

## 24-bit, 96kHz ADC with 8 Channel I/P Multiplexer

### DESCRIPTION

The WM8773 is a high performance, stereo audio ADC with an 8 channel input selector. The WM8773 is ideal for digitising multiple analogue sources for surround sound processing applications for home hi-fi, automotive and other audio visual equipment.

A stereo 24-bit multi-bit sigma delta ADC is used with an eight stereo channel input selector. Each channel has analogue domain mute and programmable gain control. Digital audio output word lengths from 16-32 bits and sampling rates from 8kHz to 96kHz are supported.

The audio data interface supports I<sup>2</sup>S, left justified, right justified and DSP digital audio formats.

The device is controlled via a 3 wire serial interface. The interface provides access to all features including channel selection, volume controls, mutes, de-emphasis and power management facilities. The device is available in a 64-pin TQFP package.

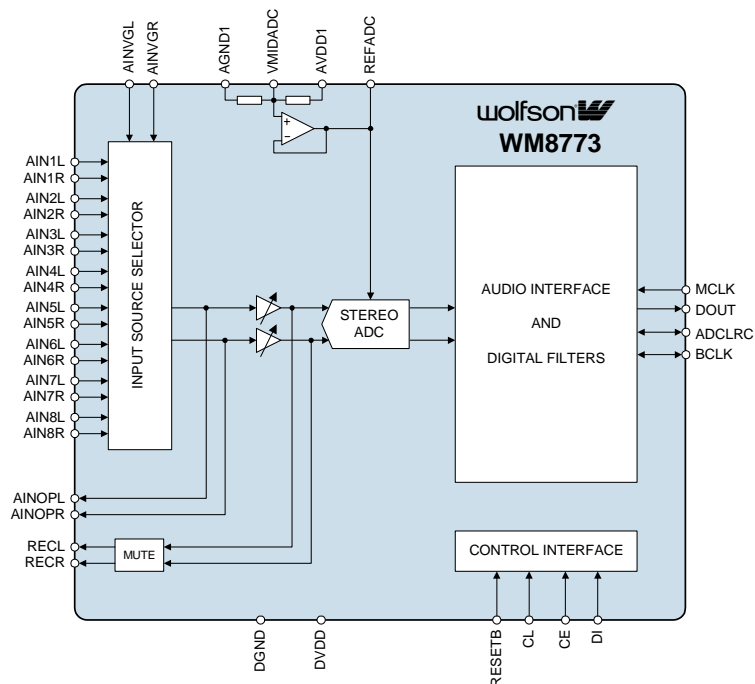
### FEATURES

- Audio Performance
  - 102dB SNR ('A' weighted @ 48kHz) ADC
- ADC Sampling Frequency: 8kHz – 96kHz
- 3-Wire SPI MPU Serial Control Interface
- Master or Slave Clocking Mode
- Programmable Audio Data Interface Modes
  - I<sup>2</sup>S, Left, Right Justified or DSP
  - 16/20/24/32 bit Word Lengths
- Analogue Record Monitor Outputs
- Eight stereo ADC inputs with analogue gain adjust from +19dB to –12dB in 1dB steps
- 2.7V to 5.5V Analogue, 2.7V to 3.6V Digital supply Operation
- 5V tolerant digital inputs

### APPLICATIONS

- Surround Sound AV Processors and Hi-Fi systems
- Automotive Audio

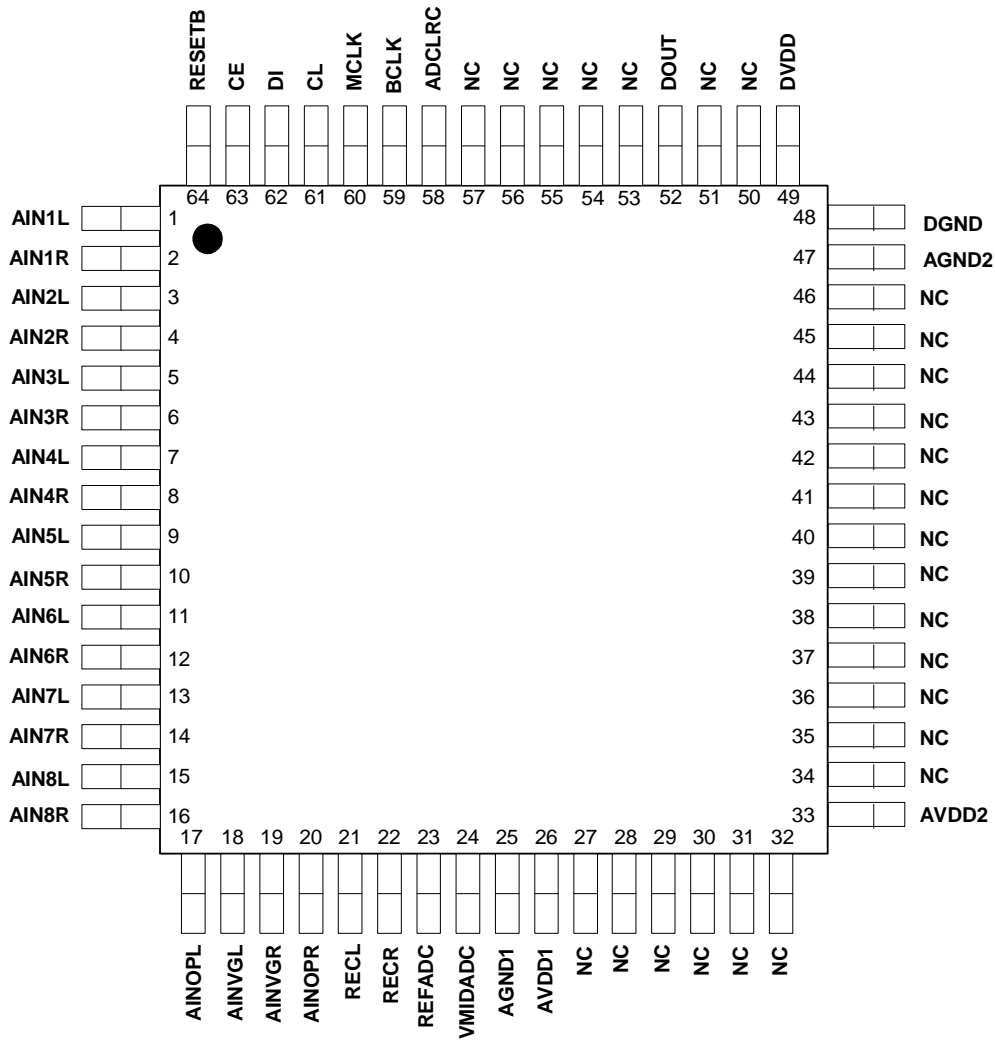
### BLOCK DIAGRAM



**PIN CONFIGURATION**

**ORDERING INFORMATION**

DEVICE	TEMP. RANGE	PACKAGE
WM8773IFT/V	-40 to +85°C	64-pin TQFP



**PIN DESCRIPTION**

<b>PIN</b>	<b>NAME</b>	<b>TYPE</b>	<b>DESCRIPTION</b>
1	AIN1L	Analogue Input	Channel 1 left input multiplexor virtual ground
2	AIN1R	Analogue Input	Channel 1 right input multiplexor virtual ground
3	AIN2L	Analogue Input	Channel 2 left input multiplexor virtual ground
4	AIN2R	Analogue Input	Channel 2 right input multiplexor virtual ground
5	AIN3L	Analogue Input	Channel 3 left input multiplexor virtual ground
6	AIN3R	Analogue Input	Channel 3 right input multiplexor virtual ground
7	AIN4L	Analogue Input	Channel 4 left input multiplexor virtual ground
8	AIN4R	Analogue Input	Channel 4 right input multiplexor virtual ground
9	AIN5L	Analogue Input	Channel 5 left input multiplexor virtual ground
10	AIN5R	Analogue Input	Channel 5 right input multiplexor virtual ground
11	AIN6L	Analogue Input	Channel 6 left input multiplexor virtual ground
12	AIN6R	Analogue Input	Channel 6 right input multiplexor virtual ground
13	AIN7L	Analogue Input	Channel 7 left input multiplexor virtual ground
14	AIN7R	Analogue Input	Channel 7 right input multiplexor virtual ground
15	AIN8L	Analogue Input	Channel 8 left input multiplexor virtual ground
16	AIN8R	Analogue Input	Channel 8 right input multiplexor virtual ground
17	AINOPL	Analogue Output	Left channel multiplexor output
18	AINVGL	Analogue Input	Left channel multiplexor virtual ground
19	AINVGR	Analogue Input	Right channel multiplexor virtual ground
20	AINOPR	Analogue Output	Right channel multiplexor output
21	RECL	Analogue Output	Left channel input mux select output
22	RECR	Analogue Output	Right channel input mux select output
23	REFADC	Analogue Output	ADC reference buffer decoupling pin; 10uF external decoupling
24	VMIDADC	Analogue Output	ADC midrail divider decoupling pin; 10uF external decoupling
25	AGND1	Supply	Analogue negative supply and substrate connection
26	AVDD1	Supply	Analogue positive supply
27		NC	No connection
28		NC	No connection
29		NC	No connection
30		NC	No connection
31		NC	No connection
32		NC	No connection
33	AVDD2	Supply	Analogue positive supply
34		NC	No connection
35		NC	No connection
36		NC	No connection
37		NC	No connection
38		NC	No connection
39		NC	No connection
40		NC	No connection
41		NC	No connection
42		NC	No connection
43		NC	No connection
44		NC	No connection
45		NC	No connection
46		NC	No connection
47	AGND2	Supply	Analogue negative supply and substrate connection
48	DGND	Supply	Digital negative supply
49	DVDD	Supply	Digital positive supply

