

**AMIS-53000 Frequency Agile Transceiver  
Data Sheet**

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## 1.0 Overview of the AMIS-53000

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The AMIS-53000 is the latest highly flexible member of AMI Semiconductor's ASTRIC™ family of single-chip wireless transceivers. It is ideally suited for low to moderate data rate, low power, sub 1GHz, narrow band, FSK/GFSK/OOK, multiple channel, wireless applications in the medical, automotive and industrial markets. The AMIS-53000 can easily be interfaced to a base band processor via a serial interface bus.

### 1.1 Applications for the AMIS-53000

- Medical Implantable Communication Systems (MICS)
- Wireless Medical Telemetry Systems
- Wireless Sensors
- RFID
- Remote Monitoring
- Access Control and Security
- Keyless Entry
- Mobile Wireless Data Terminals
- Keyless Entry
- Tire Pressure Monitors
- Wireless Toys

### 1.2 Key Features

- Medical implant communication protocol support
- Very low power single-chip CMOS transceiver
- Patented Quick Start crystal oscillator
- Low power receive Sniff Mode™
- Periodic transmit using Burst mode
- Internal low power 10kHz oscillator
- Internal self calibration functions
- SPI/I<sup>2</sup>C interface bus
- 3-wire/4-wire serial data interface
- Two analog to digital converter channels
- Internal fractional N frequency synthesizer
- On/off shift key/frequency shift key modulation/Gaussian FSK (BT = 1)
- Internal temperature sensor
- Minimal external components

### 1.3 Technical Features

- Operating voltage range: 2.2 to 3.3V
- Operating temperature range: -40° to +85°C
- Operating frequency range: 300 to 928MHz
- Data rate:
  - 1 to 19.2kbps (OOK)
  - 2 to 128kbps (FSK/GFSK)
- Transmit output power:
  - +15dBm max (high power)
  - +0dBm max (low power)
- Transmit current: 50mA typical (15dBm)

- Receiver sensitivity
  - -115dBm (OOK @ 1kbps)
  - -105dBm (FSK @ 20kbps)
- Receiver current: 12mA (continuous)
- Minimum RX energy detect time: 130uS (Sniff)

## 1.4 Circuit Overview

### 1.4.1. Transmitter

The AMIS-53000 uses a driver and class E power amplifier to output the on/off shift keyed or frequency shift keyed RF waveforms. The class E power amplifier has two output power ranges allowing more efficient output power for the one setting up to 0dBm output and the other setting for output power greater than 0dBm. The class E power amplifier can achieve output power of +12dBm to +15dBm for frequencies in the range of 300MHz to 915MHz. The output power is programmable in each of the two output power bands.

The transmit data can be NRZ or Manchester encoded. Data can also be modulated as on/off shift keyed or frequency shift keyed. Data rates for the OOK modulation can be as high as 19.2kbps. Data rates for the FSK/GFSK modulation can be as high as 128kbps. The carry frequency deviation for the FSK modulation is programmable, typically one half to one times the data rate.

The transmit data output can be wave shaped with a Gaussian format. This can reduce the occupied bandwidth of the signal.

### 1.4.2. Receiver

The AMIS-53000 has a single receiver channel and a single transmit channel, which can be connected to individual antennas or can be combined into a single antenna. The receiver uses four different methods to receive and recover data that has been on/off shift keyed or frequency shift keyed modulated. The FSK/GFSK data is recovered using either a PLL circuit or a FFT circuit along with a CDR circuit. The OOK data is recovered using an RSSI circuit along an optional CDR circuit. It is suggested that the CDR circuit be used when receiving OOK signals.

#### 1.4.2.1. On/Off Shift Key Modulation

The AMIS-53000 uses a logarithmic received signal strength indicator (RSSI) detector to recover the data from the on/off shift keyed modulated waveform. Data rates can be up to 19.2kbps. The AMIS-53000 has eight discrete data filters for common baud rates. The receiver can detect either NRZ or Manchester encoded data.

#### 1.4.2.2. Low Data Rate Frequency Shift Key Modulation

The AMIS-53000 uses a digital PLL detector to recover the data from the frequency shift keyed data below 20kbps. The recovered data waveform is applied to the clock and data recovery circuit to produce the digital data and a synchronized clock. The receiver can detect either NRZ or Manchester encoded data.

#### 1.4.2.3. High Data Rate Frequency Shift Key Modulation

The AMIS-53000 uses a fast fourier transform (FFT) to recover data from frequency shift keyed modulated waveforms when the data rate is higher than 20kbps. The data rate can be as high as 128kbps. The demodulated data waveform is applied to the clock and data recovery circuit to produce the digital data and a synchronized clock. The receiver can detect either NRZ or Manchester encoded data.

#### 1.4.2.4. Clock and Data Recovery

The AMIS-53000 uses a clock and data recovery circuit along with the frequency shift keyed or on/off shift keyed data detector circuits to recover the data stream. The CDR circuit synchronizes a clock with the data rate of the received data. This same circuit can be used with the on/off shift keyed waveform.

#### 1.4.2.5. Manchester Data Encoding

The AMIS-53000 can encode the data as NRZ or Manchester.

#### 1.4.2.6. Oscillators

The AMIS-53000 requires a single external crystal working with the internal VCO and PLL to generate frequencies from 300MHz to 928MHz. The AMIS-53000 has internal capacitors that eliminate the need for external load capacitors when using a typical 24MHz external crystal. The VCO requires an external inductor and capacitor (including internal capacitance) to produce the desired transmit or receive RF frequency. The AMIS-53000 generates the desired RF transmit and receive frequencies from 300MHz to 928MHz by selecting the proper inductor and capacitor value along with programming the frequency in the AMIS-53000. A patented Quick Start circuit is used to force the crystal oscillator on to the desired frequency in microseconds rather than in milliseconds.

A low power internal 10kHz oscillator provides the timing for Sniff, Burst and housekeeping. The AMIS-53000 self-calibration circuits can tune this oscillator to within two percent of 10kHz.

#### 1.4.2.7. Interface Serial Bus

The AMIS-53000 has separate interfaces for data and control. The transfer of TX/RX data between the AMIS-53000 and an external host/controller is done with a 3-wire serial interface or a 4-wire SPI compatible serial interface. Control information is written to the AMIS-53000 registers or read from the AMIS-53000 registers using either a 3-wire serial interface or a 2-wire I<sup>2</sup>C compatible serial interface.

Once the AMIS-53000 configuration registers have data written to them for various operational modes such as, TX, RX, Sniff or other, placing the AMIS-53000 into one of these modes is accomplished through a single write to the command register.

#### TX/RX Data Interface

The transmit or receive data interface of the AMIS-53000 can be programmed to be either a proprietary 3-wire serial interface or a 4-wire SPI compatible serial bus. The data interface can be set up to do either data transfers into a buffer in the AMIS-53000 or streaming data (data is transmitted as it is received by the AMIS-53000 or data is sent to the host/controller as it is recovered in the AMIS-53000 receiver). When using the buffered data mode, the AMIS-53000 can be the master or slave, but it must be the master to do streaming data.

#### Control Interface

Once the AMIS-53000 is first powered on, an external host/controller sets the type of interface to the AMIS-53000 (3-wire or I<sup>2</sup>C) by simply writing to the AMIS-53000 with the desired protocol. The AMIS-53000 will continue to use that interface protocol until power is removed from the AMIS-53000. The AMIS-53000 is always a slave device for the control interface.

## 2.0 Operational Specifications

### 2.1 Absolute Maximum Ratings

Table 1: Absolute Maximum Ratings

| Symbol             | Parameter            | Min.  | Max.    | Units | Notes  |
|--------------------|----------------------|-------|---------|-------|--------|
| Vdd                | DC Supply Voltage    | -0.3  | 3.6     | V     |        |
| Vin                | Input Pin Voltage    | -0.3  | Vdd+0.3 | V     |        |
| Iin                | Input Pin Current    | -10.0 | 10.0    | mA    | 25C    |
| Tstrg              | Storage Temperature  | -55   | 150     | C     | SSOP   |
| Tlead              | Lead Temperature     |       | 300     | C     | 10 SEC |
| ESD <sub>HBM</sub> | Human Body Model     |       | 2       | kV    |        |
| ESD <sub>CDM</sub> | Charged-Device Model |       | 750     | V     |        |
| ESD <sub>MM</sub>  | Machine Model        |       | 200     | V     |        |

### 2.2 Recommended Operating Conditions

Table 2: Operating Conditions

| Symbol | Parameter                     | Min. | Max. | Units | Notes             |
|--------|-------------------------------|------|------|-------|-------------------|
| Vdd    | DC Supply                     | 2.2  | 3.3  | V     |                   |
| Idd    | Dynamic Current               |      | 70   | mA    | (1) Continuous TX |
| Idds   | Standby Current (off current) |      | 2    | uA    | (2)               |
| Vss    | Ground                        | 0.0  | 0.0  | V     |                   |
| Ta     | Ambient Temperature           | -40  | 85   | C     |                   |

Notes:

- Dynamic current is with the transmitter enabled at maximum output power + 15dBm in FSK mode at 928MHz.
- Standby current is with all analog cells in power down. Other logic powered up with no clocks running. All outputs unloaded and inputs tied high or low. No floating inputs.

#### 2.2.1. Parametric Voltage and Current Levels

(Testing for the below currents assumes a static test setup with measurements performed while static data is applied to the device.)

##### 2.2.1.1. Inputs

Table 3: Pin Input Parameters

| Pin  | Vil      |          | Vih      |          | Iil (1) |         | Iih (1) |         | Notes (2)                 |
|------|----------|----------|----------|----------|---------|---------|---------|---------|---------------------------|
|      | Min. (V) | Max. (V) | Min. (V) | Max. (V) | Min. uA | Max. uA | Min. uA | Max. uA |                           |
| AI   |          |          |          |          | 0.0     | 1.0     | -1.0    | 0.0     | Analog input              |
| DISU | 0.0      | 0.3      | 0.8      | 1.0      | 0.0     | 1.0     | -30     | -90     | CMOS with pull up Schmitt |
| DISC | 0.0      | 0.3      | 0.8      | 1.0      | 0.0     | 1.0     | -1.0    | 0.0     | CMOS Schmitt              |

Notes:

- Iil and Iih are tested at Vdd = VDDmax volts. Not tested at less than room temperature.
- PU = Pull up, PD = Pull down, SC = Schmitt, SU = Schmitt & Pull up and SD = Schmitt and Pull down.
- CMOS values are 'Vin \* VDD' and TTL values are absolute voltages.



### 2.2.1.2. Outputs

Table 4: Pin Output Parameters

| Pin | Vol (1)  |          | Voh (2)  |          | Iol (1,3) |         | Ioh (2,3) |         | Notes          |
|-----|----------|----------|----------|----------|-----------|---------|-----------|---------|----------------|
|     | Min. (V) | Max. (V) | Min. (V) | Max. (V) | Min. mA   | Max. mA | Min. mA   | Max. mA |                |
| AO  |          |          |          |          |           |         |           |         | Analog outputs |
| DO  | 0        | 0.4      | Vdd-4    |          | 2         |         | -2        |         | CMOS           |

Notes:

1. Vol, Iol are tested at Vdd = VDDmin volts unless otherwise stated.
2. Voh, Ioh are tested at Vdd = VDDmin volts unless otherwise stated.
3. Polarity on currents indicates direction of current: (+) for sinking and (-) for sourcing.

### 2.2.1.3. I/O Pins

Table 5: I/O Pin Parameters

| Pin | Vil V Min. | Vil V Max. | Vih V Min. | Vih V Max. | Vol V (1) | Vol V Max. (1) | Voh V Min. (2) | Voh V (2) | Iol mA Min. (1) | Ioh mA Min. (2) | Iozl uA Max. (3) | Iozh uA Max. (3) | Notes   |
|-----|------------|------------|------------|------------|-----------|----------------|----------------|-----------|-----------------|-----------------|------------------|------------------|---------|
| AIO |            |            |            |            |           |                |                |           |                 |                 |                  |                  |         |
| DIO | 0          | 0.3        | 0.8        | 1          | 0         | 0.4            | Vdd-4          | Vdd       | 2               | -2              | 1                | -1               | Schmitt |

Notes:

1. Vol, Iol are tested at Vdd = 3.1 volts.
  2. Voh, Ioh are tested at Vdd = 3.1 volts.
  3. Ioz is tested with Vdd = 3.5 volts.
- \*\*\* Leakage on I/O pins is typically checked for +/- 10 microamps with the output device turned off and no PU or PD device present.

## 2.3 Operational Specifications

Table 6: Operational Specifications

| Parameter                              | Min.          | Typ.   | Max.  | Units | Comment  |
|--|---------------|--------|-------|-------|--|
| <b>Receiver</b>                        |               |        |       |       |  |
| Frequency Range                        | 300           |        | 928   | MHz   |  |
| Sensitivity                            | -107          |        | -114  | dBm   | @ 10kHz data rate (FSK/GFSK modulation)                            |
|  | -104          |        | -111  | dBm   | @ 10kHz data rate (OOK modulation)                                 |
| Noise Figure                           | 6.0           | 7.8    | 9.0   | dB    |  |
| IIP2                                   |               | +60    |       | dBm   | Dual tone test using RSSI  |
| IIP3                                   |               | +5     |       | dBm   | Dual tone test using RSSI  |
| Image Rejection                        | 30            | 40     | 50    | dB    | Modulated desired, single tone interferer                          |
| Input Impedance                        |               | 15-j35 |       | Ω     | @ 900MHz series equivalent   |
|  |               | 72-j62 |       | Ω     | @ 433MHz series equivalent   |
| RSSI Gain                              | 14            | 16     | 18    | mV/dB |  |
| I <sub>IN</sub>                        |               | 8      |       | mA    | Receiver current consumption at 900MHz                             |
| I <sub>SB</sub>                        | Full Shutdown | 2      |       | uA    | Standby current (no clocks enabled)                                |
|  | Crystal Mode  | 2      |       | mA    | System clock output enabled (6MHz)                                 |
| T <sub>on</sub>                        |               | 100    |       | us    | Standby to receiver on time  |
| T <sub>RX_TX</sub> /T <sub>TX_RX</sub> |               | 100    |       | us    | Transition time from RX to TX or TX to RX                          |
| <b>LAN</b>                             |               |        |       |       |  |
| Input Trim                             | 1.2           |        | 4     | pF    | Internal capacitor range for the receiver input                    |
| Output Trim                            | 0.32          |        | 0.912 | pF    | Internal capacitor range for the output of the LNA in the receiver |

Table 6: Operational Specifications (Continued)

| Parameter                 | Min.            | Typ. | Max.             | Units            | Comment   |                                      |
|---------------------------|-----------------|------|------------------|------------------|---|--------------------------------------|
| <b>Transmitter</b>        |                 |      |                  |                  |   |                                      |
| Frequency Range           | 300             |      | 928              | MHz              |   |                                      |
| T <sub>on</sub>           |                 | 100  |                  | us               | Standby to transmitter on time                      |                                      |
| T <sub>TX_RX</sub>        |                 | 100  |                  | us               | Transition time from TX to RX                       |                                      |
| OOK On/Off Ratio          |                 | 60   |                  | dB               |   |                                      |
| FSK Frequency Separation  | 0               |      | 200              | kHz              | Allowable transmit/receive peak deviation           |                                      |
| CW Output Power           | -20             |      | 15               | dBm              | Range of output power in the high power mode        |                                      |
|                           | -25             |      | 0                | dBm              | Range of output power in the low power mode         |                                      |
| P <sub>HARMONICS</sub>    |                 | 35   |                  | dBc              | With complete matching network                      |                                      |
| <b>TX PA</b>              |                 |      |                  |                  |   |                                      |
| Output Cap.               | 2               |      | 7.5              | pF               | Internal capacitor range for the PA adjustable trim |                                      |
| Output Switch R           | 5               |      | 18               | Ω                |   |                                      |
| On/Off Ratio              |                 | 60   |                  | dB               |   |                                      |
| Output Harmonics          |                 |      | -35              | dBc              | With typical 50Ω matching circuits                  |                                      |
| Operating Current         |                 | 50   | 68               | mA               | 15dBm CW  |                                      |
| Operating Current         |                 | 12   | 24               | mA               | 0dBm CW   |                                      |
| MAX Power                 | High Power      | 14   | 16.5             | 17               | dBm   | Matching network for 50Ω 928MHz high |
|                           | Low Power       | -1   | 4.5              | 5.7              | dBm   |                                      |
| MAX Power                 | High Power      | 15.8 | 16.5             | 17               | dBm   | Matching network for 50Ω 433MHz high |
|                           | Low Power       | 3    | 4                | 5                | dBm   |                                      |
| <b>Crystal Oscillator</b> |                 |      |                  |                  |   |                                      |
| Center Frequency          |                 | 24   |                  | MHz              | Trimmed   |                                      |
| Tolerance                 |                 |      | 20               | ppm              | Required crystal tolerance                          |                                      |
| Startup Time              |                 | 50   | 100              | us               | Quick Start enabled                                 |                                      |
| Startup Time              |                 | 2    | 5                | ms               | Quick Start disabled                                |                                      |
| Trim Cap                  | 0               |      | 45               | pF               | Internal trim capacitor (self calibration sets)     |                                      |
| Trim Resolution           | 145             | 160  | 175              | fF               |   |                                      |
| I <sub>dd</sub>           |                 | 800  |                  | uA               | Normal operation                                    |                                      |
| I <sub>dd</sub>           |                 | 1.5  |                  | mA               | During Quick Start                                  |                                      |
| <b>10kHz Oscillator</b>   |                 |      |                  |                  |   |                                      |
| Output Frequency          | 9.8             | 10   | 10.2             | kHz              | After trimming                                      |                                      |
| Operating Current         | 300             | 375  | 450              | nA               | After trimming                                      |                                      |
| Duty Cycle                |                 | 50   |                  | %                |   |                                      |
| <b>ADC</b>                |                 |      |                  |                  |   |                                      |
| Resolution                |                 | 8    |                  | Bits             |   |                                      |
| FSR                       | V <sub>ss</sub> |      | V <sub>ref</sub> | V                | Full scale input range                              |                                      |
| C <sub>i</sub>            |                 | 1    |                  | pF               | Input capacitance                                   |                                      |
| V <sub>ref</sub>          |                 | 2.0  |                  | V                | Internal voltage reference                          |                                      |
| Reference Offset          |                 | 1    |                  | %FSR             |   |                                      |
| f <sub>clk</sub>          | 0.01            |      | 2                | MHz              | Clock frequency                                     |                                      |
| Conversion Rate           |                 |      | 200              | KSPS             | Clock rate = 2MHz                                   |                                      |
| Conversion Time           |                 | 10   |                  | T <sub>clk</sub> |   |                                      |

Table 6: Operational Specifications (Continued)

| Parameter                 | Min. | Typ.  | Max. | Units              | Comment  |
|---------------------------|------|-------|------|--------------------|--|
| <b>Data Filter</b>        |      |       |      |                    |  |
| 3dB Down Point            | 110  | 120   | 130  | %F <sub>DATA</sub> | AM data filter bandwidth (relative to associated defined data rates) |
| <b>Temperature Sensor</b> |      |       |      |                    |  |
| Output Voltage            | 0.93 | 0.97  | 1.01 | V                  | At 27°C  |
| Voltage Range             | 0.61 | 0.97  | 1.4  | V                  | Output   |
| Slope                     |      | -5.24 |      | mv/°C              | dV/dT  |
| <b>RSSI Buffer</b>        |      |       |      |                    |  |
| Input Range               | 0    |       | Vdd  | V                  |  |
| Operating Current         | 135  | 185   | 250  | uA                 |  |
| Unity Gain-BW             | 615  | 1000  | 1700 | kHz                | 100kΩ/100pF load   |
| <b>PLL</b>                |      |       |      |                    |  |
| Reference Input Frequency |      | 12    |      | MHz                | Transmit mode (24MHz external crystal)                               |
|                           |      | 16    |      | MHz                | Receive mode (24MHz external crystal)                                |
| Resolution                |      | 91.55 |      | Hz                 | Frequency step size  |
| VCO Gain Constant         | 9.62 | 12    | 14.4 | MHz/V              | @ 400MHz   |
|                           | 25.6 | 32    | 38.4 | MHz/V              | @ 900MHz   |
| Settling Time             |      | 100   |      | us                 | Internal loop filter   |
| Phase Noise               |      | -90   | -80  | dBc/Hz             | Internal loop filter @ 10kHz offset                                  |
| Phase Noise               |      | -120  | -110 | dBc/Hz             | Internal loop filter @ 3MHz offset                                   |
| Max Spurious Level        |      | -70   | -50  | dBc                | Internal loop filter   |
| <b>POR</b>                |      |       |      |                    |  |
| Delay Time                | 28   | 43    | 60   | ms                 |  |
| Brown-out Trip            | 1.2  | 1.6   | 1.8  | V                  |  |

## 3.0 Block Diagrams

### 3.1 AMIS-53000 Overall Block Diagram

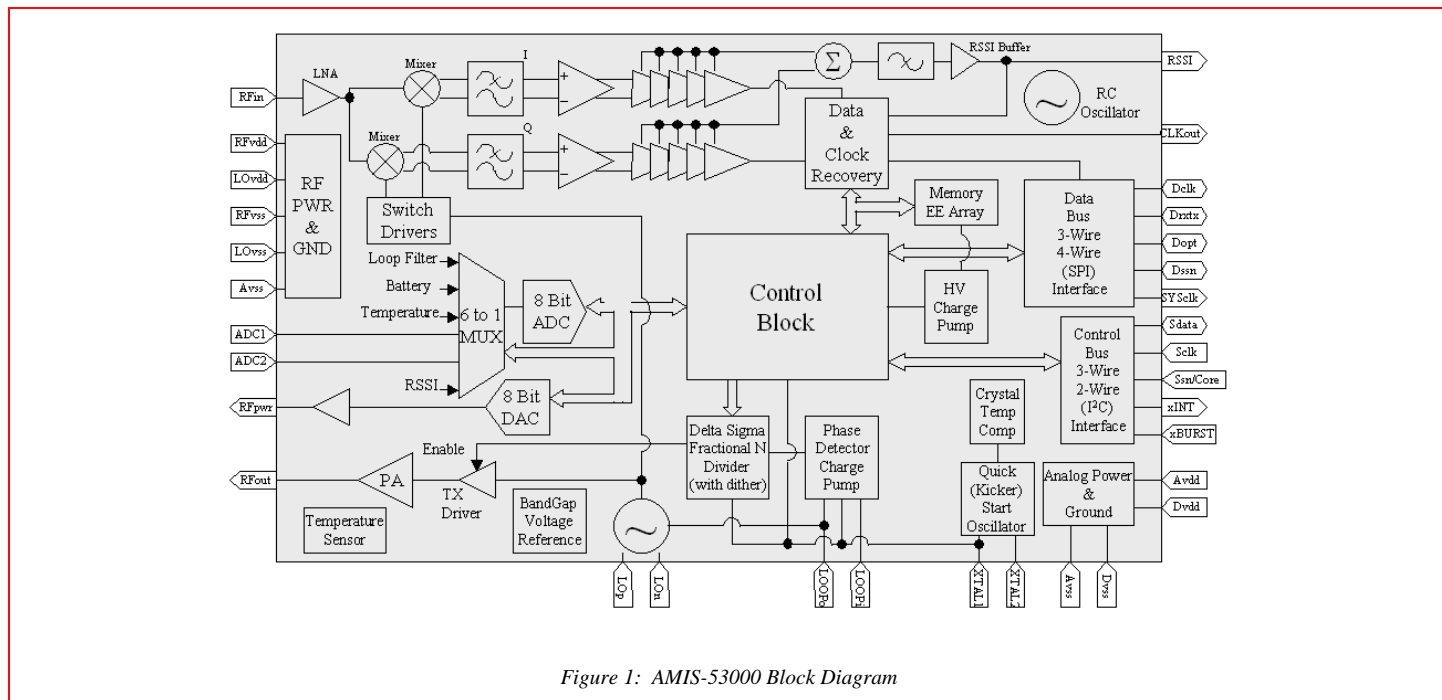


Figure 1: AMIS-53000 Block Diagram

### 3.2 Package

#### 3.2.1. Pin Definition

Table 7: Pin Definitions

| Pin# | -001   | -002   | Pin Type       | Description   |
|------|--------|--------|----------------|---|
| 1    | LNAvdd | LNAvdd | Power          | A DC short (inductor) is connected to VDD from this pin   |
| 2    | RFin   | RFin   | Analog Input   | The RF input to the receiver circuits   |
| 3    | RFvss  | RFvss  | Ground         | Ground for the RF circuits  |
| 4    | RFout  | RFout  | Analog Output  | RF transmit output  |
| 5    | RFpwr  | RFpwr  | Analog Output  | Variable DC voltage output to power the RF transmitter (requires a DC short (inductor) connection to Rfout) |
| 6    | Avdd   | Avdd   | Power          | Vdd power for the analog circuits   |
| 7    | ADC1   | ADC1   | Analog Input   | Input to the analog to digital conversion circuit   |
| 8    | ADC2   | ADC2   | Analog Input   | Input to the analog to digital conversion circuit   |
| 9    | RSSI   | RSSI   | Analog IO      | Analog voltage related to the strength of the received RF   |
| 10   | PEAK   | PEAK   | Analog         | Analog voltage for external auto-slice capacitor  |
| 11   | Avss   | Avss   | Ground         | Ground for the analog circuits  |
| 12   | XTAL2  | XTAL2  | Analog         | Connection to an external crystal   |
| 13   | XTAL1  | XTAL1  | Analog         | Connection to an external crystal   |
| 14   | INT    | INT    | Digital Output | Interrupt to external controller  |
| 15   | Dopt   | Dopt   | Digital Input  | Optional data pin for the 4-wire data interface mode  |
| 16   | Dssn   | Dssn   | Digital IO     | Active low select line for the data interface   |

Table 7: Pin Definitions (Continued)

| Pin# | -001    | -002    | Pin Type       | Description   |
|------|---------|---------|----------------|---|
| 17   | Drxtx   | Drxtx   | Digital IO     | Serial data input (transmit) or output (received)   |
| 18   | Dclk    | Dclk    | Digital IO     | Recovered clock output (data interface clock)   |
| 19   | SYSclk  | SYSclk  | Digital Output | System clock output   |
| 20   | Dvss    | Dvss    | Ground         | Ground for the digital circuits   |
| 21   | Dvdd    | Dvdd    | Power          | Vdd power for the digital circuits  |
| 22   | CoreReg | SSN     | Digital        | -001 (control and status for the serial data interface)<br>-002 (decoupling capacitor pin for the internal regulator) |
| 23   | SCLK    | SCLK    | Digital        | Bi-directional clock for the 2-wire serial interface  |
| 24   | SDATA   | SDATA   | Digital        | Bi-directional data for the 2-wire serial interface   |
| 25   | xBURST  | xBURST  | Digital Input  | Active low input interrupt that will immediately cause a Burst transmission   |
| 26   | LOOPout | LOOPout | Analog         | Output to the optional external loop filter   |
| 27   | LOOPin  | LOOPin  | Analog         | Input from the optional external loop filter  |
| 28   | LOvss   | LOvss   | Ground         | Ground for the local oscillator circuits  |
| 29   | LOn     | LOn     | Analog         | Negative side of the VCO tank   |
| 30   | LOp     | LOp     | Analog         | Positive side of the VCO tank   |
| 31   | LOvdd   | LOvdd   | Power          | Vdd for the local oscillator circuits   |
| 32   | RFvdd   | RFvdd   | Power          | Vdd power for the RF circuits   |

### 3.2.2. Block Diagram/Pin Definition

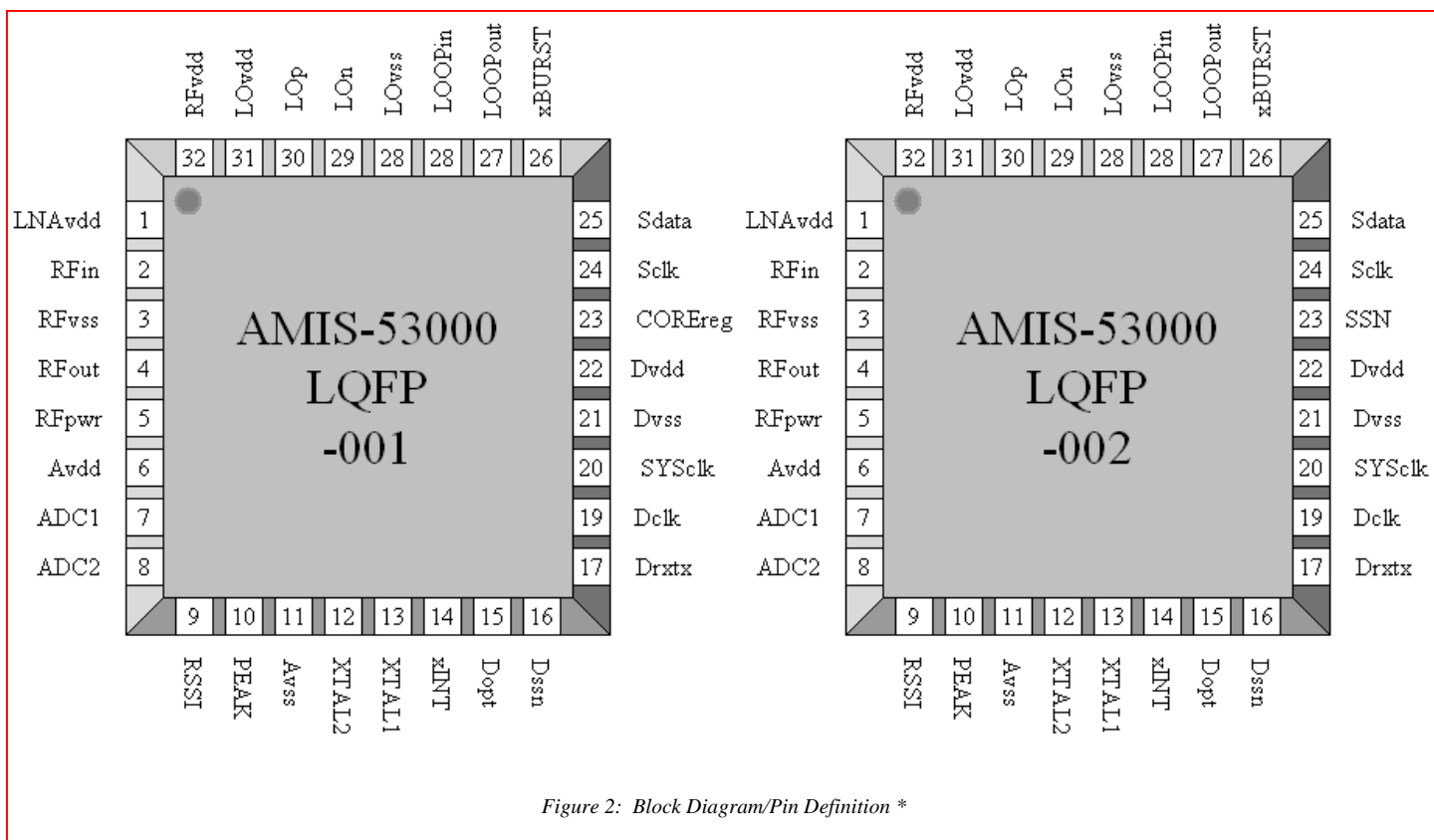


Figure 2: Block Diagram/Pin Definition \*

\* Not actual package markings. Please see marking format in 3.2.3.3.

### 3.2.3. Physical Characteristics

#### 3.2.3.1. Package Type

32 pin LQFP

#### 3.2.3.2. Package Dimensions

Table 8: AMIS-53000 LQFP Package Dimensions

| Symbol    | Min.     | Nom. | Max. | Units |
|-----------|----------|------|------|-------|
| Thickness | -        | -    | 1.60 | mm    |
| D         | 9.00 BSC |      |      | mm    |
| D1        | 7.00 BSC |      |      | mm    |
| E         | 9.00 BSC |      |      | mm    |
| E1        | 7.00 BSC |      |      | mm    |
| e         | 0.80 BSC |      |      | mm    |

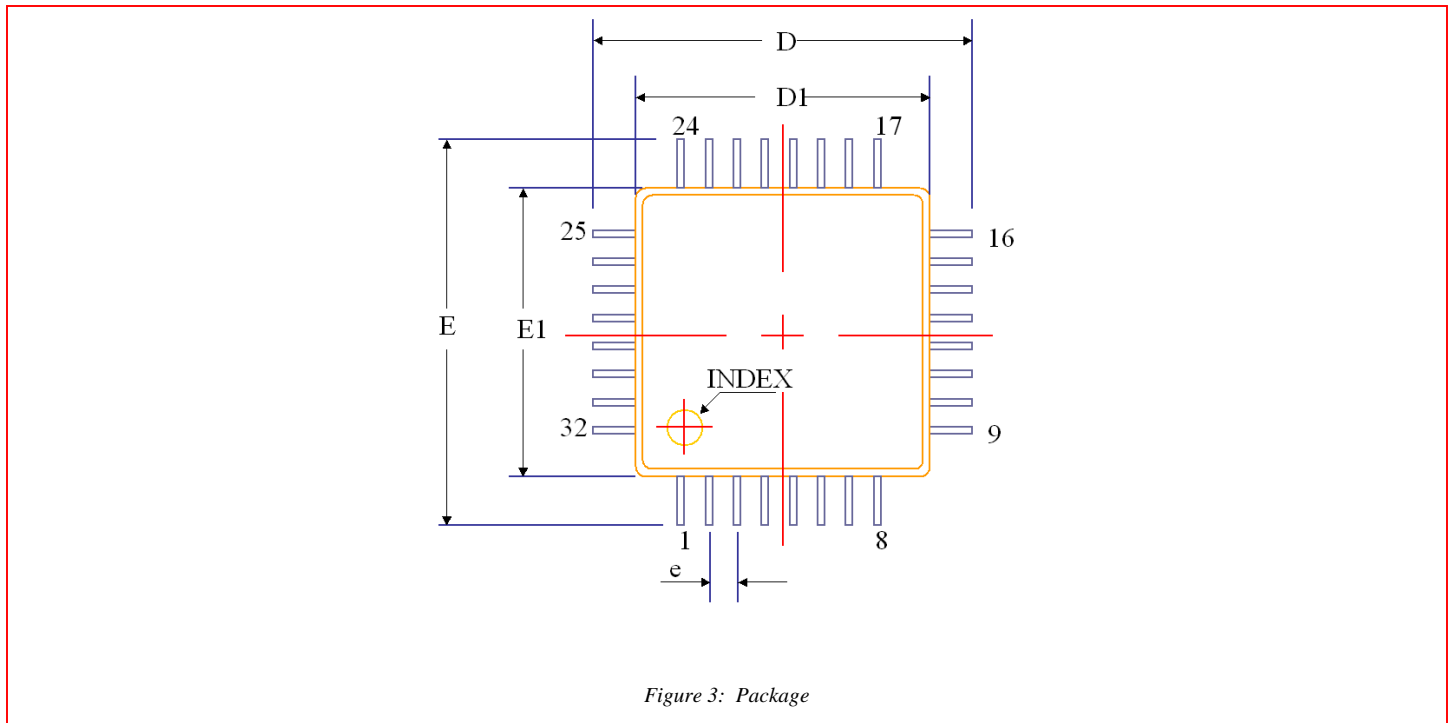


Figure 3: Package

#### 3.2.3.3. Package Marking Format

```
(AMIS Logo)
AMIS53000a
19608-bbb
XXXXYYZ
```

Where:

a is the market application

bbb is the AMIS device version

XXXXYYZ is the date and tractability code\*\*\*\* is the country of origin (found on underside of chip).

The year in which the mask work was first fixed in a semiconductor chip product may also appear.

## 4.0 Acronyms

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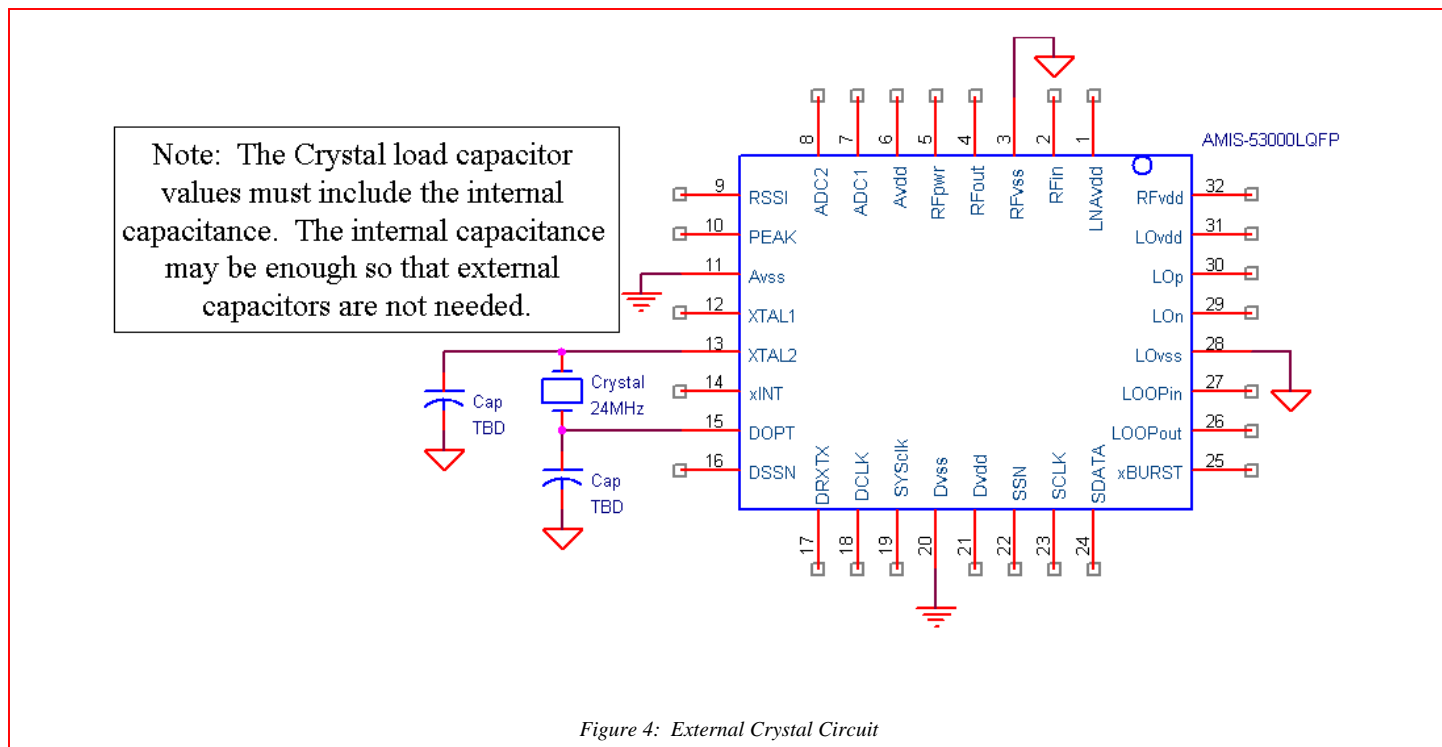
The following acronyms are used in this document.

|        |   |
|--------|---|
| AM     | Amplitude Modulated signal  |
| ASIC   | Integrated circuit designed for a single customer requirement                                 |
| ASK    | Amplitude Shift Key   |
| ASSP   | A custom integrated circuit, that may be used in general designs                              |
| ASTRIC | AMI Semiconductor's family of wireless products   |
| CCA    | Clear Channel Assessment  |
| CDR    | Clock and Data Recovery, data is recovered from the received signal using a synchronous clock |
| CRC    | Cyclic Redundancy Checking; data error checking   |
| CW     | Continuous Wave, a single frequency or modulated signal carrier                               |
| DAC    | Digital to Analog Conversion  |
| dB     | Decibels; a logarithmic measure of signal level   |
| dBm    | Logarithmic measure of signal level above a milli-watt  |
| DFFT   | Digital or Discrete Fast Fourier Transform  |
| DPLL   | Digital Phase Locked Loop circuit to create a precise frequency                               |
| EE     | Electrical Erasable Memory  |
| FFT    | Fast Fourier Transform; transform between time and frequency                                  |
| FM     | Frequency Modulated signal  |
| FSK    | Frequency Shift Key   |
| GFSK   | Gaussian Data Waveform Modulated signal   |
| IF     | Intermediate Frequency  |
| kbps   | Data rate in thousand bits per second   |
| kHz    | Frequency in kilohertz per second   |
| LO     | Local Oscillator frequency; used to convert signals between RF frequency and IF frequency     |
| LOP    | Byte indicating the length of a packet  |
| MHz    | Frequency in megahertz  |
| MICS   | Medical Implantable Communication System  |
| mV     | Milli-volts   |
| OOK    | On/Off method of creating an amplitude modulated signal                                       |
| OTA    | Transconductance Amplifier  |
| PLL    | Phase Locked Loop circuit to create a precise frequency                                       |
| POR    | Power-on-Reset is a threshold circuit for limiting operation at low voltages                  |
| RF     | Radio Frequency   |
| RSSI   | Received Signal Strength Indication; measurement of RF signal strength                        |
| SOF    | Byte indicating start of packet in data protocol  |
| VCO    | Voltage controlled variable frequency oscillator  |

## 5.0 Hardware Description

### 5.1 Frequency

The AMIS-53000 uses an internal VCO, PLL and trim capacitors with an external oscillator crystal to generate the RF frequencies for both TX and RX. The external crystal is a parallel resonant mode crystal with required loading capacitors. The AMIS-53000 contains internal load capacitors, typically sufficient for use with the suggested 24MHz. It is suggested that a 24MHz with 20ppm tolerance be used with the AMIS-53000.



