

MAS6179

This is preliminary information on a new product under development. Micro Analog Systems Oy reserves the right to make any changes without notice.



AM Receiver IC

- Tri Band Receiver IC
- High Sensitivity
- Very Low Power Consumption
- Wide Supply Voltage Range
- Power Down Control
- Control for AGC On
- High Selectivity by Crystal Filter
- Fast Startup Feature

DESCRIPTION

The MAS6179 AM-Receiver chip is a highly sensitive, simple to use AM receiver specially intended to receive time signals in the frequency range from 40 kHz to 100 kHz. Only a few external components are required for time signal receiving. The circuit has preamplifier, wide range automatic gain control, demodulator and output comparator built in. The output signal can be processed directly by an additional digital circuitry to extract the data from the received signal. The control for AGC (automatic gain

control) can be used to switch AGC on or off if necessary. MAS6179 supports up to three frequency band operation by switching between three crystal filters and two additional antenna tuning capacitors.

MAS6179 has differential input and two internal compensation capacitor options for compensating shunt capacitances of different crystals (See ordering information on page 10).

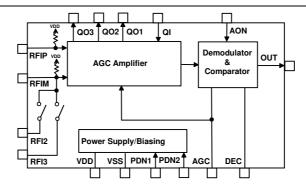
FEATURES

- Tri Band Receiver IC
- Highly Sensitive AM Receiver, 0.4 μV_{RMS} typ.
- Wide Supply Voltage Range from 1.1 V to 3.6 V
- Very Low Power Consumption
- Power Down Control
- Fast Startup
- Only a Few External Components Necessary
- Control for AGC On
- Wide Frequency Range from 40 kHz to 100 kHz
- High Selectivity by Quartz Crystal Filter
- Differential Input

APPLICATIONS

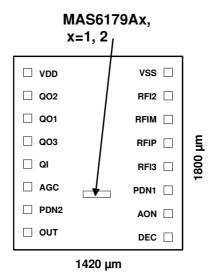
Multi Band Time Signal Receiver WWVB (USA),
 JJY (Japan), DCF77 (Germany), MSF (UK), HBG (Switzerland) and BPC (China)

BLOCK DIAGRAM





PAD LAYOUT



DIE size = 1.42 x 1.80 mm; rectangular PAD 80 μm x 80 μm

Note: Because the substrate of the die is internally connected to VSS, the die has to be connected to VSS or left floating. Please make sure that VSS is the first pad to be bonded. Pick-and-place and all component assembly are recommended to be performed in ESD protected area.

Note: Coordinates are pad center points where origin has been located in bottom-left corner of the silicon die.

| Pad Identification | Name | X-coordinate | Y-coordinate | Note |
|--|------|--------------|--------------|------|
| Power Supply Voltage | VDD | 154 μm | 1580 μm | |
| Quartz Filter Output for Crystal 2 | QO2 | 154 μm | 1393 μm | |
| Quartz Filter Output for Crystal 1 | QO1 | 154 μm | 1207 μm | |
| Quartz Filter Output for Crystal 3 | QO3 | 154 μm | 1021 μm | |
| Quartz Filter Input for Crystals | QI | 154 μm | 835 μm | |
| AGC Capacitor | AGC | 154 μm | 648 μm | |
| Power Down/Frequency Selection Input 2 | PDN2 | 154 μm | 462 μm | 1 |
| Receiver Output | OUT | 154 μm | 276 μm | 2 |
| Demodulator Capacitor | DEC | 1266 μm | 276 μm | |
| AGC On Control | AON | 1266 μm | 462 μm | 3 |
| Power Down/Frequency Selection Input 1 | PDN1 | 1266 μm | 648 μm | 1 |
| Receiver Input 3 (for Antenna Capacitor 3) | RFI3 | 1266 μm | 835 μm | |
| Positive Receiver Input | RFIP | 1266 μm | 1021 μm | 4 |
| Negative Receiver Input | RFIM | 1266 μm | 1207 μm | 4 |
| Receiver Input 2 (for Antenna Capacitor 2) | RFI2 | 1266 μm | 1393 μm | |
| Power Supply Ground | VSS | 1266 μm | 1580 μm | |

Notes:

