

STB5701

350 to 400 MHz FSK/ASK receiver (ST-RECORD01 family)

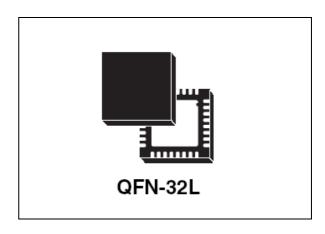
Preliminary Data

Features

- Multiband receiver: 350MHz to 400MHz
- FSK/ASK modulation selection
- Programmable multichannel
- High dynamic range with On-Chip AGC
- PLL and fully VCO integrated
- Image Rejection Mixer integrated
- Start Code Detector block (SCD)
- I2C serial interface (standard mode/fast mode)
- BiCMOS SiGe technology
- VQFN package 5 x 5 mm

Applications

- 350 to 400 MHz ISM Band System
- Set Top Box



Description

This device is a single chip FSK/ASK receiver optimized for licence-free ISM band operations from 350MHz to 400 MHz. It can easily be configured to provide the optimal solution for the user's application like Set Top Box and 350/400 MHz ISM Band Systems.

Table 1. Device summary

Paskago	Order codes	
Package	Tray	Tape & Reel
QFN32L	STB5701	STB5701TR

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1 Block diagram and pin configuration

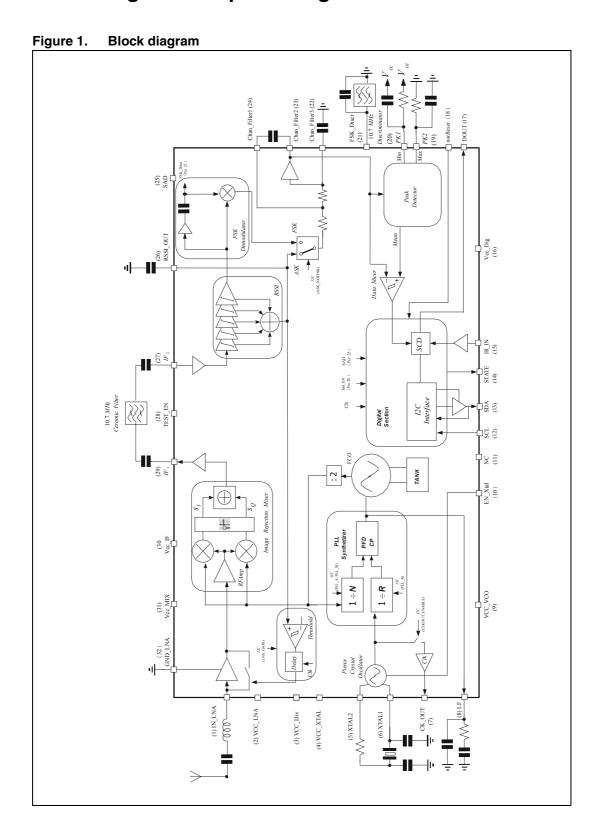


Table 2. Pin description

Pin	Name	I/O type	I/O schematic	Description
32	GND_LNA	GND	DI, LINA Free GND, LINA	Low Noise Amplifier ground
1	IN_LNA	Ţ	<u> </u>	Low Noise Amplifier input
2	Vcc_LNA	Vcc		Low Noise Amplifier supply voltage
3	Vcc_Div	Vcc		Divider supply voltage
4	Vcc_XTAL	Vcc		Crystal supply voltage
5	XTAL2	0	XTAL2	2 nd Crystal input
6	XTAL1	I	FCC FCC XTALL TO THE TOTAL TO THE TOTAL TO	1 st Crystal input
7	CK_OUT	0	The cx_out	Clock output
8	LPF	I/O	LPF Proc	Loop filter
9	Vcc_VCO	Vcc		VCO supply voltage
10	EN_XTAL	I	EN, Xual	Crystal oscillator enable
11	NC			Not connected

Table 2. Pin description (continued)

Pin	Name	I/O type	I/O schematic	Description
12	SCL	1	SCI.	I2C clock
13	SDA	-	SDA DE FEE	I2C data In/Out
14	STATE	0	Fice STATE	High when selecting UHF
15	IR_IN	-	FCC FCC FCC	From external IR source
16	Vcc_Dig	Vcc		Digital section supply voltage
17	DOUT	0	DOUT	Digital serial output
18	NotReset	I	Fice Fice Fice Fice Fice Fice Fice Fice	Global reset, active low

Table 2. Pin description (continued)

