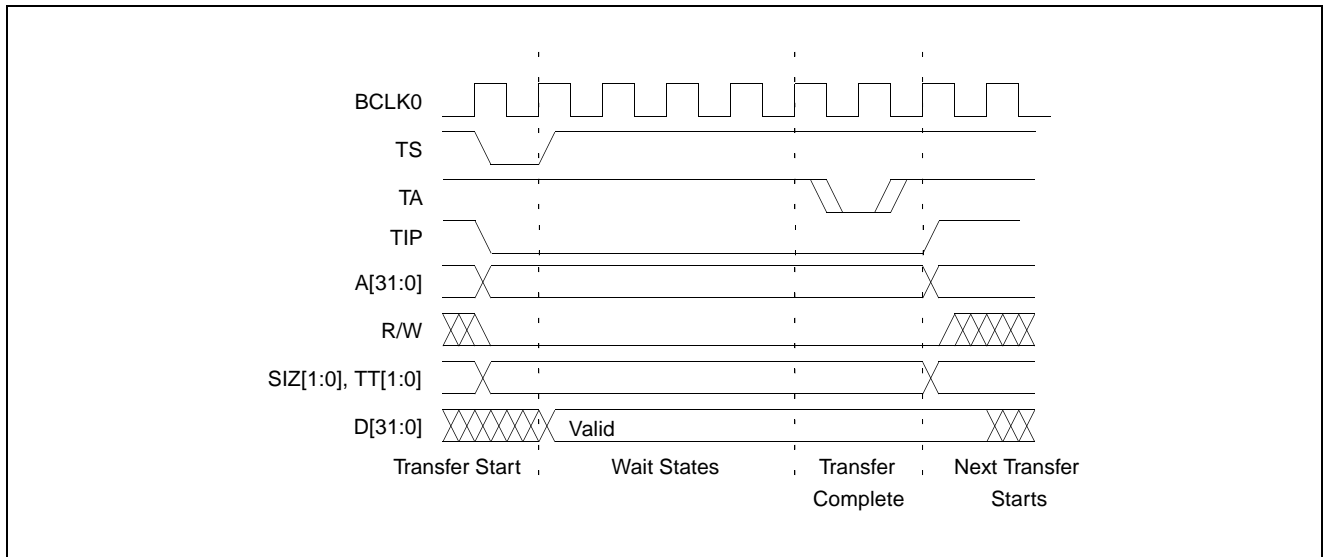


Figure 2-2: MCF5307 Memory Write Cycle

2.1.3 Burst Cycles

Burst cycles are very similar to normal cycles, except that they occur as a series of four back-to-back, 32-bit memory reads or writes. The \overline{TIP} (Transfer In Progress) output is asserted continuously through the burst. Burst memory cycles are mainly intended to fill

caches from program or data memory. They are typically not used for transfers to or from IO peripheral devices such as the S1D13A04. The MCF5307 chip selects provide a mechanism to disable burst accesses for peripheral devices which are not burst capable.

2.2 Chip-Select Module

In addition to generating eight independent chip-select outputs, the MCF5307 Chip Select Module can generate active-low Output Enable (\overline{OE}) and Write Enable (\overline{BWE}) signals compatible with most memory and x86-style peripherals. The MCF5307 bus controller also provides a Read/Write (R/\overline{W}) signal which is compatible with most 68K peripherals.

Chip selects 0 and 1 can be programmed independently to respond to any base address and block size. Chip select 0 can be active immediately after reset, and is typically used to control a boot ROM. Chip select 1 is likewise typically used to control a large static or dynamic RAM block.

Chip selects 2 through 7 have fixed block sizes of 2M bytes each. Each has a unique, fixed offset from a common, programmable starting address. These chip selects are well-suited to typical IO addressing requirements.

Each chip select may be individually programmed for:

- port size (8/16/32-bit).
- up to 15 wait states or external acknowledge.
- address space type.
- burst or non-burst cycle support.
- write protect.

Figure 2-3: “Chip Select Module Outputs Timing” illustrates a typical cycle for a memory mapped device using the GPCM of the Power PC.

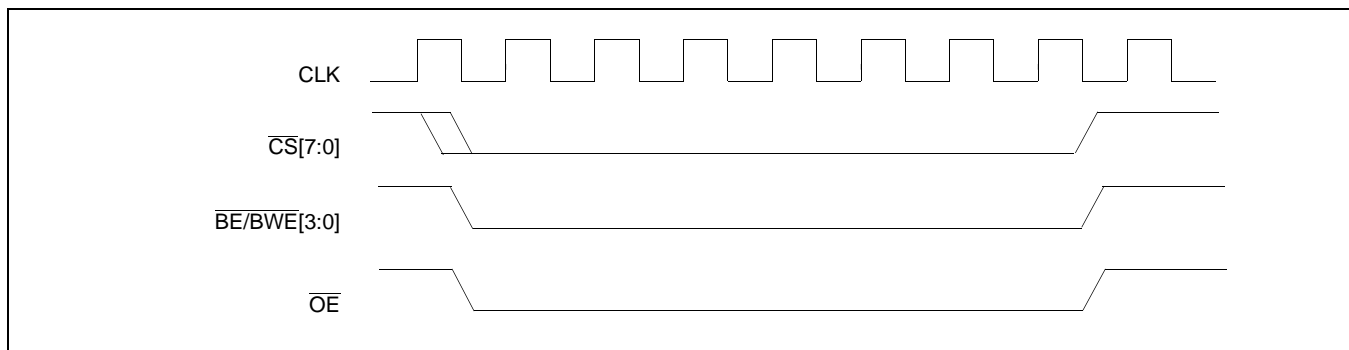


Figure 2-3: Chip Select Module Outputs Timing

3 S1D13A04 Host Bus Interface

The S1D13A04 directly supports multiple processors. The S1D13A04 implements a 16-bit Generic #1 Host Bus Interface which is most suitable for direct connection to the Motorola MFC5307 microprocessor. Generic #1 supports a Chip Select and an individual Read Enable/Write Enable for each byte.

The Generic #1 Host Bus Interface is selected by the S1D13A04 on the rising edge of RESET#. After RESET# is released, the bus interface signals assume their selected configuration. For details on the S1D13A04 configuration, see Section 4.2, "S1D13A04 Hardware Configuration" on page 14.

3.1 Host Bus Interface Pin Mapping

The following table shows the functions of each Host Bus Interface signal.

Table 3-1: Host Bus Interface Pin Mapping

S1D13A04 Pin Names	Motorola MCF5307
AB[17:0]	A[17:0]
DB[15:0]	D[31:16]
WE1#	$\overline{\text{BWE1}}$
CS#	$\overline{\text{CS4}}$
M/R#	A18
CLKI	BCLK0
BS#	Connect to IO _{VDD} from the S1D13A04
RD/WR#	$\overline{\text{OE}}$
RD#	$\overline{\text{OE}}$
WE0#	$\overline{\text{BWE0}}$
WAIT#	$\overline{\text{TA}}$
RESET#	system $\overline{\text{RESET}}$

3.2 Host Bus Interface Signals

The Host Bus Interface requires the following signals.

- CLKI is a clock input which is required by the S1D13A04 Host Bus Interface as a source for its internal bus and memory clocks. This clock is typically driven by the host CPU system clock. For this example, BCLK0 from the Motorola MCF5307 is used for CLKI.
- The address inputs AB[17:0] connect directly to the MCF5307 address bus (A[17:0]).
- DB[7:0] connects D[23:16] (the MCF5307 low order byte). DB[15:8] connects to D[31:24] (the MCF5307 high order byte). CNF4 must be set to select big endian mode.
- Chip Select (CS#) must be driven low by $\overline{\text{CS4}}$ whenever the S1D13A04 is accessed by the Motorola MCF5307.
- M/R# (memory/register) selects between memory or register accesses. This signal is generated by the external address decode circuitry. For this example, M/R# is connected to address line A18, allowing system address A18 to select between memory or register accesses.
- WE0# connects to $\overline{\text{BWE0}}$ (the low byte enable signal from the MCF5307) and must be driven low when the MCF5307 is writing the low byte to the S1D13A04.
- WE1# connects to $\overline{\text{BWE1}}$ (the high byte enable signal from the MCF5307) and must be driven low when the MCF5307 is writing the high byte to the S1D13A04.
- RD# and RD/WR# are read enables for the low-order and high-order bytes, respectively. Both signals are driven low by $\overline{\text{OE}}$ when the Motorola MCF5307 is reading data from the S1D13A04.
- WAIT# connects to $\overline{\text{TA}}$ and is a signal which is output from the S1D13A04 that indicates the host CPU must wait until data is ready (read cycle) or accepted (write cycle) on the host bus. Since host CPU accesses to the S1D13A04 may occur asynchronously to the display update, it is possible that contention may occur in accessing the S1D13A04 internal registers and/or refresh memory. The WAIT# line resolves these contentions by forcing the host to wait until the resource arbitration is complete. This signal is active low and may need to be inverted if the host CPU wait state signal is active high.
- The Bus Status (BS#) signal is not used in the bus interface for Generic #1 mode and must be tied high to IO V_{DD} .

4 MCF5307 To S1D13A04 Interface

4.1 Hardware Description

The interface between the S1D13A04 and the MCF5307 requires no external glue logic. The polarity of the WAIT# signal must be selected as active high by connecting CNF5 to IO V_{DD} (see Table 4-1: "Summary of Power-On/Reset Options," on page 14).

The following diagram shows a typical implementation of the MCF5307 to S1D13A04 interface.

