

# **SGM8541**

# **SGM8542**

# **SGM8544**

## **1.1MHz, 42 $\mu$ A, Rail-to-Rail I/O CMOS Operational Amplifier**

### **PRODUCT DESCRIPTION**

The SGM8541 (single) , SGM8542 (dual) and SGM8544 (quad) are rail-to-rail input and output voltage feedback amplifiers offering low cost. They have a wide input common-mode voltage range and output voltage swing, and take the minimum operating supply voltage down to 2.1V and the maximum recommended supply voltage is 5.5 V. All are specified over the extended -40°C to +125°C temperature range.

The SGM8541/8542/8544 provide 1.1MHz bandwidth at a low current consumption of 42 $\mu$ A per amplifier. Very low input bias currents of 0.5pA, enable SGM8541/8542 /8544 to be used for integrators, photodiode amplifiers, and piezoelectric sensors. Rail-to-rail inputs and outputs are useful to designers buffering ASIC in single-supply systems.

Applications for this amplifiers include safety monitoring, portable equipment, battery and power supply control, and signal conditioning and interfacing for transducers in very low power systems.

The SGM8541 is available in the tiny SOT23-5 and SO-8 packages. The SGM8542 comes in the miniature SO-8 and MSOP-8 packages. The SGM8544 is offered in TSSOP-16 and SO-16 packages.

### **APPLICATIONS**

ASIC Input or Output Amplifier

Sensor Interface

Piezo Electric Transducer Amplifier

Medical Instrumentation

Mobile Communication

Audio Output

Portable Systems

Smoke Detectors

Mobile Telephone

Notebook PC

PCMCIA cards

Battery –Powered equipment

### **FEATURES**

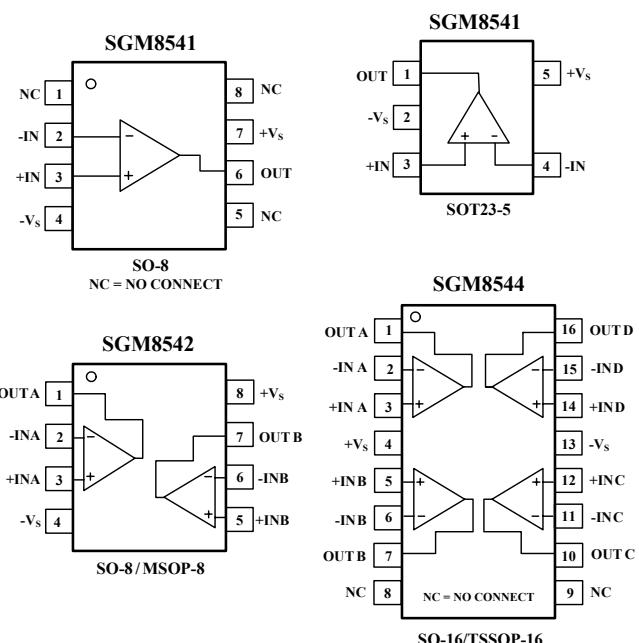
- Low Cost
- Rail-to-Rail Input and Output  
    0.8mV Typical Vos
- Unity Gain Stable
- Gain Bandwidth Product: : 1.1MHz
- Very low input bias currents : 0.5pA
- Operates on 2.1 V to 5.5 V Supplies
- Input Voltage Range = - 0.1 V to +5.6 V with Vs = 5.5 V
- Low Supply Current: 42 $\mu$ A/Amplifier
- Small Packaging

SGM8541 Available in SO-8 and SOT23-5

SGM8542 Available in SO-8 and MSOP-8

SGM8544 Available in SO-16 and TSSOP-16

### **PIN CONFIGURATIONS** (Top View)



# ELECTRICAL CHARACTERISTICS : $V_S = +5V$

(At  $R_L = 100k\Omega$  connected to  $V_S/2$ , and  $V_{OUT} = V_S/2$ , unless otherwise noted)

