

BIPOLAR ANALOG INTEGRATED CIRCUIT

μ PC8182TB

3 V, 2.9 GHz SILICON MMIC MEDIUM OUTPUT POWER AMPLIFIER FOR MOBILE COMMUNICATIONS

DESCRIPTION

The μ PC8182TB is a silicon monolithic integrated circuit designed as amplifier for mobile communications. This IC operates at 3 V. The medium output power is suitable for RF-TX of mobile communications system.

This IC is manufactured using our 30 GHz f_{max} UHS0 (Ultra High Speed Process) silicon bipolar process. This process uses direct silicon nitride passivation film and gold electrodes. These materials can protect the chip surface from pollution and prevent corrosion/migration. Thus, this IC has excellent performance, uniformity and reliability.

FEATURES

- Supply voltage : $V_{CC} = 2.7$ to 3.3 V
- Circuit current : $I_{CC} = 30$ mA TYP. @ $V_{CC} = 3.0$ V
- Medium output power : $P_{O(1dB)} = +9.5$ dBm TYP. @ $f = 0.9$ GHz
 $P_{O(1dB)} = +9.0$ dBm TYP. @ $f = 1.9$ GHz
 $P_{O(1dB)} = +8.0$ dBm TYP. @ $f = 2.4$ GHz
- Power gain : $G_P = 21.5$ dB TYP. @ $f = 0.9$ GHz
 $G_P = 20.5$ dB TYP. @ $f = 1.9$ GHz
 $G_P = 20.5$ dB TYP. @ $f = 2.4$ GHz
- Upper limit operating frequency : $f_u = 2.9$ GHz TYP. @ 3 dB bandwidth
- High-density surface mounting : 6-pin super minimold package ($2.0 \times 1.25 \times 0.9$ mm)

APPLICATION

- Buffer amplifiers on 1.9 to 2.4 GHz mobile communications system

ORDERING INFORMATION

Part Number	Package	Marking	Supplying Form
μ PC8182TB-E3	6-pin super minimold	C3F	<ul style="list-style-type: none"> • Embossed tape 8 mm wide • Pin 1, 2, 3 face the perforation side of the tape • Qty 3 kpcs/reel

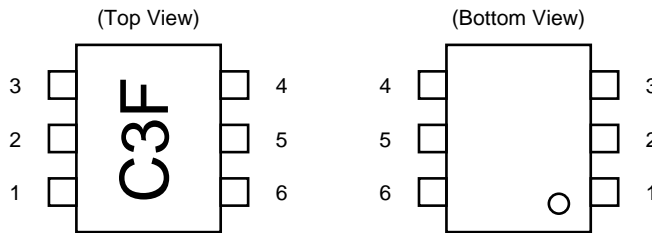
Remark To order evaluation samples, contact your nearby sales office.

Part number for sample order: μ PC8182TB

Caution Observe precautions when handling because these devices are sensitive to electrostatic discharge.

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 Not all devices/types available in every country. Please check with local NEC Compound Semiconductor Devices representative for availability and additional information.

PIN CONNECTIONS



Pin No.	Pin Name
1	INPUT
2	GND
3	GND
4	OUTPUT
5	GND
6	V _{cc}

PRODUCT LINE-UP (T_A = +25°C, V_{cc} = V_{out} = 3.0 V, Z_s = Z_L = 50 Ω)

