

## The RF MOSFET Line

# RF Power Field Effect Transistors

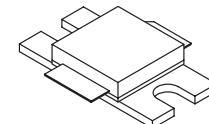
## N-Channel Enhancement-Mode Lateral MOSFETs

Designed for PCN and PCS base station applications with frequencies from 2.0 to 2.2 GHz. Suitable for FM, TDMA, CDMA and multicarrier amplifier applications. To be used in Class AB for PCN-PCS/cellular radio and WLL applications.

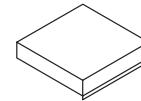
- Wideband CDMA Performance: -45 dB ACPR @ 4.096 MHz, 28 Volts  
Output Power — 3.5 Watts  
Power Gain — 14 dB  
Efficiency — 15%
- High Gain, High Efficiency and High Linearity
- Integrated ESD Protection
- Designed for Maximum Gain and Insertion Phase Flatness
- Capable of Handling 10:1 VSWR, @ 28 Vdc, 2.11 GHz, 30 Watts CW Output Power
- Excellent Thermal Stability
- Characterized with Series Equivalent Large-Signal Impedance Parameters
- In Tape and Reel. R3 Suffix = 250 Units per 32 mm, 13 Inch Reel.

**MRF21030R3**  
**MRF21030SR3**

**2.2 GHz, 30 W, 28 V**  
**LATERAL N-CHANNEL**  
**RF POWER MOSFETs**



CASE 465E-03, STYLE 1  
NI-400  
MRF21030R3



CASE 465F-03, STYLE 1  
NI-400S  
MRF21030SR3

### MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Drain-Source Voltage	V <sub>DSS</sub>	65	Vdc
Gate-Source Voltage	V <sub>GS</sub>	-0.5, +15	Vdc
Total Device Dissipation @ T <sub>C</sub> = 25°C Derate above 25°C	P <sub>D</sub>	83.3 0.48	Watts W/°C
Storage Temperature Range	T <sub>stg</sub>	-65 to +200	°C
Operating Junction Temperature	T <sub>J</sub>	200	°C

### ESD PROTECTION CHARACTERISTICS

Test Conditions	Class
Human Body Model	2 (Minimum)
Machine Model	M3 (Minimum)

### THERMAL CHARACTERISTICS

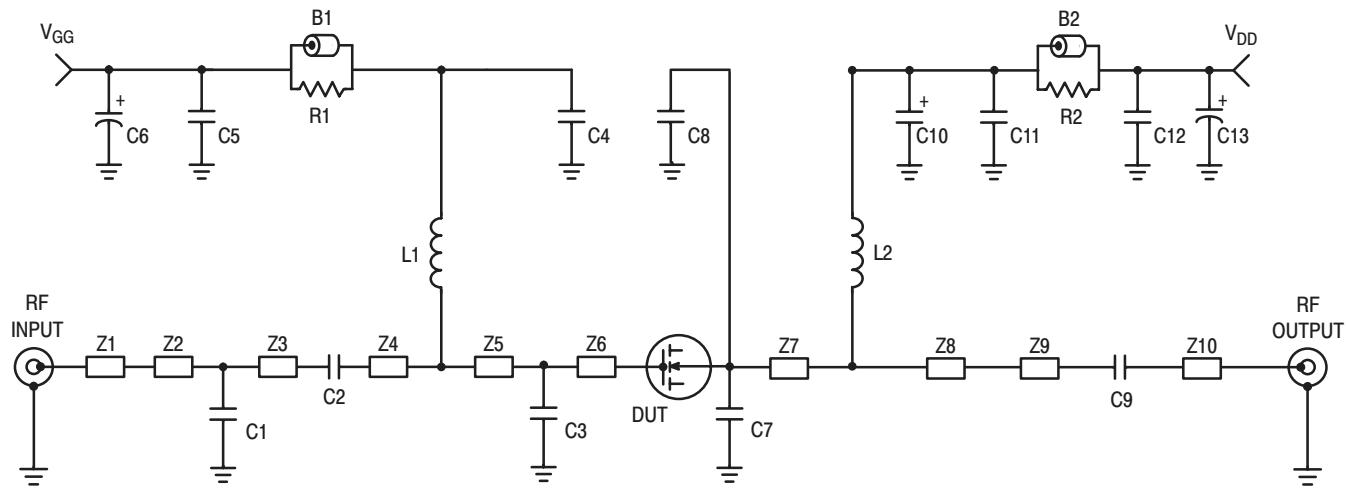
Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Case	R <sub>θJC</sub>	2.1	°C/W

NOTE – **CAUTION** – MOS devices are susceptible to damage from electrostatic charge. Reasonable precautions in handling and packaging MOS devices should be observed.

