



Micro Power Systems

# MP7643

4-Channel High-Speed Non-Inverting  
15 MHz Input Bandwidth 8-Bit  
DACs with Output Buffers and  
Parallel Digital Data Port

## FEATURES

- 4 Independent 2-Quadrant Multiplying 8-Bit DACs with Output Amplifiers
- Dual Positive (+10 V and +5 V) Supplies or Dual (+5 V) Supplies Capability
- High Speed:
  - 12.5 MHz Digital Clock Rate
  - $V_{REF}$  to  $V_{OUT}$  Settling Time: 150ns to 8-bit (typ)
  - Voltage Reference Input Bandwidth: 15 MHz
- Very Low Noise Gain Control
- Low Power: 80mW
- Low AC Voltage Reference Feedthrough
- Excellent Channel-to-Channel Isolation
- $DNL = \pm 0.5$  LSB,  $INL = \pm 1$  LSB (typ)

- DACs Matched to  $\pm 0.5\%$  (typ)
- Low Harmonic Distortion: 0.25% typical with  $V_{REF} = 1$  V p-p @ 1 MHz
- TTL/CMOS Compatible
- Latch-Up Free
- ESD Protection: 2000 V Minimum
- PDIP, SOIC & PLCC Packages Available

## APPLICATIONS

- Direct High-Frequency Automatic Gain Control
- Video AGC & CCD Level AGC
- Convergence Adjustment for High-Resolution Monitors (Workstations)
- Multiplier Replacement

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## GENERAL DESCRIPTION

The MP7643 is ideal for digital gain control of high frequency analog signals such as video, composite video and CCD. The device includes 4-channels of high speed, high bandwidth, two quadrant multiplying, 8-bit accurate digital-to-analog converter. It includes an output drive buffer per channel capable of driving a  $\pm 1$ mA (typ) to a load.  $DNL$  of better than  $\pm 0.5$  LSB is achieved with a channel-to-channel matching of typically 0.5%. Stability, matching, and precision of the DACs are achieved by using MPS' thin film technology. Excellent channel-to-channel isolation is also achieved with MPS' BiCMOS process which cannot be achieved using a typical CMOS technology.

An open loop architecture (patent pending) provides wide small signal bandwidth from  $V_{REF}$  to output up to 15 MHz (typ), fast output settling time of 150 ns, and excellent  $V_{REF}$

feedthrough isolation. In addition, low distortion in the order of 0.25% with a 1 V p-p, 1 MHz signal is achieved.

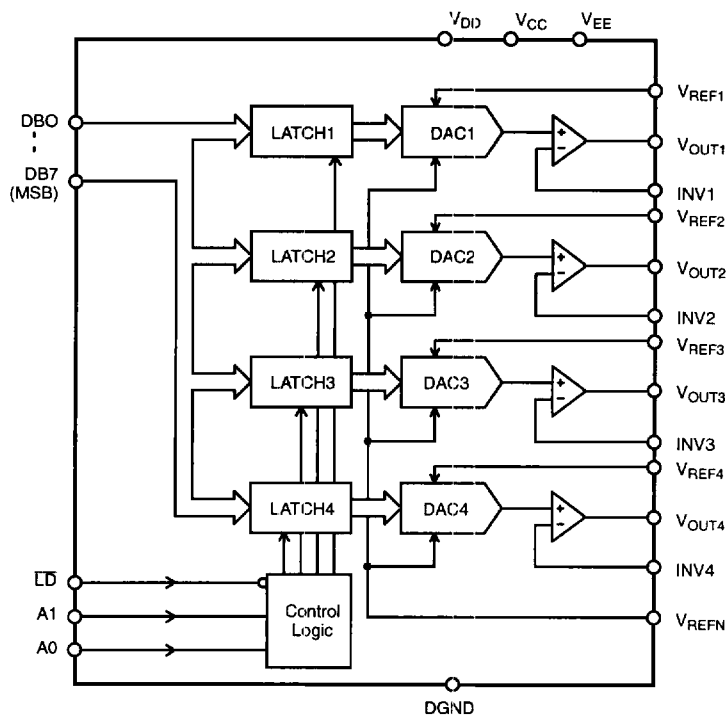
The combination of a constant input Z and the ability to vary  $V_{REF}$  within  $V_{CC} - 1.8$  V to  $V_{EE} + 1.5$  V allows flexibility for optimum system design.

The MP7643 is fabricated on a junction isolated, high speed BiCMOS (BiCMOS IV<sup>TM</sup>) process with thin film resistors. This process enables precision high speed analog/digital (mixed-mode) circuits to be fabricated on the same chip.

This device is latch-up free, and all inputs are protected to a minimum of 2000 V for ESD protection per MIL-STD-883.

Specified for operation over the commercial / industrial ( $-40$  to  $+85^{\circ}\text{C}$ ) temperature range, the MP7643 is available in Plastic dual-in-line (PDIP), Surface Mount (SOIC), and Plastic leaded chip carrier (PLCC) packages.

