

# IR9331/IR9331N V/F Converter

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## ■ Description

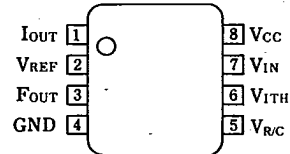
The IR9331/IR9331N is a voltage-to-frequency converters ideally suited for use in simple low-cost circuits for A/D conversion, precision F/V conversion, longterm intergration, linear frequency modulation or demodulation, and many other functions.

## ■ Features

1. Guaranteed linearity 0.01%FS (MAX.)
2. Excellent temperature stability  $\pm 30\text{ppm}/^\circ\text{C}$  (TYP.)
3. Wide dynamic range 100dB at 10kHz FS\* (MIN.)
4. Wide range of FS frequency 1~100kHz
5. Wide range of supply voltage 4~40V
6. 8-pin dual-in-line package (IR9331)  
8-pin small-outline package (IR9331N)

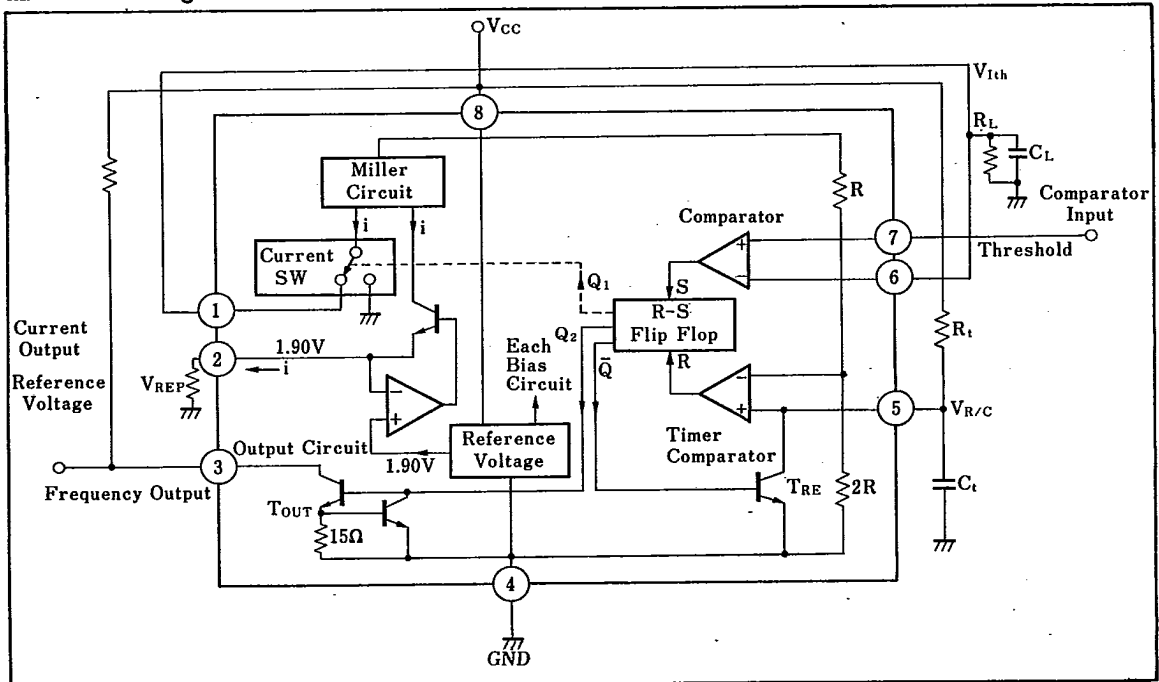
\* FS: Full Scale

## ■ Pin connections



Top View

## ■ Block Diagram



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## Absolute Maximum Ratings

(Ta=25°C)

Symbol	Condition	Rating	Unit		
Supply voltage	$V_{CC}$	40	V		
Input voltage	$V_{R/C}$	-0.2~ $V_{CC}$	V		
	$V_{Ith}$				
	$V_{IN}$				
Output short-circuit time	$t_{SG}$ to GND	Infinity	s		
	$t_{SV}$ to $V_{CC}$ , short-circuit 30mA(TYP.)	Infinity			
Power dissipation	$P_D$	Ta≤25°C	500	mW	
		IR9331N	450		
$P_D$ derating ratio	$\Delta P_D/^\circ C$	Ta>25°C	IR9331N	4.5	mW/°C
Operating temperature	$T_{opr}$		-10~+70	°C	
Storage temperature	$T_{stg}$		-55~+150	°C	

