

### FEATURES

**Low Supply Current: 600  $\mu$ A Max**

**OP07 Type Performance**

**Offset Voltage: 20  $\mu$ V Max**

**Offset Voltage Drift: 0.6  $\mu$ V/ $^{\circ}$ C Max**

**Very Low Bias Current**

**25 $^{\circ}$ C: 100 pA Max**

**-55 $^{\circ}$ C to +125 $^{\circ}$ C: 250 pA Max**

**High Common-Mode Rejection: 114 dB Min**

**Extended Industrial Temperature Range: -40 $^{\circ}$ C to +85 $^{\circ}$ C**

### GENERAL DESCRIPTION

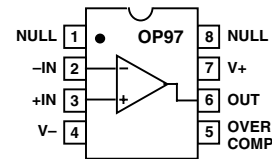
The OP97 is a low power alternative to the industry-standard OP07 precision amplifier. The OP97 maintains the standards of performance set by the OP07 while utilizing only 600  $\mu$ A supply current, less than 1/6 that of an OP07. Offset voltage is an ultralow 25  $\mu$ V, and drift over temperature is below 0.6  $\mu$ V/ $^{\circ}$ C. External offset trimming is not required in the majority of circuits.

Improvements have been made over OP07 specifications in several areas. Notable is bias current, which remains below 250 pA over the full military temperature range. The OP97 is ideal for use in precision long-term integrators or sample-and-hold circuits that must operate at elevated temperatures.

### PIN CONNECTIONS

**8-Lead PDIP (P Suffix)**

**8-Lead SOIC (S Suffix)**



Common-mode rejection and power supply rejection are also improved with the OP97, at 114 dB minimum over wider ranges of common-mode or supply voltage. Outstanding PSR, a supply range specified from  $\pm 2.25$  V to  $\pm 20$  V and the OP97's minimal power requirements combine to make the OP97 a preferred device for portable and battery-powered instruments.

The OP97 conforms to the OP07 pinout, with the null potentiometer connected between Pins 1 and 8 with the wiper to V+. The OP97 will upgrade circuit designs using 725, OP05, OP07, OP12, and 1012 type amplifiers. It may replace 741-type amplifiers in circuits without nulling or where the nulling circuitry has been removed.

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# OP97—SPECIFICATIONS

## ELECTRICAL CHARACTERISTICS (@ $V_S = \pm 15\text{ V}$ , $V_{CM} = 0\text{ V}$ , $T_A = 25^\circ\text{C}$ , unless otherwise noted.)

Parameter	Symbol	Conditions	OP97E			OP97F			Unit
			Min	Typ	Max	Min	Typ	Max	
Input Offset Voltage	$V_{OS}$			10	25		30	75	$\mu\text{V}$
Long-Term Offset Voltage Stability	$\Delta V_{OS}/\text{Time}$			0.3			0.3		$\mu\text{V}/\text{Month}$
Input Offset Current	$I_{OS}$			30	100		30	150	pA
Input Bias Current	$I_B$			$\pm 30$	$\pm 100$		$\pm 30$	$\pm 150$	pA
Input Noise Voltage	$e_n$ p-p	0.1 Hz to 10 Hz		0.5			0.5		$\mu\text{V p-p}$
Input Noise Voltage Density	$e_n$	$f_O = 10\text{ Hz}^1$		17	30		17	30	$\text{nV}/\sqrt{\text{Hz}}$
		$f_O = 1000\text{ Hz}^2$		14	22		14	22	$\text{nV}/\sqrt{\text{Hz}}$
Input Noise Current Density	$i_n$	$f_O = 10\text{ Hz}$		20			20		$\text{fA}/\sqrt{\text{Hz}}$
Large-Signal Voltage Gain	$A_{VO}$	$V_O = \pm 10\text{ V}$ ; $R_L = 2\text{ k}\Omega$	300	2000		200	2000		V/mV
Common-Mode Rejection	CMR	$V_{CM} = \pm 13.5\text{ V}$	114	132		110	132		dB
Power-Supply Rejection	PSR	$V_S = \pm 2\text{ V to } \pm 20\text{ V}$	114	132		110	132		dB
Input Voltage Range <sup>3</sup>	IVR		$\pm 13.5$	$\pm 14.0$		$\pm 13.5$	$\pm 14.0$		V
Output Voltage Swing	$V_O$	$R_L = 10\text{ k}\Omega$	$\pm 13$	$\pm 14$		$\pm 13$	$\pm 14$		V
Slew Rate	SR		0.1	0.2		0.1	0.2		V/ $\mu\text{s}$
Differential Input Resistance <sup>4</sup>	$R_{IN}$		30			30			M $\Omega$
Closed-Loop Bandwidth	BW	$A_{VCL} = 1$	0.4	0.9		0.4	0.9		MHz
Supply Current	$I_{SY}$			380	600		380	600	$\mu\text{A}$
Supply Voltage	$V_S$	Operating Range	$\pm 2$	$\pm 15$	$\pm 20$	$\pm 2$	$\pm 15$	$\pm 20$	V

### NOTES

<sup>1</sup>10 Hz noise voltage density is sample tested. Devices 100% tested for noise are available on request.

<sup>2</sup>Sample tested.

<sup>3</sup>Guaranteed by CMR test.

<sup>4</sup>Guaranteed by design.

Specifications subject to change without notice.

## ELECTRICAL CHARACTERISTICS (@ $V_S = \pm 15\text{ V}$ , $V_{CM} = 0\text{ V}$ , $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ for the OP97E/F, unless otherwise noted.)

