

CLC030 SMPTE 292M/259M Digital Video Serializer with Video and Ancilliary Data FIFOs and Integrated Cable Driver General Description The CLC030's internal circuitry is powered from +2.5V

The CLC030 SMPTE 292M/259M Digital Video Serializer with Ancilliary Data FIFO and Integrated Cable Driver is a monolithic integrated circuit that encodes, serializes and transmits bit-parallel digital video data conforming to SMPTE 125M and 267M standard definition, 10-bit wide component video and SMPTE 260M, 274M, 295M and 296M high-definition, 20-bit wide component video standards. The CLC030 operates at SMPTE 259M serial data rates of 270 Mbps, 360 Mbps, the SMPTE 344M (proposed) serial data rate of 540 Mbps; and the SMPTE 292M serial data rate of 1.485 Gbps. The serial data clock frequency is internally generated and requires no external frequency setting, trimming or filtering components*.

Functions performed by the CLC030 include: parallel-toserial data conversion, SMPTE standard data encoding, NRZ to NRZI data format conversion, serial data clock generation and encoding with the serial data, automatic video rate and format detection, ancilliary data packet storage manipulation and insertion, and serial data output driving. The CLC030 has circuitry for automatic EDH/CRC character and flag generation and insertion per SMPTE RP-165 (standard definition) or SMPTE 292M (high definition). LSB dithering is implemented according to the proposed SMPTE recommended practice. Unique to the CLC030 are its video and ancilliary data FIFOs. The video FIFO allows from 0 to 7 parallel data clock delays to be inserted in the data path for video timing purposes. The ancilliary data port and on-chip FIFO and control circuitry offer elegant handling and insertion of ancilliary data packets and checksums in the ancilliary data space. The CLC030 also has an exclusive built-in selftest (BIST) and video test pattern generator (TPG) with SD and HD component video test patterns: reference black, PLL and EQ pathologicals and colour bars in 4:3 and 16:9 raster formats for NTSC and PAL standards*. The colour bar patterns feature optional bandwidth limiting coding between the chroma and luma transitions.

The CLC030 has a unique multi-function I/O port which provides access to control and configuration signals and data. This port may be programmed to provide external access to control functions and data for use as inputs and outputs. This allows the designer greater flexibility in tailoring the CLC030 to the desired application. At power-up or after a reset command, the CLC030 is auto-configured to a default operating condition. Separate power pins for the output driver, PLL and the serializer improve power supply rejection, output jitter and noise performance. The CLC030's internal circuitry is powered from +2.5V and the I/O circuitry from a +3.3V supply. Power dissipation is typically 575 mW at 1.485Gbps including two 75 Ω AC-coupled and back-matched output loads. The device is packaged in a 64-pin TQFP.

Features

- SDTV/HDTV serial digital video standard compliant
- Supports 270 Mbps, 360 Mbps, 540 Mbps and 1.485 Gbps serial video data rates with auto-detection*
- LSB dithering option
- No external serial data rate setting or VCO filtering components required*
- Fast VCO lock time: < TBD µs at 1.485 Gbps
- Built-in self-test (BIST) and video test pattern generator (TPG)*
- Automatic EDH/CRC word and flag generation and insertion
- On-chip ancilliary data FIFO and insertion control circuitry
- Flexible control and configuration I/O port
- LVCMOS compatible data and control inputs and outputs
- 75Ω ECL-compatible, differential, serial cable-driver outputs
- 3.3V I/O power supply, 2.5V logic power supply operation
- Low power: typically 575mW
- 64-pin TQFP package
- Commercial temperature range 0°C to +70°C
- * Patent applications made or pending.

Applications

- SDTV/HDTV parallel-to-serial digital video interfaces for: — Video cameras
 - VTRs
 - Telecines
 - Digital video routers and switchers
 - Digital video processing and editing equipment
 - Video test pattern generators and digital video test equipment
 - Video signal generators

Typical Application



DS200003-1





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Absolute Maximum Ratings (Note 1) It is anticipated that this device will not be offered in

It is anticipated that this device will not be offered in a military qualified version. If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office / Distributors for availability and specifications.

CMOS I/O Supply Voltage	4.0V
SDO Supply Voltage	
(V _{DDSD} -V _{SSSD}):	4.0V
Digital Logic Supply Voltage	
(V _{DDD} -V _{SSD}):	3.0V
PLL Supply Voltage	
(V _{ddpll} –V _{sspll}):	3.0V
CMOS Input Voltage (Vi):	V _{SSIO} –0.15V to V _{DDIO} +0.15V
CMOS Output Voltage (Vo):	V _{SSIO} –0.15V to V _{DDIO} +0.15V
CMOS Input Current (single input):	
$Vi = V_{SSIO} - 0.15V:$	–5 mA
$Vi = V_{DDIO} + 0.15V:$	+5 mA
CMOS Output Source/Sink Current:	±10 mA
SDO Output Sink Current:	40 mA

Package Thermal Resistance $\theta_{\mathsf{JA}} @$ 0 LFM Airflow TBD°C/W $\theta_{\mathsf{JA}} @$ 500 LFM Airflow TBD°C/W TBD°C/W θ_{JC} -65°C to +150°C Storage Temp. Range: Junction Temperature: +150°C Lead Temperature (Soldering 4 Sec): +260°C ESD Rating (HBM): 2 kV ESD Rating (MM): 300V Transistor Count: TBD

Recommended Operating Conditions

3.3V ±150 mV
3.3V ±150 mV
2.5V ±125 mV
2.5V ±125 mV
V _{SSIO} to V _{DDIO} 0°C to +70°C

DC Electrical Characteristics

Over Supply Voltage and Operating Temperature ranges, unless otherwise specified (Notes 2, 3).

Symbol	Parameter	Conditions	Reference	Min	Тур	Max	Units
V _{IH}	Input Voltage High Level		All LVCMOS	2.0		V _{DDIO}	V
V _{IL}	Input Voltage Low Level		Inputs	V _{SSIO}		0.8	V
I _{IH}	Input Current High Level	$V_{IH} = V_{DDIO}$			+40	+60	μA
I _{IL}	Input Current Low Level	$V_{IL} = V_{SSIO}$			-1	-20	μA
V _{OH}	CMOS Output Voltage High Level	I _{OH} = -10 mA	All LVCMOS Outputs	2.4	2.7	V _{DDIO}	V
V _{OL}	CMOS Output Voltage Low Level	I _{OL} = +10 mA		V _{SSIO}	V _{SSIO} +0.3	V _{SSIO} +0.5V	V
V_{SDO}	Serial Driver Output Voltage	Test Circuit, Test Loads Shall Apply	SDO, SDO	720	800	880	mV _{P-P}
I _{DD} (3.3V)	Power Supply Current, 3.3V Supply, Total	V _{CLK} = 27 MHz, NTSC Colour Bar Pattern, Test Circuit, Test Loads Shall Apply	V _{ddio} , V _{ddsd}		30	TBD	mA
I _{DD} (3.3V)	Power Supply Current, 3.3V Supply, Total	V _{CLK} = 74.25 MHz, NTSC Colour Bar Pattern, Test Circuit, Test Loads Shall Apply	V _{ddio} , V _{ddsd}		80	TBD	mA
I _{DD} (2.5V)	Power Supply Current, 2.5V Supply, Total	V _{CLK} = 27 MHz, NTSC Colour Bar Pattern, Test Circuit, Test Loads Shall Apply	V _{ddd} , V _{ddz} , V _{ddpll}		50	TBD	mA
I _{DD} (2.5V)	Power Supply Current, 2.5V Supply, Total	V_{CLK} = 74.25 MHz, NTSC Colour Bar Pattern, Test Circuit, Test Loads Shall Apply	V _{ddd} , V _{ddz} , V _{ddpll}		125	TBD	mA

AC Electrical Characteristics

Symbol	Parameter	Conditions	Reference	Min	Тур	Max	Units
f _{VCLK}	Parallel Video Clock Frequency	(Note 11)	V _{CLK}	27		74.25	MHz
DC_V	Video Clock Duty Cycle	(Note 11) V _{CLK}		45	50	55	%
t _{JIT}	Video Clock Jitter	(Notes 8, 11)	V _{CLK}		100		ps _{P-P}
f _{ACLK}	Ancilliary Clock Frequency	(Note 11)	A _{CLK}	1 x 10 ⁻³		V _{CLK}	MHz
DC _A	Ancilliary Clock Duty Cycle	(Note 11)	A _{CLK}	45	50	55	%
t _r , t _f	Input Clock and Data Rise Time, Fall Time	10%–90%, (Note 11)	V _{CLK} , A _{CLK} , DV _N , AD _N	1.0	1.5	3.0	ns
BR _{SDO}	Serial Data Rate	(Notes 5, 6)	SDO, SDO	270		1,485	M _{BPS}
t _r , t _f	Rise Time, Fall Time	20%-80%, (Note 6)	SDO, SDO			270	ps
	Output Overshoot	(Note 4)	SDO, SDO		5		%
t _j	Serial Output Jitter, Intrinsic	270 M _{BPS} , (Notes 5, 9, 10)	SDO, SDO		0.1		UI _{P-P}
t _j	Serial Output Jitter, Intrinsic	1,485 M _{BPS} , (Notes 6, 9, 10)	SDO, SDO		0.1		UI _{P-P}
t _{LOCK}	Lock Time	(Notes 5, 7) (SD Rates)				500	μs
t _{LOCK}	Lock Time	(Notes 6, 7) (HD Rates)				500	μs
t _S	Setup Time, Video Data	Timing Diagram, (Note 11)	DV_N to V_CLK		1.5	2.0	ns
t _H	Hold Time, Video Data	Timing Diagram, (Note 11)	V_{CLK} to DV_N		1.5	2.0	ns
t _S	Setup Time, Anc. Data Port	Timing Diagram, (Note 11)	AD_N to A_{CLK}		1.5	2.0	ns
t _H	Hold Time, Anc. Data	Timing Diagram, (Note 11)	A _{CLK} to AD _N		1.5	2.0	ns

Note 1: "Absolute Maximum Ratings" are those parameter values beyond which the life and operation of the device cannot be guaranteed. The stating herein of these maximums shall not be construed to imply that the device can or should be operated at or beyond these values. The table of "Electrical Characteristics" specifies acceptable device operating conditions.

