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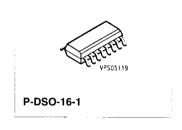
IC for Inductive Proximity Switches with 0.5 A Output Stage

TCA 705

Advance Information SPT IC

Features

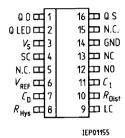
- Output current up to 500 mA
- Switch-ON and switch-OFF
- High-side-switch and low-side-switch
- Short-circuit and overload protection
- Supply voltage range from 7.5 V to 65 V
- Few external components
- LED output for "switch-ON/OFF" and short-circuit failure indication
- High switching frequency
- High noise immunity, low temperature coefficient
- Temperature protection



T	Гуре	Ordering Code	Package
7	CA 705 G	Q67000-A8310	P-DSO-16-1 (SMD)

▼ New type

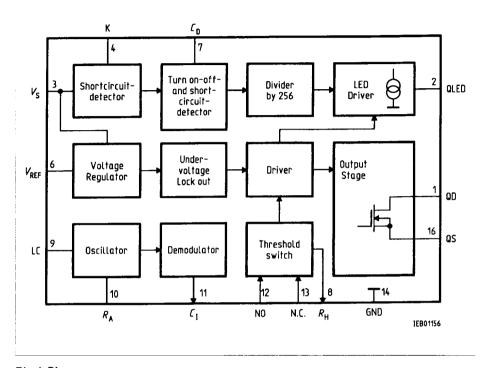
The TCA 705 is a monolithic IC in the new Smart Power Technology (SPT) for designing excellent inductive proximity switches at low cost and with little space. Features like 0.5 A output current, short-circuit, overload and temperature protection classify the TCA 705 as a versatile general-purpose circuit. The separate LED output indicates the switching state of the sensor (switch-ON / switch-OFF) and an eventual short-circuit failure.



Pin Configurations (top view)

Pin Definitions and Functions

Pin	Symbol	Function
1	QD	Output; open drain
2	Q LED	Output LED
3	$V_{\mathbb{S}}$	Supply voltage
4	sc	Short-circuit detector
5	N.C.	Not connected
6	V_{REF}	Reference voltage
7	C_{D}	Turn-ON delay / Turn-OFF delay / Short-circuit delay
8	R_{Hys}	Hysteresis
9	LC	Oscillator
10	R_{Dist}	Distance
11	C_1	Integrating capacitance
12	NO	Input driver (normally open)
13	NC	Input driver (normally closed)
14	GND	Ground
15	N.C.	Not connected
16	QS	Output; open source



Block Diagram

Pin Functions

Pins 1/16: QD, QS

A load has to be connected between QS and ground, i.e. high-side switch respectively between QD and $V_{\rm S}$, i.e. low-side switch. The current through the load is 500 mA max. To protect the IC against destruction by short-circuit or overload, a capacitor has to be connected between pin $C_{\rm D}$ and ground.

Pin 2: Q LED

Output QLED drives a LED with up to 10 mA, connected between QLED and $V_{\rm S}$. The LED indicates the switching state (switch-ON / switch-OFF) and blinks during an eventual short-circuit failure.

Pin 3: V_s

Outputs QD , QS and QLED are inhibited as long as the voltage on $V_{\rm S}$ is below approx. 7.5 V. They are enabled between approx 8.0 and 8.5 V, the basic function of the circuit is then ensured. During the turn-ON and turn-OFF of $V_{\rm S}$ there are consequently no undesirable static states. The operating data and characteristics apply upwards from 8.5 V. See pin $C_{\rm D}$ for the avoidance of erroneous pulses during oscillator built-up.

Pin 4: SC

SC serves for short-circuit sensing in the output circuit that is to be protected. The current can be sensed referred to ground or $V_{\rm S}$ (see application circuits 2 and 3). The current sensing is made by a dedicated resistor ($R_{\rm SC}$) in the output circuit. For a voltage drop \geq 0.3 V across $V_{\rm S}$ and SC or across ground and SC, all outputs are turned off after the turn-OFF delay (brief glitches on the outputs or the charging of line capacitances therefore do not trigger the short-circuit protection). After a pause of about 200 times of the turn-OFF delay, the outputs turn on again. If the short-circuit is still present, the turn-OFF cycle will start up a new.

Pin 6: V_{REF}

The internal stabilized voltage of the IC of approx. 3.5 V appears on this pin. A capacitor can be connected between V_{REF} and ground to improve the noise immunity of the overall circuit function. The output current at this pin is up to 10 mA.

Pin 7: C_D

A capacitor on this pin delays the activation of the output after the supply voltage is applied (turn-ON delay). In this way erroneous pulses are prevented on the output during built-up of the oscillator.

If $V_{\rm S}$ falls to less than 7.8 to 8 V, the output is not inhibited until after a turn-OFF delay time, this also being determined by $C_{\rm D}$. In this way the delayed turn-ON operation described above is suppressed if there are just short glitches (voltage dips) on $V_{\rm S}$. This is of particular advantage for large core diameters, because in such cases a relatively long turn-ON delay has to be selected, since this would otherwise start to run every time there was a brief voltage dip.