## Electrical Characteristics

( $V_{CC}=15V$ , Ta=25°C, Test circuit 1)

Parameter	Symbol	Condition	MIN.	TYP.	MAX.	Unit
VFC non linearity error *2	$NL_b$	$4.5V \leq V_{CC} \leq 20V$		±0.003	±0.01	%FS
	$NL_b'$	$T_{opr}(-10 \sim 70^\circ C)$		±0.006	±0.02	
	$NL_a$	$V_{CC}=15V$ , $f_{OUT}$ Test circuit 2*1		±0.10	±0.30	
Scale factor (gain)	SF	$V_{IN} = -10V$ , $R_S = 14k\Omega$	0.90	1.00	1.10	kHz/V
Gain temperature coefficient	$\alpha SF$	$4.5V \leq V_{CC} \leq 20V$ , $T_{opr}(-10 \sim 70^\circ C)$		±30		ppm/°C
Gain-power supply stability	SVR	$4.5V \leq V_{CC} \leq 10V$		0.01	0.15	%V
	SVR'	$10V \leq V_{CC} \leq 40V$		0.006	0.06	
Full scale frequency	$F_{FS}$	$V_{IN} = -10V$	10.0			kHz
Over range frequency	$F_{over}$	$V_{IN} = -11V$	10			%

### Input comparator (terminal 6 and 7)

Offset voltage	$V_{IO1}$			±3	±10	mV
	$V_{IO2}$	$T_{opr}(-10 \sim 70^\circ C)$		±4	±14	
Bias current	$I_B$			-80	-300	nA
Offset current	$I_{IO}$			±8	±100	nA
In-phase input range	$V_{ICM}$	$T_{opr}(-10 \sim 70^\circ C)$	-0.2		$V_{CC}-2.0$	V

### Timer (terminal 5)

Timer threshold voltage	$V_{th}$		0.63	0.667	0.70	( $\times V_{CC}$ )V
Input bias current	$I_{15}$	$V_{CC}=15V$ , $0V \leq V_5 \leq 9.9V$		±10		nA
	$I_{15}'$	$V_{CC}=15V$ , $V_5=10V$		200	1,000	
Saturation voltage (reset)	$V_{SAT5}$	$I=5mA$		0.22	0.5	V

### Power supply source (terminal 1)

Output current	$I_{OUT}$	$R_S=14k\Omega$ , $V_1=0V$	116	136	156	μA
$I_{OUT}$ -Voltage fluctuation	$I_{OV}$	$0V \leq V_1 \leq 10V$		0.7	1.5	μA
OFF-state leakage current	$I_{OFF}$			0.02	10.0	nA
	$I_{OFF}'$	Ta=70°C		2.0	50.0	
Operating current range	$I_{opr}$			10~500		μA

### Reference voltage (terminal 2)

Reference voltage	$V_{REF}$		1.70	1.89	2.08	$V_{DC}$
Temperature coefficient	$\alpha V_{REF}$			±60		ppm/°C
Time drift	$\alpha V_{REF}$	1,000 hours		±0.1		%

### Logic output (Terminal 3)

Saturation voltage	$V_{SAT3}$	$I=5mA$		0.15	0.50	V
	$V_{SAT3}'$	$I=3mA$		0.10	0.40	
OFF-state leakage current	$I_{OFF3}$			0.05	1.0	μA