Parameter	Symbol	Conditions	OP97E			OP97F			Unit
			Min	Typ	Max	Min	Typ	Max	
Input Offset Voltage	$V_{OS}$			25	60		60	200	$\mu\text{V}$
Average Temperature Coefficient of $V_{OS}$	$TCV_{OS}$	S-Package		0.2	0.6		0.3	2.0	$\mu\text{V}/^\circ\text{C}$
Input Offset Current	$I_{OS}$			60	250		80	750	pA
Average Temperature Coefficient of $I_{OS}$	$TCI_{OS}$			0.4	2.5		0.6	7.5	$\text{pA}/^\circ\text{C}$
Input Bias Current	$I_B$			$\pm 60$	$\pm 250$		$\pm 80$	$\pm 750$	pA
Average Temperature Coefficient of $I_B$	$TCI_B$			0.4	2.5		0.6	7.5	$\text{pA}/^\circ\text{C}$
Large Signal Voltage Gain	$A_{VO}$	$V_O = 10\text{ V}$ ; $R_L = 2\text{ k}\Omega$	200	1000		150	1000		V/mV
Common-Mode Rejection	CMR	$V_{CM} = \pm 13.5\text{ V}$	108	128		108	128		dB
Power Supply Rejection	PSR	$V_S = \pm 2.5\text{ V to } \pm 20\text{ V}$	108	126		108	128		dB
Input Voltage Range*	IVR		$\pm 13.5$	$\pm 14.0$		$\pm 13.5$	$\pm 14.0$		V
Output Voltage Swing	$V_O$	$R_L = 10\text{ k}\Omega$	$\pm 13$	$\pm 14$		$\pm 13$	$\pm 14$		V
Slew Rate	SR		0.05	0.15		0.05	0.15		V/ $\mu\text{s}$
Supply Current	$I_{SY}$			400	800		400	800	$\mu\text{A}$
Supply Voltage	$V_S$	Operating Range	$\pm 2.5$	$\pm 15$	$\pm 20$	$\pm 2.5$	$\pm 15$	$\pm 20$	V

\*Guaranteed by CMR test.

Specifications subject to change without notice.

**ABSOLUTE MAXIMUM RATINGS<sup>1</sup>**

Supply Voltage	±20 V
Input Voltage <sup>2</sup>	±20 V
Differential Input Voltage <sup>3</sup>	±1 V
Differential Input Current <sup>3</sup>	±10 mA
Output Short-Circuit Duration	Indefinite
Operating Temperature Range	
OP97E, OP97F (P, S)	−40°C to +85°C
Storage Temperature Range	−65°C to +150°C
Junction Temperature Range	−65°C to +150°C
Lead Temperature (Soldering, 60 sec)	300°C

Package Type	$\theta_{JA}$ <sup>4</sup>	$\theta_{JC}$	Unit
8-Lead PDIP (P)	103	43	°C/W
8-Lead SOIC (S)	158	43	°C/W

**NOTES**

<sup>1</sup>Absolute maximum ratings apply to both DICE and packaged parts, unless otherwise noted.

<sup>2</sup>For supply voltages less than ±20 V, the absolute maximum input voltage is equal to the supply voltage.

<sup>3</sup>The OP97's inputs are protected by back-to-back diodes. Current-limiting resistors are not used in order to achieve low noise. Differential input voltages greater than 1 V will cause excessive current to flow through the input protection diodes unless limiting resistance is used.

<sup>4</sup> $\theta_{JA}$  is specified for worst-case mounting conditions, i.e.,  $\theta_{JA}$  is specified for device in socket for PDIP package;  $\theta_{JA}$  is specified for device soldered to printed circuit board for SOIC package.

**ORDERING GUIDE**

Model	Temperature Range	Package Description	Package Option*
OP97EP	−40°C to +85°C	8-Lead PDIP	N-8
OP97FP	−40°C to +85°C	8-Lead PDIP	N-8
OP97FS	−40°C to +85°C	8-Lead SOIC	R-8
OP97FS-REEL	−40°C to +85°C	8-Lead SOIC	R-8
OP97FS-REEL7	−40°C to +85°C	8-Lead SOIC	R-8

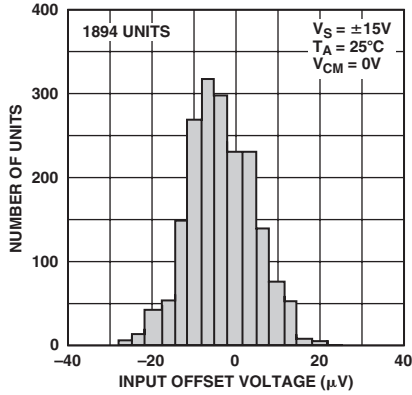
\*For outline information, see Package Information section.

**CAUTION**

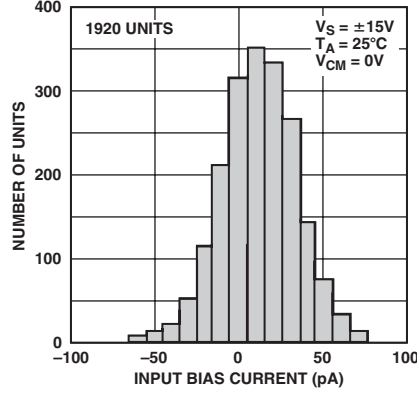
ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readily accumulate on the human body and test equipment and can discharge without detection. Although the OP97 features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.