Note 2: Current flow into device pins is defined as positive. Current flow out of device pins is defined as negative. All voltages are referenced to V_{SS} = 0V.

Note 3: Typical values are stated for V_{DDIO} = V_{DDSD} = +3.3V, V_{DDD} = V_{DDPLL} = +2.5V and T_A = +25°C.

Note 4: Spec. is guaranteed by design.

Note 5: R_L = 75 Ω , AC-coupled @ 270 M_{BPS}, R_{REF}LVL = R_{REF}PRE = 4.75 k Ω 1%, See Test Loads and Test Circuit.

Note 6: $R_L = 75\Omega$, AC-coupled @ 1,485 M_{BPS}, $R_{REF}LVL = R_{REF}PRE = 4.75 k\Omega$ 1%, See Test Loads and Test Circuit.

Note 7: Measured from rising-edge of first DV_{CLK} cycle until Lock Detect output goes high (true).

Note 8: Average value measured between rising edges computed over at least one video field.

Note 9: Intrinsic timing jitter is measured in accordance with SMPTE RP 184-1996, SMPTE RP 192-1996 and the applicable serial data transmission standard, SMPTE 259M-1997 or SMPTE 292M (proposed). A colour bar test pattern is used. The value of f_{SCLK} is 270 MHz or 360 MHz for SMPTE 259M, 540MHz for SMPTE 344M or 1,485 MHz for SMPTE 292M serial data rates. See Timing Jitter Bandpass section.

Note 10: Intrinsic jitter is defined in accordance with SMPTE RP 184-1996 as: jitter at an equipment output in the absence of input jitter. As applied to this device, the input port is V_{CLK} and the output port is SDO or \overline{SDO} .

Note 11: The signal applied to the appropriate input of this device shall conform to the stated specification.



Test Circuit



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Device Operation

The CLC030 SDTV/HDTV Serializer is used in digital video signal origination equipment: cameras, video tape recorders, telecines and video test and other equipment. It converts parallel SDTV or HDTV component digital video signals into serial format. Logic levels within this equipment are normally produced by LVCMOS logic devices. The encoder produces ECL-compatible serial digital video (SDV) signals conforming to SMPTE 259M, SMPTE 344M (proposed) or SMPTE 292M. The CLC030 operates at parallel data rates of 27.0 MHz, 36.0 MHz, 54.0 MHz and 74.25 MHz. Corresponding serial data rates are 270 Mbps, 360 Mbps, 540 Mbps and 1.485 Gbps.

VIDEO DATA PATH

The **input data register** accepts 10-bit standard definition or 20-bit high definition parallel data and associated clock signals having LVCMOS-compatible signal levels. All parallel data inputs, **DV[19:0]** and **AD[9:0]**, have internal pull-down devices. **VCLK** and **ACLK** do not have internal pull-down devices. Parallel video data may conform to any of several SMPTE standards: 125M, 267M, 260M, 274M, 295M or 296M. For HDTV data, the upper 10 bits of the DV input are luminance (luma) information and the lower 10 bits are co-lour difference (chrominance or chroma) information. For SDTV data, the lower order 10-bits contain both luma and chroma information. Output from this register feeds the video FIFO, video format detection circuit, TRS character detector, SMPTE scrambler, EDH/CRC generators, serializer/NRZI converter and the device control system.

Data from the input data register passes into an 8-register deep **video FIFO** prior to encoding and other processing. The depth of this FIFO is set by a word written into the **VIDEO FIFO Depth[2:0]** bits in the **ANC 0** control register.

The video format detector automatically determines the raster characteristics (video data format) of the parallel input data and configures the CLC030 to properly handle the data. This assures that the data will be properly formatted, that the correct data rate is selected and that ancilliary data and CRC/EDH data are correctly inserted. Indication of the standard being processed is stored in the **FORMAT 1** control data register. This format data can be programmed for output on the multi-function I/O port.

The CLC030 may be configured to operate at a single video format by writing the appropriate **FORMAT SET[4:0]** control data into the **FORMAT 0** control register. Also, the CLC030 may be configured to handle only the standard-definition data formats by setting the **SD ONLY** bit or only the highdefinition data formats by setting the **HD ONLY** bit in the **FORMAT 0** control register. When both of these bits are reset the part automatically detects the data rate and range. The **TRS character detector** processes the timing reference signals which control raster framing. The TRS detector supplies control signals to the system controller to identify the presence of the valid video data. The system controller supplies necessary control signals to the EDH/CRC control block. TRS character LSB-clipping as prescribed in ITU-R BT.601 is used. LSB-clipping causes all TRS characters with a value between 000h and 003h to be forced to 000h and all TRS characters with a value between 3FCh and 3FFh to be forced to 3FFh. Clipping is done prior to scrambling and EDH/CRC character generation.

The CLC030 incorporates circuitry that implements the proposed SMPTE recommended practice and method for LSB dithering. Control of this circuitry is via the Dither Enable bit in the VIDEO INFO 0 control register. Dithering can be selectively enabled during the vertical blanking interval by use of the V Dither Enable bit in the VIDEO INFO 0 control register. The initial condition of Dither Enable and V Dither Enable is OFF.

The **SMPTE scramblers** accept 10-bit standard definition or 20-bit high definition parallel video data and encode it using the polynomial $X^9 + X^4 + 1$ as specified in the respective standard in SMPTE 259M, SMPTE 344M (proposed) or SMPTE 292M. The data is then serialized and sent to the **NRZ-to-NRZI** converter before being output. The transmission bit order is LSB-first.

The **NRZ-to-NRZI converter** accepts NRZ serial data from the SMPTE scrambler. The data is converted to NRZI format using the polynomial (X + 1). The converter's output goes to the output cable driver amplifier.

ANCILLIARY/CONTROL DATA PATH

The Ancilliary and Control Data Port is used to load ancilliary data into the Ancilliary Data FIFO for insertion into the video data stream. This port also provides read/write access to contents of the configuration and control registers. It is also used to configure the Multi-function I/O Port. Ancilliary and control data are input via the 10-bit Ancilliary/Control Data Port, AD[9:0]. The signals RD/WR, ANC/CTRL and ACLK control data flow through the port. The utilization and flow of ancilliary data within the device is managed by a system of control bits, masks and IDs in the control data registers.

Control data are input to and output from the CLC030 using the lower-order 8 bits **AD**[7:0] of the Ancilliary/Control Data Port. This control data initializes, monitors and controls operation of the CLC030. The upper two bits **AD**[9:8] of the

port function as handshaking signals with the device accessing the port. When either a control register read or write address is being written to the port, **AD[9:8]** must be driven as 00b (0XXh, where XX are AD[7:0]). When control data is being written to the port, **AD[9:8]** must be driven as 11b (3XXh, where XX are AD[7:0]). When control data is being read from the port, the CLC030 will output **AD[9:8]** as 10b (2XXh, where XX are output data AD[7:0]) and may be ignored by the monitoring system.

Figure 1 shows the sequence of clock and control signals for reading control data from the ancilliary/control data port. Control data read mode is invoked by making the ANC/ CTRL input low and the RD/WR input high. The 8-bit address of the control register set to be accessed is input to the port on bits AD[7:0]. The address is captured on the rising edge of ACLK. When a control register read address is being written to the port, AD[9:8] must be driven as 00b (0XXh, where XX are AD[7:0]). When control data is being read from the port, the CLC030 will output AD[9:8] as 10b (2XXh, where XX are output data AD[7:0]) and may be ignored by the monitoring system. Data being output from the selected register is driven by the port immediately following the rising edge of ACLK or when the address signal is removed. For optimum system timing, the signals driving the address to the port should be removed before or at the falling edge of ACLK at the end of that address cycle. Output data remains stable until the next rising edge of ACLK and may be written into external devices at any time after the removal of the address signal.

Example: Read the Full-field Flags via the AD port. 1) Take **ANC/CTRL** to a logic-low. 2) Take **RD/WR** to a logic-high. 3) Present 001h to **AD[9:0]** as the register address. 4) Transition **ACLK** L–H. 5) Release the bus driving the AD port. 6) Read the data present on the AD port. The Full-field Flags are bits AD[4:0]. 7) To release the AD port, transition **ACLK** L–H.

Figure 2 shows the sequence of clock and control signals for writing control data to the ancilliary/control data port. The control data write mode is similar to the read mode. Control data write mode is invoked by making the ANC/CTRL input low and the RD/WR input low. The 8-bit address of the control register set to be accessed is input to the port on bits AD[7:0]. The address is captured on the rising edge of ACLK. The address data is removed on the falling edge of ACLK. Next, the control data is presented to the port bits AD[7:0] and written into the selected register on the next rising edge of ACLK. When a control register write address is being written to the port, AD[9:8] must be driven as 00b (0XXh, where XX are AD[7:0]). When control data is being written to the port, AD[9:8] must be driven as 11b (3XXh, where XX are AD[7:0]). Control data written into the registers may be read out non-destructively in most cases.

Example: Enable the TPG Mode and the SMPTE 260M colour bars as test pattern via the AD port. 1) Take **ANC/CTRL** to a logic-low. 2) Take **RD/WR** to a logic-low. 3) Present 00Dh to **AD[9:0]** as the Test 0 register address. 4) Toggle **ACLK**. 5) Present 063h to **AD[9:0]** as the register data. 6) Toggle **ACLK**.