- 1) PDN1 = PDN2 = VDD means receiver off
 - Fast start-up is triggered when the receiver is after power down controlled to power up
- 2) OUT = VSS when carrier amplitude at maximum; OUT = VDD when carrier amplitude is reduced (modulated)
 - the output is a current source/sink with $|I_{OUT}| > 5 \mu A$
 - at power down the output is pulled to VSS (pull down switch)
- 3) AON = VSS means AGC off (hold current gain level); AON = VDD means AGC on (working)
 - Internal pull-up with current < 1 μA which is switched off at power down
- 4) Receiver inputs RFIP and RFIM have both 600 kΩ biasing resistors towards VDD



ABSOLUTE MAXIMUM RATINGS

| Parameter | Symbol | Conditions | Min | Max | Unit |
|-----------------------|---------------------|------------|----------------------|----------------------|------|
| Supply Voltage | V_{DD} - V_{SS} | | -0.3 | 5.5 | V |
| Input Voltage | V _{IN} | | V _{SS} -0.3 | V _{DD} +0.3 | V |
| Power Dissipation | P _{MAX} | | | 100 | mW |
| Operating Temperature | T _{OP} | | -40 | +85 | °C |
| Storage Temperature | T _{ST} | | -55 | +150 | °C |

ELECTRICAL CHARACTERISTICS

Operating Conditions: VDD = 1.5V, Temperature = 25°C unless otherwise noted

| Parameter | Symbol | Conditions: VDD = 1.3 | Min | Тур | Max | Unit |
|---|---------------------------------------|--|-----------------------|----------------------|----------|----------|
| Operating Voltage | V_{DD} | | 1.10 | | 3.6 | V |
| Current Consumption | I _{DD} | VDD=1.5 V, Vin=0 μVrms VDD=1.5 V, Vin=20 mVrms VDD=3.6 V, Vin=0 μVrms VDD=3.6 V, Vin=20 mVrms | 40 24 | 55 40 58 43 | 80 65 | μА |
| Stand-By Current | I _{DDoff} | | | | 0.1 | μΑ |
| Input Frequency Range | f _{IN} | | 40 | | 100 | kHz |
| Minimum Input Voltage | $V_{IN min}$ | | | 0.4 | 1 | μVrms |
| Maximum Input Voltage | V _{IN max} | | 20 | | | mVrms |
| Receiver Input Resistance Receiver Input Capacitance | R_{RFI} C_{RFI} | f=40kHz77.5 kHz | | 600 0.5 | | kΩ pF |
| RFI2 Switch On Resistance RFI2 Switch Off Capacitance | R _{ON2} C _{OFF2} | VDD=1.4 V | | 5 TBD | 15 | Ω pF |
| RFI3 Switch On Resistance RFI3 Switch Off Capacitance | R _{ON3} C _{OFF3} | VDD=1.4 V | | 2.5 TBD | 15 | Ω pF |
| Input Levels $ I_{IN} $ <0.5 μ A | $V_{IL} \ V_{IH}$ | | V _{DD} -0.35 | | 0.35 | V |
| Output Current V _{OL} <0.2 V _{DD} ;V _{OH} >0.8 V _{DD} | I _{OUT} | | 5 | | | μΑ |
| Output Pulse | T _{100ms} | $1 \mu Vrms \le V_{IN} \le 20 mVrms$ | 50 | | 140 | ms |
| | T _{200ms} | 1 μVrms ≤ V _{IN} ≤ 20 mVrms | 150 | | 230 | ms |
| | T _{500ms} | 1 μVrms ≤ V _{IN} ≤ 20 mVrms | 400 | 500 | 600 | ms |
| | T _{800ms} | 1 μVrms ≤ V _{IN} ≤ 20 mVrms | 700 | 800 | 900 | ms |
| Startup Time | T _{Start} | Fast Start-up, Vin=0.4 μVrms Fast Start-up, Vin=20 mVrms | | 1.3 3.5 | | S |
| Output Delay Time | T _{Delay} | _ | | 50 | 100 | ms |

Note: Stand-by current consumption may increase if V_{\parallel} and V_{\parallel} differ from VDD and GND respectively. TBD = To Be Defined



FREQUENCY SELECTION

The frequency selection and power down control is accomplished via two digital control pins PDN1 and PDN2. The control logic is presented in table 1.