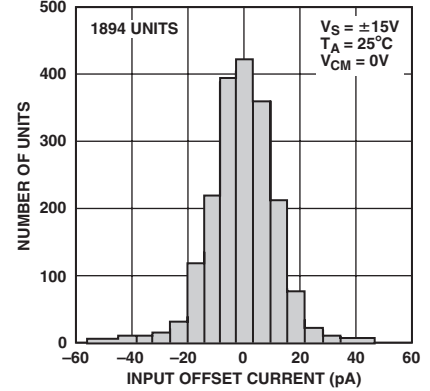
# OP97—Typical Performance Characteristics



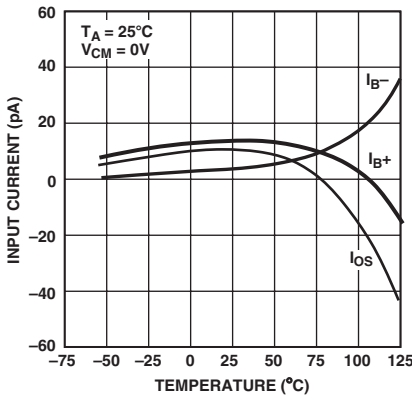
TPC 1. Typical Distribution of Input Offset Voltage



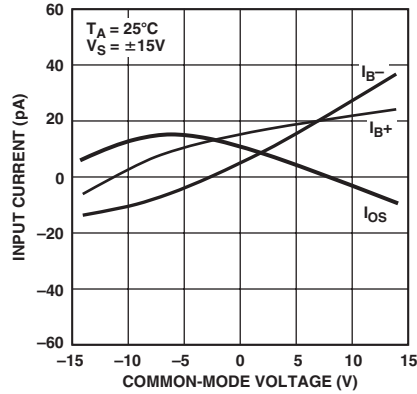
TPC 2. Typical Distribution of Input Bias Current



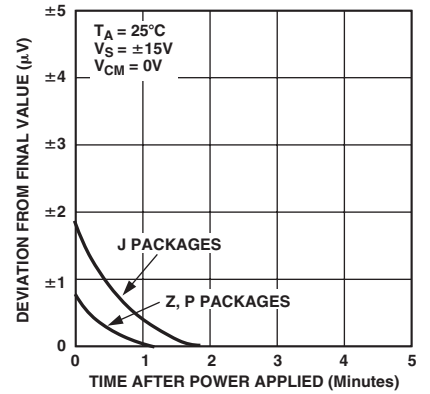
TPC 3. Typical Distribution of Input Offset Current



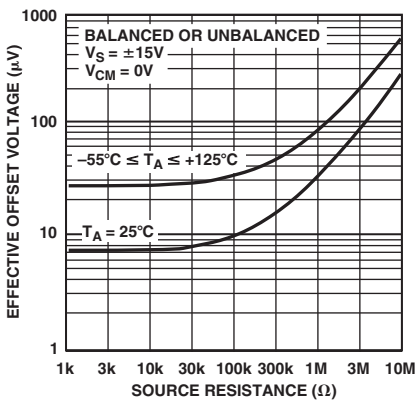
TPC 4. Input Bias, Offset Current vs. Temperature



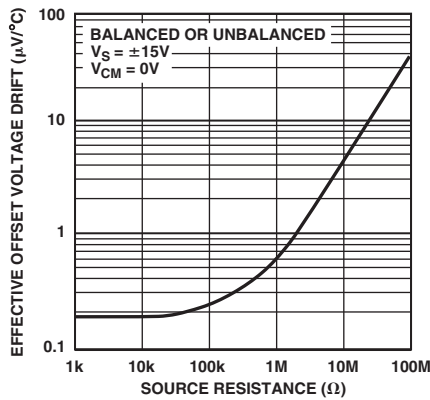
TPC 5. Input Bias, Offset Current vs. Common-Mode Voltage



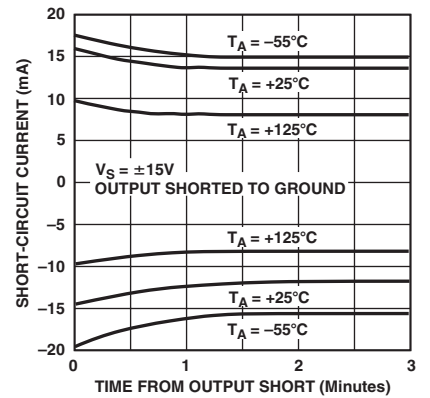
TPC 6. Input Offset Voltage Warm-Up Drift



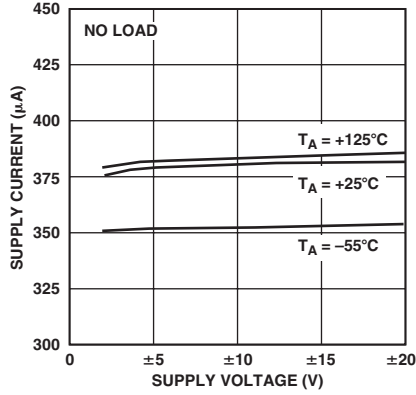
TPC 7. Effective Offset Voltage vs. Source Resistance