FIGURE 1. Control Data Read Timing (2 read and 1 write cycle shown)



FIGURE 2. Control Data Write Timing

Figure 3 shows the sequence of clock, data and control signals for writing ancilliary data to the port. In ancilliary data write mode, 10-bit **Ancilliary Data** is written into the port using bits **AD[9:0]** and routed to the ancilliary data FIFO. From the FIFO, the ancilliary data can be written into the ancilliary data spaces in the serial video data stream. Ancilliary data write mode is invoked by making the **ANC/CTRL** input high and the **RD/WR** input low. Data presented to the port on a falling edge of **ACLK** is written into the FIFO on the

next rising edge of **ACLK**. Data may only be written to the port when in the ancilliary data mode. Ancilliary data cannot be read from the port.

Admission of ancilliary data to and from the FIFO is controlled by a system of masking and control bits in the control registers. The details and functions of these control registers and bits is explained later in this datasheet.



FIGURE 3. Ancilliary Data Write Timing

MULTI-FUNCTION I/O PORT

The multi-function I/O port can be configured to provide immediate access to many control and indicator functions within the CLC030 configuration and control registers. The individual pins comprising this port are assigned as input or output for selected bits in the control data registers. The multi-function I/O port is configured by way of an 8x6-bit register bank, configuration and control registers I/O pin 0 CONFIG through I/O pin 7 CONFIG. The contents of these registers determine whether the port bits function as inputs or outputs and to which register element each port bit is assigned. Port bits may be assigned to access different register elements or any or all port bits may be assigned to access the same register element (an unlikely or unusual situation). Controls and indicators that are accessible by the port and their corresponding selection addresses are given in the I/O Pin Configuration Register Addresses, Table 6. Table 2 gives the control register bit assignments.

Example: Program multi-function I/O port bit-0 as the CRC Luma Error bit output. 1) Take **ANC/CTRL** to a logic-low. 2) Take **RD/WR** to a logic-low. 3) Present 00Fh to **AD[9:0]** as the **I/O PIN 0 CONFIG** register address. 4) Toggle **ACLK**. 5) Present 310h to **AD[9:0]** as the register data. 6) Toggle **ACLK**.

EDH/CRC SYSTEM

The CLC030 has **EDH** and **CRC character generation and insertion** circuitry. The EDH system functions as described in SMPTE Recommended Practice RP-165. The CRC system functions as specified in SMPTE 292M. The **EDH/CRC polynomial generators** accept parallel data from the input register and generate the EDH and CRC check words for insertion in the serial data. Incoming parallel data is checked for errors and the EDH flags are updated automatically. EDH check words and status flags for SDTV data are generated using the polynomial $X^{16} + X^{12} + X^6 + 1$ per SMPTE RP165. CLC030

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EDH check words are inserted in the serial data stream at the correct positions in the ancilliary data space and formatted per SMPTE 291M. Generation and automatic insertion of the EDH check words is controlled by **EDH Force** and **EDH Enable** bits in the control registers. After a reset, the initial state of all EDH and CRC check characters is 00h.

The SMPTE 292M high definition video standard employs **CRC** (cyclic redundancy check codes) error checking instead of EDH. The CRC consists of two 18-bit words generated using the polynomial $X^{18} + X^5 + X^4 + 1$ per SMPTE 292M. One CRC is used for luminance and one for chrominance data. CRC data is inserted at the required place in the video data according to SMPTE 292M. The CRCs appear in the data stream following the EAV and line number characters.

EDH and CRC errors are reported in the EDH0, EDH1, and EDH2 register sets of the configuration and control registers.

PHASE-LOCKED LOOP SYSTEM

The phase-locked loop (PLL) system generates the output serial data clock at 10x (standard definition) or 20x (high definition) the parallel data clock frequency. This system consists of a VCO, divider chain, phase-frequency detector and internal loop filter. The VCO free-running frequency is internally set. The PLL automatically generates the appropriate frequency for the serial clock rate using the parallel data clock (VCLK) frequency as its reference. Loop filtering is internal to the CLC030. The VCO has separate analog and digital power supply feeds: V_{DDPLLA} pin 62, V_{SSPLLA} pin 61, V_{DDPLLD} pin 1, and V_{SSPLLD} pin 2. These may be separately supplied power via external low-pass filters, if desired. PLL acquisition (lock) time is less than 500µs @ 1,485 Mbps. The VCO halts when VCLK signal is not present or is inactive. A LOCK DETECT indicator is available as a bit in the VIDEO INFO 0 control registers. This function also includes logic to check the stability of the device after the digital logic reset is released following PLL lock. If the system is not fully stable, the logic is automatically reset. LOCK DETECT is a logic-1 when the loop is locked and can be assigned as an output on the multifunction I/O port. The power-on default assigns LOCK DETECT as I/O Port bit 4.

SERIAL DATA OUTPUT DRIVER

The serial data outputs provide low-skew complimentary or differential signals. The output buffer design is intended to drive AC-coupled and terminated, 75Ω coaxial cables. Output levels are 800 mV_{P-P} $\pm 10\%$ into 75 Ω AC-coupled loads. The 75Ω resistors connected to the SDO outputs function both as drain-load and back-matching resistors. No series back-matching resistors should be used. The serial output level is controlled by the value of $R_{REF}LVL$ and $R_{REF}PRE$ connected to pin 53 and pin 52, respectively. The R_{REF}LVL resistor sets the nominal peak-to-peak level of the output signal to the SMPTE nominal level. The R_{REF}PRE resistor sets the value of a pre-emphasis current which is active during the rise and fall times of the HD-rate output signal. The value of $R_{\text{REF}}\text{LVL}$ is normally 4.75 KΩ, ±1%. The value of R_{REF}PRE is normally 4.75 KΩ, ±1%. The rise and fall times of this output buffer design automatically adjust and are different for the HD and SD data rate conditions. The output buffer is quiescent when the device is in an out-oflock condition. The output will become active after the VCO is locked and a valid format has been detected. Separate power feeds are provided for the serial output driver: V_{SSSD}, pins 54, 55, and 59; V_{DDSD}, pin 51; and V_{DDLS}, pin 57. **NOTE:** This output buffer is not designed or specified for driving 50 Ω or other impedance loads.

POWER-ON-RESET AND RESET INPUT

The CLC030 has an automatic, **power-on-reset** circuit. Reset initializes the device and clears TRS detection circuitry, all latches, registers, counters and polynomial generators, sets the EDH/CRC characters to 00h and disables the serial output. An active-HIGH-true, manual **reset input** is available at pin 64. The reset input has an internal pull-down device and is inactive when unconnected.

TEST PATTERN GENERATOR (TPG) AND BUILT-IN SELF-TEST (BIST)

The CLC030 includes an on-board **test pattern generator (TPG)**. Four test pattern types are available in both HD and SD formats, NTSC and PAL standards, and 4x3 and 16x9 raster sizes. The test patterns are: flat-field black, PLL pathological, equalizer (EQ) pathological and a 75%, 8-colour vertical bar pattern. The pathologicals follow recommendations contained in SMPTE RP 178-1996 regarding the test data used. The colour bar pattern has optional bandwidth limiting coding in the chroma and luma data transitions between bars. The VPG FILTER ENABLE bit in the VIDEO INFO 0 control register enables the colour bar filter function. The default condition of VPG FILTER ENABLE is OFF.

The TPG also functions as a **built-in self-test (BIST)** which can verify device functionality. The BIST function performs a comprehensive go/no-go test of the device. The test may be run using any of the HD colour bar test patterns or one of two SD test patterns, either a 270 Mb/s NTSC full-field colour bar or a PLL pathological, as the test data pattern. Data is input internally in the input data register, processed through the device and tested for errors using either the EDH system for SD or the CRC system for HD. A go/no-go indication is logged in the **Pass/Fail** bit of the **TEST 0** control register set. This bit may be assigned as an output on the multifunction I/O port.

TPG and BIST operation is initiated by loading the code for the desired test pattern into the Test Pattern Select[5:0] bits and by setting the TPG Enable bit of the TEST 0 register. In the default power-on state, TPG Enable appears as bit 7 on the multi-function I/O port. The TPG is run by making TPG Enable true and applying the appropriate frequency clock at the VCLK input. Note that when attempting to use the TPG or BIST immediately after the device has been reset or powered on, the TPG defaults to the 270Mbps SD rate and expects a VCLK clock frequency of 27MHz as input. The device must also be configured for the desired test pattern. If HD operation is desired, the device must be further configured for an HD rate and test pattern. (Note also the requirement to initialize the ancilliary data port control logic by clocking ACLK at least three (3) complete cycles before attempting to load the first register address). Table 5 gives the available test patterns and codes. The Pass/Fail bit in the control register gives device test status indication. If no errors have been detected, this bit will be set to logic-1 approximately 2 field intervals after TPG Enable is set. If errors have been detected in the internal circuitry of the CLC030, Pass/Fail will remain reset to a logic-0. The TPG or BIST is stopped by resetting TPG Enable. The serial output data is present at the SDO outputs during TPG or BIST operation.

Example: Enable the TPG Mode to use the NTSC 270Mbps SD colour bars as the BIST and TPG pattern. 1) Take **ANC/CTRL** to a logic-low. 2) Take **RD/WR** to a logic-low. 3) Present 00Bh to **AD[9:0]** as the FORMAT 0 register address. 4) Toggle **ACLK**. 5) Present 343h to **AD[9:0]** as the register data (SD, SMPTE 125M, 27MHz). 6) Toggle **ACLK**. 7) Present 00Dh to **AD[9:0]** as the TEST 0 register address. 8) Toggle **ACLK**. 9) Present 343h to **AD[9:0]** as the register data (SD-ONLY, NTSC 525/30/4x3 Colour bars). 10) Toggle **ACLK**. 11) Take **TPG ENABLE** (I/O Port, bit 7) high. The **PASS/FAIL** indicator (I/O Port, bit 6) is monitored for the result of the test.

