

# 3MHz Dual Step-Down Converters and Dual Low-Input LDOs with I<sup>2</sup>C Compatible Interface

### **ISL9305**

The ISL9305 is an integrated mini Power Management IC (mini-PMIC) ideal for applications of powering low-voltage microprocessor or multiple voltage rails with battery as input sources, such as a single Li-ion or Li-Polymer. ISL9305 integrates two high-efficiency 3MHz synchronous step-down converters (DCD1 and DCD2) and two low-input, low-dropout linear regulators (LD01 and LD02).

The 3MHz PWM switching frequency allows the use of very small external inductors and capacitors. Both step-down converters can enter skip mode under light load conditions to further improve the efficiency and maximize the battery life. For noise sensitive applications, they can also be programmed through  $\rm I^2C$  interface to operate in forced PWM mode regardless of the load current condition. The  $\rm I^2C$  interface supports on-the-fly slew rate control of the output voltage from 0.825V to 3.6V at 25mV/step size for dynamic power saving. Each step-down converter can supply up to 800mA load current. The default output voltage can be set from 0.8V to  $\rm V_{IN}$  using external feedback resistors on the adjustable version, or the ISL9305 can be ordered in factory pre-set power-up default voltages in increments of 100mV from 0.9V to 3.6V.

The ISL9305 also provides two 300mA low dropout (LDO) regulators. The input voltage range is 1.5V to 5.5V allowing them to be powered from one of the on-chip step-down converters or directly from the battery. The default LDO power-up output comes with factory pre-set fixed output voltage options between 0.9V to 3.3V.

The ISL9305 is available in a 4mmx4mm 16 Ld TQFN package.

### **Related Literature**

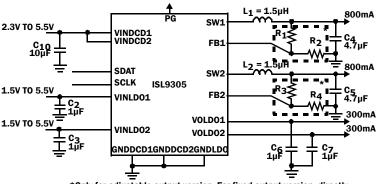
- FN7724, ISL9305H Data Sheet
- AN1564 "ISL9305IRTZEVAL1Z and ISL9305HIRTZEVAL1Z Evaluation Boards"

### **Features**

- Dual 800mA, Synchronous Step-down Converters and Dual 300mA, General-purpose LDOs
- Input Voltage Range
- 400kb/s I<sup>2</sup>C-bus Series Interface Transfers the Control Data Between the Host Controller and the ISL9305
- Adjustable Output Voltage
  - DCD1/DCD2 . . . . . . . . . . . . . . . 0.8V to V<sub>IN</sub>
- Fixed Output I<sup>2</sup>C Programmability
- LD01/LD02 Output Voltage I<sup>2</sup>C Programmability
- 50 $\mu$ A I $_{\rm Q}$  (Typ) with DCD1/DCD2 in Skip Mode; 20 $\mu$ A I $_{\rm Q}$  (Typ) for each Enabled LDO
- On-the-fly I<sup>2</sup>C Programming of DC/DC and LDO Output Voltages
- DCD1/DCD2 I<sup>2</sup>C Programmable Skip Mode Under Light Load or Forced Fixed Switching Frequency PWM Mode
- Small, Thin, 4mmx4mm TQFN Package

### **Applications**

- · Cellular Phones, Smart Phones
- · PDAs, Portable Media Players, Portable Instruments
- Single Li-ion/Li-Polymer Battery-Powered Equipment
- DSP Core Power



\*Only for adjustable output version. For fixed output version, directly connect the FB pin to the output of the buck converter.

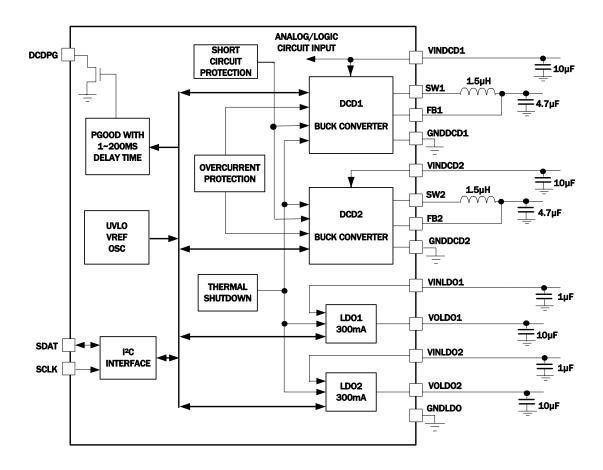
FIGURE 1. TYPICAL APPLICATION DIAGRAM

**TABLE 1. TYPICAL APPLICATION PART LIST** 

PARTS	DESCRIPTION	MANUFACTURER	PART NUMBER	SPECIFICATIONS	SIZE
L1, L2	Inductor	Sumida	CDRH2D14NP-1R5	$\textbf{1.5} \mu \text{H}/\textbf{1.80A}/\textbf{50} \text{m}\Omega$	3.0mmx3.0mmx1.55mm
C1	Input capacitor	Murata	GRM21BR60J106KE19L	10µF/6.3V	0805
C2, C3	Input capacitor	Murata	GRM185R60J105KE26D	1μF/6.3V	0603
C4, C5	Output capacitor	Murata	GRM219R60J475KE01D	4.7μF/6.3V	0805
C6, C7	Output capacitor	Murata	GRM185R60J105KE26D	1μF/6.3V	0603
R1, R2, R3, R4	Resistor	Various		1%, SMD, $0.1\Omega$	0603

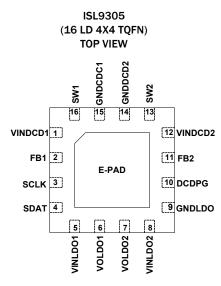
#### NOTE:

# **Block Diagram**



<sup>1.</sup> C4 and C5 are  $10\mu F/6.3V$  for VODCD less than 1V.