Table 1. Frequency selection and power down control

| PDN1 | PDN2 | RFI2 Switch | RFI3 Switch | Selected Crystal Output | Description |
|------|------|-------------|-------------|----------------------------|--|
| High | High | Open | Open | - | Power down |
| High | Low | Open | Open | QO1 | Frequency 1 |
| Low | High | Closed | Open | QO2 | Frequency 2, RFI2 capacitor connected in parallel with antenna |
| Low | Low | Closed | Closed | QO3 | Frequency 3, RFI2 and RFI3 capacitors connected in parallel with antenna |

The internal antenna tuning capacitor switches (RFI2, RFI3) and crystal filter output switches (QO1, QO2, QO3) are controlled according table 1.

If frequency 1 is selected the RFI2 and RFI3 switches are open and only crystal output QO1 is active. Antenna frequency is determined by antenna inductor L_{ANT} (see Typical Application on page 6), antenna capacitor C_{ANT1} and parasitic capacitances related to antenna inputs RFIP, RFIM, RFI2 and RFI3 (see Antenna Tuning Considerations below). Frequency 1 is the highest frequency of the three selected frequencies.

If frequency 2 is selected then RFI2 switch is closed to connect C_{ANT2} to pin RFIM in parallel with ferrite antenna and tune it to frequency 2. Then only crystal output QO2 is active. Frequency 2 is the

medium frequency of the three selected frequencies.

If frequency 3 is selected both RFI2 and RFI3 switches are closed to connect both C_{ANT2} and C_{ANT3} capacitors to RFIM pin in parallel with ferrite antenna and tune it to frequency 3. Then only crystal output QO3 is active. Frequency 3 is the lowest frequency of the three selected frequencies.

It is recommended to switch the device to power down for at least 50ms before switching to another frequency. This guarantees fast startup in switching to another frequency. During minimum 50ms power down time the AGC capacitor voltage is completely pulled to VDD and the proper fast startup conditions are met. Without proper fast startup control the startup time can be several minutes. With fast startup it is shortened typically to few seconds.



ANTENNA TUNING CONSIDERATIONS

The ferrite bar antenna having inductance L_{ANT} and parasitic coil capacitance C_{COIL} is tuned to three reception frequencies f_1 , f_2 and f_3 by parallel capacitors C_{ANT1} , C_{ANT2} and C_{ANT3} . The receiver input stage and internal antenna capacitor switches have capacitances C_{RFI} , C_{OFF2} , C_{OFF3} which affect

Frequency f₁ (highest frequency):

$$\begin{aligned} \mathbf{C}_{\mathsf{TOT1}} &= \mathbf{C}_{\mathsf{COIL}} + \mathbf{C}_{\mathsf{ANT1}} + \mathbf{C}_{\mathsf{RFI}} + \mathbf{C}_{\mathsf{OFF2}} + \mathbf{C}_{\mathsf{OFF3}} \\ f_1 &= \frac{1}{2\pi\sqrt{L_{ANT} \cdot C_{TOT1}}} \end{aligned}$$

Frequency f₂ (middle frequency):

$$\begin{split} \mathbf{C}_{\text{TOT2}} &= \mathbf{C}_{\text{COIL}} + \mathbf{C}_{\text{ANT1}} + \mathbf{C}_{\text{ANT2}} + \mathbf{C}_{\text{RFI}} + \mathbf{C}_{\text{OFF3}} \\ f_2 &= \frac{1}{2\pi\sqrt{L_{ANT} \cdot C_{TOT2}}} \end{split}$$

Frequency f₃ (lowest frequency):

$$C_{\text{TOT3}} = C_{\text{COIL}} + C_{\text{ANT1}} + C_{\text{ANT2}} + C_{\text{ANT3}} + C_{\text{RFI}}$$

$$f_3 = \frac{1}{2\pi\sqrt{L_{ANT} \cdot C_{TOT3}}}$$

the resonance frequencies. C_{OFF2} and C_{OFF3} are switch capacitances when switches are open. When switches are closed these capacitances are shorted by on resistance of the switches and they are effectively eliminated. Following relationships can be written for the three tuning frequencies.