PARAMETER	CONDITION	SGM8541/8542/8544							
		TYP	MIN/MAX OVER TEMPERATURE						UNITS
			+25°C	+25°C	0°C to 70°C	-40°C to 85°C	-40°C to 125°C		
<b>INPUT CHARACTERISTICS</b>									
Input Offset Voltage ( $V_{os}$ )			$\pm 0.8$	$\pm 3.5$	$\pm 3.7$	$\pm 4.1$	$\pm 5.4$	mV	MAX
Input Bias Current ( $I_B$ )			0.5					pA	TYP
Input Offset Current ( $I_{os}$ )			0.5					pA	TYP
Common-Mode Voltage Range ( $V_{CM}$ )	$V_S = 5.5V$		-0.1 to +5.6					V	TYP
Common-Mode Rejection Ratio(CMRR)	$V_S = 5.5V, V_{CM} = -0.1V$ to 4 V	88	72	72	72	72	dB	MIN	
	$V_S = 5.5V, V_{CM} = -0.1V$ to 5.6 V	78	60	59	59	57	dB	MIN	
Open-Loop Voltage Gain( $A_{OL}$ )	$R_L = 5K\Omega, V_o = 0.1V$ to 4.9V	90	80	80	80	78	dB	MIN	
	$R_L = 100K\Omega, V_o = 0.035V$ to 4.965V	94	85	85	85	82	dB	MIN	
Input Offset Voltage Drift ( $\Delta V_{os}/\Delta T$ )			2.7					$\mu V/^\circ C$	TYP
<b>OUTPUT CHARACTERISTICS</b>									
Output Voltage Swing from Rail	$R_L = 100K\Omega$	0.008					V	TYP	
Output Current ( $I_{out}$ )		23	20	19.1	18.8	18	mA	MIN	
<b>POWER SUPPLY</b>									
Operating Voltage Range			2.1	2.5	2.5	2.5	V	MIN	
			5.5	5.5	5.5	5.5	V	MAX	
Power Supply Rejection Ratio (PSRR)	$V_S = +2.5 V$ to + 5.5 V								
	$V_{CM} = (-V_S) + 0.5$	92	76	74	71	70	dB	MIN	
Quiescent Current / Amplifier ( $I_Q$ )	$I_{out} = 0$	42	54	59	60	64	$\mu A$	MAX	
<b>DYNAMIC PERFORMANCE</b>									
Gain-Bandwidth Product (GBP)	$C_L = 100pF$		1.1				MHz	TYP	
Slew Rate (SR)	$G = +1$ , 2V Output Step	0.52					V/ $\mu s$	TYP	
Settling Time to 0.1% ( $t_s$ )	$G = +1$ , 2 V Output Step	5.3					$\mu s$	TYP	
Overload Recovery Time	$V_{IN} \cdot \text{Gain} = V_S$	2.6					$\mu s$	TYP	
<b>NOISE PERFORMANCE</b>									
Voltage Noise Density ( $e_n$ )	$f = 1kHz$	27					$nV/\sqrt{Hz}$	TYP	
	$f = 10kHz$	20					$nV/\sqrt{Hz}$	TYP	

Specifications subject to change without notice.

## PACKAGE/ORDERING INFORMATION

MODEL	ORDER NUMBER	PACKAGE DESCRIPTION	PACKAGE OPTION	MARKING INFORMATION
SGM8541	SGM8541XN5/TR	SOT23-5	Tape and Reel, 3000	8541
	SGM8541XS/TR	SO-8	Tape and Reel, 2500	SGM8541XS
SGM8542	SGM8542XS/TR	SO-8	Tape and Reel, 2500	SGM8542XS
	SGM8542XMS/TR	MSOP-8	Tape and Reel, 3000	SGM8542XMS
SGM8544	SGM8544XS/TR	SO-16	Tape and Reel, 2500	SGM8544XS
	SGM8544XTS	TSSOP-16	Tape and Reel, 3000	SGM8544XTS

### ABSOLUTE MAXIMUM RATINGS

Supply Voltage, V+ to V- ..... 7.5 V  
 Common-Mode Input Voltage .....  $(-V_S) - 0.5 \text{ V}$  to  $(+V_S) + 0.5 \text{ V}$   
 Storage Temperature Range .....  $-65^\circ\text{C}$  to  $+150^\circ\text{C}$   
 Junction Temperature .....  $150^\circ\text{C}$   
 Operating Temperature Range .....  $-55^\circ\text{C}$  to  $+150^\circ\text{C}$   
 Package Thermal Resistance @  $T_A = 25^\circ\text{C}$   
 SOT23-5,  $\theta_{JA}$  .....  $190^\circ\text{C}/\text{W}$   
 SO-8,  $\theta_{JA}$  .....  $125^\circ\text{C}/\text{W}$   
 MSOP-8,  $\theta_{JA}$  .....  $216^\circ\text{C}/\text{W}$   
 SO-16,  $\theta_{JA}$  .....  $82^\circ\text{C}/\text{W}$   
 TSSOP-16,  $\theta_{JA}$  .....  $105^\circ\text{C}/\text{W}$   
 Lead Temperature Range (Soldering 10 sec) .....  $260^\circ\text{C}$