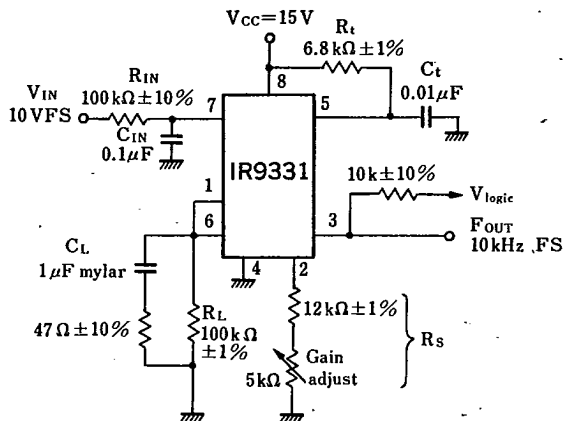
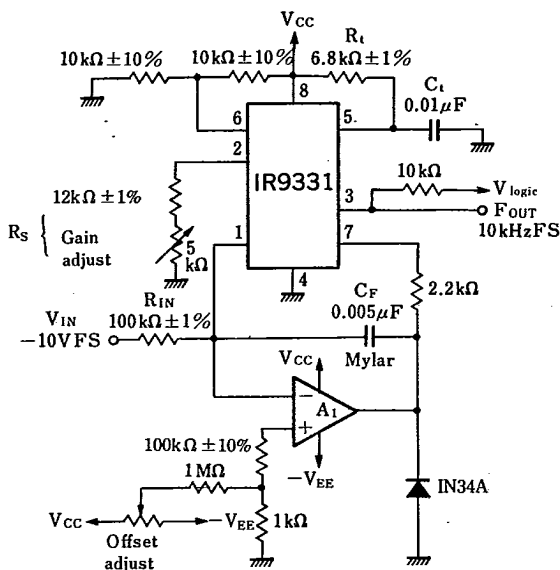
Parameter	Symbol	Condition	MIN.	TYP.	MAX.	Unit
<b>Supply current (terminal 8)</b>						
Supply current	$I_{CC}$	$V_{CC}=5V$	1.5	3.0	6.0	mA
	$I_{CC}'$	$V_{CC}=40V$	2.0	4.0	8.0	

\*1  $f_{OUT}=10\text{Hz}\sim 11\text{kHz}$ , this test alone is to be performed on test circuit 2.

\*2 Non-linearity error is defined as the deviation from  $V_{IN} \times (10\text{kHz}/-10V_{DC})$  at  $f_{OUT}=1\text{Hz}\sim 11\text{kHz}$ .  
(Full scale adjustment at 10kHz, zero adjustment at 10kHz)

## Test Circuit

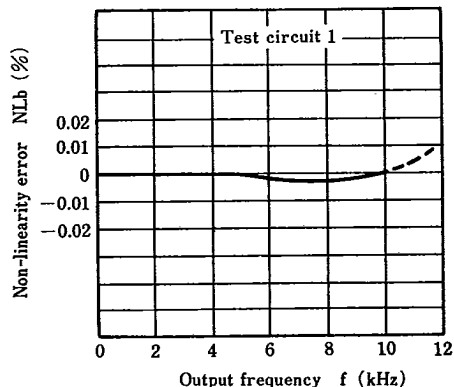
- (1) Test circuit (Precision V/F conversion circuit)      (2) Test circuit 2 (Simple V/F conversion circuit)



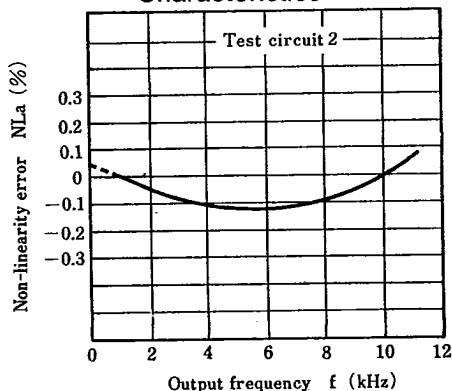
$A_1$ : Use an operational amplifier that satisfies the following conditions:  
Input offset voltage below 1mV  
Input offset current below 2nA

## Electrical Characteristics Curves (Unless otherwise specified, $V_{CC}=15V$ , $T_a=25^\circ\text{C}$ )

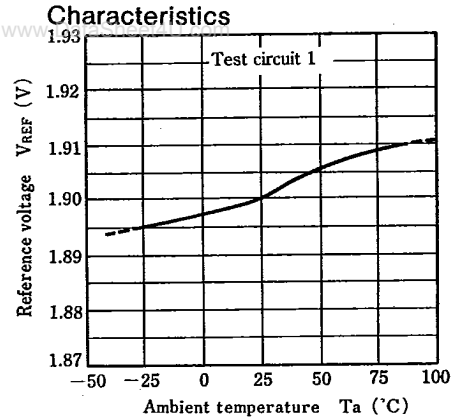
Non-linearity error—Output frequency Characteristics



Non-linearity error—Output frequency Characteristics



## Reference voltage—Ambient temperature



## Description of Operation

The IR9331 is organized mainly as an input comparator, R-S flip-flop, timer comparator, current supply, current switch 1.9V reference voltage supply and output circuit. To briefly explain the circuit operation, the feed-back of this circuit is organized in supply, current switch 1.9V reference voltage supply and output circuit. To briefly explain the circuit  $V_{IN}$  is higher,  $C_L$  will be discharged through  $R_L$  in a relatively short time to settle for a lower frequency. That is to say that it operates as a highly accurate loose coupling oscillator that produces frequencies linearly in proportion to the input voltage.

