VISUAL EXCELLENCE

## Features

- broadcast quality 10 / 8-bit 24-tap poly-phase horizontal and vertical scalar for HDTV / SDTV video images
- high performance 2D scaling processor with separate control of horizontal and vertical scaling factors and pan positions
- support for arbitrary video formats up to 2048 by 2048
- support for multiplexed and non-multiplexed Y/C video
- flexible 4:2:2 or 4:4:4 YCbCr or RGB output
- field merge / separation can be inserted / removed from progressive images using interlaced I/O
- double banked control registers for 'on-the-fly' dynamic effects
- external 3:2 / 2:2 pull-down insertion and extraction
- programmable output matrix with 6 dB gain range
- film rate features include 1080p24 and 1080PsF support
- fully programmable colour background generator
- flexible F,V,H output and TRS insertion
- seamless interface to GF9330 de-interlacer
- seamless interface to common SDRAM
- user configuration through dedicated serial interface
- 3.3 V supply


## Description

The GF9320 Scaling Processor offers 10 / 8-bit broadcast quality scaling of video images up to 2048 by 2048 pixels. The GF9320 supports arbitrary display modes to fit custom applications. Dynamic zoom and pan effects allow for a variety of aspect ratio conversion
choices while a programmable colour background generator can be customized to appropriately match the image content. A fully programmable and flexible output matrix allows for colour difference over-sampling, gain and hue controls as well as YCbCr to RGB conversions to power nearly any display device on the market.

The GF9320 also includes a vertical interpolation filter to perform stand alone cost-sensitive de-interlacing. Broadcast quality de-interlacing is offered through a seamless interface to the GF9330 and GF9331 devices.

## Applications

- HDTV Up / Down Converters
- Production Equipment
- Video Walls
- Projection Systems
- Plasma Displays
- LCD TVs
- Home Theatre Systems
- HD DVD Players


## Ordering Information

| Part Number | Package | Temp. Range |
| :---: | :---: | :---: |
| GF9320-CBW | 352 pin TBGA | $0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$ |



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## 1. Pin Description



Figure 1-1: GF9320 Pin Out

## Table 1-1: Pin Descriptions

| Symbol | Pin Grid | Type | Description |
| :---: | :---: | :---: | :---: |
| YIN[9:0] | $\begin{gathered} \mathrm{H} 1, \mathrm{~J} 4, \mathrm{~J} 3, \mathrm{~J} 2, \mathrm{~J} 1, \mathrm{~K} 2, \mathrm{~K} 1, \\ \mathrm{~L} 4, \mathrm{~L} 3, \mathrm{~L} 2 \end{gathered}$ | 1 | 10-bit multiplexed signed luminance / signed offset colour difference data input. <br> Note that either input must include TRS words. |
| CIN[9:0] | $\begin{gathered} \text { L1, M4, M3, M2, M1, N4, } \\ \text { N3, N2, N1, P1 } \end{gathered}$ | 1 | 10-bit signed offset colour difference data input. Note that theinput must include TRS words. |
| CK_IN | G2 | 1 | Input clock. <br> Note that it is equal $Y$ data rate for separate $Y$ and $C$ inputs, and is equal to $2 x$ Y data rate for multiplexed YC input. |
| CK_V | G25 | 1 | Vertical processing clock. <br> Note that it is usually the higher of CK_IN or CK_OUT. |
| CK_OUT | AA26 | 1 | Output clock. |
| FILM_FR | H2 | 1 | Input film sequence reset. |
| OUT_FRST | E1 | 1 | Output frame reset. |
| GOUT[9:0] | $\begin{aligned} & \text { K25, K26, L23, L24, L25, } \\ & \text { L26, M23, M25, M26, N25 } \end{aligned}$ | 0 | 10 / 8-bit unsigned green data output OR <br> 10 / 8-bit unsigned luminance data output OR <br> 10 / 8-bit multiplexed signed luminance / signed offset colour difference data output. |
| BOUT[9:0] | $\begin{aligned} & \text { P26, P25, P24, P23, R26, } \\ & \text { R23, T26, T25, T24, T23 } \end{aligned}$ | 0 | 10 / 8-bit unsigned blue data output OR <br> 10 / 8-bit signed offset (B-Y) data output OR <br> 10/8-bit multiplexed signed offset colour difference data output. |
| ROUT[9:0] | U26, U25, V25, V24, V23, W26, W25, W24, W23, Y26 | 0 | 10 / 8-bit unsigned red data output OR 10 / 8-bit signed offset (R-Y) data output. |
| OUT_CK | J26 | 0 | Output clock timed to clock output data. |
| OUT_F | T3 | 0 | Output format frame / field signal. <br> Note that the output is 3 clocks in advance of output video data. |
| OUT_V | T2 | 0 | Output format vertical signal. <br> Note that the output is 3 clocks in advance of output video data. |
| OUT_H | T1 | 0 | Output format horizontal signal. <br> Note that the output is 3 clocks in advance of output video data. |
| SIF_OUT | V4 | 0 | Serial interface control data out. |
| SIF_IN | V1 | 1 | Serial interface control data in. |
| SIF_CK | V2 | 1 | Serial interface clock. |
| SIF_RST | V3 | 1 | Serial interface reset. |
| RST | F1 | 1 | Power-on reset. |
| DATA_A[19:0] | A15, D14, C14, B14, A14, <br> A13, B13, A12, B12, C12, <br> D12, C11, D11, A10, B10, <br> A9, B9, C9, D9, A8 | I/O | Data bus for memory array A. |

## Table 1-1: Pin Descriptions (Continued)

| Symbol | Pin Grid | Type | Description |
| :---: | :---: | :---: | :---: |
| DATA_B[19:0] | $\begin{aligned} & \text { F25, F24, F23, E26, E25, } \\ & \text { D26, E23, C26, D24, C25, } \\ & \text { B26, B25, A26, A25, B24, } \\ & \text { C23, A24, D22, B23, A23 } \end{aligned}$ | I/O | Data bus for memory array B. |
| DATA_C[19:0] | AE14, AF14, AF13, AE13, AD13, AC13, AF12, AD12, AC12, AF11, AE11, AD11, AC11, AF10, AE10, AF8, AE8, AD8, AC8, AF7 | I/O | Data bus for memory array C. |
| DATA_D[19:0] | AA24, AA23, AB26, AB25, AC26, AC25, AB23, AD26, AC24, AD25, AE26, AF26, AE25, AF25, AE24, AD23, AF24, AC22, AE23, AF23 | 1/0 | Data bus for memory array D. |
| ADDR_A[10:0] | $\begin{gathered} \text { A3, C4, B3, A2, B2, A1, } \\ \text { B1, C2, D3, C1, E4 } \end{gathered}$ | 0 | Address bus for memory array A. |
| BA_A | D5 | 0 | SDRAM bank select for memory array A. |
| ADDR_B[10:0] | D18, C18, B18, B17, A17, D16, C16, B16, A16, D15, C15 | 0 | Address bus for memory array B. |
| BA_B | A19 | 0 | SDRAM bank select pin for memory array B. |
| ADDR_C[10:0] | AD4, AE3, AF2, AF1, AE2, AE1, AD2, AC3, AD1, AB4, AC2 | 0 | Address bus for memory array C. |
| BA_C | AF3 | 0 | SDRAM bank select pin for memory array C. |
| ADDR_D[10:0] | AD18, AE18, AF18, AE17, AF17, AC16, AD16, AE16, AF16, AC15, AD15 | 0 | Address bus for memory array D. |
| BA_D | AC18 | 0 | SDRAM bank select pin for memory array D. |
| CS_A 3 : 0 ] | D6, A5, B5, A4 | 0 | Chip select for memory array A . |
| $\overline{\mathrm{CS}}$ - $[3: 0]$ | A22, D21, C21, B21 | 0 | Chip select for memory array B. |
| CS_C[3:0] | AE5, AF4, AD5, AE4 | 0 | Chip select for memory array C. |
| CS_D[3:0] | AC21, AD21, AE21, AF21 | 0 | Chip select for memory array D. |
| $\overline{\text { RAS_A }}$ | C6 | 0 | Row address strobe for memory array A. |
| $\overline{\text { RAS_B }}$ | B19 | 0 | Row address strobe for memory array B. |
| $\overline{\text { RAS_C }}$ | AD6 | 0 | Row address strobe for memory array C. |
| $\overline{\text { RAS_D }}$ | AE19 | 0 | Row address strobe for memory array D. |
| $\overline{\text { CAS_A }}$ | B6 | 0 | Column address strobe for memory array A. |
| $\overline{\text { CAS_B }}$ | C19 | 0 | Column address strobe for memory array B. |
| $\overline{\text { CAS_C }}$ | AC6 | 0 | Column address strobe for memory array C . |

## Table 1-1: Pin Descriptions (Continued)

| Symbol | Pin Grid | Type | Description |
| :---: | :---: | :---: | :---: |
| $\overline{\text { CAS_D }}$ | AD19 | 0 | Column address strobe for memory array D. |
| $\overline{W E \_A}$ | A6 | 0 | Write enable for memory array A. |
| $\overline{\text { WE_B }}$ | D19 | 0 | Write enable for memory array B. |
| $\overline{W E \_C}$ | AF5 | 0 | Write enable for memory array C. |
| $\overline{\text { WE_D }}$ | AC19 | 0 | Write enable for memory array D . |
| CK_A | D8 | 0 | Clock for memory array A. |
| CK_B | B20 | 0 | Clock for memory array B. |
| CK_C | AF6 | 0 | Clock for memory array C. |
| CK_D | AE20 | 0 | Clock for memory array D. |
| CKEN_A | B8 | 0 | Clock enable for memory array A. |
| CKEN_B | A21 | 0 | Clock enable for memory array B. |
| CKEN_C | AE7 | 0 | Clock enable for memory array C . |
| CKEN_D | AF19 | 0 | Clock enable for memory array D. |
| DATAEN_AB | B22 | 0 | Data enable for memory arrays $A$ and $B$. |
| DATAEN_CD | AE22 | 0 | Data enable for memory arrays $C$ and D. |
| VDD | K3, C10, A11, P3, C13, U3, C17, Y3, C20, AD3, G3, AD7, AC9, AD10, AD14, AD17, C24, AD20, AF22, AD24, E24, Y24, G24, K24, C5, N24, C3, U24, R25, C7, Y1, Y2 | I | 3.3V supply. |
| GND | R24, U23, N23, K23, G26, G23, Y23, D25, AC20, AF20, AC17, D23, AC14, AC10, AC7, AC5, AC4, D20, A20, Y4, D17, B15, U4, D13, P4, B11, D10, K4, D7, G4, B4, D4, AC23, C8, H3, J25, AE6, C22, AE12, M24, AD22, G1, AB24, AA25, R1, R2, R3, R4 | 1 | Device ground. |
| NC | H26, W2, W3, J23, W4, T4, U1, U2, J24, A7, A18, AA3, AA4, W1, AB1, AB2, F26, AC1, AB3, N26, Y25, F2, V26, F3, P2, B7, D2, AD9, E3, H24, AE9, D1, AF9, E2, AE15, AF15, F4, H23, H25, H4, AA1, AA2 |  | No connection. |

## 2. Electrical Characteristics

Table 2-1: Absolute Maximum Ratings

| Parameter | Symbol | Conditions | Rated Value | Units |
| :--- | :---: | :---: | :---: | :---: |
| Power Supply Voltage | $\mathrm{V}_{\mathrm{DD}}$ |  | -0.5 to +4.6 | V |
| Input Voltage | $\mathrm{V}_{\mathrm{I}}$ | $\mathrm{V}_{1}<\mathrm{V}_{\mathrm{DD}}+0.5 \mathrm{~V}$ | -0.5 to +4.6 | V |
| Output Voltage | $\mathrm{V}_{\mathrm{O}}$ | $\mathrm{V}_{\mathrm{O}}<\mathrm{V}_{\mathrm{DD}}+0.5 \mathrm{~V}$ | -0.5 to +4.6 | V |
| Output Current | $\mathrm{I}_{\mathrm{O}}$ | 40 | mA |  |
| Operating Temperature | $\mathrm{T}_{\mathrm{A}}$ | 0 to +70 | ${ }^{\circ} \mathrm{C}$ |  |
| Storage Temperature | $\mathrm{T}_{\text {STG }}$ | -65 to +150 | ${ }^{\circ} \mathrm{C}$ |  |