**ELECTRICAL CHARACTERISTICS** ( $T_C = 25^\circ\text{C}$  unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
<b>OFF CHARACTERISTICS</b>					
Drain–Source Breakdown Voltage ( $V_{GS} = 0 \text{ Vdc}$ , $I_D = 20 \mu\text{A}$ )	$V_{(BR)DSS}$	65	—	—	Vdc
Zero Gate Voltage Drain Current ( $V_{DS} = 28 \text{ Vdc}$ , $V_{GS} = 0 \text{ Vdc}$ )	$I_{DSS}$	—	—	1	$\mu\text{Adc}$
Gate–Source Leakage Current ( $V_{GS} = 5 \text{ Vdc}$ , $V_{DS} = 0 \text{ Vdc}$ )	$I_{GSS}$	—	—	1	$\mu\text{Adc}$
<b>ON CHARACTERISTICS</b>					
Gate Threshold Voltage ( $V_{DS} = 10 \text{ Vdc}$ , $I_D = 100 \mu\text{Adc}$ )	$V_{GS(\text{th})}$	2	3	4	Vdc
Gate Quiescent Voltage ( $V_{DS} = 28 \text{ Vdc}$ , $I_D = 250 \text{ mA}$ )	$V_{GS(Q)}$	2	3.3	4.5	Vdc
Drain–Source On–Voltage ( $V_{GS} = 10 \text{ Vdc}$ , $I_D = 1 \text{ Adc}$ )	$V_{DS(\text{on})}$	—	0.29	0.4	Vdc
Forward Transconductance ( $V_{DS} = 10 \text{ Vdc}$ , $I_D = 1 \text{ Adc}$ )	$g_{fs}$	—	2	—	S
<b>DYNAMIC CHARACTERISTICS</b>					
Input Capacitance (Including Input Matching Capacitor in Package) (1) ( $V_{DS} = 28 \text{ Vdc}$ , $V_{GS} = 0$ , $f = 1 \text{ MHz}$ )	$C_{iss}$	—	98.5	—	pF
Output Capacitance (1) ( $V_{DS} = 28 \text{ Vdc}$ , $V_{GS} = 0$ , $f = 1 \text{ MHz}$ )	$C_{oss}$	—	37	—	pF
Reverse Transfer Capacitance ( $V_{DS} = 28 \text{ Vdc}$ , $V_{GS} = 0$ , $f = 1 \text{ MHz}$ )	$C_{rss}$	—	1.3	—	pF
<b>FUNCTIONAL TESTS</b> (In Motorola Test Fixture, 50 ohm system)					
Two-Tone Common–Source Amplifier Power Gain ( $V_{DD} = 28 \text{ Vdc}$ , $P_{out} = 30 \text{ W PEP}$ , $I_{DQ} = 250 \text{ mA}$ , $f_1 = 2140.0 \text{ MHz}$ , $f_2 = 2140.1 \text{ MHz}$ )	$G_{ps}$	—	13	—	dB
Two-Tone Drain Efficiency ( $V_{DD} = 28 \text{ Vdc}$ , $P_{out} = 30 \text{ W PEP}$ , $I_{DQ} = 250 \text{ mA}$ , $f_1 = 2140.0 \text{ MHz}$ , $f_2 = 2140.1 \text{ MHz}$ )	$\eta$	—	33	—	%
3rd Order Intermodulation Distortion ( $V_{DD} = 28 \text{ Vdc}$ , $P_{out} = 30 \text{ W PEP}$ , $I_{DQ} = 250 \text{ mA}$ , $f_1 = 2140.0 \text{ MHz}$ , $f_2 = 2140.1 \text{ MHz}$ )	IMD	—	-30	—	dBc
Input Return Loss ( $V_{DD} = 28 \text{ Vdc}$ , $P_{out} = 30 \text{ W PEP}$ , $I_{DQ} = 250 \text{ mA}$ , $f_1 = 2140.0 \text{ MHz}$ , $f_2 = 2140.1 \text{ MHz}$ )	IRL	—	-13	—	dB
Two-Tone Common–Source Amplifier Power Gain ( $V_{DD} = 28 \text{ Vdc}$ , $P_{out} = 30 \text{ W PEP}$ , $I_{DQ} = 250 \text{ mA}$ , $f_1 = 2110.0 \text{ MHz}$ , $f_2 = 2110.1 \text{ MHz}$ and $f_1 = 2170.0 \text{ MHz}$ , $f_2 = 2170.1 \text{ MHz}$ )	$G_{ps}$	12	13	—	dB
Two-Tone Drain Efficiency ( $V_{DD} = 28 \text{ Vdc}$ , $P_{out} = 30 \text{ W PEP}$ , $I_{DQ} = 250 \text{ mA}$ , $f_1 = 2110.0 \text{ MHz}$ , $f_2 = 2110.1 \text{ MHz}$ and $f_1 = 2170.0 \text{ MHz}$ , $f_2 = 2170.1 \text{ MHz}$ )	$\eta$	31	33	—	%
3rd Order Intermodulation Distortion ( $V_{DD} = 28 \text{ Vdc}$ , $P_{out} = 30 \text{ W PEP}$ , $I_{DQ} = 250 \text{ mA}$ , $f_1 = 2110.0 \text{ MHz}$ , $f_2 = 2110.1 \text{ MHz}$ and $f_1 = 2170.0 \text{ MHz}$ , $f_2 = 2170.1 \text{ MHz}$ )	IMD	—	-30	-27.5	dBc
Input Return Loss ( $V_{DD} = 28 \text{ Vdc}$ , $P_{out} = 30 \text{ W PEP}$ , $I_{DQ} = 250 \text{ mA}$ , $f_1 = 2110.0 \text{ MHz}$ , $f_2 = 2110.1 \text{ MHz}$ and $f_1 = 2170.0 \text{ MHz}$ , $f_2 = 2170.1 \text{ MHz}$ )	IRL	—	-13	-9	dB
Output Mismatch Stress ( $V_{DD} = 28 \text{ Vdc}$ , $P_{out} = 30 \text{ W CW}$ , $I_{DQ} = 250 \text{ mA}$ , $f = 2110 \text{ MHz}$ , $\text{VSWR} = 10:1$ , All Phase Angles at Frequency of Tests)	$\Psi$	No Degradation In Output Power Before and After Test			