PIN	NAME	TYPE	DESCRIPTION
50		NC	No connection
51		NC	No connection
52	DOUT	Digital output	ADC data output
53		NC	No connection
54		NC	No connection
55		NC	No connection
56		NC	No connection
57		NC	No connection
58	ADCLRC	Digital input/output	ADC left/right word clock
59	BCLK	Digital input/output	ADC audio interface bit clock
60	MCLK	Digital input	Master ADC clock; 256, 384, 512 or 768fs (fs = word clock frequency)
61	CL	Digital input	Serial interface clock (5V tolerant)
62	DI	Digital input	Serial interface data (5V tolerant)
63	CE	Digital input	Serial interface Latch signal (5V tolerant)
64	RESETB	Digital input	Device reset input (resets gain stages to 0dB) (5V tolerant)

**Note** : Digital input pins have Schmitt trigger input buffers and are 5V tolerant.

## ABSOLUTE MAXIMUM RATINGS

Absolute Maximum Ratings are stress ratings only. Permanent damage to the device may be caused by continuously operating at or beyond these limits. Device functional operating limits and guaranteed performance specifications are given under Electrical Characteristics at the test conditions specified.



ESD Sensitive Device. This device is manufactured on a CMOS process. It is therefore generically susceptible to damage from excessive static voltages. Proper ESD precautions must be taken during handling and storage of this device.

CONDITION	MIN	MAX
Digital supply voltage	-0.3V	+3.63V
Analogue supply voltage	-0.3V	+7V
Voltage range digital inputs (DI, CL, CE & RESETB)	DGND -0.3V	+7V
Voltage range digital inputs (MCLK, ADCLRC & BCLK)	DGND -0.3V	DVDD + 0.3V
Voltage range analogue inputs	AGND -0.3V	AVDD +0.3V
Master Clock Frequency		37MHz
Operating temperature range, T <sub>A</sub>	-40°C	+85°C
Storage temperature	-65°C	+150°C
Package body temperature (soldering 10 seconds)		+240°C
Package body temperature (soldering 2 minutes)		+183°C

### Notes:

- Analogue and digital grounds must always be within 0.3V of each other.

**RECOMMENDED OPERATING CONDITIONS**

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Digital supply range	DVDD		2.7		3.6	V
Analogue supply range	AVDD		2.7		5.5	V
Ground	AGND, DGND			0		V
Difference DGND to AGND			-0.3	0	+0.3	V

**Note:** Digital supply DVDD must never be more than 0.3V greater than AVDD.

**ELECTRICAL CHARACTERISTICS****Test Conditions**

AVDD = 5V, DVDD = 3.3V, AGND = 0V, DGND = 0V, T<sub>A</sub> = +25°C, fs = 48kHz, MCLK = 256fs unless otherwise stated.

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>Digital Logic Levels (TTL Levels)</b>						
Input LOW level	V <sub>IL</sub>				0.8	V
Input HIGH level	V <sub>IH</sub>		2.0			V
Output LOW	V <sub>OL</sub>	I <sub>OL</sub> = 1mA			0.1 x DVDD	V
Output HIGH	V <sub>OH</sub>	I <sub>OH</sub> = -1mA	0.9 x DVDD			V
<b>Analogue Reference Levels</b>						
Reference voltage	V <sub>VMID</sub>		AVDD/2 – 50mV	AVDD/2	AVDD/2 + 50mV	V
Potential divider resistance	R <sub>VMID</sub>	AVDD to VMIDADC and VMIDADC to AGND	40k	50k	60k	Ohms
<b>ADC Performance</b>						
Input Signal Level (0dB)				1.0 x AVDD/5		Vrms
SNR (Note 1,2)		A-weighted, 0dB gain @ fs = 48kHz	93	102		dB
SNR (Note 1,2)		A-weighted, 0dB gain @ fs = 96kHz		98		dB
Dynamic Range (note 2)		A-weighted, -60dB full scale input		102		dB
Total Harmonic Distortion (THD)		1kHz, 0dBFS		-90	-80	dB
		1kHz, -3dBFS		-95	-85	dB
ADC Channel Separation		1kHz Input		90		dB
Programmable Gain Step Size			0.5	1.0	1.5	dB
Programmable Gain Range		1kHz Input	-12		+19	dB
Mute Attenuation		1kHz Input, 0dB gain		97		dB
Power Supply Rejection Ratio	PSRR	1kHz 100mVpp		50		dB
		20Hz to 20kHz 100mVpp		45		dB

**Test Conditions**

AVDD = 5V, DVDD = 3.3V, AGND = 0V, DGND = 0V, T<sub>A</sub> = +25°C, fs = 48kHz, MCLK = 256fs unless otherwise stated.

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>Analogue input (AIN) to Analogue output (VOUT) (Load=10k ohms, 50pF, gain = 0dB) Record Monitor Output</b>						
0dB Full scale output voltage				1.0 x AVDD/5		Vrms
SNR (Note 1)			90	100		dB
THD		1kHz, 0dB		-90		dB
		1kHz, -3dB		-95		dB
Power Supply Rejection Ratio	PSRR	1kHz 100mVpp		50		dB
		20Hz to 20kHz 100mVpp		45		dB
Mute Attenuation (REC Output Only)		1kHz, 0dB		100		dB
<b>Supply Current</b>						
Analogue supply current		AVDD = 5V		100		mA
Digital supply current		DVDD = 3.3V		20		mA

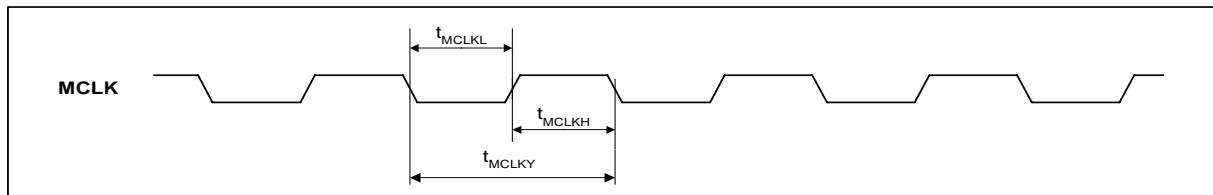
**Notes:**

- Ratio of output level with 1kHz full scale input, to the output level with all zeros into the digital input, measured 'A' weighted.
- All performance measurements done with 20kHz low pass filter, and where noted an A-weight filter. Failure to use such a filter will result in higher THD+N and lower SNR and Dynamic Range readings than are found in the Electrical Characteristics. The low pass filter removes out of band noise; although it is not audible it may affect dynamic specification values.
- VMID decoupled with 10uF and 0.1uF capacitors (smaller values may result in reduced performance).

**TERMINOLOGY**

- Signal-to-noise ratio (dB) - SNR is a measure of the difference in level between the full scale output and the output with no signal applied. (No Auto-zero or Automute function is employed in achieving these results).
- Dynamic range (dB) - DNR is a measure of the difference between the highest and lowest portions of a signal. Normally a THD+N measurement at 60dB below full scale. The measured signal is then corrected by adding the 60dB to it. (e.g. THD+N @ -60dB= -32dB, DR= 92dB).
- THD+N (dB) - THD+N is a ratio, of the rms values, of (Noise + Distortion)/Signal.
- Stop band attenuation (dB) - Is the degree to which the frequency spectrum is attenuated (outside audio band).
- Channel Separation (dB) - Also known as Cross-Talk. This is a measure of the amount one channel is isolated from the other. Normally measured by sending a full scale signal down one channel and measuring the other.
- Pass-Band Ripple - Any variation of the frequency response in the pass-band region.

