



ELECTROSTATIC SENSITIVE DEVICE
OBSERVE HANDLING PRECAUTIONS

RA45H8994M1

RoHS Compliance, 896-941MHz 45W 12.8V, 2 Stage Amp. For MOBILE RADIO

DESCRIPTION

The RA45H8994M1 is a 45-watt RF MOSFET Amplifier Module for 12.8-volt mobile radios that operate in the 896- to 941-MHz range.

The battery can be connected directly to the drain of the enhancement-mode MOSFET transistors. Without the gate voltage 1 and the gate voltage 2 ($V_{GG1}=V_{GG2}=0V$), only a small leakage current flows into the drain and the nominal output signal ($P_{out}=45W$) attenuates up to 60 dB. When fixed i.e. 3.4V, is supplied to the gate voltage 1, the output power and the drain current increase as the gate voltage 2 increases. The output power and the drain current increase substantially with the gate voltage 2 around 0V (minimum) under the condition when the gate voltage 1 is kept in 3.4V. The nominal output power becomes available at the state that V_{GG2} is 4V (typical) and 5V (maximum). At this point, V_{GG1} has to be kept in 3.4V. At $V_{GG1}=3.4V$ & $V_{GG2}=5V$, the typical gate currents are 0.4mA. This module is designed for non-linear FM modulation, but may also be used for linear modulation by setting the drain quiescent current with the gate voltages and controlling the output power with the input power.

FEATURES

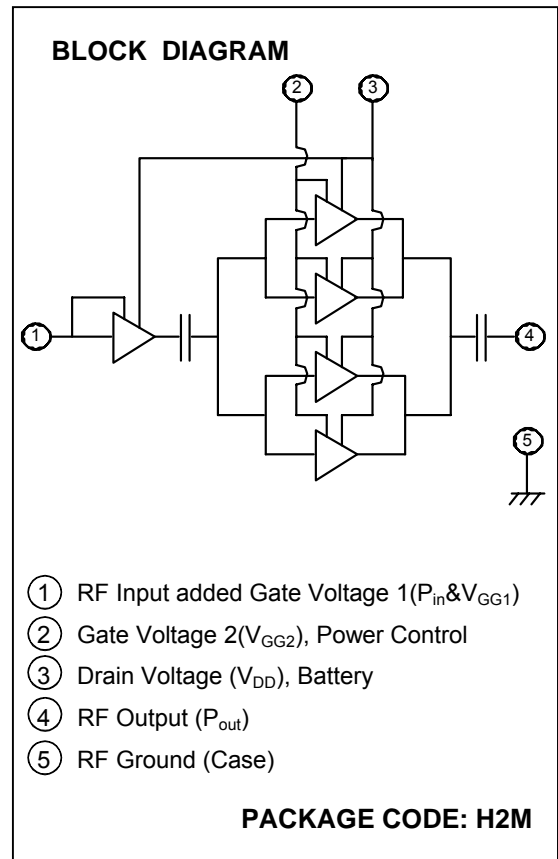
- Enhancement-Mode MOSFET Transistors ($I_{DD}=0$ @ $V_{DD}=12.8V$, $V_{GG1}=V_{GG2}=0V$)
- $P_{out}>45W$, $\eta_T>33\%$ @ $V_{DD}=12.8V$, $V_{GG1}=3.4V$, $V_{GG2}=5V$, $P_{in}=50mW$
- Broadband Frequency Range: 896-941MHz
- Metal cap structure that makes the improvements of RF radiation simple
- Low-Power Control Current $I_{GG1}+I_{GG2}=0.4mA$ (typ) @ $V_{GG1}=3.4V$, $V_{GG2}=5V$
- Module Size: 67 x 18 x 9.9 mm
- Linear operation is possible by setting the quiescent drain current with the gate voltages and controlling the output power with the input power.

RoHS COMPLIANCE

- RA45H8994M1 is a RoHS compliant product.
- RoHS compliance is indicate by the letter "G" after the Lot Marking.
- This product include the lead in the Glass of electronic parts and the lead in electronic Ceramic parts.
How ever, it is applicable to the following exceptions of RoHS Directions.
 - 1.Lead in the Glass of a cathode-ray tube, electronic parts, and fluorescent tubes.
 - 2.Lead in electronic Ceramic parts.