Table 2-2: Recommend Operating Conditions

| Parameter | Symbol | Conditions | Min. | Typ. | Max. | Units |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Power Supply Voltage | $\mathrm{V}_{\mathrm{DD}}$ |  | 3.0 | 3.3 | 3.6 | V |
| High-Level Input Voltage | $\mathrm{V}_{\mathrm{IH}}$ | TTL Interface | 2.0 | - | $\mathrm{V}_{\mathrm{DD}}$ | V |
| Low-Level Input Voltage | $\mathrm{V}_{\mathrm{IL}}$ | TTL Interface | 0.0 | - | 0.8 | V |
| Positive Trigger Voltage | $\mathrm{V}_{\mathrm{P}}$ |  | 1.5 | - | 2.7 | V |
| Negative Trigger Voltage | $\mathrm{V}_{\mathrm{N}}$ | $\mathrm{V}_{\mathrm{H}}$ |  | 0.6 | - | 1.4 |
| Hysteresis Voltage | $\mathrm{t}_{\mathrm{ri}}$ | Normal Input | 0 | - | V |  |
| Input Rise Time | $\mathrm{t}_{\mathrm{fi}}$ | Normal Input | 0 | - | 200 | ns |
| Input Fall Time |  |  |  |  | 200 | ns |

## Table 2-3: DC Characteristics

| $\mathrm{V}_{\mathbf{D D}}=3.0$ to $3.6 \mathrm{~V}, \mathrm{~T}_{A}=0$ to $70^{\circ} \mathrm{C}$, unless otherwise shown |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Symbol | Conditions | Min. | Typ. | Max. | Units |
| Static Current Consumption | $\mathrm{I}_{\text {DDS }}$ | $\mathrm{V}_{1}=\mathrm{V}_{\mathrm{DD}}$ or GND |  | 10 | 200 | $\mu \mathrm{A}$ |
| Input Leakage Current | 1 | $V_{1}=V_{D D}$ or GND |  | $\pm 10^{-4}$ | $\pm 10$ | $\mu \mathrm{A}$ |
| Low-Level Output Current | $\mathrm{I}_{\mathrm{OL}}$ | $\mathrm{V}_{\mathrm{OL}}=0.4 \mathrm{~V}$ | 12.0 | - | - | mA |
| High-Level Output Current | IOH | $\mathrm{V}_{\mathrm{OH}}=2.4 \mathrm{~V}$ | -2.0 | - | - | mA |
| Low-Level Output Voltage | $\mathrm{V}_{\text {OL }}$ | $\mathrm{l}_{\mathrm{OL}}=0 \mathrm{~mA}$ | - | - | 0.1 | V |
| High-Level Output Voltage | $\mathrm{V}_{\mathrm{OH}}$ | $\mathrm{l}_{\mathrm{OH}}=0 \mathrm{~mA}$ | $V_{D D}-0.1$ | - | - | V |
| Off-State Output Current | loz | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{DD}}$ or GND | - | - | $\pm 10$ | $\mu \mathrm{A}$ |
| Output Short-Circuit Current | los | $\mathrm{V}_{\mathrm{O}}=\mathrm{GND}$ | - | - | -250 | mA |

Table 2-4: Capacitance

| $\mathbf{T}_{\mathbf{A}}=\mathbf{2 5}^{\circ} \mathbf{C} ; \boldsymbol{f = 1 \mathbf { M H z }}$ |  |  |  |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | $\mathbf{S y m b o l}$ | Conditions | $\mathbf{M i n}$ | Typ. | Max. | Units |
| Input Capacitance | $\mathrm{C}_{1}$ | 4.0 | - | 6.4 | pF |  |
| Output Capacitance | $\mathrm{C}_{\mathrm{O}}$ | 4.0 | - | 6.0 | pF |  |
| I/O Capacitance | $\mathrm{C}_{10}$ | 4.0 | - | 6.0 | pF |  |

Table 2-5: Operating Current

| $\mathrm{V}_{\mathrm{DD}}=3.0$ to $3.6 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=0$ to $70^{\circ} \mathrm{C}$, unless otherwise shown |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Parameter | Symbol | Conditions | Min. | Typ. | Max. | Units |
| Operating Current | $I_{\text {cc }}$ | CK_IN @ 90MHz CK_OUT @ 90MHz CK_V @ 88MHz | - | - | 910 | mA |

Table 2-6: AC Characteristics - Setup \& Hold Times

| $V_{\text {DD }}=3.0$ to $3.6 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=0$ to $70^{\circ} \mathrm{C}$, unless otherwise shown |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Signal Name | Setup |  | Hold |  | Reference Clock | Units |
|  | Min. | Max. | Min. | Max. |  |  |
| YIN[9:0] | 2 | - | 1 | - | CK_IN | ns |
| CIN[9:0] | 2 | - | 1 | - | CK_IN | ns |
| FILM_FR | - | - | 2 | - | CK_IN | ns |
| SIF_IN | 1 | - | 2 | - | SIF_CK | ns |
| SIF_RST | 1 | - | 2 | - | SIF_CK | ns |
| OUT_FRST | 1 | - | 2 | - | CK_OUT | ns |
| DATA_A | 2 | - | 0 | - | CK_A | ns |
| DATA_B | 2 | - | 0 | - | CK_B | ns |
| DATA_C | 2 | - | 0 | - | CK_C | ns |
| DATA_D | 2 | - | 0 | - | CK_D | ns |

Table 2-7: AC Characteristics - Pulse Signal

| $V_{\text {D }}=3.0$ to $3.6 \mathrm{~V}, \mathrm{~T}_{\mathbf{A}}=\mathbf{0}$ to $70^{\circ} \mathrm{C}$, unless otherwise shown |  |  |  |
| :---: | :---: | :---: | :---: |
| Signal Name | Pulse Width |  | Units |
|  | Min. | Max. |  |
| $\overline{\mathrm{RST}}$ | $110^{\text {a }}$ |  | $\mu \mathrm{s}$ |

a.The minimum pulse width is for 64 Mb SDRAMs. If 16 Mb is used them $10 \mu$ s width can be used.

Table 2-8: Output Signal Timing Specifications

| Signal Name | CLK to Valid Output Delay |  | Reference Clock |
| :---: | :---: | :---: | :---: |
|  | Min. (ns) | Max. (ns) |  |
| OUT_CK | 1.81 | 3.97 | CK_OUT |
| GOUT[9:0] | 0.15 | 3 | OUT_CK |
| BOUT[9:0] | 0.15 | 3 | OUT_CK |
| ROUT[9:0] | 0.15 | 3 | OUT_CK |
| ADDR_A, $\overline{R A S \_A}, \overline{C A S \_A}, \overline{W E \_A}, \mathrm{CKEN}$ _A, DATA_A | 1.25 | 6 | CK_A |
| ADDR_B, $\overline{R A S \_B}$, $\overline{C A S \_B}, \overline{W E \_B}, \mathrm{CKEN}$-B, DATA_B | 1.25 | 6 | CK_B |
|  | 1.25 | 6 | CK_C |
| ADDR_D, $\overline{\text { RAS_D }}$, $\overline{C A S}$-D,$\overline{\text { WE_D }}$, CKEN_D, DATA_D | 1.25 | 6 | CK_D |

Table 2-9: Clock Frequency

| Clock Name | Frequency |  |
| :--- | :---: | :---: |
|  | Min. (MHz) | Max. (MHz) |
| CK_IN | 1 | 90 |
| CK_V | 1 | 88 |
| CK_OUT | 1 | 90 |
| SIF_CK | - | 90 |

## 3. Detailed Device Description

### 3.1 Device Overview

A system level block diagram is shown in the "Block Diagram" on page 1.
2 D scaling is performed by cascading two 1D-scaling filters.
If the number of horizontal input samples is greater than the number of horizontal output samples (i.e. down sampling), then it is advantageous to perform horizontal resizing first. Otherwise, horizontal resizing is performed last. This minimizes the number of operations required, reduces the intermediate image size and thus lowers the SDRAM requirements.

In addition, the SDRAMs are used for field merge or separation operations to perform simple frame rate conversions (e.g. $30 \leftrightarrow 60$ and $48 \leftrightarrow 60$ ) for film applications. This minimizes the on chip memory required to perform 2D format conversion for low-cost, high-quality format conversion.

The GF9320 has 2 fields / frames of delay depending on the selected operating mode.

Processing is performed simultaneously on 3 fields / frames. Input processing is performed on field / frame N , vertical processing is performed on field / frame ( N 1) and output processing is performed on field / frame ( $\mathrm{N}-2$ ).

The input processor decodes the input TRS to determine input video timing information. An area of the input video is selected according to the downloaded parameters. The input video is resized horizontally if down sampling is indicated.

The video is passed to picture memory control \#1 and stored in SDRAM. Field / frame ( $\mathrm{N}-1$ ) is read out of picture memory \#1, processed vertically, and stored in picture memory \#2. To process the video vertically the read address to picture memory \#1 transposes the video data while the write address to picture memory \#2 transposes the video data back. This transpose operation allows the vertical processing to be done as rows instead of columns.

Field / frame ( $\mathrm{N}-2$ ) is read out of picture memory \#2 and resized horizontally if up sampling is indicated.

The flexible output processor can be selected to perform 4:2:2 to 4:4:4 colour difference over sampling, YCbCr to RGB conversion, colour background insertion and output TRS insertion.

### 3.2 Serial Interface Control

The serial interface download control parameters are grouped into 5 sets as given in Table 3-1: Serial Interface Download Groups. All parameters may be downloaded at once or each set can be downloaded individually. This grouping allows for quick downloading of dynamic parameters (e.g. zoom, pan, gain, etc.) and only requires that the static parameters be downloaded once. Details of individual control parameters are provided in Table 3-2: Serial Interface Download Parameters.

Table 3-1: Serial Interface Download Groups

| Name | CMD ID | No. of Bytes | Number of Bits | Description | Word |
| :---: | :---: | :---: | :---: | :---: | :---: |
| All Parameters ${ }^{\text {a }}$ | $\begin{gathered} 0 \\ 00000000^{\mathrm{b}} \end{gathered}$ | 147 | 1176 | A download of all parameters. | AP[1175:0] |
| I/O Format Parameters | $\begin{gathered} 1 \\ 00100000^{b} \end{gathered}$ | 11 | $\begin{gathered} 7+14+66 \\ +1(\text { fill })=88 \end{gathered}$ | Input and static output parameters. <br> These parameters tend to remain fixed once the input and output format is selected. | IO[87:0] |
| Scaling <br> Parameters | $\begin{gathered} 2 \\ 01000000^{b} \end{gathered}$ | 19 | $149+3($ fill $)=152$ | Resizing parameters. <br> These parameters change with zoom, pan, and crop controls. | $\mathrm{RS}[151: 0]$ |
| Dynamic Output Parameters | $\begin{gathered} 3 \\ 01100000^{b} \end{gathered}$ | 20 | $157+3($ fill $)=160$ | Dynamic output parameters. <br> These parameters change with gain, H position, line advance, etc... | OD[159:0] |
| Horizontal Filter Coefficients | $\begin{gathered} 4 \\ 10000000^{b} \end{gathered}$ | 43 | $344+0($ fill $)=344$ | Horizontal filter. | HF[343:0] |
| Vertical Filter Coefficients | $\begin{gathered} 5 \\ 10100000^{b} \end{gathered}$ | 54 | $428+4($ fill $)=432$ | Vertical filter. | VF[431:0] |

a. The GF9320 download parameters are grouped into 5 sets.
b.Each group will be extended with zeros to make an integer number of bytes. In each group the LSB is sent first. So, for instance, the I/ O format parameter group sends 1 zero followed by the PROC_8_BITS bit followed by the OUT_8_BITS bit. A download of all parameters (CMD ID $=0$ ) sends the word: AP[1175:0] = IO[87:0] | RS[151:0] | OD[159:0] | HF[343:0] | VF[431:0] where "|" represents concatenation.
As with all other words the LSB of AP[1175:0] is sent first. The CMD_ID word is listed above in binary form from MSB to LSB. As with all other words the CMD_ID is sent LSB first. For example, a download of the dynamic output parameters (CMD_ID=3) sends 5 zeros followed by 2 ones followed by 1 zero followed by OD0 followed by OD1 followed by OD2....OD159.

Note that all CMD_IDs have 5 zeros as the 5 least significant bits so that each download command starts with 5 zeros.