**MASTER CLOCK TIMING**



**Figure 1 Master Clock Timing Requirements**

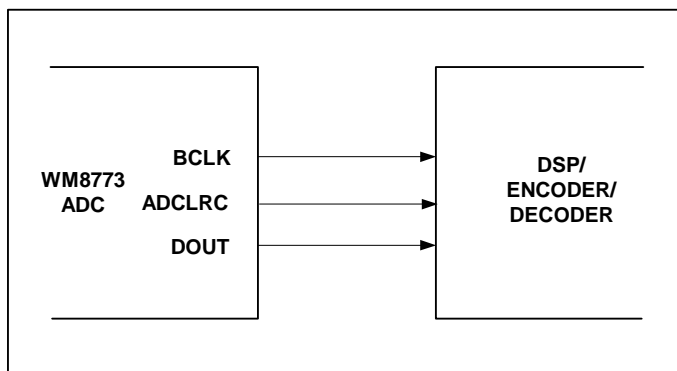
**Test Conditions**

AVDD = 5V, DVDD = 3.3V, AGND = 0V, DGND = 0V,  $T_A = +25^{\circ}C$ ,  $f_s = 48kHz$ , MCLK = 256fs unless otherwise stated.

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>System Clock Timing Information</b>						
MCLK System clock pulse width high	$t_{MCLKH}$		11			ns
MCLK System clock pulse width low	$t_{MCLKL}$		11			ns
MCLK System clock cycle time	$t_{MCLKY}$		28			ns
MCLK Duty cycle			40:60		60:40	

**Table 1 Master Clock Timing Requirements**

**DIGITAL AUDIO INTERFACE – MASTER MODE**



**Figure 2 Audio Interface - Master Mode**



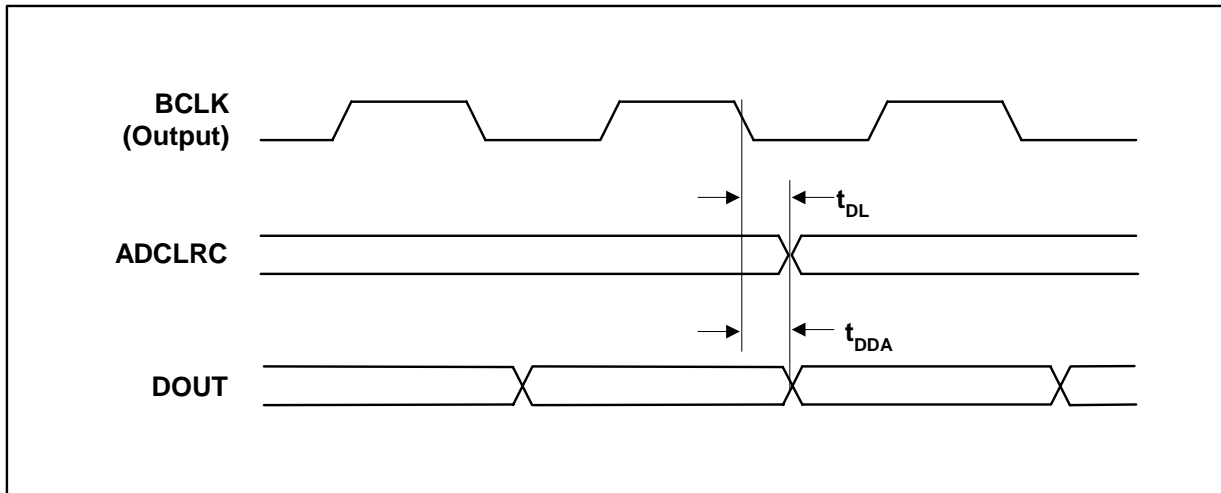


Figure 3 Digital Audio Data Timing – Master Mode

**Test Conditions**

AVDD = 5V, DVDD = 3.3V, AGND, DGND = 0V,  $T_A = +25^\circ\text{C}$ , Master Mode,  $f_s = 48\text{kHz}$ , MCLK = 256fs unless otherwise stated.

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>Audio Data Input Timing Information</b>						
ADCLRC propagation delay from BCLK falling edge	$t_{DL}$		0		10	ns
DOUT propagation delay from BCLK falling edge	$t_{DDA}$		0		10	ns

Table 2 Digital Audio Data Timing – Master Mode

**DIGITAL AUDIO INTERFACE – SLAVE MODE**

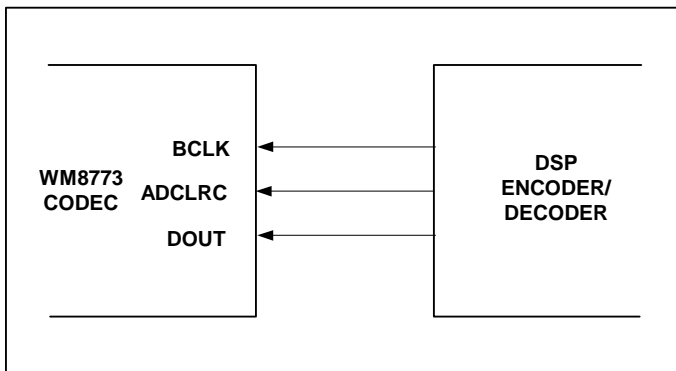


Figure 4 Audio Interface – Slave Mode

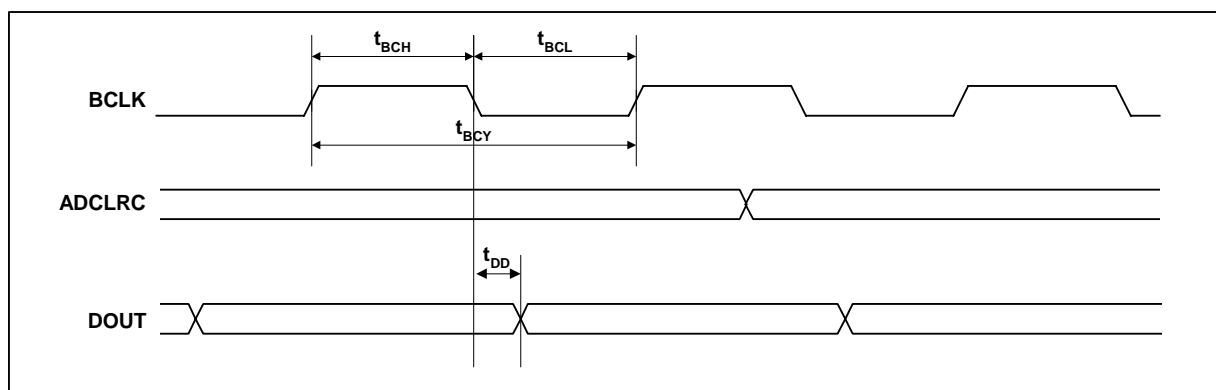


Figure 5 Digital Audio Data Timing – Slave Mode

**Test Conditions**

AVDD = 5V, DVDD = 3.3V, AGND = 0V, DGND = 0V,  $T_A = +25^\circ\text{C}$ , Slave Mode,  $f_s = 48\text{kHz}$ , MCLK = 256fs unless otherwise stated.

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>Audio Data Input Timing Information</b>						
BCLK cycle time	$t_{BCY}$		50			ns
BCLK pulse width high	$t_{BCH}$		20			ns
BCLK pulse width low	$t_{BCL}$		20			ns
DOUT propagation delay from BCLK falling edge	$t_{DD}$		0		10	ns

Table 3 Digital Audio Data Timing – Slave Mode

**Note:**

- ADCLRC should be synchronous with MCLK, although the WM8773 interface is tolerant of phase variations or jitter on these signals.