Pin	Name	I/O type	I/O schematic	Description
19	PK2	I/O	V _{cc} PK2 V _{cc}	Max. peak detector
20	PK1	I/O	Fee Property of the Property o	Min. peak detector
21	FSK_Discrim	I/O	FSK_Discr	FSK discriminator
22	Chan_Filter3	I/O	Chan_Fiber3	Channel filter
23	Chan_Filter2	I/O	Chan Filter2	Channel filter
24	Chan_Filter1	I/O	Chan_Filter1	Channel filter
25	SAD	ı	SAD - Fee	I2C address selection

Table 2. Pin description (continued)

Pin	Name	I/O type	I/O schematic	Description
26	RSSI_OUT	0	Vcc	Radio strength signal indicator output
27	IF2	1	F2 — Figure 1	External IF filter input
28	Test_EN	ı	TEST_EN Pec	Test mode enable
29	IF1	0	Vec Vec	External IF filter output
30	Vcc_IF	Vcc		IF supply voltage
31	Vcc_MIX	Vcc		Mixer supply voltage
EP	GND	GND		Ground

Figure 2. Test circuit

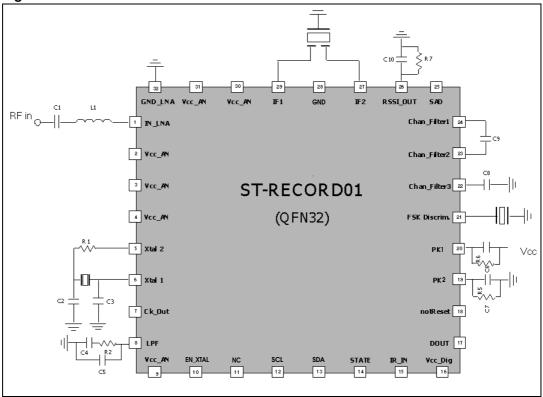


Table 3. Bill of material

Component	Size	Manufacturer	Part number	Description
C1	0603	MURATA	Series GRM39	LNA Input matching
C2	0603	MURATA	Series GRM39	Crystal load capacitor
C3	0603	MURATA	Series GRM39	Crystal load capacitor
C4	0603	MURATA	Series GRM39	Loop filter capacitor
C5	0603	MURATA	Series GRM39	Loop filter capacitor
C6	0603	MURATA	Series GRM39	Peak detector capacitor
C7	0603	MURATA	Series GRM39	Peak detector capacitor
C8	0603	MURATA	Series GRM39	FSK Discriminator Tuning
C9	0603	MURATA	Series GRM39	Channel filter capacitor
C10	0603	MURATA	Series GRM39	Channel filter capacitor
C11	0603	MURATA	Series GRM39	RSSI output low path capacitor
C12	0603	MURATA	Series GRM39	IF Filter DC Block
C13	0603	MURATA	Series GRM39	IF Filter DC Block
R1	0603	NEOHM	Series CRG0603	Resistor load capacitor
R2	0603	NEOHM	Series CRG0603	Loop filter resistor
R3	0603	NEOHM	Series CRG0603	Peak detector resistor
R4	0603	NEOHM	Series CRG0603	Peak detector resistor
R5	0603	NEOHM	Series CRG0603	RSSI output resistor
R6	0603	NEOHM	Series CRG0603	IF Filter Matching
R7	0603	NEOHM	Series CRG0603	IF Filter Matching
L1	0603	MURATA	LQP18M	LNA input matching
X1	CS10	CITIZEN	CS10- 27.000MABJTR	Crystal
F1		Toko	SK107M2N-A0-20X	Ceramic filter
F2		Toko	CDF107F-AO-022	Ceramic resonator

2 Electrical specifications

2.1 Absolute maximum ratings

Stressing the device above the rating listed in the "Absolute maximum ratings" table may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the Operating sections of this specification is not implied. Exposure to Absolute Maximum Rating conditions for extended periods may affect device reliability. Refer also to the STMicroelectronics SURE Program and other relevant quality document

Table 4. Absolute maximum ratings $(T_{amb} = 25^{\circ}C)$

Symbol	Parameter	Value	Unit
V _{CC}	Supply voltage	4.5	V
T _j	Junction temperature	-40 to 150	°C
T _{stg}	Storage temperature	-55 to + 150	οС

2.2 Thermal data

Table 5. Thermal data

Symbol	Parameter	Value	Unit
T _a	Operating temperature	0 + 70	°C
θ_{th}	Thermal resistance junction to ambient	40	°C/W

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3 Electrical characteristics

(V $_{CC}$ = 3.3 V, Z $_{\rm s}$ = 50 Ω , Ta = 25 $^{\rm o}$ C, unless otherwise specified).

