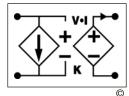


BCM[™] Bus Converter

- 48 V to 24 V V•I Chip[™] Converter
- 300 Watt (450 Watt for 1 ms)
- High density 1017 W/in³
- Small footprint 260 W/in²
- Low weight 0.5 oz (15 g)
- ZVS / ZCS isolated Sine Amplitude Converter

- Typical efficiency 96%
- 125°C operation (T_J)
- <1 µs transient response
- 3.5 million hours MTBF
- No output filtering required



Vin = 38 - 53 V Vout = 19.0 - 26.5 V Iout = 12.5 A K = 1/2 Rout = 60.0 mΩ max



RoHS

B048F240T30

B048F240M30

Product Description

The V•I Chip Bus Converter Module is a high efficiency (>96%), narrow input range Sine Amplitude Converter (SAC) operating from a 38 to 53 Vdc primary bus to deliver an isolated 19.0 V to 26.5 V secondary. The BCM may be used to power non-isolated POL converters or as an independent 19.0 - 26.5 V source. Due to the fast response time and low noise of the BCM, the need for limited life aluminum electrolytic or tantalum capacitors at the load is reduced—or eliminated—resulting in savings of board area, materials and total system cost.

The BCM achieves a power density of 1017 W/in³ in a V•I Chip package compatible with standard pick-andplace and surface mount assembly process. The V•I Chip package provides flexible thermal management through its low junction-to-board and junction-to-case thermal resistance. Owing to its high conversion efficiency and safe operating temperature range, the BCM does not require a discrete heat sink in typical applications. Low junction-to-case and junction-to-lead thermal impedances assure low junction temperatures and long life in the harshest environments.

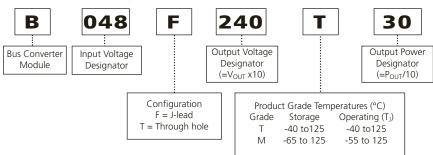
Absolute Maximum Ratings

| Parameter | Values | Unit | Notes |
|--------------------------------|-------------|------|---------------------|
| +In to -In | -1.0 to 60 | Vdc | |
| +In to -In | 100 | Vdc | For 100 ms |
| PC to -In | -0.3 to 7.0 | Vdc | |
| +Out to -Out | -0.5 to 50 | Vdc | |
| Isolation voltage | 2,250 | Vdc | Input to output |
| Output current | 14.8 | А | Continuous |
| Peak output current | 18.8 | А | For 1 ms |
| Output power | 300 | W | Continuous |
| Peak output power | 450 | W | For 1 ms |
| Case temperature | 225 | °C | During reflow MSL 5 |
| | -40 to 125 | °C | T-Grade |
| Operating junction temperature | -55 to 125 | °C | M-Grade |
| Storage temperature | -40 to 125 | °C | T-Grade |
| Storage temperature | -65 to 125 | °C | M-Grade |

Note:

 The referenced junction is defined as the semiconductor having the highest temperature. This temperature is monitored by a shutdown comparator.

Part Numbering



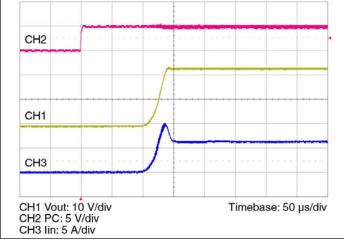


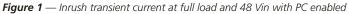
Specifications

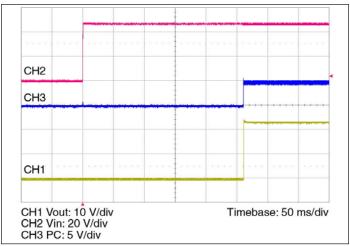
| Input (Conditions are at 48 | Vin, full load, and 25°C ambient | unless otherwise specified) |
|------------------------------------|----------------------------------|-----------------------------|
|------------------------------------|----------------------------------|-----------------------------|