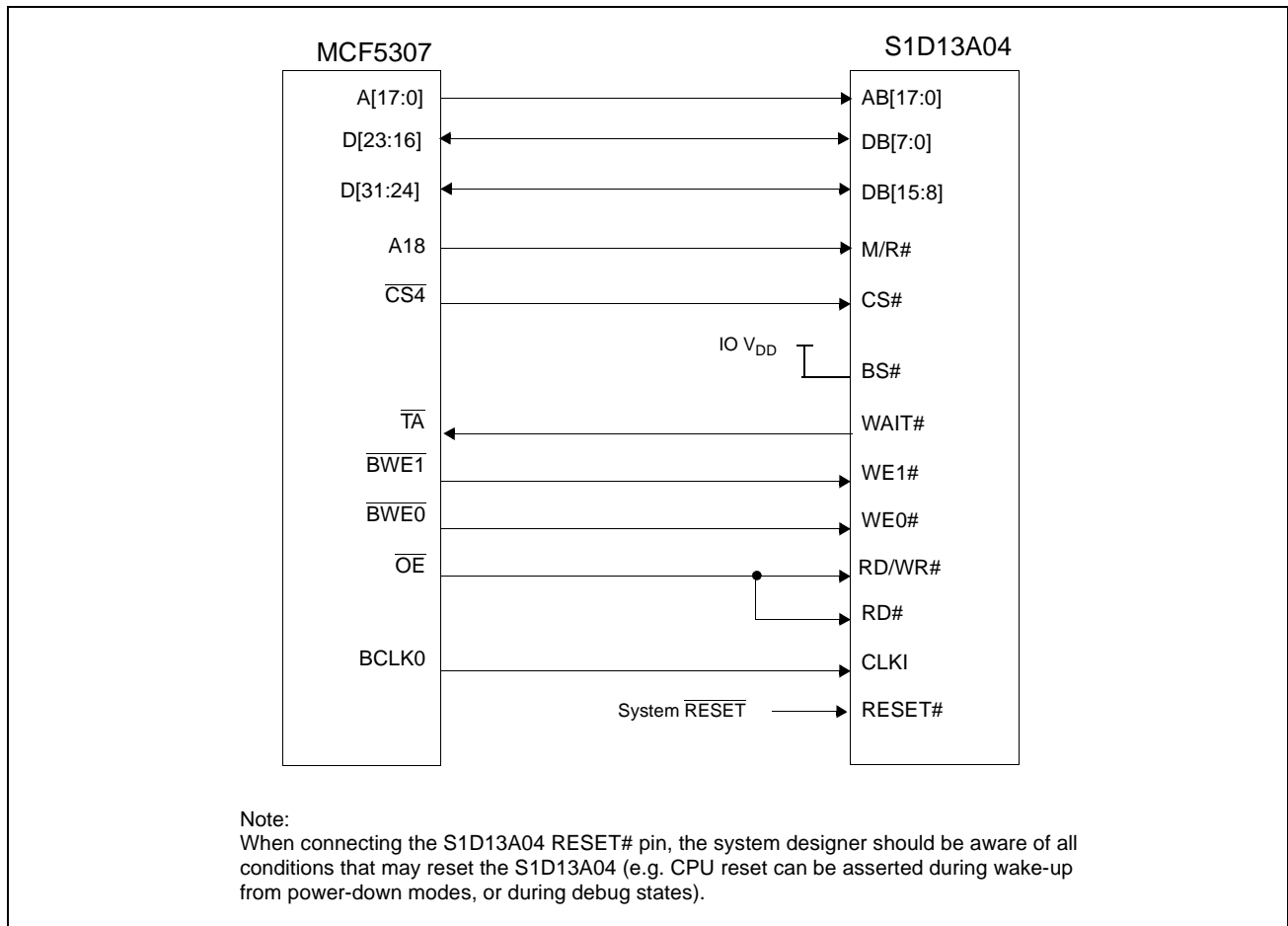


Figure 4-1: Typical Implementation of MCF5307 to S1D13A04 Interface

4.2 S1D13A04 Hardware Configuration

The S1D13A04 uses CNF6 through CNF0 to allow selection of the bus mode and other configuration data on the rising edge of RESET#. For details on configuration, refer to the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

The following table shows the configuration required for this implementation of a S1D13A04 to Motorola MFC5307 microprocessor.

Table 4-1: Summary of Power-On/Reset Options

S1D13A04 Configuration Input	Power-On/Reset State											
	1 (connected to IO V _{DD})	0 (connected to V _{SS})										
CNF4, CNF[2:0]	Select host bus interface as follows:											
	<table border="1"> <thead> <tr> <th>CNF4</th> <th>CNF2</th> <th>CNF1</th> <th>CNF0</th> <th>Host Bus</th> </tr> </thead> <tbody> <tr> <td>1</td> <td>0</td> <td>1</td> <td>1</td> <td>Generic #1, Big Endian</td> </tr> </tbody> </table>	CNF4	CNF2	CNF1	CNF0	Host Bus	1	0	1	1	Generic #1, Big Endian	
CNF4	CNF2	CNF1	CNF0	Host Bus								
1	0	1	1	Generic #1, Big Endian								
CNF3	Reserved. Must be set to 1.											
CNF5	WAIT# is active high	WAIT# is active low										
CNF6	CLKI to BCLK divide ratio 2:1	CLKI to BCLK divide ratio 1:1										
	configuration for Motorola MFC5307 microprocessor											

4.3 Register/Memory Mapping

The S1D13A04 uses two 256K byte blocks which are selected using A18 from the MCF5307 (A18 is connected to the S1D13A04 M/R# pin). The internal registers occupy the first 256K bytes block and the 160K byte display buffer occupies the second 256K byte block. These two blocks of memory are aliased over the entire 2M byte space.

Note

If aliasing is not desirable, the upper addresses must be fully decoded.

4.4 MCF5307 Chip Select Configuration

Chip Selects 0 and 1 have programmable block sizes from 64K bytes through 2G bytes. However, these chip selects would normally be needed to control system RAM and ROM. Therefore, one of the IO chip selects CS2 through CS7 is required to address the entire address space of the S1D13A04. These IO chip selects have a fixed, 2M byte block size. In the example interface, chip select 4 is used to control the S1D13A04. The CSBAR register should be set to the upper 8 bits of the desired base address.

The following options should be selected in the chip select mask registers (CSMR4/5).

- WP = 0 – disable write protect
- AM = 0 - enable alternate bus master access to the S1D13A04
- C/I = 1 - disable CPU space access to the S1D13A04
- SC = 1 - disable Supervisor Code space access to the S1D13A04
- SD = 0 - enable Supervisor Data space access to the S1D13A04
- UC = 1 - disable User Code space access to the S1D13A04
- UD = 0 - enable User Data space access to the S1D13A04
- V = 1 - global enable (“Valid”) for the chip select

The following options should be selected in the chip select control registers (CSCR4/5).

- WS0-3 = 0 - no internal wait state setting
- AA = 0 - no automatic acknowledgment
- PS (1:0) = 1:0 – memory port size is 16 bits
- BEM = 0 – Byte enable/write enable active on writes only
- BSTR = 0 – disable burst reads
- BSTW = 0 – disable burst writes

5 Software

Test utilities and display drivers are available for the S1D13A04. Full source code is available for both the test utilities and the drivers.

The test utilities are configurable for different panel types using a program called 13A04CFG (see document number X37A-B-001-xx), or by directly modifying the source. The display drivers can be customized by the OEM for different panel types, resolutions and color depths only by modifying the source.

The S1D13A04 test utilities and display drivers are available from your sales support contact (see Section 7, "*Sales and Technical Support*") or www.erd.epson.com.

6 References

6.1 Documents

- Motorola Inc., *MCF5307 ColdFire® Integrated Microprocessor User's Manual*, Motorola Publication no. MCF5307UM; available on the Internet at <http://www.mot.com/SPS/HPESD/prod/coldfire/5307UM.html>.
- Epson Research and Development, Inc., *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.
- Epson Research and Development, Inc., *S5U13A04B00C Rev. 1.0 Evaluation Board User Manual*, document number X37A-G-004-xx.
- Epson Research and Development, Inc., *S1D13A04 Programming Notes and Examples*, document number X37A-G-003-xx.

6.2 Document Sources

- Motorola Inc.: Motorola Literature Distribution Center, (800) 441-2447.
- Motorola Inc. Website: www.mot.com.
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S1D13A04 LCD/USB Companion Chip

Connecting to the Sharp HR-TFT Panels

Document Number: X37A-G-011-01

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Figure 2-4: HR-TFT Power-On/Off Sequence Timing	11
Figure 3-1: Sharp LQ031B1DDxx Gray Scale Voltage (V0-V9) Generation	14

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1 Introduction

This application note describes the hardware and software environment required to connect to the Sharp HR-TFT panels directly supported by the S1D13A04. These panels are:

- Sharp LQ031B1DDXX 160 x 160 HR-TFT panel.
- Sharp LQ039Q2DS01 320 x 240 HR-TFT panel.

The designs described in this document are presented only as examples of how such interfaces might be implemented. This application note is updated as appropriate. Please check the Epson Research and Development website at www.erd.epson.com for the latest revision of this document before beginning any development.

We appreciate your comments on our documentation. Please contact us via email at documentation@erd.epson.com.

2 Connecting to the Sharp LQ039Q2DS01 HR-TFT

2.1 External Power Supplies

The S1D13A04 provides all necessary data and control signals to connect to the Sharp LQ039Q2DS01 320 x 240 HR-TFT panel. However, it does not provide any of the voltages required for gray scaling, gate driving, or for the digital and analog supplies. Therefore, external supplies must be designed for any device utilizing the LQ039Q2DS01.

2.1.1 Gray Scale Voltages for Gamma Correction

The standard gray scale voltages can be generated using a precise resistor divider network that supplies two sets (A and B) of nine reference voltages to a National Semiconductor 9-Channel Buffer Amplifier (LMC6009). The LMC6009 buffers these nine reference voltages and outputs them to the panel column drivers. The A/B inputs allow the two sets of reference voltages to be alternated, compensating for asymmetrical gamma characteristics during row inversion. This input is controlled by the S1D13A04 output signal REV which toggles every time a horizontal sync signal is sent to the panel.

The REV signal is also used to generate the highest gray scale voltage (V0 or black) by buffering REV and shifting its maximum level to the maximum gray scale voltage (CON_POWER). CON_POWER is supplied by a National Semiconductor micropower Voltage Regulator (LP2951). Figure 2-1: “Sharp LQ039Q2DS01 Gray Scale Voltage (V0-V9) Generation” shows the schematic for gray scale voltage generation.