The internal VCO requires an external parallel LC to set the frequency for RX or TX. There is an internal capacitance that needs to be considered when selecting the values of the inductor and capacitor. The AMIS-53000 is sensitive to the positioning of the LC components in the layout of the PCB. The traces to the LC need to be as symmetrical as is possible. The location of the LC needs to be as close to the AMIS-53000 pins as is practical. A simple layout change to these parameters can mean that the AMIS-53000 VCO frequency will change causing a need to change the values of the inductor, capacitor and/or the values in the registers controlling the RF frequency. The value of the inductor and/or the capacitor may need to be adjusted to allow the AMIS-53000 to calibrate the PLL for a given frequency of operation. The layout of the printed circuit board for the inductor and capacitor should route traces connecting other components away from the inductor and capacitor pads.

The VCO in the AMIS-53000 is a differential negative resistance oscillator (DNRO), commonly found in the literature. It uses an internal voltage variable capacitor (varactor) in combination with an external L and C to provide the desired frequency. The output frequency is found simply by:

$$F_{vco} = \frac{1}{\sqrt{L_{tot} \cdot C_{tot}}} \frac{1}{2 \cdot \pi}$$

Where: L<sub>tot</sub> and C<sub>tot</sub> are the total inductance and capacitance respectively at the VCO pins. This includes the internal capacitance of approximately 2pF.



Note: These components need to be placed as close as possible to the AMIS-53000 in the layout. It is possible that the internal capacitance is enough for most designs.

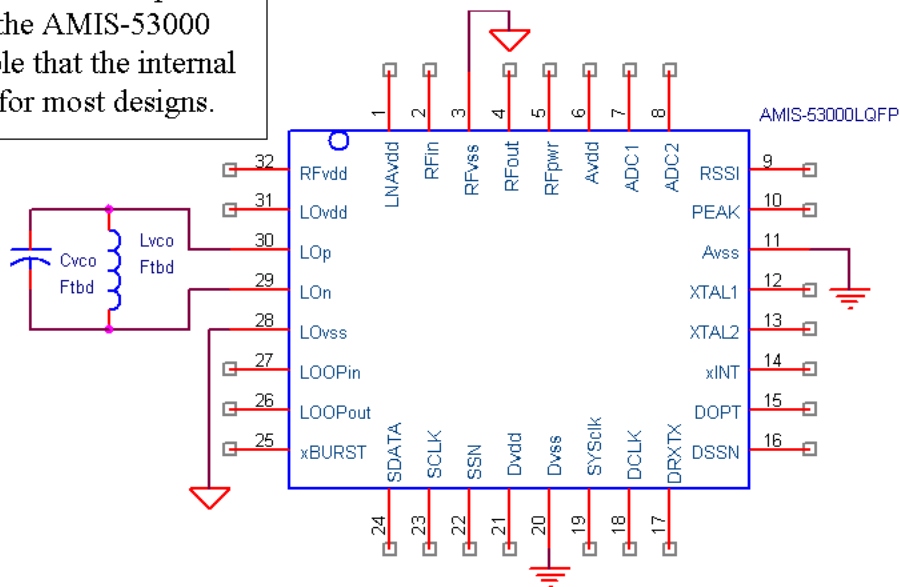


Figure 5: VCO External LC Circuit

The AMIS-53000 has an internal loop filter to work with the PLL in creating the frequency of the device. There is an option to use an external loop filter.

Table 9: Internal Loop Parameters

| Filter          | Component | Value | Units | Comments |
|-----------------|-----------|-------|-------|----------|
| Second Order    | R1        | 60    | kΩ    |          |
|                 | C1        | 64    | pF    |          |
|                 | C2        | 3     | pF    |          |
| Additional Pole | R         | 110   | kΩ    |          |
|                 | C         | 1     | pF    |          |

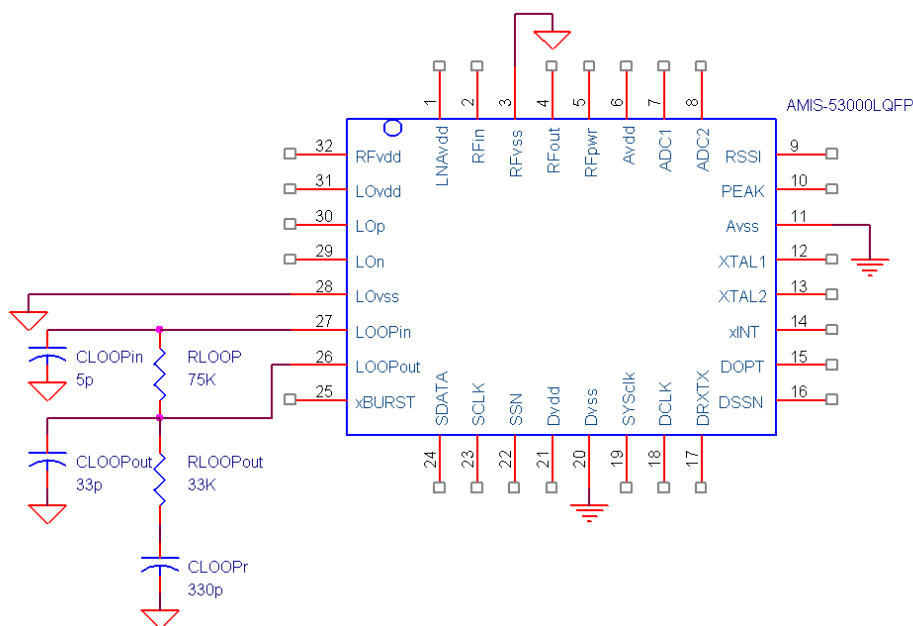


Figure 6: Optional External Loop Filter Circuit

An internal 10kHz oscillator provides timing for functions; Sniff Mode, Burst transmit and housekeeping, when the AMIS-53000 is in its lowest power mode (idle/standby). This oscillator requires no external components. The 10kHz oscillator's internal trim capacitor is trimmed by 8 bits of trim control in a self calibration. Once the trim is set, the oscillator frequency will be accurate to within two percent over specified voltages and temperatures.

## 5.2 Receiver

The AMIS-53000 has a single channel receiver. The LNA for the receiver input requires a DC connection to ground on the input (must not be an RF ground connection). The LNA for the receiver input requires a DC connection to RFVDD on the output. These connections are supplied through inductors becoming part of the matching circuit for the receiver input.

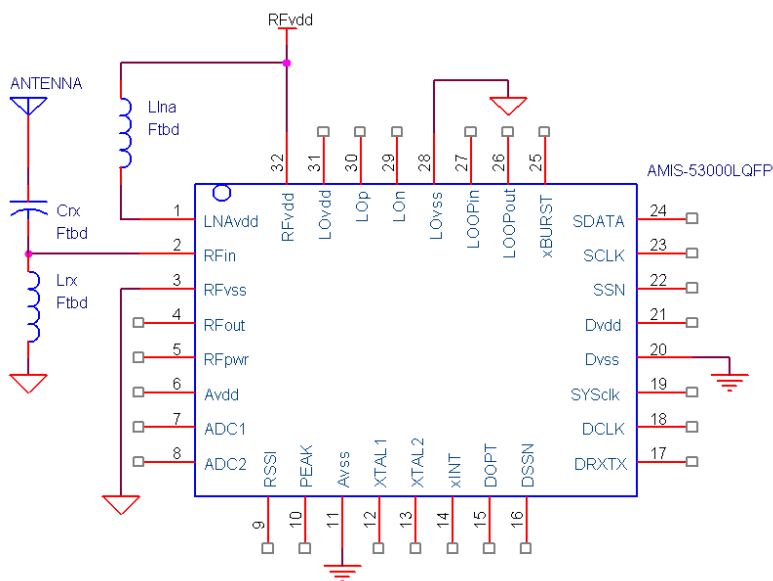


Figure 7: Receiver Input Matching Circuit

## 5.2.1. Receiver Low Noise Amplifier (LNA)

The receiver input of the AMIS-53000 is a single ended input and single ended output device. The input is matched to 50Ω using an external matching network, which provides a DC path to ground for biasing the receiver’s LNA. The output of the LNA is tuned to the desired operating frequency using an external inductor and on-chip capacitor. The inductor is also provides the LNA with DC supply voltage. On-chip tuning capacitors are binary weighted and digitally controlled.

The internal input capacitance is 1.2pF to 4pF. With this capacitance set to the mid value (register set to 0X80), the impedance of the receiver is shown in Figure 8. The internal output capacitance is 0.32pF to 0.912pF.

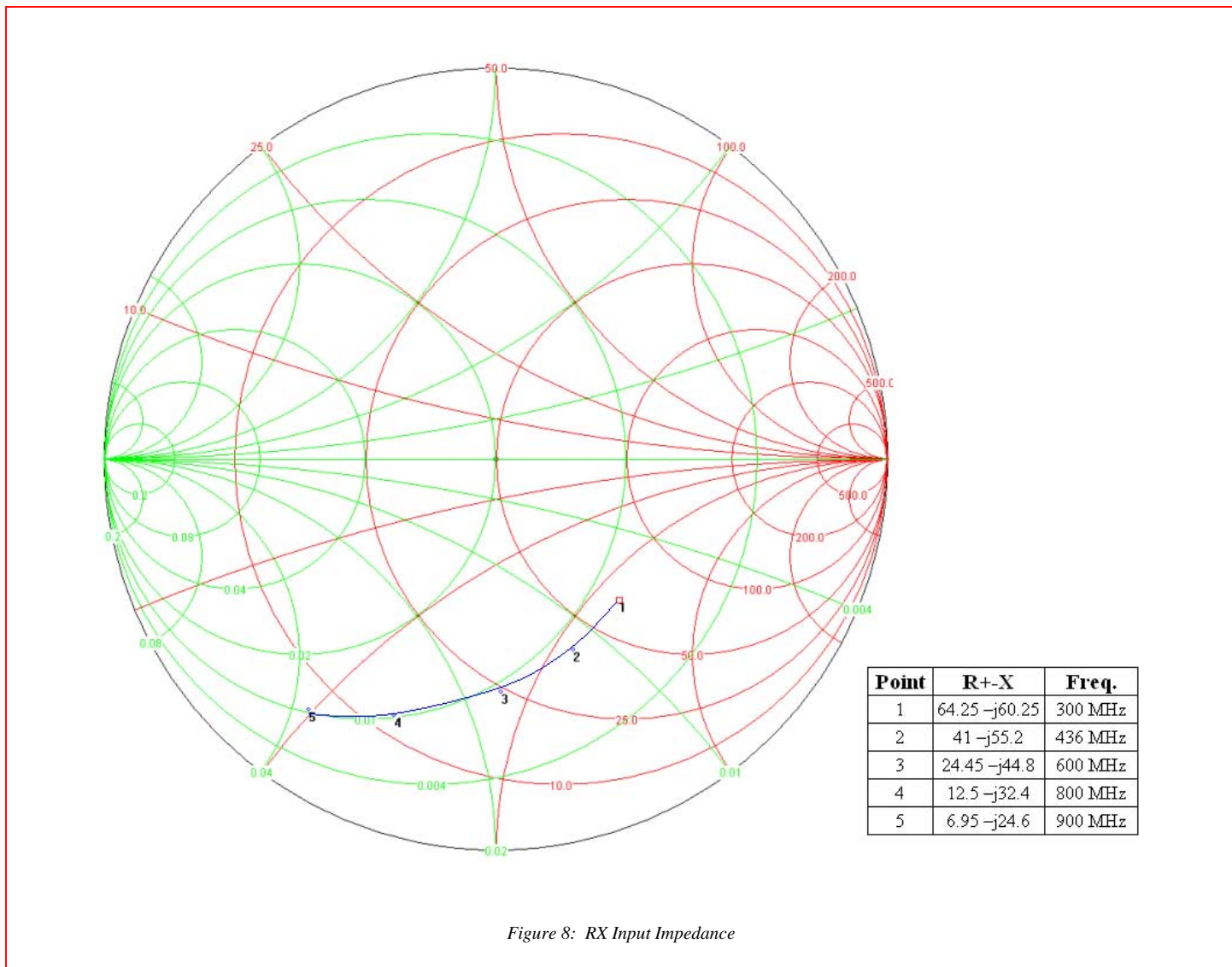


Figure 8: RX Input Impedance

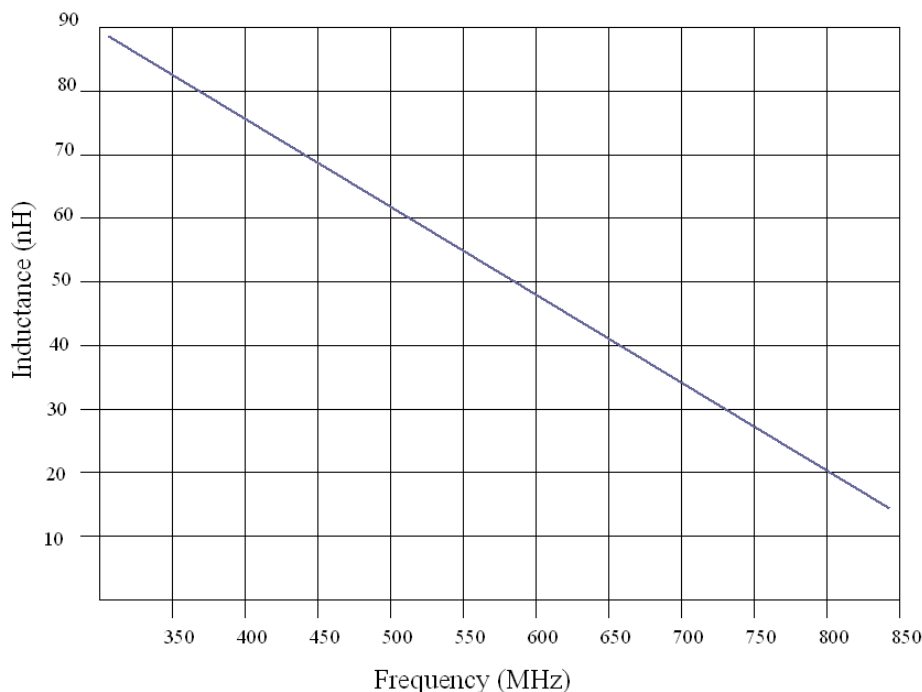


Figure 9: Receiver LNA Output Inductor Selection

### 5.2.2. IF Filter

A passive poly-phase filter and active filtering are used in the AMIS-53000 for frequency selectivity and rejection of the image frequency. It is designed to provide an optimal image rejection of 50dB at 500kHz.

### 5.2.3. Data Filter

The OOK low-pass data filter is used for additional post-detection signal filtering in accordance with the OOK signal data rate (1.2, 1.8, 2.4, 4.8, 7.2, 9.6, 14.4, 19.2kHz). The RSSI buffer is used to drive the RSSI signal off chip for external monitoring, and can also be internally configured for monitoring other signals such as the analog temp sense voltage or bandgap voltage.

## 5.3 Transmitter

The AMIS-53000 transmitter is a two-stage output amplifier. When both stages are selected, the highest output power at frequencies from 300MHz to 928MHz is +15dBm. When only one stage is used, the AMIS-53000 can output up to 0dBm with better power efficiency than when outputting the same power level with both stages. The voltage output level on the RFPWR pin controls the RF output power level of the AMIS-53000. A DC connection must be made between the RFOUT pin and the RFPWR pin. The non-linear output of the AMIS-53000 may require external components to match to a load and to reduce the spurious harmonics.

The output impedance of the AMIS-53000 can be matched to the impedance of an external load, using the spreadsheet AMIS-53RFMATCH.xls provided by AMIS. This spreadsheet is explained in the application note, AMIS-52X00 Antenna Impedance Matching Considerations. The goal of the transmit output matching with this spreadsheet is to optimize the output power while reducing the harmonic power.

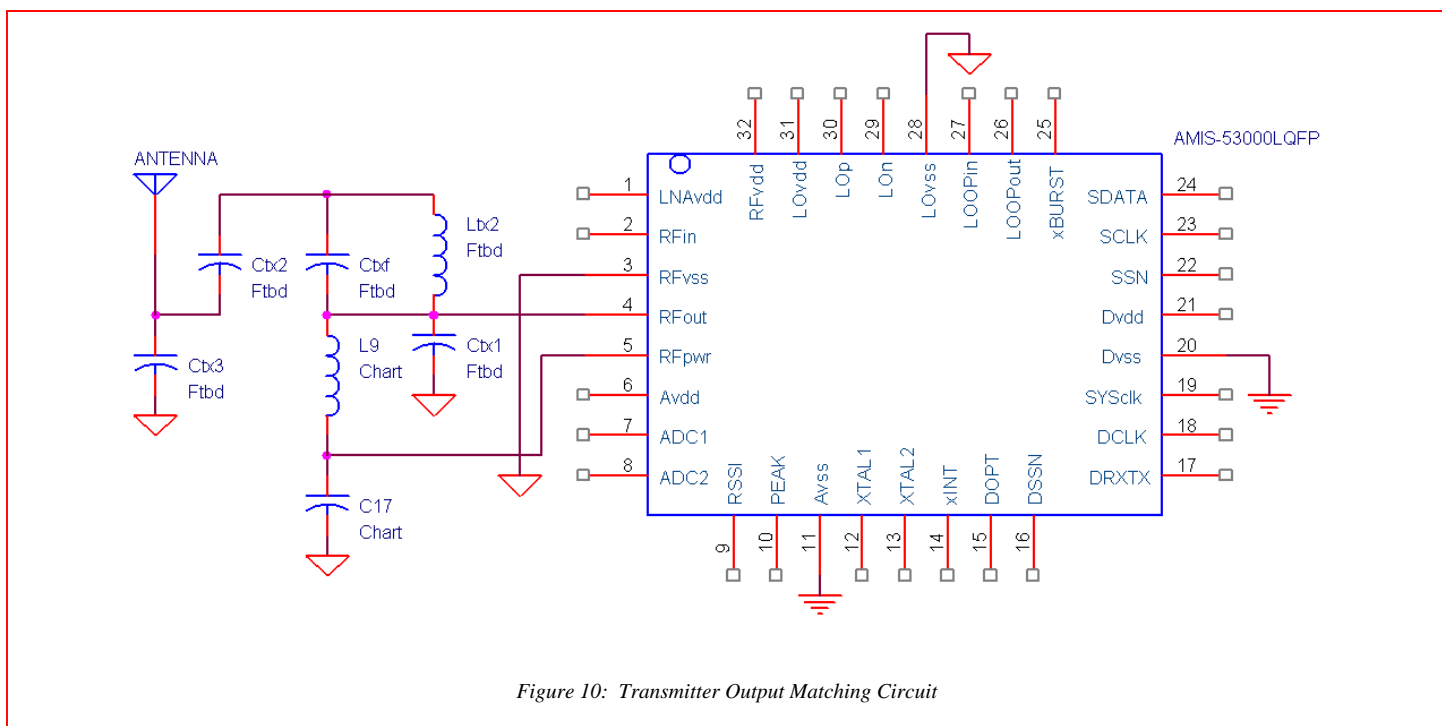


Figure 10: Transmitter Output Matching Circuit

## 5.4 Single Antenna Option

The AMIS-53000 is designed such that when the transmitter is off or the receiver is off, the pins are grounded. This provides a known impedance for the off port (transmit or receive) in combining the receiver and the transmitter to a single antenna.

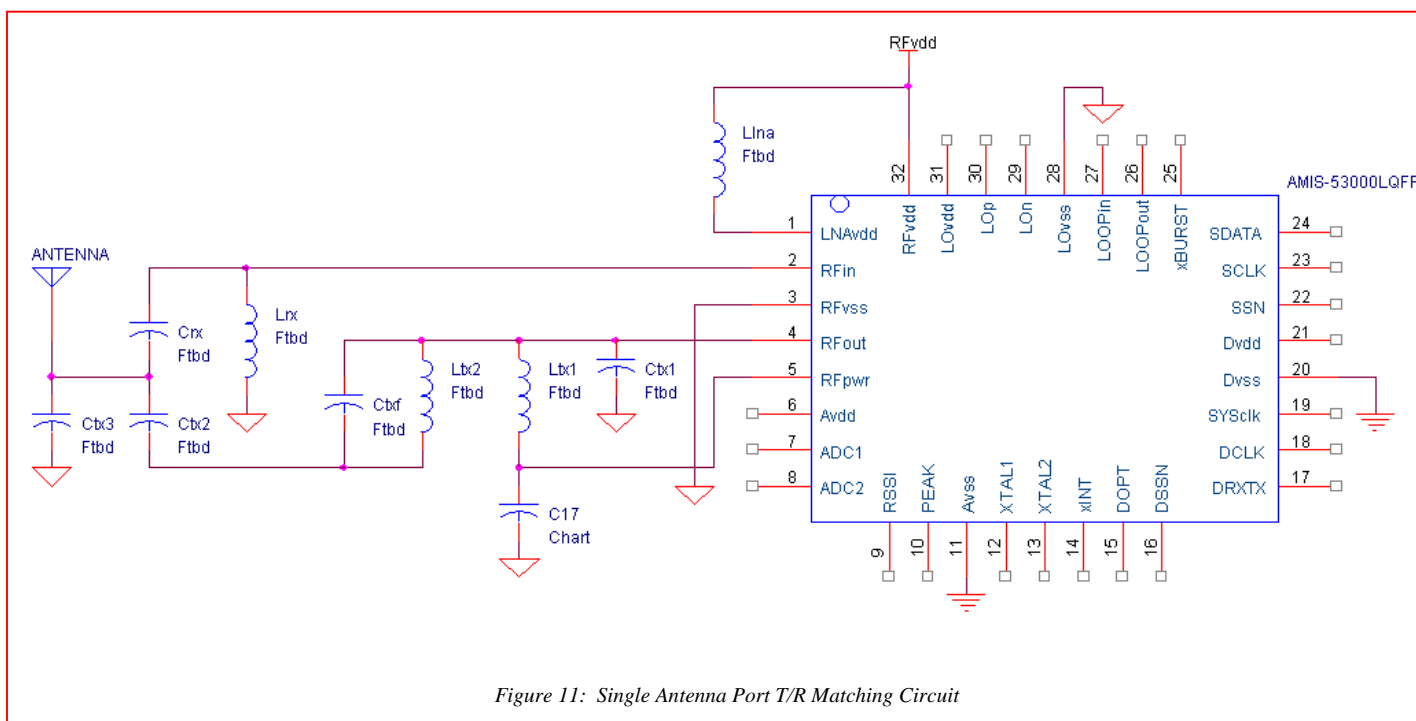


Figure 11: Single Antenna Port T/R Matching Circuit

## 5.5 Peak

The AMIS-53000 has three modes for slicing the received signal to recover the data. One method is to set a threshold value that is fixed and to which the receiver compares the recovered signal. The other two methods have the AMIS-53000 automatically setting a threshold level to which the receiver compares the recovered signal. Both of these automatic threshold methods require an external capacitor on the PEAK pin to operate.

In the averaging method, the AMIS-53000 simply adds a low pass filter with a cutoff frequency set below the data rate filter setting. This second filter extracts an average RSSI level as the data slice threshold. The capacitor on the PEAK pin sets the time constant (corner frequency) for this filter. A typical capacitor value would allow the average level to settle to 95 percent of the RSSI level in 2 bit intervals (remember that Manchester encoding may have transitions twice the data rate). The average threshold method will have chatter before a signal is received and after the signal ends which the external host/controller must be able to handle.

In the peak method the AMIS-53000 uses a peak detector to find the maximum input signal level and then sets the threshold 6dB lower than that level. The external peak capacitor is used to bleed or discharge the peak voltage in the circuit. The voltage swing on the RSSI for a typical 12dB signal to noise ratio at  $10^{-3}$  BER is 240mV. The capacitor value should not change the voltage by more than this 240mV during a string of zeros. The value is dependent on the number of zeros that are allowed in the chosen data protocol, NRZ or Manchester encoded.

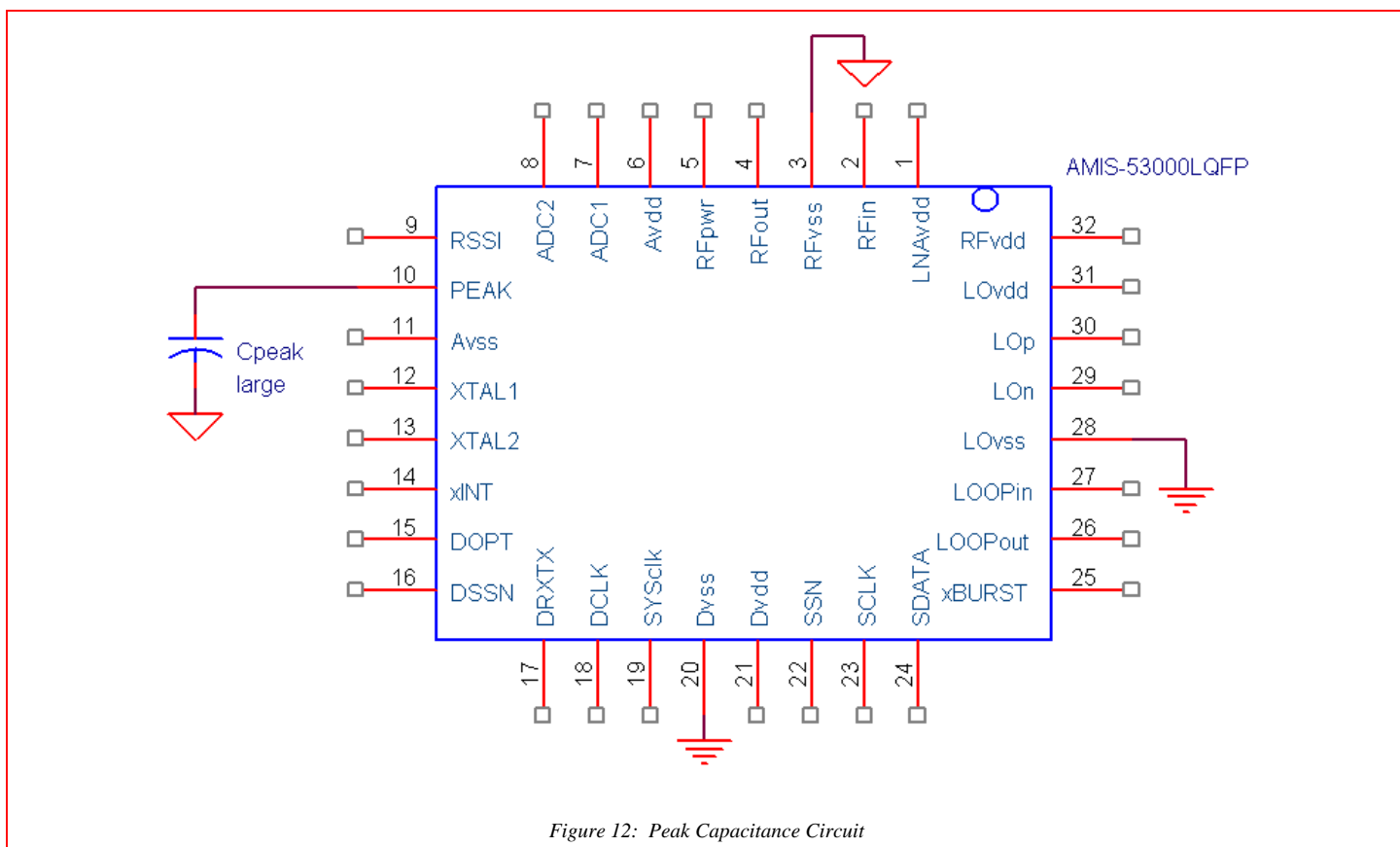


Figure 12: Peak Capacitance Circuit

## 5.6 ADC

The ADC is a successive approximation analog to digital converter, using an internal 8 bit DAC as the reference. The ADC data for the selected input channel(s) will be stored in the associated register, allowing for external access to the conversion data through the serial interface. Conversion speed is register selectable up to 128kS/s. Commands in the control register allow for single or continuous operation of the ADC.

A voltage regulator generates the 2.0V reference for the ADC and DAC based upon an internal bandgap voltage source. The ADC has six inputs, two of which are available to the designer for use in their application.

## 5.7 Control Interface Serial Bus

The AMIS-53000 uses a 3-wire or 2-wire I<sup>2</sup>C interface to communicate with the AMIS-53000 internal registers. The AMIS-53000 will automatically determine which interface to use by determining the states of the three lines; SDATA, SCLK and SSN (the interface is set when the external host/controller writes the first data to the AMIS-53000). Once the AMIS-53000 has determined the type of interface, it will continue with that configuration until power is removed from the part or the part is reset.

**I<sup>2</sup>C:** If SSN is high and an I<sup>2</sup>C start bit is detected, I<sup>2</sup>C mode is enabled.

**SPI:** If SSN is low, and a negative edge on SCLK detected, SPI mode is enabled.

The AMIS-53000 is designed to conform to the Philip Semiconductor I<sup>2</sup>C standard with the AMIS-53000 as the slave device.

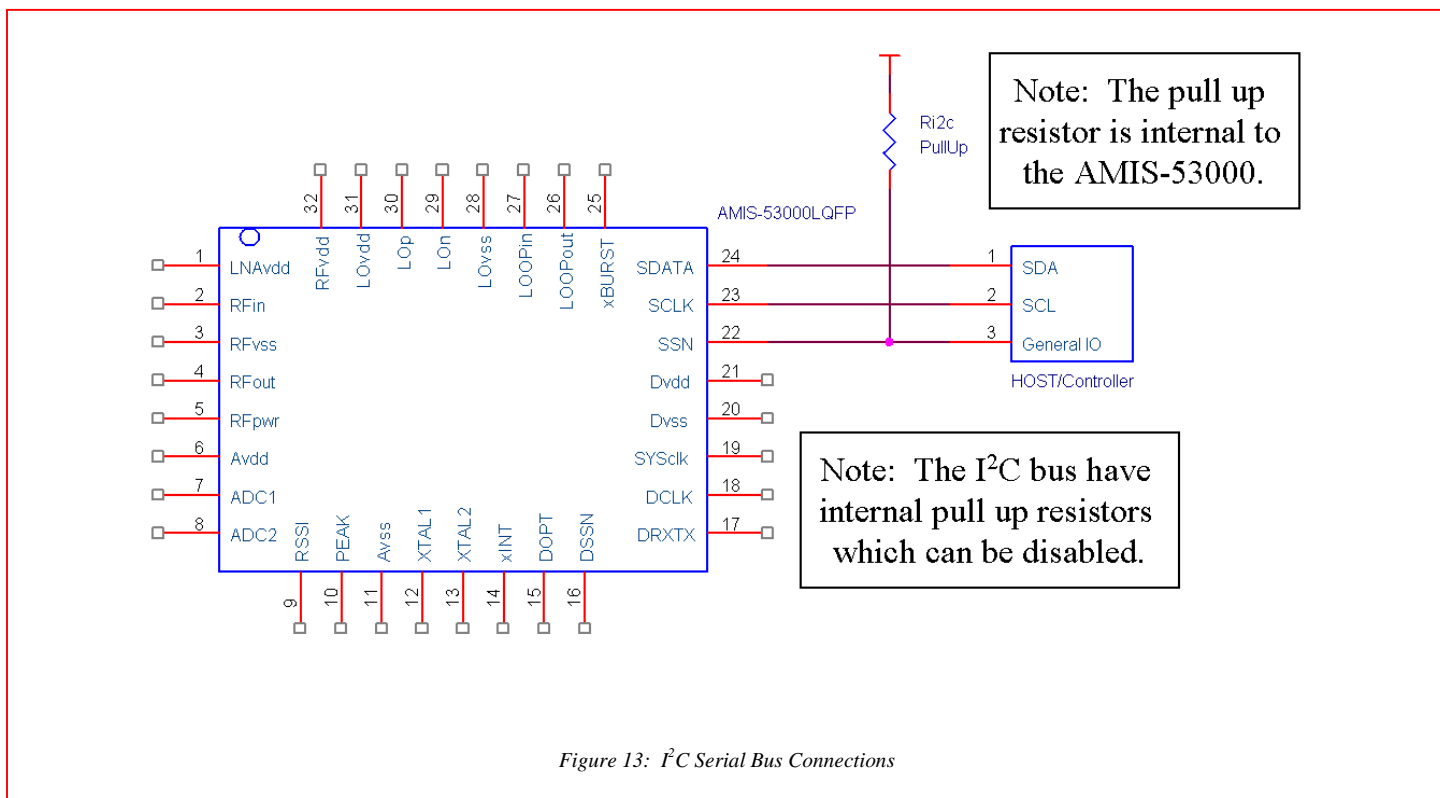


Figure 13: I<sup>2</sup>C Serial Bus Connections

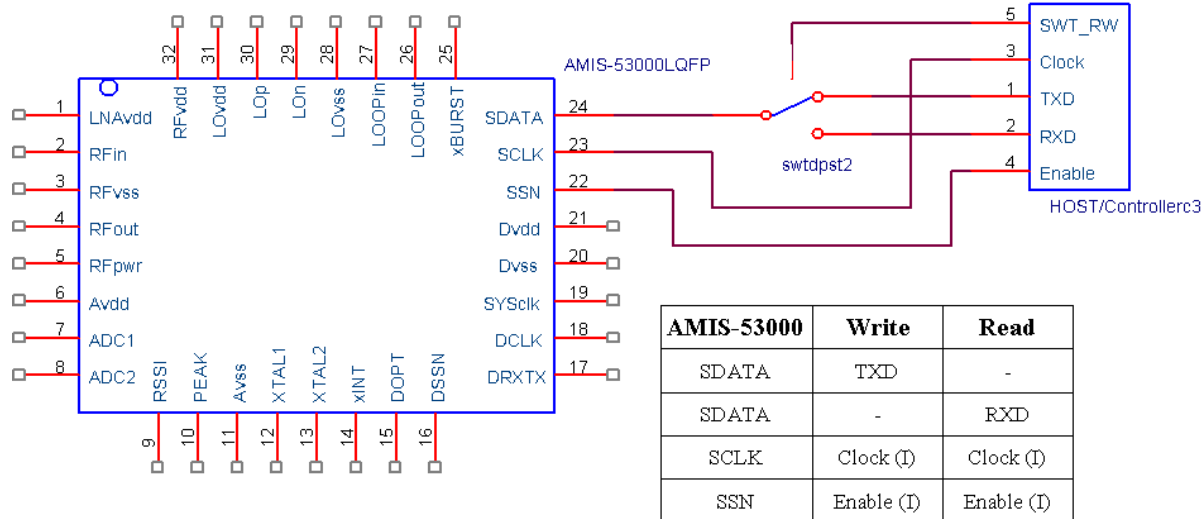


Figure 14: 3-Wire Control Bus Connections

## 5.8 TX/RX Data Interface Serial Bus

The AMIS-53000 uses a 3-wire or 4-wire SPI serial data interface to transfer data between the external host controller and the AMIS-53000. The interface is selected by writing to a register in the AMIS-53000. The DOPT line is undefined in the 3-wire interface.

The 4-wire interface of the AMIS-53000 is designed to be compatible with the definition of a standard SPI interface. The AMIS-53000 can be a slave or master device. The status of the AMIS-53000, master or slave, and the interface mode, read or write, determine the definition of the DRXTX and DOPT pin's as outputs or inputs.

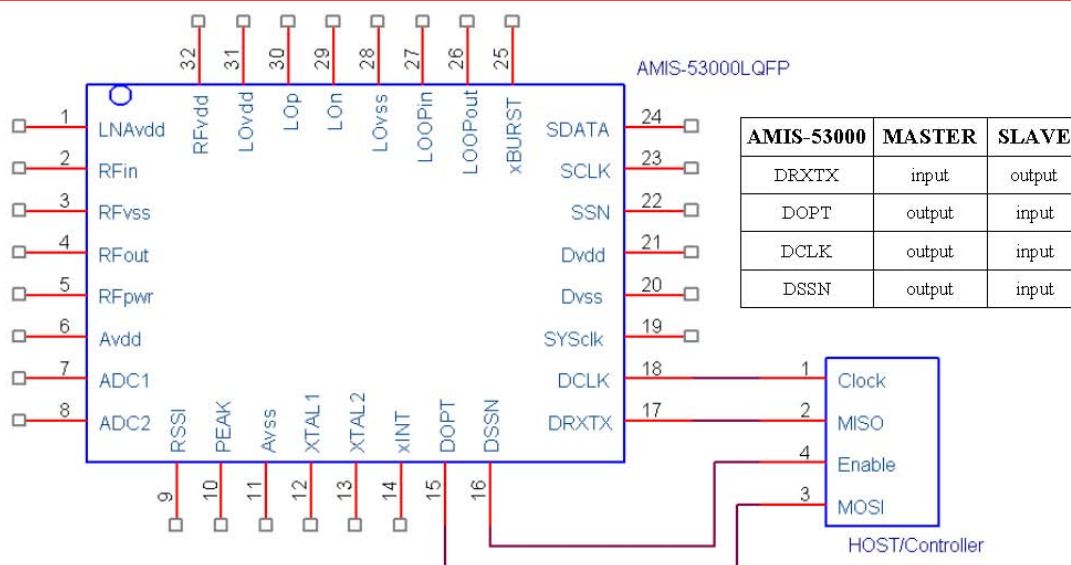


Figure 15: SPI Compatible Serial Data Interface



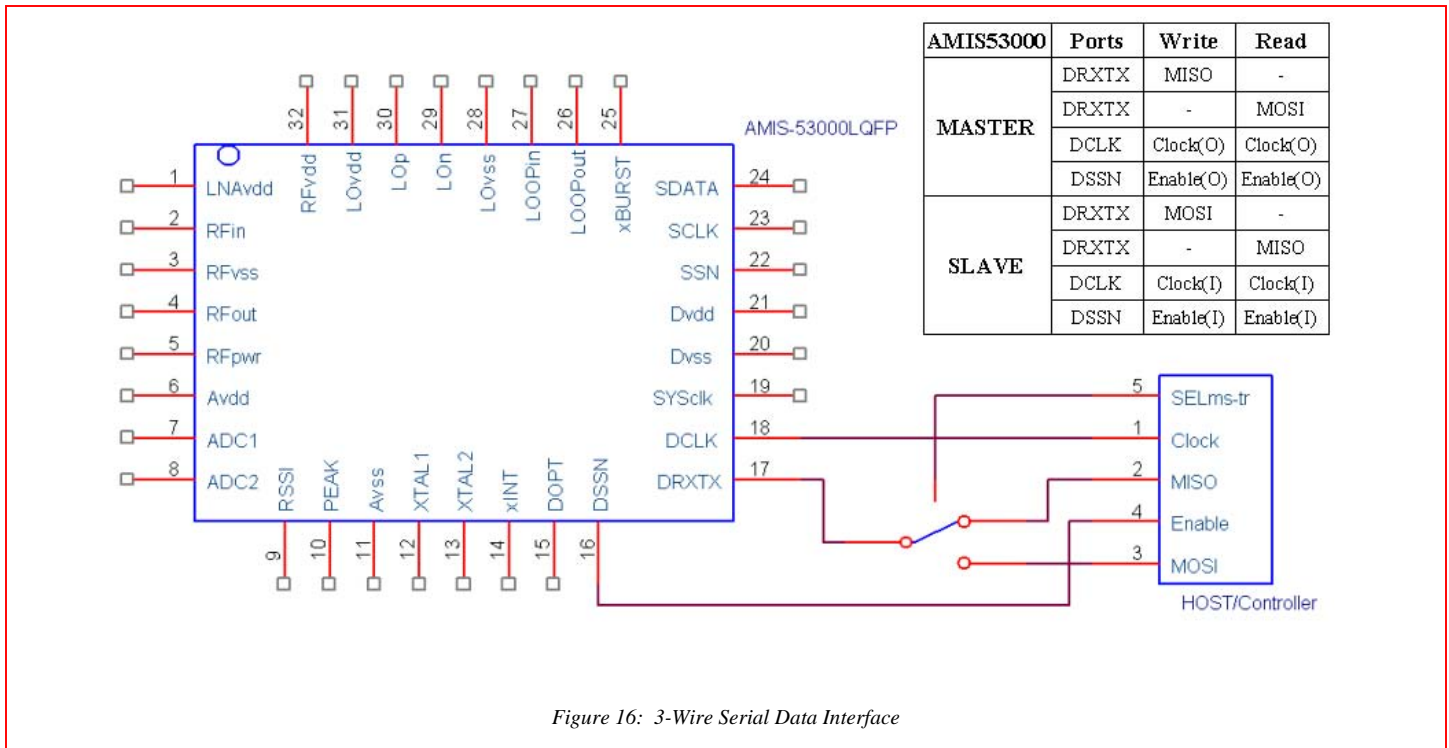


Figure 16: 3-Wire Serial Data Interface

## 5.9 System Clock

The system clock can provide a clock to the external host controller. The clock can be divided down from the 24MHz crystal frequency of the AMIS-53000. When a design desires to use the system clock as the clock to an external host/controller, the system clock can be output under the following:

- Will be output in RX or TX, unless the output is off in general options B (Bit 1:0).
- The output will start back up in idle mode after a packet is received.
- The output will start back up in housekeeping if wake up external host/controller is enabled in housekeeping configuration (Bit 6).