ESD Susceptibility  
 HBM ..... 4000V  
 MM ..... 400V

#### NOTES

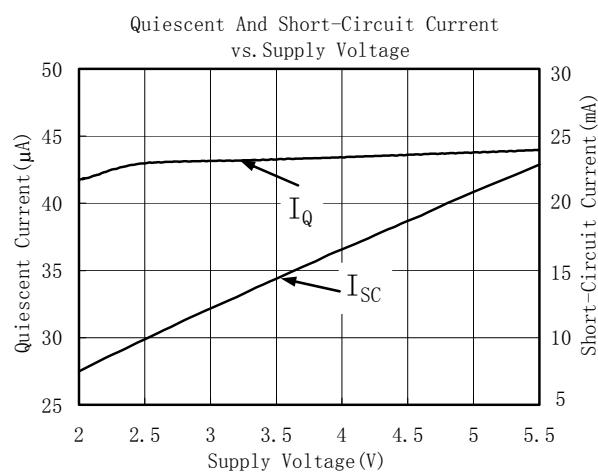
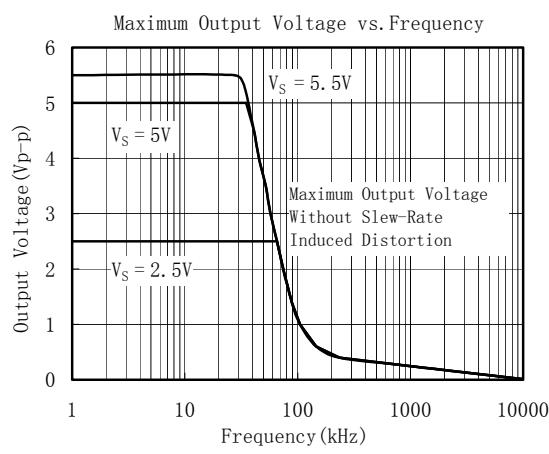
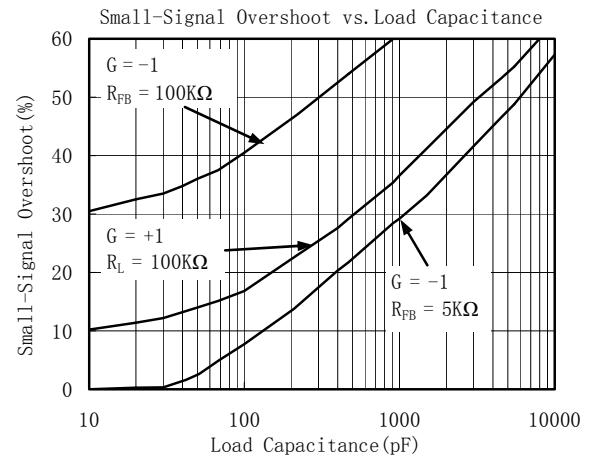
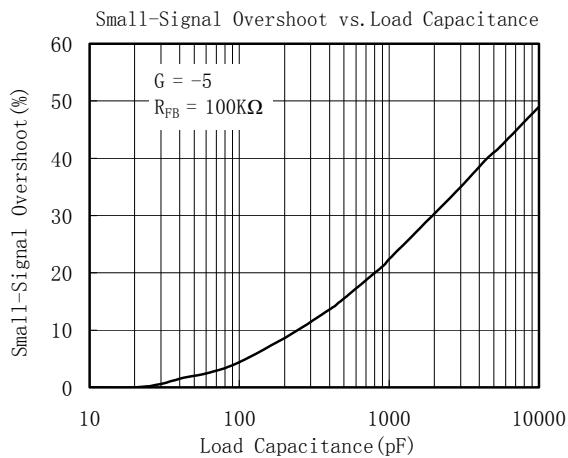
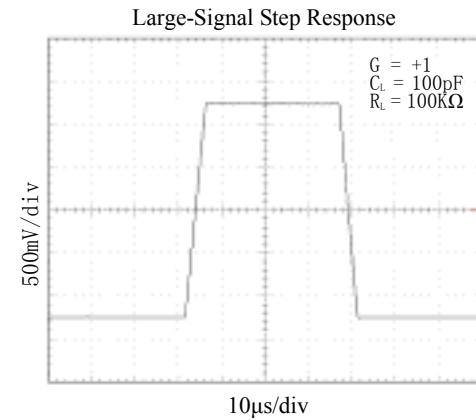
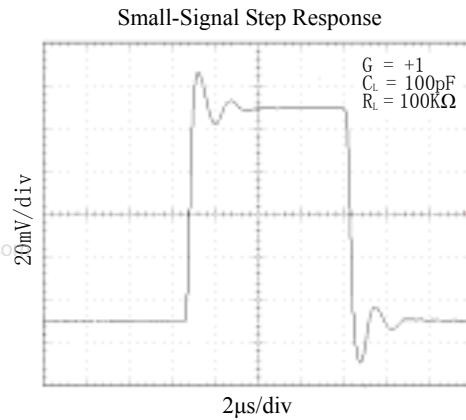
1. Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### CAUTION

This integrated circuit can be damaged by ESD. Shengbang Micro-electronics recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage. ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