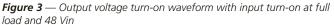
| Parameter | Min | Тур | Max | Unit | Note |
|--|------|-----|------|--------|---|
| Input voltage range | 38 | 48 | 53 | Vdc | |
| Input dV/dt | | | 1 | V/µs | |
| Input undervoltage turn-on | | | 37.4 | Vdc | |
| Input undervoltage turn-off | 32.8 | | | Vdc | |
| Input overvoltage turn-on | 53.0 | | | Vdc | |
| Input overvoltage turn-off | | | 57.5 | Vdc | |
| Input quiescent current | | 2.6 | | mA | PC low |
| Inrush current overshoot | | 3.5 | | А | Using test circuit in Figure 20; See Figure 1 |
| Input current | | | 6.8 | Adc | |
| Input reflected ripple current | | 120 | | mA p-p | Using test circuit in Figure 20; See Figure 4 |
| No load power dissipation | | 4.0 | 5.9 | W | |
| Internal input capacitance | | 1.9 | | μF | |
| Internal input inductance | | 5 | | nH | |
| Recommended external input capacitance | | 47 | | μF | 200 nH maximum source inductance; See Figure 20 |

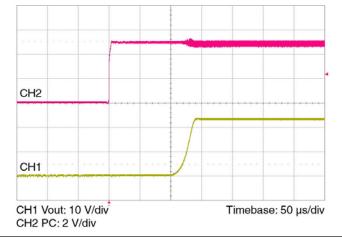
Input Waveforms

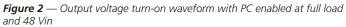


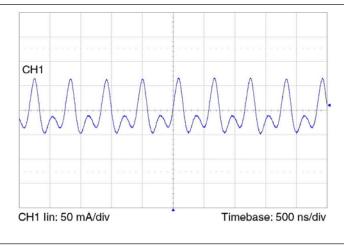














V•I Chip Bus Converter Module

Specifications (continued)

Output (Conditions are at 48 Vin, full load, and 25°C ambient unless otherwise specified)

| Parameter | Min | Тур | Max | Unit | Note |
|------------------------------------|--------|------|--------|-------|--|
| Output voltage | 19.0 | | 26.5 | Vdc | No load |
| Output voltage | 18.3 | | 25.8 | Vdc | Full load |
| | 0 | | 300 | W | 42 - 53 V _{IN} |
| Output power | 0 | | 269 | W | 38 - 53 V _{IN} |
| Rated DC current | 0 | | 14.8 | Adc | $P_{OUT} \le 300 \text{ W}$ |
| Peak repetitive power | | | 450 | W | Max pulse width 1ms, max duty cycle 10%, |
| | | | 450 | vv | baseline power 50% |
| Current share accuracy | | 5 | 10 | % | See Parallel Operation on Page 10 |
| Efficiency | | | | | |
| Half load | 95.0 | 95.7 | | % | See Figure 5 |
| Full load | 94.7 | 95.8 | | % | See Figure 5 |
| Internal output inductance | | 1.1 | | nH | |
| Internal output capacitance | | 7.7 | | μF | Effective value |
| Load capacitance | | | 300 | μF | |
| Output overvoltage setpoint | 26.5 | | | Vdc | Module will shut down |
| Output ripple voltage | | | | | |
| No external bypass | | 150 | 348 | mVp-p | See Figures 7 and 9 |
| 3.3 µF bypass capacitor | | 13 | | mVp-p | See Figure 8 |
| Short circuit protection set point | 15.1 | | | Adc | Module will shut down |
| Average short circuit current | | 0.62 | | А | |
| Effective switching frequency | 3.0 | 3.4 | 3.8 | MHz | Fixed, 1.7 MHz per phase |
| Line regulation | | | | | |
| K | 0.4950 | 1/2 | 0.5050 | | $V_{OUT} = K \bullet V_{IN}$ at no load |
| Load regulation | | | | | |
| R _{OUT} | | 43.6 | 60.0 | mΩ | |
| Transient response | | | | | |
| Voltage overshoot | | 600 | | mV | 100% load step; See Figures 10 and 11 |
| Response time | | 200 | | ns | See Figures 10 and 11 |
| Recovery time | | 1 | | μs | See Figures 10 and 11 |
| Output overshoot | | | | | |
| Input turn-on | | 0 | | mV | No output filter; See Figure 3 |
| PC enable | | 0 | | mV | No output filter; See Figure 2 |
| Output turn-on delay | | | | | |
| From application of power | | 260 | | ms | No output filter; See Figure 3 |
| From release of PC pin | | 65 | | ms | No output filter |

Output Waveforms

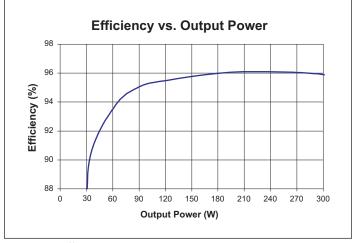


Figure 5 — Efficiency vs. output power



800-735-6200 V•I Chip Bus Converter Module

B048F240T30

Power Dissipation 14 Power Dissipation (W) 12 10 8 6 4 2 30 60 90 120 150 180 210 240 270 300 0 Output Power (W) Figure 6 — Power dissipation as a function of output power

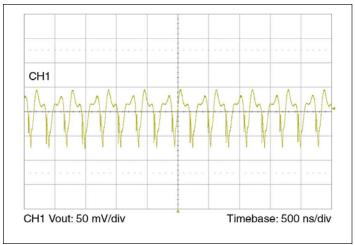


Figure 7 — Output voltage ripple at full load and 48 Vin without any external bypass capacitor.