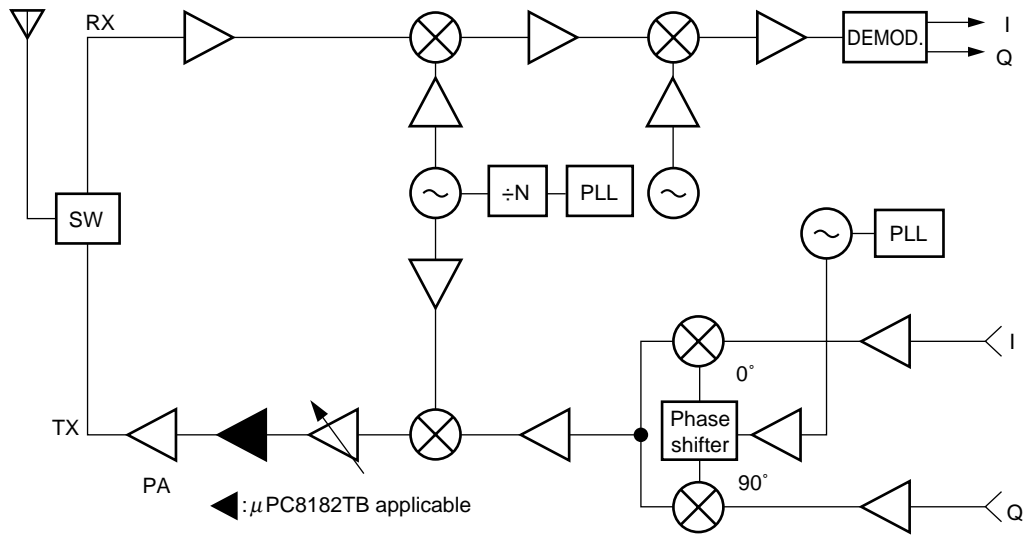
Part No.	f _u (GHz)	P _o (1 dB) (dBm)	GP (dB)	I _{cc} (mA)	Package	Marking
μPC8182TB	2.9	+9.5 @ f = 0.9 GHz +9.0 @ f = 1.9 GHz +8.0 @ f = 2.4 GHz	21.5 @ f = 0.9 GHz 20.5 @ f = 1.9 GHz 20.5 @ f = 2.4 GHz	30.0	6-pin super minimold	C3F
μPC2762T	2.9	+8.0 @ f = 0.9 GHz	13.0 @ f = 0.9 GHz	26.5	6-pin minimold	C1Z
μPC2762TB		+7.0 @ f = 1.9 GHz	15.5 @ f = 1.9 GHz		6-pin super minimold	
μPC2763T	2.7	+9.5 @ f = 0.9 GHz	20.0 @ f = 0.9 GHz	27.0	6-pin minimold	C2A
μPC2763TB		+6.5 @ f = 1.9 GHz	21.0 @ f = 1.9 GHz		6-pin super minimold	
μPC2771T	2.2	+11.5 @ f = 0.9 GHz	21.0 @ f = 0.9 GHz	36.0	6-pin minimold	C2H
μPC2771TB		+9.5 @ f = 1.5 GHz	21.0 @ f = 1.5 GHz		6-pin super minimold	
★ μPC8181TB	4.0	+8.0 @ f = 0.9 GHz +7.0 @ f = 1.9 GHz +7.0 @ f = 2.4 GHz	19.0 @ f = 0.9 GHz 21.0 @ f = 1.9 GHz 22.0 @ f = 2.4 GHz	23.0	6-pin super minimold	C3E

Remark Typical performance. Please refer to ELECTRICAL CHARACTERISTICS in detail.

Caution The package size distinguishes between minimold and super minimold.

SYSTEM APPLICATION EXAMPLE

Digital cellular telephone



Caution The insertion point is different due to the specifications of conjunct devices.

PIN EXPLANATION

Pin No.	Pin Name	Applied Voltage (V)	Pin Voltage (V) <small>Note</small>	Function and Applications	Internal Equivalent Circuit
1	INPUT	–	0.99	Signal input pin. A internal matching circuit, configured with resistors, enables 50 Ω connection over a wide band. A multi-feedback circuit is designed to cancel the deviations of h_{FE} and resistance. This pin must be coupled to signal source with capacitor for DC cut.	
4	OUTPUT	Voltage as same as V_{CC} through external inductor	–	Signal output pin. The inductor must be attached between V_{CC} and output pins to supply current to the internal output transistors.	
6	V_{CC}	2.7 to 3.3	–	Power supply pin, which biases the internal input transistor. This pin should be externally equipped with bypass capacitor to minimize its impedance.	
2 3 5	GND	0	–	Ground pin. This pin should be connected to system ground with minimum inductance. Ground pattern on the board should be formed as wide as possible. All the ground pins must be connected together with wide ground pattern to decrease impedance difference.	

Note Pin voltage is measured at $V_{CC} = 3.0$ V.

ABSOLUTE MAXIMUM RATINGS

Parameter	Symbol	Test Conditions	Ratings	Unit
Supply Voltage	V _{CC}	T _A = +25°C, pin 4 and pin 6	3.6	V
Total Circuit Current	I _{CC}	T _A = +25°C	60	mA
Power Dissipation	P _D	T _A = +85°C Note	270	mW
Operating Ambient Temperature	T _A		-40 to +85	°C
Storage Temperature	T _{stg}		-55 to +150	°C
Input Power	P _{in}	T _A = +25°C	+10	dBm

Note Mounted on double-sided copper-clad 50 × 50 × 1.6 mm epoxy glass PWB

RECOMMENDED OPERATING RANGE

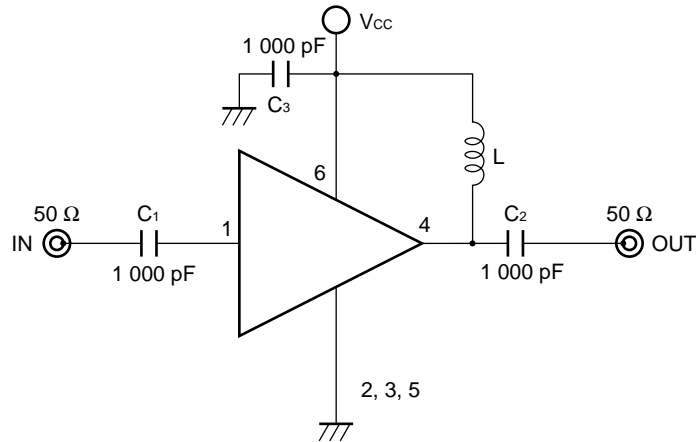
Parameter	Symbol	MIN.	TYP.	MAX.	Unit	Remarks
Supply Voltage	V _{CC}	2.7	3.0	3.3	V	Same voltage should be applied to pin 4 and pin 6.
Operating Ambient Temperature	T _A	-40	+25	+85	°C	-