## MPU INTERFACE TIMING

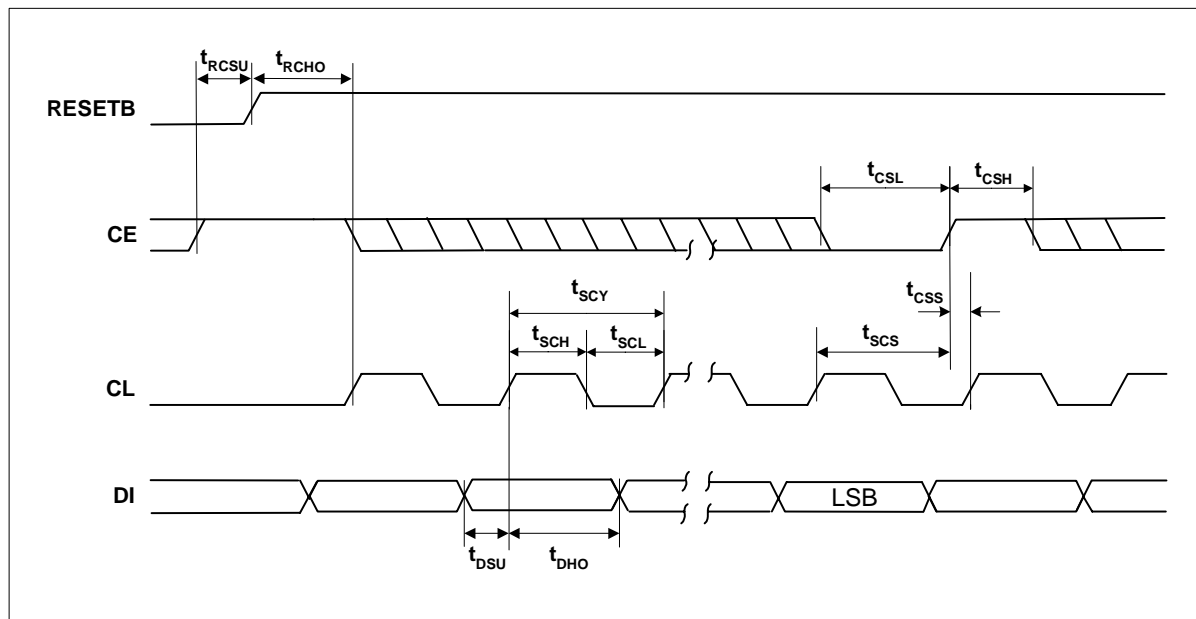


Figure 6 SPI compatible Control Interface Input Timing

Test Conditions					
AVDD = 5V, DVDD = 3.3V, AGND, DGND = 0V, T <sub>A</sub> = +25°C, f <sub>s</sub> = 48kHz, MCLK = 256fs unless otherwise stated					
PARAMETER	SYMBOL	MIN	TYP	MAX	UNIT
CE to RESETB hold time	t <sub>RCSU</sub>	20			ns
RESETB to CL setup time	t <sub>RCHO</sub>	20			ns
CL rising edge to CE rising edge	t <sub>SCS</sub>	60			ns
CL pulse cycle time	t <sub>SCY</sub>	80			ns
CL pulse width low	t <sub>SCL</sub>	30			ns
CL pulse width high	t <sub>SCH</sub>	30			ns
DI to CL set-up time	t <sub>DSU</sub>	20			ns
CL to DI hold time	t <sub>DHO</sub>	20			ns
CE pulse width low	t <sub>CSL</sub>	20			ns
CE pulse width high	t <sub>CSH</sub>	20			ns
CE rising to CL rising	t <sub>CSS</sub>	20			ns

Table 4 3-wire SPI compatible Control Interface Input Timing Information

## DEVICE DESCRIPTION

### INTRODUCTION

WM8773 is a complete stereo audio ADC with 8-channel multiplexed input.

The input multiplexor to the ADC is configured to allow large signal levels to be input to the ADC, using external resistors to reduce the amplitude of larger signals to within the normal operating range of the ADC. The ADC input PGA also allows input signals to be gained up to +19dB and attenuated down to -12dB. This allows the user maximum flexibility in the use of the ADC.

Analogue record monitor outputs are also available, to allow stereo analogue signals from any of the 8 stereo inputs to be sent to one of the two stereo outputs. This allows the user to monitor the signal that is being digitised either prior to the input programmable gain amplifier (PGA) or after gain has been applied. It is intended that the RECL/R outputs are only used to drive a high impedance buffer.

The Audio Interface may be configured to operate in either master or slave mode. In Slave mode ADCLRC and BCLK are both inputs. In Master mode ADCLRC and BCLK are all outputs.

Control of internal functionality of the device is by 3-wire serial control interface. The control interface may be asynchronous to the audio data interface as control data will be re-synchronised to the audio processing internally. CE, CL, DI and RESETB are 5V tolerant with TTL input thresholds, allowing the WM8773 to be used with DVDD = 3.3V and be controlled by a controller with 5V output.

Operation using system clock of 256fs, 384fs, 512fs or 768fs is provided. In Slave mode selection between clock rates is automatically controlled. In master mode the master clock to sample rate ratio is set by control bit ADCRATE. Sample rates (fs) from less than 8ks/s up to 96ks/s are allowed, provided the appropriate system clock is input.

The audio data interface supports right, left and I<sup>2</sup>S interface formats along with a highly flexible DSP serial port interface.

### AUDIO DATA SAMPLING RATES

In a typical digital audio system there is only one central clock source producing a reference clock to which all audio data processing is synchronised. This clock is often referred to as the audio system's Master Clock. The external master system clock can be applied directly through the MCLK input pin with no software configuration necessary. In a system where there are a number of possible sources for the reference clock it is recommended that the clock source with the lowest jitter be used to optimise the performance of the ADC.

The master clock for WM8773 supports ADC audio sampling rates from 256fs to 768fs, where fs is the audio sampling frequency (ADCLRC) typically 32kHz, 44.1kHz, 48kHz or 96kHz. The master clock is used to operate the digital filters and the noise shaping circuits.

In Slave mode the WM8773 has a master detection circuit that automatically determines the relationship between the master clock frequency and the sampling rate (to within +/- 32 system clocks). If there is a greater than 32 clocks error the interface is disabled and maintains the output level at the last sample. The master clock must be synchronised with ADCLRC, although the WM8773 is tolerant of phase variations or jitter on this clock. Table 5 shows the typical master clock frequency inputs for the WM8773.

The signal processing for the WM8773 typically operates at an oversampling rate of 128fs for the ADC. For ADC operation at 96kHz, it is recommended that the user set the ADCOSR bit. This changes the ADC signal processing oversample rate to 64fs.

SAMPLING RATE (ADCLRC)	System Clock Frequency (MHz)			
	256fs	384fs	512fs	768fs
32kHz	8.192	12.288	16.384	24.576
44.1kHz	11.2896	16.9340	22.5792	33.8688
48kHz	12.288	18.432	24.576	36.864

Table 5 System Clock Frequencies Versus Sampling Rate

In Master mode BCLK and ADCLRC are generated by the WM8773. The frequency of ADCLRC is determined by setting the required ratio of MCLK to ADCLRC using the ADCRATE control bits (Table 6).

ADCRATE[2:0]	MCLK:ADCLRC RATIO
010	256fs
011	384fs
100	512fs
101	768fs

**Table 6 Master Mode MCLK:ADCLRC ratio select**

Table 7 shows the settings for ADCRATE for common sample rates and MCLK frequencies.

SAMPLING RATE ADCLRC	System Clock Frequency (MHz)			
	256fs	384fs	512fs	768fs
	ADCRATE =010	ADCRATE =011	ADCRATE =100	ADCRATE =101
32kHz	8.192	12.288	16.384	24.576
44.1kHz	11.2896	16.9340	22.5792	33.8688
48kHz	12.288	18.432	24.576	36.864
96kHz	24.576	36.864	Unavailable	Unavailable

**Table 7 Master Mode ADC frequency selection**

BCLK is also generated by the WM8773. The frequency of BCLK depends on the mode of operation.

In 256/384/512fs modes (ADCRATE=010 or 011, 100 or 101) BCLK = MCLK/4. However if DSP mode is selected as the audio interface mode then BCLK=MCLK.

## POWERDOWN MODES

The WM8773 has powerdown control bits allowing specific parts of the WM8773 to be powered off when not being used. The 8-channel input source selector and input buffer may be powered down using control bit AINPD. When AINPD is set all inputs to the source selector (AIN1I/R to AIN8L/R) are switched to a buffered VMIDADC. Control bit ADCPD powers off the ADC and also the ADC input PGAs. Setting AINPD and ADCPD will powerdown everything except the references VMIDADC and ADCREF. These may be powered down by setting PDWN. Setting PDWN will override all other powerdown control bits. It is recommended that the 8-channel input mux and buffer and ADC are powered down before setting PDWN. The default is for all powerdown bits to be set except PDWN.

The Powerdown control bits allow parts of the device to be powered down when not in use.

## DIGITAL AUDIO INTERFACE

### MASTER AND SLAVE MODES

The audio interface operates in either Slave or Master mode, selectable using the MS control bit. In both Master and Slave modes, ADCDAT is always an output. The default is Slave mode.