(1) Part is internally matched both on input and output.



B1, B2      Short Ferrite Beads  
 C1      1 pF Chip Capacitor  
 C2      4.7 pF Chip Capacitor  
 C3      0.5 pF Chip Capacitor  
 C4      3.9 pF Chip Capacitor  
 C5, C12      0.1  $\mu$ F Chip Capacitors  
 C6, C13      470  $\mu$ F, 63 V Electrolytic Chip Capacitors  
 C7, C8      0.3 pF Chip Capacitors  
 C9      3.6 pF Chip Capacitor  
 C10      22  $\mu$ F Tantalum Chip Capacitor  
 C11      5.1 pF Chip Capacitor  
 L1, L2      12.5 nH Inductors  
 R1, R2      12  $\Omega$  Chip Resistors (1206)

Z1      0.153" x 0.087" Microstrip  
 Z2      0.509" x 0.156" Microstrip  
 Z3      0.572" x 0.087" Microstrip  
 Z4      0.509" x 0.232" Microstrip  
 Z5      0.277" x 0.143" Microstrip  
 Z6      0.200" x 0.305" Microstrip  
 Z7      0.200" x 0.511" Microstrip  
 Z8      0.510" x 0.328" Microstrip  
 Z9      0.608" x 0.081" Microstrip  
 Board      0.030" Glass Teflon<sup>®</sup>, TLX8-0300  
 Taconix ( $\epsilon_r = 2.55$ )