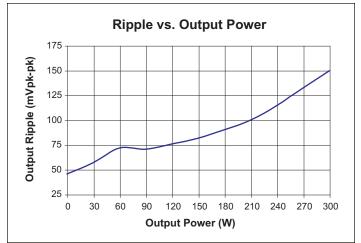


Figure 9 — Output voltage ripple vs. output power at 48 Vin without any external bypass capacitor.

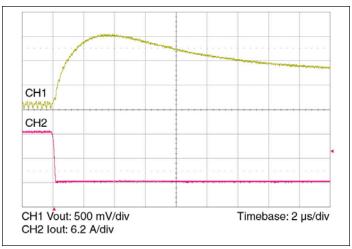


Figure 11 — 12.5-0 A load step with 100 μ F input capacitor and no output capacitor.

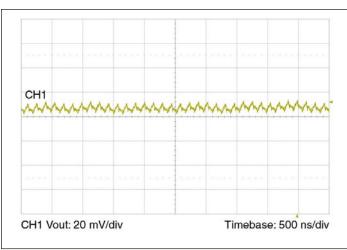


Figure 8 — Output voltage ripple at full load and 48 Vin with 3.3 μ F ceramic external bypass capacitor and 20 nH of distribution inductance.

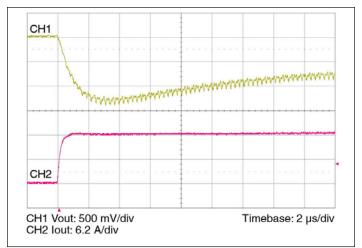


Figure 10 — 0 -12.5 A load step with 100 μ F input capacitor and no output capacitor.

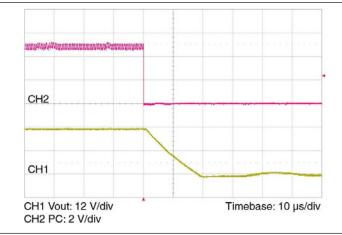
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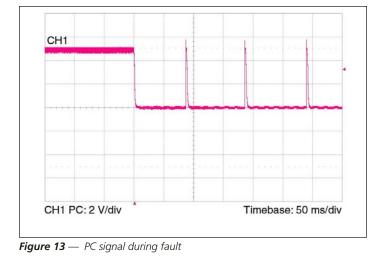
General

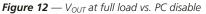
| Parameter | Min | Тур | Max | Unit | Note |
|---|-------|------------|-----|-------|--|
| MTBF | | | | | |
| MIL-HDBK-217F | | 3.5 | | Mhrs | 25°C, GB |
| Isolation specifications | | | | | |
| Voltage | 2,250 | | | Vdc | Input to output |
| Capacitance | | 3,000 | | pF | Input to output |
| Resistance | 10 | | | MΩ | Input to output |
| | | cTÜVus | | | UL/CSA 60950-1, EN 60950-1 |
| Agency approvals | | CE Mark | | | Low voltage directive |
| | | RoHS | | | |
| Mechanical | | | | | See Mechanical Drawings, Figures 15 – 18 |
| Weight | | 0.53/15 | | oz/g | |
| Dimensions | | | | | |
| Length | | 1.28/32,5 | | in/mm | |
| Width | | 0.87/22 | | in/mm | |
| Height | | 0.265/6,73 | | in/mm | |
| Thermal | | | | | |
| Over temperature shutdown | 125 | 130 | 135 | °C | Junction temperature |
| Thermal capacity | | 9.3 | | Ws/°C | |
| Junction-to-case thermal impedance $(R_{\theta JC})$ | | 1.1 | | °C/W | |
| Junction-to-board thermal impedance $(R_{\theta JB})$ | | 2.1 | | °C/W | |