Table 3-2: Serial Interface Download Parameters

| Parameter Name | No. Of Bits | Description | Word Position | Used By | Time Frame |
| :---: | :---: | :---: | :---: | :---: | :---: |
| I/O Format Parameters | 88 Total |  |  |  |  |
| Input Format Parameters | 7 Sub-total |  |  |  |  |
| IN_PROGRESSIVE | 1 | Indicates that the input is progressive. <br> 0 - Interlaced <br> 1 - Progressive | IO[87] | Input Control | 0 |
| IN_TOP_ACT_FLD | 1 | Used for interlaced formats only. Indicates which field contains the first active line in a frame. (i.e. which field is on top) <br> 0 - Field 0 is on top <br> 1 - Field 1 is on top | IO[86] | Input Control | 0 |
| IN_TOP_ACT_LONGER | 1 | Used for interlaced formats only. Indicates if the top field is one line longer than the bottom field. <br> 0 - Top and Bottom fields contain the same number of active lines <br> 1 - Top field has one more active line | IO[85] | Input Control | 0 |
| IN_YC_MUXED | 1 | Indicates if the input bus is one 10-bit bus for muxed Y\&C data. <br> 0 - Two 10-bit buses for $Y$ and $C_{b} C_{r}$ <br> $1-Y \& C$ Muxed data on a 10-bit bus | IO[84] | Input Control | 0 |
| IN_FILM_RATE | 2 | Input film frame rate. Used for film inputs only. <br> 00 - Input is from film with $3: 2$ pull-down <br> 01 - Input is from film with 2:2 pull-down <br> 10 - Input is at film rate ( $24 / 25 \mathrm{~Hz}$ ) <br> 11 - Not from film | IO[83:82] | Input Control | 0 |
| IN_REFR_LEFT | 1 | This indicates a left memory array refresh is required and normally indicates that the input is from film. <br> 0 - No refresh <br> 1 - Refresh | IO[81] | Memory Control | 0 |

Table 3-2: Serial Interface Download Parameters (Continued)

| Parameter Name | No. Of Bits | Description | Word Position | Used By | Time Frame |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Memory Configuration | 14 Sub-total |  |  |  |  |
| MEM_CONFIG_LEFT | 2 | Indicates the number of SDRAMs per array in the left bank excluding chips needed for LSBs if necessary. (i.e. 8-bit processing) <br> 00-4 chips <br> 01-3 chips <br> 10-2 chips <br> 11-1 chip | IO[80:79] | Memory Control | 0 |
| MODE_16_LEFT | 1 | Maximum number of left bank SDRAM memory rows used to store a horizontal active line. <br> 0-8 memory rows <br> 1-16 memory rows | IO[78] | Memory Control | 0 |
| MEM_CONFIG_RIGHT | 2 | Indicates the number of SDRAMs per array in the right bank excluding chips needed for LSBs if necessary. (i.e. 8 bit processing) <br> 00-4 chips <br> 01-3 chips <br> 10-2 chips <br> 11-1 chip | IO[77:76] | Memory Control | 0 |
| MODE_16_RIGHT | 1 | Maximum number of right bank SDRAM memory rows used to store a horizontal active line. <br> 0-8 memory rows <br> 1-16 memory rows | IO[75] | Memory Control | 0 |
| PIX2READ | 7 | Number of pixels to be pre-read. Vertical processing requires prereading samples so that no hits occur while processing a column of data. | IO[74:68] | Memory Control | 0 |
| OUT_REFR_RIGHT | 1 | This indicates a right memory array refresh is required and normally indicates that the input is from film. <br> 0 - No refresh <br> 1 - Refresh | IO[67] | Memory Control | 0 |

Table 3-2: Serial Interface Download Parameters (Continued)

| Parameter Name | No. Of Bits | Description | Word Position | Used By | Time Frame |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Static Output Format Parameters | 67 Sub-total |  |  |  |  |
| OUT_HLEN_TOT | 12 | The total number of samples per line. (e.g. 2200) | IO[66:55] | Output <br> Timing | 0 |
| OUT_VLEN_TOT | 12 | The total number of output lines in a frame. (e.g. 1125) | IO[54:43] | Output Timing | 0 |
| OUT_HLEN_ACT | 11 | The number of active samples per line minus 1. (e.g. 1919 implies 1920 active samples) | IO[42:32] | Output <br> Timing | 0 |
| OUT_VLEN_ACT | 11 | The number of active output lines minus 1. (e.g. 1079 implies 1080 active lines) | IO[31:21] | Output Timing | 0 |
| OUT_PROGRESSIVE | 1 | Indicates that the output is progressive. <br> 0 - Interlaced <br> 1 - Progressive | IO[20] | Output <br> Timing / Input Control | 0 |
| OUT_TOP_ACT_FLD | 1 | Used for interlaced formats only. Indicates which field contains the first active line in a frame. (i.e. which field is on top) <br> 0 - Field 0 is on top <br> 1 - Field 1 is on top | IO[19] | Output <br> Timing | 0 |
| OUT_TOP_ACT_LONGER | 1 | Used for interlaced formats only. Indicates if the top field is one line longer than the bottom field. <br> 0 - Both fields have the same number of active lines <br> 1 - Top field has one more active line | IO[18] | Output <br> Timing | 0 |
| OUT_VACT_POS | 8 | The position of the first active output line relative to the start of the frame. <br> For interlaced inputs this implies field 0 . | IO[17:10] | Output Timing | 0 |
| OUT_FLD_LONGER | 1 | Used for interlaced formats only. Indicates which field is longer. Interlaced formats contain an odd number of lines. So one field contains more lines. <br> 0 - Field 0 is longer <br> 1 - Field 1 is longer | IO[9] | Output <br> Timing | 0 |
| OUT_REF | 1 | 0 - Input TRS <br> 1- Output Reset pin on GF9320 (OUT_FRST) | IO[8] | Output <br> Timing | 0 |

Table 3-2: Serial Interface Download Parameters (Continued)

| Parameter Name | No. Of Bits | Description | Word Position | Used By | Time Frame |
| :---: | :---: | :---: | :---: | :---: | :---: |
| OUT_FILM_RATE | 2 | Output film frame rate. <br> 00 - Output has a $3: 2$ pull-down sequence <br> 01 - Output has a 2:2 pull-down sequence <br> 10 - Output is at a film rate $(24 / 25 \mathrm{~Hz})$ <br> 11 - Output is not to a film rate or sequence | IO[7:6] | Output <br> Timing / Input Control | 0 |
| OUT_MODE | 2 | Indicates output port configuration. 00-4:4:4 GBR Triple output 01-4:4:4 YC $\mathrm{Y}_{\mathrm{r}}$ Triple output 10-4:2:2 $\mathrm{YC}_{b} \mathrm{C}_{\mathrm{r}}$ Muxed single output 11-4:2:2 $\mathrm{YC}_{b} \mathrm{C}_{\mathrm{r}}$ Muxed dual output | IO[5:4] | Output | 0 |
| OUT_TRS_ON | 1 | Indicates if TRS is inserted into the output. <br> 0 - TRS not inserted <br> 1 - TRS inserted | IO[3] | Output | 0 |
| OUT_8_BITS | 1 | Indicates that the output is rounded to 8 bits. <br> 0-10-bit output <br> 1-8-bit output | $\mathrm{IO}[2]$ | Output | 0 |
| PROC_8_BITS | 1 | Indicates that H\&V processing is rounded to 8-bits. <br> 0-10-bit processing (Requires LSB memory) <br> 1-8-bit processing | $\mathrm{IO}[1]$ | Int. Filters | 0 |
| IO_FILL | 1 | Not used. | $\mathrm{IO} 0]$ |  |  |

Table 3-2: Serial Interface Download Parameters (Continued)

| Parameter Name | No. Of Bits | Description | Word Position | Used By | Time Frame |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Resizing Parameters | 152 Total |  |  |  |  |
| H_PROC_FIRST | 1 | Indicates horizontal processing is performed first. <br> 0 - Horizontal processing last <br> (H_ZOOM_RATIO < 524288) <br> 1 - Horizontal processing first <br> (H_ZOOM_RATIO >= 524288) | RS[151] | Glue Logic (Mux)/Mem Control | $\begin{aligned} & 1 \\ & 2 \end{aligned}$ |
| H_FLT_DEC | 1 | Horizontal filter decimate. <br> 0 - Non-decimate mode <br> 1 - Decimate mode | RS[150] | H Control / HBL Filter | $\begin{aligned} & 1 / 3 \\ & 1 / 3 \end{aligned}$ |
| H_ZOOM_RATIO | 22 | Horizontal zoom ratio. $\begin{aligned} & \frac{\text { IN_HLIVE } \cdot 524288}{\text { OUT_HLIVE }} \\ & \text { if } \quad \text { H_FLT_DEC }=0 \\ & \frac{\text { IN HLIVE } \cdot 524288}{\text { OUT_HLIVE } \cdot 2} \\ & \text { if } \quad \text { H_FLT_DEC }=1 \end{aligned}$ | RS[149:128] | H Control | 1/3 |
| IN_HSTART_PHASE | 7 | Indicates the starting horizontal phase to be used for resampling. | $\mathrm{RS}[127: 121]$ | H Control | 1/3 |
| IN_HSTART | 11 | Indicates the first sample to be used for resampling. | $\mathrm{RS}[120: 110]$ | Input Timing / H Control / Mem Control | $\begin{gathered} 1 \\ 1 / 3 \\ 2 \end{gathered}$ |
| IN_HSTOP | 11 | Indicates the last sample to be used for resampling. | RS[109:99] | Input Timing <br> / H Control / <br> Mem Control | $\begin{gathered} 1 \\ 1 / 3 \\ 2 \end{gathered}$ |
| OUT_HSTART | 11 | Indicates the placement of the first output sample with live data. <br> This value must be even. | RS[98:88] | Output <br> Timing / H Control / Mem Control | $\begin{gathered} 3 \\ 1 / 3 \\ 2 \end{gathered}$ |
| OUT_HSTOP | 11 | Indicates the placement of the last output sample with live data. <br> This value must be odd. | RS[87:77] | Output <br> Timing / H Control / Mem Control | $\begin{gathered} 3 \\ 1 / 3 \\ 2 \end{gathered}$ |
| V_FLT_DEC | 1 | Vertical filter decimate. <br> 0 - Non-decimate mode <br> 1 - Decimate mode | RS[76] | VBL Filter | 2 |

Table 3-2: Serial Interface Download Parameters (Continued)

| Parameter Name | No. Of Bits | Description | Word Position | Used By | Time Frame |
| :---: | :---: | :---: | :---: | :---: | :---: |
| V_ZOOM_RATIO | 22 | Vertical zoom ratio. | RS[75:54] | V Control | 2 |
|  |  | $\begin{gathered} \frac{\text { IN_VLIVE•524288 }}{\text { OUT_VLIVE }} \\ \text { if } \quad \text { V_FLT_DEC }=0 \\ \frac{\text { IN_VLIVE•524288 }}{\text { OUT_VLIVE } \cdot 2} \\ \text { if } \quad \text { V_FLT_VLIVE }=1 \end{gathered}$ |  |  |  |
| IN_VSTART_PHASE | 7 | Indicates the starting vertical phase to be used for resampling. | RS[53:47] | V Control | 2 |
| IN_VSTART | 11 | Indicates the first line to be used for resampling. | RS[46:36] | Input Timing / V Control | $\begin{aligned} & 1 \\ & 2 \end{aligned}$ |
| IN_VSTOP | 11 | Indicates the last line to be used for resampling. | RS[35:25] | Input Timing / V Control | $\begin{aligned} & 1 \\ & 2 \end{aligned}$ |
| OUT_VSTART | 11 | Indicates the placement of the first output line with live data. | RS[24:14] | Output <br> Timing / V <br> Control | $3$ |
| OUT_VSTOP | 11 | Indicates the placement of the last output line with live data. | RS[13:3] | Output <br> Timing / V <br> Control |  |
| RS_FILL | 3 | Not used. | RS[2:0] |  |  |
| Dynamic Output Parameters ${ }^{\text {a }}$ | 160 Total |  |  |  |  |
| Matrix Coefficients | 117 Sub-to <br> The matrix | efficient format is $\pm 2.10$. (i.e. 1 sign bit, | ger bits and 10 | ctional bits) |  |
| G1 | 13 | Matrix coefficient. $\mathrm{G}=\mathrm{G} 1^{*} \mathrm{Y}+\mathrm{G} 2^{*} \mathrm{C}_{\mathrm{b}}+\mathrm{G} 3^{*} \mathrm{C}_{\mathrm{r}}$ | OD[159:147] | Output | 1 |
| G2 | 13 | Matrix coefficient. $\mathrm{G}=\mathrm{G} 1^{*} \mathrm{Y}+\mathrm{G} 2^{*} \mathrm{C}_{\mathrm{b}}+\mathrm{G} 3^{*} \mathrm{C}_{\mathrm{r}}$ | OD[146:134] | Output | 1 |
| G3 | 13 | Matrix coefficient. $\mathrm{G}=\mathrm{G} 1^{*} \mathrm{Y}+\mathrm{G} 2^{*} \mathrm{C}_{\mathrm{b}}+\mathrm{G} 3^{*} \mathrm{C}_{\mathrm{r}}$ | OD[133:121] | Output | 1 |
| B1 | 13 | Matrix coefficient. $B=B 1^{*} Y+B 2^{*} C_{b}+B 3^{*} C_{r}$ | OD[120:108] | Output | 1 |
| B2 | 13 | Matrix coefficient. $B=B 1^{*} Y+B 2^{*} C_{b}+B 3^{*} C_{r}$ | OD[107:95] | Output | 1 |
| B3 | 13 | Matrix coefficient. $B=B 1^{*} Y+B 2^{*} C_{b}+B 3^{*} C_{r}$ | OD[94:82] | Output | 1 |
| R1 | 13 | Matrix coefficient. $R=R 1^{*} Y+R 2^{*} C_{b}+R 3^{*} C_{r}$ | OD[81:69] | Output | 1 |