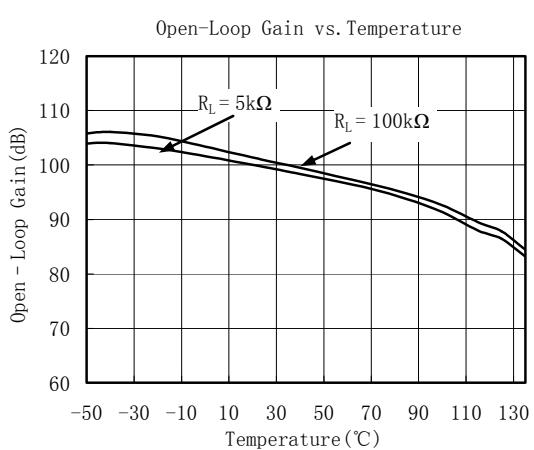
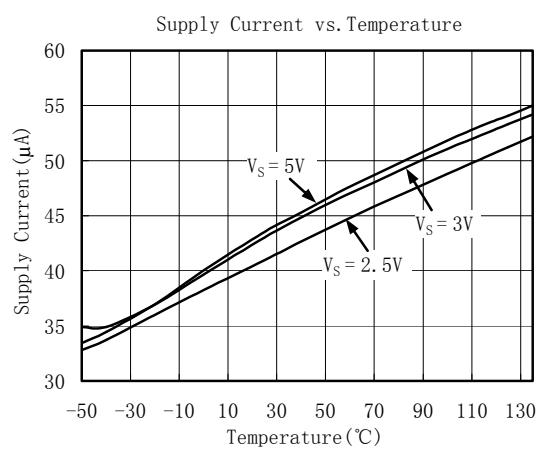
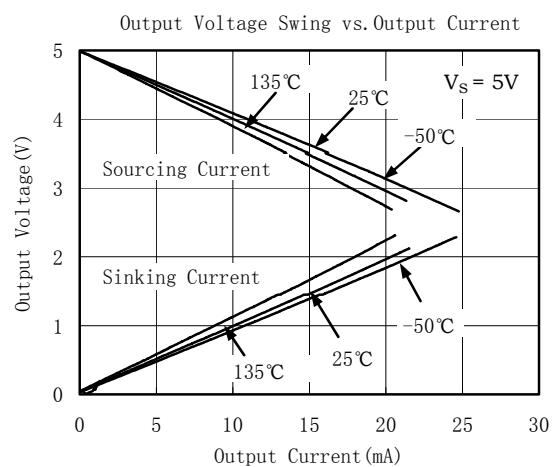
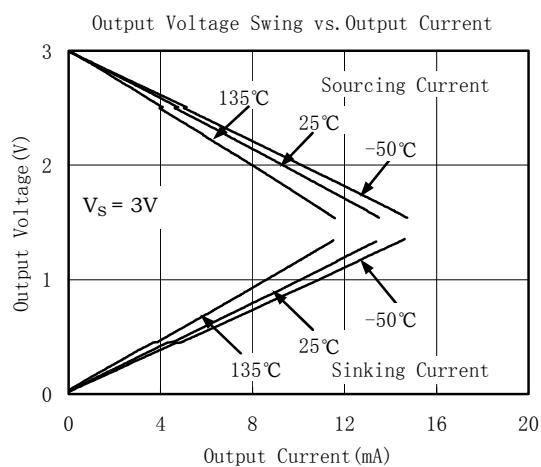
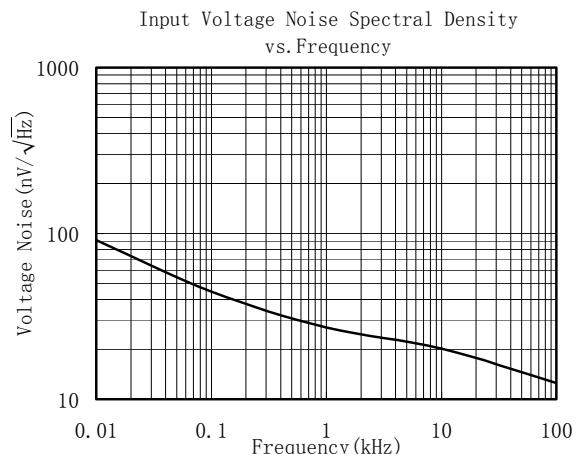
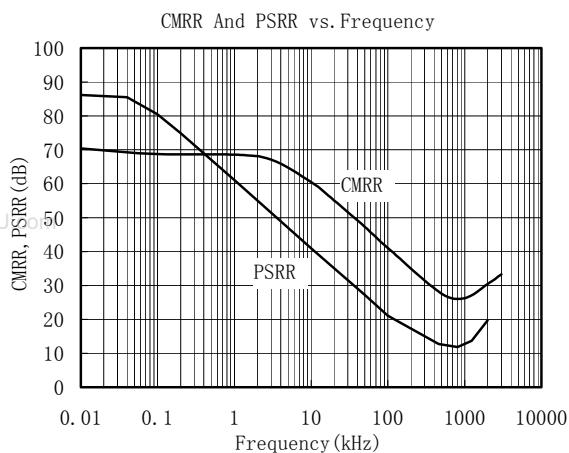
# TYPICAL PERFORMANCE CHARACTERISTICS

At  $T_A = +25^\circ\text{C}$ ,  $V_S = +5\text{V}$ , and  $R_L = 100\text{k}\Omega$  connected to  $V_S/2$ , unless otherwise noted.