ELECTRICAL CHARACTERISTICS

($T_A = +25^{\circ}\text{C}$, $V_{CC} = V_{out} = 3.0\text{ V}$, $Z_s = Z_L = 50\ \Omega$, unless otherwise specified)

Parameter	Symbol	Test Conditions	MIN.	TYP.	MAX.	Unit
Circuit Current	I_{CC}	No signal	–	30.0	38.0	mA
Power Gain	G_P	f = 0.9 GHz f = 1.9 GHz f = 2.4 GHz	19.0 18.0 18.0	21.5 20.5 20.5	25.0 24.0 24.0	dB
Noise Figure	NF	f = 0.9 GHz f = 1.9 GHz f = 2.4 GHz	– – –	4.5 4.5 5.0	6.0 6.0 6.5	dB
Upper Limit Operating Frequency	f_u	3 dB down below from gain at f = 0.1 GHz	2.6	2.9	–	GHz
Isolation	ISL	f = 0.9 GHz f = 1.9 GHz f = 2.4 GHz	28 27 26	33 32 31	– – –	dB
Input Return Loss	RL_{in}	f = 0.9 GHz f = 1.9 GHz f = 2.4 GHz	5 7 9	8 10 12	– – –	dB
Output Return Loss	RL_{out}	f = 0.9 GHz f = 1.9 GHz f = 2.4 GHz	7 8 11	10 11 14	– – –	dB
Gain 1 dB Compression Output Power	$P_{O(1dB)}$	f = 0.9 GHz f = 1.9 GHz f = 2.4 GHz	+7.0 +6.5 +5.5	+9.5 +9.0 +8.0	– – –	dBm
Saturated Output Power	$P_{O(sat)}$	f = 0.9 GHz, $P_{in} = -5\text{ dBm}$ f = 1.9 GHz, $P_{in} = -5\text{ dBm}$ f = 2.4 GHz, $P_{in} = -5\text{ dBm}$	– – –	+11.0 +10.5 +10.0	– – –	dBm

TEST CIRCUITS



COMPONENTS OF TEST CIRCUIT FOR MEASURING ELECTRICAL CHARACTERISTICS

	Type	Value
C ₁ , C ₂	Bias Tee	1 000 pF
C ₃	Capacitor	1 000 pF
L	Bias Tee	1 000 nH

EXAMPLE OF ACTUAL APPLICATION COMPONENTS

	Type	Value	Operating Frequency
C ₁ to C ₃	Chip capacitor	1 000 pF	100 MHz or higher
L	Chip inductor	100 nH	100 MHz or higher
		10 nH	2.0 GHz or higher

INDUCTOR FOR THE OUTPUT PIN

The internal output transistor of this IC consumes 20 mA, to output medium power. To supply current for output transistor, connect an inductor between the Vcc pin (pin 6) and output pin (pin 4). Select large value inductance, as listed above.

The inductor has both DC and AC effects. In terms of DC, the inductor biases the output transistor with minimum voltage drop to output enable high level. In terms of AC, the inductor make output-port-impedance higher to get enough gain. In this case, large inductance and Q is suitable.

For above reason, select an inductance of 100 Ω or over impedance in the operating frequency. The gain is a peak in the operating frequency band, and suppressed at lower frequencies.

The recommendable inductance can be chosen from example of actual application components list as shown above.

CAPACITORS FOR THE Vcc, INPUT, AND OUTPUT PINS

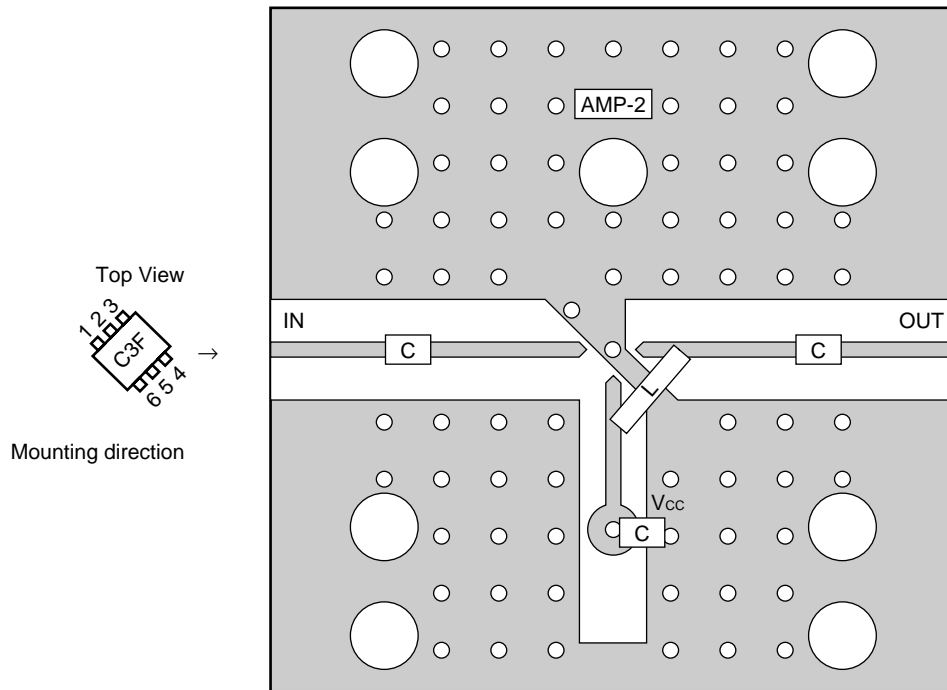
Capacitors of 1 000 pF are recommendable as the bypass capacitor for the Vcc pin and the coupling capacitors for the input and output pins.