The capacitor C_D also sets the turn-OFF delay and the pause duration in short-circuit operation.

Pin 8: R_{Hvs}

Depending on the status of the circuit, $R_{\rm Hys}$ will be high-impedance or low-impedance to ground (open collector). If the distance resistance (see $R_{\rm Dist}$) is split into two resistors $R_{\rm Dist}$ and $R_{\rm Hys}$, a distance hysteresis can be set by means of $R_{\rm Hys}$. If series hysteresis is applied, $R_{\rm Hys}$ is connected in series with $R_{\rm Dist}$ or shorted. If parallel hysteresis is applied, $R_{\rm Hys}$ is connected in parallel with $R_{\rm Dist}$ or made high-impedance (see application circuit 1).

Pin 9: LC

The resonant circuit of the proximity switch is connected between LC and ground.

Pin 10: R_{Dist}

A resistor between this pin and ground sets the current in the oscillator circuit. The greater the value of the resistor, the smaller is the current feed from the oscillator into the resonant circuit and the greater therefore is the switching distance. The greater the quality of the resonant circuit, the greater is the value of the distance resistor necessary for setting a certain switching distance.

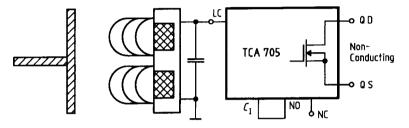
Pin 11: C₁

 C_1 can remain open; if high noise immunity is to be achieved however, this pin should be provided with a serial RC element (R_1, C_1) . If pin 7 (C_D) is not used, a correctly dimensioned RC element on pin C_1 will also prevent any erroneous pulses on the output when the supply voltage is turned on (see application circuit 1).

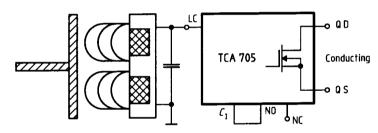
Pins 11/12/13: C_1 , NO, NC

A direct connection between C_1 and NO will result in a "normally open" output stage, i. e. switch-ON. A direct connection between C_1 and NC will result in a "normally closed" output stage, i. e. switch-OFF.

Resonant circuit, normally open, not damped:



Resonant circuit, normally open, damped:



	Not Damped	Damped
Normally Open C _I /NO	Non- Conducting	Conducting
Normally Closed \mathcal{E}_{I} /NC	Conducting	Non- Conducting

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Functional Description

This circuit is used to design inductive proximity switches. The resonant circuit of the LC oscillator is implemented with an open half-pot ferrite and a capacitor in parallel (pin LC). If a metallic target is moved closer to the open side of the half-pot ferrite, energy is drawn from the resonant circuit and the amplitude of the oscillation is decreasing accordingly. This change in amplitude is transmitted to a threshold switch by means of a demodulator and triggers the output (see operation schematic).

By means of an external distance resistor $R_{\rm A}$ on the oscillator (pin $R_{\rm Dist}$) it is possible to set the switching distance within wide limits. The optimal distances are 0.1 to 0.6 of the diameter of the half-pot ferrite in use, but extensions to both of these parameters can also be produced. The circuit also enables the setting of a path hysteresis by switching an external distance resistor ($R_{\rm H}$) via pin $R_{\rm Hys}$ (see application circuit 1).

The device includes one integrated output stage for max. 500 mA output current. It can be used as "normally open" or "normally closed" by means of programming the NO-respectively NC-input driver and as a high-side- or low-side-switch (see application circuits 2 and 3). The output stage is automatically protected against destruction by short-circuit or overload.

By means of a capacitor on pin $C_{\rm D}$ it is possible to set the response delay and the turn-OFF time of short-circuit protection. The same capacitor also defines the turn-ON delay of the output stage when the supply voltage is applied, whereby the output stage is inhibited during built-up of the oscillator. Finally $C_{\rm D}$ is used to produce a turn-OFF delay of the output stage so that the turn-ON delay is prevented from running its full length if there are brief voltage dips on $V_{\rm S}$.

Absolute Maximum Ratings

 $T_{\rm A} = -40$ to 110 °C

Parameter	Symbol	Limit	Values	Unit	Test Condition	
		min.	max.			
Supply voltage	V_{S}	- 0.3	70	٧	-	
Output voltage	V_{Q}	- 0.3	V _S + 1	٧	_	
Output current	$-I_{Q}$	-	Ī-	-	internally limited	
Current from V _{REF}	- I _{REF}	0	1	mA	_	
Voltage on SC	$V_{ m SC}$	0	V_{S}	V	-	
Current from R _{Dist}	- I _{RDist}	0	2	mA	-	
Current to R _{Hys}	I _{RHys}	0	2	mA	_	
Storage temperature	T_{stg}	- 65	125	.C	_	
Thermal resistance (system-air)	R _{th SA}		110	K/W	_	
Junction temperature	$T_{ m j}$	-	125 150	°C	– max. 70.000 h	
Capacitor on C_D	C_{V}	-	50	nF	_	