ORDERING INFORMATION:

ORDER NUMBER	SUPPLY FORM
RA45H8994M1-101	Antistatic tray, 10 modules/tray





MAXIMUM RATINGS ($T_{case}=+25^{\circ}C$, $Z_G=Z_L=50\Omega$, unless otherwise specified)

SYMBOL	PARAMETER	CONDITIONS	RATING	UNIT
V_{DD}	Drain Voltage	$V_{GG1}=3.4V \pm 7\%$, $V_{GG2}<5V$, $P_{in}=0W$	17	V
V_{GG1}	Gate Voltage 1	$V_{GG2}<5V$, $V_{DD}<12.8V$, $P_{in}=50mW$	4.5	V
V_{GG2}	Gate Voltage 2	$V_{GG1}=3.4V \pm 7\%$, $V_{DD}<12.8V$, $P_{in}=50mW$	6	V
P_{in}	Input Power	$f=896-941MHz$, $V_{GG1}=3.4V \pm 7\%$, $V_{GG2}<5V$	100	mW
P_{out}	Output Power		60	W
$T_{case(OP)}$	Operation Case Temperature Range		-30 to +100	$^{\circ}C$
T_{stg}	Storage Temperature Range		-40 to +110	$^{\circ}C$

The above parameters are independently guaranteed.

ELECTRICAL CHARACTERISTICS ($T_{case}=+25^{\circ}C$, $Z_G=Z_L=50\Omega$, unless otherwise specified)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
F	Frequency Range		896		941	MHz
P_{out1}	Output Power 1	$V_{DD}=12.8V$, $V_{GG1}=3.4V$, $V_{GG2}=5V$, $P_{in}=50mW$	45			W
ηT	Total Efficiency	$V_{DD}=12.8V$	33			%
$2f_o$	2 nd Harmonic	$V_{GG1}=3.4V$			-40	dBc
$3f_o$	3 rd Harmonic	$V_{GG2}=5V$			-40	dBc
ρ_{in}	Input VSWR	$P_{in}=50mW$			3:1	—
I_{DD}	Leakage Current	$V_{DD}=17V$, $V_{GG1}=V_{GG2}=0V$, $P_{in}=0W$			1	mA
P_{out2}	Output Power 2*	$V_{DD}=15.2V$, $V_{GG1}=3.4V$, $V_{GG2}=1V$, $P_{in}=4dBm$			2	W
—	Stability	$V_{DD}=10.0-15.2V$, $P_{in}=1-100mW$, $1.5 < P_{out} < 50W$ (V_{GG2} control, $V_{GG1}=3.4V$), Load VSWR=3:1	No parasitic oscillation			—
—	Load VSWR Tolerance	$V_{DD}=15.2V$, $P_{in}=50mW$, $P_{out}=45W$ (V_{GG2} control, $V_{GG1}=3.4V$), Load VSWR=20:1	No degradation or destroy			—

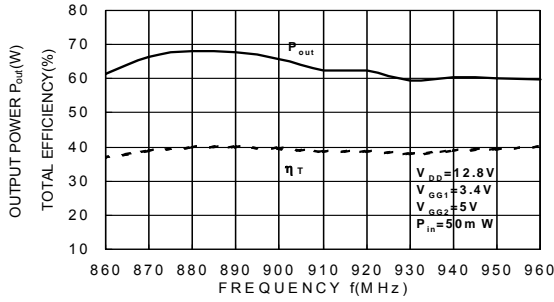
*: This is guaranteed as design value.

All parameters, conditions, ratings, and limits are subject to change without notice.

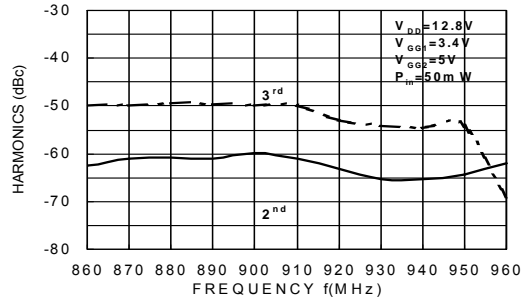


TYPICAL PERFORMANCE ($T_{case}=+25^{\circ}C$, $Z_G=Z_L=50\Omega$, unless otherwise specified)

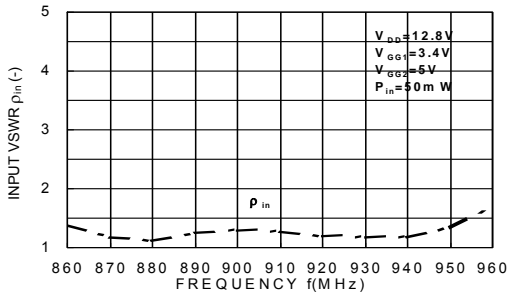
OUTPUT POWER, TOTAL EFFICIENCY, versus FREQUENCY