Table 6. Electrical characteristics

Symbol	Parameters	Test conditions		Min.	Тур.	Max.	Unit
Vcc	Supply voltage			2.7	3.3	3.6	V
f _{rx}	Receive frequency range			350		400	MHz
Δf	FSK frequency deviation			25	37.5	50	kHz
I _{rx_FSK}	FSK receive current consumption	High g	ain LNA		25		mA
I _{cc_standby}	Standby current consumption	RF Enable bit=0	En_Xtal (PIN 10) high / floating		5		mA
	, , , , , , , , , , , , , , , , , , ,	J. C	En_Xtal (PIN 10) low		20		μΑ
	FSK data rate	(Manchest	er encoding)		2	10	kbps
RLin	Input return loss				10		dB
	Max RF power input					0	dBm
P _{min}	Input sensitivity	2 kbps manchester encoding BER ≤10 ⁻³ BW = 300KHz FSK 37.5 KHz deviation	$RF_1 = 369.5MHz$ $RF_2 = 371.1MHz$ $RF_3 = 375.3MHz$ $RF_4 = 376.9MHz$ $RF_5 = 388.3MHz$ $RF_6 = 391.5MHz$ $RF_7 = 394.3MHz$ $RF_8 = 395.9MHz$		-109		dBm
			RF _X +/-30KHz		-103		dBm
	Channel blocking ⁽¹⁾		IMHz 6MHz		50 50		dBc dBc
	Image Rejection				35		dB
I2C I/O p	ins						
V _{IH}	Input logic voltage high			0.7 Vcc		3.6	V
V _{IL}	Input logic voltage low			-0.5		0.3Vcc	V

Table 6. Electrical characteristics (continued)

Symbol	Parameters	Test conditions	Min.	Тур.	Max.	Unit
V _{OL}	Output logic voltage low				0.4	V
V _{HYST}	Hysteresis of schmitt trigger for inputs		0.05 Vcc			٧
t _{of}	Output fall time from V _{IHmin} to V _{ILmax} with a bus capacitance from 10pF to 400pF		20+ 0.1C _b		250	ns
l _i	Input current each I/O pin with an input voltage between 0.1V _{cc} and 0.9V _{cc}		-10		10	μΑ
C _i	Capacitance for each I/O pin				10	pF
C _b	Capacitive load for each bus line				400	pF
I2C Stan	dard mode		1		l	
f _{SCL}	SCL clock frequency		0		100	kHz
t _{SU_STA}	Setup time for START condition		4.7			μs
t _{HD_STA}	Hold time for START condition		4			μs
t _{HIGH}	SCL high time		4			μs
t_{LOW}	SCL low time		4.7			μs
t _{SU_DAT}	DATA setup time		250			ns
t _{HD_DAT}	DATA hold time		0		3.45	μs
t _R	SDA, SCL rise time				1000	ns
t _F	SDA, SCL fall time				300	ns
t _{SU_STO}	Setup time for STOP condition		4			μs
t _{BUF}	Bus free time between STOP and START condition		4.7			μs

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Table 6. Electrical characteristics (continued)

Symbol	Parameters	Test conditions	Min.	Тур.	Max.	Unit
I2C Fast	mode					
f _{SCL}	SCL clock frequency		0		400	kHz
t _{SU_STA}	Setup time for START condition		0.6			μs
t _{HD_STA}	Hold time for START condition		0.6			μs
t _{HIGH}	SCL high time		0.6			μs
t _{LOW}	SCL low time		1.3			μs
t _{SU_DAT}	DATA setup time		100			ns
t _{HD_DAT}	DATA hold time		0		0.9	μs
t _R	SDA, SCL rise time				300	ns
t _F	SDA, SCL fall time				300	ns
t _{SU_STO}	Setup time for STOP condition		0.6			μs
t _{BUF}	Bus free time between STOP and START condition		1.3			μs

^{1.} Desired signal 10dB above the input sensitivity level, CW interferer power level increased until BER ±10⁻³

4 Functional description

4.1 General description

The STB5701 FSK/ASK receiver is a heterodyne configuration (10.7 MHz IF Frequency) and it is designed for applications in the 350MHz to 400MHz frequency range and includes ASK and FSK detectors. The synthesizer has a typical channel spacing of better than 100 kHz and uses a integrated fully VCO. With the STB5701 receiver chip, various circuit configurations can be arranged in order to meet a number of different customer requirement.