TYPICAL APPLICATION

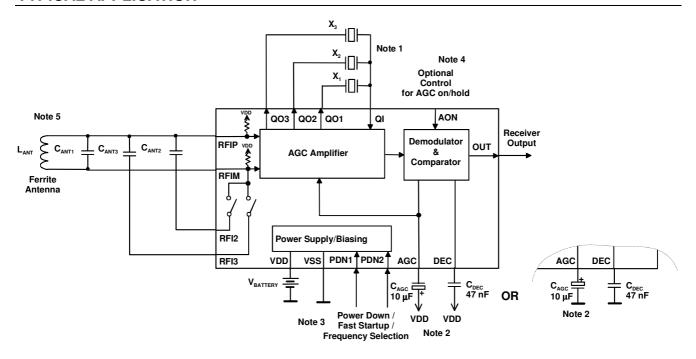


Figure 1. Application circuit of tri band receiver MAS6179.

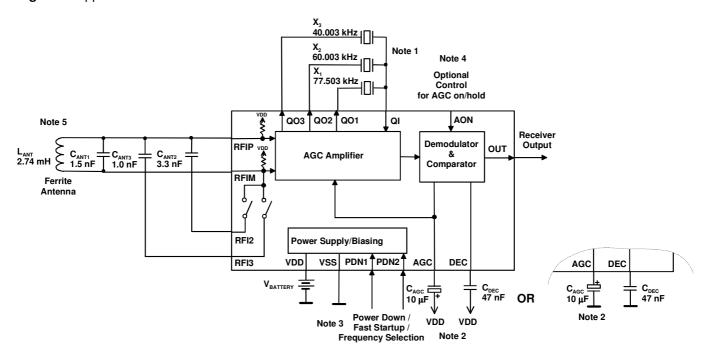


Figure 2. Example circuit of tri band receiver MAS6179 for DCF77/MSF/WWVB/JJY frequencies.



TYPICAL APPLICATION (Continued)

Note 1: Crystals

The crystals as well as ferrite antenna frequencies are chosen according to the time signal system (Table 2). The reason why the crystal frequency is about 3 Hz higher than the signal frequency is that the crystal is operated without the loading capacitor. Without loading capacitor the actual resonance frequency is about 3 Hz lower thus 77.503 kHz crystal resonates at 77.500 kHz when no loading capacitor is used.

Table 2. Time Signal System Frequencies

| Time Signal System | Location | Antenna Frequency | Recommended Crystal Frequency |
|--------------------|----------------|-------------------|-------------------------------|
| DCF77 | Germany | 77.5 kHz | 77.503 kHz |
| HBG | Switzerland | 75 kHz | 75.003 kHz |
| MSF | United Kingdom | 60 kHz | 60.003 kHz |
| WWVB | USA | 60 kHz | 60.003 kHz |
| JJY | Japan | 40 kHz and 60 kHz | 40.003 kHz and 60.003 kHz |
| BPC | China | 68.5 kHz | 68.505 kHz |

The crystal shunt capacitance C_0 should be matched as well as possible with the internal shunt capacitance compensation capacitor C_C of MAS6179. See Compensation Capacitance Options in table 3.

Table 3. Compensation Capacitance Options

| Device | Cc | Crystal Description |
|-----------|---------|---------------------------------|
| MAS6179A1 | 0.75 pF | For low C ₀ crystal |
| MAS6179A2 | 1.3 pF | For high C ₀ crystal |

It should be noted that grounded crystal package has reduced shunt capacitance. This value is about 85% of floating crystal shunt capacitance. For example crystal with 1pF floating package shunt capacitance can have 0.85pF grounded package shunt capacitance. PCB traces of crystal and external compensation capacitance should be kept at minimum to minimize additional parasitic capacitance which can cause capacitance mismatching.

Highest frequency crystal is connected to crystal output pin 1 (QO1). Medium frequency crystal is connected to crystal output pin 2 (QO2). Lowest frequency crystal is connected to crystal output pin 3 (QO3). The other pin of each crystal is connected to common crystal input pin QI.

Table 4 below presents some crystal manufacturers having suitable crystals for time signal receiver application.