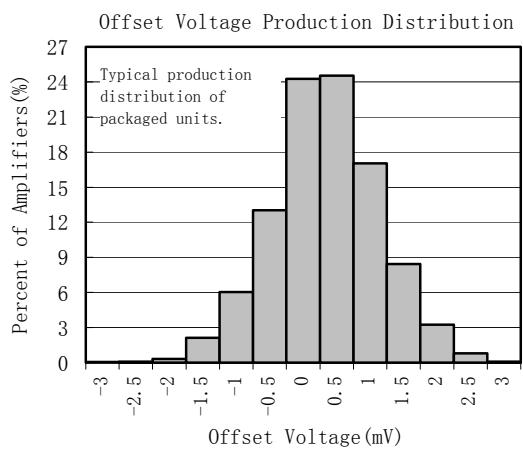
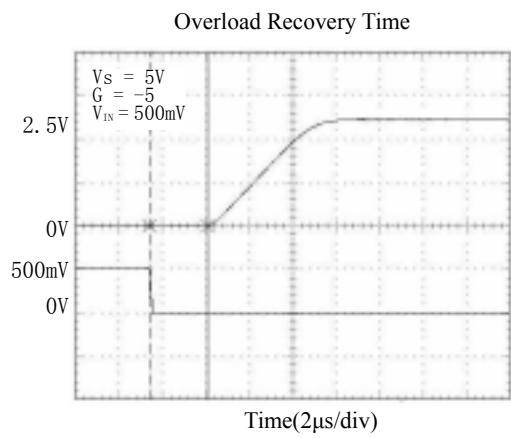
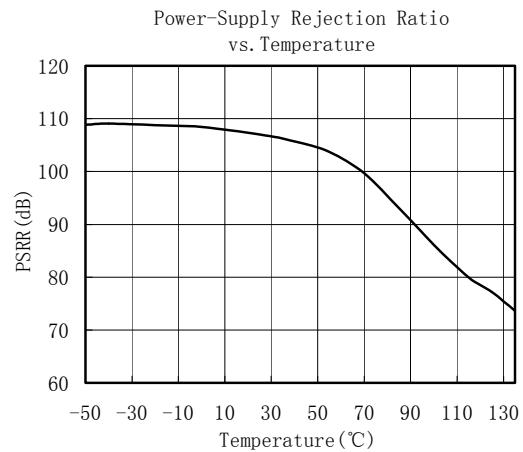
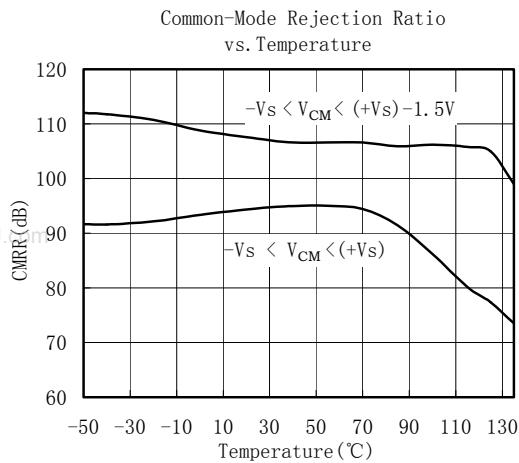
# TYPICAL PERFORMANCE CHARACTERISTICS

At  $T_A = +25^\circ\text{C}$ ,  $V_S = +5\text{V}$ , and  $R_L = 100\text{k}\Omega$  connected to  $V_S/2$ , unless otherwise noted.



# TYPICAL PERFORMANCE CHARACTERISTICS

At  $T_A = +25^\circ\text{C}$ ,  $V_S = +5\text{V}$ , and  $R_L = 100\text{k}\Omega$  connected to  $V_S/2$ , unless otherwise noted.



# APPLICATION NOTES

## Driving Capacitive Loads

The SGM854X can directly drive 250pF in unity-gain without oscillation. The unity-gain follower (buffer) is the most sensitive configuration to capacitive loading. Direct capacitive loading reduces the phase margin of amplifiers and this results in ringing or even oscillation. Applications that require greater capacitive drive capability should use an isolation resistor between the output and the capacitive load like the circuit in Figure 1. The isolation resistor  $R_{ISO}$  and the load capacitor  $C_L$  form a zero to increase stability. The bigger the  $R_{ISO}$  resistor value, the more stable  $V_{OUT}$  will be. Note that this method results in a loss of gain accuracy because  $R_{ISO}$  forms a voltage divider with the  $R_{LOAD}$ .

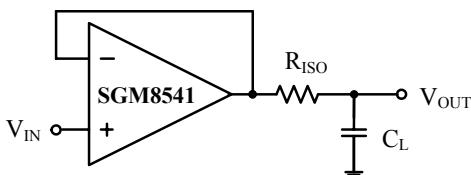


Figure 1. Indirectly Driving Heavy Capacitive Load

An improvement circuit is shown in Figure 2. It provides DC accuracy as well as AC stability.  $R_F$  provides the DC accuracy by connecting the inverting signal with the output.  $C_F$  and  $R_{ISO}$  serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving phase margin in the overall feedback loop.

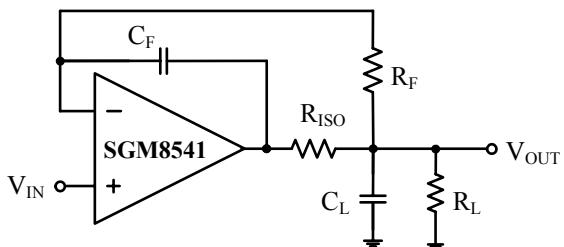


Figure 2. Indirectly Driving Heavy Capacitive Load with DC Accuracy

For no-buffer configuration, there are two others ways to increase the phase margin: (a) by increasing the amplifier's gain or (b) by placing a capacitor in parallel with the feedback resistor to counteract the parasitic capacitance associated with inverting node.