# **Pin Configuration**



# **Pin Descriptions**

	_	
PIN NUMBER (TQFN)	NAME	DESCRIPTION
1	VINDCD1	Input voltage for buck converter DCD1 and it also serves as the power supply pin for the whole internal digital/ analog circuits.
2	FB1	Feedback pin for DCD1, connect external voltage divider resistors between DCDC1 output, this pin and ground. For fixed output versions, connect this pin directly to the DCD1 output.
3	SCLK	1 <sup>2</sup> C interface clock pin.
4	SDAT	I <sup>2</sup> C interface data pin.
5	VINLD01	Input voltage for LD01.
6	VOLD01	Output voltage of LDO1.
7	VOLD02	Output voltage of LDO2.
8	VINLD02	Input voltage for LD02.
9	GNDLDO	Power ground for LD01 and LD02.
10	DCDPG	The DCDPG pin is an open-drain output to indicate the state of the DCD1/DCD2 output voltages. When both DCD1 and DCD2 are enabled, the output is released to be pulled high by an external pull-up resistor if both converter voltages are within the power-good range. The pin will be pulled low if either DCD is outside their range. When only one DCD is enabled, the state of the enabled DCD's output will define the state of the DCDPG pin. The DCDPG state can be programmed for a delay of up to 200ms before being released to rise high. The programming range is 1ms~200ms through the I <sup>2</sup> C interface.
11	FB2	Feedback pin for DCD2, connect external voltage divider resistors between DCD2 output, this pin and ground. For fixed output versions, connect this pin directly to the DCD2 output.
12	VINDCD2	Input voltage for buck converter DCD2.
13	SW2	Switching node for DCD2, connect to one terminal of the inductor.
14	GNDDCD2	Power ground for DCD2.
15	GNDDCD1	Power ground for DCD1.
16	SW1	Switching node for DCD1, connect to one terminal of the inductor.
E-pad	E-pad	Exposed Pad. Connect to system ground.

# **Ordering Information**

PART NUMBER (Notes 1, 2, 3)	PART MARKING	FBSEL DCD1 (V)	FBSEL DCD2 (V)	SLV LD01 (V)	SLV LD02 (V)	TEMP. RANGE (°C)	PACKAGE Tape and Reel (Pb-free)	PKG. DWG. #
ISL9305IRTAANLZ-T	9305I AANLZ	Adj	Adj	3.3	2.9	-40 to +85	16 Ld TQFN	L16.4x4G
ISL9305IRTBCNLZ-T	9305I BCNLZ	1.5	1.8	3.3	2.9	-40 to +85	16 Ld TQFN	L16.4x4G
ISL9305IRTBFNCZ-T	9305I BFNCZ	1.5	2.5	3.3	1.8	-40 to +85	16 Ld TQFN	L16.4x4G
ISL9305IRTWBNLZ-T	9305I WBNLZ	1.2	1.5	3.3	2.9	-40 to +85	16 Ld TQFN	L16.4x4G
ISL9305IRTWCLBZ-T	9305I WCLBZ	1.2	1.8	2.9	1.5	-40 to +85	16 Ld TQFN	L16.4x4G
ISL9305IRTWCNLZ-T	9305I WCNLZ	1.2	1.8	3.3	2.9	-40 to +85	16 Ld TQFN	L16.4x4G
ISL9305IRTWCNYZ-T	9305I WCNYZ	1.2	1.8	3.3	0.9	-40 to +85	16 Ld TQFN	L16.4x4G
ISL9305IRTWLNCZ-T	9305I WLNCZ	1.2	2.9	3.3	1.8	-40 to +85	16 Ld TQFN	L16.4x4G
ISL9305IRTZEVAL1Z	Evaluation Board	1	1	ı	1	ı	1	1

#### NOTES:

- 1. Please refer to TB347 for details on reel specifications.
- These Intersil Pb-free plastic packaged products employ special Pb-free material sets, molding compounds/die attach materials, and 100% matte
  tin plate plus anneal (e3 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations). Intersil
  Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.
- 3. For Moisture Sensitivity Level (MSL), please see device information page for ISL9305. For more information on MSL please see techbrief TB363.

### **Absolute Maximum Ratings** (Refer to ground)

SW1, SW21.5V to	6.5V
FB1, FB2	3.6V
GNDDCD1, GNDDCD2, GNDLD0	0.3۷ د
All other pins0.3V to	6.5V
ESD Ratings	
Human Body Model (Tested per JESD22-A114F)	3.5kV
Machine Model (Tested per JESD22-A115-A)	2.2kV
Charged Device Model (Tested per JESD22-C101D)	225V
Latch Up (Tested per JESD78B, Class II, Level A)	ე0mA

### **Thermal Information**

Thermal Resistance (Typical)	$\theta_{JA}$ (°C/W)	θ <sub>JC</sub> (°C/W)
16 Ld TQFN Package (Notes 4, 5)	42	5
Maximum Junction Temperature Range	4	0°C to +150°C
Recommended Junction Temperature Range	4	0°C to +125°C
Storage Temperature Range	6	5°C to +150°C
Pb-free reflow profile		. see link below
http://www.intersil.com/pbfree/Pb-FreeRe	eflow.asp	

### **Recommended Operating Conditions**

VINDCD1	2.3V to 5.5V
VINDCD2	2.3V to VINDCD1
VINLDO1 and VINLDO2	1.5V to VINDCD1
DCD1 and DCD2 Output Current	0mA to 800mA
LD01 and LD02 Output Current	0mA to 300mA
Operating Ambient Temperature	40°C to +85°C

CAUTION: Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions may adversely impact product reliability and result in failures not covered by warranty.