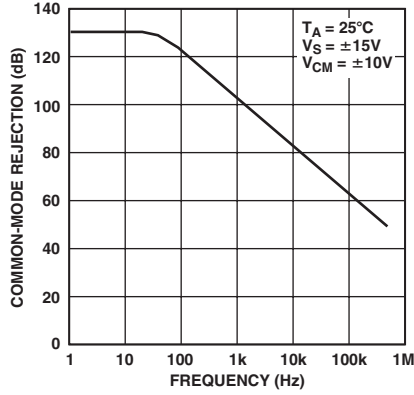
TPC 8. Effective  $TC_{V_{OS}}$  vs. Source Resistance



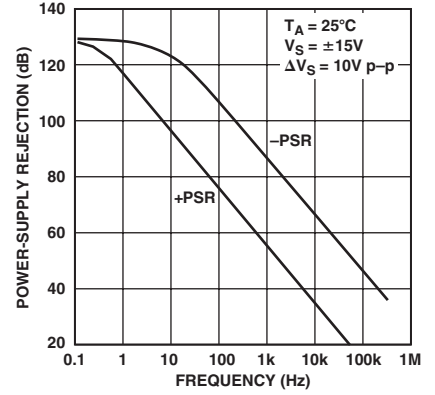
TPC 9. Short-Circuit Current vs. Time, Temperature



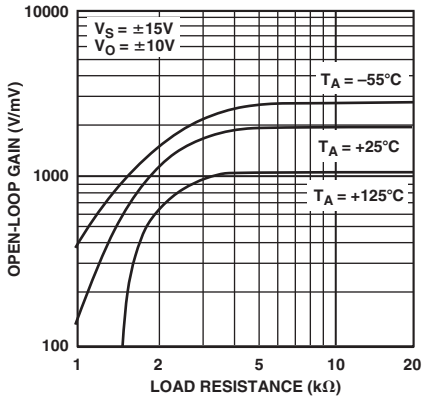
TPC 10. Supply Current vs. Supply Voltage



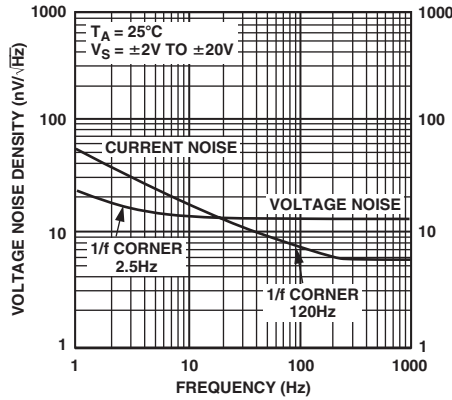
TPC 11. Common-Mode Rejection vs. Frequency



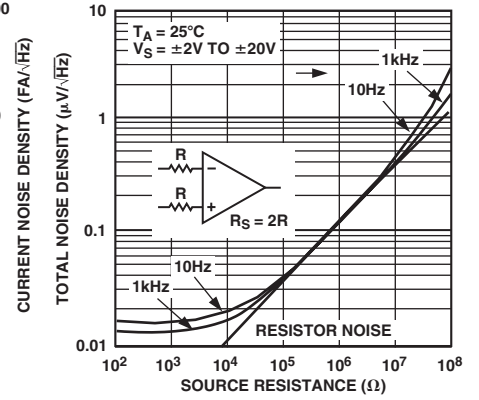
TPC 12. Power-Supply Rejection vs. Frequency



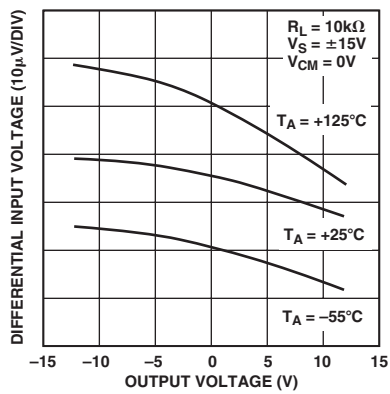
TPC 13. Open-Loop Gain vs. Load Resistance



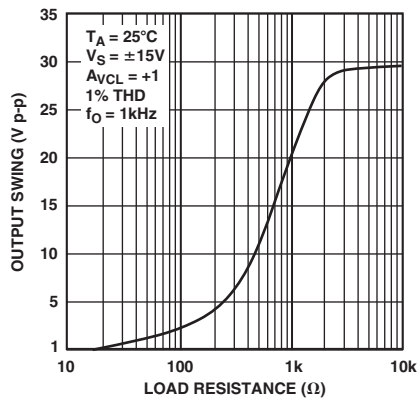
TPC 14. Noise Density vs. Frequency



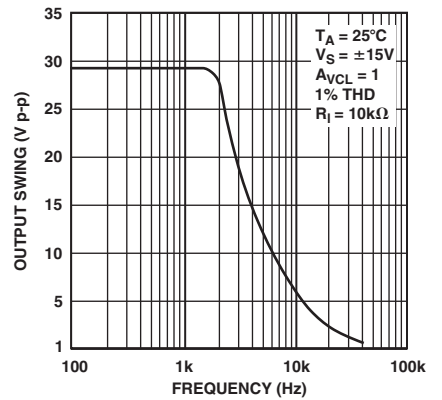
TPC 15. Total Noise Density vs. Source Resistance