CONFIGURATION AND CONTROL REGISTERS

The configuration and control registers store data to configures the operational modes of the CLC030 or which result from its operation. Many of these registers may be assigned as external I/O functions which are then available on the multi-function I/O bus. These functions are summarized in *Table 1* and detailed in *Table 2*. The power-on default condition for the multi-function I/O port is indicated in *Table 1* and detailed in *Table 6*.

Device Operation (Continued)

TABLE 1. Configuration and Control Data Register Summary

Register Function	Bits	Read or Write	Initial Condition	Available on I/O Bus	Notes
CRC Error (SD/HD)	1	R		Output	(Note 12)
CRC Error Luma	1	R		Output	
CRC Error Chroma	1	R		Output	
Full-Field Flags	5	R		No	
Active Picture Flags	5	R		No	
ANC Flags	5	R		No	
EDH Force	1	R/W	OFF	Input	
EDH Enable	1	R/W	OFF	Input	
F/F Flag Error	1	R		Output	
A/P Flag Error	1	R		Output	
ANC Flag Error	1	R		Output	
ANC Checksum Force	1	R/W	OFF	Input	
ANC Checksum Error	1	R		Output	
FIFO Empty	1	R		Output	
FIFO Full	1	R		Output	
FIFO Overrun	1	R/W	OFF	Input/Output	
Video FIFO Depth	3	R/W		No	
ANC ID	16	R/W	0000h	No	
ANC Mask	16	R/W	FFFFh	No	
MSG Track	1	R/W	OFF	No	
MSG Flush Static	1	R/W	OFF	No	
MSG Flush Dynamic	1	R/W	OFF	No	
FIFO Flush Static	1	R/W	OFF	No	
FIFO Flush Dynamic	1	R/W	OFF	No	
MSG Flush Static	1	R/W	OFF	No	
Full MSG Required	1	R/W	OFF	No	
Chksum Attach In	1	R/W	OFF	Input	
FIFO Insert Enable	1	R/W	OFF	Input	
VANC	1	R/W	OFF	No	
Switch Point 0	8	R/W	00h	No	
Switch Point 1	8	R/W	00h	No	
Switch Point 2	8	R/W	00h	No	
Switch Point 3	8	R/W	00h	No	
Format Set	5	R/W	OFF	No	
SD Only	1	R/W	OFF	No	
HD Only	1	R/W	OFF	No	
Format	5	R		Output	Format [4] (Note 12)
Н	1	R		Output	(Note 12)
V	1	R		Output	(Note 12)
F	1	R		Output	(Note 12)
Test Pattern Select	6	R/W	525/27 MHz/Black	Input	
TPG Enable	1	R/W	OFF	Input	(Note 12)
Pass/Fail	1	R		Output	(Note 12)
New Sync Position (NSP)	1	R		Output	
SAV	1	R		Output	
EAV	1	R		Output	

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unction	Bits	Read or W						
ble	1		rite	Initia	I Condition	Available o I/O Bus	n I	Notes
ble	1	R				Output	(Note 1	2)
	1	R/W			OFF	Input		
	1	R/W				Input		
able	1	R/M			OFF	Input		
	1			0.0		No		
oniig.	48	R/W		566		INO		
ed to multifunction I	O port at p	power-on.						
gic-1, OFF = logic-0	(positive i	ogic).						
		TABLE 2. Co	ontrol Reg	gister	Bit Assignme	nts		
Bit 6		Bit 5	Bit 4	ı	Bit 3	Bit 2	Bit 1	Bit 0
dress 01h)								•
EDH FORCE	E	DH ENABLE	F/F FLAG	GS(4)	F/F FLAGS(3)	F/F FLAGS(2)	F/F FLAGS(1)	F/F FLAGS(0)
dress 02h)				()		()	()	
CRC ERROR LUM		RROR CHROMA	A/P FLAG	GS(4)	A/P FLAGS(3)	A/P FLAGS(2)	A/P FLAGS(1)	A/P FLAGS(0)
dress 03h)				()		()	()	()
A/P FLAG ERROF	R ANC	FLAG ERROR	ANC FLAG	GS(4)	ANC FLAGS(3)	ANC FLAGS(2)	ANC FLAGS(1)	ANC FLAGS(0)
dress 04h)	- I			/	()	()	()	()
VIDEO		VIDEO	FIFO)	FIFO	FIFO	ANC CHECK-	ANC CHECK-
FIFO-DEPTH(1)	FIF	O-DEPTH(0)	OVERR	UN	EMPTY	FULL	SUM ERROR	SUM FORCE
dress 05h)								
ANC ID(6)		ANC ID(5)	ANC ID	0(4)	ANC ID(3)	ANC ID(2)	ANC ID(1)	ANC ID(0)
dress 06h)								
ANC ID(14)	A	NC ID(13)	ANC ID((12)	ANC ID(11)	ANC ID(10)	ANC ID(9)	ANC ID(8)
dress 07h)								
ANC MASK(6)	AN	IC MASK(5)	ANC MAS	SK(4)	ANC MASK(3)	ANC MASK(2)	ANC MASK(1)	ANC MASK(0)
dress 08h)	1							
ANC MASK(14)	AN	C MASK(13)	ANC MAS	SK(12)	ANC MASK(11)	ANC MASK(10)	ANC MASK(9)	ANC MASK(8)
dress 17h)								
ATTACH IN			DYNAN	/IC	STATIC	DYNAMIC	STATIC	MSG TRACK
dress 18h)								
reserved	IA I	NC PARITY	*BUFFE	ER	*BUFFER	*FRAME	*FRAME	VANC
leselveu		MASK	READY	' IN	READY OUT	LOCK IN	LOCK OUT	VANO
register address	09h)							
LINE(6)		LINE(5)	LINE(4	4)	LINE(3)	LINE(2)	LINE(1)	LINE(0)
register address	0Ah)							
PROTECT(3)		ROTECT(2)	PROTEC	CT(1)	PROTECT(0)	LINE(10)	LINE(9)	LINE(8)
register address	19h)			0				
LINE(6)		LINE(5)	LINE(4	4)	LINE(3)	LINE(2)	LINE(1)	LINE(0)
register address	1Ah)		DDOTTO					
PROTECT(3)	PI	ROTECT(2)	PROTEC	(1)	PROTECT(0)	LINE(10)	LINE(9)	LINE(8)
er address 0Bh)			FORMAT					
SD ONLY		HD ONLY	FORMATS	SEI(4)	FORMAT SET(3)	FORMAT SET(2)	FORMAT SET(1)	FORMAT SET(0)
		L	EODMA					
		п	FURMA	1 (4)	FURIVIAT(3)	FURIVIAT(2)	FURIVIAT(1)	FURIVIAI(U)
aaress UDN)			TFOT DAT	TEDNI	TEOT DATTERN	TEOT DATTERN	TEOT DATTERN	TEOT DATTERN
TPG ENABLE		SELECT(5)	SELECT	T(4)	SELECT(3)	SELECT(2)	SELECT(1)	SELECT(0)
ister address 0Eh	ı)							
VERT. DITHER ENABLE	V	PG FILTER ENABLE	LOCK DETEC	К СТ	EAV	SAV	NSP	reserved
	Bit 6 fress 01h) EDH FORCE fress 02h) RC ERROR LUM fress 03h) A/P FLAG ERROF fress 04h) VIDEO FIFO-DEPTH(1) fress 05h) ANC ID(6) fress 06h) ANC ID(14) fress 07h) ANC MASK(6) fress 08h) ANC MASK(14) fress 17h) CHKSUM ATTACH IN CHKSUM ATTACH IN fress 18h) reserved register address LINE(6) register address LINE(6) register address PROTECT(3) register address PROTECT(3) register address SD ONLY r address 0Bh) SD ONLY r address 0Ch) V ITPG ENABLE ister address 0EH VERT. DITHER ENABLE	Bit 6 tress 01h) EDH FORCE Iress 02h) IRC ERROR LUMA CRC EI tress 03h) A/P FLAG ERROR ANC Iress 04h) VIDEO FIFO-DEPTH(1) Iress 05h) ANC ID(6) ANC ID(14) ANC ID(14) ANC MASK(6) ANC MASK(14) AN tress 08h) ANC MASK(14) AN tress 08h) ANC MASK(14) AN tress 08h) CHKSUM register address 09h) LINE(6) register address 19h) LINE(6) register address 18h) PROTECT(3) PI r address 0Bh) SD ONLY V idress 0Dh)	Bit 6 Bit 5 Iress 01h) EDH FORCE EDH ENABLE Iress 02h) :RC ERROR LUMA CRC ERROR CHROMA Iress 03h) A/P FLAG ERROR ANC FLAG ERROR A/P FLAG ERROR ANC FLAG ERROR Ifeo-DEPTH(0) Iress 03h) VIDEO FIFO-DEPTH(0) Iress 05h) ANC ID(6) ANC ID(5) ANC ID(6) ANC ID(13) Iress 07h) ANC ID(14) ANC ID(13) Iress 07h) ANC MASK(6) ANC MASK(5) Iress 07h) ANC MASK(14) ANC MASK(13) Iress 17h) CHKSUM FULL MSG ATTACH IN REQUIRED Gress 18h) REQUIRED reserved ANC PARITY MASK register address 09h) LINE(6) LINE(5) register address 09h) LINE(6) LINE(5) register address 19h) LINE(5) register address 14h) PROTECT(3) PROTECT(2) register address 08h) SD ONLY HD ONLY r r address 0Bh) SD ONLY HD ONLY r address 0Dh) V H	TABLE 2. Control Registrer Bit 6 Bit 5 Bit 4 iress 01h) EDH FORCE EDH ENABLE F/F FLAG iress 02h) :RC ERROR LUMA CRC ERROR CHROMA A/P FLAG FLAG A/P FLAG ERROR ANC FLAG ERROR ANC FLAG FIFC Iress 03h)	TABLE 2. Control Register Bit 6 Bit 5 Bit 4 Iress 01h) EDH FORCE EDH ENABLE F/F FLAGS(4) Iress 02h) RC ERROR LUMA CRC ERROR CHROMA A/P FLAGS(4) AVP FLAG ERROR ANC FLAG ERROR ANC FLAGS(4) Iress 03h) VIDEO FIFO-DEPTH(0) OVERRUN VIDEO Soft ANC FLAG ERROR ANC FLAGS(4) Iress 04h) VIDEO FIFO-DEPTH(0) OVERRUN Iress 05h) ANC ID(5) ANC ID(4) ANC ID(5) ANC ID(12) ANC Soft ANC ID(14) ANC ID(13) ANC ID(12) Iress 05h) ANC MASK(6) ANC MASK(5) ANC MASK(4) ANC MASK(14) ANC MASK(13) ANC MASK(12) Iress 17h) CHKSUM FULL MSG FIFO FLUSH CHKSUM FULL MSG FIFO FLUSH ATTACH IN REQUIRED DYNAMIC Iress 18h) Feserved ANC PARITY *BUFFER register address 09h) LINE(5) LINE(4) Fegister address 19h) LINE(5) LINE(4) register address 19h) LINE(5) LINE(4)	TABLE 2. Control Register Bit Assignmed Bit 6 Bit 5 Bit 4 Bit 3 iress 01h) EDH FORCE EDH ENABLE F/F FLAGS(4) F/F FLAGS(3) iress 02h) CRC ERROR LUMA/CRC ERROR CHROMA A/P FLAGS(4) A/P FLAGS(3) iress 03h) A/P FLAG ERROR ANC FLAG ERROR ANC FLAGS(4) ANC FLAGS(3) iress 03h) VIDEO FIFO FIFO FIFO VIDEO VIDEO FIFO FIFO PIPO-DEPTH(1) FIFO-DEPTH(0) OVERRUN EMPTY iress 05h) ANC ID(6) ANC ID(5) ANC ID(12) ANC ID(3) ANC ID(6) ANC ID(13) ANC ID(12) ANC ID(11) Intrass 06h) ANC ID(14) ANC ID(5) ANC MASK(4) ANC MASK(3) ANC MASK(6) ANC MASK(5) ANC MASK(12) ANC MASK(11) ANC MASK(14) ANC MASK(13) ANC MASK(12) ANC MASK(11) ANC MASK(14) ANC PARITY "BUFFER "BUFFER reserved MAC PARITY "BUFFER "BUFFER reserved MAC PARITY "BUFFER READY OUT <	TABLE 2. Control Register Bit Assignments Bit 6 Bit 5 Bit 4 Bit 3 Bit 2 iress 01h) EDH FORCE EDH ENABLE F/F FLAGS(4) F/F FLAGS(3) F/F FLAGS(2) iress 02h) : <td< td=""><td>TABLE 2. Control Register Bit Assignments Bit 6 Bit 5 Bit 4 Bit 3 Bit 2 Bit 1 ress 01h) EDH FORCE EDH ENABLE F/F FLAGS(4) F/F FLAGS(3) F/F FLAGS(2) F/F FLAGS(1) ress 03h) ARC FLAG ERROR ANC FLAG ERROR ANC FLAGS(3) A/P FLAGS(2) A/P FLAGS(1) A/P FLAGS(2) A/P FL</td></td<>	TABLE 2. Control Register Bit Assignments Bit 6 Bit 5 Bit 4 Bit 3 Bit 2 Bit 1 ress 01h) EDH FORCE EDH ENABLE F/F FLAGS(4) F/F FLAGS(3) F/F FLAGS(2) F/F FLAGS(1) ress 03h) ARC FLAG ERROR ANC FLAG ERROR ANC FLAGS(3) A/P FLAGS(2) A/P FLAGS(1) A/P FLAGS(2) A/P FL