Figure 1. MRF21030 Test Circuit Schematic

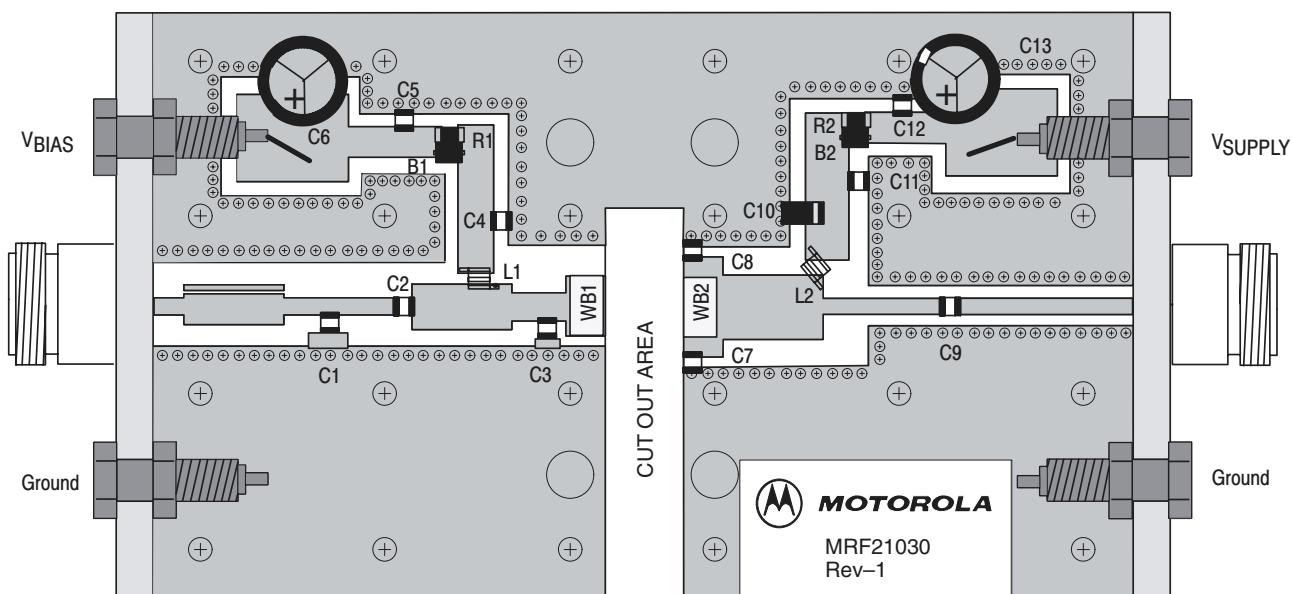
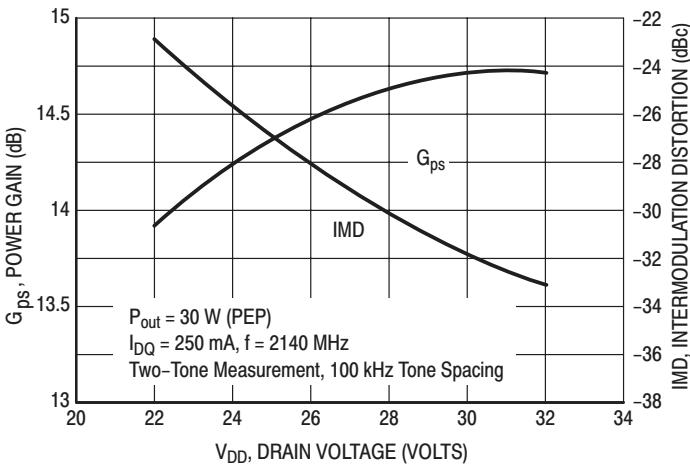