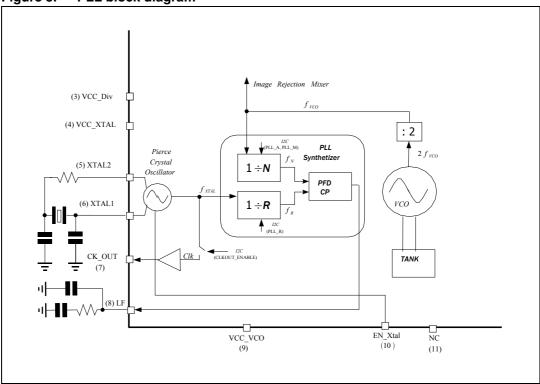
The STB5701 is housed in a VQF package 5mm x 5mm 32 leads.

The STB5701 receiver IC consists of the following building blocks:

- Phase Locked Loop Synthesizer (PLL): M-Counter, A-Counter, Prescaler, Phase-Frequency Detector (PFD), Charge Pump (CP), Voltage Controlled Oscillator (VCO).
- Programmable reference divider for crystal and channel step selection.
- Low Noise Amplifier (LNA) with AGC function for high sensitivity and dynamic range RF signal reception.
- Image Rejection Mixer for down conversion of the RF signal to the IF (without external saw image filter).
- IF amplifier to amplify and limit the IF signal and for RSSI generation.
- Phase coincidence demodulator to demodulate the IF signal.
- Operation amplifier for data slicing, filtering and ASK detection.
- Min/max peak detector to create a DC output voltage equal to the mean of the min and max peak value of the data signal.
- I2C bus to exchange data between the STB5701 and the micro.
- Start code detector (SCD) to recognize a valid data signal to prevent continuos toggling
 of data in the absence of an RF Signal due to the noise.
- Bias circuitry for band gap biasing and circuit shutdown.

4.2 Programmable Phase Locked Loop synthesizer (PLL)

Figure 3. PLL block diagram



The synthesized programmable local oscillator is a Phase Locked Loop (PLL) using a 'parallel-resonant' quartz crystal as frequency reference. The PLL block contains a phase detector, charge pump, VCO, Programmable N and R Divider and a Crystal Oscillator. The synthesized frequency (fVCO) is set by programming the 'N' Divider through the I2C interface (PINs 12, 13).

$$f_{R} = f_{XTAL} / R$$

$$f_{VCO} = N \cdot f_{N} = N \cdot f_{R} = (N / R) \cdot f_{XTAL}$$

4.2.1 Divider (a-counter, m-counter and prescaler)

The main divider (N) of the PLL contains a 3-bit A-counter, a 10-bit M-counter and an 8/9 prescaler. The divider ratio of the prescaler is controlled by the program counter and the swallow counter. During one cycle, the prescaler divides by 9 until the swallow A-counter reaches its terminal count. Afterwards the prescaler divider by 8 until the program counter reaches its terminal count. Therefore the overall feedback divider ratio can be expressed as:

$$N = 9 \cdot A + 8 \cdot (M - A)$$

The A-counter configuration represents the lower bits in the feedback divider register and the upper bits the M-counter configuration respectively. According to that, the following counter ranges are implemented:

$$0 \le A \le 7$$

$$7 \le M \le 1023$$

and therefore the range of the overall feedback divider ratio results in:

$$56 \le N \le 8191$$

The user does not need to care about the A- and M-counter settings. It is only necessary to know the overall feedback divider ratio N to program the register settings.

4.2.2 Reference divider

The reference divider reduces the frequency of the external crystal (FXTAL) to an internal reference frequency (FR) used for the phase-locked loop.

This value, corresponding to the channel step, it is set by programming a 10-bit counter through the I2C interface (PINs 12, 13).

Therefore the range of the Reference divider ratio is:

$$0 \le R \le 1023$$

4.2.3 Phase Frequency Detector

The phase detector (PFD) is a device that compares two input (fN and fR) phases, generating an output that is a measure of their phase difference.

The gain of the phase detector can be expressed as:

$$K_{PD} = I_{CP} / 2\pi$$

where ICP is the charge pump current.

If fR doesn't equal fN, the phase-error signal, after being filtered and amplified, causes the VCO frequency to deviate in the direction of fR.

4.2.4 Loop filter

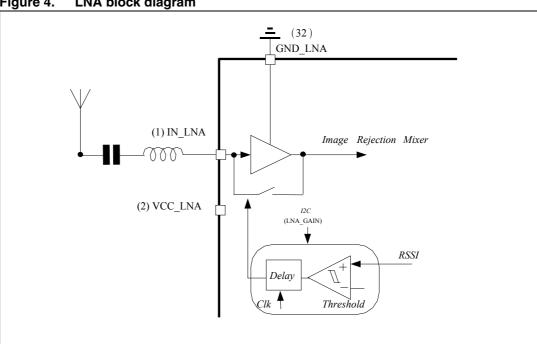
An external PLL loop filter is connected to pin LF (PIN 8). The loop filter controls the dynamic behavior of the PLL, primarily lock time and reference spur levels. Generally, the PLL lock time is a small fraction of the overall receiver start-up time. The crystal oscillator is the largest contributor to start-up time.