## Power-Supply Bypassing and Layout

The SGM854X family operates from either a single +2.5V to +5.5V supply or dual  $\pm 1.25V$  to  $\pm 2.75V$  supplies. For single-supply operation, bypass the power supply  $V_{DD}$  with a  $0.1\mu F$  ceramic capacitor which should be placed close to the  $V_{DD}$  pin. For dual-supply operation, both the  $V_{DD}$  and the  $V_{SS}$  supplies should be bypassed to ground with separate  $0.1\mu F$  ceramic capacitors.  $2.2\mu F$  tantalum capacitor can be added for better performance.

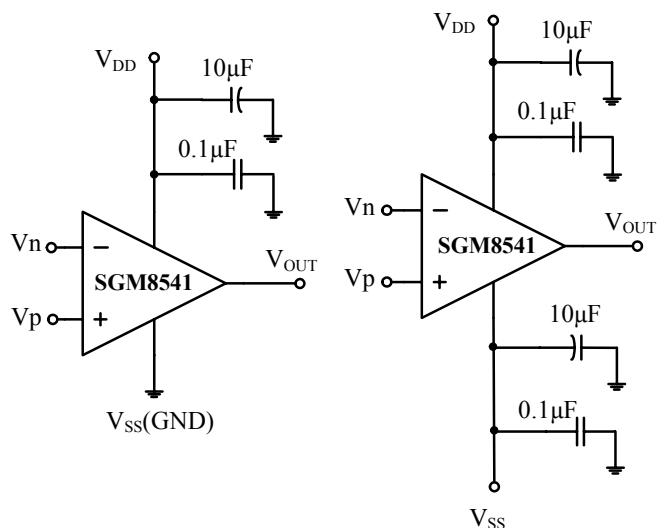


Figure 3. Amplifier with Bypass Capacitors

# Typical Application Circuits

## Differential Amplifier

The circuit shown in Figure 4 performs the difference function. If the resistors ratios are equal ( $R_4 / R_3 = R_2 / R_1$ ), then  $V_{OUT} = (V_p - V_n) \times R_2 / R_1 + V_{ref}$ .

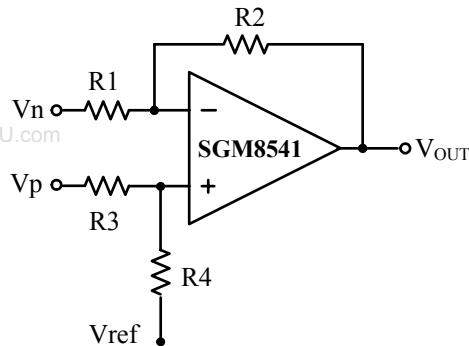


Figure 4. Differential Amplifier

## Instrumentation Amplifier

The circuit in Figure 5 performs the same function as that in Figure 4 but with the high input impedance.

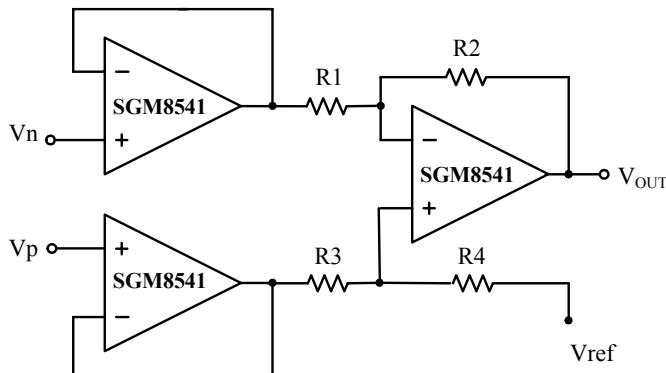


Figure 5. Instrumentation Amplifier

## Low Pass Active Filter

The low pass filter shown in Figure 6 has a DC gain of  $(-R_2 / R_1)$  and the  $-3\text{dB}$  corner frequency is  $1/2\pi R_2 C$ . Make sure the filter is within the bandwidth of the amplifier. The Large values of feedback resistors can couple with parasitic capacitance and cause undesired effects such as ringing or oscillation in high-speed amplifiers. Keep resistors value as low as possible and consistent with output loading consideration.

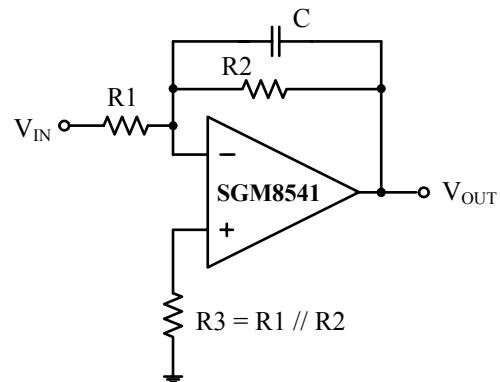
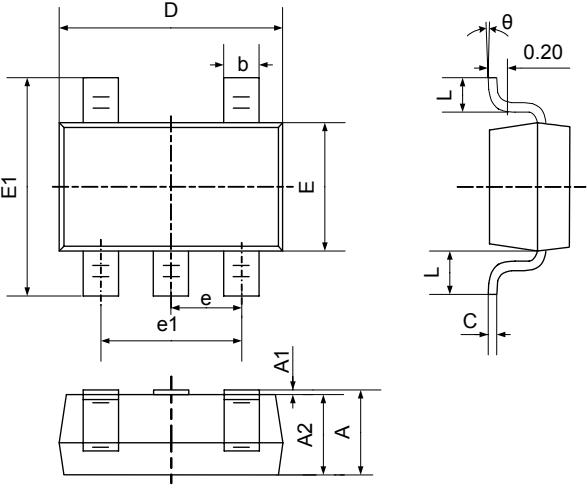