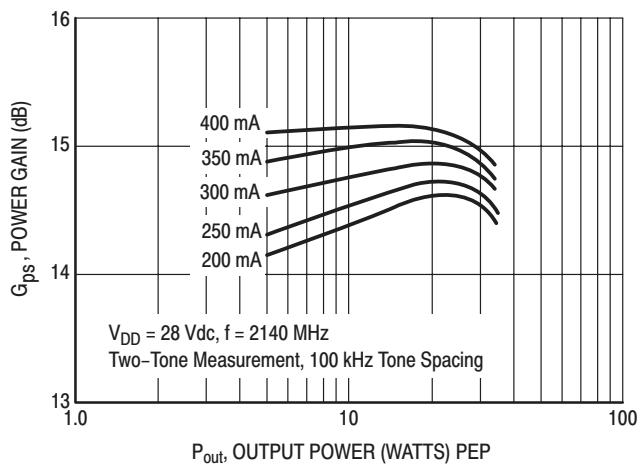
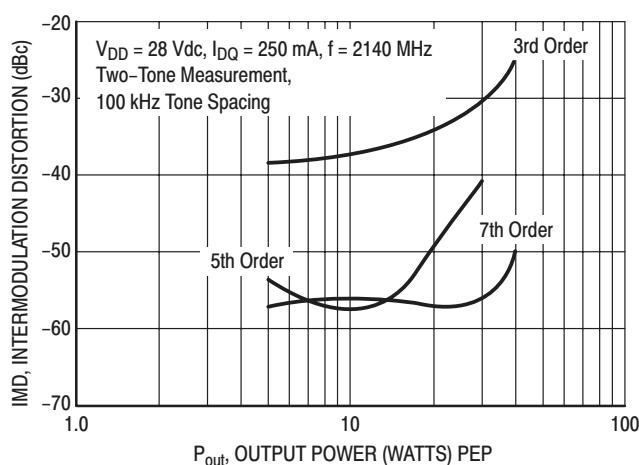
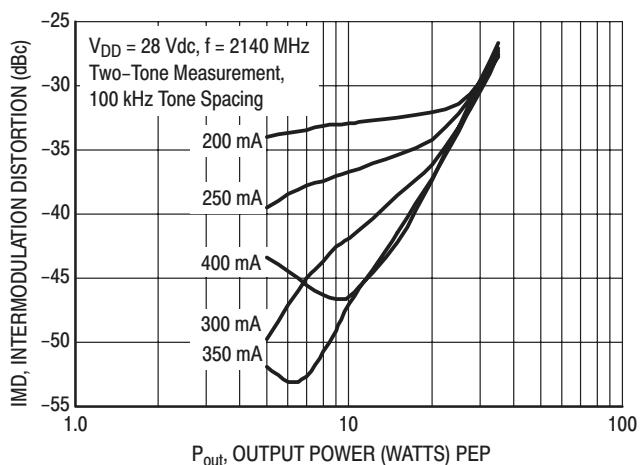