Table 4. Crystal Manufacturers and Crystal Types in Alphabetical Order for Time Signal Receiver Application

| Manufacturer | Crystal Type | Dimensions | Web Link |
|---------------|--------------|-------------|--|
| Citizen | CFV-206 | ø 2.0 x 6.0 | http://www.citizen.co.jp/tokuhan/quartz/ |
| Epson | C-2-Type | ø 1.5 x 5.0 | http://www.epsontoyocom.co.jp/english/ |
| | C-4-Type | ø 2.0 x 6.0 | |
| KDS Daishinku | DT-261 | ø 2.0 x 6.0 | http://www.kds.info/index_en.htm |
| Microcrystal | MX1V-L2N | ø 2.0 x 6.0 | http://www.microcrystal.com/ |
| - | MX1V-T1K | ø 2.0 x 8.1 | |
| Seiko | VTC-120 | ø 1.2 x 4.7 | http://www.sii-crystal.com |
| Instruments | | | |

Note 2: AGC Capacitor

The AGC and DEC capacitors must have low leakage currents due to very small signal currents through the capacitors. The insulation resistance of these capacitors should be at minimum 100 M Ω . Also probes with at least 100 M Ω impedance should be used for voltage probing of AGC and DEC pins. Electrolytic AGC capacitor should have voltage rating at least 25 V for low enough leakage. DEC capacitor can be low leakage chip capacitor.



TYPICAL APPLICATION (Continued)

Both the AGC and DEC capacitors can be connected either to VDD or to VSS. To minimize leakage currents during power down the AGC and DEC capacitors are best to be connected to VDD since in power down the AGC and DEC pins go to VDD voltage potential. In this case the positive polarity pin of electrolyte capacitor should be connected to VDD. If the capacitors are connected to VSS then the negative polarity pin of electrolyte capacitor should be connected to VSS.

Note 3: Power Down / Fast Startup Control

Both power down and fast startup are controlled using the PDN pin. The device is in power down (turned off) if PDN1 = PDN2 = VDD and in power up with other three PDN1 and PDN2 control bit combinations (see table 1 on page 4). Fast startup is triggered automatically when moving from power down to power up. The VDD must have been high before moving from power down to power up to guarantee proper operation of fast startup circuitry. Additionally the device should have been kept in power down state at least 50ms before power up. This guarantees that the AGC capacitor voltage has been completely pulled to VDD during power down. The startup time without proper fast startup control can be several minutes. With fast startup it is shortened typically to few seconds.

Note 4: Optional Control for AGC On/Hold

AON control pin has internal pull up which turns AGC circuit on all the time if AON pin is left unconnected. Optionally AON control can be used to hold and release AGC circuit. Stepper motor drive of analog clock or watch can produce disturbing amount of noise which can shift the input amplifier gain to unoptimal level. This can be avoided by controlling AGC hold (AON=VSS) during stepper motor drive periods and releasing AGC (AON=VDD) when motors are not driven. The AGC should be in hold only during disturbances and kept on other time released since due to leakage the AGC can change slowly when in hold.

Note 5: Ferrite Antenna

The ferrite antenna converts the transmitted radio wave into a voltage signal. It has an important role in determining receiver performance. Recommended antenna impedance at resonance is around 150 k Ω .

Low antenna impedance corresponds to low noise but often also to small signal amplitude. On the other hand high antenna impedance corresponds to high noise but also large signal. The optimum performance where signal-to-noise ratio is at maximum is achieved in between.

The antenna should have also some selectivity for rejecting near signal band disturbances. This is determined by the antenna quality factor which should be approximately 100. Much higher quality factor antennas suffer from extensive tuning accuracy requirements and possible tuning drifts by the temperature.

Antenna impedance can be calculated using equation 1 where f_0 , L, Q_{ant} and C are resonance frequency, coil inductance, antenna quality factor and antenna tuning capacitor respectively. Antenna quality factor Q_{ant} is defined by ratio of resonance frequency f_0 and antenna bandwidth B (equation 2).

$$\begin{split} R_{\it antenna} &= 2\pi \cdot f_0 \cdot L \cdot Q_{\it antenna} = \frac{Q_{\it antenna}}{2\pi \cdot f_0 \cdot C} = \frac{1}{2\pi \cdot B \cdot C} \\ Q_{\it antenna} &= \frac{f_0}{B} \end{split} \qquad \qquad \text{Equation 1.}$$

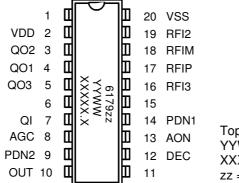
Table 5 below presents some antenna manufacturers for time signal application.