#### NOTES:

- 4. θ<sub>JA</sub> is measured in free air with the component mounted on a high effective thermal conductivity test board with "direct attach" features. See Tech Brief TB379.
- 5.  $\theta_{\text{JC}}$ , "case temperature" location is at the center of the exposed metal pad on the package underside.

**Electrical Specifications** Unless otherwise noted, all parameter limits are guaranteed over the recommended operating conditions and the typical specifications are measured at the following conditions:  $T_A = +25$ °C, VINDCD1 = 3.6V, VINDCD2 = 3.3V. For LD01 and LD02, VINLD0x = VOLD0x + 0.5V to 5.5V with VINLD0x always no higher than VINDCD1,  $L_1 = L_2 = 1.5\mu$ H,  $C_1 = 10\mu$ F,  $C_4 = C_5 = 4.7\mu$ F,  $C_2 = C_3 = C_6 = C_7 = 1\mu$ F,  $I_{OUT} = 0$ A for DCD1, DCD2, LD01 and LD02 (see Figure 1 on page 1 for more details).

PARAMETER	SYMBOL	TEST CONDITIONS	MIN (Note 6)	TYP	MAX (Note 6)	UNIT
VINDCD1, VINDCD2 Voltage Range			2.3	-	5.5	V
VINDCD1, VINDCD2 Undervoltage	V <sub>UVLO</sub>	Rising	-	2.2	2.3	V
Lockout Threshold		Falling	1.9	2.1	-	V
Quiescent Supply Current on VINDCD1	I <sub>VIN1</sub>	Only DCD1 enabled, no load and no switching on DCD1	-	40	60	μΑ
	I <sub>VIN2</sub>	Only DCD1 and LDO1 enabled, with no load and no switching on DCD1	-	65	95	μΑ
	I <sub>VIN3</sub>	Both DCD1 and DCD2 enabled, no load and no switching on both DCD1 and DCD2	-	50	75	μΑ
	I <sub>VIN4</sub>	Only LDO1 and LDO2 enabled	-	75	100	μΑ
	I <sub>VIN5</sub>	DCD1, DCD2, LD01 and LD02 are enabled, with no load and no switching on both DCD1 and DCD2	-	100	130	μА
	I <sub>VIN6</sub>	Only one DCD in forced PWM mode, no load	-	4	7.5	mA
Shutdown Supply Current	I <sub>SD</sub>	VINDCD1 = 5.5V, DCD1, DCD2, LDO1 and LDO2 are disabled through I <sup>2</sup> C interface, VINDCD1 = 4.2V	-	0.15	5	μА
Thermal Shutdown			-	155	-	°C
Thermal Shutdown Hysteresis			-	30	-	°C
DCD1 AND DCD2	1	,	ı	1	I	ı
FB1, FB2 Regulation Voltage	V <sub>FB</sub>		0.785	0.8	0.815	V
FB1, FB2 Bias Current	I <sub>FB</sub>	FB = 0.75V	-	0.001	-	μΑ

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PARAMETER SYMBOL TEST CONDITIONS		MIN (Note 6)	ТҮР	MAX (Note 6)	UNIT	
Output Voltage Accuracy		V <sub>IN</sub> = V <sub>O</sub> + 0.5V to 5.5V (minimal 2.3V), 1mA load	-3	-	+3	%
Line Regulation		V <sub>IN</sub> = V <sub>O</sub> + 0.5V to 5.5V (minimal 2.3V)	-	0.1	-	%/V
Maximum Output Current			800	-	-	mA
P-Channel MOSFET ON-resistance		V <sub>IN</sub> = 3.6V, I <sub>O</sub> = 200mA	-	0.14	0.2	Ω
		V <sub>IN</sub> = 2.3V, I <sub>O</sub> = 200mA	-	0.24	0.40	Ω
N-Channel MOSFET ON-resistance		V <sub>IN</sub> = 3.6V, I <sub>O</sub> = 200mA	-	0.11	0.2	Ω
		V <sub>IN</sub> = 2.3V, I <sub>O</sub> = 200mA	-	0.18	0.34	Ω
P-Channel MOSFET Peak Current Limit	I <sub>PK</sub>		1.075	1.3	1.6	Α
SW Maximum Duty Cycle			-	100	-	%
SW Leakage Current		V <sub>IN</sub> = 5.5V	-	0.005	1	μΑ
PWM Switching Frequency	f <sub>S</sub>		2.6	3.0	3.4	MHz
SW Minimum ON-time		VFB = 0.75V	-	70	-	ns
Bleeding Resistor			-	115	-	Ω
PG			I		l	
Output Low Voltage		Sinking 1mA, FB1 = FB2 = 0.7V	-	-	0.25	V
Rising Delay Time		Based on 1ms programmed nominal delay time	0.6	1.1	1.8	ms
Falling Delay Time		Based on 1ms programmed nonimal delay time	-	30	-	μs
PG Pin Leakage Current		PG = VINDCD1 = VINDCD2 = 3.6V	-	0.005	0.1	μΑ
PG Low Rising Threshold		Percentage of nominal regulation voltage	-	91	-	%
PG Low Falling Threshold		Percentage of nominal regulation voltage	-	87	-	%
PG High Rising Threshold		Percentage of nominal regulation voltage	-	112	-	%
PG High Falling Threshold		Percentage of nominal regulation voltage	-	109	-	%
LD01 AND LD02			I		l	
VINLDO1, VINLDO2 Supply Voltage		No higher than VINDCD1	1.5	-	5.5	V
VINLDO1, VINLDO2 Undervoltage	V <sub>UVLO</sub>	VINDCD1 = 2.3V, Rising	-	1.41	1.46	V
Lock-out Threshold		VINDCD1 = 2.3V, Falling	1.33	1.37	-	V
Internal Peak Current Limit			350	425	540	mA
Dropout Voltage		I <sub>O</sub> = 300mA, VO ≤ 2.1V	-	125	250	mV
		I <sub>O</sub> = 300mA, 2.1V < VO ≤ 2.8V	-	100	200	mV
		I <sub>O</sub> = 300mA, VO > 2.8V	-	80	170	mV
Power Supply Rejection Ratio		I <sub>0</sub> = 300mA @ 1kHz, V <sub>IN</sub> = 3.6V, VO = 2.6V, T <sub>A</sub> = +25°C	-	55	-	dB
Output Voltage Noise		V <sub>IN</sub> = 4.2V, I <sub>O</sub> = 10mA, T <sub>A</sub> = +25°C, BW = 10Hz to 100kHz	-	45	-	μV <sub>RMS</sub>