4.3 **Receiver section**

The integrated receiver is intended to be used as a single-conversion FSK/ASK receiver. It consists of a low noise amplifier, mixer, IF filter, limiter, FSK demodulator, a LPF amplifier, and a data slicer. The received strength signal indicator (RSSI) can be used for fast carrier sense detection or as amplitude shift keying, (ASK) demodulator.

4.3.1 Low Noise Amplifier (LNA)



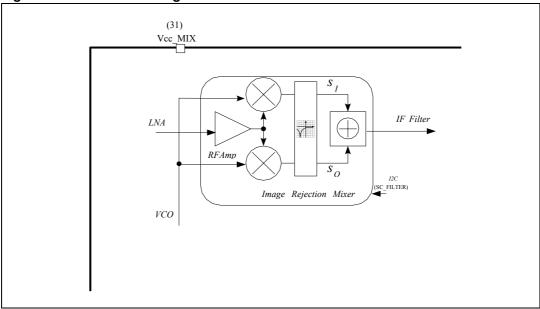


The LNA is based on a cascode topology for low-noise, high gain and good reverse isolation. The LNA output is directly connected to the mixer removing the external output matching.

It has 2 step gain managed by an internal AGC. This AGC circuit monitors the RSSI output. The AGC has a hysteresis of ~10dB.

4.3.2 Image Rejection Mixer

Figure 5. Mixer block diagram



An excellent feature of the STB5701 is the integrated image rejection mixer. This device was designed to eliminate the need for a costly front-end SAW filter for many applications. The advantage of not using a SAW filter is increased sensitivity, simplified antenna matching, less board space, and lower cost. The mixer cell is a pair of double-balanced mixers that perform an IQ down conversion of the 350-400MHz RF input to the IF (10.7MHz) with low-side injection (i.e., $f_{RF} = f_{LO} - f_{IF}$). The Image Rejection circuit combines these signals to achieve ~35dB of Image Rejection over the full temperature range. Low-side injection is required due to the on-chip Image Rejection architecture.

4.3.3 ASK/FSK demodulator and data filter

The received signal strength indicator (RSSI) voltage is proportional to the log of the down-converted RF signal at the IF limiting amplifier input. It also used as demodulator for amplitude-shift \ keying (ASK) modulation. The signal coming from the RSSI amplifier is converted into the raw data signal by the ASK/FSK demodulator. The ASK or FSK modulation selection is set by I2C interface.

The dynamic range of the RSSI amplifier is exceeded if the RF input signal is about 80 dB higher compared to the RF input signal at full sensitivity.

In FSK mode, the S/N ratio is not affected by the dynamic range of the RSSI amplifier but at FSK receive mode the RSSI output provides a field strength indication.

Coming from the RSSI the FSK signal is fed to the input of the FSK demodulator. After buffering the signal is fed to a phase discriminator. The phase shift is generated by an external 10.7MHz Ceramic discriminator connected to FSK_Discr. (PIN 21).

The FSK demodulator is intended to be used for an FSK deviation of 37.5 kHz. Lower values may be used but the sensitivity of the receiver is reduced in that condition. After demodulation a 2nd order Sallen and Key filter is provided in order to suppress unwanted frequency components.

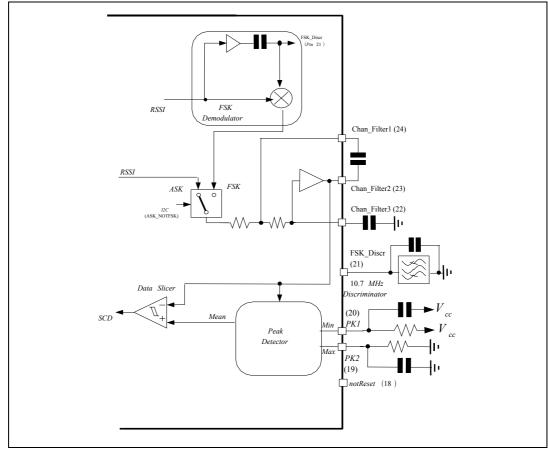


Figure 6. ASK/FSK demodulator block diagram

4.3.4 Min /Max peak detector

The peak detector embedded in the STB5701, in conjunction with an external RC filter (PINs 22, 23, 24), generates a DC output voltage equal to the mean of the min and max peak value of the data signal. The resistor provides a path for the capacitor to discharge, allowing the peak detector to dynamically follow peak changes of the data filter output voltage.