Table 3-2: Serial Interface Download Parameters (Continued)

| Parameter Name | No. Of Bits | Description | Word Position | Used By | Time Frame |
| :---: | :---: | :---: | :---: | :---: | :---: |
| R2 | 13 | Matrix coefficient. $R=R 1^{*} Y+R 2^{*} C_{b}+R 3^{*} C_{r}$ | OD[68:56] | Output | 1 |
| R3 | 13 | Matrix coefficient. $R=R 1^{*} Y+R 2^{*} C_{b}+R 3^{*} C_{r}$ | OD[55:43] | Output | 1 |
| Background Colour | 24 Sub-tota |  |  |  |  |
| Y_BKGD | 8 | Background colour for Y . Unsigned integer. | OD[42:35] | Output | 1 |
| CB_BKGD | 8 | Background colour for $\mathrm{C}_{\mathrm{b}}$. Signed integer. | OD[34:27] | Output | 1 |
| CR_BKGD | 8 | Background colour for $\mathrm{C}_{\mathrm{r}}$. Signed integer. | OD[26:19] | Output | 1 |
| Output Timing | 19 Sub-tota |  |  |  |  |
| LINE_ADV | 4 | Line advance with respect to input timing. | OD[18:15] | Output <br> Timing | $0^{\text {a }}$ |
| H_POS | 12 | Horizontal position with respect to input timing. | OD[14:3] | Output <br> Timing | $0^{\text {a }}$ |
| OD_FILL | 3 | Not used. | OD[2:0] |  |  |
| Filter Coefficients | 776 Total |  |  |  |  |
| Horizontal Filter | 344 Sub-to |  |  |  |  |
| H_Y_FLT_COEF <br> HYBANK: <br> 12\|12|11|10|10|9|9|9|9|8|8 = 107 <br> 2 filters * ( 107 bits) $=214$ | 214 | Horizontal Y filter coefficients. <br> Center coefficient is $\pm 1.10$. (i.e. 1 sign bit, 1 integer bit and 10 fractional bits) | $\begin{gathered} \text { HY[343:130] } \\ \left(\text { See footnote }{ }^{\text {b }}\right. \text { ) } \end{gathered}$ | H Y Filter | 1 |
| H_C_FLT_COEF <br> HCBANK: <br> 12\|11|9|9|8|8|8 = 65 <br> 2 filters * ( 65 bits) $=130$ | 130 | Horizontal C filter coefficients. <br> Center coefficient is $\pm 1.10$. (i.e. 1 sign bit, 1 integer bit and 10 fractional bits) | $\begin{gathered} \mathrm{HY}[129: 0] \\ \left(\text { See footnote }{ }^{\text {b }}\right. \text { ) } \end{gathered}$ | H C Filter | 1 |
| Vertical Filter | 432 Sub-to |  |  |  |  |
| V_Y_FLT_COEF <br> HYBANK: <br> 12\|12|11|10|10|9|9|9|9|8|8 = 107 <br> 2 filters * (107 bits) $=214$ | 214 | Vertical Y filter coefficients. <br> Center coefficient is $\pm 1.10$. (i.e. 1 sign bit, 1 integer bit and 10 fractional bits) | VY[431:218] <br> (See footnote ${ }^{\mathrm{c}}$ ) | V Y Filter | 1 |
| V_C_FLT_COEF <br> HCBANK: <br> 12\|12|11|10|10|9|9|9|9|8|8 = 107 <br> 2 filters * ( 107 bits) $=214$ | 214 | Vertical C filter coefficients. <br> Center coefficient is $\pm 1.10$. (i.e. 1 sign bit, 1 integer bit and 10 fractional bits) | VY[217:4] <br> (See footnote ${ }^{\mathrm{c}}$ ) | V C Filter | 1 |
| VYF_FILL | 4 | Not used. | VY[3:0] |  |  |

a.The resizing parameters, dynamic output parameters, and the filter coefficients are adjustable (dynamic). The I/O Format parameters are static, i.e. once an input and output format is selected the I/O format parameters tend to remain fixed.
b.Horizontal Coefficient Download Format: HF[343:0] = HYBANK1[106:0] | HYBANK0[106:0] | HCBANK1[64:0] | HCBANK0[64:0]

In NON-DECIMATE MODE (H_FLT_DEC = 0):
HYBANKO contains the coefficients for a 21-tap symmetric FIR filter and HCBANK0 contains the coefficients for a 13-tap symmetric FIR filter. HYBANK1 and HCBANK1 are not used when H_FLT_DEC is 0 and should contain 0s.
HYBANK0[106:0] = HYTO[11:0] | HYT1[11:0] | HYT2[10:0] | HYT3[9:0] | HYT4[9:0] | HYT5[8:0] | HYT6[8:0] | HYT7[8:0] | HYT8[8:0] |
HYT9[7:0] | HYT10[7:0]
HYBANK1[106:0] = 0
HCBANK0[64:0] = HCTO[11:0] | HCT1[10:0] | HCT2[8:0] | HCT3[8:0] | HCT4[7:0] | HCT5[7:0] | HCT6[7:0]
HCBANK1[64:0] = 0
In Decimate Mode (H_FLT_DEC = 1):
HYBANK0 and HYBANK1 contain the coefficients for a 41-tap symmetric decimation filter while HCBANK0 and HCBANK1 contain the coefficients for a 25-tap symmetric decimation filter. BANK0s contain the center tap (tap 0 ) and all odd taps (e.g. 1, 3, 5...) while the BANK1s contain a zero and all even taps (e.g. 2, 4, 6....).
HYBANKO[106:0] = HYTO[11:0] | HYT1[11:0] | HYT3[10:0] | HYT5[9:0] | HYT7[9:0] | HYT9[8:0] | HYT11[8:0] | HYT13[8:0] | HYT15[8:0] |
HYT17[7:0] | HYT19[7:0]
HYBANK1[106:0] = "000000000000" | HYT2[11:0] | HYT4[10:0] | HYT6[9:0] | HYT8[9:0] | HYT10[8:0] | HYT12[8:0] | HYT14[8:0] |
HYT16[8:0] | HYT18[7:0] | HYT20[7:0]
HCBANK0[64:0] = HCTO[11:0] | HCT1[10:0] | HCT3[8:0] | HCT5[8:0] | HCT7[7:0] | HCT9[7:0] | HCT11[7:0]
HCBANK1[64:0] = "000000000000" | HCT2[10:0] | HCT4[8:0] | HCT6[8:0] | HCT8[7:0] | HCT10[7:0] | HCT12[7:0]
Further information on the horizontal filter coefficients is given in FIR Filter Parameters (3.4.2.2 FIR Filter Parameters).
c.Vertical Coefficient Download Format:

VF[431:0] = VYBANK1[106:0] | VYBANK0[106:0] | VCBANK1[106:0] | VCBANK0[106:0] | "0000"
In Non-decimate Mode (V_FLT_DEC = 0):
VYBANK0 and VCBANK0 both contain the coefficients for a 21-tap symmetric FIR filter. VYBANK1 and VCBANK1 are not used when $V_{-} F L T$ DEC is 0 and should contain 0 s .
VYBANK0[106:0] = VYT0[11:0] | VYT1[11:0] | VYT2[10:0] | VYT3[9:0] | VYT4[9:0] | VYT5[8:0] | VYT6[8:0] | VYT7[8:0] | VYT8[8:0] | VYT9[7:0] | VYT10[7:0]
VYBANK1[106:0] = 0
VCBANK0[106:0] = VCT0[11:0] | VCT1[11:0] | VCT2[10:0] | VCT3[9:0] | VCT4[9:0] | VCT5[8:0] | VCT6[8:0] | VCT7[8:0] | VCT8[8:0] |
VCT9[7:0] | VCT10[7:0]
VCBANK1[106:0] = 0
In Decimate Mode (V_FLT_DEC = 1):
VYBANK0 and VYBANK1 together contain the coefficients for a 41-tap symmetric decimation filter while VCBANK0 and VCBANK1 together also contain the coefficients for a 41-tap symmetric decimation filter. BANK0s contain the center tap (tap 0 ) and all odd taps (e.g. $1,3,5 \ldots .19$ ) while the BANK1s contain a zero and all even taps (e.g. 2, 4, 6...20).
VYBANK0[106:0] = VYT0[11:0] | VYT1[11:0] | VYT3[10:0] | VYT5[9:0] | VYT7[9:0] | VYT9[8:0] | VYT11[8:0] |VYT13[8:0] | VYT15[8:0] | VYT17[7:0] | VYT19[7:0]
VYBANK1[106:0] = "000000000000" | VYT2[11:0] | VYT4[10:0] | VYT6[9:0] | VYT8[9:0] | VYT10[8:0] | VYT12[8:0] | VYT14[8:0] |
VYT16[8:0] | VYT18[7:0] | VYT20[7:0]
VCBANK0[106:0] = VCT0[11:0] | VCT1[11:0] | VCT3[10:0] | VCT5[9:0] | VCT7[9:0] | VCT9[8:0] | VCT11[8:0] | VCT13[8:0] | VCT15[8:0] | VCT17[7:0] | VCT19[7:0]
VCBANK1[106:0] = "000000000000" | VCT2[11:0] | VCT4[10:0] | VCT6[9:0] | VCT8[9:0] | VCT10[8:0] | VCT12[8:0] | VCT14[8:0] | VCT16[8:0] | VCT18[7:0] | VCT20[7:0]

The GF9320 parameters are downloaded using a 3-pin serial interface. The serial interface consists of a clock, data and a reset as shown in Figure 3-1: Serial Interface Download Signal Specification. The serial interface reset (SIF_RST) is provided to re-synchronise the download operation in the event that it is interrupted.


Figure 3-1: Serial Interface Download Signal Specification

### 3.3 Input Processing

The input processor decodes the input TRS from the incoming video stream. This provides input video timing information to the GF9320. An area of the input video data is selected for scaling according to the downloaded parameters (i.e. IN_HSTART, IN_HSTOP, IN_VSTART, and IN_VSTOP). This operation is called the windowing operation. Based on the input field / frame timing and the I/O format parameters a memory enable signal is generated by the input controller. This signal controls the field / frame switching of the SDRAM memory controller. Also, a frame-reset signal is sent to the output controller for use in internal lock mode (OUT_REF=0). The frame reset signal and the field / frame switch point is based on input TRS F-bit in interlaced modes (non-film). Otherwise, (i.e. progressive and all film modes) the frame reset signal and the field / frame switch point is based on one line after input TRS V-bit. Note that in film modes the frame reset signal and field / frame switch point vary according the input and output frame rates.

### 3.4 Scaling Processor

At the heart of the GF9320 is the scaling processor. It is here where the raw input image selected from the input video is translated into a raw output image of selected size according to user controlled scaling parameters. As described in 3.1 Device Overview, general 2D scaling is performed by cascading two 1D-scaling filters. This section describes both the horizontal and vertical scaling filters. A block diagram of the horizontal resizing filter is shown in Figure 3-2: Horizontal Scaling Filter. A block diagram of the vertical resizing filter is shown Figure 3-3: Vertical Scaling Filter.