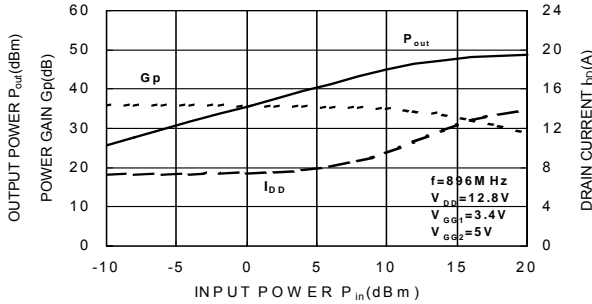
2nd, 3rd HARMONICS versus FREQUENCY



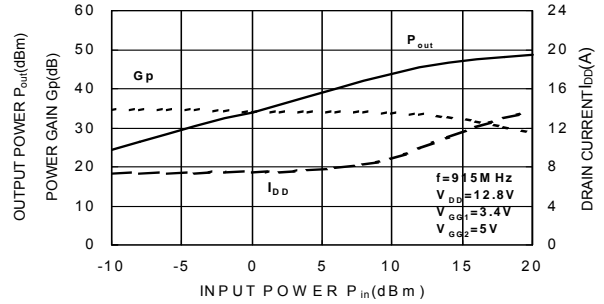
INPUT VSWR versus FREQUENCY



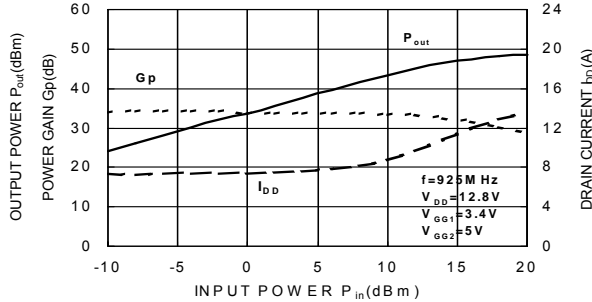
OUTPUT POWER, POWER GAIN and DRAIN CURRENT versus INPUT POWER



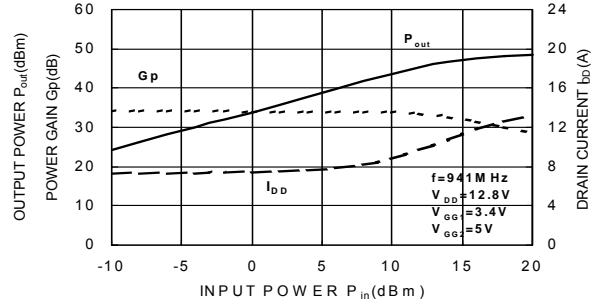
OUTPUT POWER, POWER GAIN and DRAIN CURRENT versus INPUT POWER



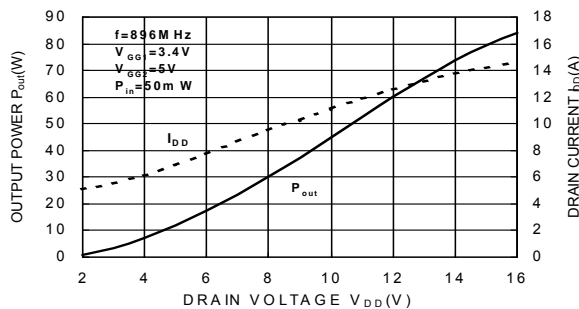
OUTPUT POWER, POWER GAIN and DRAIN CURRENT versus INPUT POWER



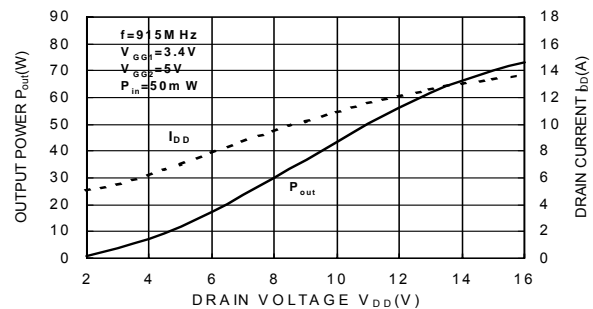
OUTPUT POWER, POWER GAIN and DRAIN CURRENT versus INPUT POWER



OUTPUT POWER and DRAIN CURRENT versus DRAIN VOLTAGE



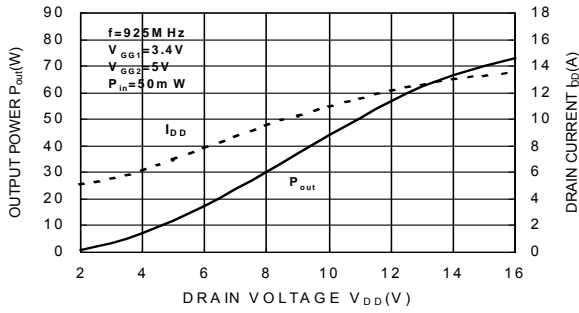
OUTPUT POWER and DRAIN CURRENT versus DRAIN VOLTAGE