4.3.5 Data slicer

The purpose of the data slicer is to take the analog output of the data filter and convert it to a digital signal. This is achieved by using a comparator and comparing the analog input to a threshold voltage. One input is supplied by the data filter output.

The other path is fed to the min/max Peak detector to derive the average value (DC component) as an adaptive slice reference which is presented to the positive comparator input. The adaptive reference allows detecting the received data over a large range of noise floor levels.

4.4 Digital control and source selection

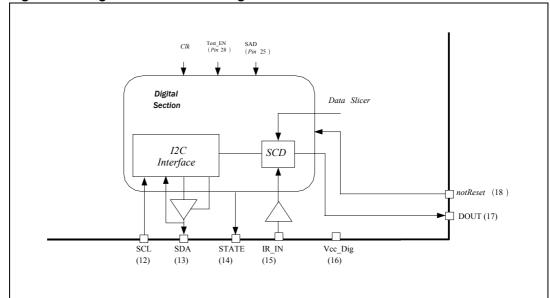


Figure 7. Digital section block diagram

The STB5701 digital section is responsible for the following functions:

- Configuration of the RF front end through software programmable config bits.
- Configuration of the PLL through software programmable registers (PLLA,M,R).
- Detection of a pre-programmed start code sequence (one of two possibilities).
- Arbitration between UHF and IR sources.

The serial data emerges from the DATA output pin (PIN 17) and the system state emerges from the STATE output pin (PIN 14).

4.4.1 State pin

The STATE output pin may be configured in one of two ways:

- To reflect the state of the start code detection block officiated = detected, 0 = searching or off), the state can be cleared by a programmable time out or a software reset.
- To operate as a true interrupt with associate status bit and a software clearing mechanism.

The state output may be optionally inverted.

4.4.2 Data pin

The DATA output pin is driven by one of two sources:

- The IR_IN pin.
- The UHF input from the RF front end.

4.4.3 MODE register

The DATA source selected at any time is governed by the MODE register, the options are as follows:

- IR_IN only, direct connection.
- UHF only, direct connection.
- IR triggered, the output is zero and switched to UHf on detection of a valid start code.
- IR/UHF arbitrated, the output comes by default from IR_IN, however on detection of a valid start code the output is switched to select the UHF source. The switch back to IR is accomplished in one of two ways.

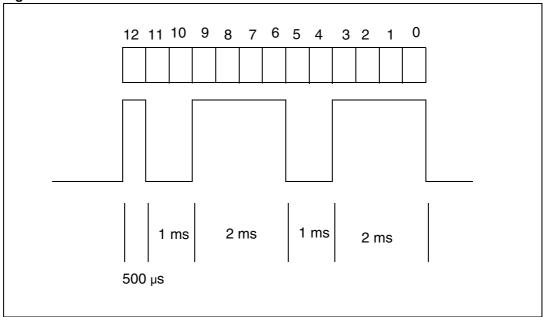
A software reset is programmed.

The time-out counter has been set and has timed out.

4.5 Start code detection example

This section explains the operation of the start code detector with reference to the start code sequence illustrated in the following figure:

Figure 8. START code



Firstly we assume the nominal symbol duration is $500 \,\mu s$, in this example there are then 13 symbol (denoted as 12 to 0 in the figure above).

If the device is operating from a 27 Mhz external clock then we program the prescaler to 27. The internal sampling clock will be 1Mhz and the sampling resolution will be 1us. To add some error tolerance we also program an upper and lower bound on the symbol duration as well as the nominal value. In this case 400 (corresponding to 400us) in SCD_symbol_min_time register, 500 (corresponding to 500us) in SCD_symbol_nom_time and 600 (corresponding to 600us) in SCD_symbol_max_time.

Now we program 0b1001111001111 into the SCD_code register and 13 (0x0d) corresponding to 13 symbols to be detected into the SCD_code_length register. The start code detection is started by setting the enable and research bits in SCD_config register to '1'.

The start code detector checks for the minimum symbol time of each symbol and the sequence in which symbols are received. If the symbol time is not respected by the input (incase of noise) the start code detection is re-initialized.

The detector is capable of simultaneously searching for two different start codes, the alternative code is written into the upper 16 bits of SCD_code register.

4.6 Registers list

Two unique I2C addresses have been reserved for the STB5701 device, the appropriate one is selected using the SAD input pin.

Table 7. I2C reserved addresses

SAD		Address					
0	0110010x	Unique to this device.					
1 (spare)	0110011x	Offique to this device.					