Table 10: System Clock Control

| Mode    | Control                | Bits | Outputs                          | Comments |
|---------|------------------------|------|----------------------------------|----------|
|         | 0X0D General Options B | 1:0  | Frequency: 12, 6, 3MHz or off    |          |
| RX      |                        |      |                                  |          |
| TX      |                        |      |                                  |          |
| Standby | General Options A      | 2:1  | POR state: standby, idle, RX, TX |          |
|         |                        | 0    | Output in standby                |          |
| Idle    | Idle Config            | 4:3  | Clock cycles before stop         |          |

## 5.10 Power and Grounds

The AMIS-53000 has four different power inputs and two different grounds. This allows the design of the AMIS-53000 in an application to separate RF power from the analog and digital power. The same applies to the grounds, where a separate ground plane for the RF grounds can reduce the amount of noise induced into the sensitive RF circuits.

## 5.11 Design Suggestions

The following schematic and layout suggests at least one way to create a printed circuit board for applications using the AMIS-53000.

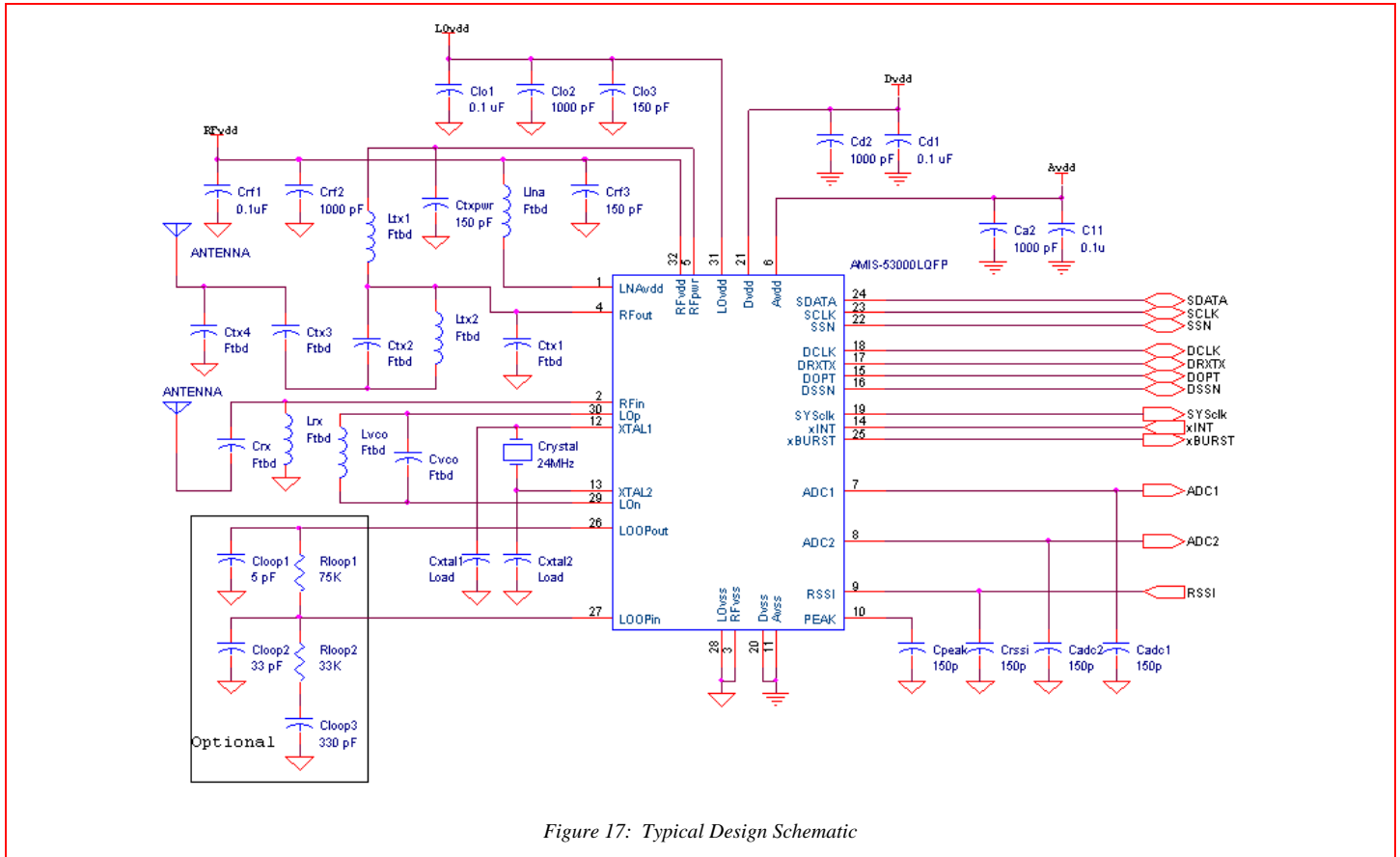


Figure 17: Typical Design Schematic

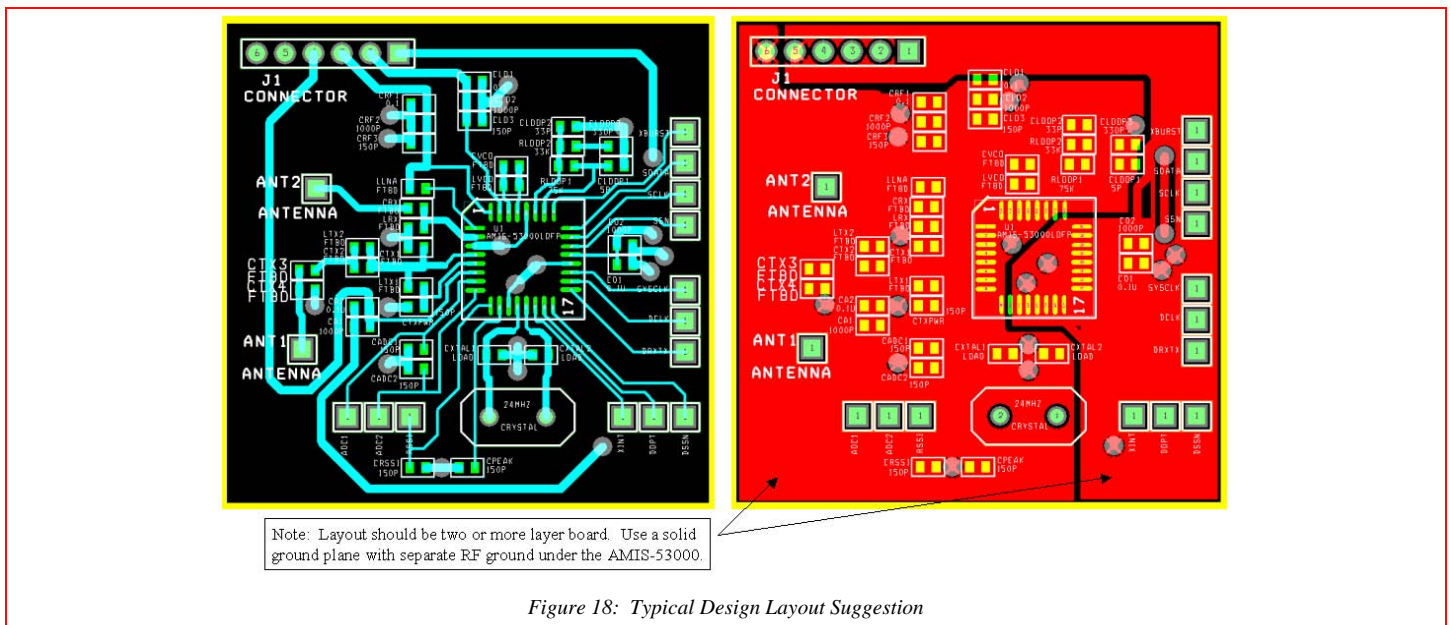


Figure 18: Typical Design Layout Suggestion

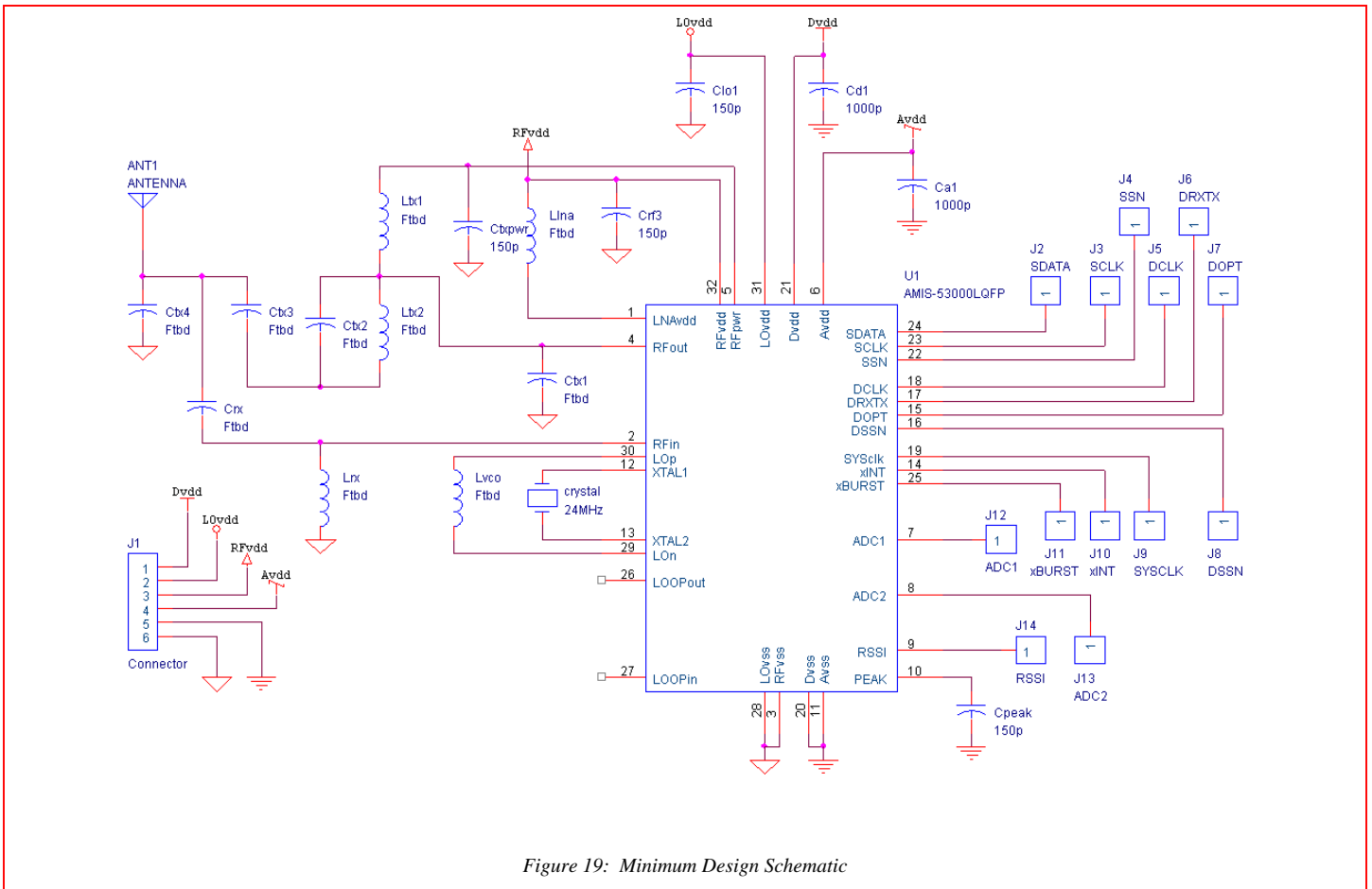


Figure 19: Minimum Design Schematic

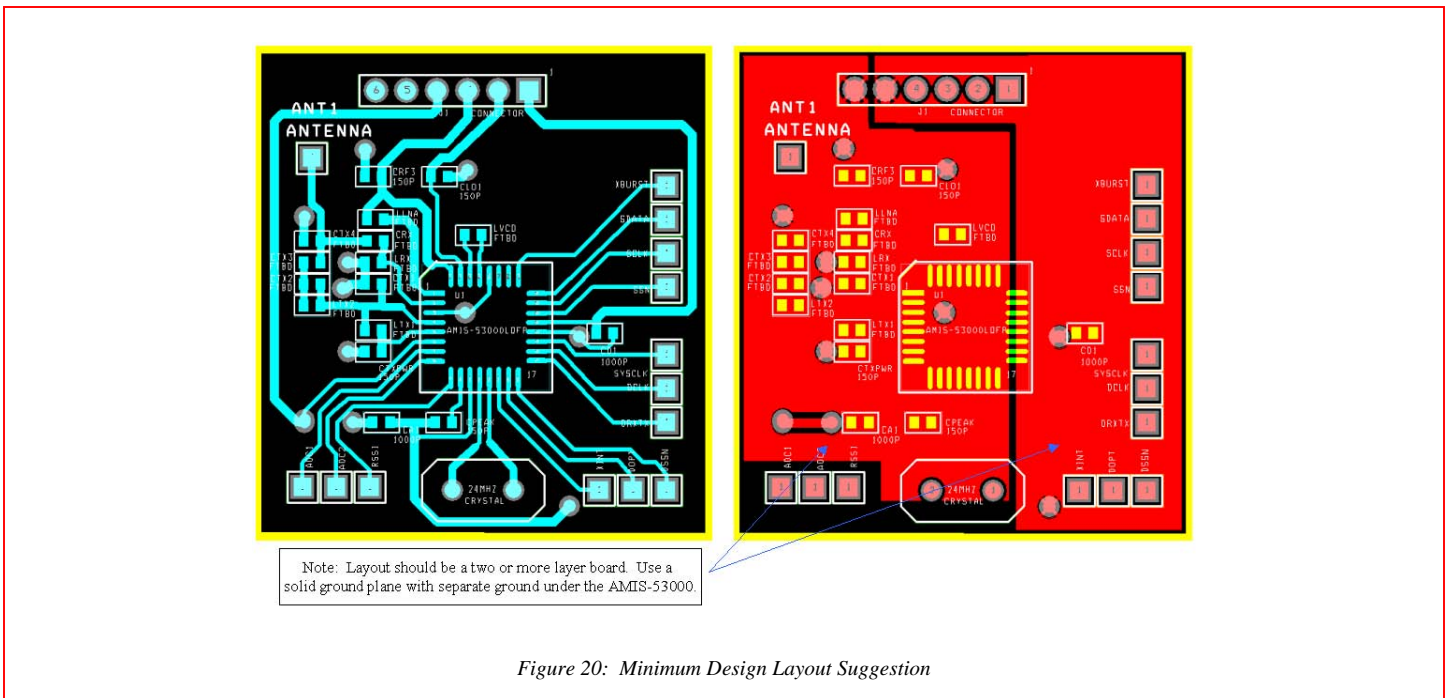


Figure 20: Minimum Design Layout Suggestion

## 6.0 User's Guide

This user's guide divides the control register description of the AMIS-53000 into functional areas; command register flow diagrams, frequency generation, receiver, transmitter, idle, data/control interfaces, Burst transmit, and MICS features.

### 6.1 Control Serial Interface Bus Description

Table 11: Control Interface Physical Configuration

| Interface        | Function | Clock |        | Data   |       | Select | AMIS-53000 |
|------------------|----------|-------|--------|--------|-------|--------|------------|
|                  |          | Pin   | Source | Output | Input |        |            |
| I <sup>2</sup> C | Control  | SCLK  | Master | SDATA  | SDATA | None   | Slave only |
| 3-Wire           | Control  | SCLK  | Master | SDATA  | SDATA | SSN    | Slave only |

The AMIS-53000 employs two different control interfaces. Communication with the AMIS-53000 control registers is through either a 3-wire bus or through a 2-wire (with third line for control/status) I<sup>2</sup>C compatible bus. The state of the control bus is detected by the AMIS-53000 at the first communication, I<sup>2</sup>C or 3-wire, and is set in that function (3-wire or I<sup>2</sup>C) as long as power remains applied to the part.

- 3-wire control communication bus
- I<sup>2</sup>C control communication bus
- AMIS-53000 is always the slave

#### 6.1.1. Control Interface Protocol

The AMIS-53000 control interface allows an external controller to write instructions to the registers of the AMIS-53000 and control the functions of the AMIS-53000. The external controller can also read the registers and status of the AMIS-53000. The control interface can be configured as a slave device in either a 2-wire I<sup>2</sup>C interface bus or a 3-wire serial interface.

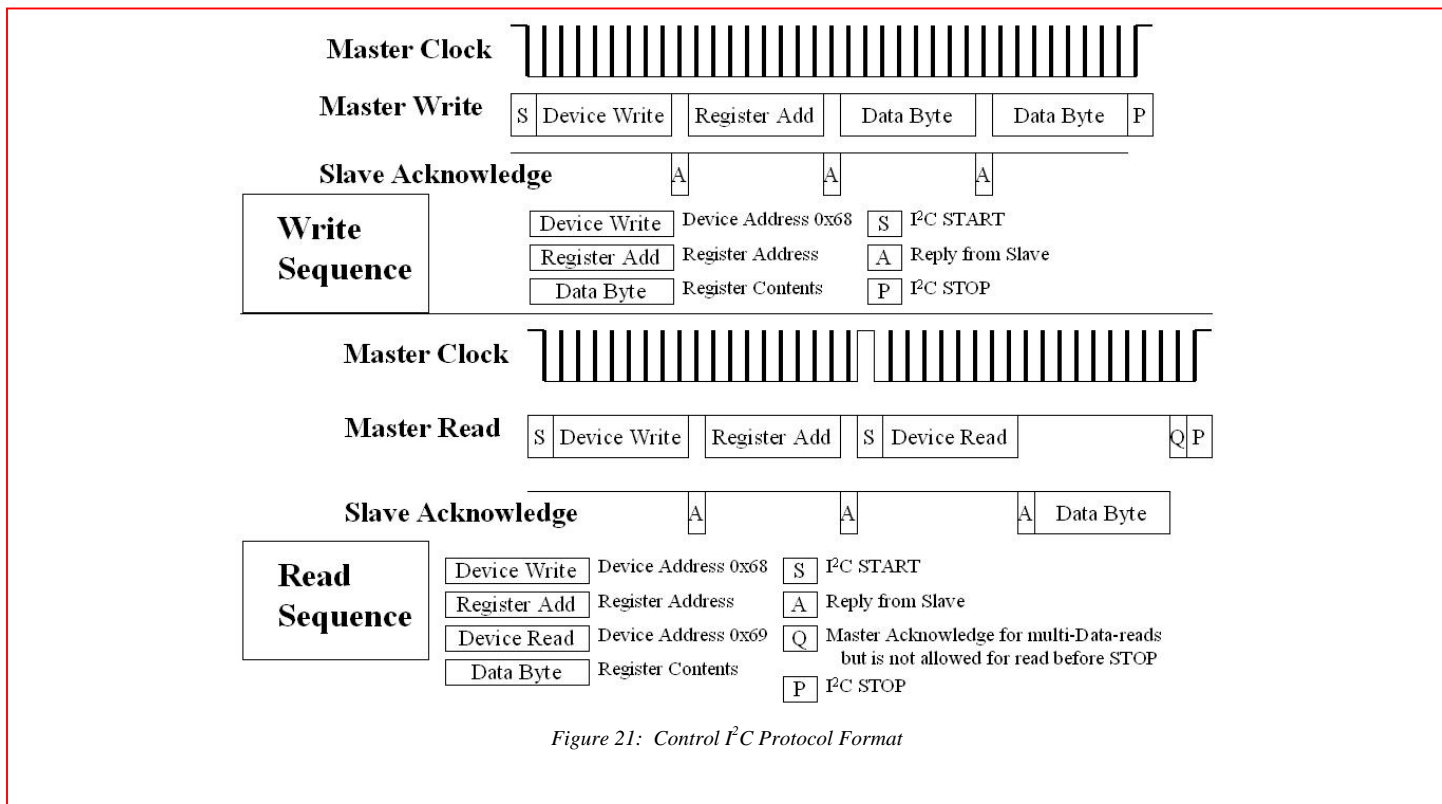


Table 12: I<sup>2</sup>C Addressing

| Address  | Description                         |
|----------|-------------------------------------|
| 0110100X | AMIS-53000 I <sup>2</sup> C Address |
| 01101000 | AMIS-53000 Write Command            |
| 01101001 | AMIS-53000 Read Command             |

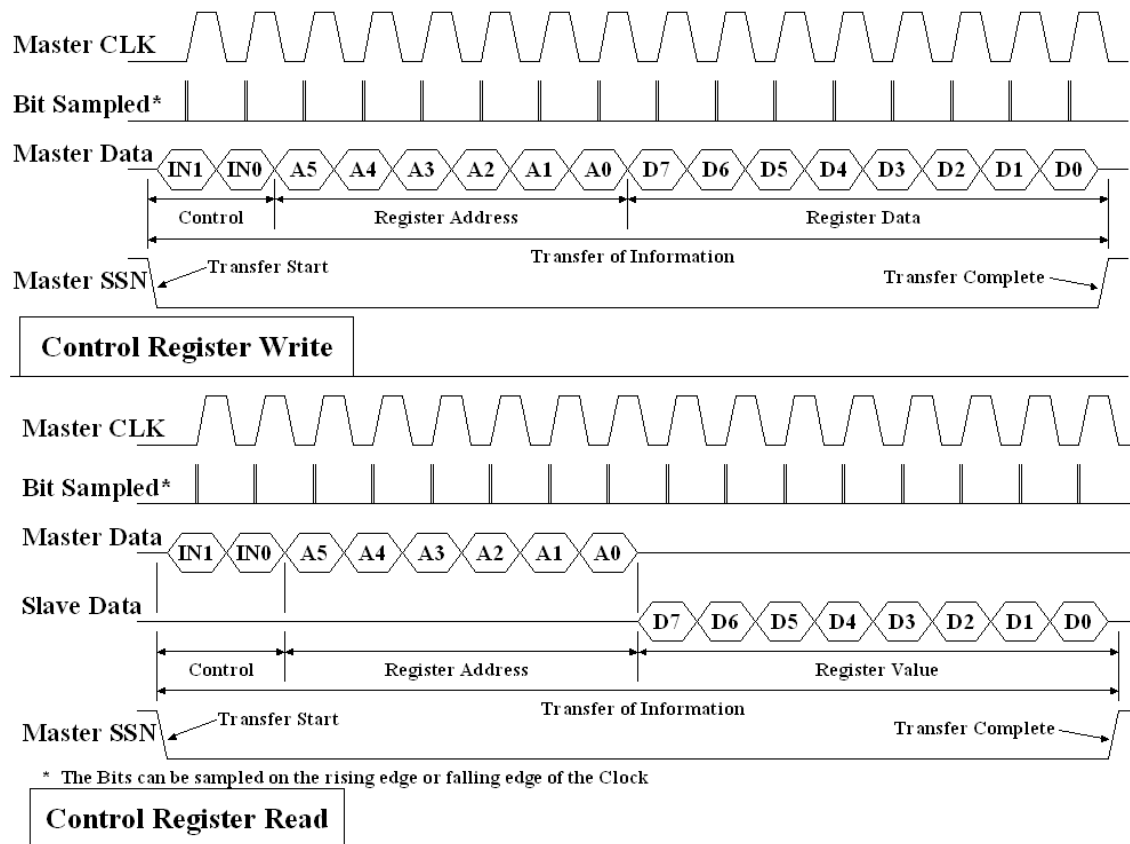


Figure 22: 3-Wire Control Protocol Format

Table 13: 3-Wire Control (IN1 and IN0)

| Control Word Bits |     | Description               |
|-------------------|-----|---------------------------|
| IN1               | IN0 |                           |
| 0                 | 0   | Single Register Read      |
| 0                 | 1   | Single Register Write     |
| 1                 | 0   | Sequential Register Read  |
| 1                 | 1   | Sequential Register Write |

- I<sup>2</sup>C device address:
  - 0x68 HEX for device write
  - 0x69 HEX for device read
- External controller can write registers

- External controller can read registers
- External controller can issue a immediate transmit via the xBURST input
- External controller can receive an interrupt (xINT) from the AMIS-53000

Setup registers descriptions:

**I<sup>2</sup>C/3-wire select**- First write to the interface sets the type of interface until AMIS-53000 is power cycled.

### 6.1.2. Serial Control Interface: Configuration

The AMIS-53000 can automatically detect the type of interface for the serial control bus. The interface pins are then given the definitions as shown in Table 14. The detection depends on the status of the AMIS-53000 pins as shown in Figure 23.

Table 14: Control Port Pin Definitions

| Pin Name | I <sup>2</sup> C Mode | 3-Wire Mode    |
|----------|-----------------------|----------------|
| SCLK     | SCL                   | SCLK           |
| SDATA    | SDA                   | R/W controlled |
| SSN      | Internal pull up      | SSN            |

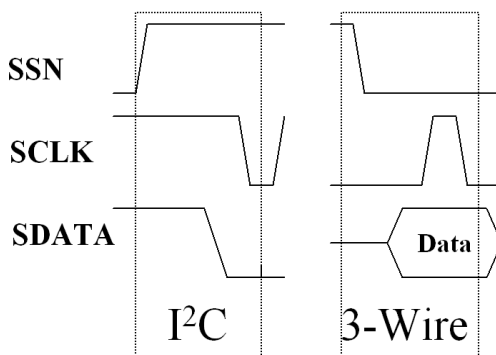


Figure 23: Control Interface Selection

Simply addressing the part with the desired protocol performs initial interface selection. After the first communication with the part, the selection is locked until power is removed from the device. The internal logic for determining which protocol to use on initial power up is as follows:

**I<sup>2</sup>C:** If SSN is high and an I<sup>2</sup>C start bit is detected, I<sup>2</sup>C mode is enabled.

**3-wire:** If SSN is low, and a negative edge on SCLK detected, 3-wire mode is enabled.

The internal pull ups on SCLK and SDATA can also be disabled for I<sup>2</sup>C applications using external pull ups.

Table 15: Control Interface Pull Up Control

| Mode             | SCLK, SDATA Pull Ups                                  | SSN Pin Configuration       |
|------------------|---|-----------------------------|
| I <sup>2</sup> C | Controlled by bit 3 of the general options A register | Not used (internal pull up) |
| 3-wire           | Controlled by bit 3 of the general options A register | SSN: normal mode            |

### 6.1.3. 3-Wire Interface Mode

The AMIS-53000 is always the slave device.

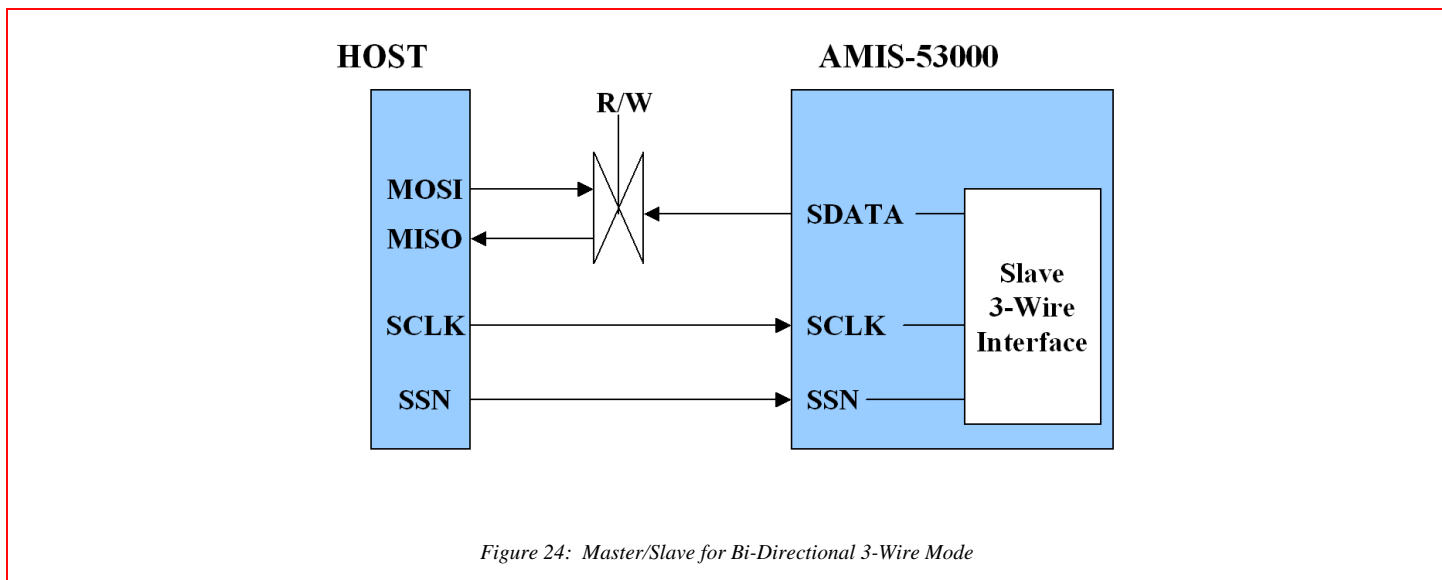


Figure 24: Master/Slave for Bi-Directional 3-Wire Mode

Figure 24 illustrates the connections between the master SPI port and the slave 3-wire port in the AMIS-53000.

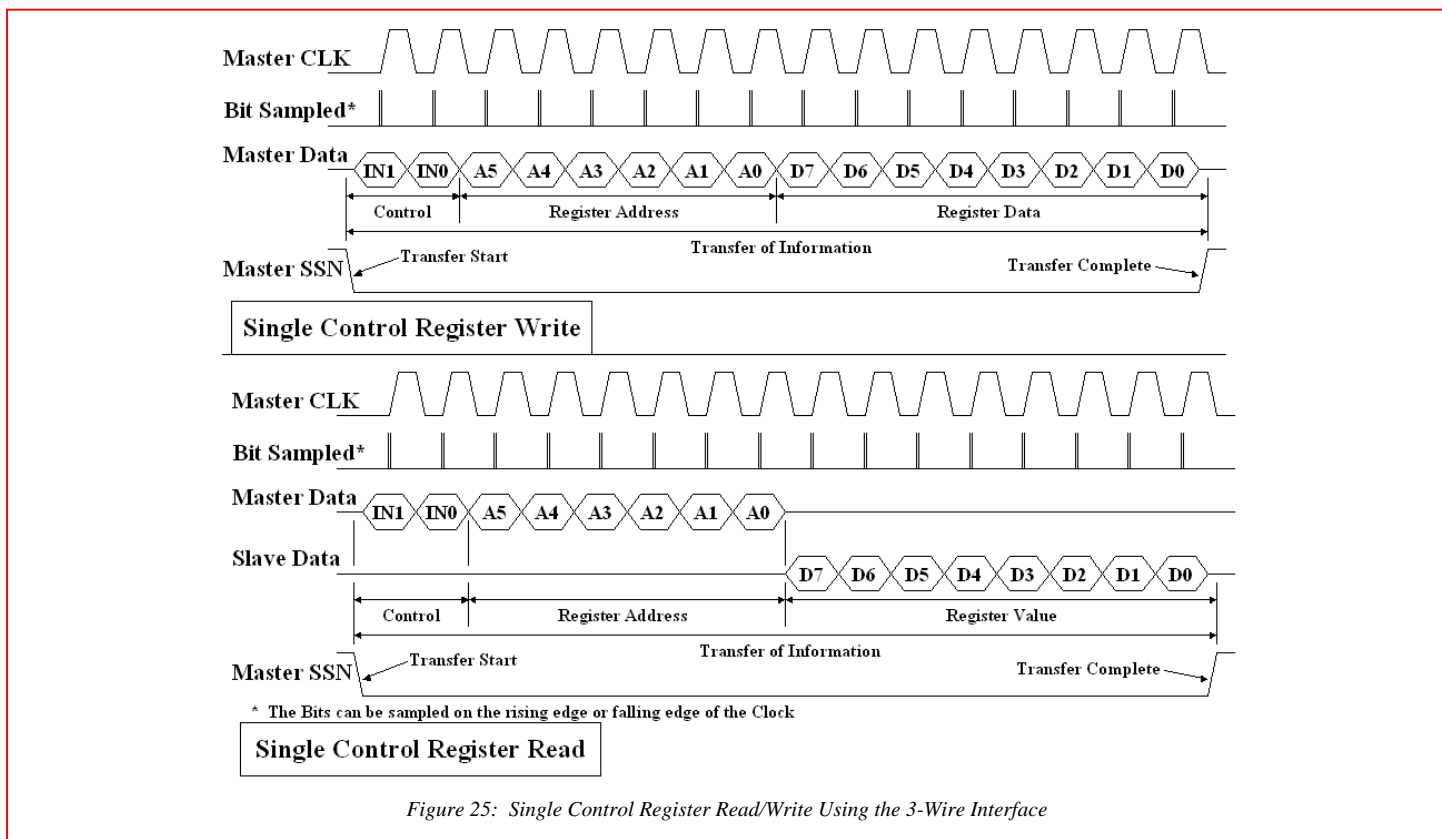


Figure 25: Single Control Register Read/Write Using the 3-Wire Interface



Figure 25 shows a single read or single write control data transfer. The operation starts with SSN transitioning low to indicate a start of transfer. The first two bits transferred are the instruction for the slave interface of the AMIS-53000, IN1 and IN0. Following the instruction are the six address bits to specify which address to read or write from. If the instruction is to write a register, the data to be written to address location A<5:0> is specified with the next 8 bits, D<7:0>. If the operation is a read, the slave output buffer is enabled at the end of the address bits, and the data bits D<7:0> are buffered out of the part MSB first.

For single read/write, the SSN line can remain active between successive read and write operations.

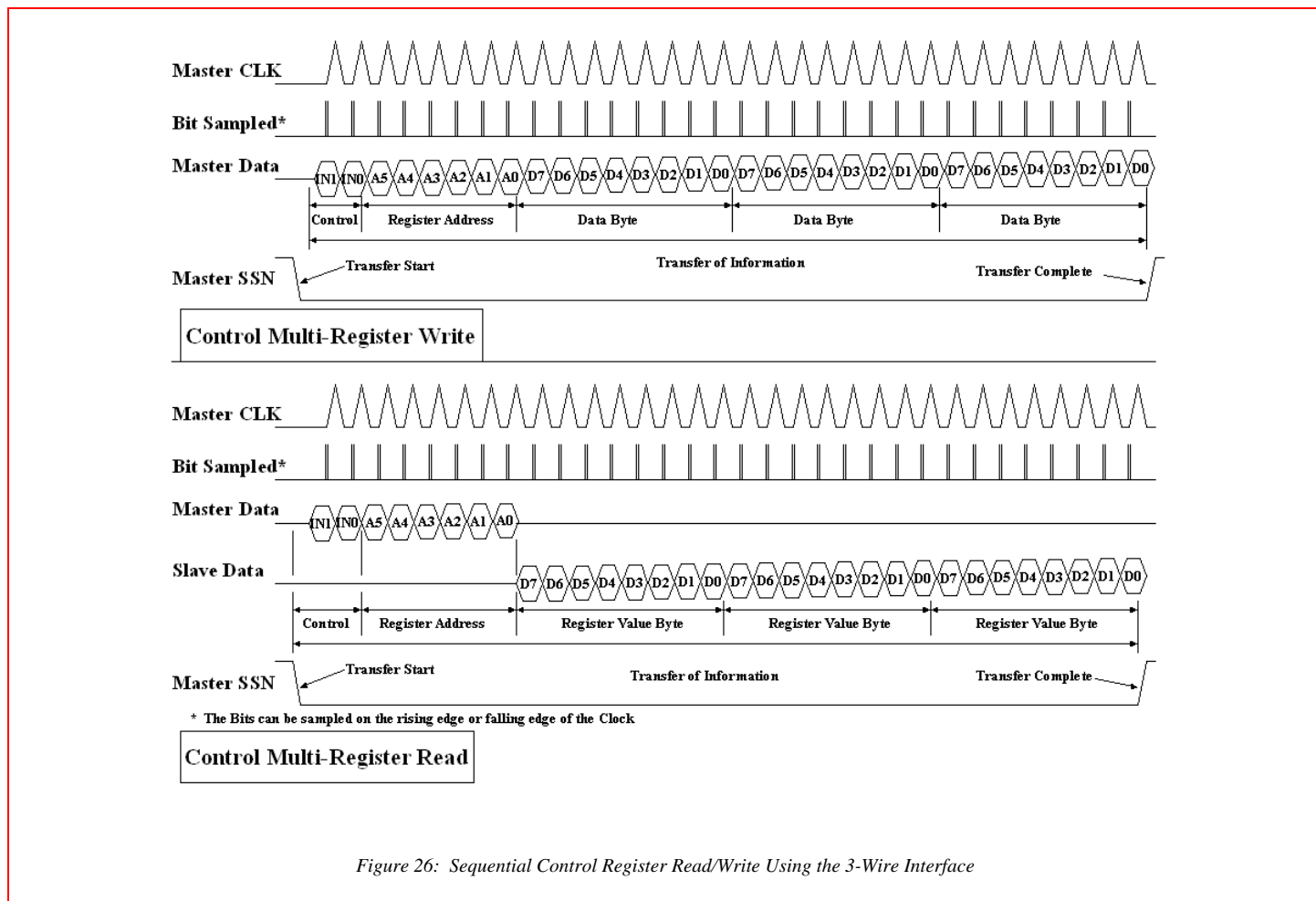


Figure 26: Sequential Control Register Read/Write Using the 3-Wire Interface

Figure 26 is a diagram for sequential reads or sequential writes for 3-wire control data transfer. The format of the instruction and address is identical to that for a single read/write operation, with the address corresponding to the first register location to read or write. The first 8 bits of data transferred correspond to the address selected. The address is internally incremented after each data byte transferred. This task is most useful for writing to or reading from variables spanning over multiple address locations such as the fractional PLL word (registers 03-05).