Device Operation (Continued)

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TABLE 2. Control Register Bit Assignments (Continued)								
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
MULTI-FUNCTION I/O BUS PIN CONFIGURATION								
I/O PIN 0 CONFIG	(register address 0	Fh)						
reserved	reserved	PIN 0 SEL[5]	PIN 0 SEL[4]	PIN 0 SEL[3]	PIN 0 SEL[2]	PIN 0 SEL[1]	PIN 0 SEL[0]	
I/O PIN 1 CONFIG	(register address 1	0h)	•					
reserved	reserved	PIN 1 SEL[5]	PIN 1 SEL[4]	PIN 1 SEL[3]	PIN 1 SEL[2]	PIN 1 SEL[1]	PIN 1 SEL[0]	
I/O PIN 2 CONFIG	(register address 1	1h)	•	•				
reserved	reserved	PIN 2 SEL[5]	PIN 2 SEL[4]	PIN 2 SEL[3]	PIN 2 SEL[2]	PIN 2 SEL[1]	PIN 2 SEL[0]	
I/O PIN 3 CONFIG (register address 12h)								
reserved	reserved	PIN 3 SEL[5]	PIN 3 SEL[4]	PIN 3 SEL[3]	PIN 3 SEL[2]	PIN 3 SEL[1]	PIN 3 SEL[0]	
I/O PIN 4 CONFIG (register address 13h)								
reserved	reserved	PIN 4 SEL[5]	PIN 4 SEL[4]	PIN 4 SEL[3]	PIN 4 SEL[2]	PIN 4 SEL[1]	PIN 4 SEL[0]	
I/P PIN 5 CONFIG (register address 14h)								
reserved	reserved	PIN 5 SEL[5]	PIN 5 SEL[4]	PIN 5 SEL[3]	PIN 5 SEL[2]	PIN 5 SEL[1]	PIN 5 SEL[0]	
I/O PIN 6 CONFIG	(register address 1	5h)		•				
reserved	reserved	PIN 6 SEL[5]	PIN 6 SEL[4]	PIN 6 SEL[3]	PIN 6 SEL[2]	PIN 6 SEL[1]	PIN 6 SEL[0]	
I/O PIN 7 CONFIG	I/O PIN 7 CONFIG (register address 16h)							
reserved	reserved	PIN 7 SEL[5]	PIN 7 SEL[4]	PIN 7 SEL[3]	PIN 7 SEL[2]	PIN 7 SEL[1]	PIN 7 SEL[0]	

Register Name	Address	Address
	Decimal	Hexadecimal
EDH 0	1	01
EDH 1	2	02
EDH 2	3	03
ANC 0	4	04
ANC 1	5	05
ANC 2	6	06
ANC 3	7	07
ANC 4	8	08
ANC 5	23	17
ANC 6	24	18
SWITCH POINT 0	9	09
SWITCH POINT 1	10	0A
SWITCH POINT 2	25	19
SWITCH POINT 3	26	1A
FORMAT 0	11	0B
FORMAT 1	12	0C
TEST 0	13	0D
VIDEO INFO 0	14	0E
I/O PIN 0 CONFIG	15	0F
I/O PIN 1 CONFIG	16	10
I/O PIN 2 CONFIG	17	11
I/O PIN 3 CONFIG	18	12
I/O PIN 4 CONFIG	19	13
I/O PIN 5 CONFIG	20	14
I/O PIN 6 CONFIG	21	15
I/O PIN 7 CONFIG	22	16

TABLE 3. Control Register Addresses

EDH REGISTERS 0, 1 AND 2 (Addresses 01h through 03h)

The CRC Error flag indicates that parallel data has been input that contains detected errors in either the EDH checksums (SD) or CRC checkwords (HD).

Updated EDH packets may be inserted by setting the EDH Force bit in the control registers. The EDH Force control bit causes the insertion of new EDH checkwords and flags into the serial output regardless of the previous condition of EDH checkwords and flags in the input parallel data. This function may be used in situations where video content has been editted thus making the previous EDH information invalid. In the case of SMPTE 292M data, the CRC check characters are recalculated and inserted automatically regardless of the presence of CRC characters in the parallel data. After the CLC030 is reset, the initial state of the CRC check characters is 00h.

The EDH Enable bit enables operation of the EDH generator function.

The EDH flags F/F FLAGS[4:0], A/P FLAGS[4:0] and ANC FLAGS[4:0] are defined in SMPTE RP 165. The EDH flags are stored in the control registers. The flags are updated automatically when the EDH function is enabled and data is being received.

The status of EDH flag errors in incoming SD parallel data are reported in the ffFlagError, apFlagError and anc-FlagError bits. The ffFlagError, apFlagError and anc-FlagError bits are the logical-OR of the corresponding EDH and EDA flags of the EDH checkwords.

CRC errors in incoming HD parallel data are reported in the CRC ERROR, CRC ERROR LUMA and CRC ERROR CHROMA bits in the control registers.

ANC REGISTERS 1 THROUGH 6 (Addresses 04h through 08h, 17h and 18h)

The V FIFO Depth[2:0] bits control the depth of the video FIFO which follows the input data latches. The depth can be set from 0 to 7 stages deep by writing the corresponding binary code into these bits. For example: to set the Video FIFO depth at two registers, load 11010XXXXXb into the

ANC 0 control register (where X represents the other functional bits of this register). To retain other data previously stored in a register, read the register's contents and logically-OR this with the new data. Then write the composite data back into the register.