Figure 3-2: Horizontal Scaling Filter


Figure 3-3: Vertical Scaling Filter

### 3.4.1 Scalar Processing

General 1D scaling is performed by cascading an FIR filter with an interpolation filter. The FIR filter is needed to band limit the input signal when the output Nyquist frequency is less than the input Nyquist frequency. The interpolation filter is used to resample the input signal to the new output rate.

### 3.4.1.1 FIR Filter

The purpose of the FIR filters is to band limit or shape the input signal. Each filter is user programmable, with the coefficients derived depending on the required frequency response. The FIR filter can be used in one of two modes: decimate and non-decimate.

Decimate mode can be used when the output rate is half the input rate. The advantage to using decimate mode is that the number of taps is approximately doubled by using two input clocks to compute one output sample. In non-decimate mode the filter is 21 taps ( 13 for horizontal colour difference due to the 4:2:2 input video structure). In decimate mode the filter is 41 taps ( 25 for horizontal colour difference).

Vertically the same modes are available however due to the 4:2:2 sampling structure both luma and colour difference have equal numbers of taps, i.e. 41 taps for decimate and 21 taps for non-decimate.

The filter operation is described by:

$$
\begin{aligned}
& \sum_{k=1} H Y T\langle 2 k\rangle \cdot\left[H Y_{I}(n-2 k)+H Y_{I}(n+2 k)\right] \\
& k=1 \\
& H C_{O}(n) \cdot 1024=\left\{\begin{array}{c}
k=6 \\
\text { HCT0_0 } H C_{I}(n)+\sum_{-} H C T\left\langle 2 \mathrm{k}-1_{-} \mathrm{k}\right\rangle \cdot\left[H C_{I}(n-k)+H C_{I}(n+k)\right] \quad \text { H_FLT_DEC }=0 \\
k=1 \\
k=10
\end{array}\right. \\
& \text { HCT0_0 } H C_{I}(n)+\sum_{k=1} H C T\left\langle 2 \mathrm{k}-1 \_\mathrm{k}\right\rangle \cdot\left[H C_{I}(n-2 k+1)+H C_{I}(n+2 k-1)\right]+ \\
& \begin{array}{ll}
k=1 \\
k=6
\end{array} \quad \text { H_FLT_DEC }^{2}=1 \\
& \sum H C T\langle 2 k\rangle \cdot\left[H C_{I}(n-2 k)+H C_{I}(n+2 k)\right] \\
& k=1
\end{aligned}
$$

$$
\begin{aligned}
& V Y_{O}(n) \cdot 1024=\left\{\begin{array}{cc}
k=10 \\
& k=1 \\
V Y T 0 \_0 \cdot V Y_{I}(n)+\sum_{-} V Y T\left\langle 2 \mathrm{k}-1_{-} \mathrm{k}\right\rangle \cdot\left[V Y_{I}(n-k)+V Y_{I}(n+k)\right] & \text { V_FLT_DEC }=0 \\
k=10 \\
\text { VYT0_0 } V Y_{I}(n)+\sum_{i} V Y T\left\langle 2 \mathrm{k}-1_{-} \mathrm{k}\right\rangle \cdot\left[V Y_{I}(n-2 k+1)+V Y_{I}(n+2 k-1)\right]+ & \\
k=1 \\
k=10 & \text { V_FLT_DEC }=1
\end{array}\right. \\
& \sum_{k=1} V Y T\langle 2 k\rangle \cdot\left[V Y_{I}(n-2 k)+V Y_{I}(n+2 k)\right] \\
& V C_{O}(n) \cdot 1024=\left\{\begin{array}{c}
k=10 \\
\text { VCT0_0 } \cdot V C_{I}(n)+\sum_{i} V C T\left\langle 2 \mathrm{k}-1_{-} \mathrm{k}\right\rangle \cdot\left[V C_{I}(n-k)+V C_{I}(n+k)\right] \\
\quad k=10 \\
\text { VCT0_0 } V C_{I}(n)+\sum_{-} V C T\left\langle 2 \mathrm{k}-1_{-} \mathrm{k}\right\rangle \cdot\left[V C_{I}(n-2 k+1)+V C_{I}(n+2 k-1)\right]+ \\
k=1 \\
k=10
\end{array}\right. \\
& \sum V C T\langle 2 k\rangle \cdot\left[V C_{I}(n-2 k)+V C_{I}(n+2 k)\right] \\
& k=1
\end{aligned}
$$

where $\mathrm{HY}(\mathrm{n}), \mathrm{HC}_{l}(\mathrm{n}), \mathrm{VY}_{l}(\mathrm{n})$ and $\mathrm{VC}_{l}(\mathrm{n})$ are the FIR filter inputs,
$H Y_{O}(n), \mathrm{HC}_{\mathrm{O}}(\mathrm{n}), \mathrm{VY}_{\mathrm{O}}(\mathrm{n})$ and $\mathrm{VC}_{\mathrm{O}}(\mathrm{n})$ are the FIR filter outputs,
HYT, HCT, VYT, and VCT are the filter coefficients as given in Table 3-3: Horizontal Filter Coefficients and Table 3-4: Vertical Filter Coefficients, and
1024 is the DC gain of the filter.
In non-decimate mode only one bank of coefficients are used (Bank 0), but in decimate mode both banks of coefficients are used (Bank 0 and Bank 1).

### 3.4.1.2 Interpolation Filter

After FIR filtering the video data is passed to the interpolation filter where the rate conversion is performed. The interpolation filter is a polyphase filter that allows the output phase to be adjusted every clock cycle. The interpolation filter contains 128 phases (64 phases for horizontal colour difference). The phase selection allows generation of an output anywhere between two inputs with $1 / 128$ input pixel resolution (1/64 for horizontal colour difference). The scaling control unit takes as input the scaling ratio (input / output), and starting phase (starting position of the first output pixel with respect to the input). With these parameters, the scaling control chooses the correct phasing sequence for the interpolator, determines which input samples should be held and for how long (up sampling), which interpolator outputs should be discarded (down sampling), and generates the new output.

### 3.4.2 Resizing Parameters

In order to understand how to program the GF9320 to perform the necessary conversions an explanation of the window parameters, the zoom parameters and the filter parameters is necessary.

### 3.4.2.1 Window Parameters

Figure 3-4: Input Window Definition - Progressive through Figure 3-7: Output Window Definition - Interlaced show how the GF9320 places a window over the input and output active video. This window is selected by using offsets from the active video area (HSTART, HSTOP, VSTART, VSTOP). Note that VSTART and VSTOP for interlaced video refers to field based offsets. The windowed portion is referred to as the live video and can cover the entire active video or just a portion of it. The size of the windowed portion is HLIVE by VLIVE pixels where:

$$
\begin{aligned}
& \text { HLIVE }=\text { HSTOP }- \text { HSTART }+1 \\
& \text { VLIVE }=\text { VSTOP }- \text { VSTART }+1
\end{aligned}
$$

For interlaced video one field may have one more active line that the other. This means that VLIVE is longer for that field. Also, in interlaced film modes VSTART and VSTOP are still field-based offsets but VLIVE is frame based since the fields are merged and processed as a frame.

The input video window is determined by IN_HSTART, IN_HSTOP, IN_VSTART, and IN_VSTOP. The size of the input windowed portion is IN_HLIVE by IN_VLIVE pixels.

The output video window is determined by OUT_HSTART, OUT_HSTOP, OUT_VSTART, and OUT_VSTOP. The size of the output windowed portion is OUT_HLIVE by OUT_VLIVE pixels.


Figure 3-4: Input Window Definition - Progressive


Figure 3-5: Input Window Definition - Interlaced


Figure 3-6: Output Window Definition - Progressive


Figure 3-7: Output Window Definition - Interlaced

### 3.4.2.2 FIR Filter Parameters

The FIR filter shape is programmable by downloading the filter coefficients. The horizontal filter coefficients and download positions are given in Table 3-3:
Horizontal Filter Coefficients. The vertical filter coefficients are given in Table 3-4: Vertical Filter Coefficients. The overall gain of the FIR filter is 1024 , but the range of coefficients is larger to permit implementation of enhancement filters. Note that the coefficients change meaning depending on the filter structure (i.e. if the filter is in decimate mode or not). The filter structure is determined by the FLT_DEC parameter. If H_FLT_DEC is 1, then the horizontal FIR filter is configured in decimate mode. If H_FLT_DEC is 0 , then the horizontal FIR filter is configured in non-decimate mode. If $V_{-}$FLT_DEC is 1 , then the vertical FIR filter is configured in decimate mode. If V_FLT_DEC is 0 , then the vertical FIR filter is configured in nondecimate mode.

Table 3-3: Horizontal Filter Coefficients

| Parameter | No. of Bits | Range | Word Position |
| :---: | :---: | :---: | :---: |
| twelve_zeros | 12 | [0, 0] | HF[343:332] |
| HYT2 | 12 | [-2048, 2047] | HF[331:320] |
| HYT4 | 11 | [-1024, 1023] | HF[319:309] |
| HYT6 | 10 | [-512, 511] | HF[308:299] |
| HYT8 | 10 | [-512, 511] | HF[298:289] |
| HYT10 | 9 | [-256, 255] | HF[288:280] |
| HYT12 | 9 | [-256, 255] | HF[279:271] |
| HYT14 | 9 | [-256, 255] | HF[270:262] |
| HYT16 | 9 | [-256, 255] | HF[261:253] |
| HYT18 | 8 | [-128, 127] | HF[252:245] |
| HYT20 | 8 | [-128, 127] | HF[244:237] |
| HYTO_0 | 12 | [-2048, 2047] | HF[236:225] |
| HYT1_1 | 12 | [-2048, 2047] | HF[224:213] |
| HYT3_2 | 11 | [-1024, 1023] | HF[212:202] |
| HYT5_3 | 10 | [-512, 511] | HF[201:192] |
| HYT7_4 | 10 | [-512, 511] | HF[191:182] |
| HYT9_5 | 9 | [-256, 255] | HF[181:173] |
| HYT11_6 | 9 | [-256, 255] | HF[172:164] |
| HYT13_7 | 9 | [-256, 255] | HF[163:155] |
| HYT15_8 | 9 | [-256, 255] | HF[154:146] |
| HYT17_9 | 8 | [-128, 127] | HF[145:138] |
| HYT19_10 | 8 | [-128, 127] | HF[137:130] |
| twelve_zeros | 12 | [0, 0] | HF[129:118] |
| HCT2 | 11 | [-1024, 1023] | HF[117:107] |
| HCT4 | 9 | [-256, 255] | HF[106:98] |
| HCT6 | 9 | [-256, 255] | HF[97:89] |
| HCT8 | 8 | [-128, 127] | HF[88:81] |
| HCT10 | 8 | [-128, 127] | HF[80:73] |
| HCT12 | 8 | [-128, 127] | HF[72:65] |
| HCTO_0 | 12 | [-2048, 2047] | HF[64:53] |
| HCT1_1 | 11 | [-1024, 1023] | HF[52:42] |
| HCT3_2 | 9 | [-256, 255] | HF[41:33] |

Table 3-3: Horizontal Filter Coefficients (Continued)

| Parameter | No. of Bits | Range | Word Position |
| :---: | :---: | :---: | :---: |
| HCT5_3 | 9 | $[-256,255]$ | HF[32:24] |
| HCT7_4 | 8 | $[-128,127]$ | HF[23:16] |
| HCT9_5 | 8 | $[-128,127]$ | HF[15:8] |
| HCT11_6 | 8 | $[-128,127]$ | HF[7:0] |