The SSN line must be de-asserted at the completion of a sequential read/write in order for the slave SPI controller to correctly interpret the next 8 bits as a command and not data.

### 6.1.4. I<sup>2</sup>C Interface

The I<sup>2</sup>C interface for the AMIS-53000 is compatible with the Philip Semiconductor I<sup>2</sup>C standard, with the AMIS-53000 as the slave device.



### 6.1.4.1. I2C Device Addressing

A control byte is the first byte received following the start condition from the master device. The control byte consists of a 7-bits for the device address, and 1-bit for a read or write command. For the AMIS-53000, the device address is '0110100' binary. The last bit of the control byte defines the operation to be performed. When set to '1', a read operation is selected. When set to '0', a write operation is selected. Following the start condition, the AMIS-53000 monitors the SDA bus checking the device type identifier being transmitted. Upon receiving its device address, the AMIS-53000 outputs an acknowledge signal on the SDA line. Depending on the state of the R/W bit, the AMIS-53000 will select a read or write operation.

### 6.1.4.2. Single Register Write

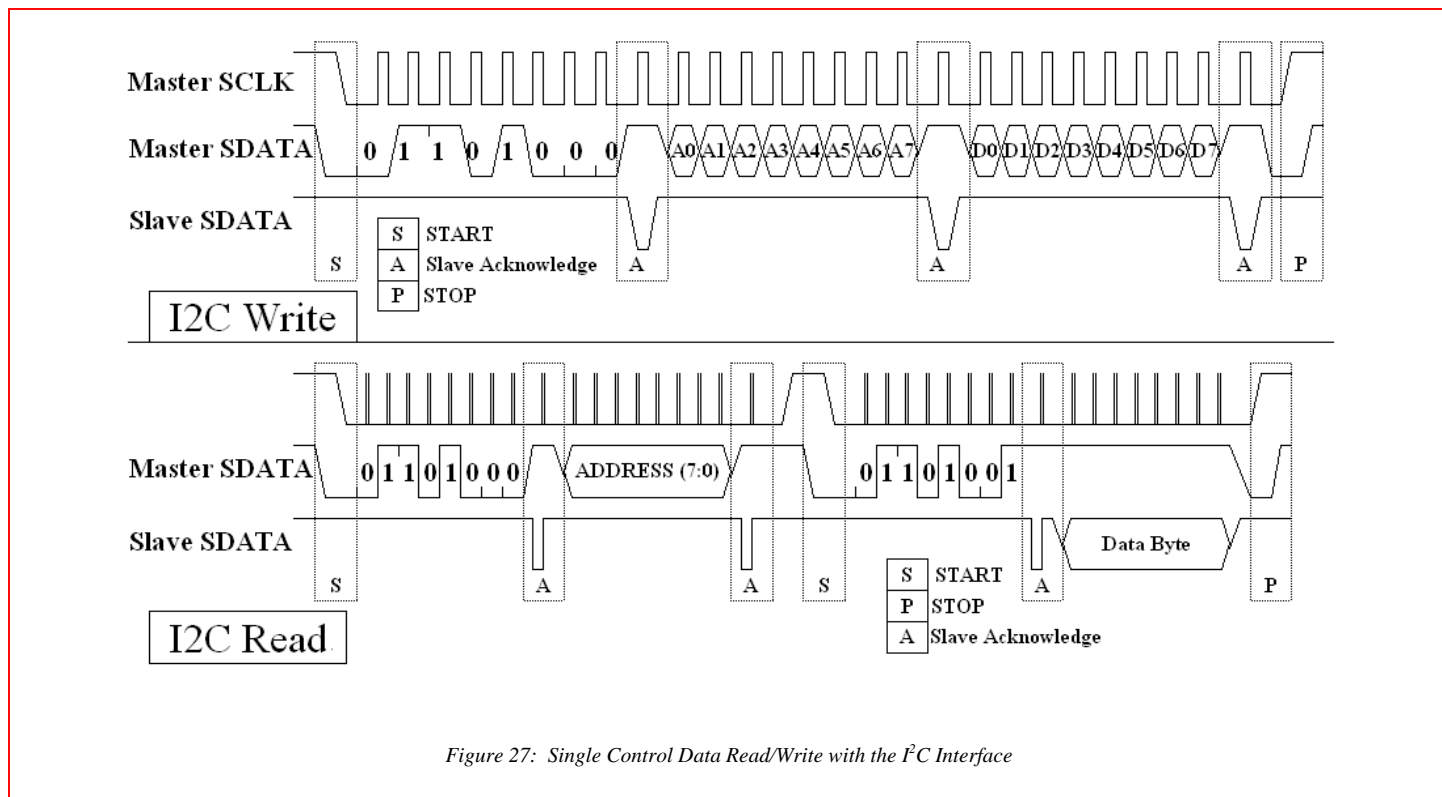


Figure 27: Single Control Data Read/Write with the I<sup>2</sup>C Interface

The master device issues the start condition, then issues the device address, and then issues the single R/W bit, a logic low state. This indicates to the addressed slave receiver that a byte with a register address will follow after the slave has generated an acknowledge bit during the ninth clock cycle. Therefore, the next byte transmitted by the master is the register address to be written with data. After receiving another acknowledge signal from the AMIS-53000, the master device will transmit the data word to be written, and the AMIS-53000 will acknowledge again. The write cycle ends with the master generating a stop condition.

A similar approach is used to read a register value. The master device issues the start condition, then issues the device address, and then issues the single R/W bit, a logic low state. This indicates to the addressed slave receiver that a byte with a register address will follow after the slave has generated an acknowledge bit during the ninth clock cycle. Therefore, the next byte transmitted by the master is the register address to be read. After receiving another acknowledge signal from the AMIS-53000, the master device will immediately follow with another start sequence, however, the R/W bit is now set high telling the slave device that the master wants the contents of the register (addressed with the write command) to be placed on the SDA bus line. After 8 bits of data are read by the master, the master does not acknowledge but sends the stop sequence.

### 6.1.4.3. Sequential Register Write

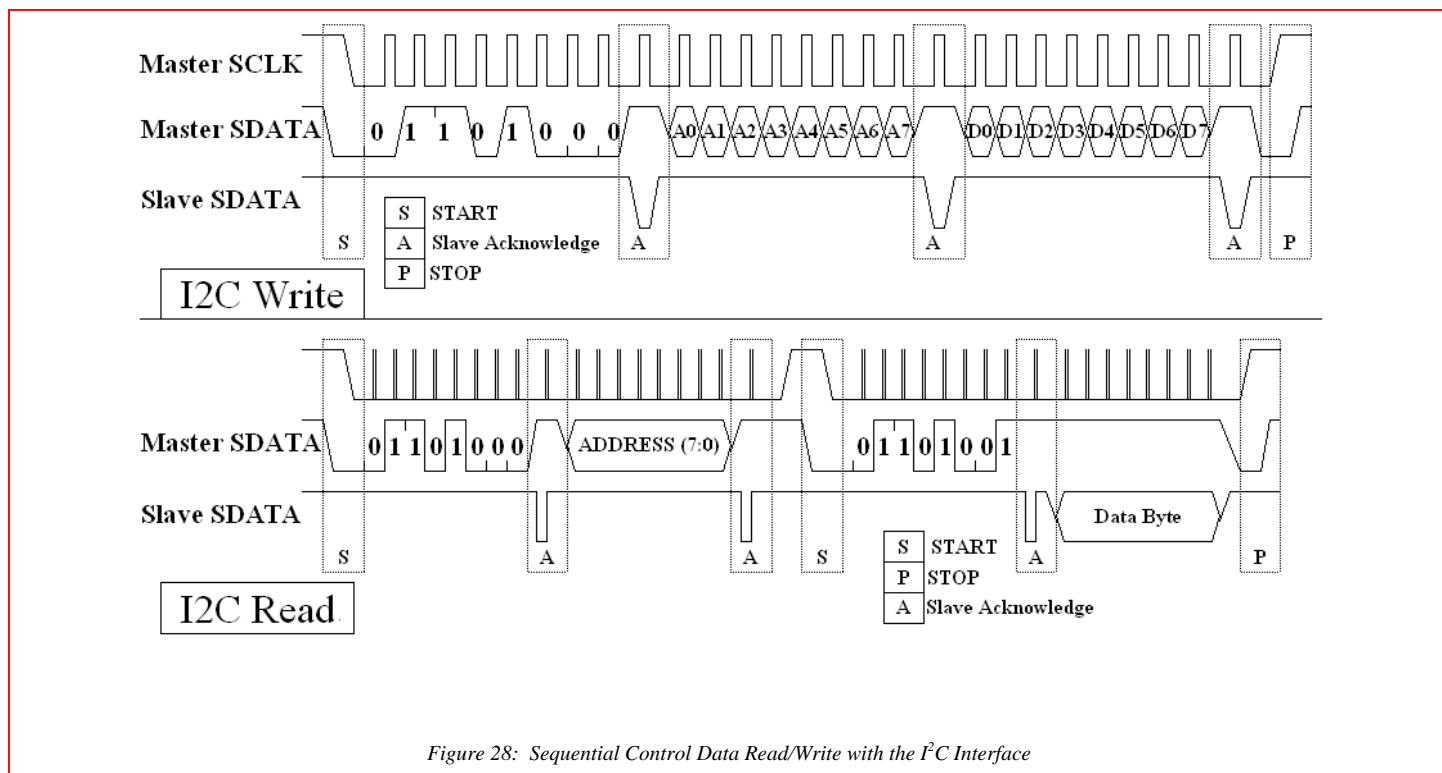


Figure 28: Sequential Control Data Read/Write with the I<sup>2</sup>C Interface

When setting the AMIS-53000 up for an application it sometimes is nice to write data to a number of registers one after the other. The write control byte, register address and first data byte are transmitted to the AMIS-53000 in the same way as in a byte write. However, instead of generating a stop condition, the master can continue to write register locations. Upon receipt of each word, the address is internally incremented by '1'. If the master should transmit more words than the AMIS-53000 has address locations, the address will roll over.

It is a similar approach to read a register value. The write control byte and register address are transmitted to the AMIS-53000 in the same way as in a byte write. After receiving another acknowledge signal from the AMIS-53000, the master device will immediately follow with another start sequence, however, the R/W bit is now set high telling the slave device that the master wants the contents of the register (addressed with the write command) to be placed on the SDA bus line. After the 8 bits are read by the master, the master acknowledges the reception. The AMIS-53000 will increment the register address and continue to output register values. After the last register value is received by the master, the master does not respond with an acknowledge but sends the stop sequence.

### 6.1.4.4. Current Address Read

The internal address counter maintains the last address addressed, incremented by '1'. If the last instruction received was to access register N, the current address read operation will read the contents from register N+1. The timing for the current address read is to send a start bit followed by the 7-bit device address, with the R/W bit set to one. The slave will acknowledge, after which the 8-bit register contents will be transmitted. The master does not acknowledge the transmission, but does generate a stop bit.

### 6.1.4.5. Interface Options

Table 16: I<sup>2</sup>C Address Auto Increment

| Register     |                   |      | Function  |
|--------------|-------------------|------|---|
| Number (HEX) | Name              | Bits |   |
| 0X0C         | General Options A | 3    | Disable the internal pull up resistors on SDATA and SSN lines                     |
| 0X0D         | General Options B | 5,4  | Select a clock rate for the interface when the AMIS-53000 is master               |
|              |                   | 3    | Disable auto increment for I <sup>2</sup> C control interface register addressing |

Additional interface options give the AMIS-53000 the flexibility to tailor the interface to specific requirements. These options are available in the interface options register, and can be stored into EE at board assembly to best suit the application.

### 6.1.4.6. Pull Up Disable

The AMIS-53000 includes built in pull up resistors for use with the I<sup>2</sup>C operation to reduce the overall system component count. The pull ups are asserted at POR until mode selection occurs. If mode is determined to be 3-wire, the pull ups are removed. If mode is determined to be I<sup>2</sup>C, this option bit determines whether the pull ups are to be removed.

## 6.2 Command Register

The AMIS-53000 contains a single 8-bit register that allows single writes to the register to place the AMIS-53000 into a desired mode. It is very important to remember that all registers associated with that mode must be preprogrammed for the single write to this register to operate correctly.

The command register allows the user application to issue a single register write to the AMIS-53000 to initiate the function listed in Table 17. When the function is started, the AMIS-53000 uses the register values associated with the function selected as the parameters of the function. These register values will be the default values or the values the user application has written to the registers before the function is called out in the command register.

Table 17: Command - 0X00 [0]

| Bit   | Command          | Comment  |  |
|-------|------------------|--|--|
| [7:4] | [3:0]            |  |  |
| 0000  | 0000             | Standby  | Put the part into standby  |
|       | 0001             | Receive  | Put the part into receive mode                                       |
|       | 0010             | Transmit   | Put the part into transmit mode                                      |
|       | 0011             | Idle   | Put the part into idle mode  |
|       | 0100             | Idle Return  | Use to return to idle after interrupt for HK or receive during sniff |
|       | 0101             | Write EE   | Write the content of the working registers into EE                   |
|       | 0110             | Read EE  | Read the contents of the EE  |
|       | 0111             | Calibrate QS Osc                                       | Calibrates the Quick Start oscillator                                |
|       | 1000             | Calibrate RC   | Calibrates the 10kHz RC oscillator                                   |
|       | 1001             | Calibrate PLL  | Calibrates the PLL   |
|       | 1010             | Calibrate LNA  | Calibrates the LNA matching  |
| 1011  | ROM2Regs         | Write the content of the ROM into the shadow registers |  |
| 1111  | Global Reset     | Resets the part completely                             |  |
| [7:6] | [5:0]            |  |  |
| 01    | XXXXXX<br>100000 | Single ADC Conversion                                  | Do an ADC one time on the channel selected loop filter output        |
| 11    | XXXXXX<br>100000 | Continuous ADC Conversions                             | Do an ADC on the channel selected continuously loop filter output    |

Standby is both a state for the transceiver, and a command given to the transceiver. The actual operation of standby mode can be either a low-power mode where all internal circuitry of the AMIS-53000 except for the interface will be disabled, or a clock-only mode where the crystal oscillator will be enabled to continue providing a system clock for an external microprocessor. A bit available in the general options A register allows selection of power-down or clock-only operation in standby mode.

Many of the instructions for the part are finite in duration and the end point of the task is controlled by the AMIS-53000, such as the calibration instructions. For these instructions, the AMIS-53000 will return to its standby mode at the completion of the task. The user may poll the Status/Flag 1 register busy bit to determine the completion of the calibration instruction. Other instructions such as receive or transmit are indefinite in length and user controlled. To return to the standby state, the AMIS-53000 waits for the standby instruction to end receive or transmit, at which point the transceiver will return to its standby state.

Note that there are two low-power modes for the AMIS-53000; standby and idle.

- Standby** allows the SYSCLK output
- Idle** is the very low power state without SYSCLK output

### 6.3 Functional Flow Diagrams

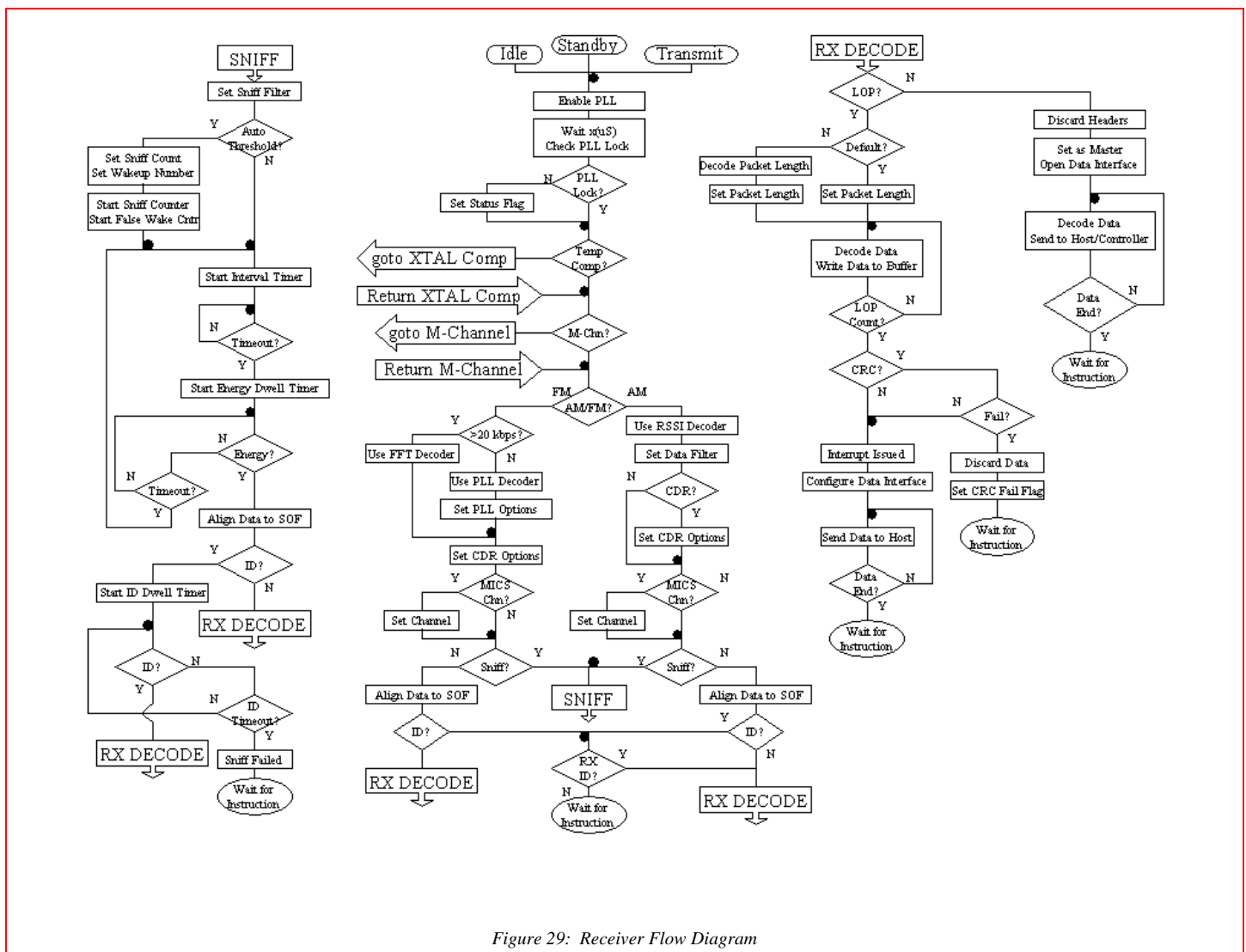


Figure 29: Receiver Flow Diagram

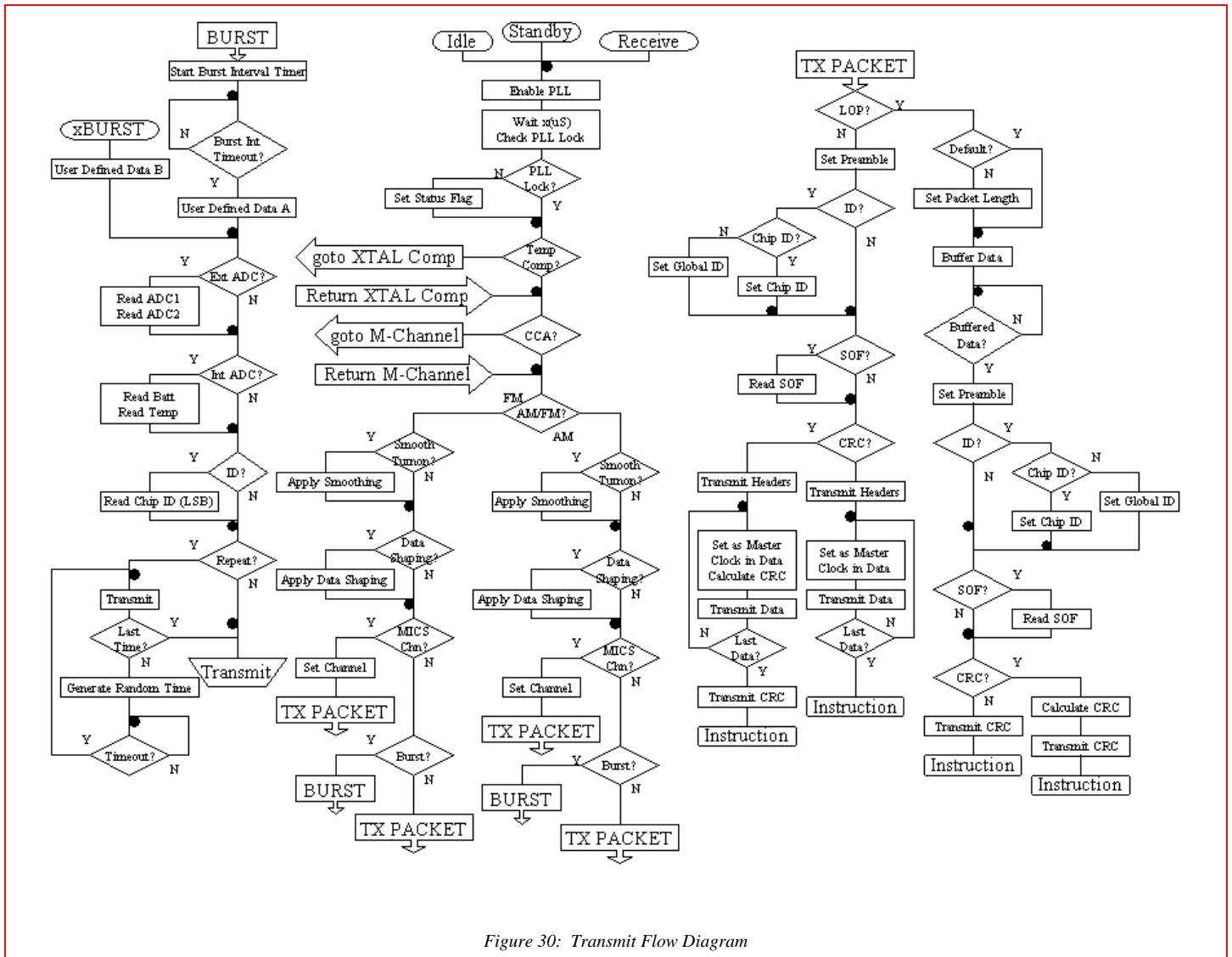


Figure 30: Transmit Flow Diagram

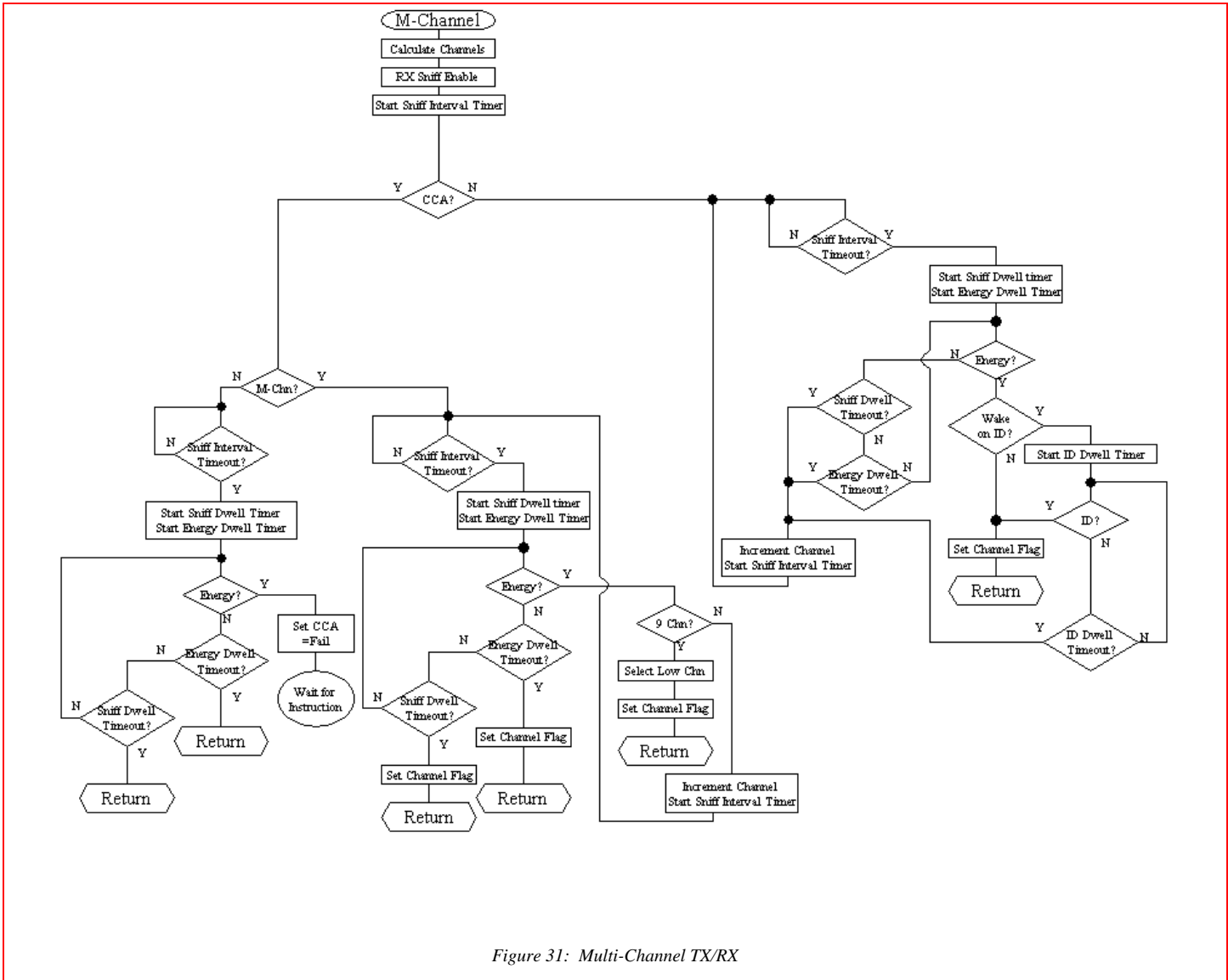


Figure 31: Multi-Channel TX/RX

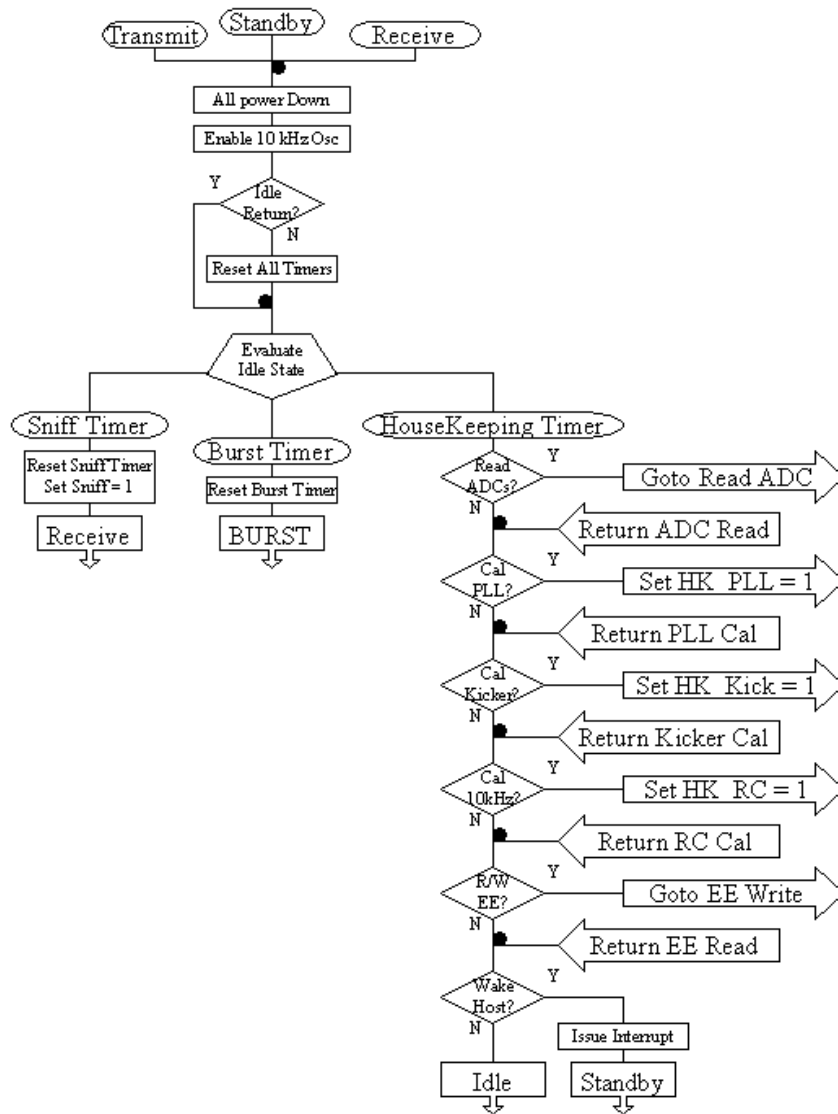


Figure 32: Idle State Flow Diagram

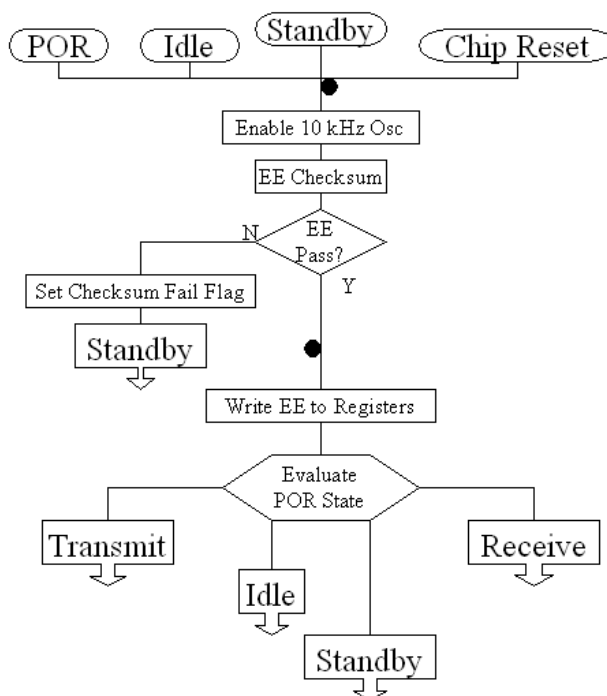


Figure 33: EE Flow Diagram

## 6.4 Frequency

The AMIS-53000 uses an internal PLL/VCO to generate the RF frequencies for both transmit and receive. Only one set of registers needs to be programmed to generate the TX frequency and to generate the LO frequency to produce the mixing frequency for converting the received signal to the low IF (about 500kHz). The AMIS-53000 can do a self-calibration, which will trim internal capacitance to tune the TX and RX frequencies. The AMIS-53000 needs to run the self-calibration (must be started by the external host/controller) at least once between startup and entering a command such as TX, RX, Sniff, etc.

### 6.4.1. Frequency Control

AMIS has developed an executable program (AMIS-53CALC.exe, available from AMIS, which generates the register values for the frequency divider and fractional word to produce a given frequency. First, the AMIS-53000 must have the correct LC for the desired frequency connected to the VCO pins. There is internal capacitance that is part of the capacitance for determining the value of the inductor. The following equation can be used to determine the approximate value of the LC components. Remember that the VCO is sensitive to the placement of the LC components - they should be placed as close to the AMIS-53000 as practical (even short traces add significant parasitics) and the traces to the components should be made symmetrical.

$$F_{vco} = \frac{1}{\sqrt{L_{tot} \cdot C_{tot}}} \cdot \frac{1}{2\pi}$$

Where: Ltot and Ctot are the total inductance and capacitance respectively at the VCO pins. This includes the internal capacitance of approximately 2pF.

The RF PLL is a 24-bit Sigma Delta based fractional N synthesizer used to provide the LO signal for receive, and a direct RF output for transmit.



Setup registers descriptions:

**RF Divider-** The RF frequency of the receiver must be configured. This is done in two steps, one is setting the RF divider and the other is setting the fractional N word.

**RF Frequency-** Program the 3 register fractional N word.

**Peak Deviation-** When the data modulation is to be FSK, the 2 register peak deviation must also be set. The deviation should be set to a value between ½ and 1 times the data rate.

**PLL-** Configure the parameters for the PLL.

**Loop Filter-** Configure the parameters for the loop filter.

### 6.4.1.1. RF Divider

Setting the RF channel frequency is done through the RF divider register, along with the RF frequency 2, 1 and 0 registers. The RF divider register is used to specify the integer portion of the divide value, and the RF frequency 2, 1 and 0 registers are used to specify the fraction. The values are calculated as follows:

$$\text{DivideVal} = \frac{F_{\text{Channel}}}{24\text{MHz}}$$

Where  $F_{\text{Channel}}$  is the desired RF center frequency. The value for the RF divider register is found by,

$$\text{Integer} = \text{round}(\text{DivideVal})$$

Where integer is the value used for RF divider. The last step is to calculate the fractional value. This is done as,

$$\text{Fraction} = (\text{DivideVal} - \text{Integer}) \cdot 262147$$

Fraction is the value to be used in the RF frequency 2, 1 and 0 registers. As an example, if the desired RF frequency channel is 903.5MHz,

$$\text{DivideVal} = \frac{903.5\text{Mhz}}{24\text{MHz}} = 37.6458333$$

$$\text{Integer} = \text{round}(\text{DivideVal}) = 38 = 0x26$$

$$\text{Fraction} = (\text{DivideVal} - \text{Integer}) \cdot 262147 = -92844 = 0xFE\_95\_54$$

For this example, the RF divider register is written to 0x26, RF frequency 2 is written to 0xFE, RF frequency 1 to 0x95, and RF frequency 0 to 0x54. This value +/- 1 is fed directly to the PLL as N0 and N1. (i.e. if 63, send 64 and 65 to the PLL)

Table 18: RF Divider - 0X05 [5]

| Bit | Name            | Comment   |
|-----|-----------------|---|
| 7:0 | RF_divide [7:0] | 00h through 0Bh: not allowed<br>1Ah: divide by 26<br>1Bh: divide by 27<br>-----<br>4Ah: divide by 74<br>4Bh: divide by 75<br>4Ch through FFh: not allowed |

### 6.4.1.2. RF Frequency 2

Table 19: RF Frequency 2 = 0X06 [6]

| Bit | Name            | Comment                         |
|-----|-----------------|---------------------------------|
| 7:0 | RF_FREQ [23:16] | Upper 8 bits of the RF fraction |

### 6.4.1.3. RF Frequency 1

Table 20: RF Frequency 1 - 0X07 [7]

| Bit | Name           | Comment                          |
|-----|----------------|----------------------------------|
| 7:0 | RF_FREQ [15:8] | Center 8 bits of the RF fraction |

### 6.4.1.4. RF Frequency 0

Table 21: RF Frequency 0 - 0X08 [8]

| Bit | Name          | Comment                         |
|-----|---------------|---------------------------------|
| 7:0 | RF_FREQ [7:0] | Lower 8 bits of the RF fraction |

### 6.4.1.5. Peak Deviation 1

The peak deviation for FSK transmissions is determined by the peak deviation 1 register and the peak deviation 0 register. This value is also used inside the DFT FSK detector. Calculation of the value for the peak deviation is straightforward:

$$PEAK\_DEV = \frac{\text{Peak\_Deviation (Hz)}}{91.5517}$$

The result of this equation, converted to Hex, is entered into the peak deviation registers.

Table 22: Peak Deviation 1 - 0X09 [9]

| Bit | Name            | Comment                            |
|-----|-----------------|------------------------------------|
| 7:0 | PEAK_DEV [15:8] | Upper 8 bits of the peak deviation |

### 6.4.1.6. Peak Deviation 0

Table 23: Peak Deviation 0 - 0X0A [10]

| Bit | Name           | Comment                            |
|-----|----------------|------------------------------------|
| 7:0 | PEAK_DEV [7:0] | Lower 8 bits of the peak deviation |

### 6.4.1.7. RF PLL Options

Contains general options for the setup of the RF PLL.

Table 24: RF PLL Options - 0X28 [40]

| Bit | Name                           | State | Comment  |
|-----|--------------------------------|-------|--|
| 7   | Kicker Calibration Status      | 1     | The kicker has been calibrated   |
|     |                                | 0     | The kicker has not been calibrated   |
| 6   | Temperature Compensation Curve | 1     | Use the Type 1 compensation for external crystal with curves similar to Type 1 (See Figure 32) |
|     |                                | 0     | Use the Type 2 compensation for external crystal with curves similar to Type 2 (See Figure 33) |
| 5   |                                |       |  |
| 4   |                                |       |  |
| 3   | Internal Loop Filter           | 1     | Enable using the internal loop filter for the PLL (used for calibration)                       |
|     |                                | 0     |  |
| 2   | Charge Pump Current            | 1     | 50uA   |
|     |                                | 0     | 25uA   |
| 1   | lvco[1:0]                      | 11    | lvco= 1.2mA  |
|     |                                | 01    | lvco= 800uA  |
|     |                                | 10    | lvco= 600uA  |
|     |                                | 00    | lvco= auto level control   |

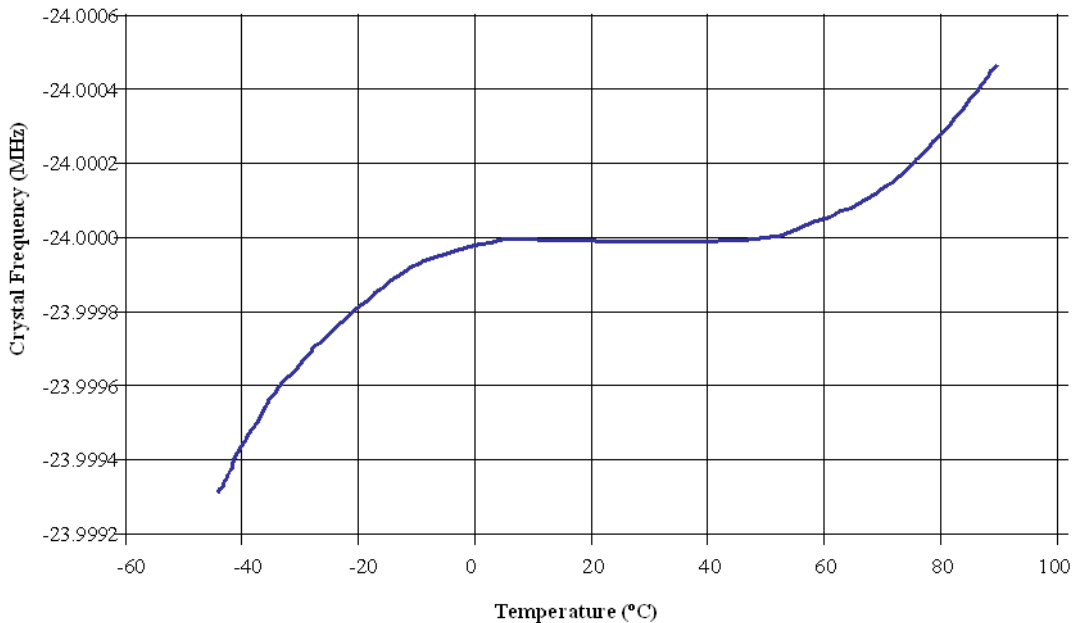


Figure 34: Typical Crystal Temperature Curve for Crystal with Type 1 Characteristics

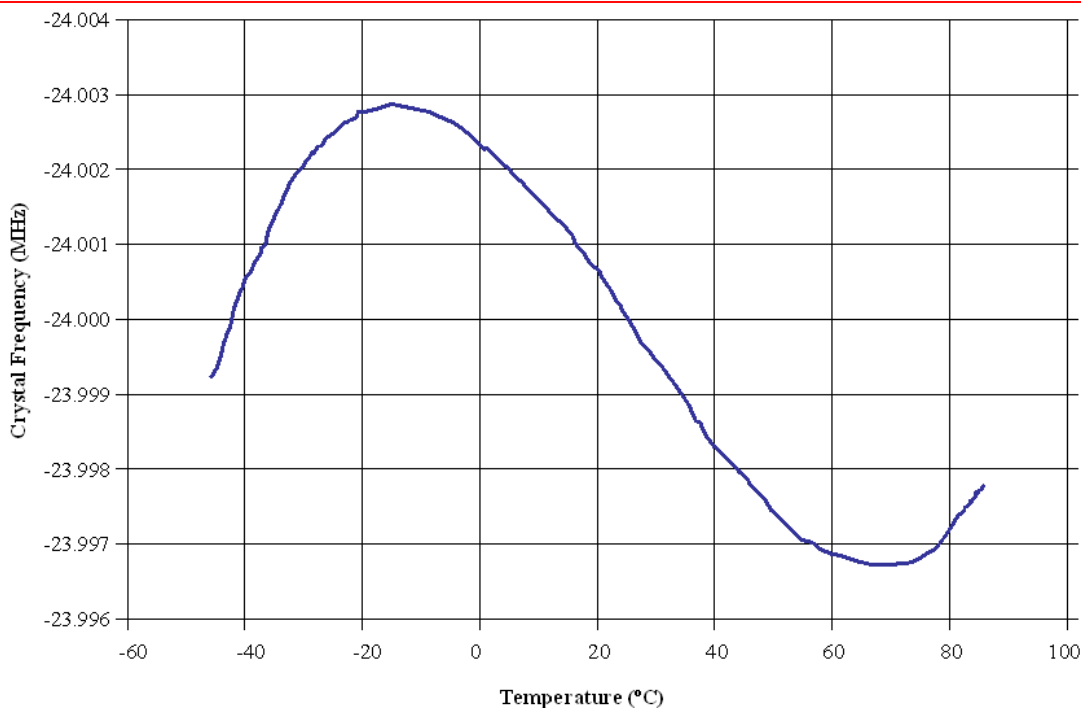


Figure 35: Typical Crystal Temperature Curve for Crystal with Type 2 Characteristics

### 6.4.1.8. Loop Filter

Table 25: Loop Filter - 0X39 [57]

| Bit | Name              | Comment              |
|-----|-------------------|----------------------|
| 7:0 | LOOP_FILTER [7:0] | Internal loop filter |

### 6.4.2. 10kHz Oscillator

The AMIS-53000 has an internal 10kHz oscillator. This oscillator is running whenever the AMIS-53000 is in standby or idle modes. This very low power oscillator provides the clock for timing functions such as Sniff receive, Burst transmit or housekeeping. The oscillator is trimmed in the calibration instructions.