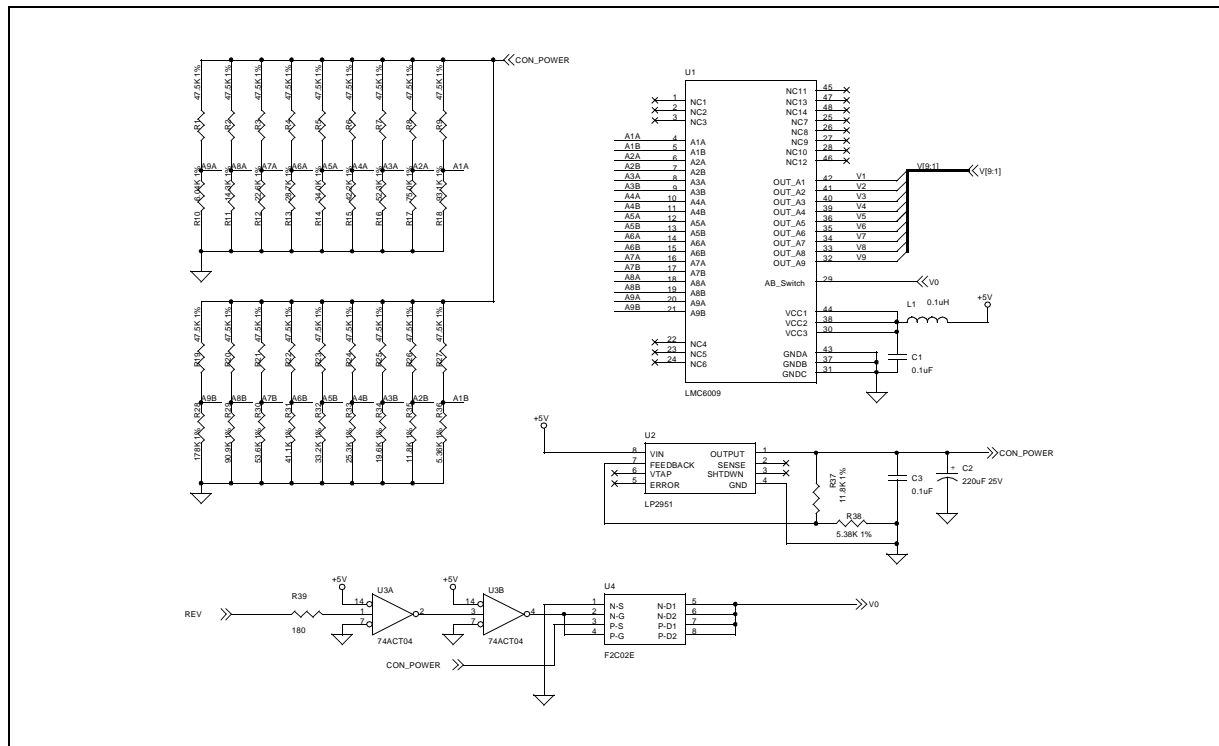


Figure 2-1: Sharp LQ039Q2DS01 Gray Scale Voltage (V0-V9) Generation

2.1.2 Digital/Analog Power Supplies

The digital power supply (VSHD) must be connected to a 3.3V supply. The analog power supply (VSHA) must be connected to a 5.0V supply.

2.1.3 DC Gate Driver Power Supplies

The gate driver high level power supply (V_{DD}) and the gate driver logic low power supply (V_{SS}) have typical values of +15V and -15V respectively. These power supplies can be provided by a Linear Technology high efficiency switching regulator (LT1172). The two power supplies can be adjusted through their allowable ranges using the potentiometer VR1.

The gate driver logic high power supply (V_{CC}) is defined as $V_{SS} + V_{SHD}$. The typical V_{CC} voltage of -11.7V can be supplied from V_{SS} using a 3.3V zener diode which provides the necessary voltage change.

Figure 2-2: "Panel Gate Driver DC Power Supplies" shows the schematic for V_{SS} , V_{DD} and V_{CC} .

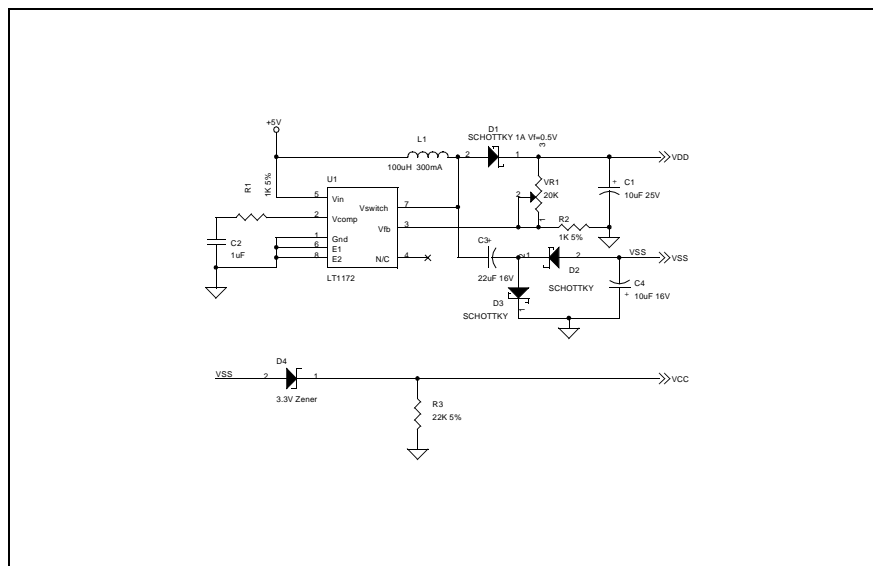


Figure 2-2: Panel Gate Driver DC Power Supplies

2.1.4 AC Gate Driver Power Supplies

The gate drive low level power supply (V_{EE}) is an AC power supply with a DC offset voltage (offset typically $-9.0V$). The AC component is the common electrode driving signal (V_{COM}) which has a voltage of $\pm 2.5V$. V_{COM} must be alternated every horizontal period and every vertical period. The S1D13A04 output signal REV accomplishes this function and generates the alternating V_{COM} signal which is superimposed onto V_{EE} . Figure 2-3: "Panel Gate Driver AC Power Supplies," on page 10 shows the schematic for generating V_{COM} and V_{EE} .

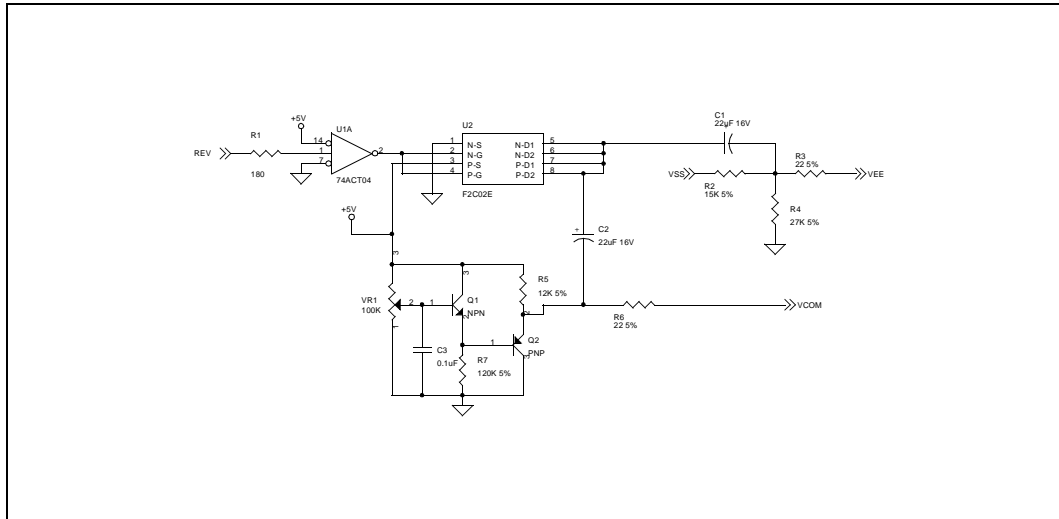


Figure 2-3: Panel Gate Driver AC Power Supplies

2.2 HR-TFT MOD Signal

The HR-TFT panel uses an input signal (MOD) to control the power-on sequencing of the panel. This HR-TFT signal should not be confused with the S1D13A04 signal DRDY (referred to as MOD for passive panels).

To power-on the HR-TFT panel, MOD must be held low until the power supply has been turned on for more than two FRAMES. To power-off the HR-TFT panel, MOD must be forced low before the power supply is turned off. This sequencing requires two software controlled GPIO pins from the S1D13A04 (see Figure 2-4: “HR-TFT Power-On/Off Sequence Timing”).

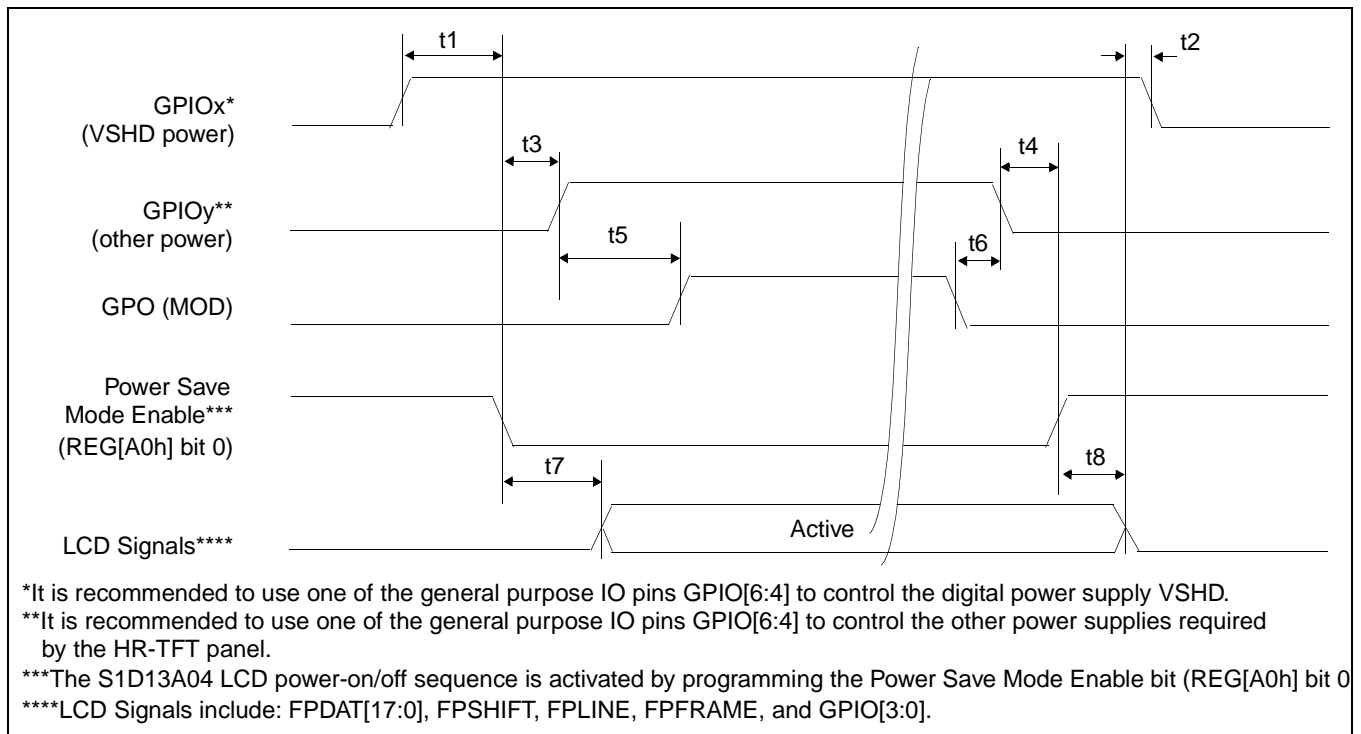


Figure 2-4: HR-TFT Power-On/Off Sequence Timing

Table 2-1: HR-TFT Power-On/Off Sequence Timing

Symbol	Parameter	Min	Max	Units
t1	LCD Power (VSHD) active to Power Save Mode disabled	0		ns
t2	LCD signals low to LCD Power (VSHD) inactive	0		ns
t3	Power Save Mode disabled to LCD Power (other) active	0		ns
t4	LCD Power (other) inactive to Power Save Mode enabled	0		ns
t5	LCD Power (other) active to MOD active	2		FRAME
t6	MOD inactive to LCD Power (other) inactive	0		ns
t7	Power Save Mode disabled to LCD signals active		20	ns
t8	Power Save Mode enabled to LCD signals low		20	ns