Operating Range

Supply voltage	V_{S}	7.5	65	٧	-
Ambient temperature Distance and hysteresis resistance	T _A	- 40	110	°C	_
$R_{\rm A}$ and $R_{\rm H}$ in series	R _A R _H	300 0	- -	$\Omega \ \Omega$	
R_{A} and R_{H} parallel	R_{A}/R_{H}	300	_	Ω	-
Output current	- I _Q	_	500	mA	_

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Characteristics

 $7.5 \text{ V} \le V_{\text{S}} \le 65 \text{ V}; T_{\text{A}} = -40 \text{ to } 110 \,^{\circ}\text{C}$

Parameter	Symbol	L	imit Va	lues	Unit
	-	min.	typ.	max.	<u>.</u>
Power Supply ($V_{ m S}$)					
Current consumption	Is		550	750	μΑ
Turn-ON threshold (output active)	V_{TON1}		8.0	8.5	V
Turn-OFF threshold (output disabled)	V_{TOFF1}	7.5	8.0		V
Hysteresis $V_{\text{TON1}} - V_{\text{TOFF1}}$	ΔV_{Hys1}	_	40	_	mV
Reference Voltage	V_{REF}	3.0	3.3	3.5	mV
Oscillator (LC, R _{Dist}) Oscillator frequency				12	MHz
Oscillator amplitude	$-\frac{f_{\text{osc}}}{A_{\text{osc}}}$	- -		3	V_{PP}
Demodulator, Threshold Sw Threshold on C_1	V_{CI}		2		· V
Hysteresis on C_1	V_{HysCI}	_ -	0.8	1_	†v
Current in C ₁	I_{Cl}	_	7	+	μA
Current from C ₁	- I _{CI}	-	6	<u>-</u>	μА
Switching frequency $(C_1 < 50 \text{ pF})$	fs		5	_	kHz
Integrating capacitance	Cı		1	1-	nF
Reference Voltage ($V_{ extsf{REF}}$)					
D-4	V_{REF}	3.0	3.3	3.5	· · · · · · · · · · · · · · · · · · ·
Reference voltage I _{REF} = 0 to 1 mA	HEF	0.0			
I _{REF} = 0 to 1 mA		!			
•		!	0.65	0.82	ms/nF
$I_{REF} = 0$ to 1 mA Turn-ON, Turn-OFF and Sho	rt-Circuit Dela	y (C _D)			

Short-circuit pause

0.35

 t_{P}

0.5

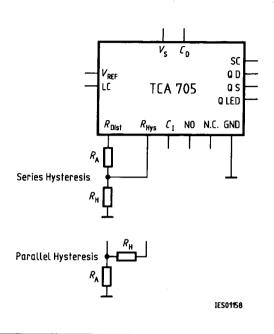
0.65

ms/nF

Characteristics (cont'd)

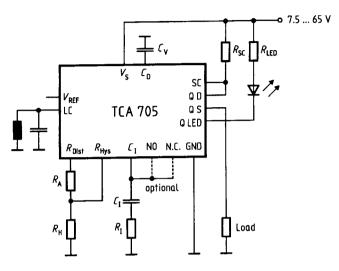
7.5 V $\leq V_{\rm S} \leq$ 65 V; $T_{\rm A} = -$ 40 to 110 $^{\circ}{\rm C}$

Parameter	Symbol	Limit Values			Unit
		min.	typ.	max.	
Outputs (QD, QS, Q LED)					
Current on Q LED	I_{QLED} R_{DSON}	0.3	_	10 0.5	mA Ω
Residual current on QD, QS	I_{QRES}		-	100	μΑ
Short-Circuit Detector (SC)					
Trigger level ref. to $V_{\rm S}$	V_{SCS}	0.255	0.3	0.345	٧
Trigger current	I _{SCS}	_		30	μΑ
Trigger level ref. to ground	$V_{\sf SCO}$	0.255	0.3	0.345	٧
Trigger current	- I _{sco}			6	μА



Application Circuit 1 Series Hysteresis / Parallel Hysteresis

For explanation see under "Pin Functions".

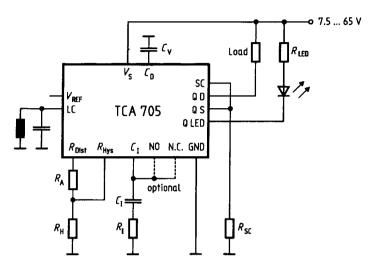


 $\label{thm:ligh-side-switch} \begin{tabular}{ll} High-side-switch, short-circuit-proof, LED indicator, normally closed (normally open). \end{tabular}$

Short-circuit sensing: $R_{SC} = \frac{0.3 \text{ V}}{\text{max.} I_{D}}$

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Application Circuit 2



Low-side-switch, short-circuit-proof, LED indicator, normally open (normally closed).

Short-circuit sensing:
$$R_{SC} = \frac{0.3 \text{ V}}{\text{max.} I_Q}$$

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Application Circuit 3