Table 3-4: Vertical Filter Coefficients

| Parameter | No. of Bits | Range | Word Position |
| :---: | :---: | :---: | :---: |
| twelve_zeros | 12 | [0, 0] | VF[431:420] |
| VYT2 | 12 | [-2048, 2047] | VF[419:408] |
| VYT4 | 11 | [-1024, 1023] | VF[407:397] |
| VYT6 | 10 | [-512, 511] | VF[396:387] |
| VYT8 | 10 | [-512, 511] | VF[386:377] |
| VYT10 | 9 | [-256, 255] | VF[376:368] |
| VYT12 | 9 | [-256, 255] | VF[367:359] |
| VYT14 | 9 | [-256, 255] | VF[358:350] |
| VYT16 | 9 | [-256, 255] | VF[349:341] |
| VYT18 | 8 | [-128, 127] | VF[340:333] |
| VYT20 | 8 | [-128, 127] | VF[332:325] |
| VYTO_0 | 12 | [-2048, 2047] | VF[324:313] |
| VYT1_1 | 12 | [-2048, 2047] | VF[312:301] |
| VYT3_2 | 11 | [-1024, 1023] | VF[300:290] |
| VYT5_3 | 10 | [-512, 511] | VF[289:280] |
| VYT7_4 | 10 | [-512, 511] | VF[279:270] |
| VYT9_5 | 9 | [-256, 255] | VF[269:261] |
| VYT11_6 | 9 | [-256, 255] | VF[260:252] |
| VYT13_7 | 9 | [-256, 255] | VF[251:243] |
| VYT15_8 | 9 | [-256, 255] | VF[242:234] |
| VYT17_9 | 8 | [-128, 127] | VF[233:226] |
| VYT19_10 | 8 | [-128, 127] | VF[225:218] |
| twelve_zeros | 12 | [0, 0] | VF[217:206] |
| VCT2 | 12 | [-2048, 2047] | VF[205:194] |
| VCT4 | 11 | [-1024, 1023] | VF[193:183] |

Table 3-4: Vertical Filter Coefficients (Continued)

| Parameter | No. of Bits | Range | Word Position |
| :---: | :---: | :---: | :---: |
| VCT6 | 10 | [-512, 511] | VF[182:173] |
| VCT8 | 10 | [-512, 511] | VF[172:163] |
| VCT10 | 9 | [-256, 255] | VF[162:154] |
| VCT12 | 9 | [-256, 255] | VF[153:145] |
| VCT14 | 9 | [-256, 255] | VF[144:136] |
| VCT16 | 9 | [-256, 255] | VF[135:127] |
| VCT18 | 8 | [-128, 127] | VF[126:119] |
| VCT20 | 8 | [-128, 127] | VF[118:111] |
| VCTO_0 | 12 | [-2048, 2047] | VF[110:99] |
| VCT1_1 | 12 | [-2048, 2047] | VF[98:87] |
| VCT3_2 | 11 | [-1024, 1023] | VF[86:76] |
| VCT5_3 | 10 | [-512, 511] | VF[75:66] |
| VCT7_4 | 10 | [-512, 511] | VF[65:56] |
| VCT9_5 | 9 | [-256, 255] | VF[55:47] |
| VCT11_6 | 9 | [-256, 255] | VF[46:38] |
| VCT13_7 | 9 | [-256, 255] | VF[37:29] |
| VCT15_8 | 9 | [-256, 255] | VF[28:20] |
| VCT17_9 | 8 | [-128, 127] | VF [19:12] |
| VCT19_10 | 8 | [-128, 127] | VF[11:4] |
| fill | 4 | [0, 0] | VF[3:0] |

### 3.4.2.3 Zoom Parameters

The zoom parameters (IN_HSTART_PHASE, IN_VSTART_PHASE, H_ZOOM_RATIO, and V_ZOOM_RATIO) specify the precise conversion from the input live video to the output live video. IN_HSTART_PHASE and IN_VSTART_PHASE allows for starting the interpolator with sub-pixel accuracy. This allows for maintaining the true center of picture when zooming and panning. The zoom ratio is approximately:

$$
\begin{aligned}
& \text { H_ZOOM_RATIO }= \begin{cases}\frac{\text { IN HLIVE } \cdot 524288}{\text { OUT_VLIVE }} & \text { H_FLT_DEC }=0 \\
\frac{\text { IN_HLIVE } \cdot 524288}{\text { OUT_HLIVE } \cdot 2} & \text { H_FLT_DEC }=1\end{cases} \\
& \text { V_ZOOM_RATIO }^{\text {H_L }}= \begin{cases}\frac{\text { IN_VLIVE } \cdot 524288}{\text { OUT_VLIVE }} & \text { V_FLT_DEC }=0 \\
\frac{\text { IN_VLIVE } \cdot 524288}{\text { OUT_VLIVE } \cdot 2} & \text { V_FLT_DEC }=1\end{cases}
\end{aligned}
$$

The above equations hold only approximately because the zoom ratio must be adjusted to maintain the true center of picture.

When using the GF9320 there is a preventable condition whereby certain memory configurations cause artifacts in the output image. It is dependent upon the vertical parameters of the output video and the number of SDRAMS employed in the right memory bank as follows:
OUT_VLIVE modulo $(64$ * N$)=(64$ * N$)-3$ or $(64 * N)-1$
Where N is the number of memories in each array of the right bank, MEM_CONFIG_RIGHT, and
OUT_VLIVE $=$ OUT_VSTOP - OUT_VSTART + 1
Artifacts can be avoided by monitoring for the condition. When detected, add or subtract one (1) from the OUT_VSTOP value while maintaining the condition:
0 <= OUT_VSTART < OUT_VSTOP
Note that both fields must be checked for this condition when the output is interlaced with one field longer.

### 3.4.3 Dynamic Zoom and Pan Considerations

The GF9320 is designed to perform frame accurate zooming and panning. Some of the downloaded zoom and pan parameters are used by multiple blocks within the GF9320. These blocks operate on the video data at different time frames. For instance, the input control block operates on the video data on frame / field (N) while the vertical scaling block operates on the video data on frame / field ( $\mathrm{N}-1$ ). Both these blocks need the IN_VSTART parameter. So, the IN_VSTART parameter must be used by the scaling block one field / frame later than the input block. Registering the IN_VSTART parameter on the field / frame boundary before the scaling block uses it does this.

While most dynamic zoom and pan situations are taken care of automatically by the GF9320, some dynamic zoom and pan conditions require special downloading.

### 3.4.3.1 H_PROC_FIRST Switching

The H_PROC_FIRST download bit is special because it actually changes the configuration of the GF9320. In particular, changing the H_PROC_FIRST bit from 1 to 0 makes the horizontal filter switch from operating on field / frame (N) to operating on field / frame ( $\mathrm{N}-2$ ) and vice versa. Note that changing the H_PROC_FIRST bit from 1 to 0 is changing from down sampling to up sampling. In order to handle this special case smoothly, a 1:1 horizontal zoom factor must be downloaded.

The recommended sequence for switching from H_PROC_FIRST equal to 1 to 0 (i.e. down sampling to up sampling) is:

1. Keep H_PROC_FIRST equal to 1 and download H_ZOOM_RATIO equal to 524,288 (down sampling).
2. Wait at least 2 frames / fields.
3. Change H_PROC_FIRST to 0 and download H_ZOOM_RATIO equal to 524,287 (up sampling).
4. Change to the desired H_ZOOM_RATIO.

The recommended sequence for switching from H_PROC_FIRST equal to 0 to 1 (i.e. up sampling to down sampling) is:

1. Keep H_PROC_FIRST equal to 0 and download a H_ZOOM_RATIO equal to 524,287 (up sampling).
2. Change H_PROC_FIRST to 1 and download H_ZOOM_RATIO equal to 524,288 (down sampling).
3. Change to the desired H_ZOOM_RATIO.

### 3.4.3.2 V_FLT_DEC Switching

The vertical filter operates on field / frame ( $\mathrm{N}-1$ ), but the vertical filter coefficients operate on field / frame ( N ). When the V_FLT_DEC is switched from 0 to 1 or vice versa, the vertical filter coefficients must be delayed by one field / frame so that they operate on the same time frame. This is necessary because the filter coefficients are used differently in decimation mode and a non-decimation filter would be used in decimation mode and vice versa. This would most directly affect the DC gain of the filter that may be perceived as a brightness change in the output video. The horizontal coefficients do not need to be delayed when switching H_FLT_DEC because the horizontal filter and the horizontal coefficients operate on the same frame / field (N). Even though H_FLT_DEC is switched, down sampling (H_PROC_FIRST=1) is indicated. Delaying the vertical filter coefficients may not be necessary depending on the application.

### 3.4.3.3 Pseudo Synchronous Film Mode Conversions

This section applies to any film mode conversion when the input frame rate or the output film rate is $3: 2$ pull-down, but the input rate is not (i.e. $48 \rightarrow 60,24 \rightarrow 60$ ). In these cases the zoom and pan update rate is restricted to every other film frame as shown in Figure 3-8: 24/24/60 Download Restrictions and Figure 3-9: 48/24/60 Download Restrictions. This is because the output circuit must be updated on an output field / frame boundary.


Do not download in the shaded regions (OE_AB=1). If the GF9320 is downloaded in the shaded region, the output circuit will be updated in the middle of an output field/frame and will cause one field/ frame of the output to be invalid.

Figure 3-8: 24/24/60 Download Restrictions

IN_FILM_RATE=1 OUT_FILM_RATE=0
INPUT PROCESSING:


Do not download in the shaded regions (OE_AB=1). If the GF9320 is downloaded in the shaded region, the output circuit will be updated in the middle of an output field/frame and will cause one field/ frame of the output to be invalid.

Figure 3-9: 48/24/60 Download Restrictions

### 3.5 SDRAM Memory Interface

### 3.5.1 Memory Interface Description

To achieve high quality scaling of images in two dimensions, separate processing has to be done in the horizontal and vertical dimensions using one dimensional filter banks. Hence, the input image has to be transposed before and after vertical processing and uses SDRAMs to achieve real-time transposition of digital video images using high quality filters.

The SDRAM controller within the GF9320 acts as the master controller of the memory arrays. To perform a transpose operation the memory controller writes the entire image from one field into the image buffer and then reads it out during the next field. Further, during film mode processing, the controller can put two consecutive image fields together and read them out in the next frame as a single progressive frame. The latter technique is used for processing film material with 3:2 pull-down. We can also separate even and odd fields from a progressive frame to create film material with 3:2 pull-down. The memory organization for transposing images at high data rates is shown in Figure 3-10: Memory Interface.


Figure 3-10: Memory Interface

The memory organization consists of four arrays of memories communicating with the GF9320. Each array can contain anywhere between one to five SDRAMs based on format conversion mode. Memory array A and B compose the left bank while memory array $C$ and $D$ compose the right bank. To achieve high bandwidth, the memory arrays are arranged in an interleaved fashion. That is, when one field in written into memory array $A$, the other field will be read out of memory array $B$. The sequence of read / write operations that takes place in non-film applications is shown in Figure 3-11: Timing Diagram of Data between GF9320 and SDRAMs for non-film Modes.


Figure 3-11: Timing Diagram of Data between GF9320 and SDRAMs for nonfilm Modes

The data from an odd field is written into memory array A during Field3. At the same time data from the previous (even) field will be read out as a transposed image from memory array $B$. The horizontal rows of data read out from memory array $B$ will then be processed (vertical processing) within the GF9320 and written into memory array D. Simultaneously, the vertically processed image data from two fields back which was written into memory array $C$ will be read out. When the image is read out from memory array $C$, it went through another image transposition so that the image is back to its original orientation. Effectively, there is a two field / frame delay when processing non-film material.

The GF9320 experiences significantly more processing time in the vertical processing section due to the bandwidth limitations of the SDRAMs. For some conversions the processing time might exceed the available time. This condition can be circumvented by either increasing the number of memories in the array or by increasing the processing clock rate.

During vertical processing, the GF9320 pre-reads (number of pixels = PIX2READ) into its internal FIFO, before the beginning of every scan line so that it can supply the pixels from the FIFO into the one-dimensional filter in an uninterrupted way. The number of pixels to be pre-read is chosen based on several I/O parameters so that it is high enough to supply data continuously to the filter but low enough to complete the vertical processing in the available time.

The PIX2READ parameter is calculated by:

$$
\operatorname{PIX} 2 \text { READ }=\operatorname{MIN}\left\{\operatorname{MAX}\left(8+\operatorname{CEIL}\left[\frac{\frac{\text { IN_VLEN_ACT }}{2-\text { IN_PROGRESSIVE }}}{64 \times(4-\text { MEM_CONFIG_LEFT })}\right], 10\right), 127\right\}
$$

Where IN_VLEN_ACT is the total number of active lines per frame,
CEIL $(x)$ is the smallest integer larger than x ;

$$
\begin{aligned}
& \operatorname{MAX}(a, b)= \begin{cases}a & \text { if } a \geq b \\
b & \text { if } a<b\end{cases} \\
& \operatorname{MAX}(a, b)= \begin{cases}a & \text { if } a \geq b \\
b & \text { if } a<b\end{cases}
\end{aligned}
$$

80Mbits or $5 \times 16$ Mbit SDRAMs are required to store $2048 \times 2048 \times 20$ bits (maximum image size). As illustrated in Figure 3-12: Architecture of Memory Array with four $1 \mathrm{M} \times 16$ and one $4 \mathrm{M} \times 4$ SDRAMs, the memory array has a 20-bit data bus path, supported by blocks of four $1 \mathrm{M} \times 16$ SDRAMs and one $4 \mathrm{Mx4}$ used in parallel, sharing a common address / control bus. 1Mx16 SDRAMs store the upper significant bits of luminance $\mathrm{Y}[10: 2$ ] and colour difference C[10:2]. 4Mx4 SDRAM stores the lower significant bits $\mathrm{Y}[1: 0]$ and $\mathrm{C}[1: 0]$.