2.3 S1D13A04 to LQ039Q2DS01 Pin Mapping

Table 2-2: S1D13A04 to LQ039Q2DS01 Pin Mapping

LCD Pin No.	LCD Pin Name	S1D13A04 Pin Name	Description	Remarks
1	VDD	-	Power supply of gate driver (high level)	See Section 2.1, "External Power Supplies" on page 8
2	VCC	-	Power supply of gate driver (logic high)	See Section 2.1, "External Power Supplies" on page 8
3	MOD	-	Control signal of gate driver	See Section 2.2, "HR-TFT MOD Signal" on page 11
4	MOD	-	Control signal of gate driver	See Section 2.2, "HR-TFT MOD Signal" on page 11
5	U/L	-	Selection for vertical scanning direction	Connect to VSHD (top / bottom scanning)
6	SPS	FPFRAME	Start signal of gate driver	
7	CLS	GPIO1	Clock signal of gate driver	
8	VSS	-	Power supply of gate driver (logic low)	See Section 2.1, "External Power Supplies" on page 8
9	VEE	-	Power supply of gate driver (low level)	See Section 2.1, "External Power Supplies" on page 8
10	VEE	-	Power supply of gate driver (low level)	See Section 2.1, "External Power Supplies" on page 8
11	VCOM	-	Common electrode driving signal	See Section 2.1, "External Power Supplies" on page 8
12	VCOM	-	Common electrode driving signal	See Section 2.1, "External Power Supplies" on page 8
13	SPL	GPIO3	Sampling start signal for left / right scanning	
14	R0	FPDAT11	Red data signal (LSB)	
15	R1	FPDAT10	Red data signal	
16	R2	FPDAT9	Red data signal	
17	R3	FPDAT2	Red data signal	
18	R4	FPDAT1	Red data signal	
19	R5	FPDAT0	Red data signal (MSB)	
20	G0	FPDAT14	Green data signal (LSB)	
21	G1	FPDAT13	Green data signal	
22	G2	FPDAT12	Green data signal	
23	G3	FPDAT5	Green data signal	
24	G4	FPDAT4	Green data signal	
25	G5	FPDAT3	Green data signal (MSB)	

Table 2-2: S1D13A04 to LQ039Q2DS01 Pin Mapping (Continued)

LCD Pin No.	LCD Pin Name	S1D13A04 Pin Name	Description	Remarks
26	B0	FPDAT17	Blue data signal (LSB)	
27	B1	FPDAT16	Blue data signal	
28	B2	FPDAT15	Blue data signal	
29	B3	FPDAT8	Blue data signal	
30	B4	FPDAT7	Blue data signal	
31	B5	FPDAT6	Blue data signal (MSB)	
32	VSHD	-	Digital power supply	See Section 2.1, "External Power Supplies" on page 8
33	DGND	V _{SS}	Digital ground	Ground pin of S1D13A04
34	PS	GPIO0	Power save signal	
35	LP	FPLINE	Data latch signal of source driver	
36	DCLK	FPSHIFT	Data sampling clock signal	
37	LBR	-	Selection for horizontal scanning direction	Connect to VSHD (left / right scanning)
38	SPR	-	Sampling start signal for right / left scanning	Right to left scanning not supported
39	VSHA	-	Analog power supply	See Section 2.1, "External Power Supplies" on page 8
40	V0	-	Standard gray scale voltage (black)	See Section 2.1, "External Power Supplies" on page 8
41	V1	-	Standard gray scale voltage	See Section 2.1, "External Power Supplies" on page 8
42	V2	-	Standard gray scale voltage	See Section 2.1, "External Power Supplies" on page 8
43	V3	-	Standard gray scale voltage	See Section 2.1, "External Power Supplies" on page 8
44	V4	-	Standard gray scale voltage	See Section 2.1, "External Power Supplies" on page 8
45	V5	-	Standard gray scale voltage	See Section 2.1, "External Power Supplies" on page 8
46	V6	-	Standard gray scale voltage	See Section 2.1, "External Power Supplies" on page 8
47	V7	-	Standard gray scale voltage	See Section 2.1, "External Power Supplies" on page 8
48	V8	-	Standard gray scale voltage	See Section 2.1, "External Power Supplies" on page 8
49	V9	-	Standard gray scale voltage (white)	See Section 2.1, "External Power Supplies" on page 8
50	AGND	V _{SS}	Analog ground	Ground pin of S1D13A04

3 Connecting to the Sharp LQ031B1DDxx HR-TFT

3.1 External Power Supplies

The S1D13A04 provides all necessary data and control signals to connect to the Sharp LQ031B1DDxx 160x160 HR-TFT panel(s). However, it does not provide any of the voltages required for the backlight, gray scaling, gate driving, or for the digital and analog supplies. Therefore, external supplies must be designed for any device utilizing the LQ031B1DDxx.

The LQ031B1DDxx (160x160) has the same voltage requirements as the LQ039Q2DS01 (320x240). All the circuits used to generate the various voltages for the LQ039Q2DS01 panel also apply to the LQ031B1DDxx panel. This section provides additional circuits for generating some of these voltages.

3.1.1 Gray Scale Voltages for Gamma Correction

The standard gray scale voltages can be generated using a precise resistor divider network as described in Section 2.1.1, “Gray Scale Voltages for Gamma Correction” on page 8. Alternately, they can be generated using a Sharp gray scale IC. The Sharp IR3E203 eliminates the large resistor network used to provide the 10 gray scale voltages and combines their function into a single IC.

The S1D13A04 output signal REV is used to alternate the gray scale voltages and connects to the SW input of the IR3E203 IC. The COM signal is used in generating the gate driver panel AC voltage, V_{COM} and is explained in Section 3.1.4, “AC Gate Driver Power Supplies” on page 15. Figure 3-1: “Sharp LQ031B1DDxx Gray Scale Voltage (V_0 - V_9) Generation” shows the circuit that generates the gray scale voltages using the Sharp IR3E203 IC.

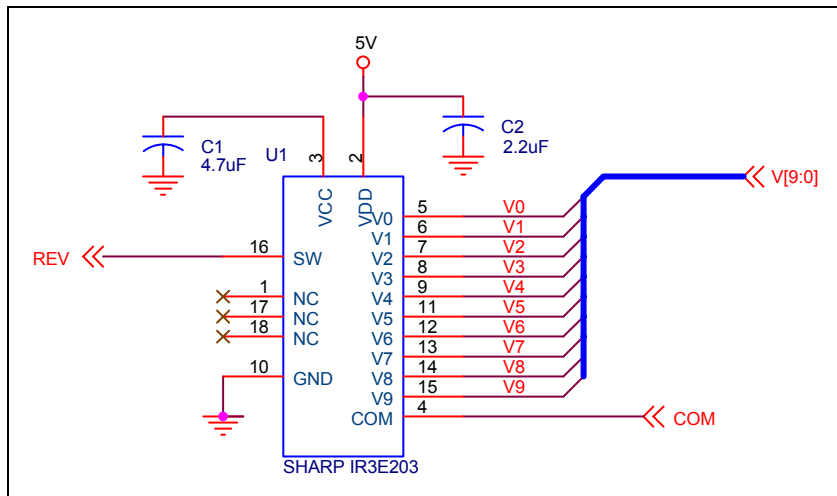


Figure 3-1: Sharp LQ031B1DDxx Gray Scale Voltage (V_0 - V_9) Generation

3.1.2 Digital/Analog Power Supplies

The digital power supply (VSHD) must be connected to a 3.3V supply. The analog power supply (VSHA) must be connected to a 5.0V supply.

3.1.3 DC Gate Driver Power Supplies

See Section 2.1.3, “DC Gate Driver Power Supplies” on page 9 and Figure 2-2: “Panel Gate Driver DC Power Supplies,” on page 9 for details on generating V_{SS} , V_{DD} , and V_{CC} .

3.1.4 AC Gate Driver Power Supplies

See Section 2.1.4, “AC Gate Driver Power Supplies” on page 10 and Figure 2-3: “Panel Gate Driver AC Power Supplies,” on page 10 for details on generating V_{EE} and V_{COM} . If the Sharp IR3E203 is used to generate the gray scale voltages, the COM signal can be connected to the input of the F2C02E MOSFET instead of the buffered REV signal.

3.2 HR-TFT MOD Signal

See Section 2.2, “HR-TFT MOD Signal” on page 11 for details on controlling the MOD signal through software.

3.3 S1D13A04 to LQ031B1DDxx Pin Mapping

Table 3-1: S1D13A04 to LQ031B1DDxx Pin Mapping

LCD Pin No.	LCD Pin Name	S1D13A04 Pin Name	Description	Remarks
1	VDD	-	Power supply of gate driver (high level)	See Section 3.1, "External Power Supplies" on page 14
2	VCC	-	Power supply of gate driver (logic high)	See Section 3.1, "External Power Supplies" on page 14
3	MOD	-	Control signal of gate driver	See Section 3.2, "HR-TFT MOD Signal" on page 15
4	MOD	-	Control signal of gate driver	See Section 3.2, "HR-TFT MOD Signal" on page 15
5	U/L	-	Selection for vertical scanning direction	Connect to VSHD (top / bottom scanning)
6	SPS	FPFRAME	Start signal of gate driver	
7	CLS	GPIO1	Clock signal of gate driver	
8	VSS	-	Power supply of gate driver (logic low)	See Section 3.1, "External Power Supplies" on page 14
9	VEE	-	Power supply of gate driver (low level)	See Section 3.1, "External Power Supplies" on page 14
10	VEE	-	Power supply of gate driver (low level)	See Section 3.1, "External Power Supplies" on page 14
11	VCOM	-	Common electrode driving signal	See Section 3.1, "External Power Supplies" on page 14
12	VCOM	-	Common electrode driving signal	See Section 3.1, "External Power Supplies" on page 14
13	SPL	GPIO3	Sampling start signal for left / right scanning	
14	R0	FPDAT11	Red data signal (LSB)	
15	R1	FPDAT10	Red data signal	
16	R2	FPDAT9	Red data signal	
17	R3	FPDAT2	Red data signal	
18	R4	FPDAT1	Red data signal	
19	R5	FPDAT0	Red data signal (MSB)	
20	G0	FPDAT14	Green data signal (LSB)	
21	G1	FPDAT13	Green data signal	
22	G2	FPDAT12	Green data signal	
23	G3	FPDAT5	Green data signal	
24	G4	FPDAT4	Green data signal	
25	G5	FPDAT3	Green data signal (MSB)	