TPC 16. Open-Loop Gain Linearity

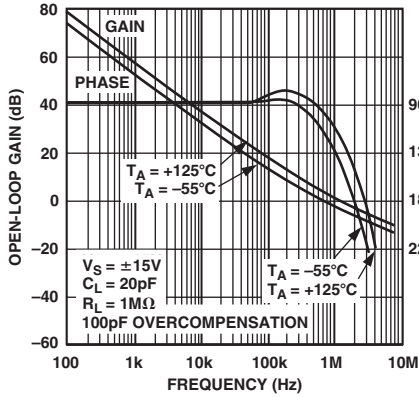


TPC 17. Maximum Output Swing vs. Load Resistance

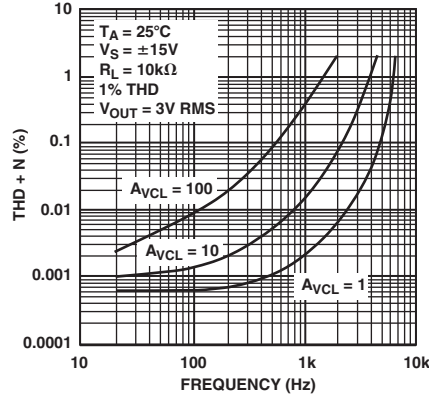


TPC 18. Maximum Output Swing vs. Frequency

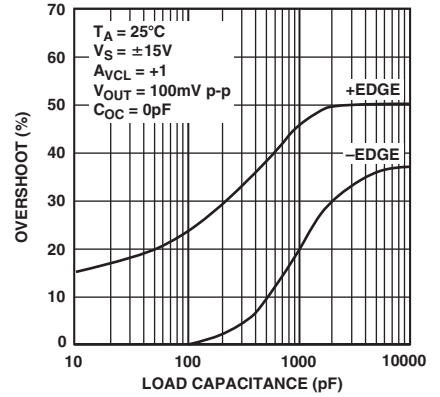
# OP97



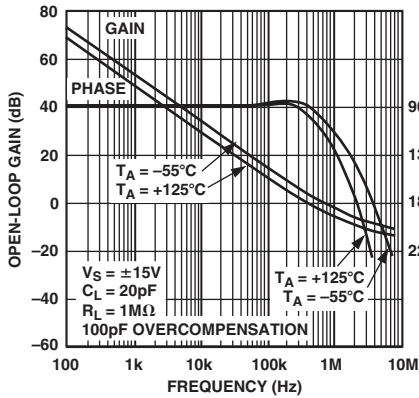
TPC 19. Open-Loop Gain, Phase vs. Frequency ( $C_{OC} = 0 \text{ pF}$ )



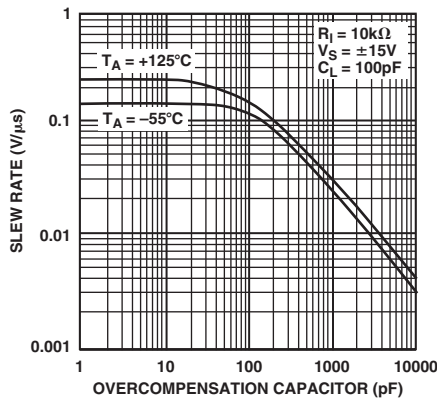
TPC 20. Total Harmonic Distortion Plus Noise vs. Frequency



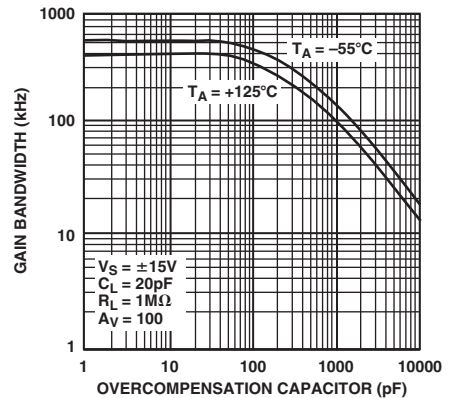
TPC 21. Small Signal Overshoot vs. Capacitive Load



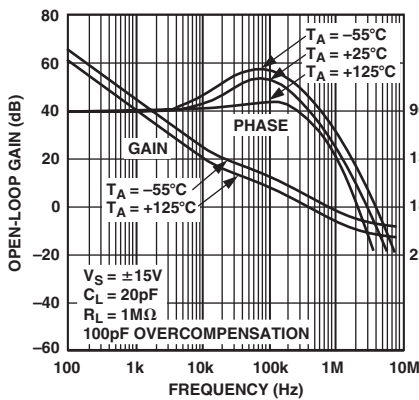
TPC 22. Open-Loop Gain, Phase vs. Frequency ( $C_{OC} = 100 \text{ pF}$ )



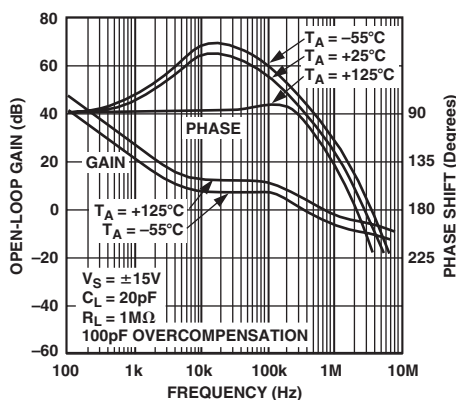
TPC 23. Slew Rate vs. Overcompensation



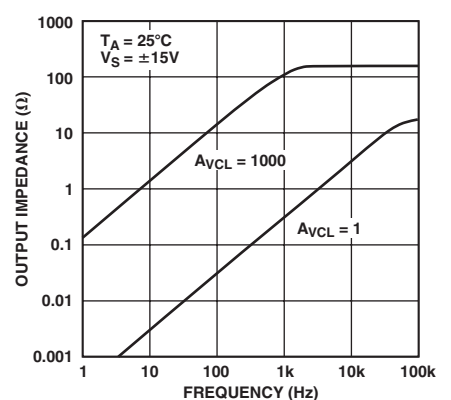
TPC 24. Gain Bandwidth Product vs. Overcompensation