Figure 6. Low Pass Active Filter

# PACKAGE OUTLINE DIMENSIONS

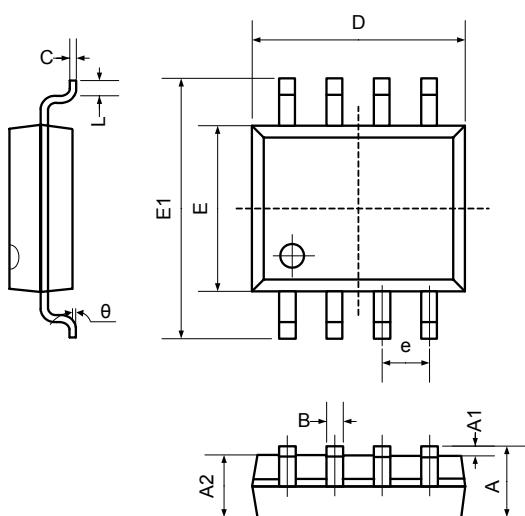
SOT23-5



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	1.050	1.250	0.041	0.049
A1	0.000	0.100	0.000	0.004
A2	1.050	1.150	0.041	0.045
b	0.300	0.400	0.012	0.016
c	0.100	0.200	0.004	0.008
D	2.820	3.020	0.111	0.119
E	1.500	1.700	0.059	0.067
E1	2.650	2.950	0.104	0.116
e	0.950TYP		0.037TYP	
e1	1.800	2.000	0.071	0.079
L	0.700REF		0.028REF	
L1	0.300	0.600	0.012	0.024
θ	0°	8°	0°	8°