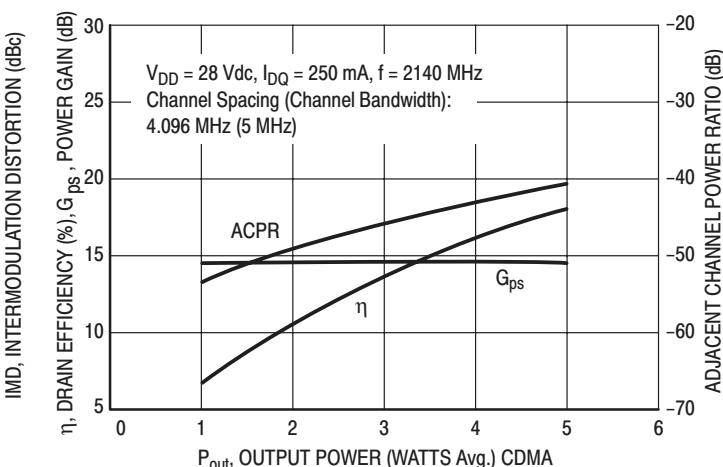
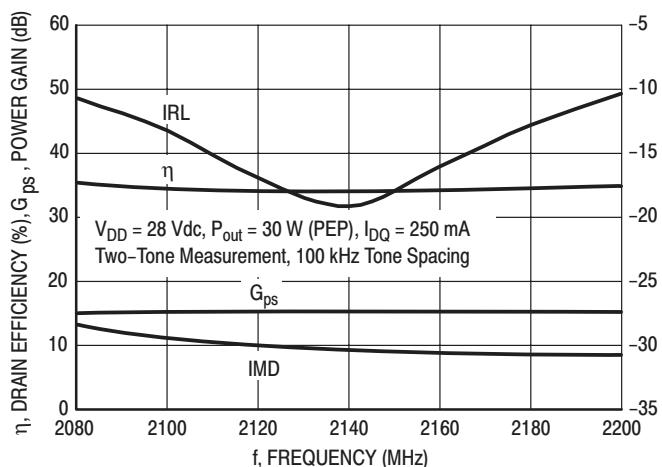
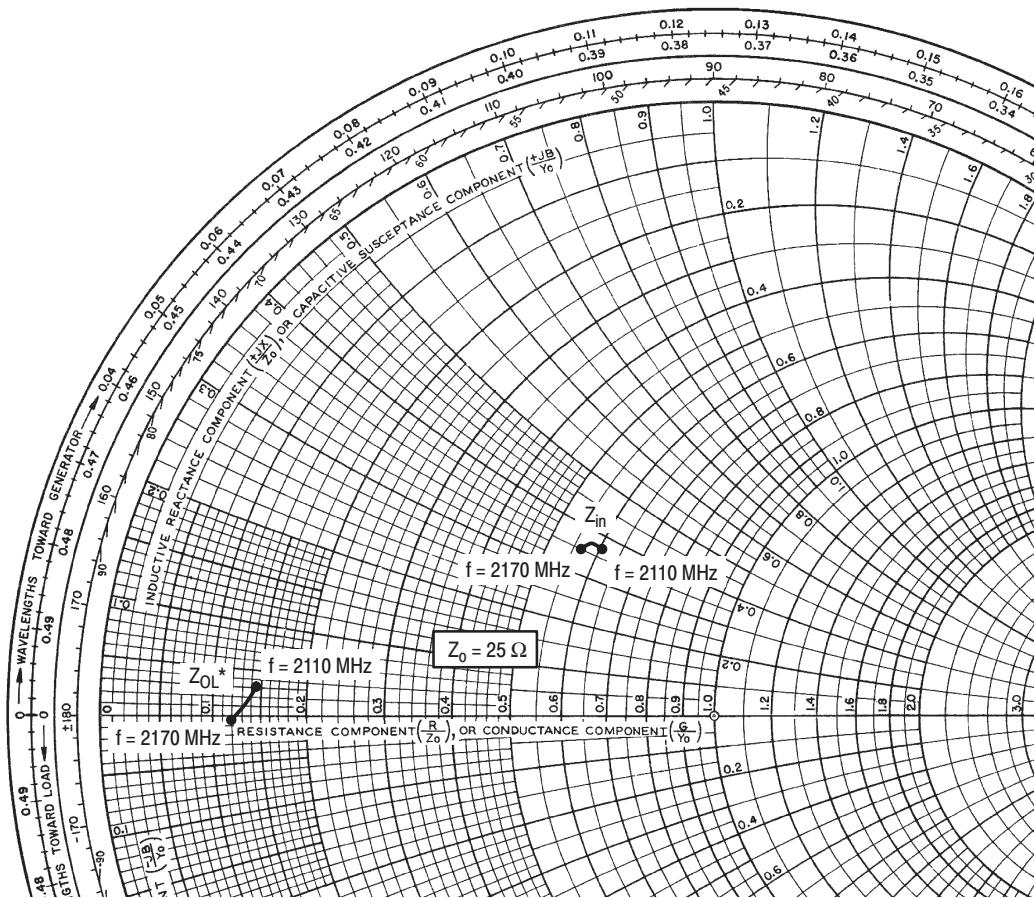