TPC 25. Open-Loop Gain, Phase vs. Frequency ( $C_{OC} = 1000 \text{ pF}$ )



TPC 26. Open-Loop Gain, Phase vs. Frequency ( $C_{OC} = 10,000 \text{ pF}$ )



TPC 27. Closed-Loop Output Resistance vs. Frequency

## APPLICATION INFORMATION

The OP97 is a low power alternative to the industry-standard precision op amp, the OP07. The OP97 may be substituted directly into OP07, OP77, 725, 112/312, and 1012 sockets with improved performance and/or less power dissipation and may be inserted into sockets conforming to the 741 pinout if nulling circuitry is not used. Generally, nulling circuitry used with earlier generation amplifiers is rendered superfluous by the OP97's extremely low offset voltage and may be removed without compromising circuit performance.

Extremely low bias current over the full military temperature range makes the OP97 attractive for use in sample-and-hold amplifiers, peak detectors, and log amplifiers that must operate over a wide temperature range. Balancing input resistances is not necessary with the OP97. Offset voltage and  $TCV_{OS}$  are degraded only minimally by high source resistance, even when unbalanced.

The input pins of the OP97 are protected against large differential voltage by back-to-back diodes. Current-limiting resistors are not used so that low noise performance is maintained. If differential voltages above  $\pm 1$  V are expected at the inputs, series resistors must be used to limit the current flow to a maximum of 10 mA. Common-mode voltages at the inputs are not restricted and may vary over the full range of the supply voltages used.

The OP97 requires very little operating headroom about the supply rails and is specified for operation with supplies as low

as  $\pm 2$  V. Typically, the common-mode range extends to within 1 V of either rail. The output typically swings to within 1 V of the rails when using a 10 k $\Omega$  load.

Offset nulling is achieved utilizing the same circuitry as an OP07. A potentiometer between 5 k $\Omega$  and 100 k $\Omega$  is connected between Pins 1 and 8 with the wiper connected to the positive supply. The trim range is between 300  $\mu$ V and 850  $\mu$ V, depending upon the internal trimming of the device.

## AC PERFORMANCE

The OP97's ac characteristics are highly stable over its full operating temperature range. Unity-gain small-signal response is shown in Figure 2. Extremely tolerant of capacitive loading on the output, the OP97 displays excellent response even with 1000 pF loads (Figure 3). In large-signal applications, the input protection diodes effectively short the input to the output during the transients if the amplifier is connected in the usual unity-gain configuration. The output enters short-circuit current limit, with the flow going through the protection diodes. Improved large-signal transient response is obtained by using a feedback resistor between the output and the inverting input. Figure 4 shows the large-signal response of the OP97 in unity gain with a 10 k $\Omega$  feedback resistor. The unity-gain follower circuit is shown in Figure 5.

The overcompensation pin may be used to increase the phase margin of the OP97 or to decrease gain-bandwidth product at gains greater than 10.

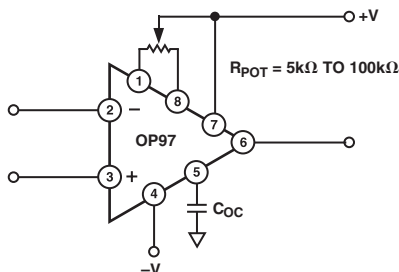


Figure 1. Optional Input Offset Voltage Nulling and Overcompensation Circuits

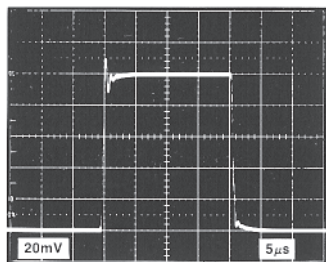


Figure 2. Small-Signal Transient Response ( $C_{LOAD} = 100$  pF,  $A_{VCL} = 1$ )

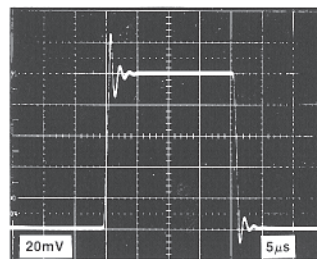


Figure 3. Small-Signal Transient Response ( $C_{LOAD} = 1000$  pF,  $A_{VCL} = 1$ )

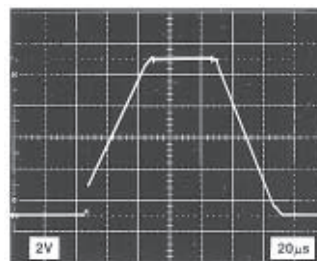


Figure 4. Large-Signal Transient Response ( $A_{VCL} = 1$ )