### NOTE:

6. Compliance to datasheet limits is assured by one or more methods: production test, characterization and/or design.

# **Theory of Operation**

### **DCD1** and **DCD2** Introduction

Both the DCD1 and DCD2 converters on ISL9305 use the peak-current-mode pulse-width modulation (PWM) control scheme for fast transient response and pulse-by-pulse current limiting. Both converters are able to supply up to 800mA load current. The default output voltage ranges from 0.8V to 3.6V depending on the factory pre-set configuration and can be programmed via the  $\rm I^2C$  interface in the range of 0.825V to 3.6V at 25mV/step with a programmable slew rate. An open-drain DCDPG (DCD Power-Good) signal is also provided to monitor the DCD1 and DCD2 output voltages. Optionally, both DCD1 and DCD2 can be programmed to be actively discharged via an on-chip bleeding resistor (typical  $115\Omega$ ) when the converter is disabled.

### Skip Mode (PFM Mode) for DCD1/DCD2

Under light load condition, the DCD1 and DCD2 can be programmed to automatically enter a pulse-skipping mode to minimize the switching loss by reducing the switching frequency. Figure 2 illustrates the skip mode operation. A zero-cross sensing circuit monitors the current flowing through the SW node for zero crossing. When it is detected to cross zero for 16 consecutive

cycles, the regulator enters the skip mode. During the 16 consecutive cycles, the inductor current could be negative. The counter is reset to zero when the sensed current flowing through the SW node does not cross zero during any cycle within the 16 consecutive cycles. Once the converter enters the skip mode, the pulse modulation is controlled by an internal comparator while each pulse cycle remains synchronized to the PWM clock. The P-Channel MOSFET is turned on at the rising edge of the clock and turned off when its current reaches ~20% of the peak current limit. As the average inductor current in each cycle is higher than the average current of the load, the output voltage rises cycle-over-cycle. When the output voltage is sensed to reach 1.5% above its nominal voltage, the P-Channel MOSFET is turned off immediately and the inductor current is fully discharged to zero and stays at zero. The output voltage reduces gradually due to the load current discharging the output capacitor. When the output voltage drops to the nominal voltage, the P-Channel MOSFET will be turned on again, repeating the previous operations.

The regulator resumes normal PWM mode operation when the output voltage is sensed to drop below 1.5% of its nominal voltage value as shown in Figure 3.

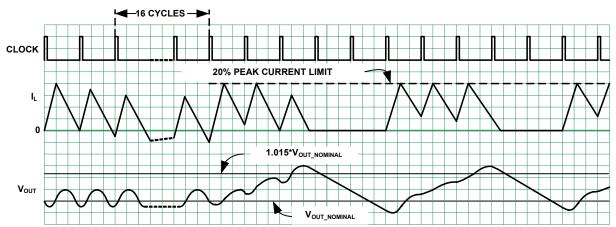


FIGURE 2. SKIP MODE OPERATION WAVEFORMS

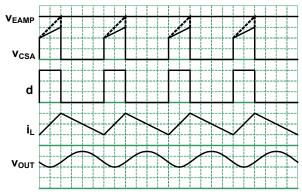


FIGURE 3. PWM OPERATION WAVEFORMS

#### **Soft-Start**

The soft-start reduces the in-rush current during the start-up stage. The soft-start block limits the current rising speed so that the output voltage rises in a controlled fashion.

#### **Overcurrent Protection**

The overcurrent protection for DCD1 and DCD2 is provided on ISL9305 for when an overload condition occurs. When the current at P-Channel MOSFET is sensed to reach the current limit, the internal protection circuit is triggered to turn off the P-Channel MOSFET immediately.

### **DCD Short-Circuit Protection**

The ISL9305 provides Short-Circuit Protection for both DCD1 and DCD2. The feedback voltage is monitored for output short-circuit protection. When the output voltage is sensed to be lower than a certain threshold, the internal circuit will change the PWM oscillator frequency to a lower frequencies in order to protect the IC from damage. The P-Channel MOSFET peak current limit remains active during this state.

### **Undervoltage Lock-out (UVLO)**

An undervoltage lock-out (UVLO) circuit is provided on ISL9305. The UVLO circuit block can prevent abnormal operation in the event that the supply voltage is too low to guarantee proper operation. The UVLO on VINDCD1 is set for a typical 2.2V with 100mV hysteresis. VINLDO1 and VINLDO2 are set for a typical 1.4V with 50mV hysteresis. When the input voltage is sensed to be lower than the UVLO threshold, the related channel is disabled.