Figure 2. MRF21030 Test Circuit Component Layout

## TYPICAL CHARACTERISTICS





$V_{DD} = 28 \text{ V}$ ,  $I_{DQ} = 250 \text{ mA}$ ,  $P_{out} = 30 \text{ W PEP}$

$f$ MHz	$Z_{in}$ $\Omega$	$Z_{OL^*}$ $\Omega$
2110	$15.3 + j9.4$	$3.7 + j0.78$
2140	$14.6 + j9.4$	$3.4 + j0.37$
2170	$14.3 + j8.8$	$3.0 - j0.13$

$Z_{in}$  = Complex conjugate of source impedance.

$Z_{OL^*}$  = Complex conjugate of the optimum load impedance at a given output power, voltage, IMD, bias current and frequency.

Note:  $Z_{OL^*}$  was chosen based on tradeoffs between gain, output power, drain efficiency and intermodulation distortion.

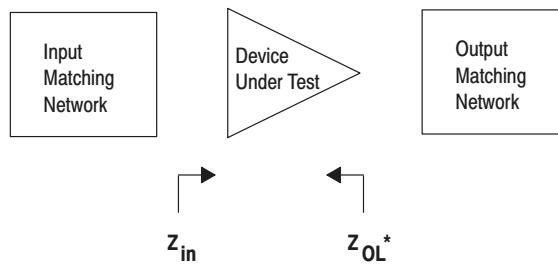
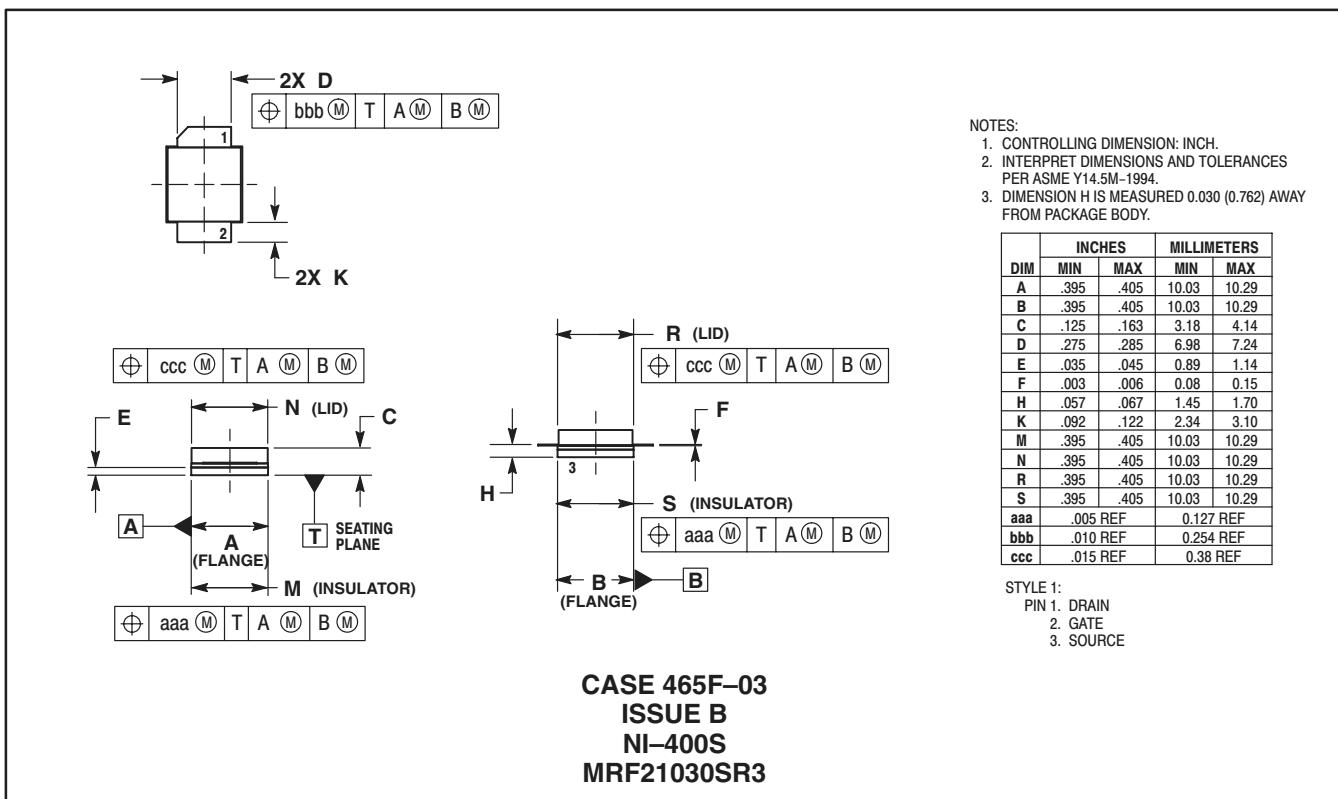
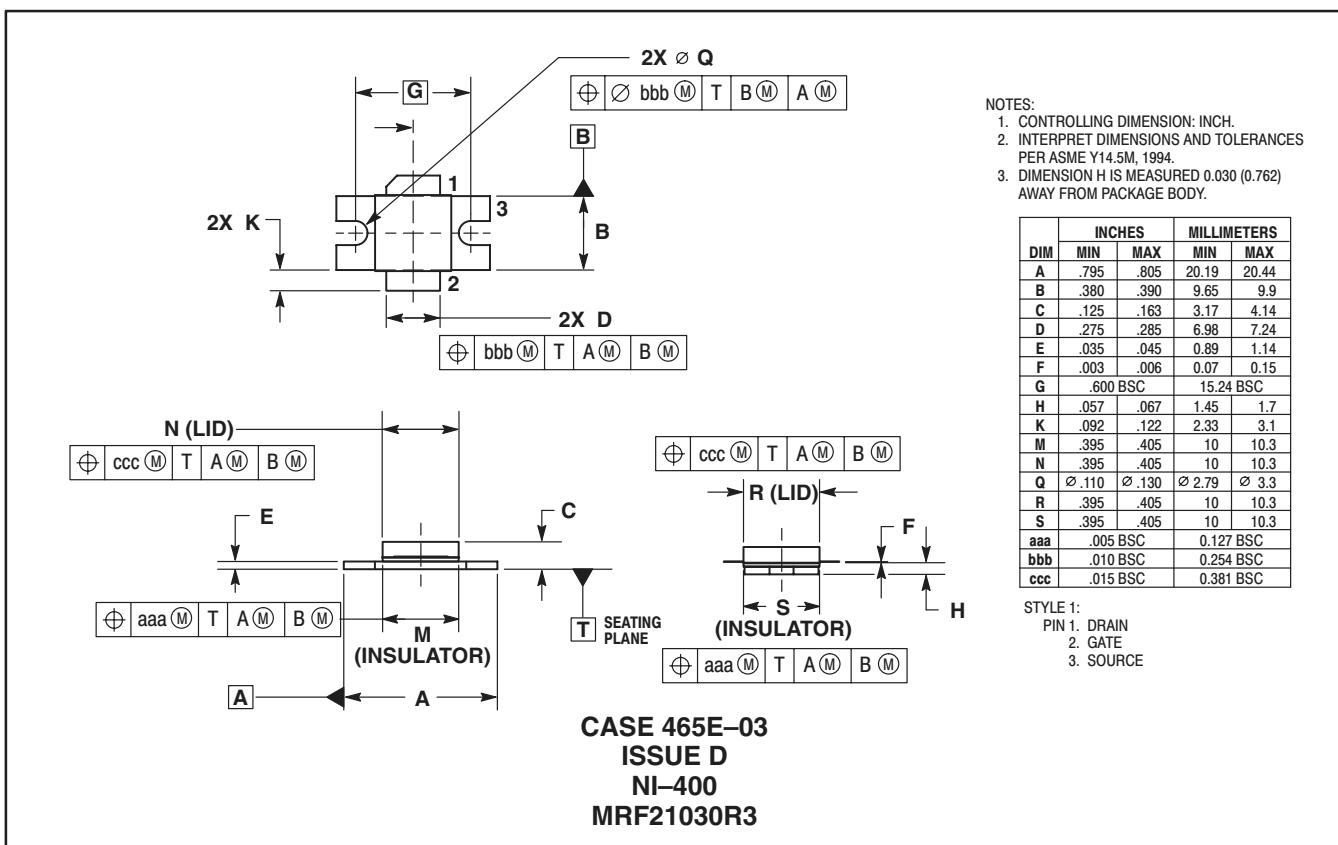


Figure 9. Series Equivalent Input and Output Impedance

# **NOTES**

## PACKAGE DIMENSIONS



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