Table 5. Antenna Manufacturers and Antenna Types in Alphaphetical Order for Time Signal Application

| Manufacturer | Antenna Type | Dimensions | Web Link |
|--------------------|-----------------|--------------|-----------------------------------|
| HR Electronic GmbH | 60716 (60kHz) | ø 10 x 60 mm | http://www.hrelectronic.com/ |
| | 60708 (77.5kHz) | | |
| Sumida | ACL80A (40kHz) | ø 10 x 80 mm | www.sumida.co.jp/jeita/XJA021.pdf |



MAS6179 SAMPLES IN SBDIL 20 PACKAGE



Top Marking Definitions: YYWW = Year Week XXXXX.X = Lot Number zz = Sample Version

PIN DESCRIPTION

| Pin Name | Pin | Туре | Function | Note |
|----------|-----|------|--|------|
| | 1 | NC | | |
| VDD | 2 | Р | Positive Power Supply | |
| QO2 | 3 | AO | Quartz Filter Output for Crystal 2 | |
| QO1 | 4 | AO | Quartz Filter Output for Crystal 1 | |
| QO3 | 5 | AO | Quartz Filter Output for Crystal 3 | |
| | 6 | NC | | 1 |
| QI | 7 | Al | Quartz Filter Input for Crystal | |
| AGC | 8 | AO | AGC Capacitor | |
| PDN2 | 9 | DI | Power Down/Frequency Selection Input 2 | 2 |
| OUT | 10 | DO | Receiver Output | 3 |
| | 11 | NC | | |
| DEC | 12 | AO | Demodulator Capacitor | |
| AON | 13 | DI | AGC On Control | |
| PDN1 | 14 | DI | Power Down/Frequency Selection Input 1 | |
| | 15 | NC | | |
| RFI3 | 16 | Al | Receiver Input 3 (for Antenna Capacitor 3) | |
| RFIP | 17 | Al | Positive Receiver Input | 5 |
| RFIM | 18 | Al | Negative Receiver Input 5 | |
| RFI2 | 19 | Al | Receiver Input 2 (for Antenna Capacitor 2) | |
| VSS | 20 | G | Power Supply Ground | |

A = Analog, D = Digital, P = Power, G = Ground, I = Input, O = Output, NC = Not Connected

Notes:

- 1) Pin 6 between QO3 and QI must be connected to VSS to eliminate DIL package lead frame parasitic capacitances disturbing the crystal filter performance. All other NC (Not Connected) pins are also recommended to be connected to VSS to minimize noise coupling.
- 2) PDN1 = PDN2 = VDD means receiver off
 - Fast start-up is triggered when the receiver is after power down controlled to power up
- 3) OUT = VSS when carrier amplitude at maximum; OUT = VDD when carrier amplitude is reduced (modulated)
 - the output is a current source/sink with $|I_{OUT}| > 5 \mu A$
 - at power down the output is pulled to VSS (pull down switch)
- 4) AON = VSS means AGC off (hold current gain level); AON = VDD means AGC on (working)
 - Internal pull-up with current $< 1 \mu A$ which is switched off at power down
- 5) Receiver inputs RFIP and RFIM have both 600 kΩ biasing resistors towards VDD



ORDERING INFORMATION

| Product Code | Product | Description | Capacitance Option |
|---------------|---|--|---------------------------|
| MAS6179A1TC00 | Tri Band AM-Receiver IC with Differential Input | EWS-tested wafer, Thickness 400 μm. | $C_{C} = 0.75 \text{ pF}$ |
| MAS6179A2TC00 | Tri Band AM-Receiver IC with Differential Input | EWS-tested wafer, Thickness 400 µm. | $C_C = 1.3 \text{ pF}$ |

Contact Micro Analog Systems Oy for other wafer thickness options.

| OCAL DISTRIBUTOR | | | | |
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