Setup registers descriptions:

**10k Oscillator Trim-** The value of the calibration for the 10kHz oscillator. (See Section 6.10.1.4)\*\*

### 6.4.3. System Clock

The AMIS-53000 provides a divided version of the external reference oscillator (typically 24MHz) as an output to an external host/controller or other circuits needing a clock.

Table 26: System Clock Control

| Register     |                   |      | Function   |
|--------------|-------------------|------|--|
| Number (HEX) | Name              | Bits |  |
| 0X0C         | General Options A | 2,1  | 11 POR starts in TX                                |
|              |                   |      | 10 POR starts in RX                                |
|              |                   |      | 01 POR starts in idle                              |
|              |                   |      | 00 POR starts in standby                           |
|              |                   | 0    | Standby mode with system clock output              |
| 0X0D         | General Options B | 1,0  | 11 External XTAL reference divided by 2 (12MHz)    |
|              |                   |      | 10 External XTAL reference divided by 4 (6MHz)     |
|              |                   |      | 01 External XTAL reference divided by 8 (3MHz)     |
|              |                   |      | 00 System clock off                                |
| 0X10         | Idle Config       | 4,3  | 11 System clock continues for 1024 clock cycles    |
|              |                   |      | 10 System clock continues for 512 clock cycles     |
|              |                   |      | 01 System clock continues for 256 clock cycles     |
|              |                   |      | 00 System clock shuts down after idle command ASAP |

Setup registers descriptions:

**Crystal Oscillator Trim-** The value of the calibration for the 10kHz oscillator. (See Section 6.10.1.1)\*\*

#### 6.4.4. Quick Start

The AMIS-53000 includes the ASTRIC family patented Quick Start oscillator. This circuit uses a “kicker” to force the crystal oscillator close to the final desired frequency. This reduces the time required for the crystal oscillator to settle to the RF frequency.

Table 27: Kicker Calibration

| Register     |                |      | Function                  |
|--------------|----------------|------|---------------------------|
| Number (HEX) | Name           | Bits |                           |
| 0X28         | RF PLL Options | 7    | Kicker calibration status |

Setup registers descriptions:

**Quick Start Trim-** The value of the calibration for the kicker. (See Section 6.10.1.3)\*\*

#### 6.4.5. Self Calibration

The AMIS-53000 has internal trim functions for the PLL, TX PLL, RX PLL, 10kHz oscillator, and kicker (Quick Start). A self calibration is started by writing an instruction to the command register. This self calibration needs to be done at least once after the AMIS-53000 has been powered on and before the AMIS-53000 is placed into any mode such as transmit or receive. The application should monitor the status registers and trim value registers to determine that the calibration was successful.

Table 28: Self Calibration Command

| Register     |                     |       | Function                             |
|--------------|---------------------|-------|--------------------------------------|
| Number (HEX) | Name                | Code  |                                      |
| 0X00         | Command             | 0X07  | Calibrate the Quick Start (kicker)   |
|              |                     | 0X08  | Calibrate the 10kHz oscillator       |
|              |                     | 0X09  | Calibrate the PLL                    |
|              |                     | 0X0A  | Calibrate the LNA                    |
| 0X1B         | Housekeeping Config | Bit 2 | Calibrate PLL during HK              |
|              |                     | Bit 1 | Calibrate 10kHz oscillator during HK |
|              |                     | Bit 0 | Calibrate kicker during HK           |

Setup registers descriptions:

**Status-** Contains the results of calibrations, instructions and activity in the AMIS-53000.

**Software State-** Shows the current mode of the AMIS-53000.

### 6.4.5.1. Status/Flag1

The purpose of the Status1 register is to provide information back to the host on the status of the part. This register should be queried at the completion of calibration sequences to ensure proper operation. The flags will be reset when the register is read.

Checksum indicates whether an attempt to read or write the EE has failed due to an incorrect CheckSum.

Instruction enable indicates that the AMIS-53000 is ready to receive an instruction. This can be used to insure that the AMIS-53000 does not miss a command instruction due to the AMIS-53000 not being ready. Along with the busy flag, these status flags can police the flow of commands to the AMIS-53000.

Table 29: Status/Flag1 - 0X01 [1]

| Bit | Name               | State | Comment  |
|-----|--------------------|-------|--|
| 7   | PLL xLock          | 1     | PLL out of lock on startup (RX, TX, Sniff, Burst)                      |
|     |                    | 0     |  |
| 6   | TX PLL Cal         | 1     | PLL calibration for transmit failed                                    |
|     |                    | 0     |  |
| 5   | RX PLL Cal         | 1     | PLL calibration for receive failed                                     |
|     |                    | 0     |  |
| 4   | RC Cal             | 1     | 10kHz RC oscillator calibration failed                                 |
|     |                    | 0     |  |
| 3   | Quick Start Cal    | 1     | Quick Start calibration failed   |
|     |                    | 0     |  |
| 2   | Checksum           | 1     | EE CheckSum failed   |
|     |                    | 0     |  |
| 1   | Instruction Enable | 1     | The AMIS-53000 is in a state of operation that can accept instructions |
|     |                    | 0     |  |
| 0   | ADC Done           | 1     | ADC conversion complete  |
|     |                    | 0     |  |

### 6.4.5.2. Status/Flag2

The Status2 register provides information on the operating status of the part. The busy bit is asserted for any of the following reasons:

**Calibration:** The busy bit will remain high for the duration of a calibration sequence. Status2 can be repeatedly polled during a calibration sequence to determine when it's complete.

**Read/Write EE:** While the AMIS-53000 is reading from or writing to the EE, the busy bit will remain set.

**Buffered RX:** When in receive mode, and a valid chip ID is found, the AMIS-53000 will begin processing of this packet. During the time the packet is being processed, the busy bit will be set high.

**Buffered TX:** After the command is given for transmit with the buffered packet option enabled, the busy bit will remain high until the part has completed the actual transmission of the packet.

**Housekeeping:** Busy is asserted during a housekeeping cycle.

**Burst TX:** Busy is asserted during a Burst transmission.

Status2 also contains information on the reason an interrupt was issued to the external host. The CCA channel status bits provide information back to the host on which channel within the MICS band is being used for communication. After a CCA enabled transmission, these bits will be set to indicate which channel was used. For MICS enabled receivers performing multi-channel Sniff, these bits are used to indicate the channel upon which either energy or an entire packet was found.

If CCA is enabled for operation other than MICS, bit 0 (CCA failed) is used to indicate whether or not the channel is clear. The flags will be reset when the register is read.

Table 30: Status/Flag2 - 0X02 [2]

| Bit | Name              | State | Comment                               |
|-----|-------------------|-------|---------------------------------------|
| 7:4 | CCA Channel [3:0] |       | Indicates channel selected during CCA |
| 3:1 | Interrupt Type    | 111   | RX CRC failed                         |
|     |                   | 110   | Receive energy dwell timer timed out  |
|     |                   | 101   | CCA failed                            |
|     |                   | 100   | Transmit complete                     |
|     |                   | 011   | Buffer data for TX                    |
|     |                   | 010   | Data has been received                |
|     |                   | 001   | Housekeeping                          |
|     |                   | 000   | Low battery                           |
| 0   | Busy              | 1     | AMIS-53000 is busy                    |
|     |                   | 0     |                                       |

### 6.4.5.3. Software State

Displays the current mode of the AMIS-53000. This status register can be used to monitor the activity of the AMIS-53000.

Table 31: Software State - 0X3C [60]

| Bit | Name           | State | Comment                                     |
|-----|----------------|-------|---|
| 7:0 | Software State | 1111  | Undetermined                                |
|     |                | 1011  | Startup                                     |
|     |                | 1010  | Copying ROM data to registers               |
|     |                | 1001  | Calibrating LNA                             |
|     |                | 1000  | Calibrating PLL                             |
|     |                | 0111  | Calibrating 10kHz oscillator                |
|     |                | 0110  | Calibrating Quick Start (kicker) oscillator |
|     |                | 0101  | Reading EE data                             |
|     |                | 0100  | Writing EE data                             |
|     |                | 0011  | Idle  |
|     |                | 0010  | Transmitting                                |
|     |                | 0001  | Receiving                                   |
|     |                | 0000  | Standby                                     |

## 6.5 Receiver

The AMIS-53000 receiver is designed for either on/off shift key (AM) modulated signals or frequency shift key (FM) modulated signals. The receiver includes all the circuitry to recover data from either the OOK or the FSK modulated signal carrier. The receiver operates on fixed frequencies in the operating frequency range of 300 to 928MHz using an internal fractional N PLL to set the frequency. The receiver can reduce power consumption using the Sniff Mode to acquire the incoming signal. The receiver can set a user defined fixed threshold for data detection or it can form a threshold from the incoming signal for determining the presence of signal and the state of the recovered waveform. The receiver can use a synchronous data detector to extract the data clock and the data from the incoming signal (FSK modulation always uses this method of data detection).

- OOK modulation (AM)
  - Manchester encoding option
  - CDR data detection option (recommended that this be used)
  - Common data rates from 1.2kbps to 19.2kbps or user defined
- FSK/GFSK modulation (FM)

- Manchester encoding
- FFT or PLL demodulation (depends on data rate)
- Common data rates from 1.2kbps to 128kbps or user defined

The AMIS-53000 receiver is a low IF frequency single down conversion, sub-sampling, image rejection architecture with a common AM/FM IF chain. Three demodulators are used for signal detection with additional post-detection and filtering capabilities for data recovery.

- A complex FFT demodulator is used for FSK signals with data rates > 20kbps.
- A digital PLL demodulator is used for FSK signals with data rates less than 20kbps.
- A logarithmic (RSSI) detector is used for OOK/ASK signals.

### 6.5.1. Receiver Circuit Brief Overview

**Clock and Data Recovery:** The AMIS-53000 can extract a synchronous clock signal from the received data. In this mode, the data in the received signal is detected, filtered and then fed into the clock and data recovery block where additional digital filtering is performed. The waveform is sampled using a data clock in the AMIS-53000 to synchronously recover the data. Signal sensitivity is improved and the recovered data jitter is reduced by this method.

**LO Frequency:** A sub-sampling LO frequency architecture is implemented that down converts the incoming RF signal to the IF frequency of about 500kHz. The LO frequency is produced from the internal VCO frequency. The frequency design of the LO signal reduces the power consumption of the AMIS-53000 and simplifies the receiver, achieving reliable, quadrature LO signal generation.

**IF Topology:** The receiver implements a quadrature down-conversion architecture improving image rejection and creating the signals required for the complex FFT FSK signal detection. The receiver uses this quadrature down-conversion and a combination of passive and active poly-phase filtering to provide image suppression.

**Sniff Signal Acquisition:** As with earlier ASTRIC devices, the AMIS-53000 can reduce the receiver power requirements by implementing the Sniff Mode for RF signal detection. Sniff Mode is a method using the Quick Start oscillator to quickly wake the receiver, check for signal energy and return to sleep or start the receive function. The Quick Start can start the receiver crystal oscillator in as little as 10 micro-seconds. Using this fast start time, the Sniff Mode can turn on the receiver, check for signal energy and return to sleep in as little as 130 micro-seconds. More information about this Sniff Mode is in Section 6.7.2.

Table 32: Receive Command

| Register     |         |      | Function  |
|--------------|---------|------|---|
| Number (HEX) | Name    | Code |   |
| 0X00         | Command | 0X01 | Instruction to place the AMIS-53000 into receive (remember that all parameters for receive must be set before issuing this command) |

Setup registers descriptions:

**RX Config-** Options for the receiver must be set.

**RF Frequency-** The RF frequency of the receiver must be configured. (See Section 6.4.1.1)



## 6.5.1.1. RX Configuration

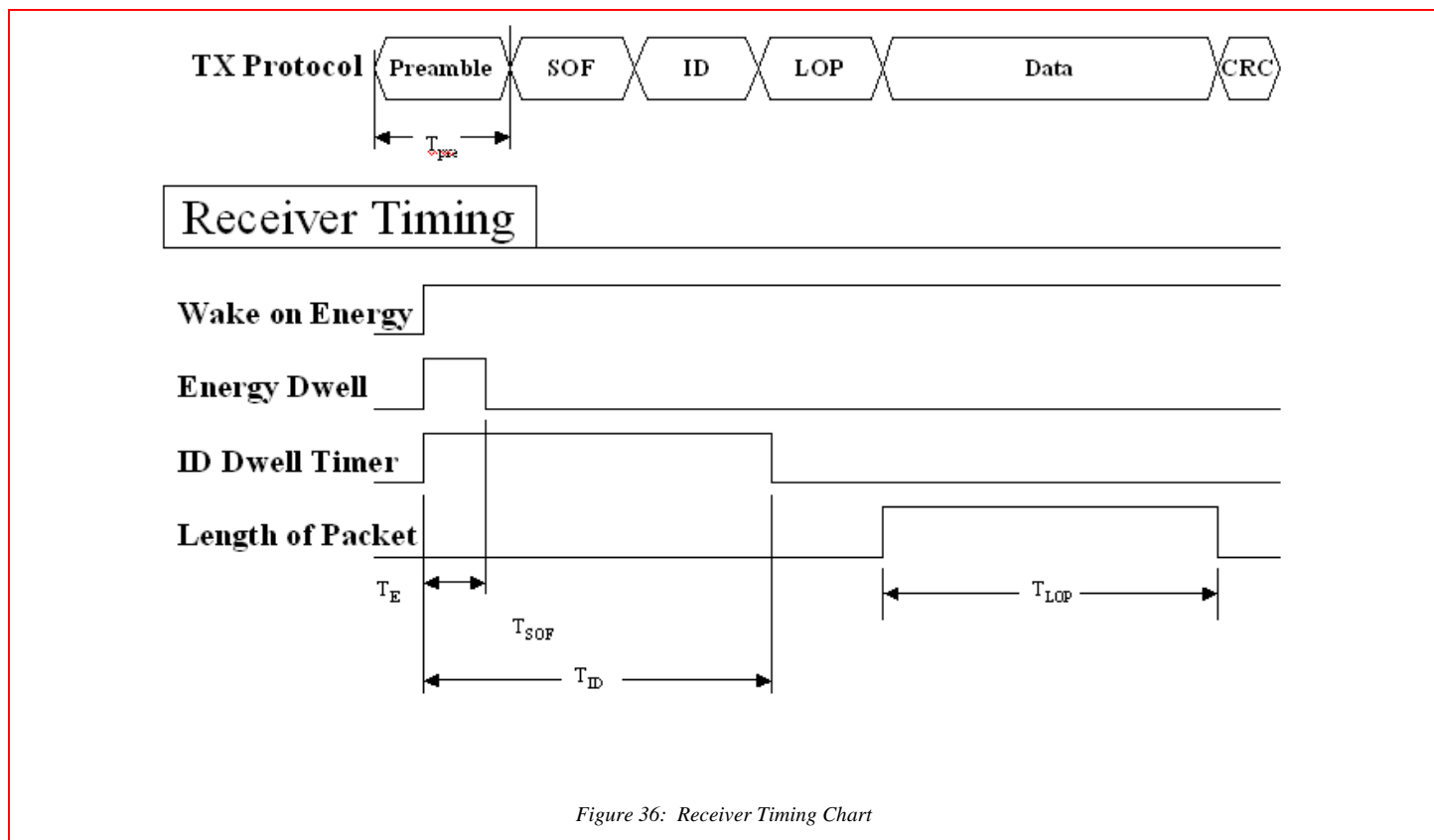


Figure 36: Receiver Timing Chart

Table 33: Receiver Timing

| Symbol    | Timing           |                  |      |        | Comments   |
|-----------|------------------|------------------|------|--------|--|
|           | Min.             | Typ.             | Max. | Units  |  |
| $T_{PRE}$ | MOD <sup>1</sup> | MOD <sup>1</sup> |      | D Bits | The TX preamble should be made long enough to allow the receiver to acquire the signal                             |
| $T_E$     | 0                |                  | 255  | D Bits | The energy dwell timer should be set long enough to allow the receiver to detect the energy <sup>2</sup>           |
| $T_{ID}$  | 0                |                  | 255  | D Bits | The ID dwell timer should be set long enough to allow the receiver to detect the chip ID or global ID <sup>2</sup> |
| $T_{LOP}$ | 1                | Default          | 255  | D Bits | The length of packet will turn the receiver off after the number of data bits is received                          |

- Notes:
- The need for a preamble and the type of preamble is determined by the data modulation selected.
  - The dwell timers need to be long enough to allow the receiver to stay active from the time it turns on due to energy and the time that the desired event occurs. However, making this number the maximum may in the case of false energy detection or signal corruption may waste system power unnecessarily.

The RX Config register is used to set options for receive mode operation.

**Wake on Energy:** When enabled, the CDR circuit is held in reset until the energy threshold is met. This option can be used to make the normal receiver function to perform similar to sniff. The energy dwell timer is used to determine how long the receiver will stay on checking for energy (With FF in the energy dwell time register, the receiver will stay on until the threshold is met).

**Gate on Energy:** This option can be used in FM receive mode only, and will gate the data interface while the energy on RSSI is below the energy threshold.

**AM\_FM\_RX:** Sets the mode of operation for receive.

**Force MICS Channel:** When this bit is set, the bits in Status2 used to show which channel the radio is on can be overwritten to force a particular channel.

Table 34: RX Config - 0X0E [14]

| Bit | Name               | State | Comment   |
|-----|--------------------|-------|---|
| 7   | RSSI Active        | 1     | RSSI output during RX                                   |
|     |                    | 0     |   |
| 6   | Multi Channel      | 1     | Sniff cycle performed at multiple pre-defined channels  |
|     |                    | 0     |   |
| 5   | LNA Mode[1:0]      | 11    | No operation defined                                    |
| 4   |                    | 10    | Linear mode   |
|     |                    | 01    | High gain mode  |
|     |                    | 00    | Normal gain mode  |
| 3   | Force MICS Channel | 1     | Force receive mode on a specific MICS channel (status2) |
|     |                    | 0     |   |
| 2   | AM_FM_RX           | 1     | AM receive mode   |
|     |                    | 0     | FM receive mode   |
| 1   | Gate on Energy     | 1     | Clock and data outputs gated for RSSI<energy threshold  |
|     |                    | 0     |   |
| 0   | Wake on Energy     | 1     | Clock and data disabled until energy threshold met      |

### 6.5.1.2. Energy Threshold

Sets the threshold for either wake on energy or Sniff Mode. If the automatic noise floor detection is enabled in Sniff, the AMIS-53000 will overwrite the contents of this register each time a new threshold is calculated.

Table 35: Energy Threshold - 0X15 [21]

| Bit | Name        | Comment  |
|-----|-------------|--|
| 7:0 | E_Threshold | Energy detect threshold in DAC Mode = E_THRESHOLD*7.8125mV |

6.5.1.3 Receiver Parameters

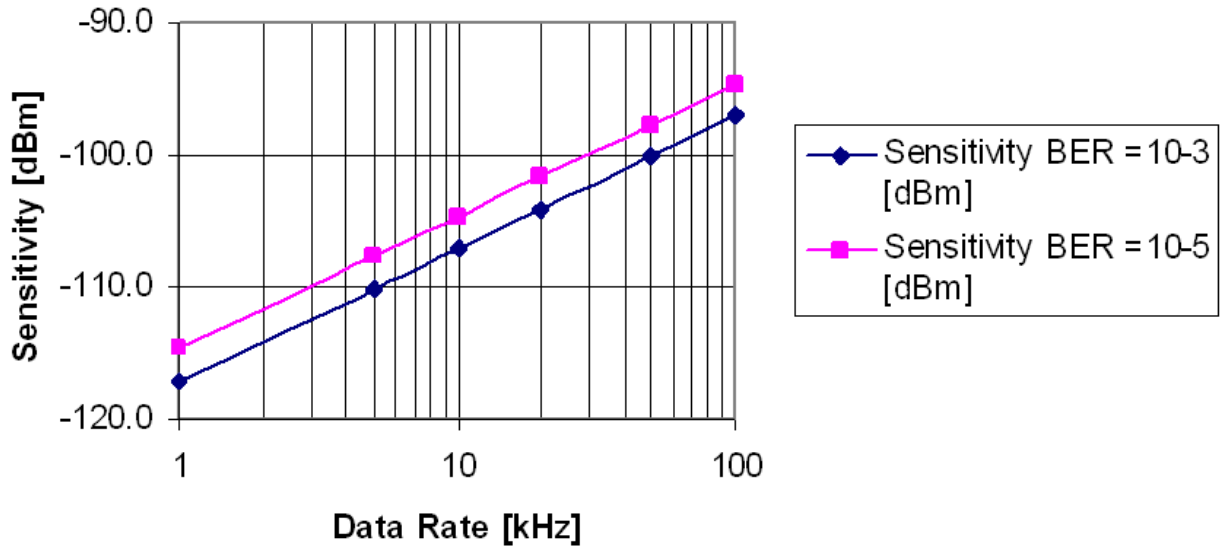


Figure 37: Receiver Sensitivity vs. Data Rate

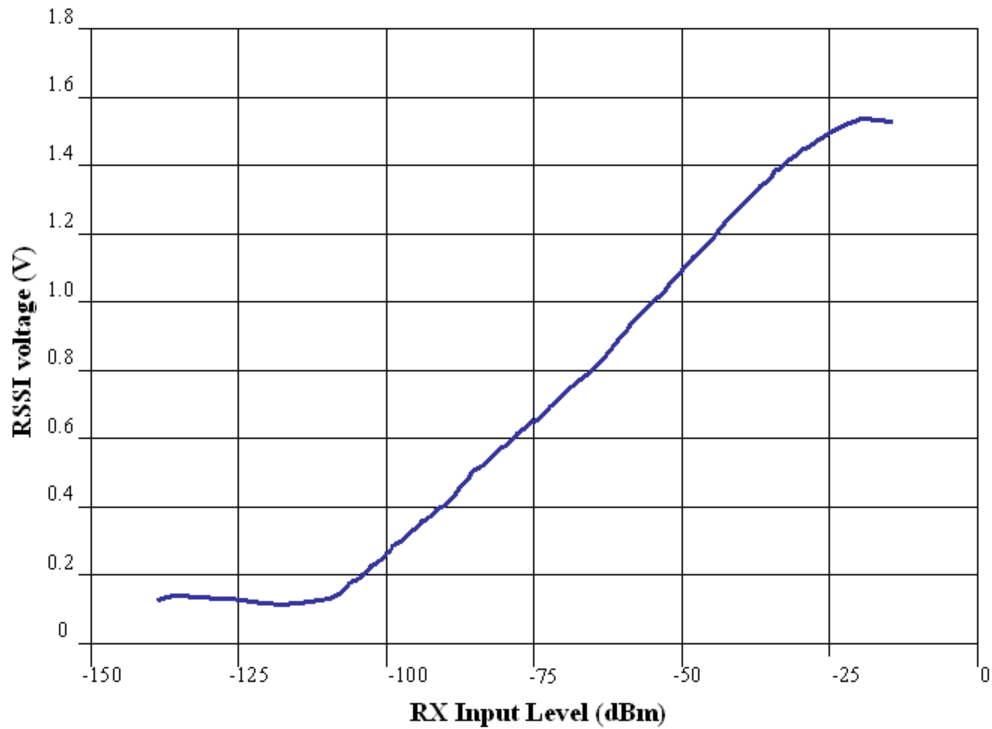


Figure 38: RSSI Curve

### 6.5.1.4. Data Recovery

Table 36: Receiver Configuration

| RX Configuration Registers |     |                    |                 |                       |                  |                    | Data Range  | Preamble     | CDR              | SOF             |
|----------------------------|-----|--------------------|-----------------|-----------------------|------------------|--------------------|-------------|--------------|------------------|-----------------|
| 0X0E RX Config             |     | 0X1F CDR Options A |                 | 0X0B Data Rate Format |                  | 0X1E Filter Slice  |             |              |                  |                 |
| Bit 2                      | Mod | Bit 0              | Demod           | Bit 3                 | Code             | Bit 1:0            |             |              |                  |                 |
| 0                          | FM  | 0                  | FFT             | 0                     | NRZ <sup>1</sup> | SLICE <sup>2</sup> | >20k – 128k | 10Pattern    | Yes <sup>3</sup> | No              |
|                            |     |                    |                 | 1                     | Man              | SLICE <sup>2</sup> | >20k – 128k | 10Pattern    | Yes <sup>3</sup> | No <sup>4</sup> |
|                            |     | 1                  | PLL             | 0                     | NRZ <sup>1</sup> | SLICE <sup>2</sup> | <20k        | 10Pattern    | Yes <sup>3</sup> | No              |
|                            |     |                    |                 | 1                     | Man              | SLICE <sup>2</sup> | <20k        | All 1 or 0's | Yes <sup>3</sup> | 0X55<br>0XAA    |
| 1                          | AM  | 0                  | NA <sup>6</sup> | 0                     | NRZ              | SLICE <sup>2</sup> | 1k – 19.2k  | CW           | Opt <sup>7</sup> | 0X55            |
|                            |     |                    |                 | 1                     | Man              | SLICE <sup>2</sup> | 1k – 19.2k  | CW           | Opt <sup>7</sup> | 0X0A            |
|                            |     | 1                  | NA <sup>6</sup> | 0                     | NRZ              | SLICE <sup>2</sup> | 1k – 19.2k  | CW           | Opt <sup>7</sup> | 0X55            |
|                            |     |                    |                 | 1                     | Man              | SLICE <sup>2</sup> | 1k – 19.2k  | CW           | Opt <sup>7</sup> | 0X0A            |

Notes:

- Long strings of 1's or 0's will degrade the performance of the CDR circuits.
- SLICE can use the parameters in Table 38.
- Yes indicates that CDR is always used to detect the data. (See Table 41 and Table 42 for CDR parameters)
- Manchester encoded data requires a SOF. A unique SOF (suggested 0x66) is used to bit-align the Manchester decoder to the recovered data.
- 0X55 or 0X0A SOF have the following requirements:
  - 0x55 The following is suggested:
    - It is suggested that CDR with fast phase alignment be enabled
    - It is suggested that CDR with activity check be enabled with 8 or 16 bit times
    - It is suggested that the preamble be long enough to trigger activity check (10 or 20 bit times)
  - 0x0A The following is suggested:
    - It is suggested that CDR with fast phase alignment be enabled
    - It is suggested that CDR with activity check be enabled with 4 bit times
    - It is suggested that the preamble be long enough to trigger activity check
- NA indicates that the parameter is not available for the AM/OOK modulated signals.
- OPT indicates that CDR is an option to detect the data however, it is recommended that CDR be used. (See Table 41 and Table 42 for CDR parameters)

### AM Data Recovery with RSSI

The logarithmic AM detector, used with OOK/ASK modulated signals, produces an RSSI output signal with approximately 18mV/dB output level. A low pass filter provides additional filtering matched to the AM signal data rate (1.2, 1.8, 2.4, 4.8, 7.2, 9.6, 14.4, 19.2kbps). This filtered signal is sampled and compared to the slice threshold to recover the digital data. The slice threshold can be set to a fixed value or it can use a signal tracking circuit to set a peak or average threshold.

The RSSI output signal can also be applied to a clock and data recovery circuit, which synchronizes an AMIS-53000 internal clock with the incoming data rate (See CDR Operation).

#### Setup registers descriptions:

**Slice Threshold-** The slice operation needs to be selected, fixed or automatic. The value for the fixed threshold needs to be set.

**Filter/Slice-** Filter settings and slice mode need to be selected.

#### Slice Threshold

Sets the data slice level for AM reception. This threshold is used when the slice method sets in the AM data filter and slice options is set to DAC.

Table 37: Slice Threshold - 0X1D [29]

| Bit | Name            | Comment   |
|-----|-----------------|---|
| 7:0 | SL_THRESH [7:0] | AM slice threshold in DAC Mode = SL_THRESH*7.8125mV |

### AM Data Filter and Slice Options

This register contains settings for determining the method of slicing for AM receive mode.

**AM\_FILTER:** Sets the post-detection filter bandwidth for RSSI during AM receive. These filter bandwidths are set for the corresponding data rates. These filter settings can also be used with custom data rates.

**AM\_HYST:** Sets the amount of hysteresis in the AM slice comparator.

**AM\_SLICE:** Used to select the method for providing a reference to the AM slice comparator. DAC Mode: This is a fixed threshold level programmed into the slice threshold register.

**Average Mode:** This is an automatic threshold level where the AMIS-53000 sets the threshold level to the average level of the RSSI signal. An external capacitor is required on the PEAK pin to set a bandwidth for the low pass filter response of the averaging circuit. Due to the potential for long strings of 1's or 0's in NRZ data this threshold method is not recommended for use with NRZ data.

**Peak Mode:** This is similar to the average mode, but only the highest level is determined from the incoming signal and the AMIS-53000 sets the threshold to a level 6dB lower than the peak value. An external capacitor on the PEAK pin determines a bleed off (discharge) rate for the peak detector circuit.

Table 38: AM Data Filter and Slice Options - 0X1E [30]

| Bit   | Name           | State | Comment  |
|-------|----------------|-------|--|
| [7:5] | AM_FILTER[2:0] | 000   | RSSI filter bandwidth = 300Hz                          |
|       |                | 001   | RSSI filter bandwidth = 600Hz                          |
|       |                | 010   | RSSI filter bandwidth = 1.2kHz                         |
|       |                | 011   | RSSI filter bandwidth = 2.4kHz                         |
|       |                | 100   | RSSI filter bandwidth = 4.8kHz                         |
|       |                | 101   | RSSI filter bandwidth = 9.6kHz                         |
|       |                | 110   | RSSI filter bandwidth = 19.2kHz                        |
|       |                | 111   | RSSI filter bandwidth = 38.4kHz                        |
| 4     | NU             |       |  |
| [3:2] | AM_SLICE[1:0]  | 00    | DAC mode: Slice threshold set in register 1Bh          |
|       |                | 01    | Average mode: AM threshold set using averaging filter  |
|       |                | 10    | Peak detect mode: AM threshold set using peak detector |
|       |                | 11    | DAC mode: Slice threshold set in register 1Bh          |
| [1:0] | AM_HYST[1:0]   | 00    | 0mV slice hysteresis                                   |
|       |                | 01    | 25mV slice hysteresis                                  |
|       |                | 10    | 50mV slice hysteresis                                  |
|       |                | 11    | 100mV slice hysteresis                                 |

### FM FFT

The AMIS-53000 receiver uses a FFT function to recover data from a FM/FSK modulated signal when the data rate is higher than 20kbps. The FFT detector uses a two-bit DFFT to demodulate the incoming IF signal. This circuit also uses the same clock recovery block (CDR) as the AM detector (See Section 6.1.4) to detect the data.

A pattern of 1's and 0's is required as a preamble. A SOF is not required unless Manchester encoded data is used, requiring a unique preamble to bit-align the Manchester decoder to the recovered data.

Setup registers descriptions:

**Data Rate-** Can be specified in the discrete data rate register, or specified as a 16-bit word for a user defined data rate. (See Section 7.1.3)

**Peak Deviation-** The peak deviation register stores the value to be used for both transmit and receive. In the FFT FM receive mode, this value is used to set-up the FFT bins. (See Sections 6.4.1.5 and 6.4.1.6)

**FM PLL Detector Loop Filter-** For the discrete data rates, the values for the loop filter coefficients are pre-programmed. For user defined data rates, this value needs to be calculated.

PLL Detector Loop Filter Setting

A program (AMIS-53CALC.exe available from AMIS) has been created to aid in the design of loop filter settings.

Table 39: PLL Detector Loop Filter Setting - 0X2B [43]

| Bit | Name         | Comment                 |
|-----|--------------|-------------------------|
| 7:0 | PLL_CO [7:0] | PLL loop filter setting |

**FM PLL (Low Data Rate FSK)**

The AMIS-53000 uses a PLL function to recover the data from a FM/FSK modulated signal with data rates 20kbps or lower. This circuit uses an A DPLL for demodulation, the output of which is fed to the AM CDR circuit to recover the clock, and additionally filter the output data.

The preamble sent by the AMIS-53000 when configured as NRZ FM is a repeating sequence of 1's and 0's. This gives the CDR circuit and PLL demodulator sufficient edges to acquire lock. Hence, for the NRZ case it is unnecessary to include a SOF byte. In Manchester mode, the preamble is specified as all 1's (or 0's). This gives the clock recovery circuit the most edges for lock acquisition. However, due to the ambiguity of the preamble, a SOF byte is necessary for the Manchester decoding block. The suggested SOF for this is either #55h or #AAh. The length of preamble necessary for this mode is dependant upon the loop bandwidth for the clock recovery PLL.

Setup registers descriptions:

**Data Rate-** Can be specified in the discrete data rate register, or specified as a 16-bit word for a user defined data rate. (See Section 7.1.3)

**Peak Deviation-** The peak deviation register stores the value to be used for both transmit and receive. In the FFT FM receive mode, this value is used to set-up the FFT bins. (See Sections 6.4.1.5 and 6.4.1.6)

**M PLL Detector Loop Filter-** For the discrete data rates, the values for the loop filter coefficients are preprogrammed. For user defined data rates, this value needs to be calculated. (See PLL Detector Loop Filter Setting)

*6.5.1.5. Clock and Data Recovery*

The AMIS-53000 device performs clock and data recovery for both AM/OOK/ASK and FM/FSK signals. An internal clock in the AMIS-53000 is programmed to be nearly the same rate as the expected data rate in the incoming signal. This clock is then synchronized to the incoming data rate by extracting a clock from the data. This loop recovery method recovers data without much of the jitter and noise associated with wireless communication links.

Before launching headlong into the operation of the detectors, and how to set them up, it is instructive to review the following related registers, setup options and their functions.

Setup registers descriptions:

**Fast Phase Alignment:** In both the AM and PLL based FM modes (lower data rate), the AMIS-53000 can be configured to quickly acquire phase lock on incoming data. The pattern necessary for the fast phase alignment is simply '1010'. This function can be enabled in the CDR options A register. With this function enabled, the CDR circuit will operate with minimum power consumption until the '1010' sequence is received. A 32-bit correlation is used to not only recognize the 1010 pattern, but also to instantaneously provide a phase correction to the clock recovery circuit allowing very fast (less than 4 bit) lock times locking the incoming data.

**Activity Check:** This function can be used in conjunction with the fast phase alignment to reset the clock and data recovery block back into its minimal power consumption mode when no transitions are detected on the data line for a specified period. The check can be configured for 4, 8 or 16 bit times.

**Over-Sampling Clock (Ts Clock):** All three detectors use the Ts clock as the sampling clock for the transition from analog to digital. This clock should be set to the highest rate possible, but not greater than 400x the data rate, to ensure adequate phase information. For the discrete data rates, this value is pre-programmed for those rates.

**Data Rate Clamp:** The data rate clamp restricts the clock recovery circuit from wandering when an actual signal is not present, and the phase error signals being generated come only from noise. Small fractional values for the clamp can lead to longer lock times since the clock recovery PLL may not be able to make as large of a correction as is necessary all at once.

**Channel Clamp:** This clamping circuit is used to hold the low data rate FSK PLL detector within the specified limits to prevent the PLL from wandering in the absence of signal.

### CDR Operation

This circuit utilizes an all digital PLL (ADPLL) to recover the clock from the raw sliced data. The slicer output is integrated over a bit time to provide a phase error, and the sign of the integration is used to determine the data symbol.

When using the AMIS-53000 in AM mode with any of the packet framing options enabled, it is necessary to have the SOF byte for proper start-up of the AM CDR circuit. It is recommended that the CDR is set up with the activity check, and fast phase alignment features enabled for the packet framing modes. The preamble that the AMIS-53000 will transmit in AM mode is CW, hence the SOF byte is used by the fast phase alignment feature in the CDR to acquire lock. The suggested SOF for AM NRZ format is #55h. This will provide the most transitions for the clock recovery circuit to acquire lock prior to the incoming packet. For Manchester operation, the suggested SOF is #0A. This will provide early transitions for phase lock, and 4 bits to align the Manchester decode.

Because there are no transitions during the preamble in AM mode, the CDR relies on the fast phase alignment for acquiring lock. As this is the case, the length of the preamble can be quite short as long as the activity check is enabled. The preamble should be long enough to trip the activity detection circuitry such that the fast phase alignment circuit is reset at the beginning of the SOF. This guarantees that the fast phase alignment will kick in during the SOF. The suggested length for the preamble is 4 BT's for Manchester with activity check set to 4 BT's, and 10 or 20 for NRZ, with activity check set to 8 or 16 respectively. Note in the NRZ case, enabling activity check will require the data be formatted to guarantee at least one transition in the data during the length of activity check (i.e. every 8 or 16 BT's).

Setup registers descriptions:

**ID Dwell-** Set a time that the CDR circuit will continue to search for the chip ID.

**CDR Config-** Set the parameters for the clock and data recovery circuits.

**Data Rate-** Can be specified in the discrete data rate register, or specified as a 16-bit word for a user defined data rate. (See Section 7.1.3)

**Start of Frame-** Byte used to tell the AMIS-53000 receiver that data will start. (See Section 7.1.6)

**Clock Recovery Loop Filter-** For the discrete data rates, the values for the loop filter coefficients are pre-programmed. For user defined data rates, this value needs to be calculated.

### Chip ID Dwell Timer

Used to specify how long the clock and data recovery circuit will stay active after energy has been detected, looking for a valid chip ID. The part will look for either the chip ID or the global ID.

Table 40: Chip ID Dwell Timer - 0X14 [20]

| Bit | Name          | Comment   |
|-----|---------------|---|
| 7:0 | C_DWELL [7:0] | 00h: Code dwell timer disabled (for standard receive wake on code)<br>01h – FFh: Code dwell time = C_DWELL*bit time * 8 |

## CDR Options A

This register contains settings for determining the clock and data recovery parameters.

- DR\_Clamp: Limits the CDR frequency drift between data packets.
- Channel\_Clamp: Restricts the bandwidth to the channel bandwidth.
- Activity Check: Sets a number of bit times that the CDR circuit will shut down if there is no data present.
- Fast Phase Alignment: Forces the CDR circuit to quickly synchronize to the incoming data.
- FM Mode: The FM detector used in the receiver depends on the data rate of the incoming signal.