In Slave mode (MS=0) ADCLRC and BCLK are inputs to the WM8773 (Figure 7). ADCLRC is sampled by the WM8773 on the rising edge of BCLK. ADC data is output on DOUT and changes on the falling edge of BCLK. By setting control bit BCLKINV the polarity of BCLK may be reversed so that ADCLRC is sampled on the falling edge of BCLK and DOUT changes on the rising edge of BCLK.

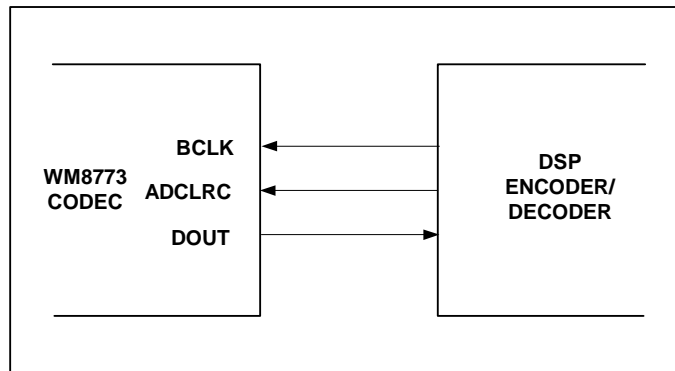


Figure 7 Slave Mode

In Master mode (MS=1) ADCLRC and BCLK are outputs from the WM8773 (Figure 8). ADCLRC and BITCLK are generated by the WM8773. ADCDAT is output on DOUT and changes on the falling edge of BCLK. By setting control bit BCLKINV the polarity of BCLK may be reversed so that DOUT changes on the rising edge of BCLK.

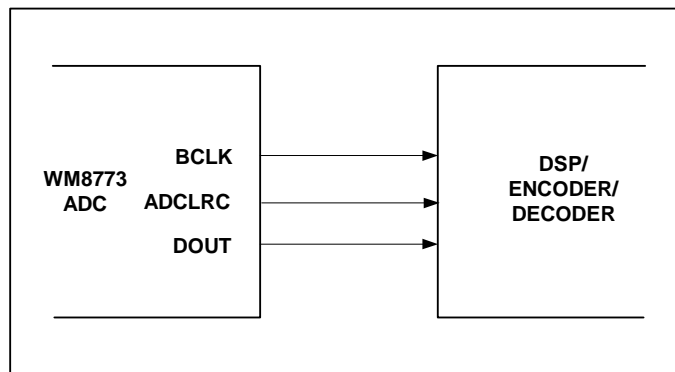


Figure 8 Master Mode

### AUDIO INTERFACE FORMATS

Audio data is output from the ADC filters, via the Digital Audio Interface. 5 popular interface formats are supported:

- Left Justified mode
- Right Justified mode
- I<sup>2</sup>S mode
- DSP Early mode
- DSP Late mode

All 5 formats send the MSB first and support word lengths of 16, 20, 24 and 32 bits, with the exception of 32 bit right justified mode, which is not supported.

In left justified, right justified and I<sup>2</sup>S modes, the digital audio interface outputs ADC data on DOUT. Audio Data for each stereo channel is time multiplexed with ADCLRC indicating whether the left or right channel is present. ADCLRC is also used as a timing reference to indicate the beginning or end of the data words.

In left justified, right justified and I<sup>2</sup>S modes, the minimum number of BCLKs per ADCLRC period is 2 times the selected word length. ADCLRC must be high for a minimum of word length BCLKs and low for a minimum of word length BCLKs. Any mark to space ratio on ADCLRC is acceptable provided the above requirements are met.

The ADC data may also be output in DSP early or late modes, with ADCLRC used as a frame sync to identify the MSB of the first word. The minimum number of BCLKs per ADCLRC period is 2 times the selected word length

**LEFT JUSTIFIED MODE**

In left justified mode, the MSB of the ADC data is output on DOUT and changes on the same falling edge of BCLK as ADCLRC and may be sampled on the rising edge of BCLK. ADCLRC is high during the left samples and low during the right samples (Figure 9).

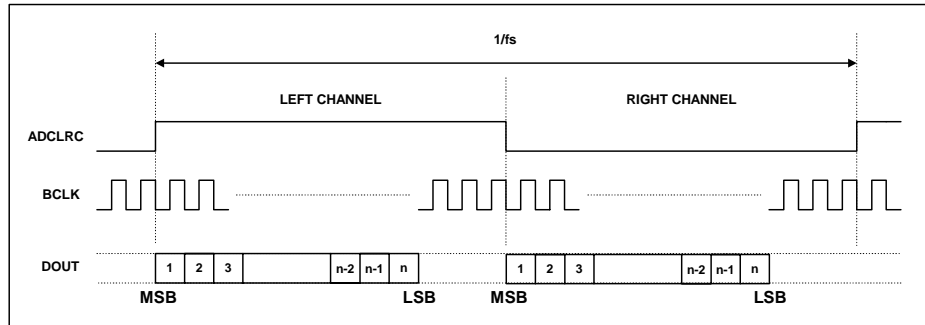


Figure 9 Left Justified Mode Timing Diagram

**RIGHT JUSTIFIED MODE**

In right justified mode, the LSB of the ADC data is output on DOUT and changes on the falling edge of BCLK preceding an ADCLRC transition and may be sampled on the rising edge of BCLK. ADCLRC is high during the left samples and low during the right samples).

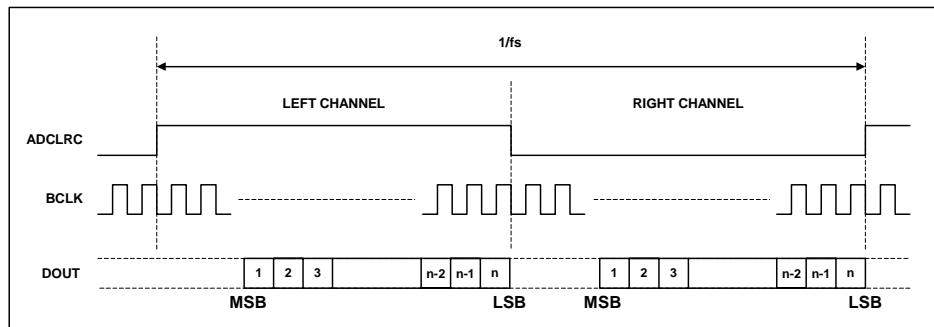


Figure 10 Right Justified Mode Timing Diagram

**I<sup>2</sup>S MODE**

In I<sup>2</sup>S mode, the MSB of the ADC data is output on DOUT and changes on the first falling edge of BCLK following an ADCLRC transition and may be sampled on the rising edge of BCLK. ADCLRC is low during the left samples and high during the right samples.

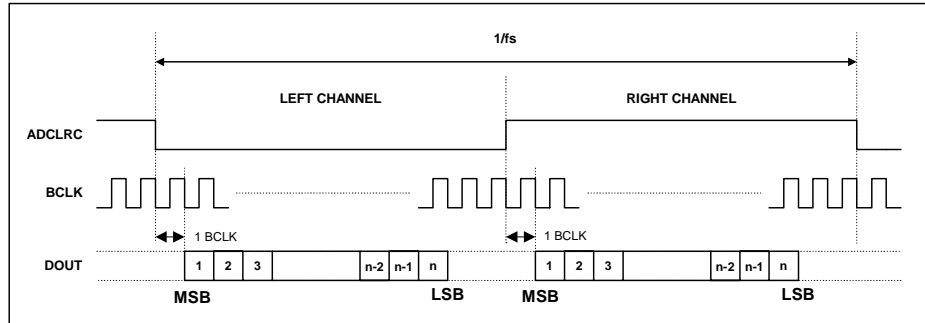


Figure 11 I<sup>2</sup>S Mode Timing Diagram

**DSP EARLY MODE**

The MSB of the left channel ADC data is output on DOUT and changes on the first falling edge of BCLK following a low to high ADCLRC transition and may be sampled on the rising edge of BCLK. The right channel ADC data is contiguous with the left channel data (Figure 12)

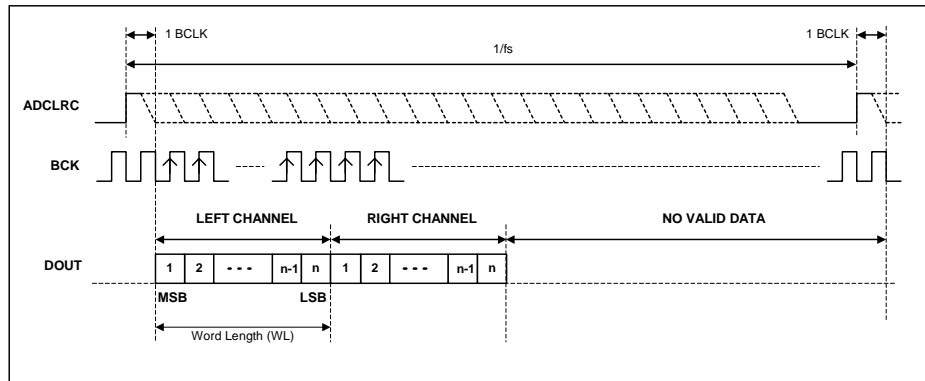


Figure 12 DSP Early Mode Timing Diagram – ADC Data Output



**DSP LATE MODE**

The MSB of the left channel ADC data is output on DOUT and changes on the same falling edge of BCLK as the low to high ADCLRC transition and may be sampled on the rising edge of BCLK. The right channel ADC data is contiguous with the left channel data (Figure 13).