Auxiliary Pins (Conditions are at 48 Vin, full load, and 25°C ambient unless otherwise specified)

| Parameter | Min | Тур | Max | Unit | Note |
|------------------------|-----|-----|-----|------|---|
| Primary control (PC) | | | | | |
| DC voltage | 4.8 | 5.0 | 5.2 | Vdc | |
| Module disable voltage | 2.4 | 2.5 | | Vdc | |
| Module enable voltage | | 2.5 | 2.6 | Vdc | |
| Current limit | 2.4 | 2.5 | 2.9 | mA | Source only |
| Enable delay time | | 65 | | ms | |
| Disable delay time | | 20 | | μs | See Figure 12, time from PC low to output low |









Rev. 2.3

+In / -In – DC Voltage Input Ports

The V•I Chip input voltage range should not be exceeded. An internal under / over voltage lockout function prevents operation outside of the normal operating input range. The BCM turns on within an input voltage window bounded by the "Input under-voltage turn-on" and "Input over-voltage turn-off" levels, as specified. The V•I Chip may be protected against accidental application of a reverse input voltage by the addition of a rectifier in series with the positive input, or a reverse rectifier in shunt with the positive input located on the load side of the input fuse.

The connection of the V•I Chip to its power source should be implemented with minimal distribution inductance. If the interconnect inductance exceeds 100 nH, the input should be bypassed with a RC damper to retain low source impedance and stable operation. With an interconnect inductance of 200 nH, the RC damper may be 47 μ F in series with 0.3 Ω . A single electrolytic or equivalent low-Q capacitor may be used in place of the series RC bypass.

PC – Primary Control

The Primary Control port is a multifunction node that provides the following functions:

Enable / Disable – If the PC port is left floating, the BCM output is enabled. Once this port is pulled lower than 2.4 Vdc with respect to –In, the output is disabled. This action can be realized by employing a relay, opto-coupler, or open collector transistor. Refer to Figures 1-3, 12 and 13 for the typical enable / disable characteristics. This port should not be toggled at a rate higher than 1 Hz. The PC port should also not be driven by or pulled up to an external voltage source.

<u>Primary Auxiliary Supply</u> – The PC port can source up to 2.4 mA at 5.0 Vdc. The PC port should never be used to sink current.

<u>Alarm</u> – The BCM contains circuitry that monitors output overload, input over voltage or under voltage, and internal junction temperatures. In response to an abnormal condition in any of the monitored parameters, the PC port will toggle. Refer to Figure 13 for PC alarm characteristics.

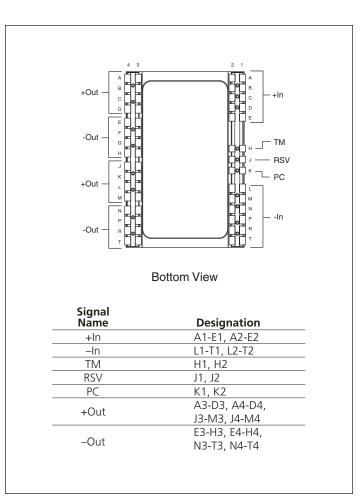
TM and RSV – Reserved for factory use.

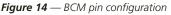
+Out / -Out – DC Voltage Output Ports

Two sets of contacts are provided for the +Out port. They must be connected in parallel with low interconnect resistance. Similarly, two sets of contacts are provided for the –Out port. They must be connected in parallel with low interconnect resistance. Within the specified operating range, the average output voltage is defined by the Level 1 DC behavioral model of Figure 21. The current source capability of the BCM is rated in the specifications section of this document.

The low output impedance of the BCM reduces or eliminates the need for limited life aluminum electrolytic or tantalum capacitors at the input of POL converters.

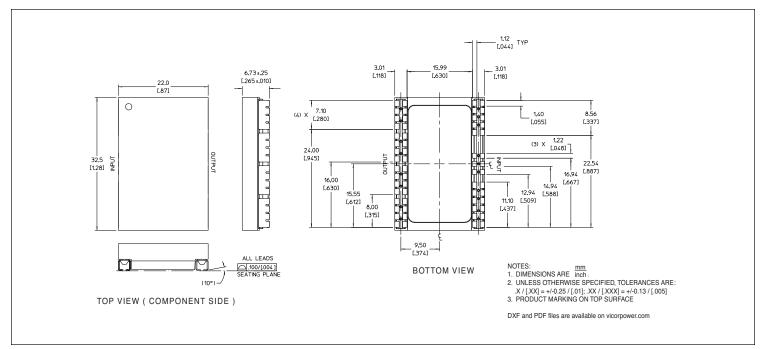
Total load capacitance at the output of the BCM should not exceed the specified maximum. Owing to the wide bandwidth and low output impedance of the BCM, low frequency bypass capacitance and significant energy storage may be more densely and efficiently provided by adding capacitance at the input of the BCM.