### **DCDPG (DCD Power-Good)**

ISL9305 offers an open-drain Power-Good signal with programmable delay time for monitoring the converters DCD1 and DCD2 output voltages status.

When both DCD1 and DCD2 are enabled and their output voltages are within the power-good window, an internal power-good signal is issued to turn off the open-drain MOSFET so the DCDPG pin voltage can be externally pulled high after a programmed delay time. If either DCD1 or DCD2 output voltages or both of them are not within the power-good window, the DCDPG outputs an open-drain logic low signal after the programmed delay time.

When there is only one DCD converter (either DCD1 or DCD2) is enabled, then the DCDPG only indicates the status of this active DCD converter. For example, if only DCD1 converter is enabled and DCD2 converter is disabled, when DCD1 output is within the power-good window, internal power-good signal will be issued to turn off the open-drain MOSFET so the DCDPG pin voltage is externally pulled high after the programmed delay time. If output voltage of DCD1 is outside the power-good window, the DCDPG outputs an open-drain logic low signal after the programmed delay time. It is similar when only DCD2 is enabled and DCD1 is disabled. When both converters are disabled, DCDPG always outputs the open-drain logic low signal.

### **Low Dropout Operation**

Both DCD1 and DCD2 converters feature the low dropout operation to maximize the battery life. When the input voltage drops to a level that the converter can no longer operate under switching regulation to maintain the output voltage, the P-Channel MOSFET is completely turned on (100% duty cycle). The dropout voltage under such a condition is the product of the load current and the ON-resistance of the P-Channel MOSFET. Minimum required input voltage V<sub>IN</sub> under such condition is the sum of output voltage plus the voltage drop across the inductor and the P-Channel MOSFET switch.

# Active Output Voltage Discharge For DCD1/DCD2

The ISL9305 offers a feature to actively discharge the output voltage of DCD1 and DCD2 via an internal bleeding resistor (typical 115 $\Omega$ ) when the channel is disabled. This feature is enabled by default, thus outputs can be disabled individually through programming the control bit in DCD\_PARAMETER register.

### **Thermal Shutdown**

The ISL9305 provides built-in thermal protection function with thermal shutdown threshold temperature set at  $+155\,^{\circ}$ C with  $+25\,^{\circ}$ C hysteresis (typical). When the die temperature is sensed to reach  $+155\,^{\circ}$ C, the regulator is completely shut down and as the temperature is sensed to drop to  $+130\,^{\circ}$ C (typical), the device resumes normal operation starting from the soft-start.

### **Board Layout Recommendations**

The ISL9305 is a high frequency switching charger and hence the PCB layout is a very important design practice to ensure a satisfactory performance.

The power loop is composed of the output inductor L, the output capacitor  $C_{OUT}$ , the SW pin and the PGND pin. It is important to make the power loop as small as possible and the connecting traces among them should be direct, short and wide; the same practice should be applied to the connection of the VIN pin, the input capacitor  $C_{IN}$  and PGND.

The switching node of the converter, the SW pin, and the traces connected to this node are very noisy, so keep the voltage feedback trace and other noise sensitive traces away from these noisy traces.

The input capacitor should be placed as close as possible to the VIN pin. The ground of the input and output capacitors should be connected as close as possible as well. In addition, a solid ground plane is helpful for a good EMI performance.

The ISL9305 employs a thermal enhanced TQFN package with an exposed pad. The exposed pad should be properly soldered on thermal pad of the board in order to remove heat from the IC. The thermal pad should be big enough for 9 vias as shown in Figure 4.

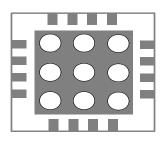


FIGURE 4. EXPOSED THERMAL PAD

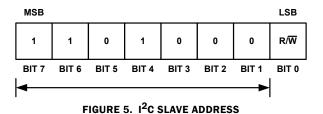
# I<sup>2</sup>C Compatible Interface

The ISL9305 offers an  $I^2C$  compatible interface, using two pins: SCLK for the serial clock and SDAT for serial data respectively. According to the  $I^2C$  specifications, a pull-up resistor is needed for the clock and data signals to connect to a positive supply. When the ISL9305 and the host use different supply voltages, the pull-up resistors should be connected to the higher voltage rail.

Signal timing specifications should satisfy the standard I<sup>2</sup>C bus specification. The maximum bit rate is 400kb/s and more details regarding the I<sup>2</sup>C specifications can be found from Philips.

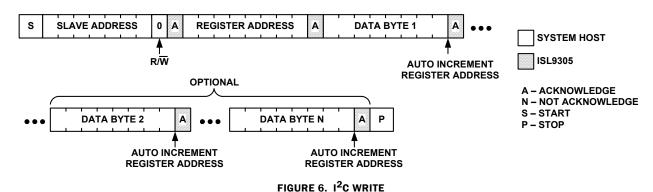
# I<sup>2</sup>C Slave Address

The ISL9305 serves as a slave device and the 7-bit default chip address is 1101000, as shown in Figure 5 According to the  $I^2C$  specifications, here the value of Bit 0 determines the direction of the message ("0" means "write" and "1" means "read").



### I<sup>2</sup>C Protocol

Figures 6, 7, and 8 show three typical I<sup>2</sup>C-bus transaction protocols.