The bypass capacitor connected to the Vcc pin is used to minimize ground impedance of Vcc pin. So, stable bias can be supplied against Vcc fluctuation.

The coupling capacitors, connected to the input and output pins, are used to cut the DC and minimize RF serial impedance. Their capacitance are therefore selected as lower impedance against a 50 Ω load. The capacitors thus perform as high pass filters, suppressing low frequencies to DC.

To obtain a flat gain from 100 MHz upwards, 1 000 pF capacitors are used in the test circuit. In the case of under 10 MHz operation, increase the value of coupling capacitor such as 10 000 pF. Because the coupling capacitors are determined by equation, $C = 1/(2\pi Rfc)$.

ILLUSTRATION OF THE TEST CIRCUIT ASSEMBLED ON EVALUATION BOARD



COMPONENT LIST

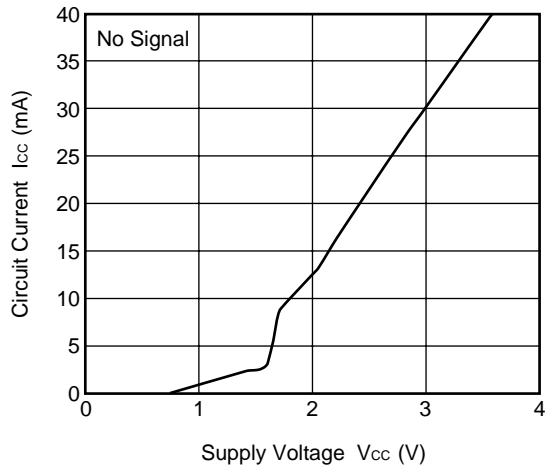
	Value
C	1 000 pF
L	Example: 10 nH

Notes

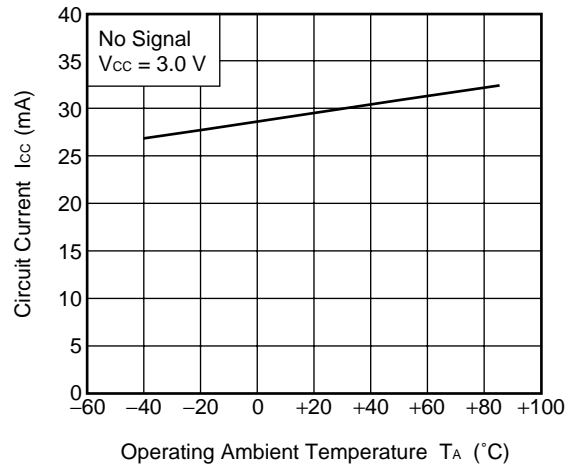
1. 30 × 30 × 0.4 mm double-sided copper-clad polyimide board.
2. Back side: GND pattern
3. Solder plated on pattern
4. ○ ○ : Through holes

TYPICAL CHARACTERISTICS ($T_A = +25^\circ\text{C}$, unless otherwise specified)

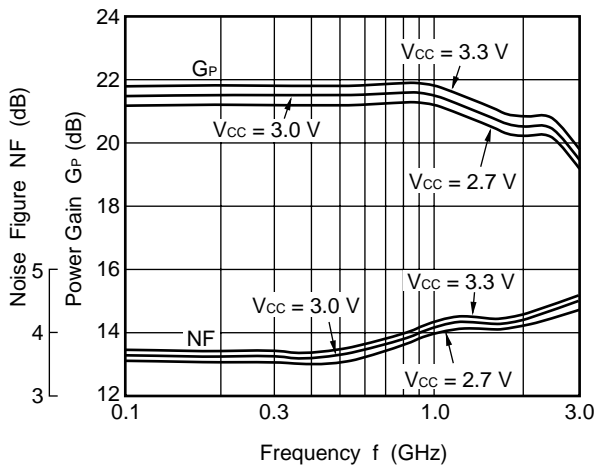
CIRCUIT CURRENT vs. SUPPLY VOLTAGE



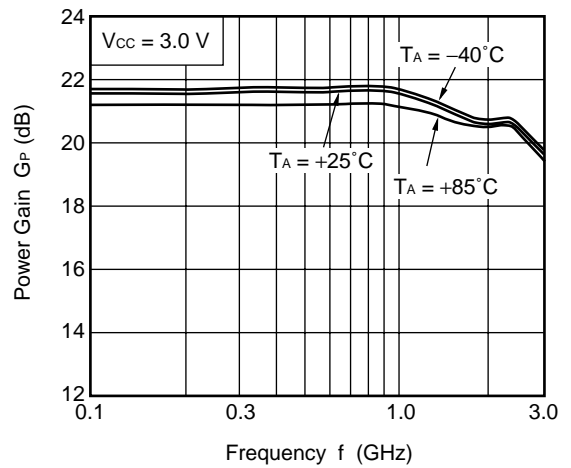
CIRCUIT CURRENT vs. OPERATING AMBIENT TEMPERATURE