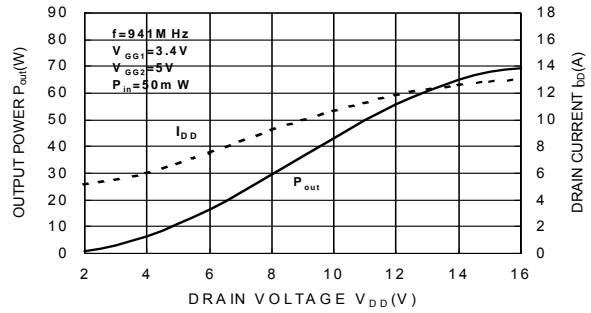


TYPICAL PERFORMANCE ($T_{case}=+25^{\circ}C$, $Z_G=Z_L=50\Omega$, unless otherwise specified)

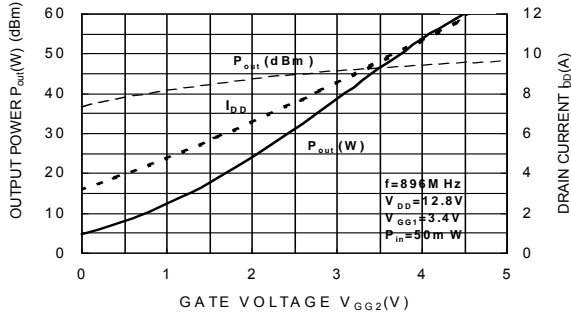
OUTPUT POWER and DRAIN CURRENT
versus DRAIN VOLTAGE



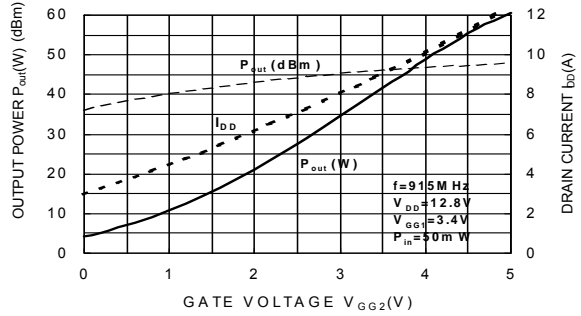
OUTPUT POWER and DRAIN CURRENT
versus DRAIN VOLTAGE



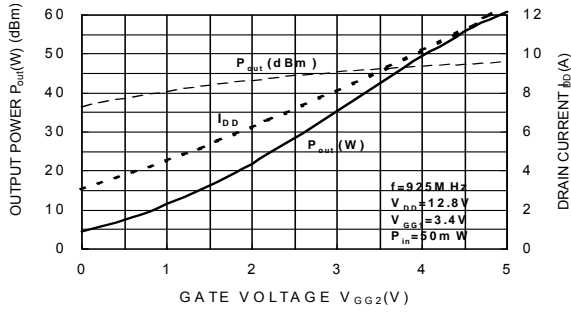
OUTPUT POWER and DRAIN CURRENT
versus GATE VOLTAGE 2



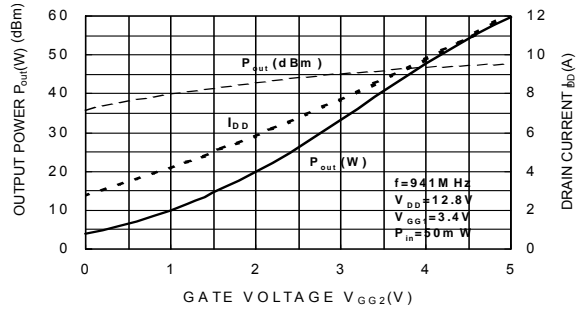
OUTPUT POWER and DRAIN CURRENT
versus GATE VOLTAGE 2



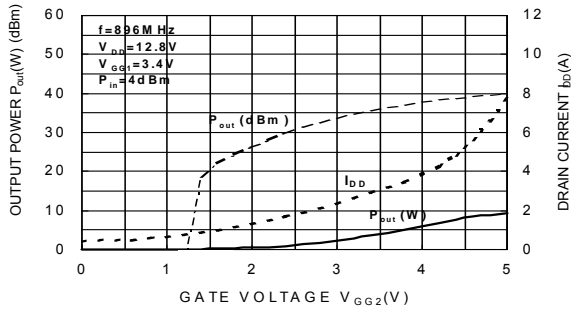
OUTPUT POWER and DRAIN CURRENT
versus GATE VOLTAGE 2



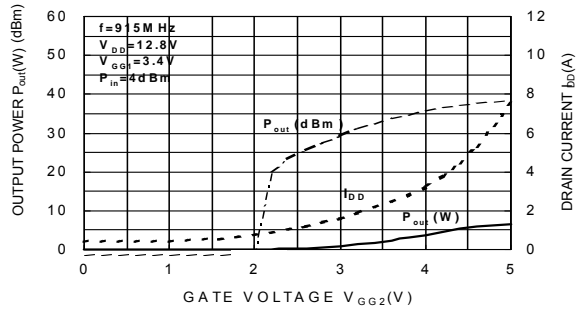
OUTPUT POWER and DRAIN CURRENT
versus GATE VOLTAGE 2