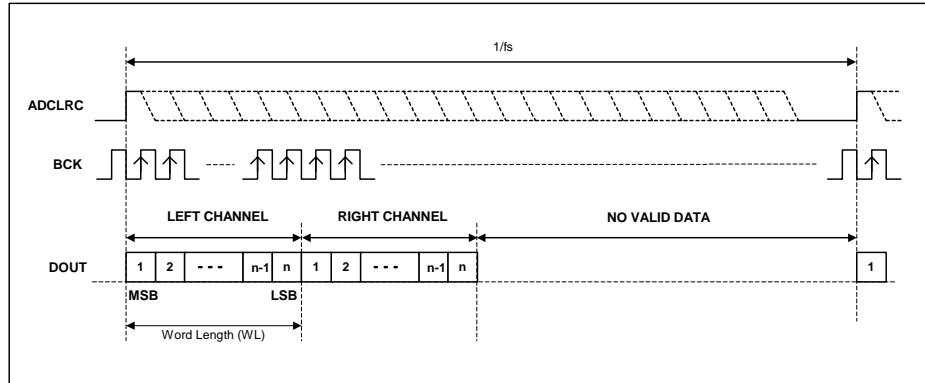


Figure 13 DSP Late Mode Timing Diagram – ADC Data Output

**CONTROL INTERFACE OPERATION**

The WM8773 is controlled using a 3-wire SPI compatible serial configuration.

The control interface is 5V tolerant, meaning that the control interface input signals CE, CL and DI may have an input high level of 5V while DVDD is 3V. Input thresholds are determined by DVDD. RESETB is also 5V tolerant.

**3-WIRE (SPI COMPATIBLE) SERIAL CONTROL MODE**

DI is used for the program data, CL is used to clock in the program data and CE is used to latch the program data. DI is sampled on the rising edge of CL. The 3-wire interface protocol is shown in Figure 14.

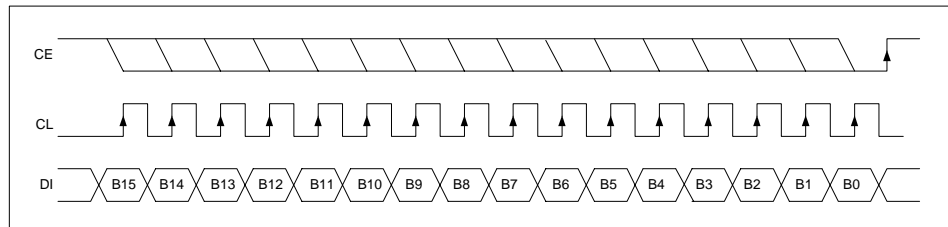


Figure 14 3-wire SPI compatible Interface

**Note:**

1. B[15:9] are Control Address Bits
2. B[8:0] are Control Data Bits
3. CE is edge sensitive – the data is latched on the rising edge of CE.

## CONTROL INTERFACE REGISTERS

### DIGITAL AUDIO INTERFACE CONTROL REGISTER

Interface format is selected via the FMT[1:0] register bits:

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
10110 Interface Control	1:0	FMT[1:0]	10	Interface format Select 00 : right justified mode 01: left justified mode 10: I <sup>2</sup> S mode 11: DSP (early or late) mode

In left justified, right justified or I<sup>2</sup>S modes, the LRP register bit controls the polarity of ADCLRC. If this bit is set high, the expected polarity of ADCLRC will be the opposite of that shown Figure 9, and. Note that if this feature is used as a means of swapping the left and right channels, a 1 sample phase difference will be introduced. In DSP modes, the LRP register bit is used to select between early and late modes.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
10110 Interface Control	2	LRP	0	In left/right/I <sup>2</sup> S modes: ADCLRC Polarity (normal) 0 : normal ADCLRC polarity 1: inverted ADCLRC polarity
				In DSP mode: 0 : Early DSP mode 1: Late DSP mode

By default, ADCLRC is sampled on the rising edge of BCLK and should ideally change on the falling edge. Data sources that change ADCLRC on the rising edge of BCLK can be supported by setting the BCP register bit. Setting BCP to 1 inverts the polarity of BCLK to the inverse of that shown in Figure 9, Figure 10, Figure 11, Figure 12, and Figure 13.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
10110 Interface Control	3	BCP	0	BCLK Polarity (DSP modes) 0 : normal BCLK polarity 1: inverted BCLK polarity

The WL[1:0] bits are used to control the word length.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
10110 Interface Control	5:4	WL[1:0]	10	Word Length 00 : 16 bit data 01: 20 bit data 10: 24 bit data 11: 32 bit data

**Note:**

1. If 32-bit mode is selected in right justified mode, the WM8773 defaults to 24 bits.
2. In 24 bit I<sup>2</sup>S mode, any width of 24 bits or less is supported provided that ADCLRC is high for a minimum of 24 BCLKs and low for a minimum of 24 BCLKs.

Control bit MS selects between audio interface Master and Slave Modes. In Master mode ADCLRC and BCLK are outputs and are generated by the WM8773. In Slave mode ADCLRC and BCLK are inputs to WM8773.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
10111 Interface Control	8	MS	0	Audio Interface Master/Slave Mode select: 0 : Slave Mode 1: Master Mode

#### MASTER MODE ADCLRC FREQUENCY SELECT

In Master mode the WM8773 generates ADCLRC and BCLK. These clocks are derived from master clock and the ratio of MCLK to ADCLRC and are set by ADCRATE.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
0111 ADCLRC frequency select	2:0	ADCRATE[2:0]	010	Master Mode MCLK:ADCLRC ratio select: 010: 256fs 011: 384fs 100: 512fs 101: 768fs

#### ADC OVERSAMPLING RATE SELECT

For ADC operation at 96kHz it is recommended that the user set the ADCOSR bit. This changes the ADC signal processing oversample rate to 64fs.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
10111 ADC Oversampling Rate	3	ADCOSR	0	ADC oversampling rate select 0: 128x oversampling 1: 64x oversampling

#### MUTE MODES

Each ADC channel has an individual mute control bit, which mutes the input to the ADC. In addition both channels may be muted by setting ADCMUTE.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
11001 ADC Mute	7	ADCMUTE	0	ADC MUTE Left and Right 0 : Normal Operation 1: mute ADC left and ADC right
11001 ADC Mute Left	5	MUTE	0	ADC Mute select 0 : Normal Operation 1: mute ADC left
11010 ADC Mute Right	5	MUTE	0	ADC Mute select 0 : Normal Operation 1: mute ADC right

The Record outputs may be enabled by setting RECEN, where RECEN enables the RECL and RECR outputs.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
10100 REC Enable	5	RECEN	0	REC Output Enable 0 : REC output muted 1: REC output enabled

### POWERDOWN MODE AND ADC DISABLE

Setting the PDWN register bit immediately powers down the WM8773, including the references, overriding all other powerdown control bits. All trace of the previous input samples are removed, but all control register settings are preserved. When PDWN is cleared the digital filters will be reinitialised. It is recommended that the 8-channel input mux and buffer, and ADC are powered down before setting PDWN.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
11000 Powerdown Control	0	PDWN	0	Power Down Mode Select: 0 : Normal Mode 1: Power Down Mode

The ADC may also be powered down by setting the ADCD disable bit. Setting ADCD will disable the ADC and select a low power mode. The ADC digital filters will be reset and will reinitialise when ADCD is reset.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
11000 ADC Powerdown Control	1	ADCD	1	ADC Disable: 0 : Normal Mode 1: Power Down Mode

### ADC GAIN CONTROL

Control bits LAG[4:0] and RAG[4:0] control the ADC input gain, allowing the user to attenuate the ADC input signal to match the full-scale range of the ADC. The gain is independently adjustable on left and right inputs. Left and right inputs may also be independently muted. The LRBOTH control bit allows the user to write the same attenuation value to both left and right volume control registers. The ADC volume and mute also applies to the bypass signal path.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
11001 Attenuation ADCL	4:0	LAG[4:0]	01100 (0dB)	Attenuation data for Left channel ADC gain in 1dB steps. See Table 8
	5	MUTE	0	Mute for Left channel ADC: 0: Mute off 1: Mute on
	6	LRBOTH	0	Setting LRBOTH will write the same gain value to LAG[4:0] and RAG[4:0]
11010 Attenuation ADCR	4:0	RAG[4:0]	01100 (0dB)	Attenuation data for right channel ADC gain in 1dB steps. See Table 8
	5	MUTE	0	Mute for Right channel ADC: 0: Mute off 1: Mute on
	6	LRBOTH	0	Setting LRBOTH will write the same gain value to RAG[4:0] and LAG[4:0]

**ADC INPUT GAIN**

Registers LAG and RAG control the left and right channel gain into the stereo ADC in 1dB steps from +19dB to -12dB Table 8 shows how the attenuation levels are selected from the 5-bit words.