# OP97

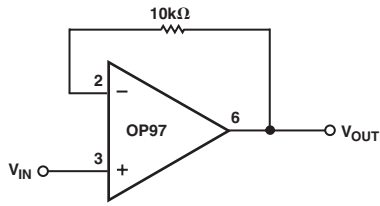


Figure 5. Unity-Gain Follower

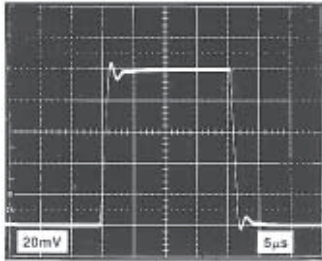


Figure 6. Small-Signal Transient Response with Over-compensation ( $C_{LOAD} = 1000 \text{ pF}$ ,  $A_{VCL} = 1$ ,  $C_{OC} = 220 \text{ pF}$ )

## GUARDING AND SHIELDING

To maintain the extremely high input impedances of the OP97, care must be taken in circuit board layout and manufacturing. Board surfaces must be kept scrupulously clean and free of moisture. Conformal coating is recommended to provide a humidity barrier. Even a clean PC board can have 100 pA of leakage currents between adjacent traces, so that guard rings should be used around the inputs. Guard traces are operated at a voltage close to that on the inputs, so that leakage currents become minimal. In noninverting applications, the guard ring should be connected to the common-mode voltage at the inverting input (Pin 2). In inverting applications, both inputs remain at ground, so that the guard trace should be grounded. Guard traces should be made on both sides of the circuit board.

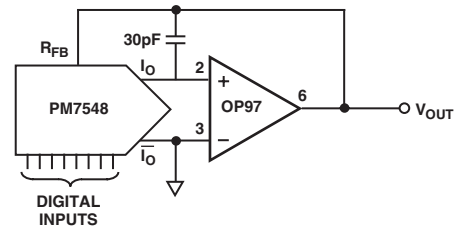


Figure 7. DAC Output Amplifier

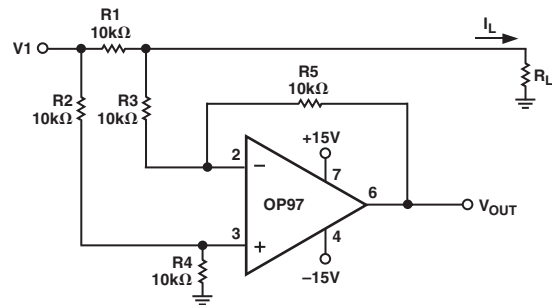


Figure 8. Current Monitor

High impedance circuitry is extremely susceptible to RF pickup, line frequency hum, and radiated noise from switching power supplies. Enclosing sensitive analog sections within grounded shields is generally necessary to prevent excessive noise pickup. Twisted-pair cable will aid in rejection of line frequency hum.

The OP97 is an excellent choice as an output amplifier for higher resolution CMOS DACs. Its tightly trimmed offset voltage and minimal bias current result in virtually no degradation of linearity, even over wide temperature ranges.

Figure 8 shows a versatile monitor circuit that can typically sense current at any point between the  $\pm 15 \text{ V}$  supplies. This makes it ideal for sensing current in applications such as full bridge drivers where bidirectional current is associated with large common-mode voltage changes. The 114 dB CMRR of the OP97 makes the amplifier's contribution to common-mode error negligible, leaving only the error due to the resistor ratio inequality. Ideally,  $R2/R4 = R3/R5$ .

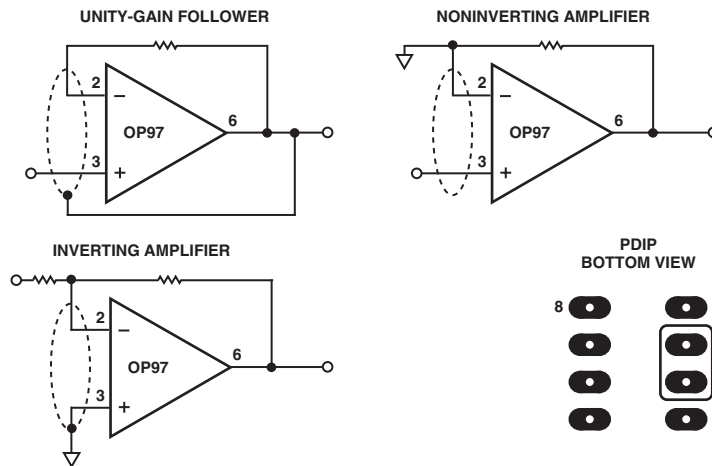


Figure 9. Guard Ring Layout and Connections



The digitally programmable gain amplifier shown in Figure 10 has 12-bit gain resolution with 10-bit gain linearity over the range of -1 to -1024. The low bias current of the OP97 maintains this linearity, while C1 limits the noise voltage bandwidth allowing accurate measurement down to microvolt levels.

DIGITAL IN	GAIN (Av)
4095	-1.00024
2048	-2
1024	-4
512	-8
256	-16
128	-32
64	-64
32	-128
16	-256
8	-512
4	-1024
2	-2048
1	-4096
0	Open Loop

Many high speed amplifiers suffer from less-than-perfect low frequency performance. A combination amplifier consisting of a high precision, slow device like the OP97 and a faster device such as the OP44 results in uniformly accurate performance from dc to the high frequency limit of the OP44, which has a gain-bandwidth product of 23 MHz. The circuit shown in Figure 11 accomplishes this, with the OP44 providing high frequency amplification and the OP97 operating on low frequency signals and providing offset correction. Offset voltage and drift of the circuit are controlled by the OP97.