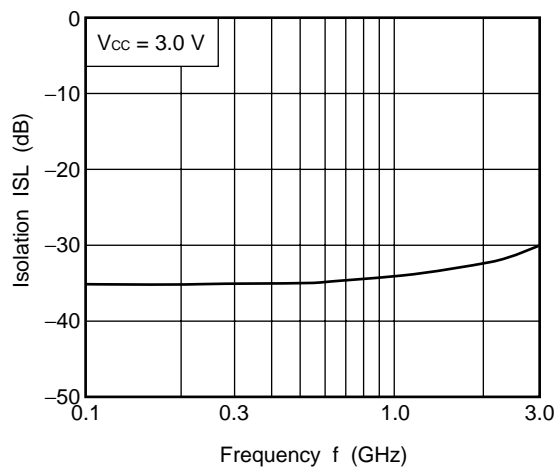
NOISE FIGURE, POWER GAIN vs. FREQUENCY



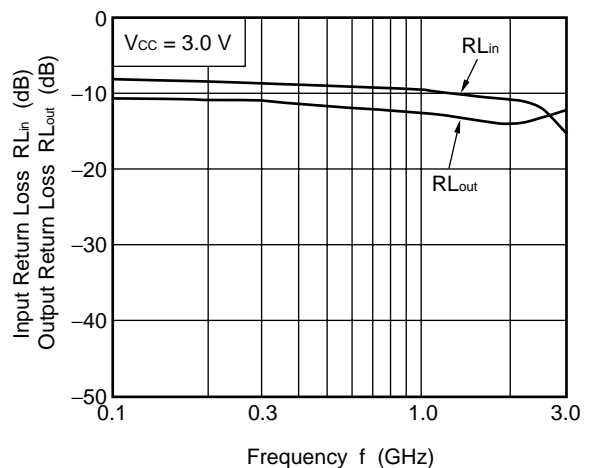
POWER GAIN vs. FREQUENCY



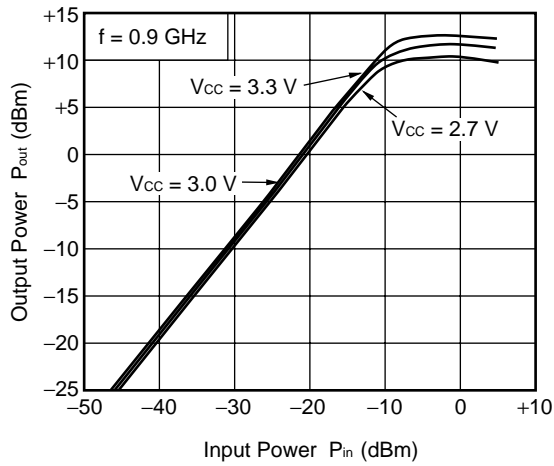
ISOLATION vs. FREQUENCY



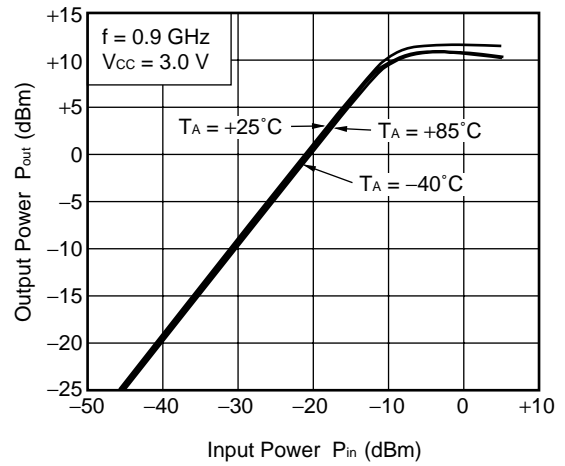
INPUT RETURN LOSS, OUTPUT RETURN LOSS vs. FREQUENCY