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Mechanical Drawings





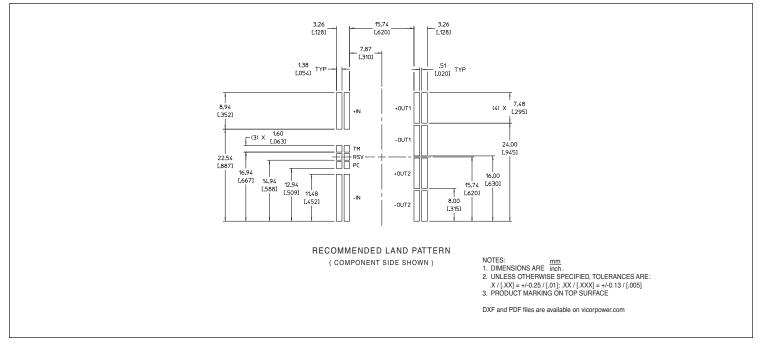


Figure 16 — BCM PCB land layout information



Mechanical Drawings (continued)

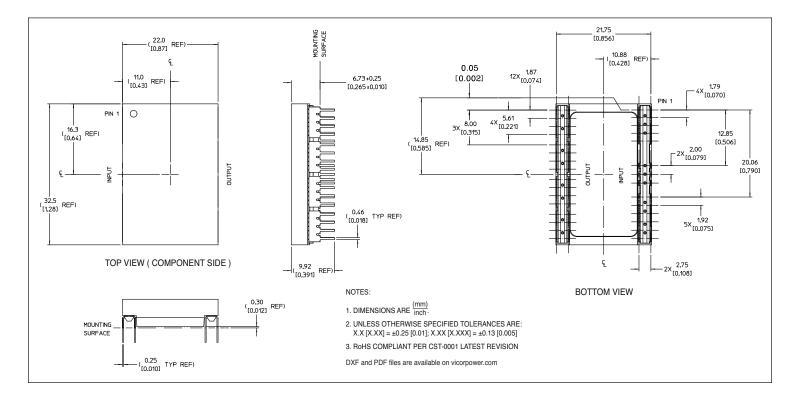
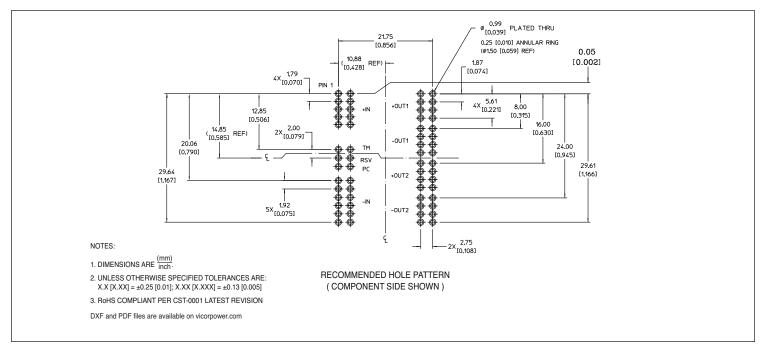
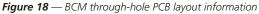


Figure 17 — BCM through-hole mechanical outline







Configuration Options

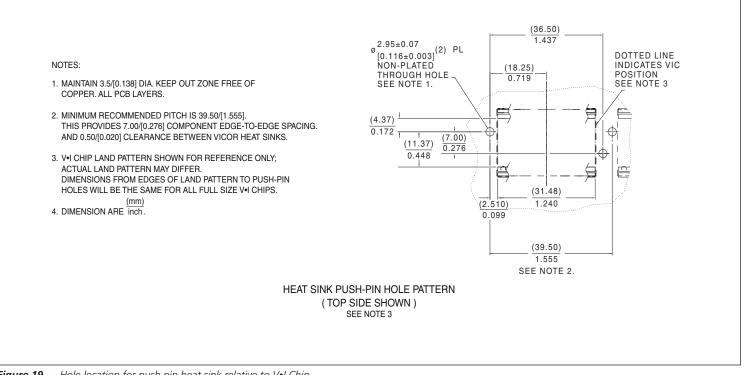


Figure 19 — Hole location for push pin heat sink relative to V•I Chip



Parallel Operation

The BCM will inherently current share when operated in an array. Arrays may be used for higher power or redundancy in an application.