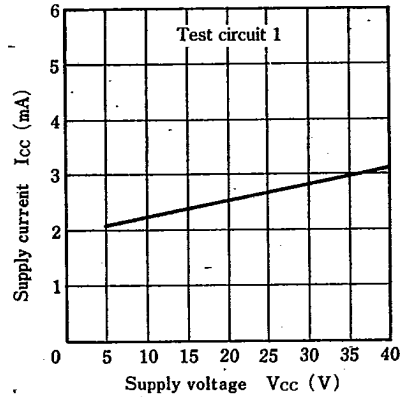
Following is a detailed description.

Suppose that the voltage  $V_{Ith}$  (terminal 6) becomes as satisfies  $V_{Ith} < V_{IN}$ . The input comparator compares  $V_{Ith}$  and  $V_{IN}$  to set the R-S flip-flop. The  $Q_1$  output of F.F closes the current switch and starts charging  $C_L$  with the current  $i$ . At the same time the  $Q_2$  output turns on the frequency output transistor ( $T_{OUT}$ ) while the  $\bar{Q}$  turns off the reset transistor ( $T_{RE}$ ). From this moment on  $C_T$  will continue to get charged logarithmically toward  $V_{CC}$ . When the voltage of  $C_T$  has come up to  $2/3 V_{CC}$ , the timer comparator applies reset output to F.F. The time taken so far is about  $1.1R_tC_t$  ( $1.1 = \ln 0.333...$ )

Even if the timer comparator generate reset output, the F.F will remain set so long as  $V_{Ith} \leq V_{IN}$ , in which it will continue being charged well beyond  $2/3 V_{CC}$  until it gets to the state where  $V_{Ith} > V_{IN}$ . This condition arises on power-up or when an excessively higher signal gets in to have the output frequency 0. It will, however, go back to normal if  $V_{IN}$  restores within the operating range.

F.F will not be reset until the reset output is produced and a condition is reached as satisfies  $V_{Ith}$

## Supply current—Supply voltage



$V_{IN}$ . The current switch opens to have  $C_L$  start discharging (until it reaches a point where  $V_{Ith} > V_{IN}$ ). Simultaneously with the resetting of F.F,  $T_{RE}$  turns on to have  $C_T$  discharge itself. Also  $T_{OUT}$  turns off. The number of the repetition of this cycle above over and over again in a second is the frequency as defined.

How to work out the output frequency

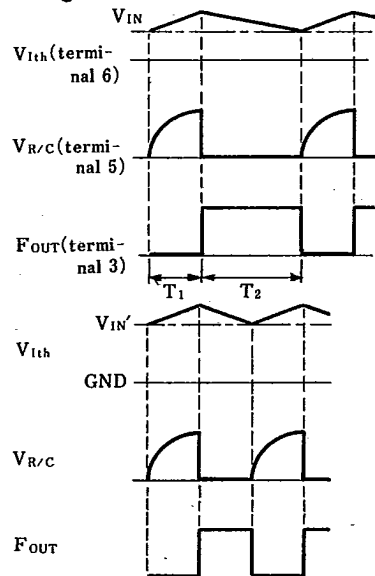
$$f_{OUT} = \frac{1}{T_1 + T_2}, \quad i = V_{REF}/R_S$$

$$T_1 = -R_t C_t \ln(1/3) \approx 1.1 R_t C_t \quad \text{Charging time for } C_L$$

$$T_2 = \frac{(i - V_{IN}/R_L) R_L}{V_{IN}} T_1 \quad \text{Charging time for } C_T$$