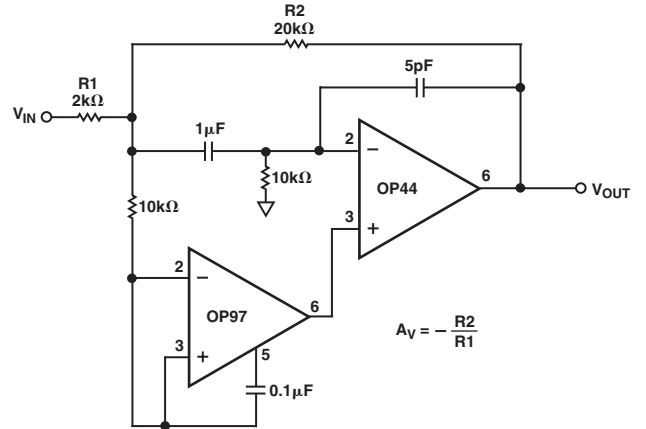


Figure 11. Combination High-Speed, Precision Amplifier

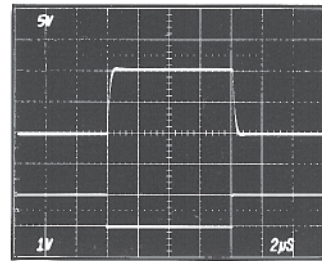


Figure 12. Combination Amplifier Transient Response

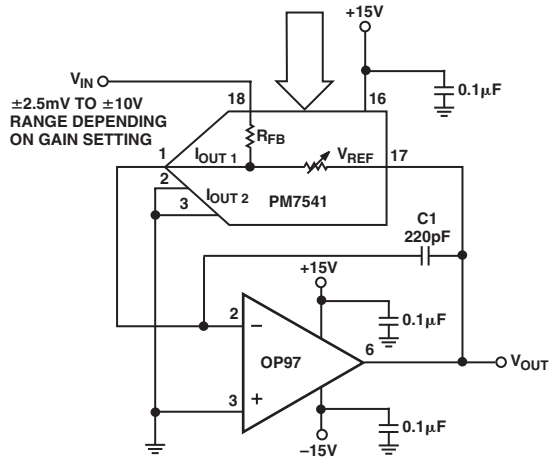
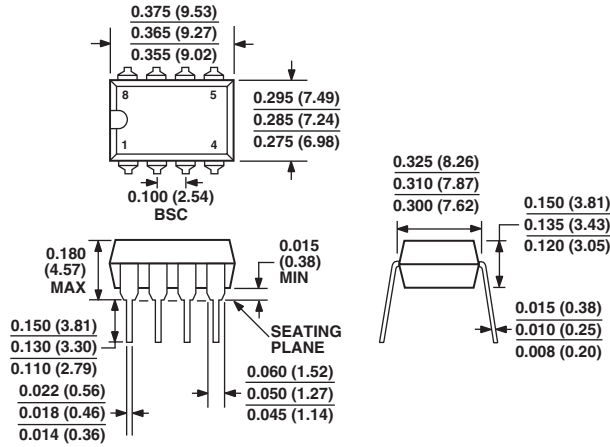


Figure 10. Precision Programmable Gain Amplifier

OUTLINE DIMENSIONS

8-Lead Plastic Dual In-Line Package [PDIP]  
 P-Suffix  
 (N-8)

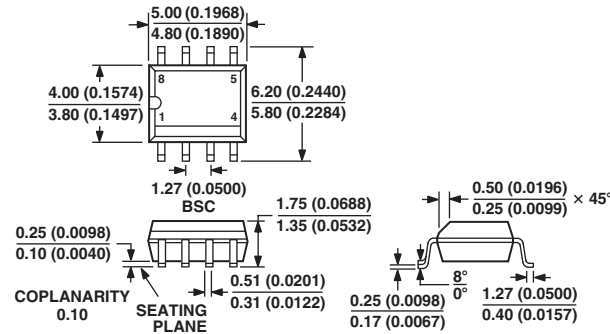
Dimensions shown in inches and (millimeters)



COMPLIANT TO JEDEC STANDARDS MO-095AA  
 CONTROLLING DIMENSIONS ARE IN INCHES; MILLIMETER DIMENSIONS  
 (IN PARENTHESES) ARE ROUNDED-OFF INCH EQUIVALENTS FOR  
 REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN

8-Lead Standard Small Outline Package [SOIC]  
 Narrow Body  
 S-Suffix  
 (R-8)

Dimensions shown in millimeters and (inches)



COMPLIANT TO JEDEC STANDARDS MS-012AA  
 CONTROLLING DIMENSIONS ARE IN MILLIMETERS; INCH DIMENSIONS  
 (IN PARENTHESES) ARE ROUNDED-OFF MILLIMETER EQUIVALENTS FOR  
 REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN

## Revision History

<b>Location</b>	<b>Page</b>
<b>7/03—Data Sheet changed from REV. D to REV. E.</b>	
Deleted H-08A .....	Universal
Deleted Q-8 .....	Universal
Deleted E-20A .....	Universal
Deleted DIE CHARACTERISTICS .....	4
Deleted WAFER TEST LIMITS .....	4
Updated TPC 14 .....	5
Updated OUTLINE DIMENSIONS .....	10
<b>1/02—Data Sheet changed from REV. C to REV. D.</b>	
Edits to Absolute Maximum Ratings .....	3
Edits to Ordering Guide .....	3
Deleted DICE CHARACTERISTICS .....	3
Deleted WAFER TEST LIMITS .....	3
Edits to APPLICATION INFORMATION .....	7



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