Flags for **FIFO EMPTY**, **FIFO FULL** and **FIFO OVERRUN** are available in the configuration and control register set. These flags can also be assigned as inputs and outputs on the multi-function I/O port. The **FIFO OVERRUN** flag indicates that an attempt to write data into a full FIFO has occurred. When **FIFO FLUSH DYNAMIC** or **MSG FLUSH DYNAMIC** are enabled, the **FIFO OVERRUN** function is superceded. When **FIFO OVERRUN** is active and not superceded, it can be reset by reading the bit's status via the Ancilliary/Command port. To be used properly, **FIFO OVER-RUN** should be assigned as an output on the multi-function I/O port and monitored by the host system. Otherwise, inadvertent loss of ancilliary packet data could occur.

The ANC Checksum Force bit, under certain conditions, enables the overwriting of ancilliary data checksums received in the parallel ancilliary data. Calculation and insertion of new ancilliary data checksums is controlled by the ANC Checksum Force bit. If a checksum error is detected (calculated and received checksums do not match) and the ANC Checksum Force bit is set, a new checksum will be inserted in the ancilliary data replacing the previous one. If a checksum error is detected and the ANC Checksum Force bit is not set, the checksum mismatch is reported via the ANC Checksum Error bit.

Ancilliary data checksums may be received in the incoming parallel ancilliary data. Alternatively they may be calculated and inserted automatically by the CLC030. The CHK-SUM ATTACH IN bit in the control registers when set to a logic-1 indicates that the checksum is to be supplied in the incoming data. When the CHKSUM ATTACH IN bit is set, checksums for incoming data are calculated and checked against received checksums. Calculation and insertion of new ancilliary data checksum is controlled by the ANC Checksum Force bit in the configuration and control registers. If a checksum error is detected (calculated and received checksums do not match) and the ANC Checksum Force bit is set, a new checksum will be inserted in the ancilliary data replacing the previous one. If a checksum error is detected and the ANC Checksum Force bit is not set, the checksum mismatch is reported via the ANC CHECKSUM ERROR bit in the control registers.

The **ANC Checksum Error** bit indicates that the received ancilliary data checksum did not agree with the CLC030's internally generated checksum. This bit is available as an output on the multifunction I/O port.

Admission of ancilliary data packets into the FIFO is controlled by the **ANC MASK**[15:0] and **ANC ID**[15:0] bits in the control registers. The **ANC ID**[15:0] normally is set to a valid 16-bit code used for component ancilliary data packet identification as specified in SMPTE 291M-1998. The **ANC MASK**[15:0] is a 16-bit word that can be used to selectively control loading of packets with specific IDs (or ID ranges) into the FIFO. When the **ANC MASK**[15:0] is set to FFFFh, packets with any ID can be loaded into the FIFO. When any bit or bits of the **ANC MASK**[15:0] are set to a logic-1, the corresponding bit or bits of the **ANC ID**[15:0] are a don'tcare when matching IDs of incoming packets. When the **ANC MASK**[15:0] is set to 0000h, the ANC ID of incoming packets must match exactly, bit-for-bit the **ANC ID**[15:0] set in the control register for the packets to be loaded into the FIFO. The initial value of the **ANC MASK[15:0]** is FFFFh and the **ANC ID[15:0]** is 0000h.

The **ANC PARITY MASK** bit when set disables parity checking for the DATA ID (DID) and SECONDARY DATA ID (SDID) in the ANC data packet. When reset, parity checking is enabled, and, if a parity error occurs, the packet will not be loaded.

The **FIFO INSERT ENABLE** bit in the control registers enables insertion of ancilliary data stored in the FIFO into the serial data stream. Data insertion is enabled when this bit is set to a logic-1. This bit can be used to delay automatic insertion of data into the serial data stream.

The CLC030 can keep track of up to 8 ancilliary packets in the FIFO. Incoming packet length versus available space in the FIFO is also tracked. The **MSG TRACK** bit in the control registers, when set, enables tracking of packets in the FIFO. **MSG TRACK** also enables several other functions for control of packet traffic in the FIFO: **FIFO FLUSH DYN**, **FIFO FLUSH STAT**, **MSG FLUSH DYN**, and **MSG FLUSH STAT**.

With message tracking enabled and **FIFO FLUSH DYN** set to a logic-1, if a FIFO full condition is encountered, all existing message packets in the FIFO will be flushed. The current message packet will be left intact. When **FIFO FLUSH DYN** is not set and a FIFO full condition is encountered, the FIFO will overrun and the **FIFO OVERRUN** flag will be set. **FIFO FLUSH DYN** remains set until cleared.

With message tracking enabled, setting the **FIFO FLUSH STAT** bit to a logic-1 flushes the FIFO. Data may not be loaded into the FIFO during **FIFO FLUSH STAT** execution. **FIFO FLUSH STAT** is automatically reset after this operation is complete.

With message tracking enabled and **MSG FLUSH DYN** set to a logic-1, the oldest message packet in the FIFO will be flushed when the next message is written to the FIFO. **MSG FLUSH DYN** remains set until cleared.

With message tracking enabled and **MSG FLUSH STAT** set to a logic-1, the oldest message packet in the FIFO is flushed when data is not being written to the FIFO. **MSG FLUSH STAT** is automatically reset after this operation is complete.

The **FULL MSG REQ** (full message required) bit in the control registers, when set, instructs the CLC030 to insert only complete packets residing in the FIFO into the serial data stream. When this bit is not set, messages of any length, incomplete or partial, will be inserted into the serial data stream. This function is not affected by **MSG TRACK**. This function can be used to prevent overrunning available space in the FIFO.

The **VANC** bit in the control registers, when set to a logic-1, enables insertion of ancilliary data during the vertical blanking interval (both active video and horizontal blanking portions of the line).

SWITCH POINT REGISTERS 0 THROUGH 3 (Addresses 09h, 0Ah, 19h and 1Ah)

The **Line[10:0]** and **Protect[4:0]** bits define the vertical switching point. The vertical switching point for component digital standard definition formats is defined in SMPTE RP 168-1993. The vertical switching point for high-definition formats has the same basic definition. However, since the vertical switching point has not been standardized among the various high-definition rasters, these registers provide a convenient means whereby the vertical switching point may be programmed by the user.

FORMAT REGISTERS 0 AND 1 (Addresses 0Bh and 0Ch)

The CLC030 may be set to process a single video format by writing the appropriate data into the **FORMAT 0** register. The **Format Set[4:0]** bits configure the raster format parameters to recognize and process only the specified type of standard or high definition format. These formats and codes are detailed in *Table 4*. Bit **Format Set[4]** when set indicates that HD data is to be processed. When reset, SD data is indicated. **Format Set[3]** when set indicates that PAL data is to be processed. When reset NTSC data is to be processed. **Format Set[2:0]** correspond to one of the sub-standards given in the table.

The CLC030 can automatically determine the format of the incoming parallel data. The result of this operation is stored in the **FORMAT 1** register. The **Format[4:0]** bits identify which of the many possible video data standards that the CLC030 can process is being received. These format codes follow the same arrangement as for the **Format Set[4:0]** bits. These formats and codes are given in *Table 4*. Bit **Format[4]** when set indicates that HD data is being processed. When reset, SD data is indicated. **Format[3]** when set indicates that PAL data is being processed. When reset NTSC data is being processed. **Format[2:0]** correspond with one of the sub-standards given in the table.

TABLE 4. Video Raster Format Parameters

Format Code [4,3,2,1,0]	Format	Spec.	Frame Rate	Lines	Active Lines	Samples	Active Samples
00001	SDTV, 54	RP 174	601	525	507/487*	3432	2880
00010	SDTV, 36	SMPTE 267	601	525	507/487*	2288	1920
00011	SDTV, 27	SMPTE 125	601	525	507/487*	1716	1440
01001	SDTV, 54	ITU-R BT 601.5	501	625	577	3456	2880
01010	SDTV, 36	ITU-R BT 601.5	501	625	577	2304	1920
01011	SDTV, 27	ITU-R BT 601.5	501	625	577	1728	1440
10001	HDTV, 74.25	SMPTE 260	301	1125	1035	2200	1920
10010	HDTV, 74.25	SMPTE 274	301	1125	1080	2200	1920
10011	HDTV, 74.25	SMPTE 274	30P	1125	1080	2200	1920
11001	HDTV, 74.25	SMPTE 274	251	1125	1080	2640	1920
11010	HDTV, 74.25	SMPTE 274	25P	1125	1080	2640	1920
11100	HDTV, 74.25	SMPTE 295	251	1250	1080	2376	1920
11101	HDTV, 74.25	SMPTE 274	24P	1125	1080	2750	1920
10100	HDTV, 74.25	SMPTE 296	60P	750	720	1650	1280

The **HD Only** bit when set to a logic-1 locks the CLC030 into the high definition data range and frequency. In systems designed to handle only high definition signals, enabling **HD Only** reduces the time required for the CLC030 to establish frequency lock and determine the HD format being processed.

The **SD Only** bit when set to a logic-1 locks the CLC030 into the standard definition data ranges and frequencies. In systems designed to handle only standard definition signals, enabling **SD Only** reduces the time required for the CLC030 to establish frequency lock and determine the format being processed. When **SD Only** and **HD Only** are set to logic-0, the device operates in SD/HD mode.

The **H**, **V**, and **F** bits correspond to input TRS data bits 6, 7 and 8, respectively. The meaning and function of this data is the same for both standard definition (SMPTE 125M) and high definition (SMPTE 292M luminance and colour difference) video data. Polarity is logic-1 equals HIGH-true. These bits are registered for the duration of the applicable field.

TEST 0 REGISTER (Address 0Dh)

The **Test Pattern Select** bits determine which test pattern is output when the Test Pattern Generator (TPG) mode or the Built-in Self-Test (BIST) mode is enabled. *Table 5* gives the codes corresponding to the various test patterns. All HD

colour bars test patterns are BIST data. SD BIST test patterns are: NTSC, 27MHz, 4x3 Colour Bars and PAL, 27MHz, 4x3 PLL Pathological.