Table 3-1: S1D13A04 to LQ031B1DDxx Pin Mapping (Continued)

LCD Pin No.	LCD Pin Name	S1D13A04 Pin Name	Description	Remarks
26	B0	FPDAT17	Blue data signal (LSB)	
27	B1	FPDAT16	Blue data signal	
28	B2	FPDAT15	Blue data signal	
29	B3	FPDAT8	Blue data signal	
30	B4	FPDAT7	Blue data signal	
31	B5	FPDAT6	Blue data signal (MSB)	
32	VSHD	-	Digital power supply	See Section 3.1, "External Power Supplies" on page 14
33	DGND	V _{SS}	Digital ground	Ground pin of S1D13A04
34	PS	GPIO0	Power save signal	
35	LP	FPLINE	Data latch signal of source driver	
36	DCLK	FPSHIFT	Data sampling clock signal	
37	LBR	-	Selection for horizontal scanning direction	Connect to VSHD (left / right scanning)
38	SPR	-	Sampling start signal for right / left scanning	Right to left scanning not supported
39	VSHA	-	Analog power supply	See Section 3.1, "External Power Supplies" on page 14
40	V0	-	Standard gray scale voltage (black)	See Section 3.1, "External Power Supplies" on page 14
41	V1	-	Standard gray scale voltage	See Section 3.1, "External Power Supplies" on page 14
42	V2	-	Standard gray scale voltage	See Section 3.1, "External Power Supplies" on page 14
43	V3	-	Standard gray scale voltage	See Section 3.1, "External Power Supplies" on page 14
44	V4	-	Standard gray scale voltage	See Section 3.1, "External Power Supplies" on page 14
45	V5	-	Standard gray scale voltage	See Section 3.1, "External Power Supplies" on page 14
46	V6	-	Standard gray scale voltage	See Section 3.1, "External Power Supplies" on page 14
47	V7	-	Standard gray scale voltage	See Section 3.1, "External Power Supplies" on page 14
48	V8	-	Standard gray scale voltage	See Section 3.1, "External Power Supplies" on page 14
49	V9	-	Standard gray scale voltage (white)	See Section 3.1, "External Power Supplies" on page 14
50	AGND	V _{SS}	Analog ground	Ground pin of S1D13A04

4 Test Software

Test utilities and display drivers are available for the S1D13A04. Full source code is available for both the test utilities and the drivers.

The test utilities are configurable for different panel types using a program called 13A04CFG (see document number X37A-B-001-xx), or by directly modifying the source. The display drivers can be customized by the OEM for different panel types, resolutions and color depths only by modifying the source.

The S1D13A04 test utilities and display drivers are available from your sales support contact (see Section 6, “*Sales & Technical Support*”) or at www.erd.epson.com.

5 References

5.1 Documents

- Sharp Electronics Corporation, *LQ039Q2DS01 Specification*.
- Sharp Electronics Corporation, *LQ031B1DDxx Specification*.
- Epson Research and Development, Inc., *S1D13A04 Hardware Functional Specification*, Document Number X37A-A-001-xx.
- Epson Research and Development, Inc., *S1D13A04 Programming Notes and Examples*, Document Number X37A-G-003-xx.
- Epson Research and Development, Inc., *S5U13A04B00C Rev. 1.0 ISA Bus Evaluation Board User Manual*, Document Number X37A-G-004-xx.

5.2 Document Sources

- Sharp Electronics Corporation Website: www.sharpmeg.com.
- Epson Research and Development Website: www.erd.epson.com.

6 Sales & Technical Support

6.1 EPSON LCD/USB Companion Chips (S1D13A04)

Japan

Seiko Epson Corporation
Electronic Devices Marketing Division
421-8, Hino, Hino-shi
Tokyo 191-8501, Japan
Tel: 042-587-5812
Fax: 042-587-5564
<http://www.epson.co.jp>

North America

Epson Electronics America, Inc.
150 River Oaks Parkway
San Jose, CA 95134, USA
Tel: (408) 922-0200
Fax: (408) 922-0238
<http://www.eea.epson.com>

Taiwan

Epson Taiwan Technology
& Trading Ltd.
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Tel: 02-2717-7360
Fax: 02-2712-9164
<http://www.epson.com.tw/>

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Epson Hong Kong Ltd.
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Tel: 2585-4600
Fax: 2827-4346
<http://www.epson.com.hk/>

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Epson Europe Electronics GmbH
Riesstrasse 15
80992 Munich, Germany
Tel: 089-14005-0
Fax: 089-14005-110
<http://www.epson-electronics.de>

Singapore

Epson Singapore Pte., Ltd.
No. 1
Temasek Avenue #36-00
Millenia Tower
Singapore, 039192
Tel: 337-7911
Fax: 334-2716
<http://www.epson.com.sg/>

6.2 Sharp HR-TFT Panel

<http://www.sharpmeg.com>

EPSON®



S1D13A04 LCD/USB Companion Chip

Interfacing to the Motorola MC68VZ328 Dragonball Microprocessor

Document Number: X37A-G-012-01

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1 Introduction

This application note describes the hardware and software environment required to interface the S1D13A04 LCD/USB Companion Chip and the Motorola MC68VZ328 Dragonball VZ microprocessor.

The designs described in this document are presented only as examples of how such interfaces might be implemented. This application note is updated as appropriate. Please check the Epson Research and Development website at www.erd.epson.com for the latest revision of this document before beginning any development.

We appreciate your comments on our documentation. Please contact us via email at documentation@erd.epson.com.

2 Interfacing to the MC68VZ328

2.1 The MC68VZ328 System Bus

The Motorola MC68VZ328 “Dragonball VZ” is the third generation in the Dragonball microprocessor family. The Dragonball VZ is an integrated controller designed for handheld products. It is based upon the FLX68000 microprocessor core and uses a 24-bit address bus and 16-bit data bus. The Dragonball VZ is faster than its predecessors and the DRAM controller now supports SDRAM. The bus interface consists of all the standard MC68000 bus interface signals except \overline{AS} , plus some new signals intended to simplify the interface to typical memory and peripheral devices. The 68000 signals are multiplexed with IrDA, SPI and LCD controller signals.

The MC68000 bus control signals are well documented in the Motorola user manuals, and are not be described here. The new signals are as follows.

- Output Enable (\overline{OE}) is asserted when a read cycle is in progress. It is intended to connect to the output enable control signal of a typical static RAM, EPROM, or Flash EPROM device.
- Upper Write Enable and Lower Write Enable (\overline{UWE} / \overline{LWE}) are asserted during memory write cycles for the upper and lower bytes of the 16-bit data bus. They may be directly connected to the write enable inputs of a typical memory device.

2.2 Chip-Select Module

The MC68VZ328 can generate up to 8 chip select outputs which are organized into four groups (A through D).

Each chip select group has a common base address register and address mask register allowing the base address and block size of the entire group to be set. In addition, each chip select within a group has its own address compare and address mask register to activate the chip select for a subset of the group’s address block. Each chip select may also be individually programmed to control an 8 or 16-bit device. Lastly, each chip select can either generate from 0 through 6 wait states internally, or allow the memory or peripheral device to terminate the cycle externally using the standard MC68000 \overline{DTACK} signal.

Chip select groups A and B are used to control ROM, SRAM, and Flash memory devices and have a block size of 128K bytes to 16M bytes. Chip select A0 is active immediately after reset and is a global chip select so it is typically used to control a boot EPROM device. A0 ceases to decode globally once its chip select registers are programmed. Groups C and D are special in that they can also control DRAM interfaces. These last two groups have block size of 32K bytes to 4M bytes.

3 S1D13A04 Host Bus Interface

The S1D13A04 directly supports multiple processors. The S1D13A04 implements a Dragonball Host Bus Interface which directly supports the Motorola MC68VZ328 microprocessor.

The Dragonball Host Bus Interface is selected by the S1D13A04 on the rising edge of RESET#. After RESET# is released, the bus interface signals assume their selected configuration. For details on the S1D13A04 configuration, see Section 4.2, “S1D13A04 Hardware Configuration” on page 12.

3.1 Host Bus Interface Pin Mapping

The following table shows the functions of each Host Bus Interface signal.

Table 3-1: Host Bus Interface Pin Mapping

S1D13A04 Pin Names	Motorola MC68VZ328
AB[17:0]	A[17:0]
DB[15:0]	D[15:0]
WE1#	\overline{UWE}
CS#	\overline{CSx}
M/R#	External Decode
CLKI	CLKO
BS#	Connect to IO _{VDD} from the S1D13A04
RD/WR#	Connect to IO _{VDD} from the S1D13A04
RD#	\overline{OE}
WE0#	\overline{LWE}
WAIT#	\overline{DTACK}
RESET#	System \overline{RESET}

3.2 Host Bus Interface Signals

The Host Bus Interface requires the following signals.

- CLKI is a clock input required by the S1D13A04 Host Bus Interface as a source for its internal bus and memory clocks. This clock is typically driven by the host CPU system clock. For this example, CLK0 from the Motorola MC68VZ328 is used for CLKI.
- The address inputs AB[17:0], and the data bus DB[15:0], connect directly to the MC68VZ328 address (A[17:0]) and data bus (D[15:0]), respectively. CNF4 must be set to one to select big endian mode.
- Chip Select (CS#) must be driven low by one of the Dragonball VZ chip select outputs from the chip select module whenever the S1D13A04 is accessed by the MC68VZ328.
- M/R# (memory/register) selects between memory or register accesses. This signal is generated by the external address decode circuitry. For this example, M/R# is connected to address line A18, allowing system address A18 to select between memory or register accesses.
- WE0# connects to $\overline{\text{LWE}}$ (the low data byte write strobe enable of the MC68VZ328) and is asserted when valid data is written to the low byte of a 16-bit device.
- WE1# connects to $\overline{\text{UWE}}$ (the upper data byte write strobe enable of the MC68VZ328) and is asserted when valid data is written to the high byte of a 16-bit device.
- RD# connects to $\overline{\text{OE}}$ (the read output enable of the MC68VZ328) and is asserted during a read cycle of the MC68VZ328 microprocessor.
- RD/WR# is not used for the Dragonball host bus interface and must be tied high to IO V_{DD} .
- WAIT# connects to $\overline{\text{DTACK}}$ and is a signal which is output from the S1D13A04 indicating the MC68VZ328 must wait until data is ready (read cycle) or accepted (write cycle) on the host bus. MC68VZ328 accesses to the S1D13A04 may occur asynchronously to the display update.
- BS# is not used for the Dragonball host bus interface and must be tied high to IO V_{DD} .