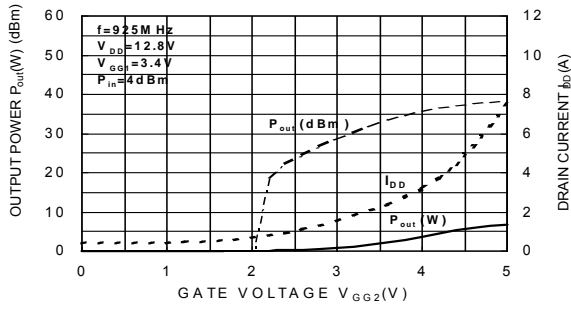
OUTPUT POWER and DRAIN CURRENT
versus GATE VOLTAGE 2



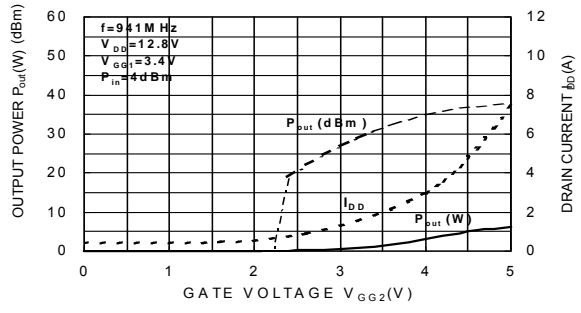
OUTPUT POWER and DRAIN CURRENT
versus GATE VOLTAGE 2