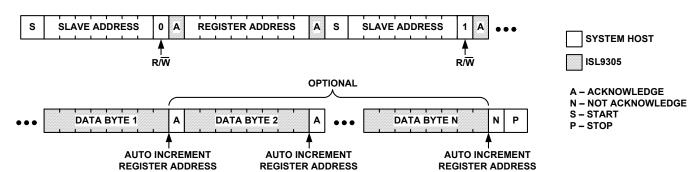


FIGURE 7. I<sup>2</sup>C READ SPECIFYING REGISTER ADDRESS

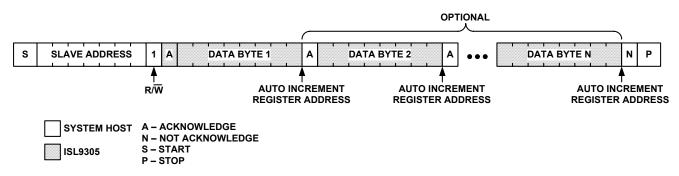


FIGURE 8. I<sup>2</sup>C READ NOT SPECIFYING REGISTER ADDRESS

# I<sup>2</sup>C Control Registers

All the registers are reset at initial start-up.

### **DCD OUTPUT VOLTAGE CONTROL REGISTER**

DCD10UT, address 0x00h; DCD20UT, address 0x01h

TABLE 2. BUCK CONVERTERS OUTPUT VOLTAGE CONTROL REGISTER

BIT	NAME	ACCESS	RESET	DESCRIPTION
В7	Reserve	-	0	Refer to Table 3
В6	DCDxOUT-6	R/W	0	
B5	DCDxOUT-5	R/W	0	
В4	DCDxOUT-4	R/W	1	
В3	DCDxOUT-3	R/W	0	Refer to Table 3
B2	DCDxOUT-2	R/W	0	
B1	DCDxOUT-1	R/W	0	
во	DCDxOUT-0	R/W	0	

### **ISL9305**

TABLE 3. DCD1 AND DCD2 OUTPUT VOLTAGE SETTING

DCDOUT <7:0>	DCD OUTPUT VOLTAGE (V)						
00	0.825	20	1.625	40	2.425	60	3.225
01	0.850	21	1.650	41	2.450	61	3.250
02	0.875	22	1.675	42	2.475	62	3.275
03	0.900	23	1.700	43	2.500	63	3.300
04	0.925	24	1.725	44	2.525	64	3.325
05	0.950	25	1.750	45	2.550	65	3.350
06	0.975	26	1.775	46	2.575	66	3.375
07	1.000	27	1.800	47	2.600	67	3.400
08	1.025	28	1.825	48	2.625	68	3.425
09	1.050	29	1.850	49	2.650	69	3.450
0A	1.075	2A	1.875	4A	2.675	6A	3.475
ОВ	1.100	2B	1.900	4B	2.700	6B	3.500
OC	1.125	2C	1.925	4C	2.725	6C	3.525
0D	1.150	2D	1.950	4D	2.750	6D	3.550
0E	1.175	2E	1.975	4E	2.775	6E	3.575
0F	1.200	2F	2.000	4F	2.800	6F	3.600
10	1.225	30	2.025	50	2.825		
11	1.250	31	2.050	51	2.850		
12	1.275	32	2.075	52	2.875		
13	1.300	33	2.100	53	2.900		
14	1.325	34	2.125	54	2.925		
15	1.350	35	2.150	55	2.950		
16	1.375	36	2.175	56	2.975		
17	1.400	37	2.200	57	3.000		
18	1.425	38	2.225	58	3.025		
19	1.450	39	2.250	59	3.050		
<b>1</b> A	1.475	ЗА	2.275	5A	3.075		
1B	1.500	3B	2.300	5B	3.100		
10	1.525	3C	2.325	5C	3.125		
1D	1.550	3D	2.350	5D	3.150		
1E	1.575	3E	2.375	5E	3.175		
1F	1.600	3F	2.400	5F	3.200		

# LDO1 AND LDO2 OUTPUT VOLTAGE CONTROL REGISTERS

LD010UT, address 0x02h and LD020UT, address 0x03h.

### TABLE 4. LDOX OUTPUT VOLTAGE CONTROL REGISTERS

BIT	NAME	ACCESS	RESET	DESCRIPTION
В7	Reserve	-	0	Refer to Table 5 for
В6	Reserve	-	0	output voltage settings
В5	LD0x0UT-5	R/W	0	
В4	LD0x0UT-4	R/W	0	
В3	LD0x0UT-3	R/W	1	
B2	LD0x0UT-2	R/W	1	
B1	LD0x0UT-1	R/W	0	
В0	LD0x0UT-0	R/W	0	

TABLE 5. LDOX OUTPUT VOLTAGE SETTINGS

LDOOUT <7:0>	LDO OUTPUT VOLTAGE (V)	LDOOUT <7:0>	LDO OUTPUT VOLTAGE (V)	LD00UT <7:0>	LDO OUTPUT VOLTAGE (V)	LD00UT <7:0>	LDO OUTPUT VOLTAGE (V)
00	0.90	10	1.70	20	2.50	30	3.30
01	0.95	11	1.75	21	2.55	31	3.35
02	1.00	12	1.80	22	2.60	32	3.40
03	1.05	13	1.85	23	2.65	33	3.45
04	1.10	14	1.90	24	2.70	34	3.50
05	1.15	15	1.95	25	2.75	35	3.55
06	1.20	16	2.00	26	2.80	36	3.60
07	1.25	17	2.05	27	2.85		
08	1.30	18	2.10	28	2.90		
09	1.35	19	2.15	29	2.95		
OA	1.40	1A	2.20	2A	3.00		
ОВ	1.45	1B	2.25	2B	3.05		
ос	1.50	10	2.30	20	3.10		
0D	1.55	1D	2.35	2D	3.15		
0E	1.60	1E	2.40	2E	3.20		
OF	1.65	<b>1</b> F	2.45	2F	3.25	•	