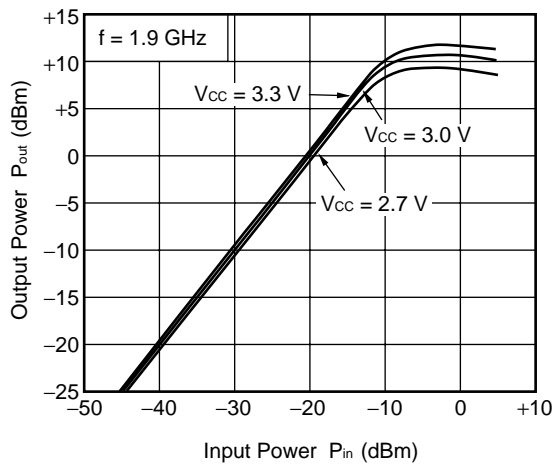
OUTPUT POWER vs. INPUT POWER



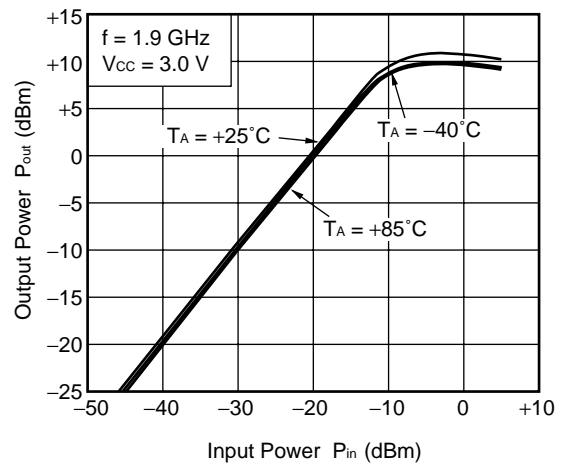
OUTPUT POWER vs. INPUT POWER



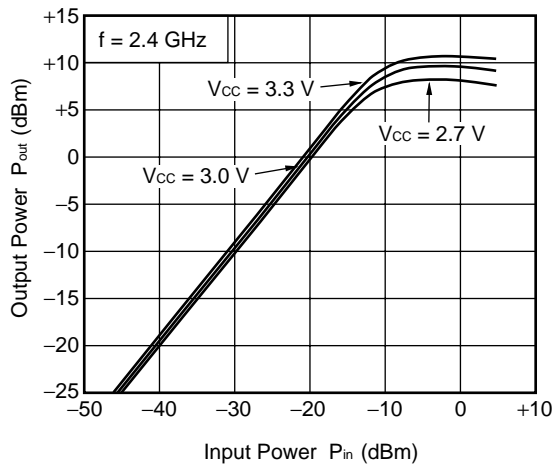
OUTPUT POWER vs. INPUT POWER



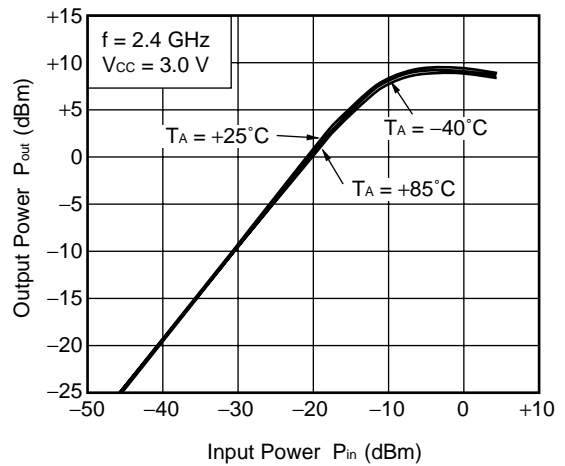