OUTPUT POWER and DRAIN CURRENT
versus GATE VOLTAGE 2

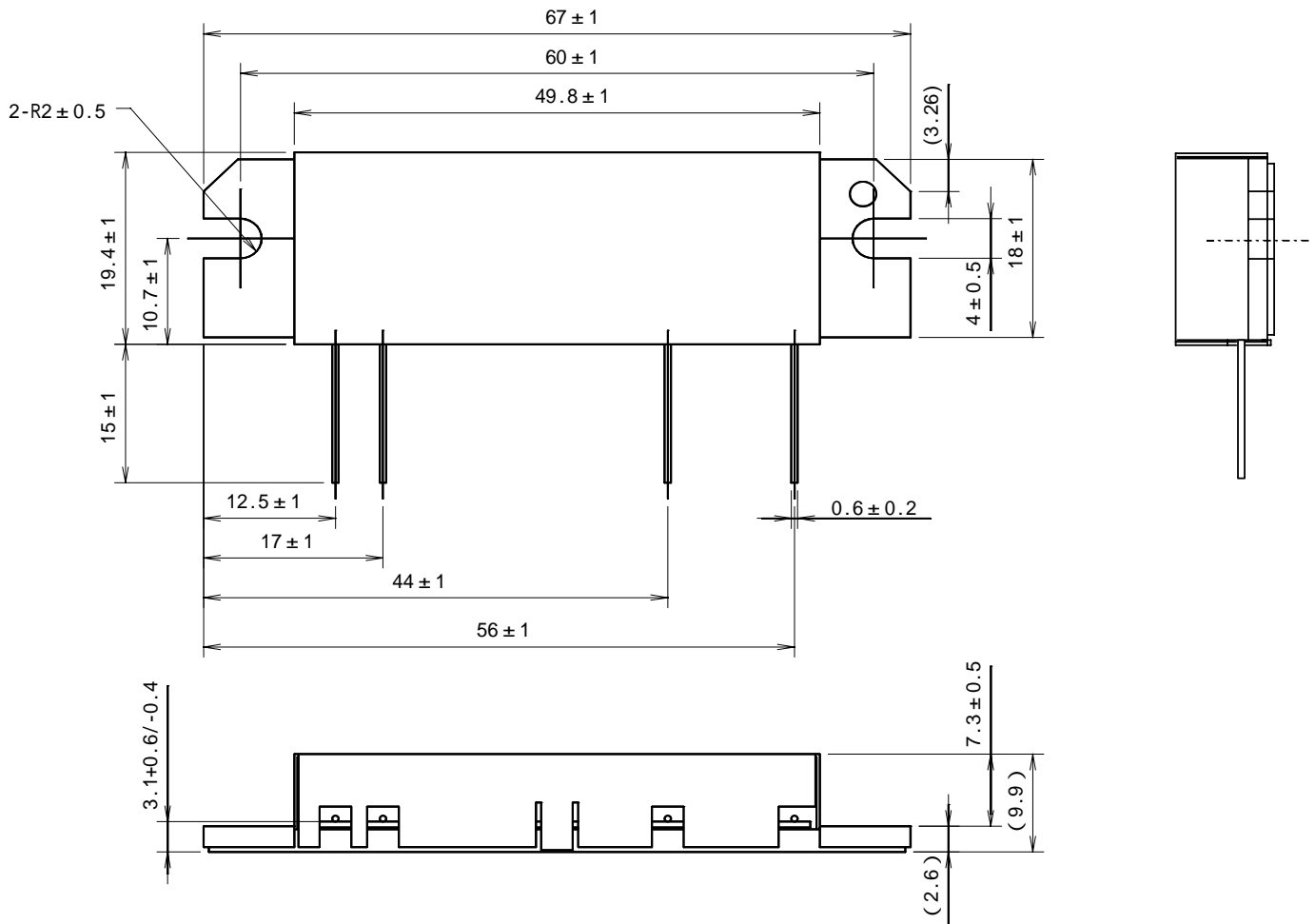


OUTPUT POWER and DRAIN CURRENT
versus GATE VOLTAGE 2





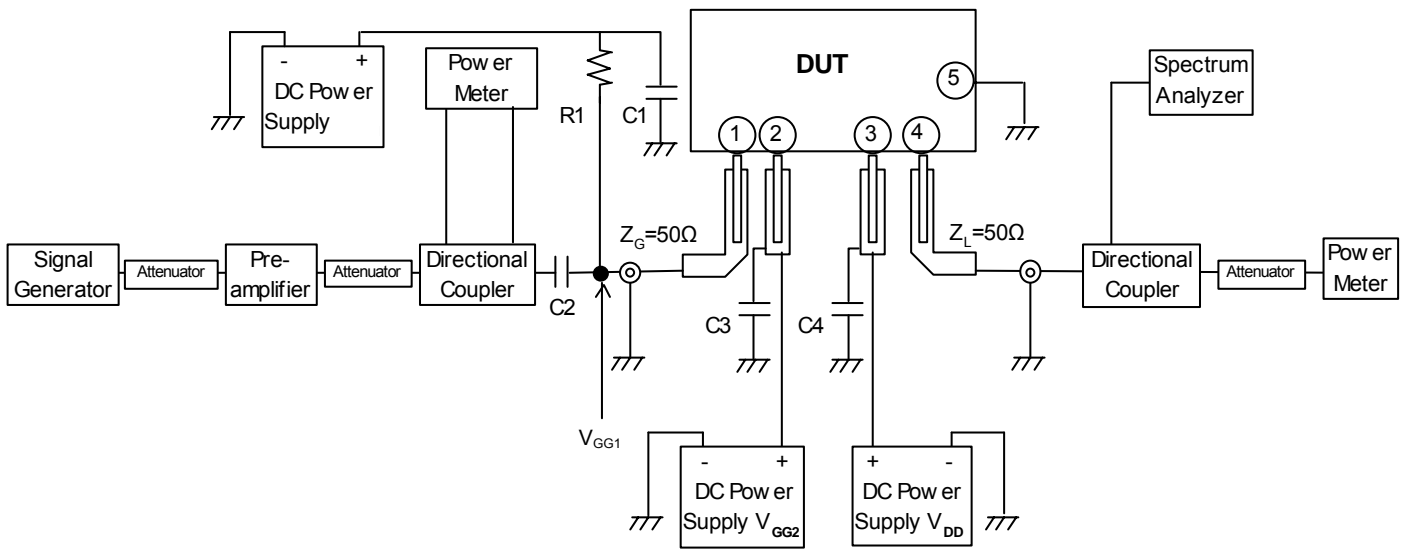
OUTLINE DRAWING (mm)



- ① RF Input added Gate Voltage 1 (P_{in} & V_{GG1})
- ② Gate Voltage 2 (V_{GG2})
- ③ Drain Voltage (V_{DD})
- ④ RF Output (P_{out})
- ⑤ RF Ground (Case)