L/RAG[6:0]	ATTENUATION LEVEL
0	-12dB
:	:
01100	0dB
:	:
11111	+19dB

**Table 8 ADC Gain Control****ADC HIGHPASS FILTER DISABLE**

The ADC digital filters contain a digital highpass filter. This defaults to enabled and can be disabled using software control bit ADCHPD.

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
10110 ADC control	8	ADCHPD	0	ADC Highpass filter disable: 0: Highpass filter enabled 1: Highpass filter disabled

**ADC INPUT MUX AND POWERDOWN CONTROL**

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
11011 ADC mux and powerdown control	2:0	LMX[2:0]	000	ADC left channel input mux control bits (see Table 9)
	6:4	RMX[2:0]	000	ADC right channel input mux control bits (see Table 9)
	8	AINPD	1	Input mux and buffer powerdown 0: Input mux and buffer enabled 1: Input mux and buffer powered down

Register bits LMX and RMX control the left and right channel inputs into the stereo ADC. The default is AIN1. However if the analogue input buffer is powered down, by setting AINPD, then all 8-channel mux inputs are switched to buffered VMIDADC.

LMX[2:0]	LEFT ADC INPUT	RMX[2:0]	RIGHT ADC INPUT
000	AIN1L	000	AIN1R
001	AIN2L	001	AIN2R
010	AIN3L	010	AIN3R
011	AIN4L	011	AIN4R
100	AIN5L	100	AIN5R
101	AIN6L	101	AIN6R
110	AIN7L	110	AIN7R
111	AIN8L	111	AIN8R

**Table 9 ADC Input Mux Control**

## SOFTWARE REGISTER RESET

Writing to register 11111 will cause a register reset, resetting all register bits to their default values.

## REGISTER MAP

The complete register map is shown below. The detailed description can be found in the relevant text of the device description. The WM8773 can be configured using the Control Interface. All unused bits should be set to '0'.

REGISTER	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0	DEFAULT
R20(14h)	0	0	1	0	1	0	0	x	x	x	RECEN	x	x	x	x	x	00000000
R22(16h)	0	0	1	0	1	1	0	ADCHPD	x	x	WL[1:0]		BCP	LRP	FMT[1:0]		000100010
R23(17h)	0	0	1	0	1	1	1	MS	x	x	x	x	ADCOSR	ADCRATE[2:0]			000010010
R24(18h)	0	0	1	1	0	0	0	x	x	x	x	x	x	x	ADCD	PWDN	000111110
R25(19h)	0	0	1	1	0	0	1	x	ADCMUTE	LRBOT H	MUTE	LAG[4:0]					000001100
R26(1Ah)	0	0	1	1	0	1	0	x	x	LRBOT H	MUTE	RAG[4:0]					000001100
R27(1Bh)	0	0	1	1	0	1	1	AINPD	x	RMX[2:0]			x	LMX[2:0]			100000000
R31(1Fh)	0	0	1	1	1	1	1	RESET									not reset
	ADDRESS							DATA									DEFAULT

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
10100 Mute	5	RECEN	0	REC Output Enable 0 : REC output muted 1: REC output enabled
10110 Interface Control	1:0	FMT[1:0]	10	Interface format select 00: right justified mode 01: left justified mode 10: I <sup>2</sup> S mode 11: DSP mode
	2	LRP	0	ADCLRC Polarity or DSP Early/Late mode select Left Justified / Right Justified / I <sup>2</sup> S 0: Standard ADCLRC Polarity 1: Inverted ADCLRC Polarity
				DSP Mode 0: Early DSP mode 1: Late DSP mode
	3	BCP	0	BITCLK Polarity 0: Normal - ADCLRC sampled on rising edge of BCLK; DOUT changes on falling edge of BCLK. 1: Inverted - ADCLRC sampled on falling edge of BCLK; DOUT changes on rising edge of BCLK.
	5:4	WL[1:0]	10	Input Word Length 00: 16-bit Mode 01: 20-bit Mode 10: 24-bit Mode 11: 32-bit Mode (not supported in right justified mode)
	8	ADCHPD	0	ADC Highpass Filter Disable: 0: Highpass Filter enabled 1: Highpass Filter disabled
10111 Master Mode control	2:0	ADCRATE[2:0]	010	Master Mode MCLK:ADCLRC ratio select: 010: 256fs 011: 384fs 100: 512fs
	3	ADCOSR	0	ADC oversample rate select 0: 128x oversampling 1: 64x oversampling
	8	MS	0	Maser/Slave interface mode select 0: Slave Mode – ADCLRC and BCLK are inputs 1: Master Mode – ADCLRC and BCLK are outputs
11000 Powerdown Control	0	PWDN	0	Chip Powerdown Control (works in together with ADCD) 0: All circuits running, outputs are active 1: All circuits in power save mode, outputs muted
	1	ADCD	1	ADC powerdown: 0: ADC enabled 1: ADC disabled
11001 Attenuation ADCL	4:0	LAG[4:0]	01100 (0dB)	Attenuation data for left channel ADC gain in 1dB steps
	5	MUTE	0	Mute for Left channel ADC: 0: Mute off 1: Mute on
	6	LRBOTH	0	Setting LRBOTH will write the same gain value to LAG[4:0] and RAG[4:0]
	7	ADCMUTE	0	Mute for Left and Right channel ADC: 0: Mute off 1: Mute on

REGISTER ADDRESS	BIT	LABEL	DEFAULT	DESCRIPTION
11010 Attenuation ADCR	4:0	RAG[4:0]	01100 (0dB)	Attenuation data for right channel ADC gain in 1dB steps
	5	MUTE	0	Mute for Right channel ADC: 0: Mute off 1: Mute on
	6	LRBOTH	0	Setting LRBOTH will write the same gain value to RAG[4:0] and LAG[4:0]
11011 ADC mux control	2:0	LMX[2:0]	000	ADC left channel input mux control bits
	6:4	RMX[2:0]	000	ADC right channel input mux control bits
	8	AINPD	1	Input mux and buffer powerdown 0: Input mux and buffer enabled 1: Input mux and buffer powered down
11111 Software reset	[8:0]	RESET	Not reset	Writing to this register will apply a reset to the device registers.

Table 10 Register Map Description



### DIGITAL FILTER CHARACTERISTICS

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>ADC Filter</b>					
Passband	±0.01 dB	0		0.4535fs	
	-6dB		0.5fs		
Passband ripple				±0.01	dB
Stopband		0.5465fs			
Stopband Attenuation	f > 0.5465fs	-65			dB
Group Delay			22		fs

Table 11 Digital Filter Characteristics

### ADC FILTER RESPONSES

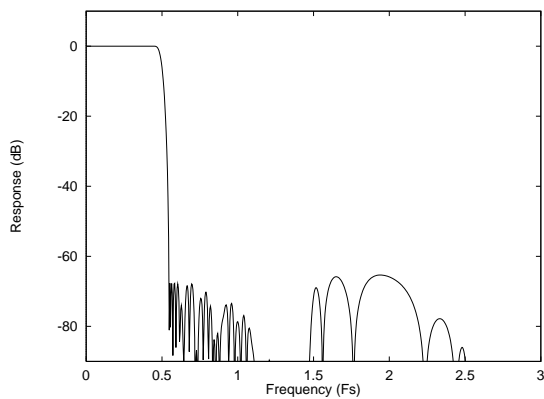


Figure 15 ADC Digital Filter Frequency Response

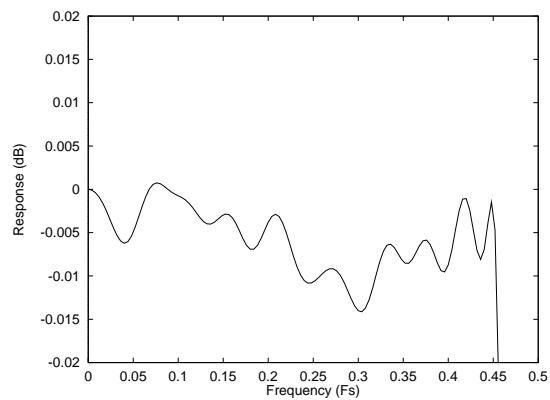


Figure 16 ADC Digital Filter Ripple

### ADC HIGH PASS FILTER

The WM8773 has a selectable digital highpass filter to remove DC offsets. The filter response is characterised by the following polynomial.