OUTPUT POWER vs. INPUT POWER



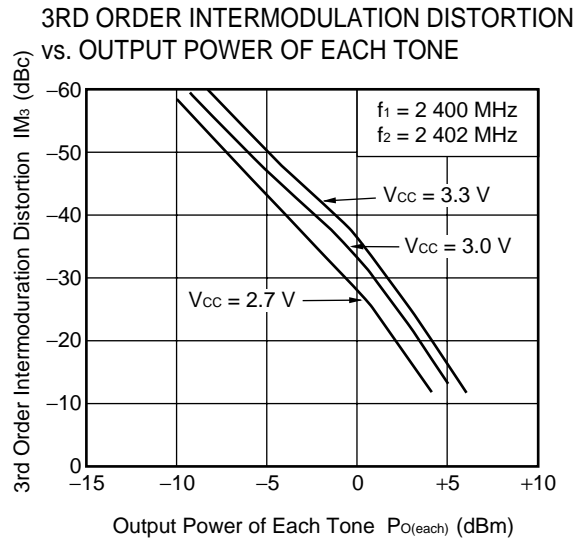
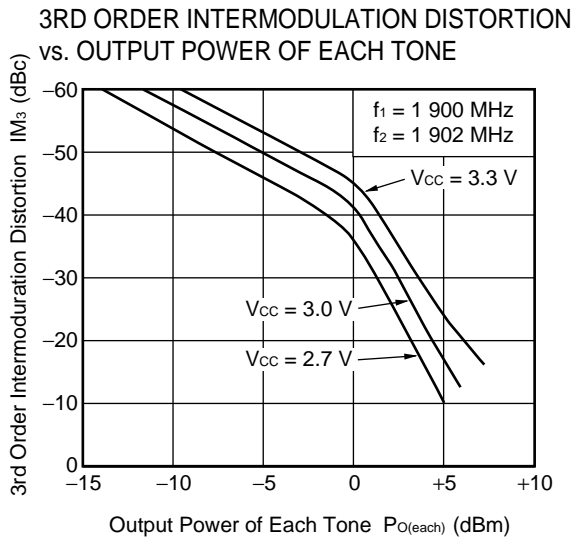
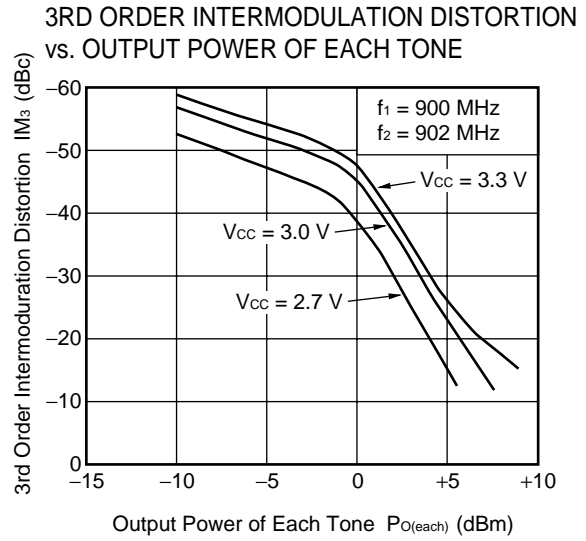
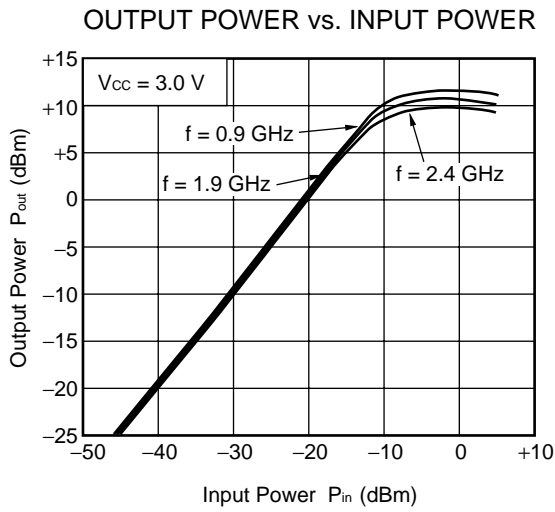
OUTPUT POWER vs. INPUT POWER