The **TPG Enable** bit when set to a logic-1 enables the Test Pattern Generator function and built-in self-test (BIST). This bit is mapped to I/O port bit 7 in the default condition.

The **Pass/Fail** bit indicates the result of the built-in self-test. This bit is a logic-1 for a pass condition. This bit is mapped to I/O port bit 6 in the default condition.

st Pattern Select Word Bits >	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit (
	1=HD	1=Progressive			00=Black	
Video Dector Standard		0=Interlaced			01=PLL Pa	th.
video Raster Standard	0=SD	1=PAL	1		10=EQ Pat	h.
		0=NTSC			11=Colour	Bars
1125 Line, 74.25 MHz, 30 Frame I	nterlaced Com	ponent (SMPTE 260M)			
Ref. Black	1	0	0	0	0	0
PLL Path.	1	0	0	0	0	1
EQ Path.	1	0	0	0	1	0
Colour Bars	1	0	0	0	1	1
1125 Line, 74.25 MHz, 30 Frame I	nterlaced Com	ponent (SMPTE 274M)			
Ref. Black	1	0	0	1	0	0
PLL Path.	1	0	0	1	0	1
EQ Path.	1	0	0	1	1	0
Colour Bars	1	0	0	1	1	1
1125 Line, 74.25 MHz, 25 Frame I	nterlaced Com	ponent (SMPTE 274M	l)			
Ref. Black	1	0	1	0	0	0
PLL Path.	1	0	1	0	0	1
EQ Path.	1	0	1	0	1	0
Colour Bars	1	0	1	0	1	1
1125 Line, 74.25 MHz, 25 Frame I	nterlaced Com	ponent (SMPTE 295M)			
Ref. Black	1	0	1	1	0	0
PLL Path.	1	0	1	1	0	1
EQ Path.	1	0	1	1	1	0
Colour Bars	1	0	1	1	1	1
1125 Line, 74.25 MHz, 30 Frame F	Progressive Co	mponent (SMPTE 274	4M)			
Ref. Black	1	1	0	0	0	0
PLL Path.	1	1	0	0	0	1
EQ Path.	1	1	0	0	1	0
Colour Bars	1	1	0	0	1	1
1125 Line, 74.25 MHz, 25 Frame F	Progressive Co	mponent (SMPTE 274	4M)			
Ref. Black	1	1	0	1	0	0
PLL Path.	1	1	0	1	0	1
EQ Path.	1	1	0	1	1	0
Colour Bars	1	1	0	1	1	1
1125 Line, 74.25 MHz, 24 Frame F	Progressive Co	mponent (SMPTE 274	4M)			
Ref. Black	1	1	1	0	0	0
PLL Path.	1	1	1	0	0	1
EQ Path.	1	1	1	0	1	0
Colour Bars	1	1	1	0	1	1
750 Line, 74.25 MHz, 60 Frame Pr	ogressive Com	ponent (SMPTE 296	V)		11	
Ref. Black	1	1	1	1	0	0
PLL Path.	1	1	1	1	0	1
EQ Path.	1	1	1	1	1	0
Colour Bars	1	1	1	1	1	1

Device	Operation	(Continued)
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TADLE J. TEST I diletti Selection Codes (Continued
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Т	ABLE 5. Test	Pattern Selection (Codes (Contin	ued)		
Test Pattern Select Word Bits >	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
525 Line, 30 Frame, 27 MHz, NTSC	4x3 (SMPTE	125M)				
Ref. Black	0	0	0	0	0	0
PLL Path.	0	0	0	0	0	1
EQ Path.	0	0	0	0	1	0
Colour Bars (SD BIST)	0	0	0	0	1	1
625 Line, 25 Frame, 27 MHz, PAL	4x3 (ITU-T BT.	601)				
Ref. Black	0	1	0	0	0	0
PLL Path. (SD BIST)	0	1	0	0	0	1
EQ Path.	0	1	0	0	1	0
Colour Bars	0	1	0	0	1	1
525 Line, 30 Frame, 36 MHz, NTSC	16x9 (SMPTE	125M)				
Ref. Black	0	0	0	1	0	0
PLL Path.	0	0	0	1	0	1
EQ Path.	0	0	0	1	1	0
Colour Bars	0	0	0	1	1	1
625 Line, 25 Frame, 36 MHz, PAL	16x9 (ITU-T BT	.601)	1		1	
Ref. Black	0	1	0	1	0	0
PLL Path.	0	1	0	1	0	1
EQ Path.	0	1	0	1	1	0
Colour Bars	0	1	0	1	1	1
525 Line, 30 Frame, 54 MHz (NTSC	;)		•			
Ref. Black	0	0	1	0	0	0
PLL Path.	0	0	1	0	0	1
EQ Path.	0	0	1	0	1	0
Colour Bars	0	0	1	0	1	1
625 Line, 25 Frame, 54 MHz (PAL)	-	1	1	1		
Ref. Black	0	1	1	0	0	0
PLL Path.	0	1	1	0	0	1
EQ Path.	0	1	1	0	1	0
Colour Bars	0	1	1	0	1	1

Note:SD BIST patterns are NTSC 4x3 Colour Bars and PAL 4x3 PLL Pathological. HD BIST patterns are colour bars for each format.

VIDEO INFO 0 REGISTER (Address 0Eh)

The NSP (New Sync Position) bit indicates that a new or out-of-place TRS character has been detected in the input data. This bit is set to a logic-1 and remains set for at least one horizontal line period or unless re-activated by a subsequent new or out-of-place TRS. It is reset by an EAV TRS character. {Activation of this function flushes the existing state of the machine reseting the EDH/CRC generator, SMPTE polynomial generator, serializer and NRZ-NRZI converter.

The EAV (end of active video) and SAV (start of active video) bits track the occurrence of the corresponding TRS characters.

The lock detect indicator is registered as a control signal and is a logic-1 when the loop is locked. This bit may be programmed as an output on the multi-function I/O bus. This bit is mapped to I/O port bit 4 in the default condition.

The VPG Filter Enable bit when set enables operation of the Video Pattern Generator filter. Operation of this filter causes the insertion of transition codes in the chroma and luma data of colour bar test patterns where these patterns change from one bar to the next. This filter reduces the magnitude of out-of-band frequency products which can be produced by abrupt transitions in the chroma and luma data when fed to D-to-A converters and picture monitors.

I/O PIN 0 THROUGH 7 CONFIGURATION REGISTERS (Addresses 0Fh through 16h)

The I/O Pin Configuration Registers are used to map the bits of the multi-function I/O port to selected bits of the Configuration and Control Registers. Table 6 gives the addresses of the Configuration and Control register functions that may be mapped to the port and their corresponding addresses. Pin[n] Config bit 5 controls whether the port pin is input or output. The port pin will be an input when this bit is set and an output when reset. Input-only functions may not be configured as outputs and vice versa.

Example: Program, via the AD port, I/O port bit 0 as output for the CRC Luma Error bit in the control registers. 1) Take **ANC/CTRL** to a logic-low. 2) Take **RD/WR** to a logic-low. 3) Present 00Fh to AD[9:0] as the I/O PIN 0 CONFIG register

address. 4) Toggle **ACLK**. 5) Present 010h to **AD[9:0]** as the register data, the bit address of the CRC Luma Error bit in the control registers. 6) Toggle **ACLK**.

TABLE 6. I/O Configuration Register Addresses for Control Register Functions

Deviator Dit		Bit	Address	Pin # SEL		Davida Ora Otativa		
Register Bit	[5]	[4]]	[3]	[2]	[1]	[0]	I/P or O/P	Power-On Status
reserved	0	0	0	0	0	0	Output	
FF Flag Error	0	0	0	0	0	1	Output	
AP Flag Error	0	0	0	0	1	0	Output	
ANC Flag Error	0	0	0	0	1	1	Output	
CRC Error (SD/HD)	0	0	0	1	0	0	Output	I/O Port Bit 5
reserved	0	0	0	1	0	1	Output	
reserved	0	0	0	1	1	0	Output	
reserved	0	0	0	1	1	1	Output	
reserved	0	0	1	0	0	0	Output	
reserved	0	0	1	0	0	1	Output	
reserved	0	0	1	0	1	0	Output	
reserved	0	0	1	0	1	1	Output	
reserved	0	0	1	1	0	0	Output	
SAV	0	0	1	1	0	1	Output	
EAV	0	0	1	1	1	0	Output	
NSP	0	0	1	1	1	1	Output	
CRC Luma Error	0	1	0	0	0	0	Output	
CRC Chroma Error	0	1	0	0	0	1	Output	
F	0	1	0	0	1	0	Output	I/O Port Bit 0
V	0	1	0	0	1	1	Output	I/O Port Bit 1
Н	0	1	0	1	0	0	Output	I/O Port Bit 2
Format[0]	0	1	0	1	0	1	Output	
Format[1]	0	1	0	1	1	0	Output	
Format[2]	0	1	0	1	1	1	Output	
Format[3]	0	1	1	0	0	0	Output	
Format[4]	0	1	1	0	0	1	Output	I/O Port Bit 3 (SD/HD)
FIFO Full	0	1	1	0	1	0	Output	I/O Port Bit 6
FIFO Empty	0	1	1	0	1	1	Output	
Lock Detect	0	1	1	1	0	0	Output	I/O Port Bit 4
Pass/Fail	0	1	1	1	0	1	Output	
FIFO Overrun	0	1	1	1	1	0	Output	
ANC Chksum Error	0	1	1	1	1	1	Output	
EDH Force	1	0	0	0	0	0	Input	
Test Pattern Select[0]	1	0	0	0	0	1	Input	
Test Pattern Select[1]	1	0	0	0	1	0	Input	
Test Pattern Select[2]	1	0	0	0	1	1	Input	
Test Pattern Select[3]	1	0	0	1	0	0	Input	
Test Pattern Select[4]	1	0	0	1	0	1	Input	
Test Pattern Select[5]	1	0	0	1	1	0	Input	
EDH Enable	1	0	0	1	1	1	Input	
TPG Enable	1	0	1	0	0	0	Input	I/O Port Bit 7
reserved	1	0	1	0	0	1	Input	
Chksum Attach In	1	0	1	0	1	0	Input	
reserved	1	0	1	0	1	1	Input	