Current sharing accuracy is maximized when the source and load impedance presented to each BCM within an array are equal. The recommended method to achieve matched impedances is to dedicate common copper planes within the PCB to deliver and return the current to the array, rather than rely upon traces of varying lengths. In typical applications the current being delivered to the load is larger than that sourced from the input, allowing traces to be utilized on the input side if necessary. The use of dedicated power planes is, however, preferable.

The BCM power train and control architecture allow bi-directional power transfer, including reverse power processing from the BCM output to its input. Reverse power transfer is enabled if the BCM input is within its operating range and the BCM is otherwise enabled. The BCM's ability to process power in reverse improves the BCM transient response to an output load dump.

Input Impedance Recommendations

To take full advantage of the BCM capabilities, the impedance presented to its input terminals must be low from DC to approximately 5 MHz. The source should exhibit low inductance (less than 100 nH) and should have a critically damped response. If the interconnect inductance exceeds 100 nH, the BCM input pins should be bypassed with an RC damper (e.g., 47 μF in series with 0.3 ohm) to retain low source impedance and stable operations. Given the wide bandwidth of the BCM, the source response is generally the limiting factor in the overall system response.

Anomalies in the response of the source will appear at the output of the BCM multiplied by its K factor. The DC resistance of the source should be kept as low as possible to minimize voltage deviations. This is especially important if the BCM is operated near low or high line as the over/under voltage detection circuitry could be activated.

Input Fuse Recommendations

V•I Chips are not internally fused in order to provide flexibility in configuring power systems. However, input line fusing of V•I Chips must always be incorporated within the power system. A fast acting fuse should be placed in series with the +In port.

Application Notes

For BCM and V•I Chip application notes on soldering, thermal management, board layout, and system design click on the link below:

http://www.vicorpower.com/technical_library/application_information/chips/



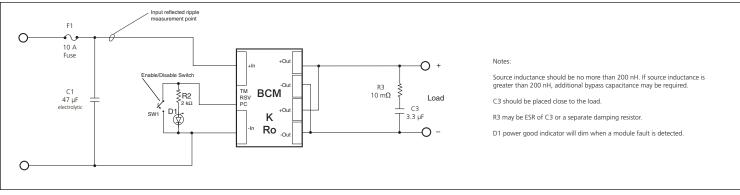


Figure 20 — BCM test circuit

V•I Chip Bus Converter Level 1 DC Behavioral Model for 48 V to 24 V, 300 W

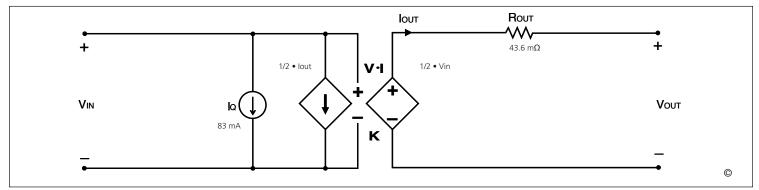


Figure 21 — This model characterizes the DC operation of the V•I Chip bus converter, including the converter transfer function and its losses. The model enables estimates or simulations of output voltage as a function of input voltage and output load, as well as total converter power dissipation or heat generation.

V•I Chip Bus Converter Level 2 Transient Behavioral Model for 48 V to 24 V, 300 W

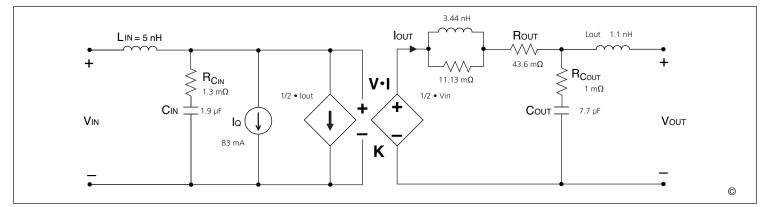


Figure 22 — This model characterizes the AC operation of the V•I Chip bus converter including response to output load or input voltage transients or steady state modulations. The model enables estimates or simulations of input and output voltages under transient conditions, including response to a stepped load with or without external filtering elements.



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V•I Chip Bus Converter Module

B048F240T30

Rev. 2.3