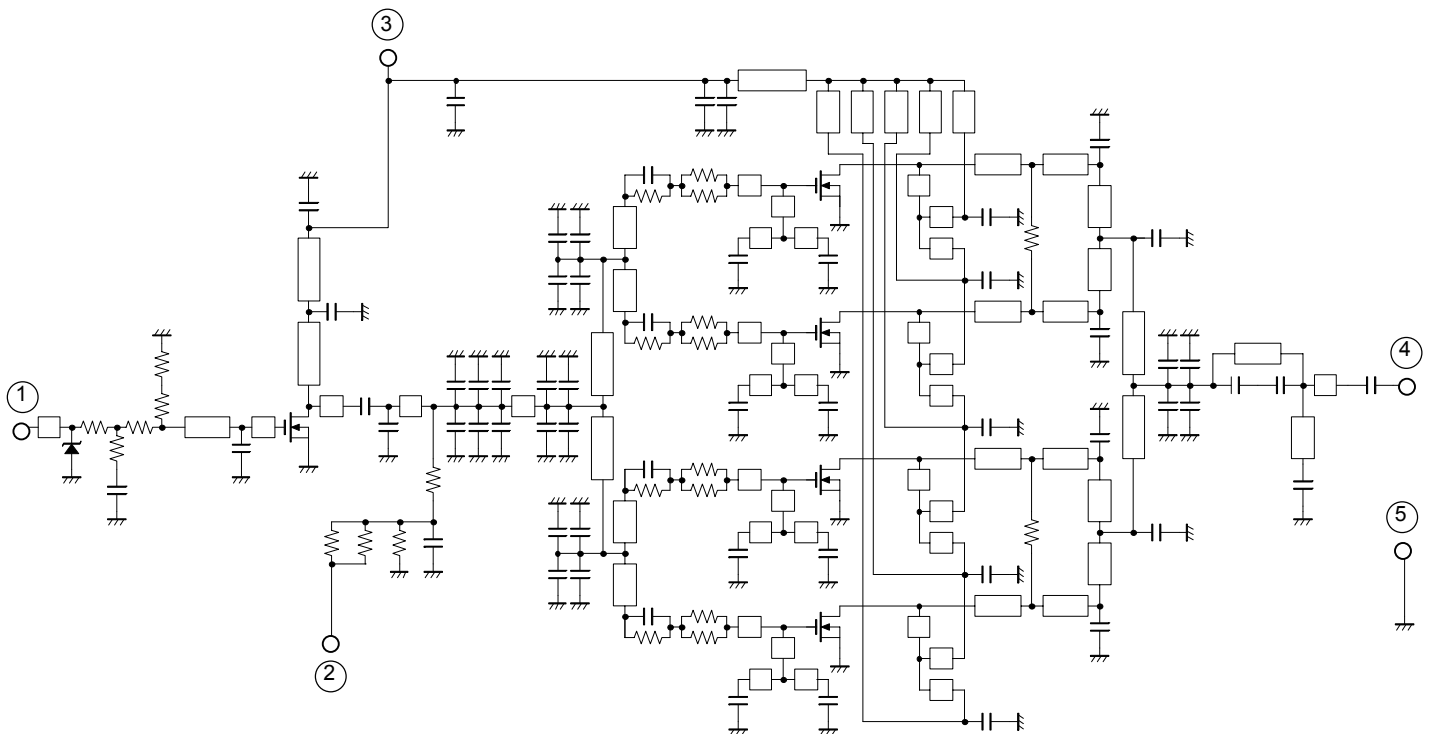
TEST BLOCK DIAGRAM



C1: 4700pF, C2: 1000pF, R1: suitable. Please refer the detail below.
C3, C4: 4700pF, 22uF in parallel
 $V_{GG1}=3.4V$

- ① RF Input added Gate Voltage 1 (P_{in} & V_{GG1})
- ② Gate Voltage 2 (V_{GG2})
- ③ Drain Voltage (V_{DD})
- ④ RF Output (P_{out})
- ⑤ RF Ground (Case)

EQUIVALENT CIRCUIT



NOTE: Resistance between Gate Voltage 1, where RF is input, and ground equals to 15k ohm.

External resistance connected to V_{GG1} ; impedance between P_{in} & V_{GG1} and ground needs to make high impedance that doesn't prevent RF characteristic on this module.



PRECAUTIONS, RECOMMENDATIONS, and APPLICATION INFORMATION:

Construction:

This module consists of a glass-epoxy substrate soldered onto a copper flange. For mechanical protection, a metal cap is attached (which makes the improvement of RF radiation easy). The MOSFET transistor chips are die bonded onto metal, wire bonded to the substrate, and coated with resin. Lines on the substrate (eventually inductors), chip capacitors, and resistors form the bias and matching circuits. Wire leads soldered onto the glass-epoxy substrate provide the DC and RF connection.

Following conditions must be avoided:

- a) Bending forces on the glass-epoxy substrate (for example, by driving screws or from fast thermal changes)
- b) Mechanical stress on the wire leads (for example, by first soldering then driving screws or by thermal expansion)
- c) Defluxing solvents reacting with the resin coating on the MOSFET chips (for example, Trichloroethylene)
- d) Frequent on/off switching that causes thermal expansion of the resin
- e) ESD, surge, overvoltage in combination with load VSWR, and oscillation

ESD:

This MOSFET module is sensitive to ESD voltages down to 1000V. Appropriate ESD precautions are required.