### **DCD1 AND DCD2 CONTROL REGISTER**

DCD\_PARAMETER, address 0x04h

### TABLE 6. DCD\_PARAMETER REGISTER

BIT	NAME	ACCESS	RESET	DESCRIPTION
В7	-	-	0	Reserved
В6	DCD_PHASE	R/W	0	DCD1 and DCD2 PWM switch selection. 0-in phase; 1 to 180° out-of-phase.
B5	DCD2_ULTRA	R/W	0	Ultrasonic feature under PFM mode for DCD2. 0-disabled; 1-enabled.
B4	DCD1_ULTRA	R/W	0	Ultrasonic feature under PFM mode for DCD1. 0-disabled; 1-enabled.
В3	DCD2_BLD	R/W	1	Selection of DCD2 for active output voltage discharge when disabled. 0-disabled; 1-enabled.
B2	DCD1_BLD	R/W	1	Selection of DCD1 for active output voltage discharge when disabled. 0-disabled; 1-enabled.
B1	DCD2_MODE	R/W	1	Selection on DCD2 of auto PFM/PWM mode (= 1) or forced PW mode (= 0).
во	DCD1_MODE	R/W	1	Selection on DCD1 of auto PFM/PWM mode (= 1) or forced PW mode (= 0).

### **SYSTEM CONTROL REGISTER**

SYS\_PARAMETER, address 0x05h

### TABLE 7. SYS\_PARAMETER REGISTER

BIT	NAME	ACCESS	RESET	DESCRIPTION
В7	-	-	0	Reserved
В6	I <sup>2</sup> C_EN	R/W	0	I <sup>2</sup> C function enable. 0-disabled; 1-enabled
В5	DCDPOR_1	R/W	1	DCDPOR Delay Time Setting,
В4	DCDPOR_0	R/W	0	DCDPOR[1:0]: 00 to 1ms 01 to 50ms 10 to 150ms 11 to 200m
В3	LDO2_EN	R/W	1	LD02 enable selection. 0-disable, 1-enable.
B2	LDO1_EN	R/W	1	LD01 enable selection. 0-disable, 1-enable
B1	DCD2_EN	R/W	1	DCD2 enable selection. 0-disable, 1-enable.
во	DCD1_EN	R/W	1	DCD2 enable selection. 0-disable, 1-enable

# DCD OUTPUT VOLTAGE SLEW RATE CONTROL REGISTER

DCD\_SRCTL, address 0x06h

TABLE 8.

BIT	NAME	ACCESS	RESET	DESCRIPTION
В7	DCD2SR_2	R/W	0	DCD2 Slew Rate Setting,
В6	DCD2SR_1	R/W	0	DCD2SR[2:0]: 000 to 0.225mV/µs
B5	DCD2SR_0	R/W	1	001 to 0.45mV/μs 010 to 0.90mV/μs 011 to 1.8mV/μs 100 to 3.6mV/μs 101 to 7.2mV/μs 110 to 14.4mV/μs 111 to immediate
В4	Reserve	-	0	Reserved
В3	DCD1SR_2	R/W	0	DCD1 Slew Rate Setting,
B2	DCD1SR_1	R/W	0	DCD1SR[2:0]: 000 to 0.225mV/µs
B1	DCD1SR_0	R/W	1	001 to 0.45mV/μs 010 to 0.90mV/μs 011 to 1.8mV/μs 100 to 3.6mV/μs 101 to 7.2mV/μs 110 to 14.4mV/μs 111 to immediate
В0	Reserve	-	0	Reserved

# **Typical Operating Conditions**

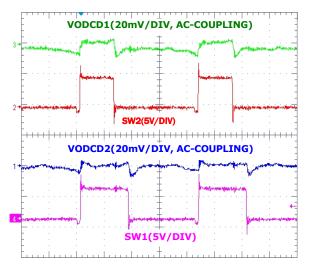


FIGURE 9. DCD OUTPUT VOLTAGE RIPPLE ( $V_{IN}$  = 4.2V, FULL LOAD AT DCD1 AND DCD2)

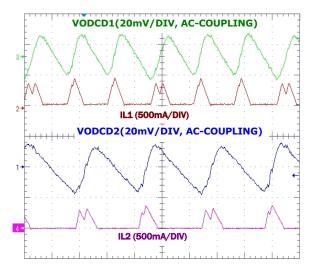


FIGURE 10. DCD OUTPUT VOLTAGE RIPPLE ( $V_{IN} = 4.2V$ , PFM MODE)

# **Typical Operating Conditions (Continued)**

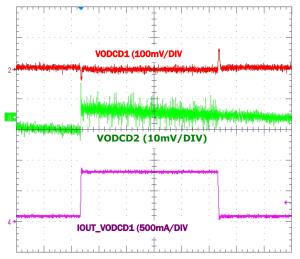


FIGURE 11. DCD OUTPUT TRANSIENT RESPONSE (V<sub>IN</sub> = 4.2V, LOAD STEP: 80mA to 800mA)

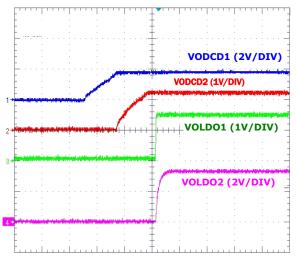


FIGURE 13. START-UP SEQUENCY (VIN = 4.2V, NO LOAD)