4.7 Register summary

The I2C mapped address space of the STB5701 is given below:

Table 8. I2C address breakdown

Base address	Group	Description
0x00	Top Level	See table 9 below
0x80	SCD Block	See table 10 below

Table 9. I2C mapped top level registers - summary

Address	Name	Reset value	Function
0x00	DIG_config	0x00000000	Digital configuration
0x04	RF_config	0x00000000	Configure RF front end
0x08	PLL_A	0x00000000	PLL post divider
0x0C	PLL_M	0x00000000	PLL feedback divider
0x10	PLL_R	0x00000000	PLL reference divider

Table 10. I2C mapped SCD registers - summary

Address	Address Name		Function
0x80	Scd_config	0x00000000	SCD configuration register.
0x84	Scd_status	0x00000000	SCD status can be read from this register.
0x88	Scd_code	0x00000000	Expected Start code is stored in this register.
0x8C	Scd_code_length	0x00000000	Length of the start code is stored in this register.
0x90	Scd_symbol_min_time	0x00000000	Minimum time of the symbol is stored in this register.
0x94	Scd_symbol_max_time	0x00000000	Maximum time of the symbol is stored in this register.
0x98	Scd_symbol_nom_time	0x00000000	Nominal symbol time is stored in this register.

Table 10. I2C mapped SCD registers - summary (continued)

Address	Name	Reset value	Function
0x9C	Scd_prescaler	0x00000001	Value of prescaler for sampling is stored.
0xA0	Interrupt_enable	0x00000000	Interrupt enable register value is stored in this register.
0xA4	Interrupt_clear	0x00000000	Interrupt clear register value is stored in this register.
0xA8	Timeout	0x00000000	Timeout for SCS based on prescaler tick.

4.8 Detailed display description

Table 11. DIG_config register format

	DIG_Config		0x00 R/W		
Bit	Bit field	Reset state	Function		
1:0	MODE	00	00: UHF direct 01: IR direct 10: UHF triggered (DOUT = 0 when no trigger) 11: IR or UHF triggered		
2	STATE_SEL	0	STATE is driven by the SCD interrupt output STATE is driven by the SCD detect state		
3	STATE_INV	0	1: Invert the STATE output		
4	STATE_DRIVE	0	0: Open drain 1: Push pull		
5	CLKOUT_ENABLE	0	1: Enable clock output buffer		
31:6	Reserved	0x00	0x00		

Table 12. RF_Config register format

RF_Config			0x04 R/W		
Bit	Bit field	Reset state	Function		
0	RF_ENABLE	0	Enable front end Disable front end		
1	ASK_NOTFSK	0	1: ASK detection 0: FSK detection		
2	LNA_GAIN	0	1: Enable AGC 0: Disable AGC (high gain)		
3	Not used	0			
31:4	Reserved	0x00	0x00		

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Table 13. PLL_A register

	PLL_A		0x08	R/W
Bit	Bit field	Reset state	Function	
2:0	PLL_A	SAD = 0: 0x1 SAD= 1: 0x7	PLL post divider value. Reset value determined by level on SAD input.	
31:3	Reserved	0x00	0x00	

Table 14. PLL_M register format

PLL_M			0x0C	R/W
Bit	Bit field	Reset state	Function	
9:0	PLL_M	SAD = 0: 0x1D5 SAD= 1: 0x1CD	PLL feedback divider value. Reset value determined by level on SAD	input.
31:10	Reserved	0x00	0x00	

Table 15. PLL_R register format

PLL_R			0x10	R/W
Bit	Bit field	Reset state	Function	
9:0	PLL_R	000	PLL reference divider value	
31:10	Reserved	0x00	0x00	

Table 16. SCD_config register format

SCD_config			0x80 R/W
Bit	Bit field	Reset state	Function
0	Enable	0	Enables the start code detection circuit By passes SCD, UHF in fed to UHF out
1	Re-search	0	1: Start a fresh research. Reset automatically to '0' on re-search start
2	Soft_rst	0	1: Resets all the counters and shift register
3	Reset_shift_reg	0	1: Reset only shift register. Status register is not affected.
31:4	Reserved	0x00	0x00

Table 17. SCD_status register format

S	CD_status	0x84		R
Bit	Bit field	Reset state	Function	
0	Detect	0	1: Start code detected, UHF in fed to UHF out	
1:2	Reserved	0	00	
3	Alternative	0	1: The alternative code was detected	
31:4	Reserved	0x00	0x00	

Table 18. SCD_code register format

SCD_code			0x88	R/W
Bit	Bit field	Reset state	Function	
15:0	Code	0x00	Start code to be detected	
31:16	Alt_code	0x00	Alternative start code to be detected	

Note: This register holds each start code to be detected.