Mounting:

A thermal compound between module and heat sink is recommended for low thermal contact resistance and to reduce the bending stress on the glass-epoxy substrate caused by the temperature difference to the heat sink. The module must first be screwed to the heat sink, then the leads can be soldered to the printed circuit board. M3 screws are recommended with a tightening torque of 0.4 to 0.6 Nm.

Soldering and Defluxing:

This module is designed for manual soldering.

The leads must be soldered after the module is screwed onto the heat sink.

The temperature of the lead (terminal) soldering should be lower than 350°C and shorter than 3 second.

Ethyl Alcohol is recommend for removing flux. Trichloroethylene solvents must not be used (they may cause bubbles in the coating of the transistor chips which can lift off the bond wires).

Thermal Design of the Heat Sink:

At $P_{out}=45W$, $V_{DD}=12.8V$ and $P_{in}=50mW$ each stage transistor operating conditions are:

Stage	P_{in} (W)	P_{out} (W)	$R_{th(ch-case)}$ (°C/W)	$I_{DD} @ \eta_T=33\%$ (A)	V_{DD} (V)
1 st	0.05	3.0	3.5	0.62	12.8
2 nd	3.0	45.0	0.6	9.96	

The channel temperatures of each stage transistor $T_{ch} = T_{case} + (V_{DD} \times I_{DD} - P_{out} + P_{in}) \times R_{th(ch-case)}$ are:

$$T_{ch1} = T_{case} + (12.8V \times 0.62A - 3.0W + 0.05W) \times 3.5°C/W = T_{case} + 17.5 °C$$

$$T_{ch2} = T_{case} + (12.8V \times 9.96A - 45.0W + 3.0W) \times 0.6°C/W = T_{case} + 51.3 °C$$

For long-term reliability, it is best to keep the module case temperature (T_{case}) below 90°C. For an ambient temperature $T_{air}=60°C$ and $P_{out}=45W$, the required thermal resistance $R_{th(case-air)} = (T_{case} - T_{air}) / ((P_{out} / \eta_T) - P_{out} + P_{in})$ of the heat sink, including the contact resistance, is:

$$R_{th(case-air)} = (90°C - 60°C) / (45W/33\% - 45W + 0.05W) = 0.33 °C/W$$

When mounting the module with the thermal resistance of 0.33 °C/W, the channel temperature of each stage transistor is:

$$T_{ch1} = T_{air} + 47.5 °C$$

$$T_{ch2} = T_{air} + 81.3 °C$$

The 175°C maximum rating for the channel temperature ensures application under derated conditions.



Output Power Control:

Depending on linearity, the following three methods are recommended to control the output power:

a) Non-linear FM modulation at high power operating:

By the gate voltages (V_{GG1} and V_{GG2}).

When the gate voltages are close to zero, the nominal output signal ($P_{out}=45W$) is attenuated up to 60 dB and only a small leakage current flows from the battery into the drain.

(On the following, V_{GG1} has to be kept in 3.4V.)

Around $V_{GG2}=0V$ (minimum), the output power and drain current increases substantially.

Around $V_{GG2}=4V$ (typical) to $V_{GG2}=5V$ (maximum), the nominal output power becomes available.

b) Linear AM modulation:

By RF input power P_{in} .

(On the following, V_{GG1} has to be kept in 3.4V.)

V_{GG2} is used to set the drain's quiescent current for the required linearity.

Oscillation:

To test RF characteristics, this module is put on a fixture with two bias decoupling capacitors each on gate and drain, a 4.700 pF chip capacitor, located close to the module, and a 22 μF (or more) electrolytic capacitor.

When an amplifier circuit around this module shows oscillation, the following may be checked:

a) Do the bias decoupling capacitors have a low inductance pass to the case of the module?

b) Is the load impedance $Z_L=50\Omega$?

c) Is the source impedance $Z_G=50\Omega$?

Frequent on/off switching:

In base stations, frequent on/off switching can cause thermal expansion of the resin that coats the transistor chips and can result in reduced or no output power. The bond wires in the resin will break after long-term thermally induced mechanical stress.

Quality:

Mitsubishi Electric is not liable for failures resulting from base station operation time or operating conditions exceeding those of mobile radios.

This module technology results from more than 20 years of experience, field proven in tens of millions of mobile radios. Currently, most returned modules show failures such as ESD, substrate crack, and transistor burnout, which are caused by improper handling or exceeding recommended operating conditions. Few degradation failures are found.

Keep safety first in your circuit designs!

Mitsubishi Electric Corporation puts the maximum effort into making semiconductor products better and more reliable, but there is always the possibility that trouble may occur. Trouble with semiconductors may lead to personal injury, fire or property damage. Remember to give due consideration to safety when making your circuit designs, with appropriate measures such as (i) placement of substitutive, auxiliary circuits, (ii) use of non-flammable material, or (iii) prevention against any malfunction or mishap.

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