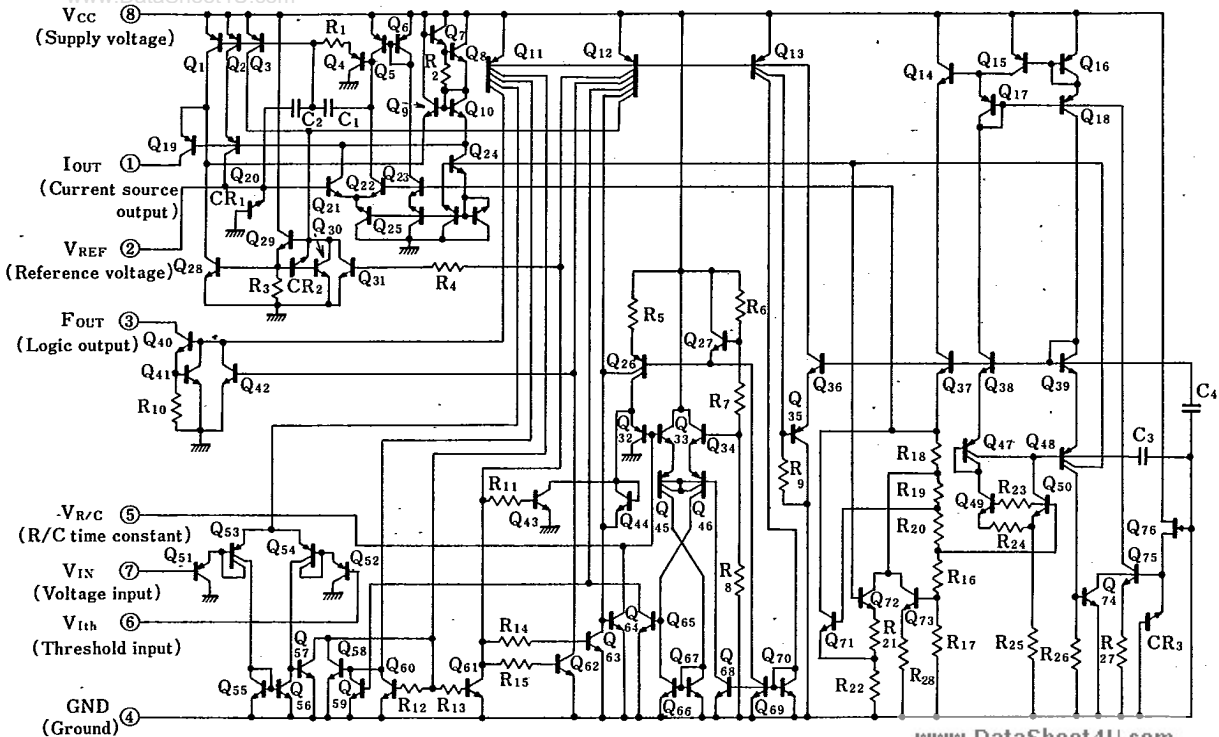
$$f_{OUT} = \frac{V_{IN}}{i R_L T_1} = \frac{V_{IN}}{V_{REF}} \cdot \frac{R_S}{R_L} \cdot \frac{1}{1.1 R_t C_t}$$

## Timing Chart



# ■ Equivalent Circuit

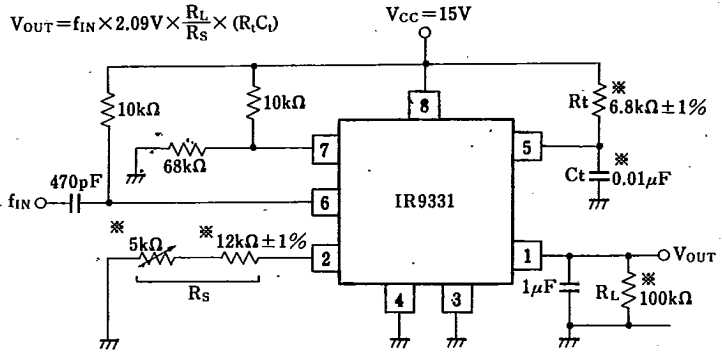
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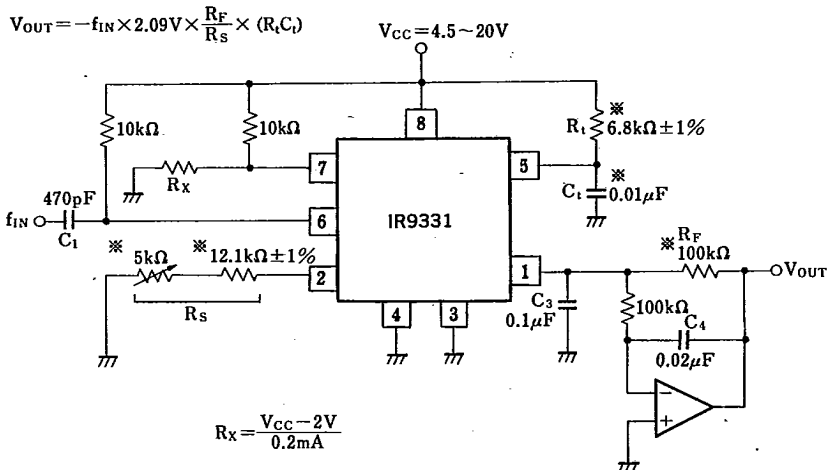
# Application Circuit Example