Table 19. SCD_code_lenth register format

SCD_code_lenth			0x8C	R/W
Bit	Bit field	Reset state	Function	
4:0	Code_length	0	Length of the start code	
15:5	Reserved	0x00	0x00	
20:16	Alt_code_length	0	Length of the alternative start code	
31:21	Reserved	0x00	0x00	

Note:

The length of the start code is stored in this register. If the start code length is 10 symbol then 10 has to be written into this register. Writing 0x00 disables the SCD.

Table 20. SCD_symbol_min_time register format

SCD_symbol_min_time			0x90	R/W
Bit	Bit field	Reset state	Function	
15:0	Min_time	0x00	Minimum symbol time	
31:16	Reserved	0x00	0x00	

Note:

The minimum time of the symbol in terms of pre-scaler ticks is stored in this register. If any symbol violates the minimum symbol time, the SCD process is re-initialized. If a value 0x10 is written into this register, the symbol min. time is 16 pre-scaler clock periods.

Table 21. SCD_symbol_max_time register format

SCD_symbol_max_time			0x94 R/W	
Bit	Bit field	Reset state	Function	
15:0	Max_time	0x00	Maximum symbol time	
31:16	Reserved	0x00	0x00	

Note:

The maximum time of the symbol is stored in this register. Any changes in the input data are allowed only between symbol minimum time and symbol maximum time. The symbol time counting is done by a clock (enable pulse) which ia output of pre-scaler. If a value 0x10 is written into this register, the symbol max time is 16 pre-scaler clock periods.

Table 22. SCD_symbol_nom_time register format

SCD_sy	SCD_symbol_nom_time		0x98	R/W
Bit	Bit field	Reset state	Function	
15:0	Nom_time	0x00	Nominal symbol time	
31:16	Reserved	0x00	0x00	

Note:

The nominal time of the symbol in terms of pre-scaler ticks is stored in this register. This value is used to register a new symbol when consecutive symbols with same logical value are received. If a value 0x10 is written into this register, the symbol nominal time is 16 pre-scaler clock periods.

Table 23. SCD_prescaler register format

SCI)_prescaler		0x9C	R/W
Bit	Bit field	Reset state Function		
15:0	prescaler	0x01	Pre scaler division value is stored in the	nis register
31:16	Reserved	0x00	0x00	

Note:

The nominal time of the symbol is stored in this register.

Table 24. Interrupt_enable register format

Interrupt_enable			0xA0	R/W
Bit	Bit field	Reset state	Function	
0	scd_detected	0	Enable interrupt Disable interrupt	
31:1	Reserved	0x00	0x00	

Note:

The interrupt enable register value is stored in this register.

Table 25. Interrupt_clear register format

Interrupt_clear			0xA4	R/W
Bit	Bit field	Reset state	Function	
0	scd_clear_int	0	Clear interrupt No change on interrupt	
31:1	Reserved	0x00	0x00	

Note:

The interrupt clear register value is stored in this register.

Table 26. Timeout register format

Bitcount			0xA8	R/W
Bit	Bit field	Reset state	leset state Function	
23:0	Timeout	0x00	Timeout value, duration is based on prescaler tick.	
31:24	Reserved	0x00	0x00	

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5 Package information

5.1 ECOPACK® packages

In order to meet environmental requirements, ST offers these devices in ECOPACK[®] packages. These packages have a Lead-free second-level interconnect. The category of Second-Level Interconnect is marked on the package and on the inner box label, in compliance with JEDEC Standard JESD97. The maximum ratings related to soldering conditions are also marked on the inner box label.

ECOPACK is an ST trademark. ECOPACK specifications are available at: www.st.com.

5.2 QFN32L (5mm x 5mm) information

DIMENSIONS IN mm

BOTTOM VIEW

EXPOSED PAD

EXPOSED PAD

L 32x

D D/2
TOP VIEW

A3

SEATING PLANE

LEADS COPLAMARITY

Figure 9. QFN32L (5mm x 5mm) dimensions

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Table 27. QFN32L (5mm x 5mm) mechanical data

	mm		
	Min.	Тур.	Max.
А	0.80	0.90	1.00
A1	0.00	0.02	0.05
А3		0.20	
b	0.18	0.25	0.30
D	4.85	5.00	5.15
D2	3.50	3.60	3.70
E	4.85	5.00	5.15
E2	3.50	3.60	3.70
е		0.50	
L	0.30	0.40	0.50
ddd			0.05

Revision history STB5701

6 Revision history

Table 28. Document revision history

Date	Revision	Changes
01-Oct-2007	1	Initial release.

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