Table 41: CDR OptionsA - 0X1F [31]

| Bit | Name               | State | Comment   |
|-----|--------------------|-------|---|
| 7,6 | DR Clamp<1>        | 11    | 1/64  |
|     |                    | 10    | 1/32  |
|     |                    | 01    | 1/16  |
|     |                    | 00    | 1/8   |
| 5,4 | Channel Clamp      | 11    | +150  |
|     |                    | 10    | +100  |
|     |                    | 01    | +50   |
|     |                    | 00    | +16   |
| 3,2 | Activity Check <1> | 11    | Reset after 16 bit times of no activity                 |
|     |                    | 10    | Reset after 8 bit times of no activity                  |
|     |                    | 01    | Reset after 4 bit times of no activity                  |
|     |                    | 00    | Activity check disabled, clock recovery will always run |
| 1   | FPA Enable         | 1     | The CDR circuit will perform fast phase alignment       |
|     |                    | 0     | CDR always running                                      |
| 0   | FM Mode            | 1     | PLL   |
|     |                    | 0     | FFT   |

## CDR Options B

The sample clock values are written from ROM with the discrete data rate selected. The sample rate should be as fast as possible without exceeding 400 samples per bit time.

Table 42: CDR OptionsB - 0X20 [32]

| Bit | Name         | State | Comment           |
|-----|--------------|-------|-------------------|
| 7   | CDR Reset    | 1     | CDR is held reset |
|     |              | 0     |                   |
| 6   | NU           |       |                   |
| 5   | NU           |       |                   |
| 4   | NU           |       |                   |
| 3:0 | Sample Clock | 1011  | 45kHz             |
|     |              | 1010  | 90kHz             |
|     |              | 1001  | 187.5kHz          |
|     |              | 1000  | 375kHz            |
|     |              | 0111  | 750kHz            |
|     |              | 0110  | 1.5M              |
|     |              | 0101  | 3M                |
|     |              | 0100  | 6M                |
|     |              | 0011  | 8M                |
|     |              | 0010  | 12M               |
|     |              | 0001  | 16M               |
|     |              | 0000  | 24M               |



[Clock Recovery Loop Filter Setting](#)

A program (AMIS-53CALC.exe available from AMIS) has been created to aid in the design of the CDR loop filter settings.

Table 43: Clock Recovery Loop Filter Setting - 0X2C [44]

| Bit | Name         | Comment                       |
|-----|--------------|-------------------------------|
| 7:0 | CDR_CO [7:0] | Clock recovery filter setting |

## 6.6 Transmitter

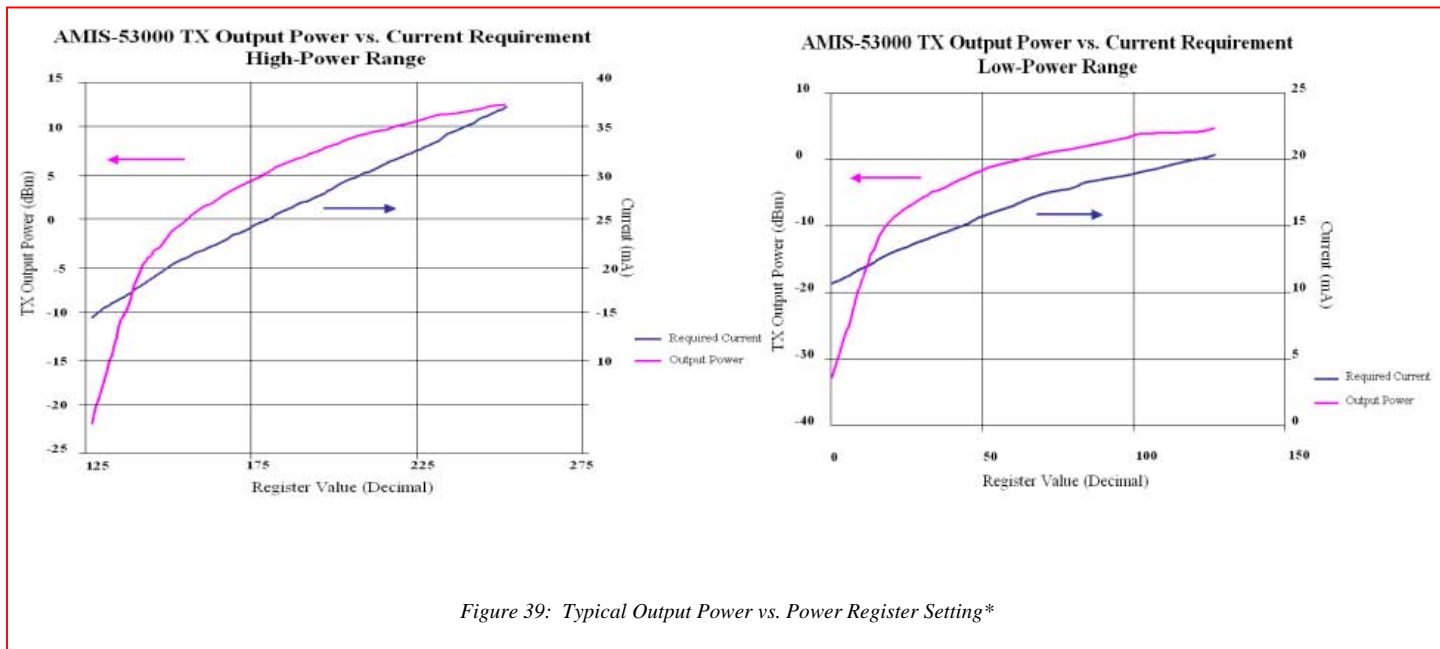


Figure 39: Typical Output Power vs. Power Register Setting\*

\*Note: Curve is for output matched to 50Ω.

Table 44: Transmit Command

| Register     |         |      | Function  |
|--------------|---------|------|---|
| Number (HEX) | Name    | Code |   |
| 0X00         | Command | 0X02 | Instruction to place the AMIS-53000 into transmit (remember that all parameters for transmit must be set before issuing this command) |

The AMIS-53000 uses a switching class E power amplifier as the high power output driver. The high power PA can be bypassed to allow a high efficiency at a lower output power. The output drivers are turned on and off directly in AM/OOK/ASK data modulation. A direct modulation PLL is used to form the FM/FSK signal for transmission. The PLL loop runs at half of the desired transmit frequency to provide excellent on/off ratio for AM, and to lower current consumption in the PLL.

- OOK modulation (AM)
- FSK modulation (FM)
- Burst mode transmit
- FM wave shaping
- High power and low power range

Setup registers descriptions:

**Data Rate-** Can be specified in the discrete data rate register, or specified as a 16-bit word for a user defined data rate. (See Section 7.1.3)

**RF Frequency-** The RF frequency of the transmitter must be configured. (See section 6.4.1)

**Output Power-** Sets the RF output level for the AMIS-53000.

**Peak Deviation-** The FM deviation must be set for FM operation. (See section 6.4.1.5 and 6.4.1.6)

**TX Config-** Options for the transmitter must be set.

**Preamble-** Set a reasonable length of preamble to insure that the receiver can detect the signal.

### 6.6.1. TX Config

General options for transmit.

**AM\_FM\_TX:** Used to set the mode for transmit.

**Shaping:** When enabled, the FSK transitions are digitally shaped in the fractional N PLL with a pre-programmed sequence. The filter for the shaping is Gaussian with a BT=1. This reduces the high frequency content of the data waveform that modulates the carrier.

**ID for TX:** This option specifies which ID the AMIS-53000 will transmit when the use ID bit in general options A is enabled. Global is used to transmit to all devices, chip ID will transmit the transmitters chip ID.

**Buffered TX:** When this option is enabled in conjunction with use ID and LOP enable, the packet for transmission is loaded into internal RAM prior to the RF being enabled.

**Force MICS Channel:** When this bit is set to 1, bits[7:4] of Status2 can be written to force the transmitter to operate on a specific MICS channel.

**Clear Channel Assessment:** When one of the CCA modes of operation are enabled, the AMIS-53000 will enable its receiver to first check for the presence of energy on the specified channel before transmitting.

Table 45: TX Config - 0X0F [15]

| Bit | Name                     | State | Comment  |
|-----|--------------------------|-------|--|
| 7,6 | Clear Channel Assessment | 11    | Not allowed  |
|     |                          | 10    | MICS multi-channel assessment (pre-defined channels)                               |
|     |                          | 01    | Single clear channel assessment prior to transmit (any frequency)                  |
|     |                          | 00    | No clear channel assessment performed, normal operation                            |
| 5   | Force MICS Channel       | 1     |  |
|     |                          | 0     |  |
| 4   | Smooth Turn On           | 1     | Enable smooth power up of PA (reduces the spurious response of the TX on power up) |
|     |                          | 0     |  |
| 3   | NU                       |       |  |
| 2   | ID for TX                | 1     | Use the defined chip ID from the chip ID register                                  |
|     |                          | 0     | Use the defined global ID from the chip ID register                                |
| 1   | Shaping                  | 1     | Gaussian FM data shaping enabled   |
|     |                          | 0     | FM data shaping disabled   |
| 0   | AM_FM_TX                 | 1     | AM transmit mode   |
|     |                          | 0     | FM transmit mode   |

## 6.6.2. Output Power

**C\_POWER:** Coarse output power control for the power amplifier. When set high, the max out is 15dBm - when set low, the max output is 0dBm.

**F\_POWER:** Fine output power control for the PA. These 7 bits control the voltage on the RFPWR pin via an internal DAC. F\_POWER is sent to the DAC as the upper 6 bits. Lower 2 bits are 0.

Table 46: Output Power - 0X18 [24]

| Bit | Name          | Comment   |
|-----|---------------|---|
| 7   | C_POWER       | Coarse output power selection sent to PA                          |
| 6:0 | F_POWER [6:0] | Fine output power control (upper 7 bits of the DAC, LSB of DAC=0) |

## 6.6.3. Preamble Length

This byte is used to define the length of preamble to send prior to data in both transmit and Burst mode transmit.

Table 47: Preamble Length - 0X1A [26]

| Bit | Name              | Comment   |
|-----|-------------------|---|
| 7   | PreambleLen [7:0] | Length, in bit times, of CW (AM), or '10 (FM) sent prior to preamble in Burst |

Table 48: Suggested Preambles

| Modulation             | Preamble     | Comment   |
|------------------------|--------------|---|
| AM NRZ with CDR        | CW           | SOF required and suggested as 0X55 <sup>1</sup>                     |
| AM Manchester with CDR | CW           | SOF required and suggested as 0x0A <sup>2</sup>                     |
| FM FFT                 | 1/0 pattern  | No SOF is required  |
| FM PLL (NRZ)           | 1/0 pattern  | No SOF is required  |
| FM PLL (Manchester)    | All 1s or 0s | SOF is required and suggested to be 55(Hex) or AA(Hex) <sup>3</sup> |

- Notes:
1. When using SOF with NRZ data, it is suggested that fast phase alignment is enabled, activity check is set to 8 or 16 and preamble length is 10 or 20 bit times.
  2. When using SOF with Manchester coded data, it is suggested that fast phase alignment is enabled, activity check is set to 4 bit times and preamble length is set long enough to trigger the activity check.
  3. The length of this preamble is dependent on the loop bandwidth of the recovery clock PLL.

## 6.6.4. FM Transmit Data Shaping

The AMIS-53000 allows the user to enable data shaping of the data waveform to improve the RF spectral efficiency. When enabled, the clock recovery NCO is used to provide an internal clock at 16 times the selected data rate. This clock is used to cycle through a pre-defined pattern whenever a transition is detected on the TX input. The shaping pattern is Gaussian with a BT=1. The intermediate values for the shaping are determined from the peak deviation register when an external host/controller writes the 'ROM 2 REGS' to the AMIS-53000 command register.

## 6.7 Idle

Table 49: Idle Command

| Register     |         |      | Function              |
|--------------|---------|------|-----------------------|
| Number (HEX) | Name    | Code |                       |
| 0X00         | Command | 0X03 | Enable the idle state |
|              |         | 0X04 | Return to idle state  |

Table 50: Idle Configuration

| Register     |                   |      | Function                                   |
|--------------|-------------------|------|--|
| Number (HEX) | Name              | Bits |  |
| 0X02         | Status/Flag2      | 0    | AMIS-53000 is busy                         |
| 0X0C         | General Options A | 2:1  | Select state that AMIS-53000 enters on POR |
|              |                   | 0    | Output system clock in standby mode        |
| 0X0D         | General Options B | 1, 0 | Select the system clock frequency          |

Note that there are two low-power modes for AMIS-53000 – standby and idle. Standby allows the SYSCLK output. Idle is the very low power state without SYSCLK output.

Table 51: Idle Modes

| Idle Tasks     | Description  |
|----------------|--|
| Sniff          | Receiver periodic wake up and RF energy detection check  |
| Burst Transmit | Periodic wake and transmit function  |
| Housekeeping   | Used to perform periodic temperature correction, calibration or to wake external host/controller |
| Standby        | Low-power mode with no activity may be programmed to continue to output system clock             |

The AMIS-53000 allows for a low-power mode. Power requirements are reduced when the very low power 10kHz oscillator is the clock for the device. This oscillator runs the timers for either the Sniff wake up timer, the Burst transmit timer or the housekeeping wakeup timer. However, even when the low power 10kHz oscillator is running, the AMIS-53000 can still provide a clock signal output to an external host/controller device such as a microprocessor after reception of a valid data packet or a housekeeping cycle where the AMIS-53000 has been programmed to issue a wake up to the external host/controller.

The AMIS-53000 will return to this low power idle state after activities such as transmit, receive or the various timers are done and the external host/controller writes the 'Return to Idle' instruction to the AMIS-53000 command register.

Setup registers descriptions:

**Idle Config-** The options for idle mode must be set.

### 6.7.1. Idle Config

The idle configuration register is used to specify which periodic tasks are performed once the idle command is given in register 0. Any combination of Sniff, Burst and housekeeping can be enabled.

**INT to DSSN Timing:** This is the time between the AMIS-53000 issuing an interrupt to indicate that data is ready and the time that data transfer starts with the assertion of the DSSN signal.

**Idle to System Clock Stop:** Once the AMIS-53000 has completed a task and is issued the idle command, the system clock can produce additional clock cycles to allow the external host/controller to finish its tasks.

Table 52: Idle Config - 0X10 [16]

| Bit | Name   | State      | Comment   |
|-----|--|------------|---|
| 7:5 | Wait timing between INT and DSSN for Sniff Receive                 | 111        | DSSN going active is delayed 8 bit times from INT                         |
|     |  | ...<br>000 | DSSN is immediately active after INT                                      |
| 4,3 | Wait timing between idle instruction and system clock shutoff time | 11         | System clock outputs 1024 clock cycles before shutdown after idle command |
|     |  | 10         | System clock outputs 512 clock cycles before shutdown after idle command  |
|     |  | 01         | System clock outputs 256 clock cycles before shutdown after idle command  |
|     |  | 00         | System clock immediately shuts down after idle command                    |
| 2   | Housekeeping Enable  | 1          | Periodic housekeeping enabled (HK settings must be set)                   |
|     |  | 0          |   |
| 1   | Burst Enable   | 1          | Periodic Burst mode enabled (BURST settings must be set)                  |
|     |  | 0          |   |
| 0   | Sniff Enable   | 1          | Periodic Sniff Mode enabled (Sniff settings must be set)                  |
|     |  | 0          |   |

### 6.7.2. Sniff Mode Operation

Table 53: Sniff Mode Configuration

| Register     |             |      | Function   |
|--------------|-------------|------|--|
| Number (HEX) | Name        | Bits |  |
| 0X0E         | RX Config   | 1    | Force the receiver to not output Clk&Data < energy level |
|              |             | 0    | Set the receiver to wake when at energy level threshold  |
| 0X10         | Idle Config | 2    | Enable periodic Sniff Mode                               |

The quick start technology enables the AMIS-53000 to operate its receiver in a mode called Sniff Mode. As implemented in the AMIS-53000, this Sniff Mode can wake the receiver and acquire the transmitted message in as little as 130 microseconds.

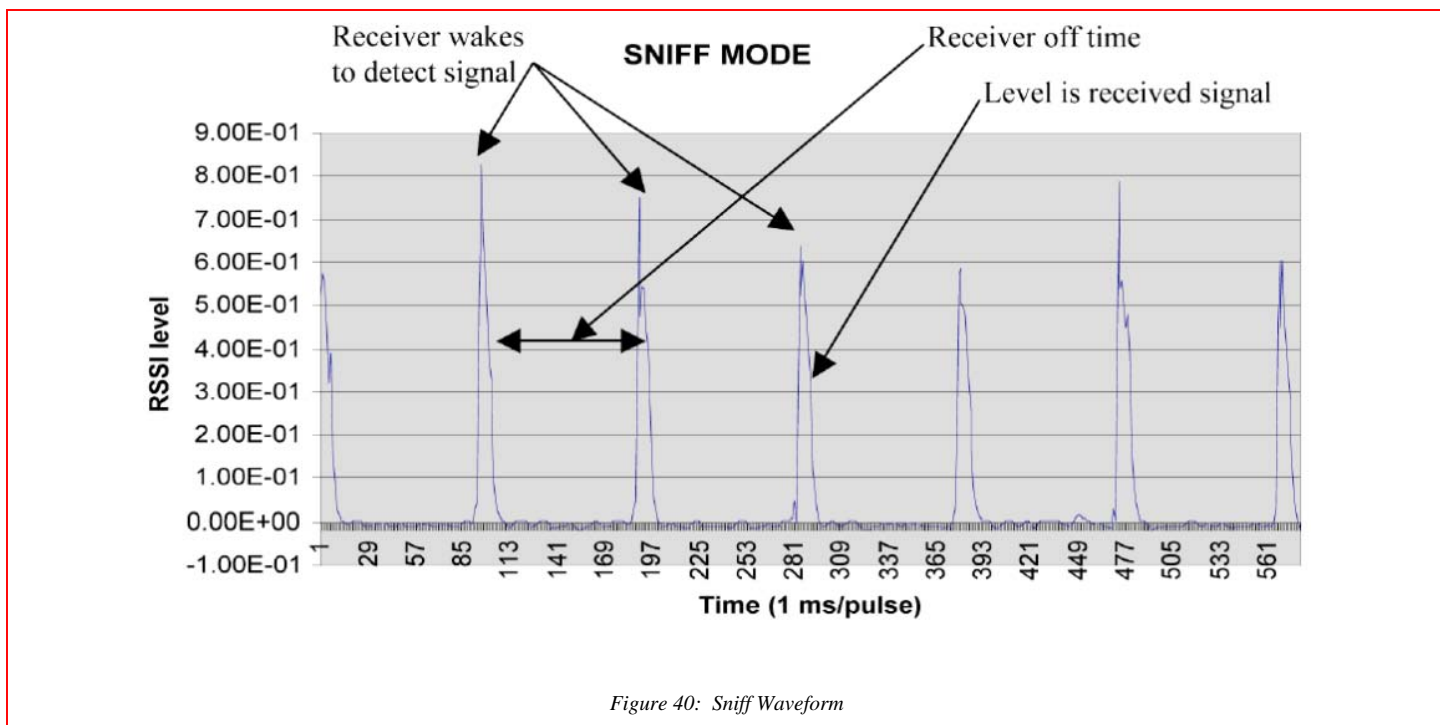


Figure 40: Sniff Waveform

The Sniff Mode of operation puts the AMIS-53000 receiver into a cyclic mode of sleeping and periodically waking to check for received signal energy. When energy is detected the receiver is placed in receive mode and the AMIS-53000 attempts to recover data. The receiver in Sniff Mode can be configured to check for a valid ID. The failure to receive a valid ID will cause the receiver to go back to sleep.

The AMIS-53000 receiver average supply current can be estimated as:

$$I_{AC} = I_{IN} \frac{\tau}{T + \tau} + I_{SM} \frac{T}{T + \tau}$$

Where,  $I_{IN}$  is receiver current consumption in continuous receive mode, equal to or less than 12mA;

$I_{SM}$  is receiver current consumption in sleep mode, equal to or less than 2uA;

$\tau$  is programmable receiver energy scan impulse on time (sniff time), approximately equal to 130µs minimum;

$T$  is programmable receiver off time period length between receiver energy scan impulses (sniff mode interval).

Setup registers descriptions:

**Sniff Config**-This sets the options in the Sniff Mode.

**Sniff Interval**- Set the time interval between receiver wakeups in the Sniff Mode.

**Energy Threshold**- The threshold for detecting the incoming RF energy must be set (see Section 6.5.1.2).

**Energy Dwell Time**- Set the time interval that the receiver will remain active looking for RF energy detection.

**Code Dwell Time**- Once RF energy has been detected the receiver can determine if the message has the unique ID for that receiver. The time interval for looking for this ID must be set. (See Chip ID Dwell Timer)

**Threshold**- The number of times a wake up is received can be monitored for false wake ups and the energy threshold adjusted to account for the noise level.

### 6.7.2.1. Sniff Config

The Sniff config register is used to set additional options for the operation of the part during a Sniff cycle, beyond those set in the RX config register.

**SNIFF\_FILTER**: Setting for the energy detection filter. This allows for different settings of the AM data filter between Sniff and receive. The energy dwell timer needs to be extended long enough to allow for this filter to settle during the Sniff cycle.

**Auto-Threshold Count Value**: Number of Sniff cycles used to determine whether to raise or lower the energy threshold for sniff. Any value other than 0x00 in this register will enable the auto-threshold function.

**Multi Channel**: This bit is used in MICS mode of operation to enable the AMIS-53000 to scan the nine pre-defined MICS channels programmed into the AMIS-53000.

**Sniff Interval Resolution**: Determines the clock for the Sniff interval timer.

Table 54: Sniff Config - 0X11 [17]

| Bit   | Name                       | State | Comment  |
|---|----------------------------|-------|--|
| 7:5   | SNIFF_FILTER               | 000   | RSSI filter bandwidth = 300Hz  |
|   |                            | 001   | RSSI filter bandwidth = 600Hz  |
|   |                            | 010   | RSSI filter bandwidth = 1.2kHz   |
|   |                            | 011   | RSSI filter bandwidth = 2.4kHz   |
|   |                            | 100   | RSSI filter bandwidth = 4.8kHz   |
|   |                            | 101   | RSSI filter bandwidth = 9.6kHz   |
|   |                            | 110   | RSSI filter bandwidth = 19.2kHz  |
|   |                            | 111   | RSSI filter bandwidth = 38.4kHz  |
| 4,3   | Auto-Threshold Count Value | 11    | 100  |
|   |                            | 10    | 500  |
|   |                            | 01    | 100  |
|   |                            | 00    | Disable  |
| This is the number of Sniff cycles to count false wake ups due to the noise level, so that the threshold level can be adjusted. It is adjusted higher when the number of false wake ups exceeds the wake up target number. It is adjusted down when there are fewer false wake ups than the target number. Entering a number other than 00 in this register will enable the auto-threshold. |                            |       |  |
| 2   | NU                         |       |  |
| 1   | NU                         |       |  |
| 0   | Sniff Interval Resolution  | 1     | Sniff interval timer resolution is 100ms (interval between Sniff signal detection events is (Sniff interval) times (Sniff resolution))       |
|   |                            | 0     | Sniff interval timer resolution is 500us 100ms (interval between Sniff signal detection events is (Sniff interval) times (Sniff resolution)) |

### 6.7.2.2. Sniff Interval Timer

Used to specify the period (time between Sniff events) of the Sniff operation. The Sniff interval is this value times the Sniff interval resolution value (set in the Sniff config register).

Table 55: Sniff Interval Timer - 0X12 [18]

| Bit | Name            | Comment  |
|-----|-----------------|--|
| 7:0 | SNIFF_INT [7:0] | Sniff interval timer = SNIFF_INT * Sniff interval timer resolution |

### 6.7.2.3. Energy Dwell Timer

Length of time receiver will stay on in a Sniff cycle checking for the presence of a signal. Also used for a receive command when the wake on energy bit is asserted in RX config. When used for a MICS market device, this may need to be set to 10 milli-seconds for CCA to be compatible with the MICS standard.

Table 56: Energy Dwell Timer - 0X13 [19]

| Bit | Name         | Comment  |
|-----|--------------|--|
| 7:0 | E_DWELL[7:0] | 00h: Energy dwell timer not used, energy determined by an impulse sample<br>01h – FEh: Energy dwell time = E_DWELL * 100us<br>FFh: Receiver remains on until energy threshold is met |

### 6.7.2.4. Energy Threshold

Sets the threshold for either wake on energy or Sniff Mode signal acquisition. If the automatic noise floor detection is enabled in Sniff, the AMIS-53000 will overwrite the contents of this register each time a new threshold is calculated.

### Automatic Threshold Optimization in Sniff Mode

An option available in the Sniff Mode config register will enable the AMIS-53000 to automatically adjust the energy threshold of the AMIS-53000 to optimize the sensitivity of the Sniff Mode. The target number of wake ups (register 0X2F) allows the user to specify a value for the number of false wake ups the AMIS-53000 is allowed with the selected configuration of the Sniff Mode (by bits [4:3] in the SNIFF\_CONFIG register) number of Sniff cycles.

As an example:

If the number of Sniff cycles is set to 500 (Sniff config [Bit 4:3]),

If the target number of false wake ups is set to 50 (target number wake ups),

Then over the course of the next 500 Sniff cycles the radio is triggered falsely by energy:

More than 50 times, the threshold will be increased to reduce the sensitivity,

Less than 50 times, the threshold will be decreased to increase the sensitivity.

Using this option to set the threshold for energy detection can have a dramatic impact on the life of battery powered devices, as the AMIS-53000 will adjust to changing levels of background noise while still maintaining maximum sensitivity, and not wasting power by continually waking and processing background noise.

Additionally, the energy threshold setting can be monitored by an external host/controller. The amount the threshold will increase or decrease is based on Table 57, with the order of the rows in the same order as the AMIS-53000 will evaluate the conditions.

Table 57: Auto Threshold Adjust

| Condition   | Adjustment   |
|---|--|
| Select the number of Sniff cycles as the test period  |  |
| Enter the desired number of false wake ups per period | Note: a false wake up is when the receiver detects energy but fails to detect the ID in the packet |
| Then if wake ups/period > target                      | Threshold is increased   |
| Then if wake ups/period < target                      | Threshold is decreased   |

The threshold of the AMIS-53000 can vary over a wide range from one device to the next, due to design, manufacturing tolerances, and environment changes; temperature and voltage. This automatic threshold optimization can be used to adjust the threshold by monitoring the level of false wake ups due to background noise. The adjustment in this fashion can reduce the effects of design and manufacturing on the threshold setting of the AMIS-53000.

Setup registers descriptions:

**Target-** Set a value for the allowed number of false wake ups desired.

### Target Threshold

The number of false wake ups occurring during a period of time is used to automatically adjust the energy threshold in the Sniff operation. This allows the AMIS-53000 to automatically adjust its input level to compensate for, manufacturing, components, environment, temperature, and/or voltage.

Table 58: Target Wake Ups - 0X2F [47]

| Bit | Name         | Comment   |
|-----|--------------|---|
| 7:0 | Target [7:0] | The number of wake-ups that the Sniff circuit will try to adjust the threshold to not have more false wake ups or less missed signal detections. This register allows the number to be 0 to 255, but this number should always be less than the number of Sniff count (Sniff config bit 4:3). |



### 6.7.3. Burst Transmit Data

Table 59: Burst Transmission Configuration

| Register     |             |      | Function                   |
|--------------|-------------|------|----------------------------|
| Number (HEX) | Name        | Bits |                            |
| 0X10         | Idle Config | 1    | Enable Burst transmissions |

Burst mode of transmission is a function that can cause the AMIS-53000 to transmit a message at a programmed time interval or by asserting the xBURST pin to active (the AMIS-53000 must be in the xBURST mode), causing the AMIS-53000 to immediately transmit a message. The Burst mode can also be started by enabling the Burst mode in the idle register and then writing to the command register to set the AMIS-53000 into idle mode. The xBURST modes sets a timed automatic transmission of register values (a message) or ADC conversion values.

Setup registers descriptions:

**Burst Config-** Set the Burst transmission parameters.

**Burst Interval-** Set the time interval between Burst transmissions.

**User DataA-** Message for timer initiated transmissions.

**User DataB-** Message for xBURST initiated transmissions.

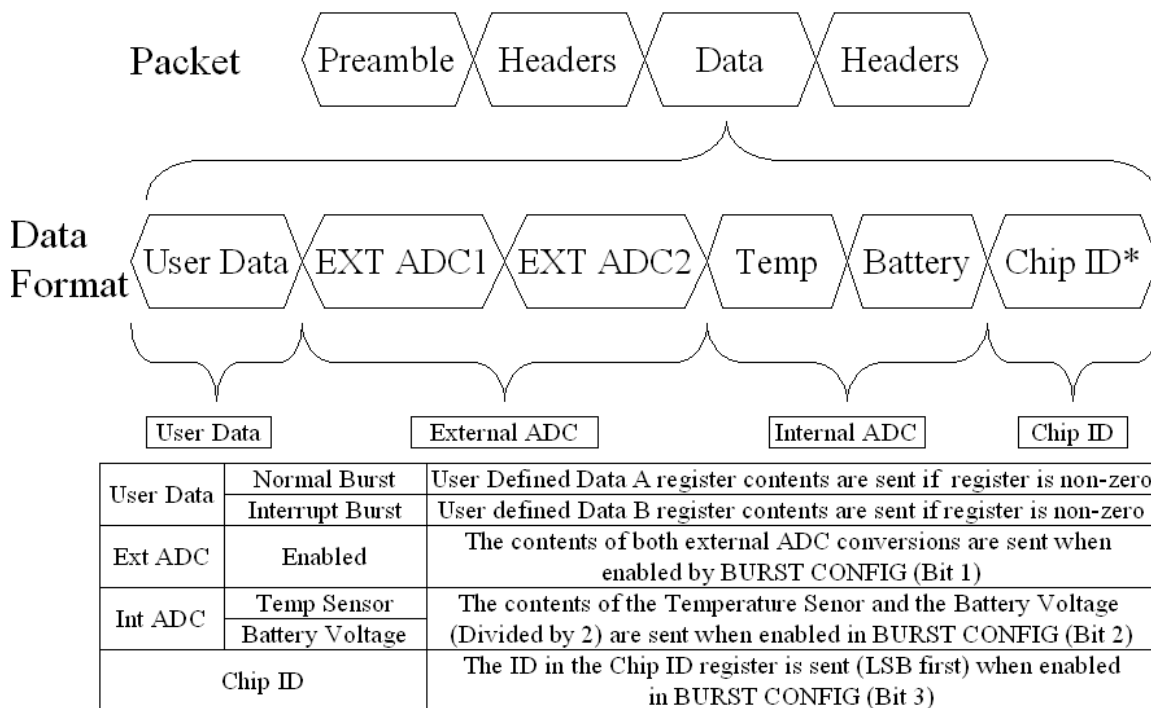


Figure 41: Data Packet Format Showing Order for Burst TX Content

Used to set the options for the Burst mode of operation:

**R\_BURST:** These bits set the number of times the packet is to be repeated each Burst interval. This can be used to increase the probability that all packets will get through when several Burst transmitters are located in the same area.

**Send Chip ID:** This option is included to allow for the operating case of having multiple transmitters sending to a single receiver. The multiple transmitters will need to be configured to send the global chip ID in the TX config options so that the receiver will wake on each transmitters packet. Sending the chip ID as part of the payload allows the receiver to differentiate the packets.

**Send Internal ADC Data:** When enabled, the AMIS-53000 will perform conversions on the battery voltage, and temperature sensor and include these conversions as part of the packet payload.

**Send External ADC Data:** When enabled, the AMIS-53000 will perform conversions on the two external ADC inputs and include these conversions as part of the packet payload.

**Burst Interval Resolution:** Used to define the clock frequency for the Burst interval timer.

Table 60: Burst Config - 0X16 [22]

| Bit | Name                      | State | Comment                                   |
|-----|---------------------------|-------|---|
| 7   |                           | 1     |   |
|     |                           | 0     |   |
| 6   |                           | 1     |   |
|     |                           | 0     |   |
| 5:4 | R_BURST[1:0]              | 00    | Packet is sent one time                   |
|     |                           | 01    | Packet is repeated once                   |
|     |                           | 10    | Packet is repeated two times              |
|     |                           | 11    | Packet is repeated three times            |
|     |                           |       | Repeat interval*                          |
| 3   | Send Chip ID              | 1     | Chip ID is included as part of the packet |
|     |                           | 0     |   |
| 2   | Send Internal ADC Data    | 1     | Data for temperature and battery is sent  |
|     |                           | 0     |   |
| 1   | Send External ADC Data    | 1     | Data for EXT1, EXT2 is sent               |
|     |                           | 0     |   |
| 0   | Burst Interval Resolution | 1     | Burst interval timer resolution is 15s    |
|     |                           | 0     | Burst interval timer resolution is 50ms   |

NOTE: When the Burst transmission is repeated the interval between transmissions is a random time period produced in a random number generator with the chip ID value used to seed the random number generator.

### 6.7.3.1. Burst Interval

Defines the period for the normal Burst transmission to occur. This is a cyclic mode and the AMIS-53000 will transmit the contents of a register at the end of each interval. This interval is fixed by the register value unlike the random time interval when the transmission is repeated.

Table 61: Burst Interval - 0X17 [23]

| Bit | Name            | Comment   |
|-----|-----------------|---|
| 7:0 | BURST_INT [7:0] | Burst interval= (BURST_INT+1)*Burst interval resolution |

### 6.7.3.2. User Defined Data A

Table 62: User Defined Data A - 0X2D [45]

| Bit | Name            | Comment   |
|-----|-----------------|---|
| 7:0 | USE_DATAA [7:0] | Optional data to be sent in normal Burst mode if non-zero |

### 6.7.3.3. User Defined Data B

Table 63: User Defined Data B = 0X2E [46]

| Bit | Name            | Comment   |
|-----|-----------------|---|
| 7:0 | USE_DATAB [7:0] | Optional data to be sent in interrupt triggered burst if non-zero |

### 6.7.4. Housekeeping

Table 64: Housekeeping Enable

| Register     |             |      | Function                            |
|--------------|-------------|------|-------------------------------------|
| Number (HEX) | Name        | Bits |                                     |
| 0X10         | Idle Config | 0    | Enable housekeeping timed functions |

Housekeeping is another periodic operation mode, which can be used to periodically perform operations such as oscillator calibrations, PLL calibration, EE refresh, or temperature compensation. It can also be used to periodically wake an external host/controller to allow it to perform whatever tasks it may need to. The housekeeping configuration register contains the options to specify what is to occur during a housekeeping cycle, and the housekeeping interval timer is used to control how frequently the wake up occurs.

Setup registers descriptions:

**Housekeeping Config-** Set the Burst transmission parameters.

**Housekeeping Interval-** Set the time interval between Burst transmissions.

#### 6.7.4.1. Housekeeping Config

The housekeeping configuration register is used to specify which tasks the AMIS-53000 should perform during a housekeeping cycle.

**HK Interval Resolution:** Used to specify the clock frequency for the housekeeping interval timer.

**Wake:** When enabled, the AMIS-53000 will enable the system clock output, and issue an interrupt to an external controller.

**Write EE:** This option is used to store the values of any calibrations that may have been performed during housekeeping. The entire working register bank will be written to EE.

**Read INT ADC Channels:** The AMIS-53000 will do conversions on the battery voltage and temperature sensor. This can be used as a method to periodically update the temperature compensation loop.

**Read EXT ADC Channels:** When enabled, the two external ADC inputs will be converted during a housekeeping cycle.

**Cal PLL, RC, Kicker:** Allows periodic calibration of the oscillators of the AMIS-53000.

Table 65: Housekeeping Config - 0X1B [27]

| Bit | Name                   | State | Comment   |
|-----|------------------------|-------|---|
| 7   | HK Interval Resolution | 1     | Five minute resolution setting  |
|     |                        | 0     | One second resolution setting   |
| 6   | Wake                   | 1     | Issue interrupt and enable clock to external host/controller              |
|     |                        | 0     |   |
| 5   | Write EE               | 1     | Current register data written to EE (performed after all Cal's complete)* |
|     |                        | 0     |   |
| 4   | Read INT ADC Channels  | 1     | Temp sensor and battery inputs converted                                  |
|     |                        | 0     |   |
| 3   | Read EXT ADC Channels  | 1     | EXT1 and EXT2 ADC inputs converted  |
|     |                        | 0     |   |
| 2   | Cal PLL                | 1     | PLL calibration performed during housekeeping                             |
|     |                        | 0     |   |
| 1   | Cal RC                 | 1     | RC oscillator calibration performed during housekeeping                   |
|     |                        | 0     |   |
| 0   | Cal Kicker             | 1     | Kicker calibration performed during housekeeping                          |
|     |                        | 0     |   |

Note: \* Bit 5 is set high to allow the EE to be automatically written after a calibration is complete.

### 6.7.4.2. Housekeeping Interval

Sets the interval (time that the AMIS-53000 is in sleep mode) for the housekeeping routine.

Table 66: Housekeeping Interval - 0X1C [28]

| Bit | Name         | Comment  |
|-----|--------------|--|
| 7:0 | HK_INT [7:0] | Housekeeping interval = (HK_INT+1)* (HK interval resolution) |

## 6.8 Idle Return

Table 67: Idle Return

| Register     |         |      | Function  |
|--------------|---------|------|---|
| Number (HEX) | Name    | Code |   |
| 0X00         | Command | 0X04 | In most conditions, the AMIS-53000 must be returned to idle mode at the end of a task by this command |

This command is used to put the part back into idle mode. It should be used by the host to place the AMIS-53000 back into idle mode after the AMIS-53000 has interrupted the host for reception of a packet in Sniff, or to end a housekeeping cycle.

## 6.9 EE

The AMIS-53000 uses internal EE memory to store register settings (either default factory settings or user defined settings).

Table 68: Calibration Results

| Register     |                     |      | Function  |
|--------------|---------------------|------|---|
| Number (HEX) | Name                | Bits |   |
| 0X00         | Command             | 0X05 | Write the register contents to EE                           |
|              |                     | 0X06 | Read the contents of the EE                                 |
| 0X01         | Status/Flag1        | 2    | EE checksum status  |
| 0X1B         | Housekeeping Config | 5    | Write register data to EE (auto after calibration complete) |

### 6.9.1. Write EE

The serial interface provides a means to read and write the working registers of the AMIS-53000. To retain the information held by these registers, on-board EE is provided to store all of the register contents needed for operation. The write EE command copies the current contents of the working registers into EE, along with a CheckSum. The CheckSum is used to verify that the content of the EE is valid when the EE is dumped back into the registers.

### 6.9.2. Load EE

The load EE command will refresh the contents of the working registers with the values stored in EE, if the EE CheckSum is valid. If the EE CheckSum fails an error bit will be set in the Status2 register.

## 6.10 Calibrate

Table 69: Calibration Results

| Register     |                      |      | Function                                     |
|--------------|----------------------|------|--|
| Number (HEX) | Name                 | Bits |  |
| 0X00         | Command              | 0X07 | Perform a Quick Start oscillator calibration |
|              |                      | 0X08 | Perform a 10kHz RC oscillator calibration    |
|              |                      | 0X09 | Perform a PLL calibration                    |
|              |                      | 0X0A | Perform a Quick Start oscillator calibration |
| 0X01         | Status/Flag1         | 6    | TX PLL calibration status                    |
|              |                      | 5    | RX PLL calibration status                    |
|              |                      | 4    | 10kHz RC oscillator calibration status       |
|              |                      | 3    | Quick Start oscillator calibration status    |
| 0X1B         | Housekeeping Config  | 2    | Perform the PLL calibration                  |
|              |                      | 1    | Perform the 10kHz RC oscillator calibration  |
|              |                      | 0    | Perform the kicker calibration               |
| 0X33         | Kicker Slope Options | 4    | Kicker calibration status                    |

Setup registers descriptions:

**Trim-** Shows the trim value for the circuit.

**PLL Trim Target-** Set a value that the PLL trim tries to achieve in calibration.

## 6.10.1. Internal Trim

### 6.10.1.1. Crystal Trim

Table 70: Crystal Trim - 0X21 [33]

| Bit | Name            | Comment  |
|-----|-----------------|--|
| 7:0 | XTAL_TRIM [7:0] | 24MHz internal trim caps; FF is max capacitance, 00 is min |

### 6.10.1.2. LNA Trim

Table 71: LNA Trim - 0X22 [34]

| Bit | Name         | Comment  |
|-----|--------------|--|
| 7:4 | LNA_OUT[3:0] | LNA output tank cap trim F is max cap              |
| 3:0 | LNA_IN [3:0] | LNA input shunt capacitor trim, F is max, 0 is min |

### 6.10.1.3. Quick Start Oscillator Trim

This register contains the value of the trim from the self calibration.