OUTPUT POWER vs. INPUT POWER



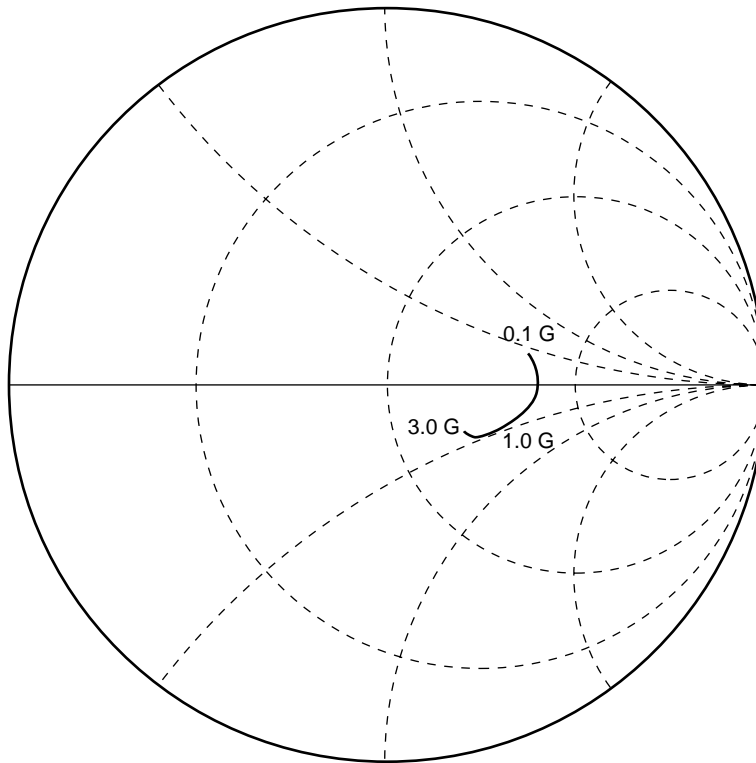
★



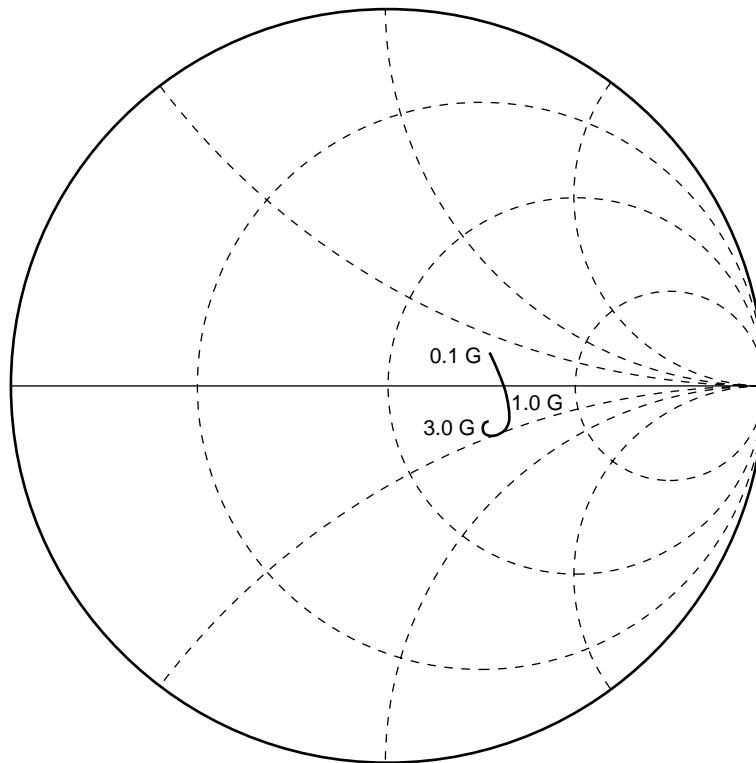
Remark The graphs indicate nominal characteristics.

SMITH CHART ($V_{CC} = V_{out} = 3.0\text{ V}$)

S₁₁-FREQUENCY



S₂₂-FREQUENCY



★ S-PARAMETERS

S-parameters/Noise parameters are provided on the NEC Compound Semiconductor Devices Web site in a form (S2P) that enables direct import to a microwave circuit simulator without keyboard input.

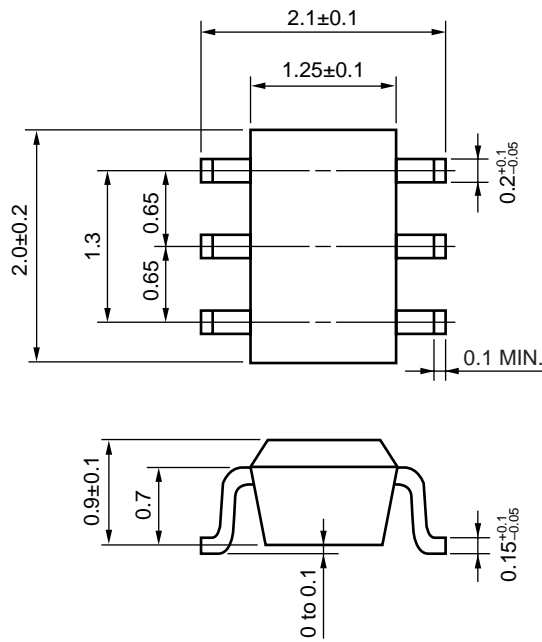
Click here to download S-parameters.

[RF and Microwave] → [Device Parameters]

URL <http://www.csd-nec.com/>

PACKAGE DIMENSIONS

6-PIN SUPER MINIMOLD (UNIT: mm)



NOTES ON CORRECT USE

- (1) Observe precautions for handling because of electro-static sensitive devices.
- (2) Form a ground pattern as widely as possible to minimize ground impedance (to prevent undesired oscillation).
All the ground pins must be connected together with wide ground pattern to decrease impedance difference.
- (3) The bypass capacitor should be attached to the V_{CC} pin.
- (4) The inductor must be attached between V_{CC} and output pins. The inductance value should be determined in accordance with desired frequency.
- (5) The DC cut capacitor must be attached to input and output pin.

RECOMMENDED SOLDERING CONDITIONS

This product should be soldered and mounted under the following recommended conditions. For soldering methods and conditions other than those recommended below, contact your nearby sales office.

Soldering Method	Soldering Conditions	Condition Symbol
Infrared Reflow	Peak temperature (package surface temperature) : 260°C or below Time at peak temperature : 10 seconds or less Time at temperature of 220°C or higher : 60 seconds or less Preheating time at 120 to 180°C : 120±30 seconds Maximum number of reflow processes : 3 times Maximum chlorine content of rosin flux (% mass) : 0.2%(Wt.) or below	IR260
VPS	Peak temperature (package surface temperature) : 215°C or below Time at temperature of 200°C or higher : 25 to 40 seconds Preheating time at 120 to 150°C : 30 to 60 seconds Maximum number of reflow processes : 3 times Maximum chlorine content of rosin flux (% mass) : 0.2%(Wt.) or below	VP215
Wave Soldering	Peak temperature (molten solder temperature) : 260°C or below Time at peak temperature : 10 seconds or less Preheating temperature (package surface temperature) : 120°C or below Maximum number of flow processes : 1 time Maximum chlorine content of rosin flux (% mass) : 0.2%(Wt.) or below	WS260
Partial Heating	Peak temperature (pin temperature) : 350°C or below Soldering time (per side of device) : 3 seconds or less Maximum chlorine content of rosin flux (% mass) : 0.2%(Wt.) or below	HS350

Caution Do not use different soldering methods together (except for partial heating).

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M8E 00.4-0110

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► **Technical issue**

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