Figure 3-12: Architecture of Memory Array with four 1Mx16 and one 4Mx4 SDRAMs

All elements in the array can be simultaneously selected for command execution by activating the chip select signals or commands can be directed to a particular element in the array by activating the chip select signal for that element and deactivating the chip select signal for the others. Figure 3-13: Memory Array Architecture with Four 4MX16 and One 16MX4 SDRAMs shows the pin connections (and slightly different addressing requirements) needed for more common 64M SDRAMs within a memory array. The number of memories for a given format conversion remains the same independent of memory (16M or 64M) being used.


Figure 3-13: Memory Array Architecture with Four 4MX16 and One 16MX4 SDRAMs

To reduce system cost, the memory array architecture is made scalable. That is, when transposing smaller image sizes or when processing 8-bit images, a lesser number of SDRAMs per memory array are required.

Table 3-5: Minimum SDRAM Configurations for Mode 8 (default mode) shows the memory requirements for various format conversions.

Table 3-5: Minimum SDRAM Configurations for Mode 8 (default mode)

| Image Width (max) x max (Input Image Height, Output Image Height) ${ }^{\text {a }}$ | Number of SDRAMs required / ARRAY |  | Download Parameters |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | [Y, C] 10-bits | [Y, C] 8-bits | MDL ${ }^{\text {b }}$ | MCL ${ }^{\text {c }}$ | MDR ${ }^{\text {d }}$ | MCR ${ }^{\mathbf{e}}$ |
| $2048 \times 2048$ | $\begin{gathered} 4(1 \mathrm{M} \times 16) \text { and } 1 \\ (4 \mathrm{M} \times 4) \end{gathered}$ | 4 (1Mx16) | 0 | 00 | 0 | 00 |
| $2048 \times 1536$ | $\begin{gathered} 3(1 \mathrm{Mx16)} \text { and } 1 \\ (4 \mathrm{M} \times 4) \end{gathered}$ | 3 (1Mx16) | 0 | 01 | 0 | 01 |
| $2048 \times 1024$ | $\begin{gathered} 2(1 \mathrm{M} \times 16) \text { and } 1 \\ (4 \mathrm{Mx4}) \end{gathered}$ | 2 (1Mx16) | 0 | 10 | 0 | 10 |
| $2048 \times 512$ | $\begin{gathered} 1(1 \mathrm{Mx16)} \text { and } 1 \\ (4 \mathrm{Mx4}) \end{gathered}$ | 1 (1Mx16) | 0 | 11 | 0 | 11 |

a. $\max (a, b)=a$ when $a>=b$, else $b$ when $a<b$.
b.MDL stands for the parameter MODE_16_LEFT.
c.MCL stands for the parameter MEM_CONFIG_LEFT.
d.MDR stands for the parameter MODE_16_RIGHT.
e.MCR stands for the parameter MEM_CONFIG_RIGHT.

Figure 3-14: Architecture of Memory Array with Lesser Number of SDRAMs per Array shows the memory array architecture when the number of memories is reduced to 2 SDRAMS and one SDRAM (8 bit processing) per memory array.


Figure 3-14: Architecture of Memory Array with Lesser Number of SDRAMs per Array

### 3.5.2 SDRAM Specifications

The speed grade of the SDRAM is chosen depending on the processing clock frequency. For example, if the processing clock is running at 74.25 MHz , SDRAM with a speed grade of -10 or 100 MHz should be selected.

### 3.5.3 Special Processing

### 3.5.3.1 Model 16

To further decrease the memory requirements at the expense of processing time an additional mode is available. Table 3-6: Minimum SDRAM Configurations for Mode 16 summarizes the memory requirements for various format conversions in this mode.

Table 3-6: Minimum SDRAM Configurations for Mode 16

| Image Width (max) x max (Input Image Height, Output Image Height) ${ }^{\text {a }}$ | Number of SDRAMs required / ARRAY |  |  | Download Parameters |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | [Y, C] 10-bits | [Y, C] 8-bits | MDL ${ }^{\text {b }}$ | MCL ${ }^{\text {c }}$ | MDR ${ }^{\text {d }}$ | MCR ${ }^{\text {e }}$ |
| $1024 \times 2048$ | $\begin{aligned} & 2(1 \mathrm{M} \times 16) \text { and } 1 \\ & (4 \mathrm{M} \times 4) \end{aligned}$ | 2 (1Mx16) | 1 | 10 | 1 | 10 |
| $1024 \times 1024$ | $\begin{gathered} 1(1 \mathrm{M} \times 16) \text { and } 1 \\ (4 \mathrm{Mx4}) \end{gathered}$ | 1 (1Mx16) | 1 | 11 | 1 | 11 |

a.max $(\mathrm{a}, \mathrm{b})=\mathrm{a}$ when $\mathrm{a}>=\mathrm{b}$, else b when $\mathrm{a}<\mathrm{b}$.
b.MDL stands for the parameter MODE_16_LEFT.
c.MCL stands for the parameter MEM_CONFIG_LEFT.
d.MDR stands for the parameter MODE_16_RIGHT.
e.MCR stands for the parameter MEM_CONFIG_RIGHT.

### 3.5.4 Film Processing

During film processing the GF9320 uses the external signals FILM_FR and OUT_FRST to encode or decode the 3:2 pull-down sequence. The timing of these signals for different modes (film and non-film) is shown in Figure 3-15: 60/60/60 Processing through Figure 3-40: $24 \mathrm{p} / 24 \mathrm{p} / 48$ p Processing. When the input video is from a film material with 3:2 pull-down, the GF9320 processes the image vertically after combining the even and odd fields to achieve better quality resizing. Duplicate fields in the input sequence are rejected by not writing into the memory. Note that in film modes memory switching does not occur at every field / frame boundary. It depends on the input and output film formats. For example, when the input is video with 3:2 pull-down, the left array of memories switch only after even and odd fields have been put together. The switching point is shown in the timing diagram by DATAEN_AB and DATAEN_CD signals that are, respectively, the output enable signals for left and right arrays. The GF9320 achieves 3:2 pull-down at the output by separately reading out the even and odd fields.

The film sequences shown in Figure 3-15: 60/60/60 Processing through Figure 340: $24 p / 24 p / 48 p$ Processing are not the only film frame sequences that the GF9320 can generate. Note that an $X$ through an input video sequence through Figure 3-16: 30i/24p/24i Processing to Figure 3-34: 48p/24p/24p Processing denotes a discarded frame when performing 3:2 pulldown compensation. Other input / output film sequences are possible. The input control uses the rising edge of FILM_FR to set the input film sequence and the film frame reset sent to the output controller. The first TRS V-bit after the rising edge of FILM_FR marks the beginning of a 3:2 (starting with 3 ) or $2: 2$ film sequences.

IN_FILM_RATE=3

## INPUT PROCESSING:



VERTICAL PROCESSING:


OUTPUT PROCESSING:


Figure 3-15: 60/60/60 Processing

IN_FILM_RATE=0
OUT_FILM_RATE=1
IN_PROGRESSIVE=0 OUT_PROGRESSIVE=0
INPUT PROCESSING:


VERTICAL PROCESSING:


OUTPUT PROCESSING:


Figure 3-16: 30i/24p/24i Processing

IN_FILM_RATE=0
IN_PROGRESSIVE=1 OUT_PROGRESSIVE=1

INPUT PROCESSING:


## VERTICAL PROCESSING:



OUTPUT PROCESSING:


Figure 3-17: 60p/24p/24p Processing

IN_FILM_RATE=0
OUT_FILM_RATE=2
IN_PROGRESSIVE=0 OUT_PROGRESSIVE=1

INPUT PROCESSING:


## VERTICAL PROCESSING:



OUTPUT PROCESSING:


Figure 3-18: 30i/24p/24p Processing


VERTICAL PROCESSING:


OUTPUT PROCESSING:


Figure 3-19: 60p/24p/30i Processing


VERTICAL PROCESSING:


OUTPUT PROCESSING:

*** Film sequence at the output is same as the input. (OUT_REF =1)
Figure 3-20: 60p/24p/30i (OUT_REF=1) Processing


VERTICAL PROCESSING:


## OUTPUT PROCESSING:



OUT_FRST

VIDEO SEQUENCE
**FILM SEQUENCE

** Film sequence is not maintained in the output.
Figure 3-21: 30i/24p/30i Processing

***Film sequence at the output is same as the input. (OUT_REF=1)
Figure 3-22: 30i/24p/30i (OUT_REF=1) Processing


INPUT PROCESSING:


VERTICAL PROCESSING:


OUTPUT PROCESSING:


Figure 3-23: 30i/24p/48p Processing

IN_FILM_RATE=0
OUT_FILM_RATE=1 IN_PROGRESSIVE=1
INPUT PROCESSING:


VERTICAL PROCESSING:


OUTPUT PROCESSING:


Figure 3-24: 60p/24p/48p Processing


INPUT PROCESSING:


OUTPUT PROCESSING:


Figure 3-25: 60p/24p/24i Processing


Figure 3-26: 60p/24p/60p Processing


Figure 3-27: 30i/24p/60p Processing


VERTICAL PROCESSING:


OUTPUT PROCESSING:

FIELD/FRAME
PULSE
OUT_FRST

VIDEO SEQUENCE FILM SEQUENCE


Figure 3-28: 24i/24p/30i Processing
 INPUT PROCESSING:


VERTICAL PROCESSING:


OUTPUT PROCESSING:


Figure 3-29: 48p/24p/30i Processing

IN_FILM_RATE=1
OUT_FILM_RATE=0 IN_PROGRESSIVE=1
INPUT PROCESSING:


VERTICAL PROCESSING:


OUTPUT PROCESSING:


Figure 3-30: 48p/24p/60p Processing


INPUT PROCESSING:


Figure 3-31: 24i/24p/60p Processing


Figure 3-32: 24i/24p/24i Processing


Figure 3-33: 48p/24p/24i Processing


Figure 3-34: 48p/24p/24p Processing


Figure 3-35: 24i/24p/24p Processing


Figure 3-36: 24p/24p/30i Processing


Figure 3-37: 24p/24p/24i Processing


Figure 3-38: 24p/24p/24p Processing


Figure 3-39: 24p/24p/60p Processing


Figure 3-40: 24p/24p/48p Processing

During film processing there is a possibility that for some conversions the GF9320 could violate the refresh period ( 64 ms ) of the SDRAM. If a violation is found (IN_REFR_LEFT or OUT_REFR_RIGHT = ' 1 '), then the appropriate (left / right) refresh bit should be activated in the download stream of parameters to the GF9320. Alternatively, Table 3-7: Input and Output Formats Requiring Refresh can be used to determine which input and output formats require refresh bits to be active.

Table 3-7: Input and Output Formats Requiring Refresh
Input Format IN_REFR_LEFT=1 Output Format OUT_REFR_RIGHT=1

| $24 p / 25 p$ | $24 p / 25 p$ |
| :---: | :---: |
| $48 p / 50 p$ |  |

### 3.5.5 Processing Delay

Processing delay for video through the GF9320 depends on the conversion. Table 3-8: Processing Delay for Various Conversions shows the processing delay for different film and non-film modes.