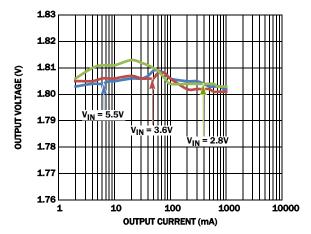


FIGURE 15. DCD OUTPUT VOLTAGE vs LOAD (V<sub>OUT</sub> = 1.2V, PFM/PWM)

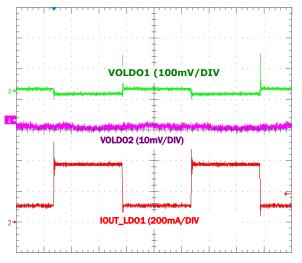


FIGURE 12. LD0 OUTPUT TRANSIENT RESPONSE ( $V_{IN}$  = 4.2V, STEP LOAD: 30mA TO 300mA)

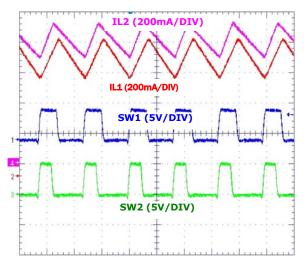


FIGURE 14. DCD1 and DCD2 SWITCHING WAVEFORM (V $_{\rm IN}$  = 5V, FULL LOAD ON TWO CHANNELS)

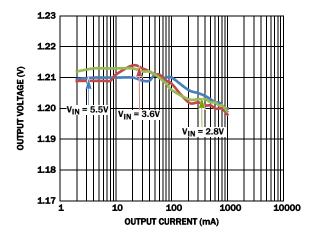


FIGURE 16. DCD OUTPUT VOLTAGE vs LOAD ( $V_{OUT} = 1.2V$ , PFM/PWM)

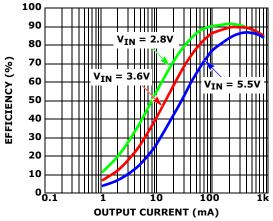


FIGURE 17. EFFICIENCY vs OUTPUT CURRENT (V<sub>OUT</sub> = 1.8V, FORCED PWM MODE)

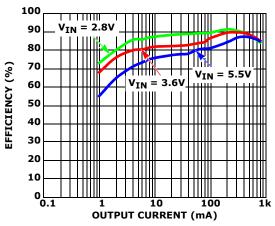


FIGURE 18. EFFICIENCY vs OUTPUT CURRENT (V<sub>OUT</sub> = 1.8V, PFM to PWM)

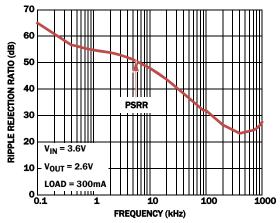


FIGURE 19. RIPPLE REJECTION RATIO vs FREQUENCY

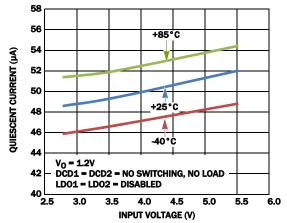


FIGURE 20. QUIESCENT CURRENT vs INPUT VOLTAGE

### **Revision History**

The revision history provided is for informational purposes only and is believed to be accurate, but not warranted. Please go to web to make sure you have the latest Rev.

DATE	REVISION	CHANGE
11/8/10	FN7605.0	Initial Release

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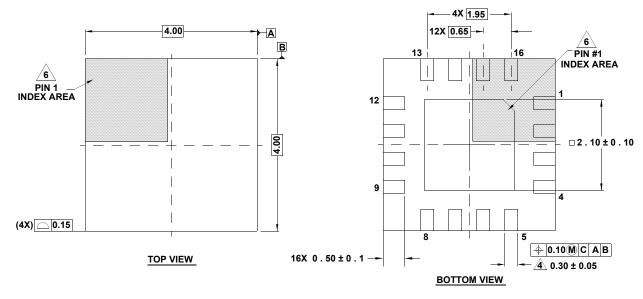
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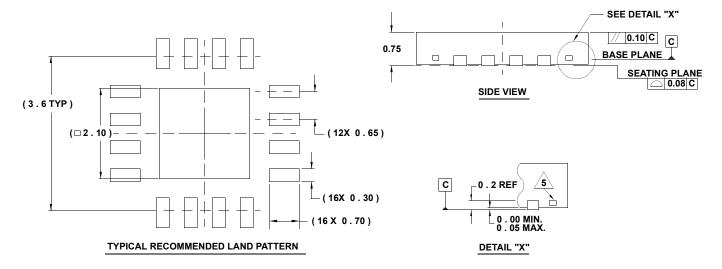
<sup>\*</sup>For a complete listing of Applications, Related Documentation and Related Parts, please see the respective device information page on intersil.com: ISL9305

### **Package Outline Drawing**

### L16.4x4G

16 LEAD THIN QUAD FLAT NO-LEAD PLASTIC PACKAGE Rev 0, 4/10





### NOTES:

- Dimensions are in millimeters.
   Dimensions in ( ) for Reference Only.
- ${\bf 2.} \quad {\bf Dimensioning\ and\ tolerancing\ conform\ to\ ASME\ Y14.5m-1994}.$
- 3. Unless otherwise specified, tolerance : Decimal  $\pm$  0.05
- <u>4.</u> Dimension applies to the metallized terminal and is measured between 0.15mm and 0.30mm from the terminal tip.
- 5. Tiebar shown (if present) is a non-functional feature.
- 6. The configuration of the pin #1 identifier is optional, but must be located within the zone indicated. The pin #1 identifier may be either a mold or mark feature.
- 7. JEDEC reference drawing: MO220K.