4 MC68VZ328 to S1D13A04 Interface

4.1 Hardware Description

The interface between the S1D13A04 and the MC68VZ328 does not require any external glue logic. Chip select module B is used to provide the S1D13A04 with a chip select and A18 is used to select between memory and register accesses.

In this example, the \overline{DTACK} signal is made available for the S1D13A04. Alternately, the S1D13A04 can guarantee a maximum cycle length that the Dragonball VZ handles by inserting software wait states (see Section 4.2.2, “MC68VZ328 Chip Select and Pin Configuration” on page 13). A single resistor is used to speed up the rise time of the WAIT# (\overline{DTACK}) signal when terminating the bus cycle.

The following diagram shows a typical implementation of the MC68VZ328 to S1D13A04 using the Dragonball host bus interface. For further information on the Dragonball Host Bus interface and AC Timing, refer to the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

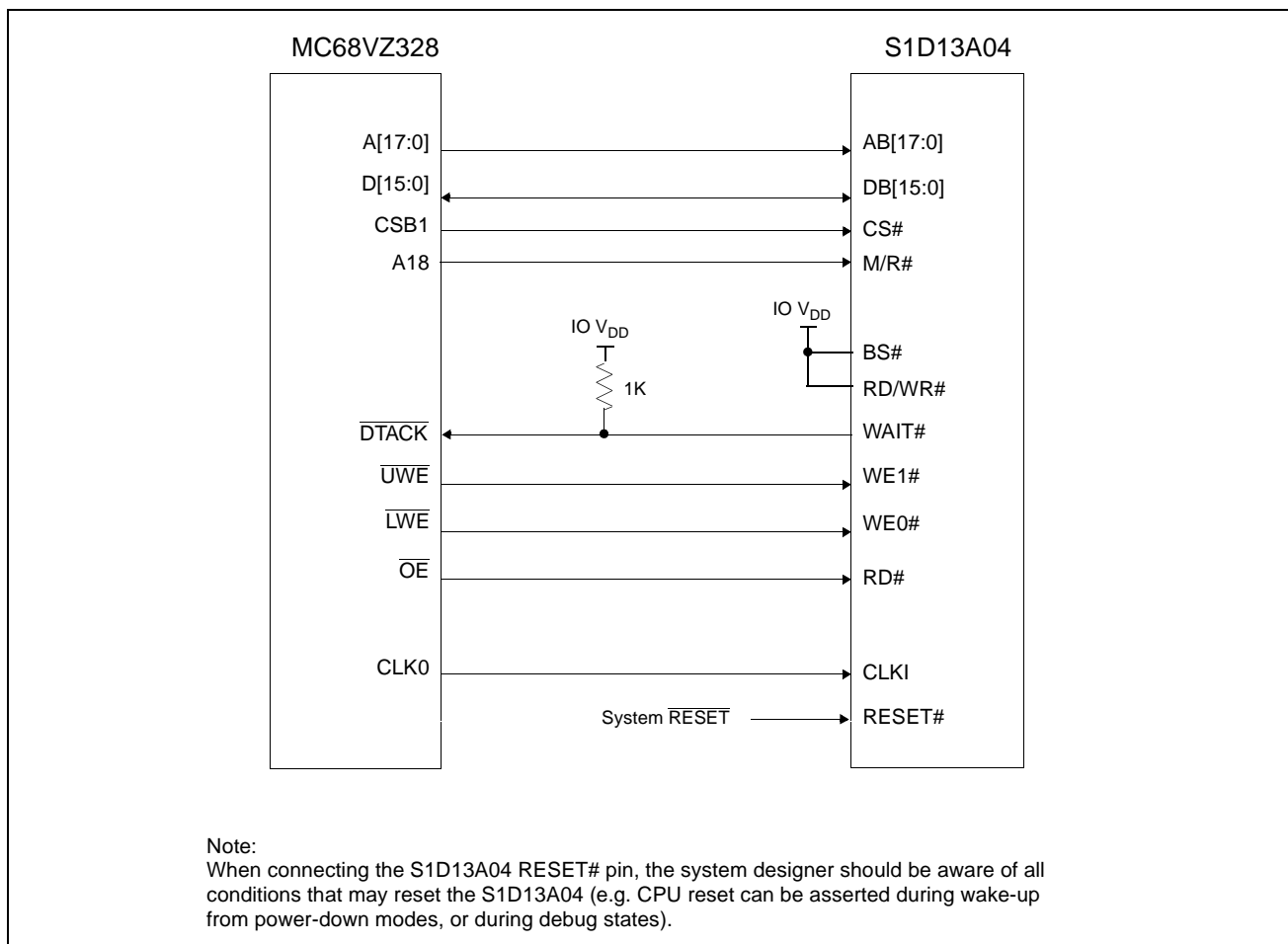


Figure 4-1: Typical Implementation of MC68VZ328 to S1D13A04 Interface

4.2 S1D13A04 Hardware Configuration

The S1D13A04 uses CNF6 through CNF0 to allow selection of the bus mode and other configuration data on the rising edge of RESET#. For details on configuration, refer to the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

The following table shows the configuration required for this implementation of a S1D13A04 to Motorola MC68VZ328 microprocessor.

Table 4-1: Summary of Power-On/Reset Options

S1D13A04 Configuration Input	Power-On/Reset State											
	1 (connected to IO V _{DD})	0 (connected to V _{SS})										
CNF4, CNF[2:0]	Select host bus interface as follows:											
	<table border="1"> <thead> <tr> <th>CNF4</th> <th>CNF2</th> <th>CNF1</th> <th>CNF0</th> <th>Host Bus</th> </tr> </thead> <tbody> <tr> <td>1</td> <td>1</td> <td>1</td> <td>0</td> <td>Dragonball Interface, Big Endian</td> </tr> </tbody> </table>	CNF4	CNF2	CNF1	CNF0	Host Bus	1	1	1	0	Dragonball Interface, Big Endian	
CNF4	CNF2	CNF1	CNF0	Host Bus								
1	1	1	0	Dragonball Interface, Big Endian								
CNF3	Reserved. Must be set to 1.											
CNF5	WAIT# is active high	WAIT# is active low										
CNF6	CLKI to BCLK divide ratio 2:1	CLKI to BCLK divide ratio 1:1										

configuration for Motorola MC68VZ328 microprocessor

4.2.1 Register/Memory Mapping

The S1D13A04 requires two 256K byte segments in memory for the display buffer and its internal registers. To accommodate this block size, it is preferable (but not required) to use one of the chip selects from groups A or B. Groups A and B can have a size range of 128K bytes to 16M bytes and groups C and D have a size range of 32K bytes to 16M bytes. Therefore, any chip select other than CSA0 would be suitable for the S1D13A04 interface.

In the example interface, chip select CSB1 controls the S1D13A04. A 512K byte address space is used by setting the SIZ bits of Chip Select Register B (FFFFF116h) to 512k bytes. The S1D13A04 internal registers occupy the first 256K byte block and the 160K byte display buffer is located in the second 256K byte block. A18 from the MC68VZ328 is used to select between these two 256K byte blocks.

4.2.2 MC68VZ328 Chip Select and Pin Configuration

The chip select used to map the S1D13A04 (in this example CSB1) must have its RO (Read Only) bit set to 0, its BSW (Bus Data Width) set to 1 for a 16-bit bus, and the WS (Wait states) bits should be set to 111b to allow the S1D13A04 to terminate bus cycles externally with \overline{DTACK} . The \overline{DTACK} pin function must be enabled with Register FFFFF433, Port G Select Register, bit 0.

If Chip Select Group B is used as the chip select module for the S1D13A04, SRAM timing must be enabled by setting the Chip Select Control Register 1 (FFFFF10Ah) SR16 bit = 0b.

Early cycle detection for static memory must be disabled by setting the Chip Select Control Register 2 (FFFFF10Ch) ECDS bit = 0b.

If \overline{DTACK} is not used, then the WS bits should be set to either 4, 6, 10, or 12 software wait states depending on the divide ratio between the S1D13A04 MCLK and BCLK. The WS bits should be set as follows.

Table 4-2: WS Bit Programming

S1D13A04 MCLK to BCLK Divide Ratio	WS Bits (wait states)
MCLK = BCLK	4
MCLK = BCLK ÷ 2	6
MCLK = BCLK ÷ 3	10
MCLK = BCLK ÷ 4	12

5 Software

Test utilities and display drivers are available for the S1D13A04. Full source code is available for both the test utilities and the drivers.

The test utilities are configurable for different panel types using a program called 13A04CFG (see document number X37A-B-001-xx), or by directly modifying the source. The display drivers can be customized by the OEM for different panel types, resolutions and color depths only by modifying the source.

The S1D13A04 test utilities and display drivers are available from your sales support contact (see Section 7, “*Sales and Technical Support*”) or at www.erd.epson.com.

6 References

6.1 Documents

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- Epson Research and Development, Inc., *S1D13A04 Hardware Functional Specification*, Document Number X37A-A-001-xx.
- Epson Research and Development, Inc., *S5U13A04B00C Rev. 1.0 Evaluation Board User Manual*, Document Number X37A-G-004-xx.
- Epson Research and Development, Inc., *S1D13A04 Programming Notes and Examples*, Document Number X37A-G-003-xx.

6.2 Document Sources

- Motorola Inc. Literature Distribution Center: (800) 441-2447.
- Motorola Inc. Website: www.mot.com.
- Epson Research and Development Website: www.erd.epson.com.

7 Sales and Technical Support

7.1 EPSON LCD/USB Companion Chips (S1D13A04)

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<http://www.epson.com.sg/>

7.2 Motorola MC68VZ328 Processor

- Motorola Design Line, (800) 521-6274.
- Local Motorola sales office or authorized distributor.

EPSON®



S1D13A04 LCD/USB Companion Chip

Interfacing to the Intel StrongARM SA-1110 Microprocessor

Document Number: X37A-G-013-01

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1 Introduction

This application note describes the hardware and software environment required to provide an interface between the S1D13A04 LCD/USB Companion Chip and the Intel StrongARM SA-1110 Microprocessor.

The designs described in this document are presented only as examples of how such interfaces might be implemented. This application note is updated as appropriate. Please check the Epson Research and Development website at www.erd.epson.com for the latest revision of this document before beginning any development.

We appreciate your comments on our documentation. Please contact us via email at documentation@erd.epson.com.

2 Interfacing to the StrongARM SA-1110 Bus

2.1 The StrongARM SA-1110 System Bus

The StrongARM SA-1110 microprocessor is a highly integrated communications microprocessor that incorporates a 32-bit StrongARM RISC processor core. The SA-1110 is ideally suited to interface to the S1D13A04 LCD controller and provides a high performance, power efficient solution for embedded systems.