Table 3-8: Processing Delay for Various Conversions

## Conversion

Delay (Input Frames / Fields)
[frame / field modes - 60Hz V processing]
Note: All other frame rates are identical with appropriate time scaling

| $30 i \rightarrow 30 i$ | $1 / 30$ seconds (2 fields) |
| :---: | :---: |
| $30 i \rightarrow 60 p$ | $1 / 30$ seconds $(2$ fields $)$ |
| $60 p \rightarrow 30 i$ | $1 / 30$ seconds (2 frames) |
| $60 p \rightarrow 60 p$ | $1 / 30$ seconds $(2$ frames $)$ |

[2:2 modes - 30Hz V processing]

| $30 i \rightarrow 30 p$ | $1 / 15$ seconds (4 fields) |
| ---: | :---: |
| $60 p \rightarrow 30 p$ | $1 / 15$ seconds (4 frames) |
| $30 p \rightarrow 30 i$ | $1 / 15$ seconds $(2$ frames $)$ |
| $30 p \rightarrow 60 p$ | $1 / 15$ seconds (2 frames) |
| $30 i \rightarrow 30 i$ | $1 / 15$ seconds $(4$ fields $)$ |
| $30 i \rightarrow 60 p$ | $1 / 15$ seconds $(4$ fields $)$ |

[3:2 modes - 24Hz V processing]

| $30 \mathrm{i} \rightarrow 30 \mathrm{i}$ | $1 / 10-1 / 12$ seconds $(5-6$ fields $)$ |
| ---: | ---: |
| $30 \mathrm{i} \rightarrow 60 \mathrm{p}$ | $1 / 10-1 / 12$ seconds $(5-6$ fields $)$ |
| $30 \mathrm{i} \rightarrow 24 \mathrm{p}$ | $3 / 40-1 / 12$ seconds $(4.5-5$ fields $)$ |
| $30 \mathrm{i} \rightarrow 24 \mathrm{i}$ | $3 / 40-1 / 12$ seconds $(4.5-5$ fields $)$ |
| $60 p \rightarrow 24 p$ | $3 / 40-1 / 12$ seconds $(4.5-5$ fields $)$ |
| $60 p \rightarrow 24 i$ | $3 / 40-1 / 12$ seconds $(4.5-5$ fields $)$ |
| $24 i \rightarrow 30 i$ | $3 / 40-1 / 12$ seconds $(3.6-4$ fields $)$ |
| $24 i \rightarrow 60 p$ | $3 / 40-1 / 12$ seconds $(3.6-4$ fields $)$ |
| $24 p \rightarrow 30 i$ | $3 / 40-1 / 12$ seconds $(1.8-2$ frames $)$ |
| $24 p \rightarrow 60 p$ | $3 / 40-1 / 12$ seconds $(1.8-2$ frames $)$ |

### 3.5.6 Pin Descriptions

The GF9320 uses the transpose memory bus interface signals to communicate with external memory (SDRAMs). The GF9320 is the master device on the bus interface and it controls the timing of the address and data flow. Each signal in the bus interface is described as follows:

### 3.5.6.1 Address Bus

## ADDR_A[10:0], ADDR_B[10:0], ADDR_C[10:0], ADDR_D[10:0]

The address bus is shared by all the memories in the array. The address bus bit BA_A (Bank Select) selects which bank is to be active in memory array. BA_A low selects bank $A$ and $B A \_A$ high selects bank $B$ within the memory. During a bank activate command cycle, ADDR_A[10:0] defines the row address when sampled at the rising clock edge. During a read / write cycle, ADDR_A[9:0] defines the column address when sampled at the rising clock edge. In addition to the column address ADDR_A10 is used to invoke auto-precharge operation.

Similarly, ADDR_B[10:0], ADDR_C[10:0], ADDR_D[10:0] form the address bus of memory arrays $B, C$ and $D$ respectively.

### 3.5.6.2 Data Bus

## DATA_A[19:0], DATA_B[19:0], DATA_C[19:0], DATA_D[19:0]

The data bus is bi-directional. Valid data is driven on the data bus by the GF9320 during write cycle, which is accepted back by the GF9320 during the read cycles. These cycles involve transfers of bursts of data between the SDRAM core and registers of GF9320. Luminance data Y[9:2] are available on DATA_A/B/C/ D [19:12] while least significant bits $\mathrm{Y}[1: 0]$ are available on DATA_A/B/C/D[3:2]. Colour difference data $C[9: 2$ ] are available on DATA_A/B/C/D[11:4] while least significant bits $C[1: 0]$ are available on DATA_A/B/C/D[1:0].

### 3.5.6.3 Command Bus

[ $\left.\left.\overline{\text { RAS_A }^{\prime}}, \overline{\text { CAS_A }^{\prime}}, \overline{\text { WE_A }}\right], \overline{\text { RAS_B }}, \overline{\text { CAS_B }}, \overline{\text { WE_B }}\right],\left[\overline{R A S \_C}, \overline{\text { CAS_C }}\right.$, WE_C], [RAS_D, CAS_D, WE_D]

These bus signals are asserted by the GF9320 when commands have to be executed on the SDRAM memory array A. Similarly, [ $\left.\overline{R A S} B, \overline{C A S \_B}, \overline{W E} B\right]$, $[\overline{R A S} C, \overline{C A S} C, \overline{W E} C]$ and $\left[\overline{R A S \_D}, \overline{C A S \_D}, \overline{W E \_D}\right]$ are asserted to execute commands on memory array $B, C$ and $D$ respectively. These signals are considered valid only if the respective $\overline{\mathrm{CS}}$ pin is low during the active edge of the clock.

## CKEN_A, CKEN_B, CKEN_C, CKEN_D

CKEN_A, CKEN_B, CKEN_C and CKEN_D are used to drive memory arrays A, B, $C$ and $D$ respectively. CKEN input suspends data (i.e. read data remains valid and write data is inhibited) during an active read or write. The GF9320 activates CKEN_A and CKEN_B signals during field / frame write cycle to drop pixels. CKEN_C and CKEN_D are activated during field / frame read cycle to hold pixel values. These signals are considered valid only if the respective $\overline{\mathrm{CS}}$ pin is low during the active edge of the clock.

## 

The CS_A $[3: 0]$ signals from the GF9320 allows selection of individual or multiple SDRAMs within the memory array A. The appropriate SDRAM(s) is selected when the respective $\overline{\mathrm{CS}} \mathrm{A}[3: 0]$ pin is active low on the rising edge of clock. $\overline{\mathrm{CS}} \mathrm{B}[3: 0]$, $\overline{C S}$ C $[3: 0]$ and CS_D[3:0] select SDRAMs within memory arrays B, C and D respectively.

## DATAEN_AB, DATAEN_CD

These signals are driven by the GF9320 only during start-up to prevent data contention. When sampled high, it places the data bus buffers within the SDRAM in a high impedance state. After successful initialization, DATAEN_AB and DATAEN_CD stay low until the next power-up reset. DATAEN_AB is shared by memories in banks $A$ and $B$, while DATAEN_CD is shared by memories in banks C and D.

## CK_A, CK_B, CK_C, CK_D

CK_A, CK_B, CK_C and CK_D are clock signals, which drive the SDRAMs clock pins in memory array $\mathrm{A}, \mathrm{B}, \mathrm{C}$ and D respectively.

### 3.6 Output Processor

A block diagram of the output processor is shown in Figure 3-41: Output Processor Block Diagram. The output processor consists of three major functions:

1. Colour difference over-sample
2. Matrix conversion
3. Output format

The colour difference over-sample function is necessary for colour matrix conversion and to provide a 4:4:4 output. The colour difference over-sample block also performs colour background insertion and horizontal edge shaping. Horizontal edge shaping is done to eliminate overshoot on edges when the scaled output does not fill the entire output raster. That is, when OUT_HSTART is greater that 0 for left edge shaping and when OUT_HSTOP is less than OUT_HLEN_ACT for right edge shaping. A programmable flat matte colour background is inserted into the output non-live video. Note that the colour background is inserted prior to the matrix conversion. This means that the downloaded background colour is in the input colour space coordinates.


Figure 3-41: Output Processor Block Diagram

The matrix block performs the following operations:

$$
\begin{gathered}
\mathrm{YMAT}_{\mathrm{OUT}}=\frac{\mathrm{G} 1 \cdot\left(\mathrm{YMAT}_{\mathrm{IN}}-64\right)+\mathrm{G} 2 \cdot \mathrm{CBMAT}_{\mathrm{IN}}+\mathrm{G} 3 \cdot \mathrm{CRMAT}_{\mathrm{IN}}}{1024}+64 \\
\operatorname{PBMAT}_{\mathrm{OUT}}=\frac{\mathrm{B} 1 \cdot\left(\mathrm{YMAT}_{\mathrm{IN}}-64\right)+\mathrm{B} 2 \cdot \mathrm{CBMAT}_{\mathrm{IN}}+\mathrm{B} 3 \cdot \mathrm{CRMAT}_{\mathrm{IN}}}{1024}+\mathrm{C}_{\mathrm{OFFSET}}
\end{gathered}
$$

$$
\mathrm{PRMAT}_{\mathrm{OUT}}=\frac{\mathrm{R} 1 \cdot\left(\mathrm{YMAT}_{\mathrm{IN}}-64\right)+\mathrm{R} 2 \cdot \mathrm{CBMAT}_{\mathrm{IN}}+\mathrm{R} 3 \cdot \mathrm{CRMAT}_{\mathrm{IN}}}{1024}+\mathrm{C}_{\mathrm{OFFSET}}
$$

where $\mathrm{YMAT}_{I N}$, CBMAT $_{I N}$ and CRMAT ${ }_{I N}$ are the inputs to the matrix;
YMAT $_{\text {OUT }}$, CBMAT $_{\text {OUT }}$ and CRMAT OUT are the outputs of the matrix; and G1, G2, G3, B1, B2, B3, R1, R2 and R3 are the matrix coefficients;
$\mathrm{C}_{\text {OFFSET }}$ is given by

$$
\mathrm{C}_{\text {OFFSET }}= \begin{cases}64 & \text { OUT_MODE }=0 \\ 512 & \text { otherwise }\end{cases}
$$

and 1024 is the gain of the matrix.
The matrix coefficients provide +6 dB of range for gain adjustments. The $\mathrm{C}_{\mathrm{b}}$ and $\mathrm{C}_{\mathrm{r}}$ components at the input to the matrix are in 2's complement format. The B and R components at the output of the matrix are unsigned in GBR output mode (OUT_MODE=0) and are offset binary in $\mathrm{YC}_{b} \mathrm{C}_{\mathrm{r}}$ output mode (OUT_MODE=1, 2 or 3). The matrix coefficients are completely programmable and are downloaded as described in the 3.2 Serial Interface Control.

The output format block formats the data into one, two or three channels according to the OUT_MODE parameter and inserts the output format TRS. If TRS is enabled, the data is clipped to 4 and 1019 for 10-bits or 1 and 254 for 8 -bits.

### 3.7 Output Timing Control

The output timing and control block determines the output video data timing. This block contains horizontal and vertical counters based on the output format parameters. The output timing is adjusted relative to the reference by using the LINE_ADV and H_POS parameters. The output reference is either the input TRS (if OUT_REF = 0) or the OUT_FRST pin on the GF9320 (if OUT_REF = 1). This provides for internal or external lock capability. The LINE_ADV parameter advances the output video data by LINE_ADV output lines. The H_POS parameter delays the output video data by $H_{-}$POS samples. The range of $H_{-} P O S$ is one output line or OUT_HLEN_TOT samples.

Only limited ranges of input / output timing relationships are available by using the GF9320. In general, there are 2 fields / frames of delay through the GF9320. It is not possible for the GF9320 to have an output timing relationship such that the last active output line occurs after the SDRAM field / frame switch point.

## 4. Package Dimensions



Figure 4-1: Package Dimensions

## 5. Revision History

| Version | ECR | Date | Changes and / or Modifications |
| :---: | :---: | :---: | :---: |
| 7 | 134925 | November 2004 | Corrections to address pins: ADDR_A[10:0], ADDR_B[10:0], ADDR_C[10:0], ADDR_D[10:0], BA_A, BA_B, BA_C and BA_D. |
| 6 | 133502 | June 2004 | Changed BOUT description. Changed template. |
| 5 |  | November 2002 | Add OUT_VLIVE issue and workaround - Change $\overline{\text { RST }}, \overline{\mathrm{SIF}}$ RST, $\overline{\mathrm{CS}}, \overline{\mathrm{RAS}}, \overline{\mathrm{CAS}}$ and $\overline{\mathrm{WE}}$ pin descriptions to active low. Make same changes throughout DS. -Make 60I->30i changes for consistency. |
| 4 |  | May 2002 | Updating GF9320. |
| 3 |  | September 2001 | Remove all "Preliminary \& Confidential" water marks \& references with the document. |
| 2 |  | July 2001 | Correction required for both Figure 15 and the table titled "Output Signal Timing Specification". |
| 1 |  | June 2001 | Correction to figure on page 2 \& other improvements. |
| 0 |  | June 2001 | Creating Preliminary Data Sheet. |



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