## (1) Simple F/V conversion



Full-scale : 10kHz  
Non-linearity : ±0.06%

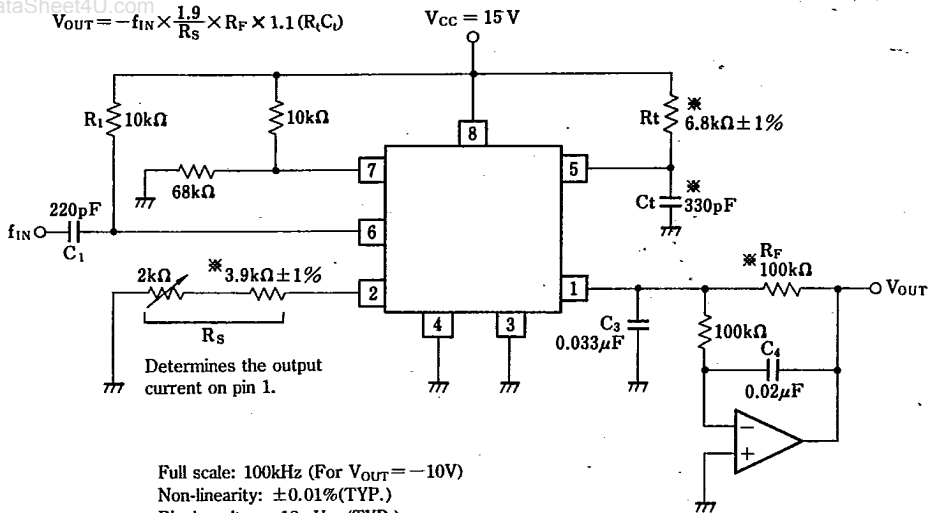
## (2) High grade F/V conversion



Full-scale : 10kHz  
Non-linearity : ±0.01%

\* Use resistors with reduced coefficient of temperature.

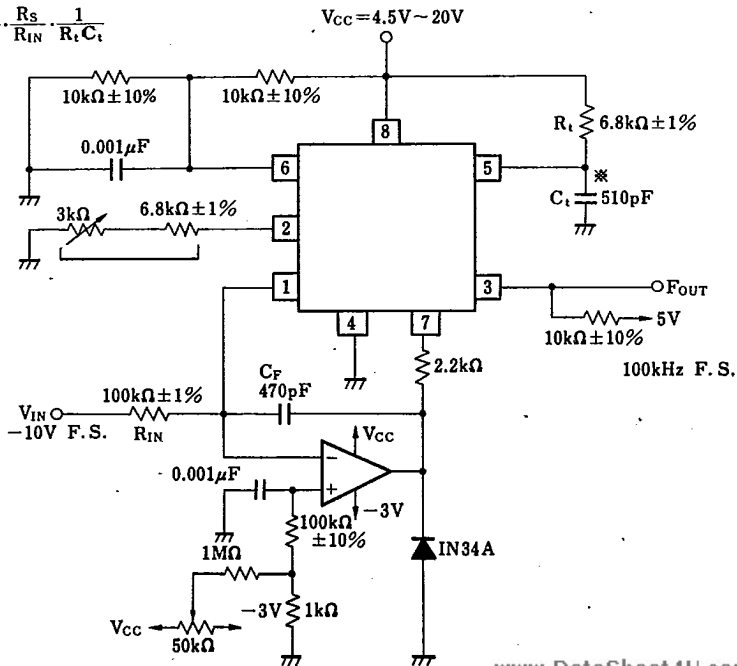
### (3) High grade F/V conversion



※ Use resistors with reduced coefficient of temperature.

### (4) High grade F/V conversion (100kHz full-scale)

$$f_{OUT} = \frac{-V_{IN} \cdot R_S}{2.09V \cdot R_{IN} \cdot R_t \cdot C_t}$$



※ Use resistors with reduced coefficient of temperature.