Table 72: Quick Start Oscillator Trim - 0X23 [35]

| Bit | Name          | Comment   |
|-----|---------------|---|
| 7:0 | QS_TRIM [7:0] | Trim for the Quick Start (kicker), this register is written to by the calibration circuit |

### 6.10.1.4. 10K Oscillator Trim

This register contains the value of the trim from the self calibration.

Table 73: 10kHz Oscillator Trim - 0X24 [36]

| Bit | Name          | Comment  |
|-----|---------------|--|
| 7:0 | RC_TRIM [7:0] | Trim for the 10kHz oscillator, this register is written by the calibration circuit |

### 6.10.1.5. Analog Trim1

This is an internal use register with no user defined meaning. This register is set at the factory and changing the value will cause the AMIS-53000 to not operate.

### 6.10.1.6. Analog Trim2

This is an internal use register with no user defined meaning. This register value is set at the factory and changing the value will cause the AMIS-53000 to not operate.

### 6.10.1.7. RF PLL Trim

This register contains the value of the trim from the self calibration. This register is valuable to monitor to determine if the PLL trim calibration passed, but with the highest trim value or lowest trim value which indicates that the trim is nearly all the way to the edge of the calibration.

Table 74: RF PLL Trim - 0X27 [39]

| Bit | Name                   | State            | Comment  |
|-----|------------------------|------------------|--|
| 7   |                        |                  |  |
| 6:4 | TX Mode PLL Trim Value | 111<br>--<br>000 | Max VCO trim value from the self calibration<br>Min VCO trim value from the self calibration |
| 3   |                        |                  |  |
| 2:0 | RX Mode PLL Trim Value | 111<br>--<br>000 | Max VCO trim value from the self calibration<br>Min VCO trim value from the self calibration |

### 6.10.1.8. PLL Target Value

This register is used by the AMIS-53000 during self calibration. This is an internal use register with no user defined meaning.

### 6.10.2. Calibrate Quick Start Oscillator

The Quick Start oscillator must be calibrated prior to operations such as Sniff or Burst transmit. This command will perform an internal calibration of the oscillator, write the result to the Quick Start trim register, issue a calibration complete flag, as well as a calibration good/bad indicator. This command can be issued from any valid state that accepts changes in the instructions.

### 6.10.3. Calibrate 10kHz Oscillator

In any of the idle modes of operation, an internal 10kHz oscillator is used as the timekeeping reference for the interval timers. The calibrate 10kHz oscillator command will enable the crystal oscillator to create an accurate time base to use for the calibration of this oscillator, and then perform the calibration and store the result. A Cal done and status flag will be issued upon completion of the calibration. This command can be issued from any valid state that accepts changes in the instructions.

### 6.10.4. Calibrate PLL

Calibrate the PLL performs a calibration for the PLL in both transmit and receive mode. The PLL status register reports the calibration value for both modes, as well as the status for the calibration. This command can be issued from any valid state that accepts changes in the instructions.

### 6.10.5. Calibrate LNA

This command turns on the RF receiver chain and optimizes both the LNA output tuning structure and the LNA input matching trim for maximum signal level on RSSI. This is a calibration typically performed at board assembly in the presence of a known RF signal. The AMIS-53000 will auto-tune both the input and output internal variable capacitances of the LNA to optimize gain, compensating for the tolerance of external components for the match.

## 6.11 ROM 2 REGS

This command starts internal AMIS-53000 processes such as:

**Multi-Channel:** Calculates the frequency information to form nine channels of 300kHz bandwidth. Four of these channels are at higher frequencies than the programmed RF frequency and four channels are at lower frequencies.

**Wave Shaping:** Calculates the voltage steps to form the Gaussian wave shaping of the data for the data rate selected.

**Defined Data Rates:** Calculates all the parameters for the selected data rate (selected from the defined data rates).

**RF Frequency:** Calculates the parameters needed to set the TX/RX frequency

**AM Filters:** Calculates parameters for the filter defined by the data rate selection (defined data rates).

## 6.12 Chip Reset

Resets the entire chip, similar to a POR. This operation will reset the unlock test MUX register.

## 6.13 ADC Conversion

Table 75: ADC Configuration

| Register     |                     |          | Function   |
|--------------|---------------------|----------|--|
| Number (HEX) | Name                | Bits     |  |
| 0X00         | Command             | 010xxxxx | Perform a single ADC conversion (see Table 84)     |
|              |                     | 110xxxxx | Perform continuous ADC conversions (see Table 85)  |
| 0X01         | Status/Flag1        | 0        | The ADC conversion has completed                   |
| 0X1B         | Housekeeping Config | 4        | Do an ADC conversion for the internal measurements |
|              |                     | 3        | Do an ADC conversion for the external measurements |

The AMIS-53000 contains an 8 bit analog to digital converter. This ADC can measure the voltage on a number of internal functions, battery voltage, temperature, received signal strength indication voltage, and loop filter voltage. The results of these conversions are available through reading the registers where that data is stored or by using the feature of the Burst transmission to send that information to another node. The AMIS-53000 also contains two ADC channels available on the device pins. The ADC can convert signals at a conversion rate up to 128k samples/second.

Setup registers descriptions:

**Temp-** Contains the value from the last ADC of the internal temperature sensor.

**Battery-** Contains the value from the last ADC of the internal battery voltage (divided by 2).

**RSSI-** Contains the value from the last ADC of the signal level sample in the receiver.

**ADC1-** Contains the value from the last ADC of the external analog input.

**ADC2-** Contains the value from the last ADC of the external analog input.

**Loop Filter-** Contains the value from the last ADC of the loop filter voltage.

### 6.13.1. ADC Conversion Results

#### 6.13.1.1. Temp ADC

Table 76: Temp ADC - 0X34 [52]

| Bit | Name           | Comment                        |
|-----|----------------|--------------------------------|
| 7:0 | TEMP_ADC [7:0] | Temperature sensor ADC reading |



### 6.13.1.2. Battery ADC

Table 77: Battery ADC - 0X35 [53]

| Bit | Name           | Comment  |
|-----|----------------|--|
| 7:0 | BATT_ADC [7:0] | Battery voltage ADC reading ( $V_{\text{bat}}/2$ ) |

### 6.13.1.3. RSSI

Table 78: RSSI - 0X36 [54]

| Bit | Name           | Comment                  |
|-----|----------------|--------------------------|
| 7:0 | RSSI_ADC [7:0] | RSSI voltage ADC reading |

### 6.13.1.4. External Input 1 ADC

Table 79: External Input1 ADC - 0X37 [55]

| Bit | Name           | Comment                      |
|-----|----------------|------------------------------|
| 7:0 | EXT1_ADC [7:0] | External input 1 ADC reading |

### 6.13.1.5. External Input 2 ADC

Table 80: External Input2 ADC - 0X38 [56]

| Bit | Name           | Comment                      |
|-----|----------------|------------------------------|
| 7:0 | EXT2_ADC [7:0] | External input 2 ADC reading |

### 6.13.1.6. Loop Filter

Table 81: Loop Filter - 0X39 [57]

| Bit | Name            | Comment              |
|-----|-----------------|----------------------|
| 7:0 | LOOP_FILT [7:0] | Internal loop filter |

## 6.13.2. Single ADC Conversion

The single conversion command performs an ADC conversion on the channel specified as part of the command. Once complete, a flag is set, and the 8 bit data for the conversion is available in its associated register.

Table 82: Single ADC Conversions

| Register     |         |            |            | Function  |
|--------------|---------|------------|------------|---|
| Number (HEX) | Name    | Bits (7:6) | Bits (5:0) |   |
| 0X00         | Command | 01         | 000001     | Perform an ADC on the external input 1            |
|              |         | 01         | 000010     | Perform an ADC on the external input 2            |
|              |         | 01         | 000100     | Perform an ADC on the internal temperature sensor |
|              |         | 01         | 001000     | Perform an ADC on the Internal battery voltage    |
|              |         | 01         | 010000     | Perform an ADC on the receiver RSSI               |
|              |         | 01         | 100000     | Perform an ADC on the loop filter                 |

### 6.13.3. Continuous ADC Conversion

This command can be given to the radio and operate in parallel with transmit or receive. This mode can also be entered into from an idle state. In this mode the specified ADC channel is continuously converted, and its associated register is continuously over written.

Table 83: Single ADC Conversions

| Register     |         |            |            | Function  |
|--------------|---------|------------|------------|---|
| Number (HEX) | Name    | Bits (7:6) | Bits (5:0) |   |
| 0X00         | Command | 11         | 000001     | Perform continuous ADC on the external input 1            |
|              |         | 11         | 000010     | Perform continuous ADC on the external input 2            |
|              |         | 11         | 000100     | Perform continuous ADC on the internal temperature sensor |
|              |         | 11         | 001000     | Perform continuous ADC on the internal battery voltage    |
|              |         | 11         | 010000     | Perform continuous ADC on the receiver RSSI               |
|              |         | 11         | 100000     | Perform continuous ADC on the loop filter                 |

## 7.0 Data Interface

Table 84: Control/Data Interface Physical Configuration

| Interface        | Function | Clock |        | Data   |       | Select | AMIS-53000 | Data Buffering |
|------------------|----------|-------|--------|--------|-------|--------|------------|----------------|
|                  |          | Pin   | Source | Output | Input |        |            |                |
| I <sup>2</sup> C | Control  | SCLK  | Master | SDATA  | SDATA | None   | Slave only | N/A            |
| 3 -Wire          | Control  | SCLK  | Master | SDATA  | SDATA | SSN    | Slave only | N/A            |
|                  | Data     | DCLK  | Master | DRXTX  | DRXTX | DSSN   | Master     | Optional       |
|                  |          | DCLK  | Master | DRXTX  | DRXTX | DSSN   | Slave      | Buffered only  |
| 4-Wire           | Data     | DCLK  | Master | DRXTX  | DOPT  | DSSN   | Master     | Optional       |
|                  |          | DCLK  | Master | DRXTX  | DOPT  | DSSN   | Slave      | Buffered only  |

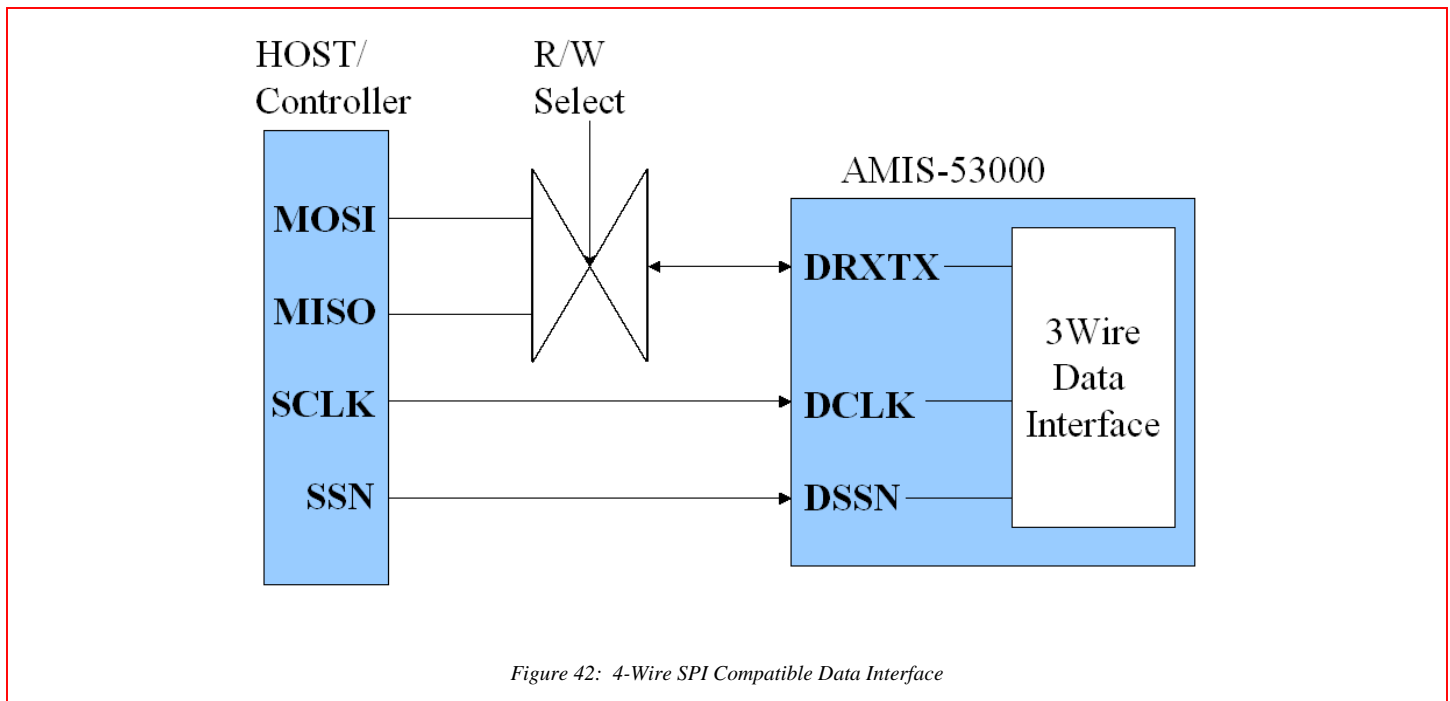


Figure 42: 4-Wire SPI Compatible Data Interface

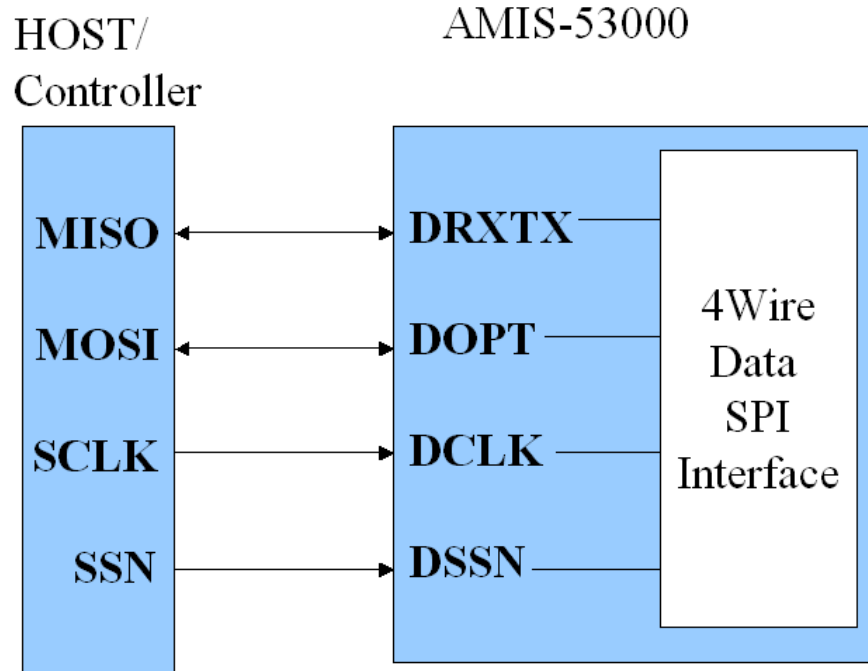


Figure 43: 3-Wire Serial Data Interface

The AMIS-53000 employs two different data interfaces. Transmit and receive data is exchanged with an external controller through either a 3-wire or a 4-wire SPI like interface. Selecting the interface, 3-wire or 4-wire is done by writing to bit 7 of the general options B register.

- T* SPI serial data interface
- T* 3-wire serial data interface
- T* AMIS-53000 can be the slave or master

Setup registers descriptions:

**Chip Address-** Allows a unique value to be transmitted and received with the data packet to identify a unique radio.

**Fixed Data Rates-** Select the options for one of several fixed data rates.

**General Options A-** Configure the interface options.

**General Options B-** Configure the interface options.

**Start of Frame-** Set a code value that indicates the start of a data packet.

**Preamble Length-** Select a type of preamble and set the length in bits. (see Section 6.6.3)

**Custom Data Rates-** Configure parameters for data rates that are not one of the fixed data rates.

**CRC Polynomial-** Value of the CRC polynomial.

**Default Length of Packet-** Set a default length for packets.

**Broad Cast ID-** A general chip ID allowing for transmissions to be received by all radios.

### 7.1.1. Chip Address MSB1

The 16 bit ID that can be used for several purposes in the AMIS-53000 is stored in the chip address MSB, and chip address LSB registers.

Table 85: Chip Address1 - 0X03 [3]

| Bit | Name            | Comment                    |
|-----|-----------------|----------------------------|
| 7:0 | Chip_Add [15:8] | Upper byte of chip address |

### 7.1.2. Chip Address LSB

Table 86: Chip Address0 - 0X04 [4]

| Bit | Name           | Comment                    |
|-----|----------------|----------------------------|
| 7:0 | Chip_Add [7:0] | Lower byte of chip address |

### 7.1.3. Data Rate/Format

The data rate/format register is used to select the data rate and format for both receive and transmit. The DDRATE[2:0] option bits allow selection of one of eight pre-programmed data rates. When one of the discrete data rates is selected, the ROM2REGS command is used to load clock and data recovery settings for the desired data rate into their associated registers.

The Manchester option bit configures the AMIS-53000 to transmit and receive in the Manchester encoded format, while the data interface remains NRZ.

If a data rate other than one of the available discrete rates is desired, the user should set the use custom bit, and then program the custom data rate register for the desired data rate. When the use custom data rate option is enabled, it is up to the user to set the correct sample clock frequency in the CDR options B register, set clock recovery loop filter settings, and if using the PLL based FSK detector, set the PLL detector loop filter.

Note: For data rates that are near one of the pre-defined data rates, a discrete data rate could first be chosen, the ROM2REGS command given to load all of the settings for the various blocks for that data rate, and then the custom data rate option enabled and the new data rate information entered.

For example, if the desired data rate is 100kbps, set DDRATE to 110 for 96kbps operation. Next, issue the ROM2REGS command in the command register. All of the proper settings for the clock and data recovery circuit for a 96k data rate will be loaded into the working registers from ROM (sample clock frequency, clock recovery loop filter settings). Finally, enable the use custom option, and program data rate 1, and 0 with the value for a 100k data rate.

Custom frequency is set in data rate 1 and data rate 0. If custom is 0, ROM contents for selected discrete data rate are loaded into data rate 1 and data rate 0.

Table 87: Data Rate/Format - 0X0B [11]

| Bit | Name         | State | Comment                             |
|-----|--------------|-------|-------------------------------------|
| 7   | NU           |       |                                     |
| 6   | NU           |       |                                     |
| 5   | NU           |       |                                     |
| 4   | Use Custom   | 1     | Enables user programmable data rate |
|     |              | 0     |                                     |
| 3   | Manchester   | 1     | Manchester encoding selected        |
|     |              | 0     | NRZ encoding selected               |
| 2:0 | DDRATE [2:0] | 000   | 1.2kbps                             |
|     |              | 001   | 2.4kbps                             |
|     |              | 010   | 4.8kbps                             |
|     |              | 011   | 9.6kbps                             |
|     |              | 100   | 19.2kbps                            |
|     |              | 101   | 57.6kbps                            |
|     |              | 110   | 96kbps                              |
|     |              | 111   | 128kbps                             |

### 7.1.4. General Options A

The general options A register contains a number of options that specify the operation of the part in its various modes.

**Standby Mode:** Determines whether the crystal oscillator is enabled during standby. For applications relying on the AMIS-53000 to provide an external host/controller with a system clock, this bit should be enabled, and is the default state.

**POR State:** Specifies the power on state of the device. Once this has been stored into EE, the device will power up in the chosen state after the EE has been shadowed into the working registers.

**Pull up Disable:** For applications not using an open drain type driver to drive the register interface pins (SDATA, SCLK and SSN) the pull ups on these pins can be disabled via this option bit to save power.

**Temperature Compensation:** When enabled, the ADC output for the temperature sensor is used to compensate the RF center frequency for crystal frequency error. A new correction factor is calculated each time the ADC performs a new conversion on the temperature sensor.

**CRC Enable:** Enables internal CRC checking in RX, and appends a CRC in TX.

**Length of Packet Enable:** Allows buffering of packets, also allows CRC when enabled.

**Use ID in RX and TX:** When enabled, in receive mode the part will not output data until a valid ID is found, and in TX, the part will automatically send preamble and chip ID before enabling the data interface.

Table 88: General OptionsA - 0X0C [12]

| Bit   | Name                     | State                | Comment  |
|-------|--------------------------|----------------------|--|
| 7     | Use ID in RX and TX      | 1                    | Wake on ID in RX, send ID in TX                      |
|       |                          | 0                    |  |
| 6     | Length of Packet Enable  | 1                    | Enables the part to frame packets                    |
|       |                          | 0                    |  |
| 5     | CRC Enable               | 1                    | Enables CRC (packet length must be enabled)          |
|       |                          | 0                    |  |
| 4     | Temperature Compensation | 1                    | RF center frequency temperature compensation enabled |
|       |                          | 0                    | Temperature compensation is disabled                 |
| 3     | Pull up Disable          | 1                    | Pull ups on IIC clock and data and SSN pins disabled |
|       |                          | 0                    |  |
| [2:1] | POR State                | 00<br>01<br>10<br>11 | Standby<br>Idle<br>RX<br>TX                          |
| 0     | Standby Mode             | 1                    | Crystal only mode, system clock output active        |
|       |                          | 0                    | Low-power standby mode                               |

### 7.1.5. General Options B

General options B contains more option bits for the general setup and operation of the AMIS-53000.

**System Clock Output Frequency:** Sets the frequency of the output clock on the SYSclk pin when enabled.

**RRTX Sampling Edge:** Specifies which edge of DCLK should be used to sample the RRTX pin.

**Auto Increment Disable:** When enabled, a multiple address read or write command on the register interface will read/write only the address given in the command multiple times.

**Data Interface Clock Frequency:** Sets the clock frequency for the data interface when the AMIS-53000 is configured to be the master of the data interface. For modes in which the AMIS-53000 does not buffer the packet, the interface speed will always be the data rate, regardless of this setting.

**Data Interface Slave/Master:** Specifies whether the AMIS-53000 is the master or slave for the data interface.

**4-Wire Data Interface:** Enables the 4-wire SPI data interface. When low, RRTX is bi-directional.

Table 89: General OptionsB - 0X0D [13]

| Bit | Name                           | State                | Comment  |
|-----|--------------------------------|----------------------|--|
| 7   | 4-Wire Data Interface          | 1                    | Enabled  |
|     |                                | 0                    |  |
| 6   | Data Interface Slave/Master    | 1                    | AMIS-53000 is slave  |
|     |                                | 0                    | AMIS-53000 is master, clock speed determined by bits 5, 4          |
| 5,4 | Data Interface Clock Frequency | 11<br>10<br>01<br>00 | 1MHz<br>500kHz<br>100kHz<br>Baud clock                             |
| 3   | NU                             |                      |  |
| 2   | RRTX Sampling Edge             | 1                    | Data bits are sampled on the rising edge of DCLK on the interface  |
|     |                                | 0                    | Data bits are sampled on the falling edge of DCLK on the interface |
| 1,0 | System Clock Output Frequency  | 11                   | 12MHz (24MHz external crystal)                                     |
|     |                                | 10                   | 6MHz (24MHz external crystal)                                      |
|     |                                | 01                   | 3MHz (24MHz external crystal)                                      |
|     |                                | 00                   | Off  |

## 7.1.6. Start of Frame

The start of frame byte is transmitted when this register is non-zero. It's used as an aid for the receiver clock and data recovery circuit in modes where the fast phase alignment feature is enabled.

Table 90: Start of Frame - 0X19 [25]

| Bit | Name      | Comment  |
|-----|-----------|--|
| 7:0 | SOF [7:0] | 8-bit code sent prior to chip ID in TX and Burst |

## 7.1.7. Data Rate 1

The data rate 1 and data rate 0 registers are used to set user defined data rates. These registers are loaded from ROM when a discrete data rate is selected. The following equation is used to calculate the value for CUST\_DR:

$$\text{CUST\_DR} = \text{DataRate} \cdot \frac{2^{22}}{F_{\text{sample\_clock}}}$$

where DataRate is the desired data rate, and  $F_{\text{sample\_clock}}$  is the frequency selected for the sample clock. This register is loaded with the discrete rate if selected.

Table 91: Data Rate1 - 0X29 [41]

| Bit | Name           | Comment   |
|-----|----------------|---|
| 7:0 | CUST_DR [15:8] | Upper byte of user defined data rate/discrete data rate |

## 7.1.8. Data Rate 0

Table 92: Data Rate0 - 0X2A [42]

| Bit | Name          | Comment   |
|-----|---------------|---|
| 7:0 | CUST_DR [7:0] | Lower byte of user defined data rate/discrete data rate |

## 7.1.9. CRC Polynomial

This register allows a designer to change the CRC Polynomial used in the AMIS-53000. The register represents the presence of the powers in the CRC equation. For example:

The Polynomial  $x^8+x^5+x^2+x+1$  is encoded by assuming the polynomial will always have a high order bit.

So the binary representation is: 1 0010 0111

This is set as the value 0X27 (HEX) in the register

(See "Koopman, P. & Chakravarty, T., " Cyclic Redundancy Code (CRC) Polynomial Selection For Embedded Networks" DSN04, June 2004." for more information.)

Table 93: CRC Poly - 0X30 [48]

| Bit | Name           | Comment              |
|-----|----------------|----------------------|
| 7:0 | CRC_POLY [7:0] | CRC polynomial value |



### 7.1.10. Default Length of Packet

This register allows a default value for the LOP (5 bytes) such that the AMIS-53000 does not have to send the LOP with a buffered packet.

Table 94: Default LOP - 0X31 [49]

| Bit | Name              | Comment   |
|-----|-------------------|---|
| 7:0 | DEFAULT_LOP [7:0] | Default value for the length of packet to be used in buffered TX/RX |

### 7.1.11. Broadcast ID 1

Many applications in the wireless market make use of a broadcast function where the master node in a system can transmit to all wireless nodes in the network without addressing each node individually, but still not broadcasting to nodes in another network.

Table 95: Broadcast ID1 - 0X3A [58]

| Bit | Name             | Comment                          |
|-----|------------------|----------------------------------|
| 7:0 | Global_ID1 [7:0] | Lower byte of the global address |

### 7.1.12. Broadcast ID 0

Table 96: Broadcast ID0 - 0X3B [59]

| Bit | Name             | Comment                          |
|-----|------------------|----------------------------------|
| 7:0 | Global_ID0 [7:0] | Upper byte of the global address |

## 7.2 TX/RX Data Interface Protocol

The AMIS-53000 TX/RX data format can be streaming data where the transmitter transmits each bit of data as it is received or it can be packetized. Packetized data can be in packets up to 256 bytes. Packetized data can add a preamble, start of frame, identification code, length of packet, and CRC error CheckSum.

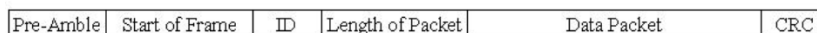


Figure 44: Data Protocol Format

- Streaming or packetized data
- Buffer size is 256 byte maximum
- Packet overhead

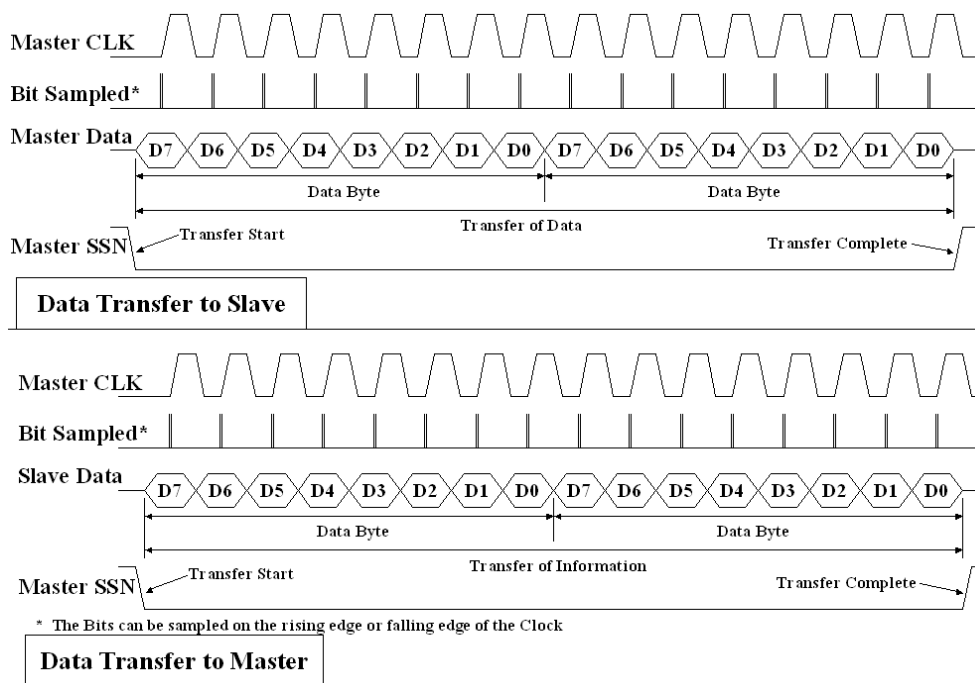


Figure 45: 3-Wire Data Transfer Protocol

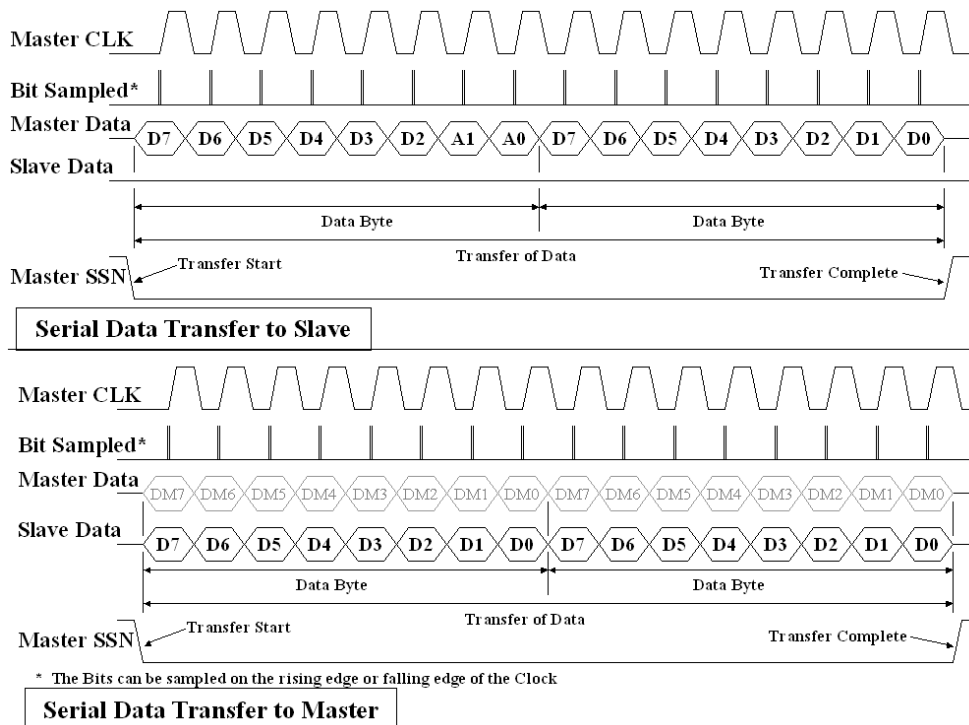


Figure 46: 4-Wire Data Transfer Protocol

Table 97: TX/RX Data Protocols

| Modulation   | Detector | CDR              | Preamble     | SOF              | ID/LOP/CRC   | Slice      |
|--------------|----------|------------------|--------------|------------------|--------------|------------|
| AM           | RSSI     | Opt <sup>1</sup> | CW           | Yes <sup>2</sup> | See Table 98 | Fixed/Auto |
| FM (<20kbps) | PLL      | Yes              | 1 0 pattern  | -----            | See Table 98 | -----      |
| FM (>20kbps) | FFT      | Yes              | <sup>3</sup> | <sup>4</sup>     | See Table 98 | -----      |

Notes:

1. The use of the CDR function to recover the data is recommended for AM/OOK modulation.
2. The SOF for AM modulation is suggested to be 55 (HEX) for NRZ and 0A (HEX) for Manchester encoded data.
3. The preamble for FM (FFT) with NRZ data is a 1 0 repeating pattern. The preamble for FM (FFT) with Manchester encoded data is all 1s or all 0s.
4. A SOF is only required for FM (FFT) when the data is Manchester encoded. The suggested SOF is a pattern of 55 (HEX) or AA (HEX).

Table 98: Interface Data Protocols

| TX/RX Data Protocol |     | Interface Data Protocol |                     | Comments  |
|---------------------|-----|-------------------------|---------------------|---|
| LOP                 | CRC | Interface               | Data                |   |
| N                   | N   | Active                  | Stream <sup>1</sup> | Data is streamed out the interface as it is received                |
| N                   | N   | Active*                 | Stream              | * Data is streamed out the interface starting with the wakeup on ID |
| Y                   | N   | Interrupt               | Buffered            | An interrupt is issued when data reception is complete              |
| Y                   | Y   | Interrupt               | Buffered            | An interrupt is issued when data reception is complete              |

Notes:

1. When the interface uses streaming data, the AMIS-53000 must be the master.

The serial data interface for the AMIS-53000 can be configured to be a 3-wire interface or a 4-wire SPI interface. The AMIS-53000 can be configured to act as a master or a slave for both receive and transmit operation. Bit 2 in the general options B register allows the user to select whether DATA will be sampled on the rising, or falling edge of DCLK. The setting for the sampling polarity applies to all modes.

Table 99: Serial Data Interface Configuration

| General Options B |       |       | Data Port Configuration |            |             |                         |        |        |       |
|-------------------|-------|-------|-------------------------|------------|-------------|-------------------------|--------|--------|-------|
| Bit 7             | Bit 6 | Bit 2 | # Port Pins             | AMIS-53000 | Edge Sample | Pin Function Definition |        |        |       |
|                   |       |       |                         |            |             | DCLK                    | DSSN   | DRXTX  | DOPT  |
| 0                 | 0     | X     | 3                       | Master     | X           | Output                  | Output | I/O    | X     |
| 0                 | 1     | X     | 3                       | Slave      | X           | Input                   | Input  | I/O    | X     |
| 1                 | 0     | X     | 4                       | Master     | X           | Output                  | Output | Output | Input |
| 1                 | 1     | X     | 4                       | Slave      | X           | Input                   | Input  | Output | Input |
| X                 | X     | 0     | X                       | X          | Falling     |                         |        |        |       |
| X                 | X     | 1     | X                       | X          | Rising      |                         |        |        |       |

### 7.2.1. AMIS-53000 in Master Mode

In receive mode, the DSSN pin will transition low when the AMIS-53000 has received data. Immediately following the transition of DSSN, the AMIS-53000 will provide a synchronized bit clock on DCLK, and the received data will appear on DRXTX.

In transmit mode, the transition of DSSN is used to signal an external host/controller that the AMIS-53000 is ready for transmit data and is ready to receive that data on the DRXTX pin. Immediately following the transition of DSSN, the AMIS-53000 will provide a synchronous clock on DCLK for the host/controller to use for loading transmit data into the AMIS-53000.

## 7.2.2. AMIS-53000 in Slave Mode

The AMIS-53000 cannot be the slave for streaming data. The requirements of adding header information such as preamble or SOF requires that the AMIS-53000 be in control of the data interface transfer. The receiver has similar requirements with removing the header information.

As the slave for the data interface, the AMIS-53000 will simply issue an interrupt to the external host indicating data is available after a data packet has been received.

For buffered transmit operation, the AMIS-53000 will issue an interrupt indicating it is ready to load the packet. After the packet is received by the AMIS-53000, the transmitter is enabled, any packet formatting is done and the packet is sent.

## 7.2.3. Manchester Operation

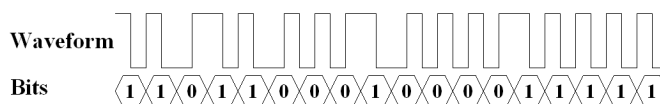


Figure 47: Manchester Coded Data

If the Manchester option is selected in the data rate and format register, the AMIS-53000 will internally encode and decode both transmit and receive data respectively. The format for the signal interface remains NRZ in this mode.

## 7.2.4. Packet Framing

Three options bits located in the general options A register determine the method with which the transceiver will process packets: use ID, length of packet (LOP) enable and cyclic redundancy check (CRC) enable. An additional bit in the TX config register enables the buffered TX mode of operation.

Table 100: Register Configuration Bits

| General Options A |       |       | TX Config |       | SOF       | Use        | Comments   |
|-------------------|-------|-------|-----------|-------|-----------|------------|--|
| Bit 7             | Bit 6 | Bit 5 | Bit 3     | Bit 2 | All       |            |  |
| 0                 | X     | X     | X         | X     | X         | No ID      |  |
| 1                 | X     | X     | X         | X     | X         | ID         |  |
| X                 | 0     | X     | X         | X     | X         | No LOP     |  |
| 1                 | 1     | X     | X         | X     | X         | LOP/ID     | The ID must be enabled with LOP for data alignment   |
| X                 | X     | 0     | X         | X     | X         | No CRC     |  |
| 1                 | 1     | 1     | X         | X     | X         | CRC/ID/LOP |  |
| X                 | 1     | X     | 1         | X     | X         | Buffer/LOP | TX must have LOP enabled when using data buffering   |
| X                 | X     | X     | 0         | X     | X         | No Buffer  |  |
| X                 | X     | X     | X         | 0     | X         | Chip ID    |  |
| X                 | X     | X     | X         | 1     | X         | Global ID  |  |
| X                 | X     | X     | X         | X     | 0x00      | No SOF     |  |
| ?                 | X     | X     | X         | X     | All other | SOF        | Certain configurations require SOF to detect ID byte |

## 7.2.5. Use ID

Table 101: ID

| Register     |                   |      | Function   |
|--------------|-------------------|------|--|
| Number (HEX) | Name              | Bits |  |
| 0X0C         | General Options A | 7    | Wake on ID in RX/send ID with TX                                   |
| 0X0F         | TX Config         | 2    | Select either the chip ID or global ID to be used in transmissions |
| 0X16         | Burst Config      | 3    | Send ID with Burst packet  |

The chip ID is a 16 bit word which can be programmed in registers 3 and 4. In receive mode, when the use ID bit in general options A is set, the AMIS-53000 will not begin exporting or buffering data until a valid ID matching the value stored in the chip address registers is received. The ID is used in more advanced modes of operation for byte alignment. In addition to waking on its own unique ID, the AMIS-53000 will also wake on a pre-defined global chip ID. The default value for the global ID is in the register table. This value can be overwritten, but is not stored in EE so care must be taken when overwriting the value.

With the use ID bit enabled in transmit mode, the AMIS-53000 will transmit the chip ID prior to enabling the data interface. An additional option bit in the TX config register allows selection of either the chip ID or global ID value for transmit.

In either transmit or receive, when the use ID bit is enabled without LOP enabled, the AMIS-53000 will not buffer data. Hence when enabled stand alone, the data interface must be configured with the AMIS-53000 as the master.

## 7.2.6. Length of Packet Enable

The length of packet enable (LOP) bit located in general options A, enables the AMIS-53000 to buffer packets. The use ID bit must be used in conjunction with LOP to allow the receiver to byte align on incoming data.

In receive mode with the LOP enabled, the AMIS-53000 will interpret the first byte following either a valid chip ID, or global ID to be the length of the incoming packet. This byte specifies the number of bytes following the LOP to be received (non-inclusive of the CRC if enabled). When enabled, the AMIS-53000 will buffer the incoming packet into internal RAM. Following reception of the last byte of the packet, an interrupt is issued on the interrupt pin, and depending on the configuration of the data interface, the packet will either be sent out of the data interface by the AMIS-53000 as master, or wait for the external host/controller to stream the packet out as the master.