2.1.1 StrongARM SA-1110 Overview

The SA-1110 system bus can access both variable-latency IO and memory devices. The SA-1110 uses a 26-bit address bus and a 32-bit data bus which can be used to access 16-bit devices. A chip select module with six chip select signals (each accessing 64M bytes of memory) allows selection of external devices. Only chip selects 3 through 5 (nCS[5:3]) may be used to select variable-latency devices which use RDY to extend access cycles. These chip selects are individually programmed in the SA-1110 memory configuration registers and can be configured for either a 16 or 32-bit data bus.

Byte steering is implemented using the four signals nCAS[3:0]. Each signal selects a byte on the 32-bit data bus. For example, nCAS0 selects bits D[7:0] and nCAS3 selects bits D[31:24]. For a 16-bit data bus, only nCAS[1:0] are used with nCAS0 selecting the low byte and nCAS1 selecting the high byte. The SA-1110 can be configured to support little or big endian mode.

2.1.2 Variable-Latency IO Access Overview

A data transfer is initiated when a memory address is placed on the SA-1110 system bus **and** a chip select signal (nCS[5:3]) is driven low. If all byte enable signals (nCAS[3:0]) are driven low, then a 32-bit transfer takes place. If only nCAS[1:0] are driven low, then a word transfer takes place through a 16-bit bus interface. If only one byte enable is driven low, then a byte transfer takes place on the respective data lines.

During a read cycle, the output enable signal (nOE) is driven low. A write cycle is specified by driving nOE high and driving the write enable signal (nWE) low. The cycle can be lengthened by driving RDY high for the time needed to complete the cycle.

2.1.3 Variable-Latency IO Access Cycles

The first nOE assertion occurs two memory cycles after the assertion of chip select (nCS3, nCS4, or nCS5). Two memory cycles prior to the end of minimum nOE or nWE assertion (RDF+1 memory cycles), the SA-1110 starts sampling the data ready input (RDY).

Samples are taken every half memory cycle until **three consecutive samples** (at the rising edge, falling edge, and following rising edge of the memory clock) indicate that the IO device is ready for data transfer. Read data is latched one-half memory cycle after the third successful sample (on falling edge). Then nOE or nWE is deasserted on the next rising edge and the address may change on the subsequent falling edge. Prior to a subsequent data cycle, nOE or nWE remains deasserted for RDN+1 memory cycles. The chip select and byte selects (nCAS[1:0] for 16-bit data transfers), remain asserted for one memory cycle after the final nOE or nWE deassertion of the burst.

The SA-1110 is capable of burst cycles during which the chip select remains low while the read or write command is asserted, precharged and reasserted repeatedly.

Figure 2-1: illustrates a typical variable-latency IO access read cycle on the SA-1110 bus.

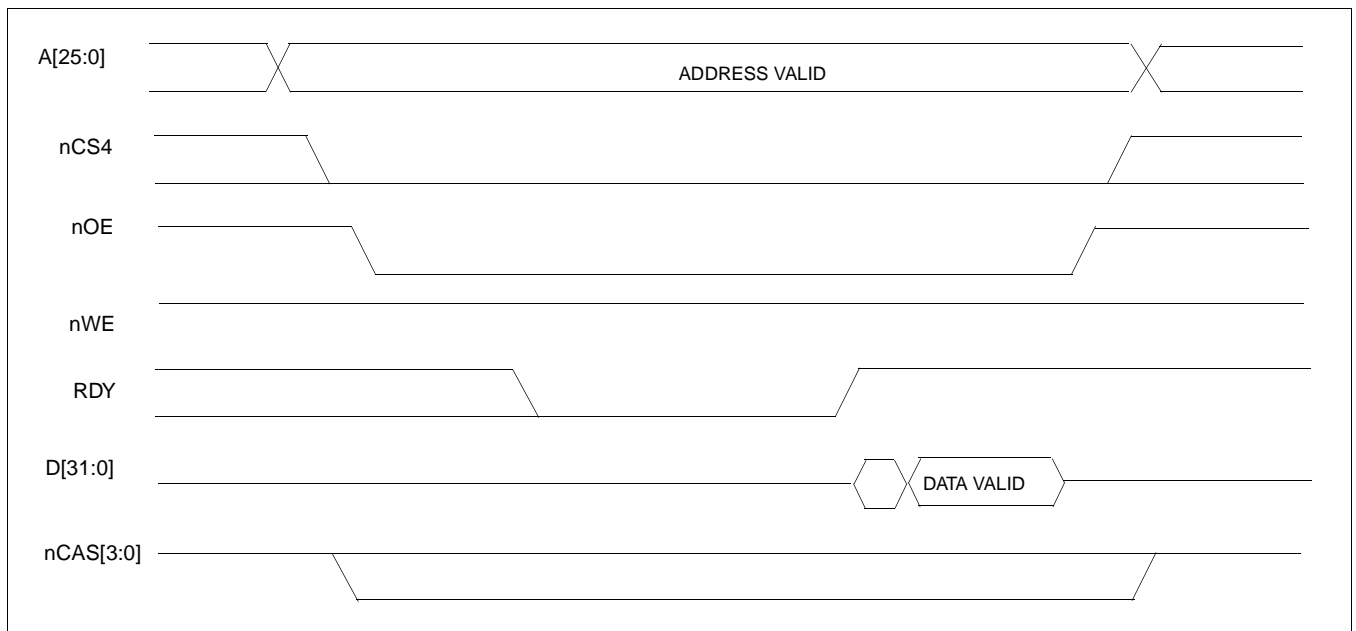
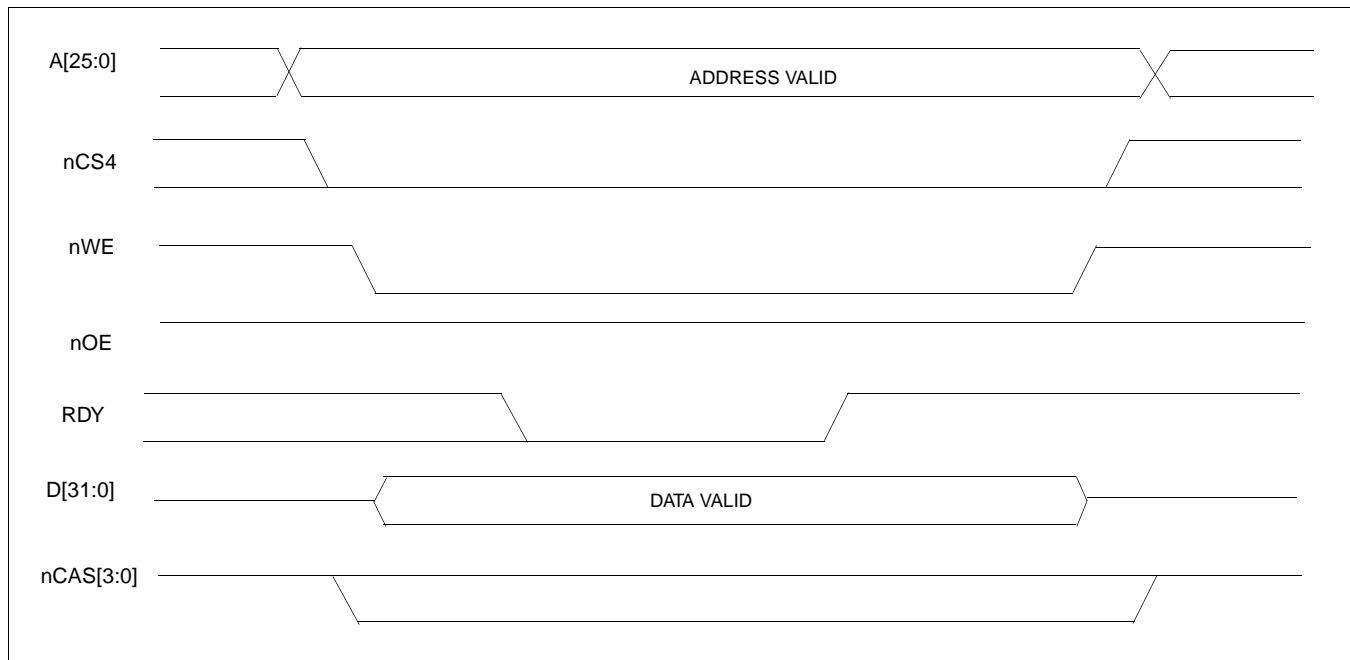


Figure 2-1: SA-1110 Variable-Latency IO Read Cycle

Figure 2-2: illustrates a typical variable-latency IO access write cycle on the SA-1110 bus.

*Figure 2-2: SA-1110 Variable-Latency IO Write Cycle*

3 S1D13A04 Host Bus Interface

The S1D13A04 directly supports multiple processors. The S1D13A04 implements a 16-bit Generic #2 Host Bus Interface which is most suitable for direct connection to the SA-1110.

The Generic #2 Host Bus Interface is selected by the S1D13A04 on the rising edge of RESET#. After releasing reset the bus interface signals assume their selected configuration. For details on S1D13A04 configuration, see Section 4.2, “S1D13A04 Hardware Configuration” on page 14.

3.1 Host Bus Interface Pin Mapping

The following table shows the functions of each Host Bus Interface signal.

Table 3-1: Host Bus Interface Pin Mapping

S1D13A04 Pin Name	SA-1110
AB[17:0]	A[17:0]
DB[15:0]	D[15:0]
WE1#	nCAS1
M/R#	A18
CS#	nCS4
CLKI	SDCLK2
BS#	Connect to IO _{VDD} from the S1D13A04
RD/WR#	Connect to IO _{VDD} from the S1D13A04
RD#	nOE
WE0#	nWE
WAIT#	RDY
RESET#	system RESET

3.2 Host Bus Interface Signal Descriptions

The S1D13A04 Generic #2 Host Bus Interface requires the following signals.