# PACKAGE OUTLINE DIMENSIONS

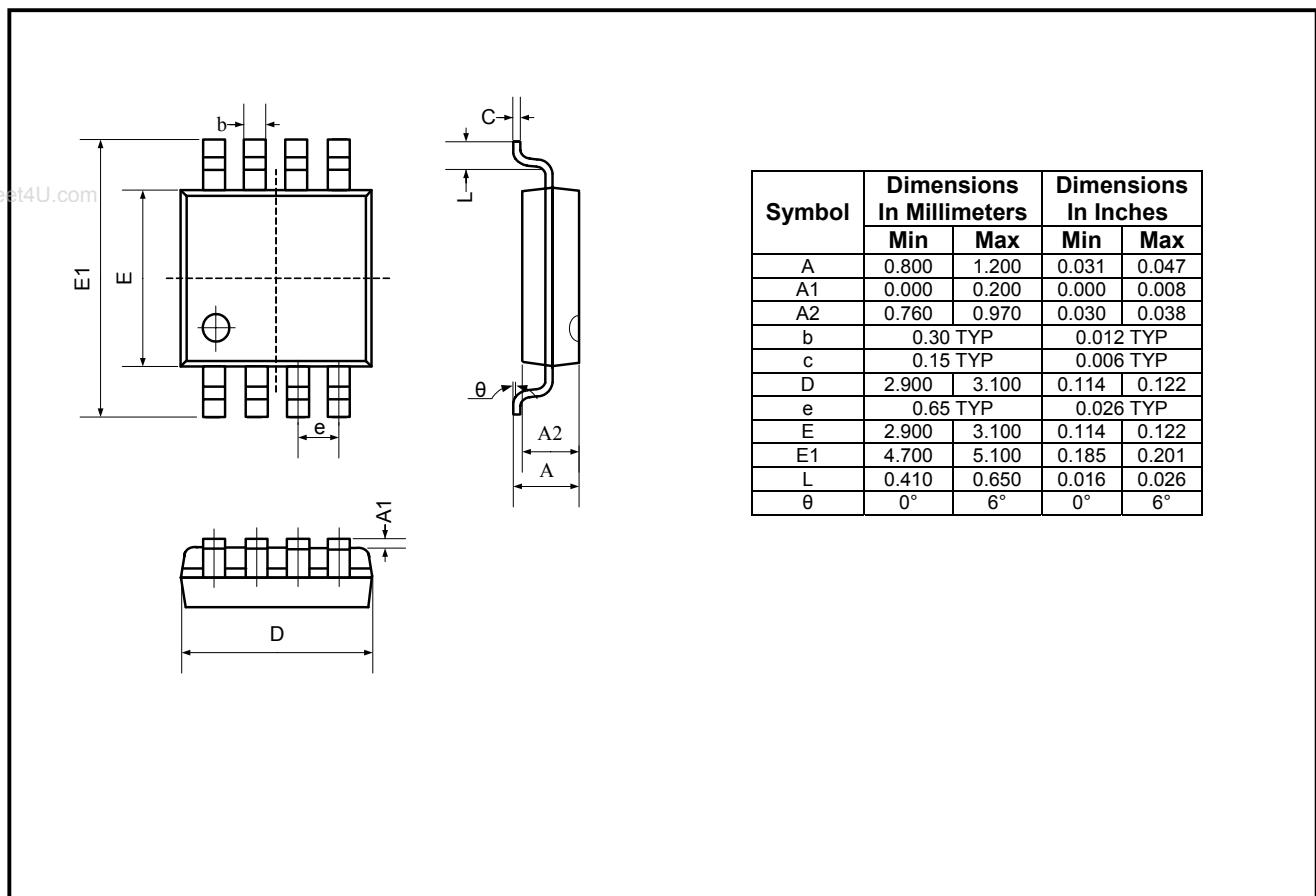
SO-8



Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	1.350	1.750	0.053	0.069
A1	0.100	0.250	0.004	0.010
A2	1.350	1.550	0.053	0.061
B	0.330	0.510	0.013	0.020
C	0.190	0.250	0.007	0.010
D	4.780	5.000	0.188	0.197
E	3.800	4.000	0.150	0.157
E1	5.800	6.300	0.228	0.248
e	1.270TYP		0.050TYP	
L	0.400	1.270	0.016	0.050
θ	0°	8°	0°	8°

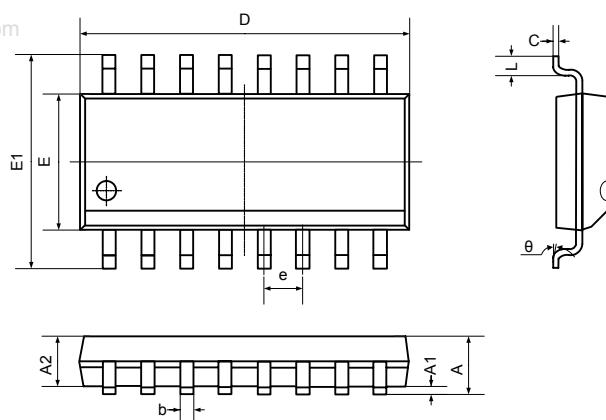
# PACKAGE OUTLINE DIMENSIONS

## MSOP-8



# PACKAGE OUTLINE DIMENSIONS

SO-16

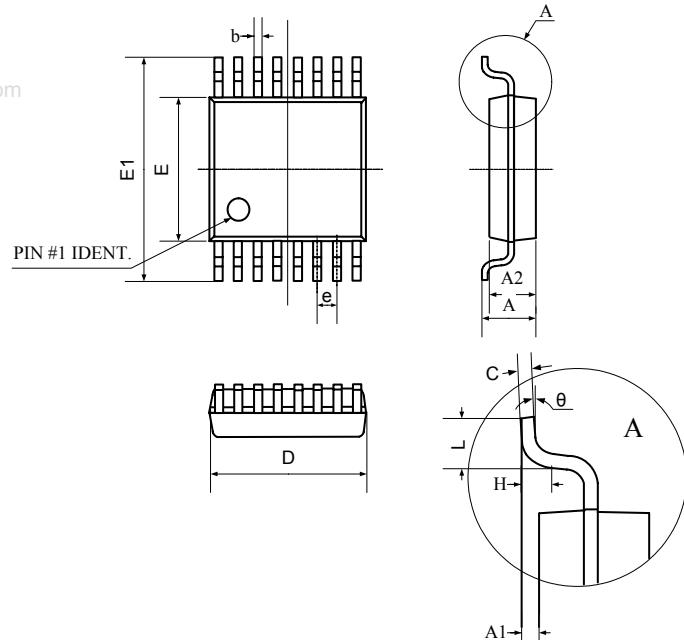


Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	1.350	1.750	0.053	0.069
A1	0.100	0.250	0.004	0.010
A2	1.350	1.550	0.053	0.061
b	0.330	0.510	0.013	0.020
c	0.170	0.250	0.007	0.010
D	9.800	10.20	0.386	0.402
E	3.800	4.000	0.150	0.157
E1	5.800	6.200	0.228	0.244
e	1.270 (BSC)		0.050 (BSC)	
L	0.400	1.270	0.016	0.050
θ	0°	8°	0°	8°

# PACKAGE OUTLINE DIMENSIONS

## TSSOP-16

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Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
D	4.900	5.100	0.193	0.201
E	4.300	4.500	0.169	0.177
b	0.190	0.300	0.007	0.012
c	0.090	0.200	0.004	0.008
E1	6.250	6.550	0.246	0.258
A			1.100	0.043
A2	0.800	1.000	0.031	0.039
A1	0.020	0.150	0.001	0.006
e	0.65 (BSC)		0.026 (BSC)	
L	0.500	0.700	0.020	0.028
H	0.25(TYP)		0.01(TYP)	
θ	1°	7°	1°	7°

## REVISION HISTORY

Location	Page
<b>11/06— Data Sheet changed from REV.A to REV.B</b> Changes to ABSOLUTE MAXIMUM ATINGS .....	3

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## Shengbang Microelectronics Co, Ltd

Unit 3, ChuangYe Plaza  
No.5, TaiHu Northern Street, YingBin Road Centralized Industrial Park  
Harbin Development Zone  
Harbin, HeiLongJiang 150078  
P.R. China  
Tel.: 86-451-84348461  
Fax: 86-451-84308461  
[www.sg-micro.com](http://www.sg-micro.com)