Having the LOP enabled in transmit mode allows for the use of the buffered TX packet option in transmit, or the AMIS-53000 can still act as master and process the packet on the fly. With LOP enabled, and buffered TX disabled, the AMIS-53000 must be the master for the data interface. In this mode, the preamble and chip ID (or global ID) will be sent before the data interface is activated. Once the DSSN is pulled low by the AMIS-53000, the first byte received into the part is expected to be the LOP byte. Transmission continues until the AMIS-53000 has determined that all bytes have been received, at which point the data interface is disabled, and the AMIS-53000 will return to standby. When buffered TX is enabled, after the transmit instruction is given to the AMIS-53000, an interrupt from the AMIS-53000 will be issued indicating the part is ready to load in the data packet. The actual loading of the data packet depends on the data interface setup as to whether the AMIS-53000 is master or slave. The first byte is again expected to be the LOP byte. After the complete packet is loaded into the radio, the RF will be enabled, the preamble and chip ID transmitted, followed by the packet. After completion of the transmission, the AMIS-53000 will return to standby.

## 7.2.7. CRC Enable

The CRC enable located in general options A is the final tier of intelligence for the AMIS-53000 packet handling capability. In order for the AMIS-53000 to do CRC checking, this option must be used in conjunction with both use ID and LOP enable. Operation of the interface for both receive and transmit with the CRC enabled is no different from that explained under the LOP enabled section. With the CRC enabled, the AMIS-53000 will append the calculated CRC in transmit as the last byte. In receive mode, interrupts to the external controller will only be issued for packets passing the CRC.

## 7.2.8. SOF Byte

Table 102: Suggested SOF

| Modulation   | Detector | Coding      | Preamble       | SOF                  |
|--------------|----------|-------------|----------------|----------------------|
| AM           | RSSI     | NRZ         | CW             | 55 (HEX)             |
|              |          | Manchester  | CW             | 0A (HEX)             |
| FM (<20kbps) | PLL      | 1 0 pattern | 1 0 pattern    | Not required         |
| FM (>20kbps) | FFT      | NRZ         | Repeating 1/0  | Not required         |
|              |          | Manchester  | All 1's or 0's | 55 (HEX) or AA (HEX) |

Depending on whether the mode of operation is AM or FM, NRZ or Manchester, it may be necessary for a SOF byte to precede the chip ID. This byte is user programmable, and is used to ensure proper CDR operation and bit alignment prior to reception of the chip ID. When the contents of the SOF byte register are loaded to any non-zero value, this byte will be transmitted prior to the chip ID. For modes not requiring the SOF byte, setting this register to 00h will prohibit transmission of this byte. More information on when the SOF byte is required is in the clock and data recovery section.

## 7.2.9. Timing Diagrams for Various Packet Framing Modes

### 7.2.9.1. Use ID Enabled, No CRC, No Packet Length

Table 103: Receive

| Parameter | Action  |
|-----------|---|
| ID        | Data interface immediately ready after ID detected                |
| No LOP    | Radio stays in RX until instructed to change                      |
|           | Data is shifted out data port as received                         |
| No CRC    | Error checking is not performed and CRC is not attached to packet |

Table 104: Transmit

| Parameter | Action   |
|-----------|--|
| Transmit  | Transmit command immediately powers the transmitter on                                   |
|           | Transmits preamble   |
| ID        | Transmits the ID   |
|           | Starts data interface and uses a synchronous clock to clock in the TX data (master only) |
|           | Transmits each bit as received   |
|           | Transmitter returns to standby after transmission complete                               |

Use:

- ID = 1 (TX: send ID/RX Wake on ID)
- CRC Enable = 0 (no CRC check)
- LOP Enable = 0 (no Length of Packet)

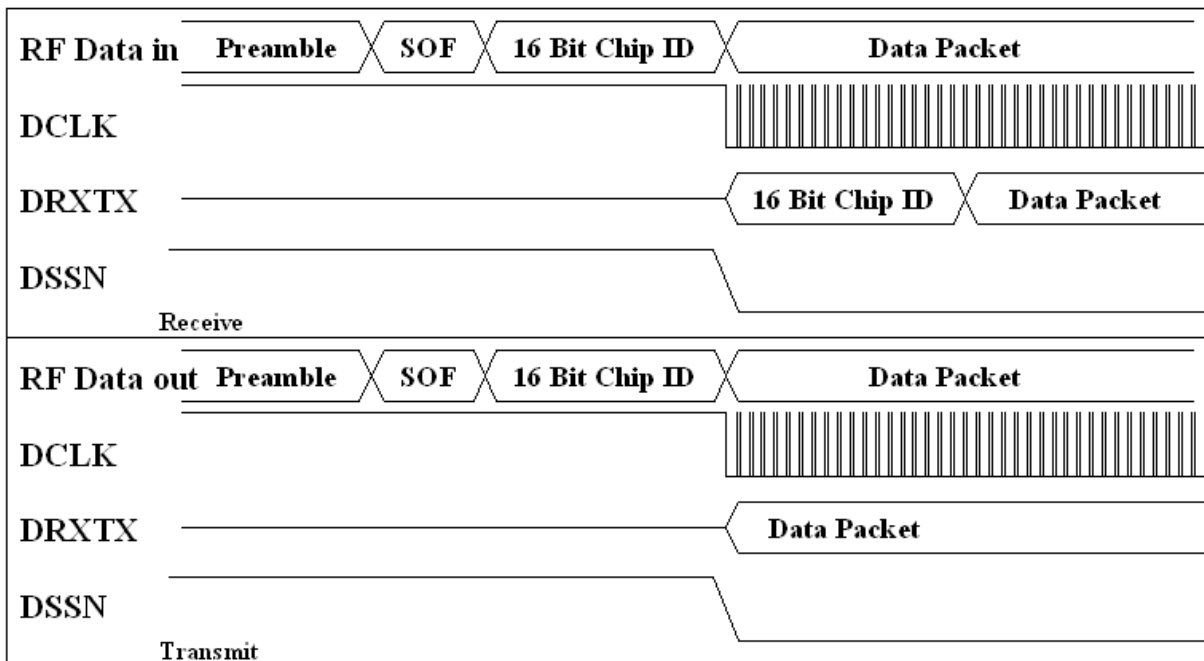


Figure 48: Data Interface Protocol (ID only)

### 7.2.9.2. Use ID Enabled, No CRC, LOP Enabled

Table 105: Receive

| Parameter | Action  |
|-----------|---|
| ID        | Data interface immediately ready after ID detected  |
| LOP       | Receiver loads rest of packet into buffer memory  |
|           | After last data byte is received, radio returns to the previous state                               |
|           | An interrupt is issued to the external controller/microprocessor                                    |
|           | Data is transferred out the port with the AMIS-53000 as master or the external controller as master |
| No CRC    | Error checking is not performed and CRC is not attached to packet                                   |

Table 106: Transmit

| Parameter         | Action  |
|-------------------|---|
| Buffered Transmit | If buffered transmit is selected, the AMIS-53000 will open the data interface and transfer all TX data into memory with AMIS-53000 as master or external controller as master |
| Transmit          | Transmit command (or end of TX data transfer) immediately powers the transmitter on   |
|                   | Transmits preamble (length of preamble as specified)  |
| ID                | Transmits the SOF and the ID  |
|                   | Starts data interface and uses a synchronous clock to clock in the TX data (master only) or clocks data out of memory (buffered TX)   |
|                   | After the packet is transmitted, the transmitter returns to standby state   |

Use:  
 ID = 1 (TX: send ID/RX Wake on ID)  
 CRC Enable = 0 (no CRC check)  
 LOP Enable = 1 (use Length of Packet)

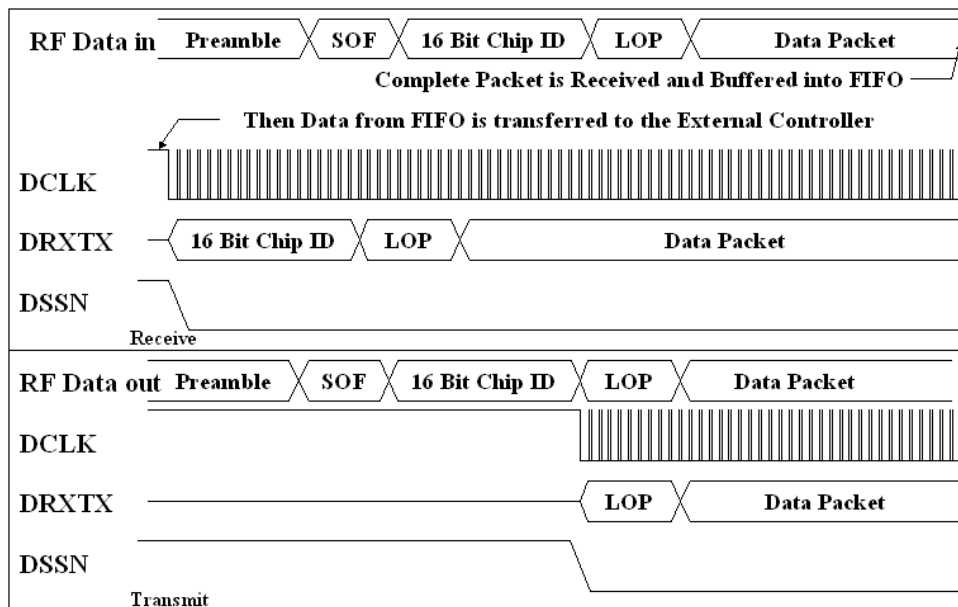


Figure 49: Data Interface Protocol (ID and LOP)

### 7.2.9.3. Use ID Enabled, CRC Enabled, LOP Enabled

Table 107: Receive

| Parameter   | Action   |
|-------------|--|
| ID          | Data interface immediately ready after ID detected   |
| CRC         | As soon as the ID is validated, the CRC starts processing the data   |
| LOP         | The LOP is received  |
|             | The rest of the received data packet is buffered into memory   |
| CRC Invalid | The last byte is the CRC and if invalid, the receiver waits for a command from the external host/controller, or if the receiver came from idle it will return to receive |
| CRC Valid   | Receiver returns to previous state and an interrupt is issued to the external controller   |
|             | The data interface is started and the data is sent to the controller, except for the CRC   |

Table 108: Transmit

| Parameter         | Action  |
|-------------------|---|
| Buffered Transmit | If buffered transmit is selected, the AMIS-53000 will open the data interface and transfer all TX data into memory with AMIS-53000 as master or external controller as master |
| Transmit          | Transmit command (or end of TX data transfer) immediately powers the transmitter on   |
|                   | Transmits preamble (length of preamble as specified)  |
| ID                | Transmits the SOF and the ID  |
| CRC               | The CRC begins processing the data with the ID  |
|                   | Starts data interface and uses a synchronous clock to clock in the TX data (master only) or clocks data out of memory (buffered TX)   |
| LOP               | The first byte is defined to be the LOP of the packet   |
| CRC Byte          | At the end of the packet, the data stops and the CRC value is sent  |
|                   | After the packet is transmitted, the transmitter waits for a command from the external host/controller  |



Use:

- ID = 1 (TX: send ID/RX Wake on ID)
- CRC Enable = 1 (use CRC check)
- LOP Enable = 1 (use Length of Packet)

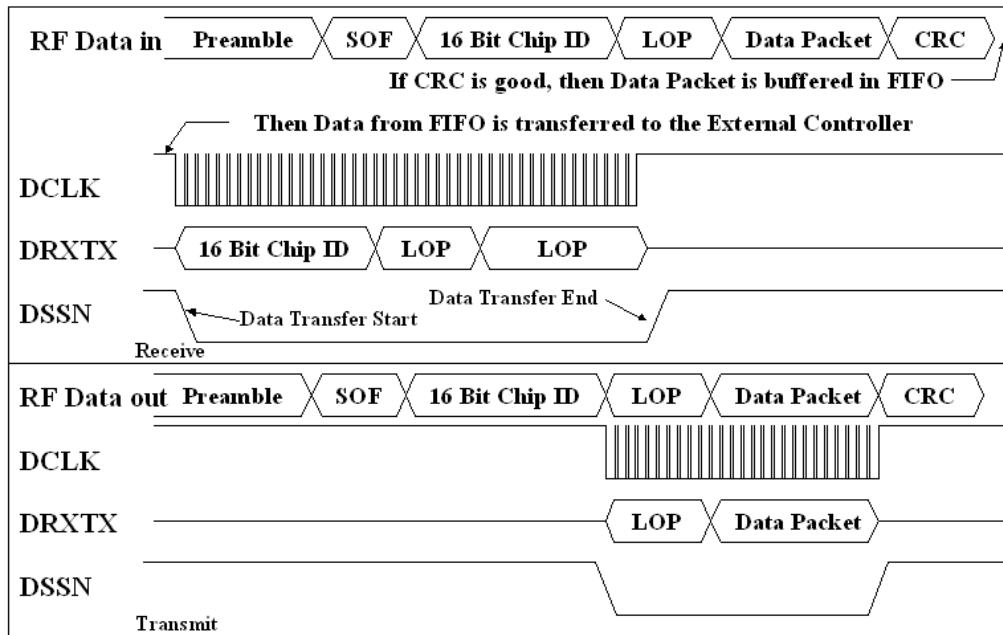


Figure 50: Data Interface Protocol (ID, LOP and CRC)

## 8.0 General System Functions

### 8.1 Pull up Disable

The AMIS-53000 includes built in pull-up resistors for use with the I<sup>2</sup>C operation to reduce the overall system component count. The pull ups are asserted at POR until mode selection occurs. If mode is determined to be 3-wire, the pull ups are removed. If mode is determined to be I<sup>2</sup>C, this option bit determines whether the pull ups are to be removed.

Table 109: I<sup>2</sup>C Pull Up Control

| Register     |                   |      | Function   |
|--------------|-------------------|------|--|
| Number (HEX) | Name              | Bits |  |
| 0X0C         | General Options A | 3    | Disable internal pull up resistors on I <sup>2</sup> C bus |

### 8.2 Brown-Out POR

The brown-out POR serves two purposes. The first is to provide a POR signal to reset the digital when power is initially applied to the part. The second is to provide a POR should the voltage on the supply drift below normal operating range to prevent a brown-out condition.

Table 110: Power-on-Reset Start Up State

| Register     |                   |      | Function   |
|--------------|-------------------|------|------------|
| Number (HEX) | Name              | Bits |            |
| 0X0C         | General Options A | 2,1  | 00 Standby |
|              |                   |      | 01 Idle    |
|              |                   |      | 10 RX      |
|              |                   |      | 11 TX      |

### 8.3 Temperature Sensor

The temperature sensor is created by using a Darlington pair of PNP transistors. The two transistors create a 5mV/°C slope that can be sensed with an analog to digital converter. Without amplification, an 8 bit ADC with a 2V reference voltage will have a resolution better than two degrees.

The temperature sensor can be trimmed to an accuracy of 3°C. As the trim is increased, the output voltage also increases.

The temperature voltage relationship is given by:

$$V = 1.1172 - T \cdot 0.00531$$

Where V is the output voltage and T is the temperature in Celsius.

Setup registers descriptions:

**ADC Temperature-** Register shows the value of the temperature sensor ADC. (See Section 6.13.1.1)

### 8.3.1. Crystal Temperature Compensation

An on-chip temperature sensor combined with an on-chip A/D and a look-up table enable the part to maintain RF frequency accuracy within  $\pm 2.5\text{ppm}$  over all operating voltages and temperatures (-45°C to 85°C). This function can be enabled via configuration bit 4 located in general options A. When this function is enabled, a new calculation for the center frequency word will be performed whenever the temperature sensor storage register is updated with a new value. Therefore it is possible to update the compensation value either in housekeeping, as part of a Burst transmit cycle, or as controlled externally by issuing the instruction to perform an ADC measurement of the temp sensor.

Table 111: RF Frequency Temperature Compensation

| Register     |                   |      | Function                                     |
|--------------|-------------------|------|--|
| Number (HEX) | Name              | Bits |  |
| 0X0C         | General Options A | 4    | Enable RF frequency temperature compensation |

## 8.4 Software

The version of the AMIS-53000 is written to a register at the end of the manufacturing process. This code can help AMIS wireless product support when there is an issue with the AMIS-53000.

Setup registers descriptions:

**AMIS ID-** Register contains a code showing the version of the AMIS-53000.

### 8.4.1. AMIS Part Revision Code

Table 112: AMIS Part Revision Code - 0X41 [65]

| Bit | Name          | Comment                           |
|-----|---------------|-----------------------------------|
| 7:0 | AMIS_ID [7:0] | Revision status of the AMIS-53000 |

## 9.0 Built-in Test Functions

The AMIS-53000 has a number of test registers. These registers are not available to the general user of the AMIS-53000. However, many of these registers control test features that are useful in the development of applications using the AMIS-53000.

Setup registers descriptions:

**Test Unlock-** A special code is required to unlock the functions of the test registers.

**Test-** Registers that route signals to pins for monitoring or turn internal circuits off for test.

### 9.1 TM Unlock Register

The developer designing the AMIS-53000 may desire to use some of the test modes to monitor the operation of the AMIS-53000 or to determine the activity of some parameter. These registers are locked from use by a code word. To unlock the test registers contact AMIS wireless product support to obtain the code. Enter this code in the unlock register to access the test registers. This register will be reset with a reset of the part and thus will lock the user out of the test registers.

Setup registers descriptions:

**Test Unlock-** Register contains a code about the state the AMIS-53000 is operating in.

\*\* Registers with the \*\* mark can be trimmed if the test registers are unlocked.

Table 113: Test Unlock Code - 0X40 [64]

| Bit | Name         | Comment  |
|-----|--------------|--|
| 7:0 | UNLOCK [7:0] | Code to unlock operation of the test registers (contact AMIS for the code to unlock the test register functions) |

### 9.2 Test Registers

The following registers allow for signals to be routed to pins for monitoring. They also turn functions in the AMIS-53000 on and off for measuring the operational parameters of the AMIS-53000.

#### 9.2.1. IF Amp Manual Trim A

This register is used for factory testing of the AMIS-53000 and has no user functions.

#### 9.2.2. IF Amp Manual Trim B

This register is used for factory testing of the AMIS-53000 and has no user functions.

#### 9.2.3. PLL Manual Trim

This register is used for factory testing of the AMIS-53000 and has no user functions.

### 9.2.4. PLL Test Modes

This register is used for factory testing of the AMIS-53000 and has no user functions.

### 9.2.5. Power Down RF Sections

This register is used for factory testing of the AMIS-53000 and has no user functions.

### 9.2.6. Analog Test Mode

**Digital Pad Test:** All digital pads except system clock out, and xInterrupt are 1mA I/O with pull ups and Schmitt triggers. The SYSclk and xInterrupt pads are 2mA outputs.

**Auto Increment Disable:** This disables the automatic incrementing of the I<sup>2</sup>C register addresses. It can allow repeated writes to the same register, useful for adjusting a parameter to optimize its value.

Table 114: Analog Test Mode - 0X47 [71]

| Bit                              | Name                                | State             | Comment   |                   |                                  |                                  |                                     |
|----------------------------------|-------------------------------------|-------------------|---|-------------------|----------------------------------|----------------------------------|-------------------------------------|
| 7                                | CAP_TRIM                            | 1                 | Enable the test mode for determination of capacitance trim value  |                   |                                  |                                  |                                     |
|                                  |                                     | 0                 |   |                   |                                  |                                  |                                     |
| 6                                | Pipe ADC1 to Data Filter            | 1                 | Enable the ADC1 input channel as a direct input to the data filter  |                   |                                  |                                  |                                     |
|                                  |                                     | 0                 | Normal operation  |                   |                                  |                                  |                                     |
| 5                                | Brown-Out Power Down                | 1                 | Override the brown-out POR to allow test at any voltage   |                   |                                  |                                  |                                     |
|                                  |                                     | 0                 |   |                   |                                  |                                  |                                     |
| 4                                | Auto Increment Disable              | 1                 | Address increment disabled (IIC only)   |                   |                                  |                                  |                                     |
|                                  |                                     | 0                 |   |                   |                                  |                                  |                                     |
| 3                                |                                     | 1                 |   |                   |                                  |                                  |                                     |
|                                  |                                     | 0                 |   |                   |                                  |                                  |                                     |
| 2                                | Ignore XTAL Control                 | 1                 | Ignore crystal control (digital clock gating)   |                   |                                  |                                  |                                     |
|                                  |                                     | 0                 |   |                   |                                  |                                  |                                     |
| 1                                | Dig Pad Test A                      | 1                 | <table border="0"> <tr> <td>Enabled:<br/>Input</td> <td>Enabled:<br/>Corresponding output</td> </tr> <tr> <td>SSSN<br/>DClock<br/>DSSN<br/>xBurst</td> <td>Doptional<br/>RXTX<br/>xINT<br/>sysClk</td> </tr> </table> | Enabled:<br>Input | Enabled:<br>Corresponding output | SSSN<br>DClock<br>DSSN<br>xBurst | Doptional<br>RXTX<br>xINT<br>sysClk |
|                                  |                                     | Enabled:<br>Input | Enabled:<br>Corresponding output  |                   |                                  |                                  |                                     |
| SSSN<br>DClock<br>DSSN<br>xBurst | Doptional<br>RXTX<br>xINT<br>sysClk |                   |   |                   |                                  |                                  |                                     |
| 0                                | Disabled                            |                   |   |                   |                                  |                                  |                                     |
| 0                                | Dig Pad Test B                      | 1                 | <table border="0"> <tr> <td>Enabled:<br/>Input</td> <td>Enabled:<br/>Corresponding output</td> </tr> <tr> <td>Doptional<br/>xBurst<br/>RXTX</td> <td>SSSN<br/>xINT<br/>DSSN</td> </tr> </table>                       | Enabled:<br>Input | Enabled:<br>Corresponding output | Doptional<br>xBurst<br>RXTX      | SSSN<br>xINT<br>DSSN                |
|                                  |                                     | Enabled:<br>Input | Enabled:<br>Corresponding output  |                   |                                  |                                  |                                     |
| Doptional<br>xBurst<br>RXTX      | SSSN<br>xINT<br>DSSN                |                   |   |                   |                                  |                                  |                                     |
| 0                                | Disabled                            |                   |   |                   |                                  |                                  |                                     |

### 9.2.7. RF Test Modes

This register is used for factory testing of the AMIS-53000 and has no user functions.

### 9.2.8. Analog Test MUX

This register is used for factory testing of the AMIS-53000 and has no user functions.

## 9.2.9. RF Test MUX

This register is used for factory testing of the AMIS-53000 and has no user functions.

## 9.2.10. Digital Test MUX A

Table 115: Digital Test MUX A - 0X4B [75]

| Bit | Test Pin    | Comment     |      |                             |      |                     |
|-----|-------------|-------------|------|-----------------------------|------|---------------------|
| 7:4 | MUX to SCLK | 1111 - 0000 | 1111 | Space Q channel CLK         | 0111 | Energy dwell enable |
|     |             |             | 1110 | Mark Q channel CLK          | 0110 | PLL increment       |
|     |             |             | 1101 | NC                          | 0101 | TX enable           |
|     |             |             | 1100 | Data Q channel              | 0100 | 10kHz clock         |
|     |             |             | 1011 | PN code from $\Sigma\Delta$ | 0011 | Software state [3]  |
|     |             |             | 1010 | Start                       | 0010 | Bandgap ready       |
|     |             |             | 1001 | Analog data out             | 0001 | ADC CLK             |
|     |             |             | 1000 | PLL detect/data out         | 0000 | Normal/system clock |
| 3:0 | MUX to Dopt | 1111 - 0000 | 1111 | Space I channel CLK         | 0111 | Code dwell enable   |
|     |             |             | 1110 | Mark I channel CLK          | 0110 | PLL decrement       |
|     |             |             | 1101 | PLL detect/NCO out          | 0101 | Kicker              |
|     |             |             | 1100 | Energy detected             | 0100 | PLL Z               |
|     |             |             | 1011 | Data I channel              | 0011 | Software state [2]  |
|     |             |             | 1010 | RF PLL CLK feedback         | 0010 | PLL xReset          |
|     |             |             | 1001 | Is locked (encoder)         | 0001 | ADC power down      |
|     |             |             | 1000 | TX data                     | 0000 | D optional          |

## 9.2.11. Digital Test MUX B

Table 116: Digital Test MUX B - 0X4C [76]

| Bit | Test Pin      | Comment     |      |                        |      |                        |
|-----|---------------|-------------|------|------------------------|------|------------------------|
| 7:4 | MUX to xINT   | 1111 - 0000 | 1111 | Encoder in             | 0111 | Cal done kicker        |
|     |               |             | 1110 | Decoder in             | 0110 | PLL in range           |
|     |               |             | 1101 | Sniff                  | 0101 | INT0                   |
|     |               |             | 1100 | $\Sigma\Delta$ output  | 0100 | Transmit done          |
|     |               |             | 1011 | RF PLL (reference CLK) | 0011 | Software state [1]     |
|     |               |             | 1010 | Brown-out output       | 0010 | Xtal on                |
|     |               |             | 1001 | Receive done           | 0001 | ADC done               |
|     |               |             | 1000 | TS CLK                 | 0000 | xInterrupt             |
| 3:0 | MUX to xBurst | 1111 - 0000 | 1111 | Recovered clock        | 0111 | PA enable              |
|     |               |             | 1110 | Decoder out            | 0110 | PLL cal timer overflow |
|     |               |             | 1101 | Encoder out            | 0101 | PLL cal enable         |
|     |               |             | 1100 | CDR enable             | 0100 | xtal PD                |
|     |               |             | 1011 | Baud clock (CDR out)   | 0011 | Software state [0]     |
|     |               |             | 1010 | CRC failed             | 0010 | isStopMode             |
|     |               |             | 1001 | RX enable              | 0001 | Watch dog reset        |
|     |               |             | 1000 | NC                     | 0000 | SSN normal mode        |

## 9.2.12. Digital Test MUX C

Table 117: Digital Test MUX C - 0X4D [77]

| Bit | Test Pin        | Comment     |      |                       |      |                      |
|-----|-----------------|-------------|------|-----------------------|------|----------------------|
| 7:4 | MUX to Data SSN | 1111 - 0000 | 1111 | Encoder in            | 0111 | RAMBist bad          |
|     |                 |             | 1110 | Decoder in            | 0110 | EE BIST done         |
|     |                 |             | 1101 | EE BIST good          | 0101 | EE BIST bad          |
|     |                 |             | 1100 | EE low voltage detect | 0100 | Busy                 |
|     |                 |             | 1011 | CPENA                 | 0011 | Instruction enable   |
|     |                 |             | 1010 | ROM BIST done         | 0010 | Bandgap power down   |
|     |                 |             | 1001 | ROM BIST bad          | 0001 | XTAL                 |
|     |                 |             | 1000 | RAM BIST done         | 0000 | Data SSN normal mode |
| 3:0 |                 | 1111- 0000  |      |                       |      |                      |
|     |                 |             |      |                       |      |                      |
|     |                 |             |      |                       |      |                      |
|     |                 |             |      |                       |      |                      |
|     |                 |             |      |                       |      |                      |
|     |                 |             |      |                       |      |                      |
|     |                 |             |      |                       |      |                      |
|     |                 |             |      |                       |      |                      |

## 9.2.13. Digital Test Mode A

This register is used for factory testing of the AMIS-53000 and has no user functions.

## 9.2.14. Digital Test Mode B

This register is used for factory testing of the AMIS-53000 and has no user functions.

## 9.2.15. Digital Test Mode C

This register is used for factory testing of the AMIS-53000 and has no user functions.

## 9.2.16. Digital Test Mode D

This register is used for factory testing of the AMIS-53000 and has no user functions.

## 9.2.17. Memory Test Mode Address

This register is used for factory testing of the AMIS-53000 and has no user functions.

## 9.2.18. Memory Test Mode Data

This register is used for factory testing of the AMIS-53000 and has no user functions.

## 10.0 Register Definition

Table 118 below contains the address for all of the internal registers. Once the EE has been written, the POR states for the registers become the data last written. Should the CheckSum fail, all registers will return to the POR state shown, and an error flag will be written to a status register.

Table 118: Register List

| R/W | Address |     | Register Name      | Description                                       | POR State | EE | Section  |
|-----|---------|-----|--------------------|---|-----------|----|----------|
|     | Hex     | Dec |                    |   |           |    |          |
| R/W | 0x00    | 0   | Command            | Instruction register                              | 0000_0000 |    | 6.2      |
| R/W | 0x01    | 1   | Status1            | Part status, flags                                | 0000_0000 |    | 6.4.5.1  |
| R/W | 0x02    | 2   | Status2            | Part status, flags                                | 0000_0000 |    | 6.4.5.2  |
| R/W | 0x03    | 3   | Chip Address 1     | Upper 8 bits of chip address                      | 0000_0000 | X  | 7.1.1    |
| R/W | 0x04    | 4   | Chip Address 0     | Lower 8 bits of chip address                      | 0000_0000 | X  | 7.1.2    |
| R/W | 0x05    | 5   | RF Divider         | Integer portion of RF frequency                   | 0000_0000 | X  | 6.4.1.1  |
| R/W | 0x06    | 6   | RF Frequency 2     | Upper 8 bits of RF fraction                       | 0000_0000 | X  | 6.4.1.2  |
| R/W | 0x07    | 7   | RF Frequency 1     | Center 8 bits of RF fraction                      | 0000_0000 | X  | 6.4.1.3  |
| R/W | 0x08    | 8   | RF Frequency 0     | Lower 8 bits of RF fraction                       | 0000_0000 | X  | 6.4.1.4  |
| R/W | 0x09    | 9   | Peak Deviation 1   | Upper 8 bits of FM deviation                      | 0000_0000 | X  | 6.4.1.5  |
| R/W | 0x0A    | 10  | Peak Deviation 0   | Lower 8 bits of FM deviation                      | 0000_0000 | X  | 6.4.1.6  |
| R/W | 0x0B    | 11  | Data Rate / Format | Set discrete data rate and encoding option        | 0000_0000 | X  | 7.1.3    |
| R/W | 0x0C    | 12  | General Options A  | General options for interface, POR state, etc.    | 0000_0000 | X  | 7.1.4    |
| R/W | 0x0D    | 13  | General Options B  | General options for interface, POR state, etc.    | 0000_0000 | X  | 7.1.5    |
| R/W | 0x0E    | 14  | RX Config          | Receiver options                                  | 0000_0000 | X  | 6.5.1.1  |
| R/W | 0x0F    | 15  | TX Config          | Transmit options                                  | 0000_0000 | X  | 6.6.1    |
| R/W | 0x10    | 16  | Idle Config        | Idle mode options                                 |           | X  | 6.7.1    |
| R/W | 0x11    | 17  | Sniff Config       | Sniff Mode options                                | 1011_0100 | X  | 6.7.2.1  |
| R/W | 0x12    | 18  | Sniff Interval     | Interval between Sniff cycles                     | 0000_1010 | X  | 6.7.2.2  |
| R/W | 0x13    | 19  | Energy Dwell Time  | Length of time to dwell in Sniff Mode             | 0000_0000 | X  | 6.7.2.3  |
| R/W | 0x14    | 20  | Code Dwell Timer   | Number of bit times to wait for code after energy | 0000_0000 | X  | 6.5.1.5  |
| R/W | 0x15    | 21  | Energy Threshold   | Threshold for wake on RSSI, Sniff and CCA         | 0000_0000 | X  | 6.5.1.2  |
| R/W | 0x16    | 22  | Burst Config       | Burst transmit options                            | 0000_0000 | X  | 6.7.3    |
| R/W | 0x17    | 23  | Burst Interval     | Interval timer for burst transmit                 | 0001_1000 | X  | 6.7.3.1  |
| R/W | 0x18    | 24  | Output Power       | Output power                                      | 0001_0000 | X  | 6.6.2    |
| R/W | 0x19    | 25  | Start of Frame     | Byte used for burst transmit/CDR wake up          | 0001_0000 | X  | 7.1.6    |
| R/W | 0x1A    | 26  | Preamble Length    | Length of CW, or '10' repeated in Burst/TX (BT's) | 0001_0000 | X  | 6.6.3    |
| R/W | 0x1B    | 27  | HK Config          | Housekeeping options register                     |           | X  | 6.7.4.1  |
| R/W | 0x1C    | 28  | HK Interval        | Interval timer for housekeeping                   |           | X  | 6.7.4.2  |
| R/W | 0x1D    | 29  | Slice Threshold    | Energy threshold for AM DAC mode data slice       |           | X  | 6.5.1.4  |
| R/W | 0x1E    | 30  | Filter/Slice       | AM/RSSI filter setting and AM slice mode          |           | X  | 6.5.1.4  |
| R/W | 0x1F    | 31  | CDR Options A      | Clock and data recovery options A                 |           | X  | 6.5.1.5  |
| R/W | 0x20    | 32  | CDR Options B      | Clock and data recovery options B                 | 1000_0000 | X  | 6.5.1.5  |
| R/W | 0x21    | 33  | Crystal Trim       | Crystal trim                                      | 0000_0000 | X  | 6.10.1.1 |
| R/W | 0x22    | 34  | LNA Trim           | LNA input and output matching trim                | 0000_0000 | X  | 6.10.1.2 |
| R/W | 0x23    | 35  | Quick Start Trim   | Quick Start oscillator trim                       | 0000_0000 | X  | 6.10.1.3 |



Table 118: Register List (Continued)

| R/W | Address |     | Register Name      | Description                                  | POR State | EE | Section  |
|-----|---------|-----|--------------------|--|-----------|----|----------|
|     | Hex     | Dec |                    |  |           |    |          |
| R/W | 0x24    | 36  | 10k Osc Trim       | 10kHz oscillator trim                        | 0000_0000 | X  | 6.10.1.4 |
|     | 0x25    | 37  | Analog Trim        | Bandgap and temp sensor trim                 | 0000_0000 | X  | 6.10.1.5 |
|     | 0x26    | 38  | Analog Trim 2      | Capacitance trim                             | 0010_0100 | X  | 6.10.1.6 |
|     | 0x27    | 39  | RF PLL Trim        | PLL calibration storage register             | 0100_0100 | X  | 6.10.1.7 |
|     | 0x28    | 40  | RF PLL Options     | RF PLL options register                      | 0000_0000 | X  | 6.4.1.7  |
|     | 0x29    | 41  | Data Rate 1        | User defined data rate upper bits            | 0000_0000 | X  | 7.1.7    |
|     | 0x2A    | 42  | Data Rate 0        | User defined data rate lower bits            | 0000_0000 | X  | 7.1.8    |
|     | 0x2B    | 43  | PLL Loop Co        | User defined PLL detector bandwidth          | 0000_0000 | X  | 6.5.1.4  |
|     | 0x2C    | 44  | CDR Loop Co        | User defined clock recovery loop             | 0000_0000 | X  | 6.5.1.5  |
|     | 0x2D    | 45  | User Data          | Transmitted on normal interval burst         | 0000_0000 | X  | 6.7.3.2  |
|     | 0x2E    | 46  | User Data          | Transmitted on interrupt triggered Burst     |           | X  | 6.7.3.3  |
|     | 0x2F    | 47  | TargNumWakeUps     | Target number of wake ups for                |           | X  | 6.7.2.4  |
|     | 0x30    | 48  | CRCPoly            | CRC polynomial register                      |           | X  | 7.1.9    |
|     | 0x31    | 49  | DefaultLOP         | Default LOP register                         |           | X  | 7.1.10   |
|     | 0x32    | 50  | Checksum           | EEPROM checksum                              |           | X  | 6.9.1    |
|     | 0x33    | 51  |                    |  |           |    |          |
|     | 0x34    | 52  | Temp ADC           | Storage register for the temp sensor reading | 0000_0000 |    | 6.13.1.1 |
|     | 0x35    | 53  | Battery ADC        | Storage register for the battery reading     | 0000_0000 |    | 6.13.1.2 |
|     | 0x36    | 54  | RSSI ADC           | Storage register for the RSSI reading        | 0000_0000 |    | 6.13.1.3 |
|     | 0x37    | 55  | EXT1 ADC           | Storage register for the EXT1 input          | 0000_0000 |    | 6.13.1.4 |
|     | 0x38    | 56  | EXT2 ADC           | Storage register for the EXT2 input          | 0000_0000 |    | 6.13.1.5 |
|     | 0x39    | 57  | Loop Filter        |  |           |    | 6.4.1.8  |
|     | 0x3A    | 58  | Global Chip ID1    |  |           |    | 7.1.11   |
|     | 0x3B    | 59  | Global Chip ID0    |  |           |    | 7.1.12   |
|     | 0x3C    | 60  | Software State     |  |           |    | 6.4.5.3  |
|     | 0x3D    | 61  |                    |  |           |    |          |
|     | 0x3E    | 62  |                    |  |           |    |          |
|     | 0x3F    | 63  |                    |  |           |    |          |
|     | 0x40    | 64  | Unlock Reg         |  | 1010_0101 |    | 9.1      |
|     | 0x41    | 65  | AMIS ID Code       |  | 0011_0001 |    | 8.4.1    |
|     | 0x42    | 66  | IF Amp Trim A      |  |           |    | 9.2.1    |
|     | 0x43    | 67  | IF Amp Trim B      |  |           |    | 9.2.2    |
|     | 0x44    | 68  | Manual PLL Trim    |  |           |    | 9.2.3    |
|     | 0x45    | 69  | PLL Test Mode      |  |           |    | 9.2.4    |
|     | 0x46    | 70  | PDtestRF           |  |           |    | 9.2.5    |
|     | 0x47    | 71  | Analog Test Mode   |  |           |    | 9.2.6    |
|     | 0x48    | 72  | RFTM               |  |           |    | 9.2.7    |
|     | 0x49    | 73  | Analog Test Mux    |  |           |    | 9.2.8    |
|     | 0x4A    | 74  | RF Test Mux        |  |           |    | 9.2.9    |
|     | 0x4B    | 75  | Digital Test Mux A |  | 0000_0000 |    | 9.2.10   |
|     | 0x4C    | 76  | Digital Test Mux B |  | 0000_0000 |    | 9.2.11   |
|     | 0x4D    | 77  | Digital Test Mux C |  | 0000_0000 |    | 9.2.12   |
|     | 0x4E    | 78  | DTM A              |  | 0000_0000 |    | 9.2.13   |
|     | 0x4F    | 79  | DTM B              |  | 0000_0000 |    | 9.2.14   |
|     | 0x50    | 80  | DTM C              |  | 0000_0000 |    | 9.2.15   |
|     | 0x51    | 81  | DTM D              |  |           |    | 9.2.16   |
|     | 0x52    | 82  | MTM Address        |  |           |    | 9.2.17   |
|     | 0x53    | 83  | MTM Data           |  |           |    | 9.2.18   |
|     | 0x54    | 84  | PLLCalTarget       | PLL calibration target value                 | 0000_0000 |    | 6.10.1.8 |

## 11.0 Applications

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## 12.0 Ordering Information

Table 119: Ordering Information

| Ordering Code           | Product Name   | Package Type | Operating Temperature Range | Industry Application          | Differentiating Feature            | Shipping Configuration           |
|-------------------------|----------------|--------------|-----------------------------|-------------------------------|------------------------------------|----------------------------------|
| 19608-001-XTP (or -XTD) | AMIS-53000-I/A | 32 LQFP      | -40C to 85C                 | Industrial, Automotive, Other | Extra Low Power                    | Tape & Reel (-XTP); Tubes (-XTD) |
| 19608-002-XTP (or -XTD) | AMIS-53000-I/A | 32 LQFP      | -40C to 85C                 | Industrial, Automotive, Other | SPI Interface; Ganged Transceivers | Tape & Reel (-XTP); Tubes (-XTD) |
| 19637-001-XTP (or -XTD) | AMIS-53000-M   | 32 LQFP      | -40C to 85C                 | Medical                       | Extra Low Power                    | Tape & Reel (-XTP); Tubes (-XTD) |
| 19637-002-XTP (or -XTD) | AMIS-53000-M   | 32 LQFP      | -40C to 85C                 | Medical                       | SPI Interface; Ganged Transceivers | Tape & Reel (-XTP); Tubes (-XTD) |

## 13.0 Company or Product Inquiries

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