Device Operation (Continued)

TABLE 6. I/O Configuration Register Addresses for Control Register Functions (Continued)

			Dit	Addrose	Din # SEI	[n]	in or regi		
Regis	ter Bit	[5]		121 F21	FIII # 3EL	. [11] [[4]	[0]	I/P or O/P	Power-On Status
	Enable	1	0	1	[4]	0	0	Input	
Dither Ena		1	0	1	1	0	1	Input	
Eroming E		1	0	1	1	1	0	Input	
		1	0	1	1	1	1	Input	
FIFO Inser		I	0	I	I	I	I	input	
Pin De	escriptio	ns							
Pin		Name		Descrip	tion				
1	V _{DDPLLD}			Positive	Power Su	pply Input	t (2.5V su	pply, PLL Logic)	
2	V _{SSPLLD}			Negative	e Power S	upply Inpu	ut (2.5V s	upply, PLL Logic)	
3	IO0			Multi-Fu	nction I/O	Port			
4	IO1			Multi-Fu	nction I/O	Port			
5	DV0			Parallel	Video Inp	ut (HD=Ch	nroma, SE	=Luma & Chroma	a)
6	DV1			Parallel	Video Inp	ut (HD=Ch	nroma, SE	=Luma & Chroma	a)
7	DV2			Parallel	Video Inp	ut (HD=Cł	nroma, SE	=Luma & Chroma	a)
8	DV3			Parallel	Video Inp	ut (HD=Ch	nroma, SE)=Luma & Chroma	a)
9	DV4			Parallel	Video Inp	ut (HD=Ch	nroma, SE	=Luma & Chroma	a)
10	V _{SSD}			Negative	Power S	upply Inpu	ut (2.5V s	upply, Digital Logic	c)
11	DV5			Parallel	Video Inpi	ut (HD=Ch	nroma, SE)=Luma & Chroma	a)
12	DV6			Parallel	Video Inpi	ut (HD=Ch	nroma, SE)=Luma & Chroma	a)
13	DV7			Parallel	Video Inpi	ut (HD=Ch	nroma, SE)=Luma & Chroma	a)
14	DV8			Parallel	Video Inpi	ut (HD=Ch	nroma, SE	=Luma & Chroma	a)
15	DV9			Parallel	Video Inpi	ut (HD=Ch	nroma, SE)=Luma & Chroma) 1)
16	Vood			Positive	Power Su	iuani vlaa	t (2.5V su	pply. Digital Logic)
17	Veed			Negative	Power S	upplv Inpi	ut (2.5V s	upply. Digital Logi	c)
18	DV10			Parallel	Video Inpi	ut (HD=Lu	ima)		- /
19	DV11			Parallel	Video Inpi	ut (HD=Lu	ima)		
20	DV12			Parallel	Video Inpi	ut (HD=Lu	ima)		
21	DV13			Parallel	Video Inpi	ut (HD=Lu	ima)		
22	DV14			Parallel	Video Inpi	ut (HD=Lu	ima)		
23	Varia			Positive	Power Su	nolv Input	t (3.3V su	nnly I/O)	
24	DV15			Parallel	Video Inni	ut (HD=I u	(0.01 00 (ma)	ppi), "O)	
25	DV16			Parallel	Video Inpi	ut (HD=Lu	ima)		
26	DV17			Parallel	Video Inni	ut (HD=Lu	ima)		
27	DV18			Parallel	Video Inni	ut (HD=Lu	ima)		
28	DV19			Parallel	Video Inni	ut (HD=Lu	ima)		
29	Vooro			Negative	Power S	upply Inni	ut (3.3V s	vlagu	
30	102			Multi-Fu	nction I/O	Port			
	103			Multi-Fu	nction I/O	Port			
32	104			Multi-Fu	nction I/O	Port			
33	105			Multi-Fu	nction I/O	Port			
34	106			Multi-Fu	nction I/O	Port			
35	107			Multi-Fu	nction I/O	Port			
36				Ancillian	/Control (Clock Incu	ıt		
37				Positivo	Power Su		(2 5\/ eu	nnly Digital Logic	1
20				Ancillion			. (2.0 V SU	איזי, דישונמו בטטונ	1
30				Ancilliar					
39				Ancilliary					
40	AD2			Ancilliary	//Control L	Jata Input			

Pin	Name	Description	
41	AD3	Ancilliary/Control Data Input	
42	AD4	Ancilliary/Control Data Input	
43	V _{SSD}	Negative Power Supply Input (2.5V supply, Digital Logic)	
44	AD5	Ancilliary/Control Data Input	
45	AD6	Ancilliary/Control Data Input	
46	AD7	Ancilliary/Control Data Input	
47	AD8	Ancilliary/Control Data Input	
48	AD9	Ancilliary/Control Data Input	
49	RD/WR	Ancilliary/Control Data Port Read/Write Control Input	
50	ANC/CTRL	Ancilliary/Control Data Port Function Control Input	
51	V _{DDSD}	Positive Power Supply Input (3.3V supply, Output Driver)	
52	R _{REF} PRE	Output Preemphasis Reference Resistor (4.75 KΩ, 1% Nom.)	
53	R _{REF} LVL	Output Level Reference Resistor (4.75 KΩ, 1% Nom.)	
54	V _{SSSD}	Negative Power Supply Input (3.3V supply, Output Driver)	
55	V _{SSSD}	Negative Power Supply Input (3.3V supply, Output Driver)	
56	SDO	Serial Data True Output	
57	V _{DDLS}	Positive Power Supply Input (3.3V supply, Level Shift)	
58	SDO	Serial Data Complement Output	
59	V _{SSLS}	Negative Power Supply Input (3.3V supply, Level Shift)	
60	V _{DDZ}	Positive Power Supply Input (2.5V supply, Serializer)	
61	V _{SSPLLA}	Negative Power Supply Input (2.5V supply, PLL Analog)	
62	V _{DDPLLA}	Positive Power Supply Input (2.5V supply, PLL Analog)	
63	VCLK	Video Data Clock Input	
64	Reset	Manual Reset Input (High True)	

Note: All LVCMOS inputs have internal pull-down devices.

Application Circuit

Application Information

A typical application circuit for the CLC030 is shown in the Application Circuit diagram. This circuit demonstrates the capabilities of the CLC030 and allows its evaluation in a native configuration. An assembled demonstration board is available, part number SD030EVK. The board may be ordered through any of National's sales offices. Complete circuit board layouts and schematics, including Gerber photoplot files, for the SD030EVK are available on National's WEB site in the application information for this device. For latest availability information, please see: www.national.com/appinfo/interface.

PCB Layout and Power System Bypass Recommendations

Circuit board layout and stack-up for the CLC030 should be designed to provide noise-free power to the device. Good layout practice also will separate high frequency or high level inputs and outputs to minimize unwanted stray noise pickup, feedback and interference. Power system performance may be greatly improved by using thin dielectrics (4 to 10 mils) for power/ground sandwiches. This increases the intrinsic capacitance of the PCB power system which improves power supply filtering, especially at high frequencies, and makes the value and placement of external bypass capacitors less critical. External bypass capacitors should include both RF ceramic and tantalum electrolytic types. RF capacitors may use values in the range 0.01 µF to 0.1 µF. Tantalum capacitors may be in the range 2.2 µF to 10 µF. Voltage rating for tantalum capacitors should be at least 5X the power supply voltage being used. It is recommended practice to use two vias at each power pin of the CLC030 as well as all RF bypass capacitor terminals. Dual vias reduce the interconnect inductance by up to half, thereby reducing interconnect inductance and extending the effective frequency range of the bypass components.

The outer layers of the PCB may be flooded with additional $V_{\rm SS}$ (ground) plane. These planes will improve shielding and isolation as well as increase the intrinsic capacitance of the

power supply plane system. Naturally, to be effective, these planes must be tied to the V_{SS} power supply plane at frequent intervals with vias. Frequent via placement also improves signal integrity on signal transmission lines by providing short paths for image currents which reduces signal distortion. The planes should be pulled back from all transmission lines and component mounting pads a distance equal to the width of the widest transmission line or the thickness of the dielectric separating the transmission line from the internal power or ground plane(s) whichever is greater. Doing so minimizes effects on transmission line impedances and reduces unwanted parasitic capacitances at component mounting pads.

The CLC030 uses two power supply voltages, 2.5 and 3.3 volts. These supplies connect to the device through seven sets of independent power input pins. The function and system supplied through these is given in the Pin Description Table. The power supply voltages normally share a common 0 volt or ground return system. Either a split plane or separate power planes can be used to supply the positive voltages to the device.

In especially noisy power supply environments, such as is often the case when using switching power supplies, separate filtering may be used at the CLC030's PLL analog, PLL digital and serial output driver power pins. The CLC030 was designed for this situation. The digital section, PLL and output driver power supply feeds are independent. See the Pin Description Table and the Connection Diagram for details. Supply filtering may take the form of L-section or pisection, L-C filters in series with these V_{DD} inputs. Such filters are available in a single package from several manufacturers. Despite being independent feeds, all device power supplies should be applied simultaneously as from a common source. The CLC030 is free from power supply latch-up caused by circuit-induced delays between the device's separate power feed systems.



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