- CLKI is a clock input which is required by the S1D13A04 Host Bus Interface as a source for its internal bus and memory clocks. This clock is typically driven by the host CPU system clock. For this example, it is driven by one of the SA-1110 signals SDCLK1 or SDCLK2 (The example implementation in this document uses SDCLK2). For further information, see Section 4.3, “*StrongARM SA-1110 Register Configuration*” on page 15.
- The address inputs AB[17:0], and the data bus DB[15:0], connect directly to the SA-1110 address bus (A[17:0]) and data bus (D[15:0]), respectively. CNF4 must be set to select little endian mode.
- M/R# (memory/register) selects between memory or register accesses. This signal is generated by the external address decode circuitry. For this example, M/R# is connected to address line A18, allowing system address A18 to select between memory or register accesses.
- Chip Select (CS#) must be driven low by nCS_x (where x is the SA-1110 chip select used) whenever the S1D13A04 is accessed by the SA-1110.
- WE1# connects to nCAS1 (the high byte enable signal from the SA-1110) which in conjunction with address bit 0 allows byte steering of read and write operations.
- WE0# connects to nWE (the write enable signal from the SA-1110) and must be driven low when the SA-1110 is writing data to the S1D13A04.
- RD# connects to nOE (the read enable signal from the SA-1110) and must be driven low when the SA-1110 is reading data from the S1D13A04.
- WAIT# connects to RDY and is a signal output from the S1D13A04 that indicates the SA-1110 must wait until data is ready (read cycle) or accepted (write cycle) on the host bus. Since SA-1110 accesses to the S1D13A04 may occur asynchronously to the display update, it is possible that contention may occur in accessing the S1D13A04 internal registers and/or display buffer. The WAIT# line resolves these contentions by forcing the host to wait until the resource arbitration is complete.
- The Bus Start (BS#) and RD/WR# signals are not used for this Host Bus Interface and should be tied high (connected to IOV_{DD}).
- The RESET# (active low) input of the S1D13A04 may be connected to the system RESET.

4 StrongARM SA-1110 to S1D13A04 Interface

4.1 Hardware Description

The SA-1110 microprocessor provides a variable latency I/O interface that can be used to support an external LCD controller. By using the Generic # 2 Host Bus Interface, no glue logic is required to interface the S1D13A04 and the SA-1110.

A pull-up resistor is attached to WAIT# to speed up its rise time when terminating a cycle.

BS# (bus start) and RD/WR# are not used by the Generic #2 Host Bus Interface and should be tied high (connected to IO V_{DD}).

The following diagram shows a typical implementation of the SA-1110 to S1D13A04 interface.

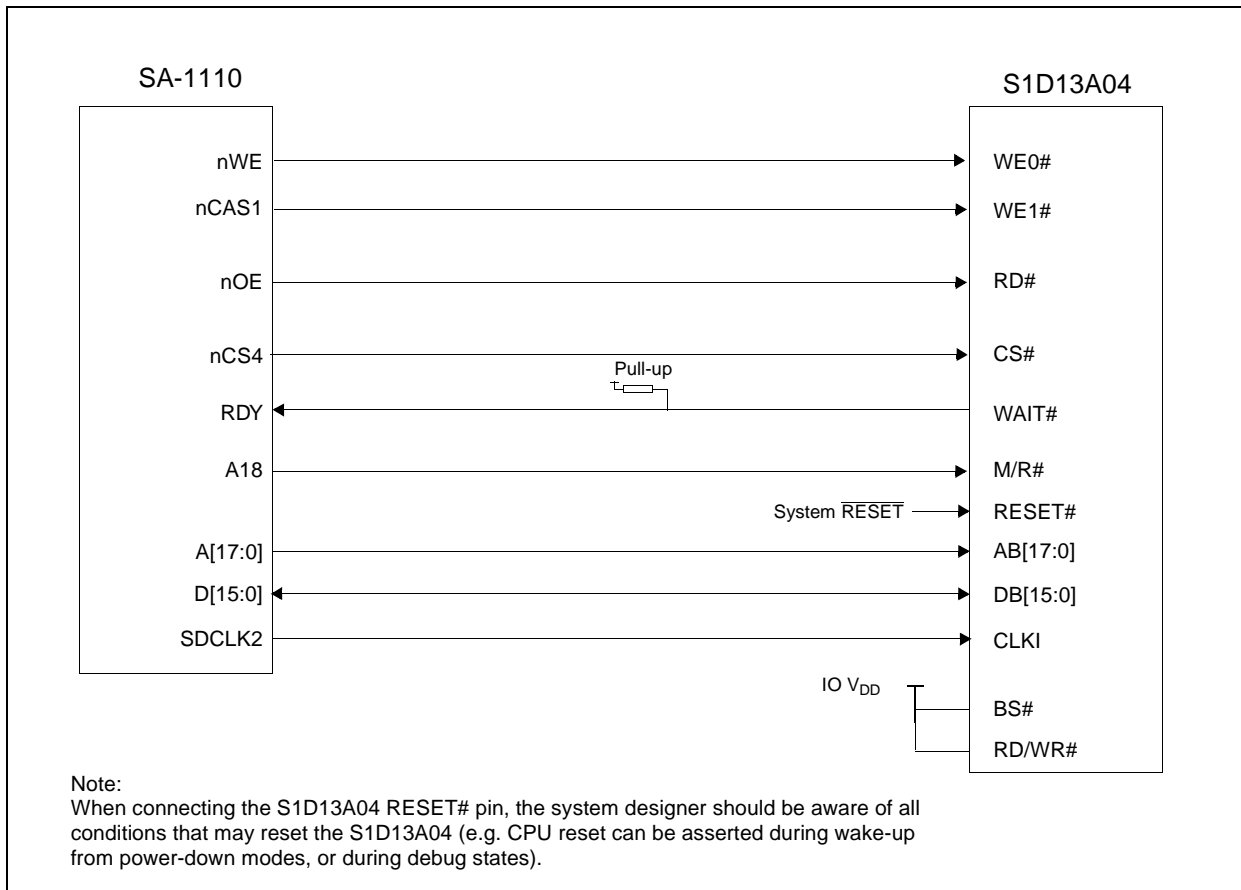


Figure 4-1: Typical Implementation of SA-1110 to S1D13A04 Interface

4.2 S1D13A04 Hardware Configuration

The S1D13A04 uses CNF6 through CNF0 to allow selection of the bus mode and other configuration data on the rising edge of RESET#. For details on configuration, refer to the *S1D13A04 Hardware Functional Specification*, document number X37A-A-001-xx.

The following table shows the configuration required for this implementation of a S1D13A04 to SA-1110 interface.

Table 4-1: Summary of Power-On/Reset Options

S1D13A04 Configuration Input	Power-On/Reset State											
	1 (connected to IO V _{DD})	0 (connected to V _{SS})										
CNF4, CNF[2:0]	Select host bus interface as follows: <table border="1"> <thead> <tr> <th>CNF4</th> <th>CNF2</th> <th>CNF1</th> <th>CNF0</th> <th>Host Bus</th> </tr> </thead> <tbody> <tr> <td>0</td> <td>1</td> <td>0</td> <td>0</td> <td>Generic #2, Little Endian</td> </tr> </tbody> </table>		CNF4	CNF2	CNF1	CNF0	Host Bus	0	1	0	0	Generic #2, Little Endian
CNF4	CNF2	CNF1	CNF0	Host Bus								
0	1	0	0	Generic #2, Little Endian								
CNF3	Reserved. Must be set to 1.											
CNF5	WAIT# is active high	WAIT# is active low										
CNF6	CLKI to BCLK divide ratio 2:1	CLKI to BCLK divide ratio 1:1										
	configuration for SA-1110 microprocessor											

4.3 StrongARM SA-1110 Register Configuration

The SA-1110 requires configuration of several of its internal registers to interface to the S1D13A04 Generic #2 Host Bus Interface.

- The Static Memory Control Registers (MSC[2:0]) are read/write registers containing control bits for configuring static memory or variable-latency IO devices. These registers correspond to chip select pairs nCS[5:4], nCS[3:2], and nCS[1:0] respectively. Each of the three registers contains two identical CNFG fields, one for each chip select within the pair. Since only nCS[5:3] controls variable-latency IO devices, MSC2 and MSC1 should be programmed based on the chip select used.

Parameter RTx<1:0> should be set to 01b (selects variable-latency IO mode).

Parameter RBWx should be set to 1 (selects 16-bit bus width).

Parameter RDFx<4:0> should be set according to the maximum desired CPU frequency as indicated in the table below.

Table 4-2: RDFx Parameter Value versus CPU Maximum Frequency

CPU Frequency (MHz)	RDFx
57.3 - 85.9	1
88.5 - 143.2	2
147.5 - 200.5	3
206.4 - 221.2	4

Parameter RDNx<4:0> should be set to 0 (minimum command precharge time).

Parameter RRRx<2:0> should be set to 0 (minimum nCSx precharge time).

- The S1D13A04 endian mode is set to little endian. To program the SA-1110 for little endian set bit 7 of the control register (register 1) to 0.
- The CLKI signal input to the S1D13A04 from one of the SDCLK[2:1] pins is a derivative of the SA-1110 internal processor speed (either divide by 2 or 4). The S1D13A04 Generic #2 Host Bus Interface has a maximum BCLK of 50MHz. Therefore, if the processor clock is higher than 100MHz, either divide the BCLK input using the S1D13A04 configuration pin CNF6 (see Table 4-1: “Summary of Power-On/Reset Options”) or set SDCLK1/SDCLK2 to CPU clock divided by four using the DRAM Refresh Control Register (MDREFR bit 26 = 1 for SDCLK2, MDREFR bit 22 = 1 for SDCLK1).

4.4 Register/Memory Mapping

The S1D13A04 is a memory-mapped device. The SA-1110 uses the memory assigned to a chip select (nCS4 in this example) to map the S1D13A04 internal registers and display buffer. The S1D13A04 uses two 256K byte blocks which are selected using A18 from the SA-1110 (A18 is connected to the S1D13A04 M/R# pin). The internal registers occupy the first 256K bytes block and the 160K byte display buffer occupies the second 256K byte block.

Each variable-latency IO chip select is assigned 128M Bytes of address space. Therefore; if nCS4 is used the S1D13A04 registers will be located at 4000 0000h and the display buffer will be located at 4004 0000h. These blocks are aliased over the entire 128M byte address space.

Note

If aliasing is not desirable, the upper addresses must be fully decoded.

5 Software

Test utilities and display drivers are available for the S1D13A04. Full source code is available for both the test utilities and the drivers.

The test utilities are configurable for different panel types using a program called 13A04CFG (see document number X37A-B-001-xx), or by directly modifying the source. The display drivers can be customized by the OEM for different panel types, resolutions and color depths only by modifying the source.

The S1D13A04 test utilities and display drivers are available from your sales support contact (see Section 7, “*Sales and Technical Support*”) or www.erd.epson.com.

6 References

6.1 Documents

- Intel Corporation, *StrongARM® SA-1110 Microprocessor Advanced Developer's Manual*, Order Number 278240-001.
- Epson Research and Development, Inc., *S1D13A04 Hardware Functional Specification*, Document Number X37A-A-001-xx.
- Epson Research and Development, Inc., *S1D13A04 Programming Notes and Examples*, Document Number X37A-G-003-xx.
- Epson Research and Development, Inc., *S5U13A04B00C Rev. 1.0 ISA Bus Evaluation Board User Manual*, Document Number X37A-G-004-xx.

6.2 Document Sources

- Intel Developers Website: <http://developer.intel.com>.
- Intel Literature contact: 1(800) 548-4725.
- Epson Research and Development Website: www.erd.epson.com.

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