$$H(z) = \frac{1 - z^{-1}}{1 - 0.9995z^{-1}}$$

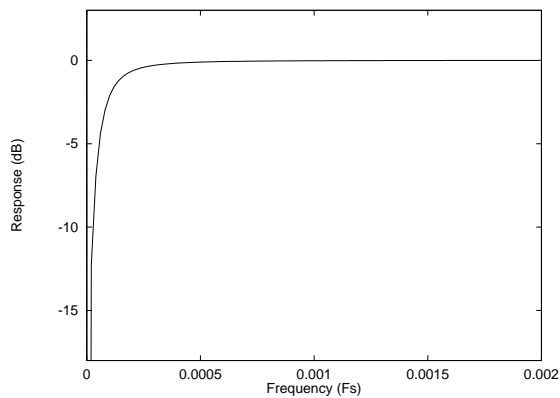


Figure 17 ADC Highpass Filter Response

## EXTERNAL CIRCUIT CONFIGURATION

In order to allow the use of 2V rms and larger inputs to the ADC inputs, a structure is used that uses external resistors to drop these larger voltages. This also increases the robustness of the circuit to external abuse such as ESD pulse.

Figure 18 shows the ADC input multiplexor circuit with external components allowing 2Vrms inputs to be applied.

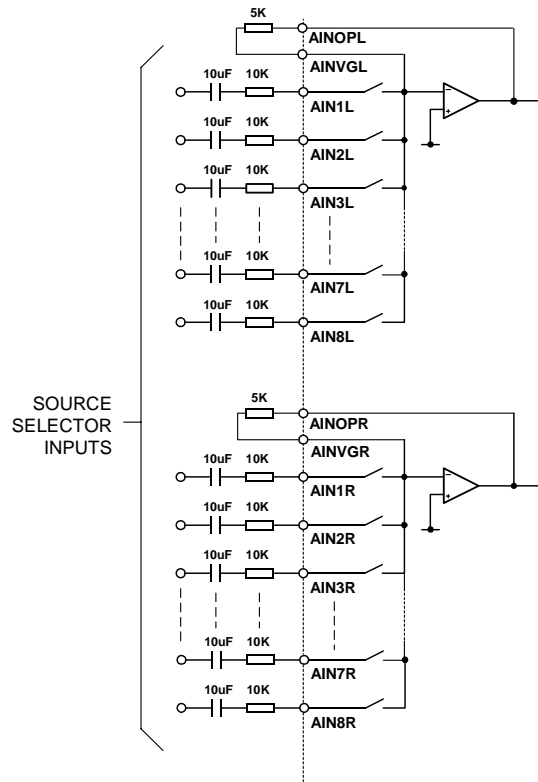
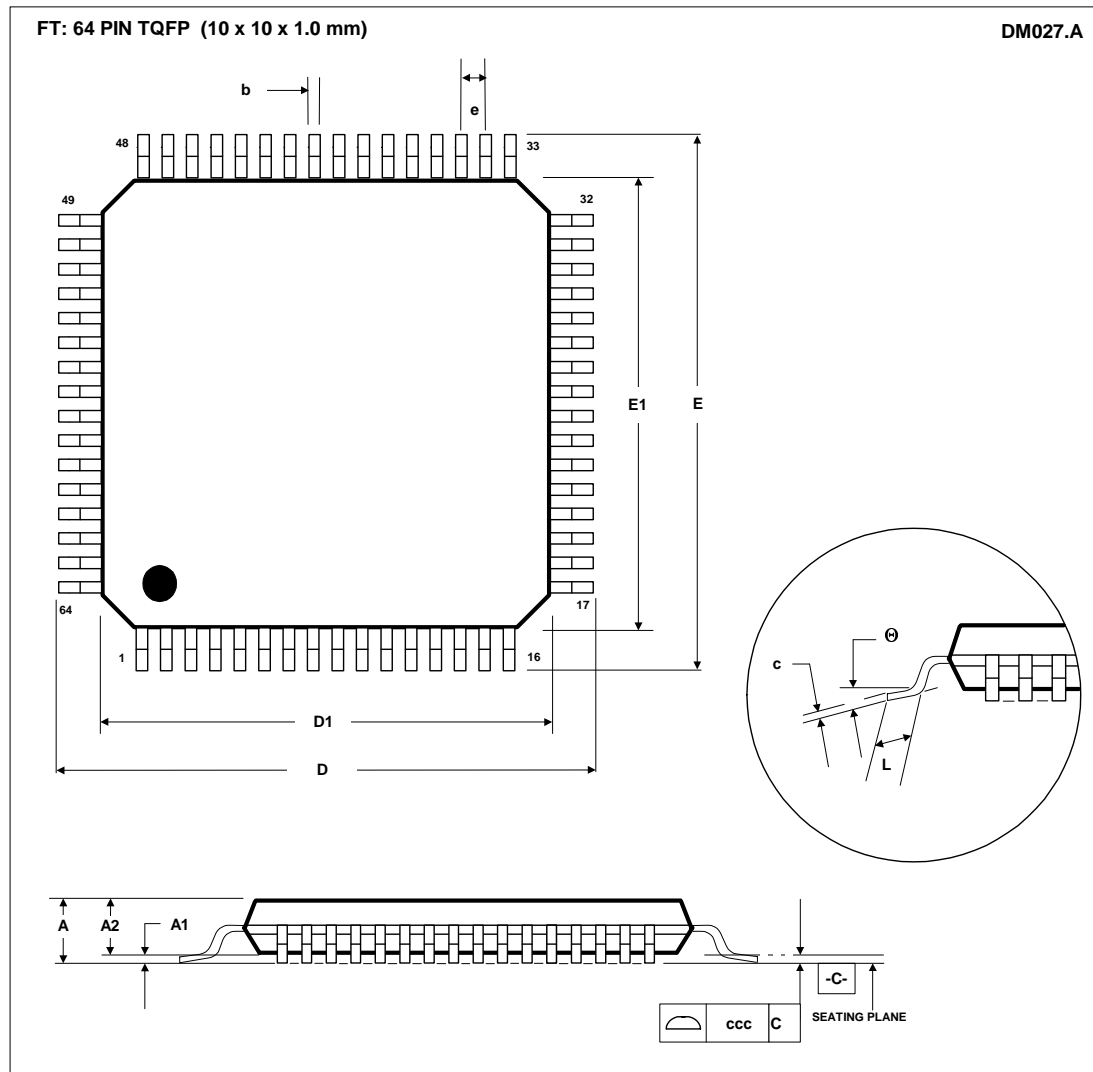


Figure 18 ADC Input Multiplexor Configuration

**PACKAGE DIMENSIONS**



Symbols	Dimensions (mm)		
	MIN	NOM	MAX
A	-----	-----	1.20
A <sub>1</sub>	0.05	-----	0.15
A <sub>2</sub>	0.95	1.00	1.05
b	0.17	0.22	0.27
c	0.09	-----	0.20
D	12.00 BSC		
D <sub>1</sub>	10.00 BSC		
E	12.00 BSC		
E <sub>1</sub>	10.00 BSC		
e	0.50 BSC		
L	0.45	0.60	0.75
$\theta$	0°	3.5°	7°
Tolerances of Form and Position			
ccc	0.08		
REF:	JEDEC.95, MS-026		

NOTES:  
 A. ALL LINEAR DIMENSIONS ARE IN MILLIMETERS.  
 B. THIS DRAWING IS SUBJECT TO CHANGE WITHOUT NOTICE.  
 C. BODY DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSION. NOT TO EXCEED 0.25MM.  
 D. MEETS JEDEC.95 MS-026, VARIATION = ACD. REFER TO THIS SPECIFICATION FOR FURTHER DETAILS.

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**REVISION HISTORY**

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