## SIMPLIFIED BLOCK DIAGRAM

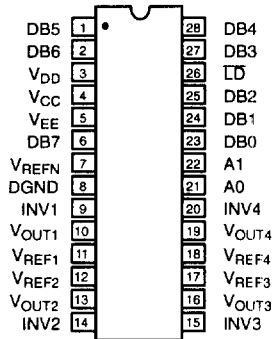


## ORDERING INFORMATION

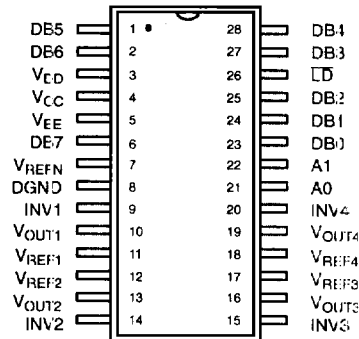
Package Type	Temperature Range	Part No.	INL (LSB)	DNL (LSB)	Gain Error (% FSF)
SOIC	-40 to +85°C	MP7643AS	±1	±0.5	±1.5
Plastic Dip	-40 to +85°C	MP7643AN	±1	±0.5	±1.5
PLCC	-40 to +85°C	MP7643AP	±1	±0.5	±1.5



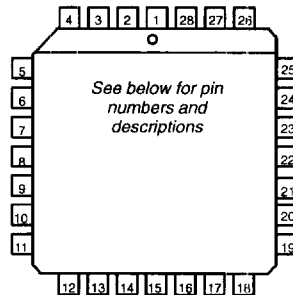
**PIN CONFIGURATIONS**



**28 Pin PDIP (0.300")  
NN28**



**28 Pin SOIC (EIAJ, 0.335")  
R28**



**28 Pin PLCC  
P28**

**PIN OUT DEFINITIONS**

PIN NO.	NAME	DESCRIPTION
1	DB5	Data Input Bit 5
2	DB6	Data Input Bit 6
3	V <sub>DD</sub>	Digital Positive Supply
4	V <sub>CC</sub>	Analog Positive Supply
5	V <sub>EE</sub>	Analog Negative Supply
6	DB7	Data Input Bit 7
7	V <sub>REFN</sub>	Negative Reference Input
8	DGND	Digital Ground
9	INV1	Inverting Input 1
10	V <sub>OUT1</sub>	DAC 1 Output
11	V <sub>REF1</sub>	DAC 1 Positive Reference input
12	V <sub>REF2</sub>	DAC 2 Positive Reference Input
13	V <sub>OUT2</sub>	DAC 2 Output
14	INV2	Inverting Input 2

PIN NO.	NAME	DESCRIPTION
15	INV3	Inverting Input 3
16	V <sub>OUT3</sub>	DAC 3 Output
17	V <sub>REF3</sub>	DAC 3 Positive Reference Input
18	V <sub>REF4</sub>	DAC 4 Positive Reference Input
19	V <sub>OUT4</sub>	DAC 4 Output
20	INV4	Inverting Input 4
21	A0	DAC Address Bit 0
22	A1	DAC Address Bit 1
23	DB0	Data Input Bit 0
24	DB1	Data Input Bit 1
25	DB2	Data Input Bit 2
26	LD	Load Data to Selected DAC
27	DB3	Data Input Bit 3
28	DB4	Data Input Bit 4

## ELECTRICAL CHARACTERISTICS TABLE FOR DUAL SUPPLIES

Unless Otherwise Noted:  $V_{DD} = 5\text{ V}$ ,  $V_{CC} = +5\text{ V}$ ,  $V_{EE} = -5\text{ V}$ ,  $V_{REF} = 3\text{ V}$  and  $-3\text{ V}$ ,  $T = 25^\circ\text{C}$ ,  
Output Load = No Resistive Load,  $V_{REFN} = DGND = 0\text{ V}$

Parameter	Symbol	25°C			Tmin to Tmax		Units	Test Conditions/Comments
		Min	Typ	Max	Min	Max		
<b>DC CHARACTERISTICS</b>								
Resolution (All Grades)	N	8					Bits	FSR = Full Scale Range (1)
Differential Non-Linearity	DNL		±0.5				LSB	
Integral Non-Linearity	INL		±1				LSB	
Monotonicity			Guaranteed					
Gain Error	GE		±1.5				% FSR	
Zero Scale Offset	Z <sub>OFS</sub>		±20				mV	
Output Drive Capability	I <sub>O</sub>		±1				mA	
<b>REFERENCE/INV INPUTS</b>								
Impedance of V <sub>REF</sub>	REF	4	6	8			kΩ	V <sub>REF</sub> Max Swing is V <sub>REFN</sub> ±3 V
Voltage Range	V <sub>RR</sub> Pos.		V <sub>CC</sub> -1.8				V	
	V <sub>RR</sub> Neg.		V <sub>EE</sub> +1.5				V	
INV DC Voltage Range	V <sub>RR</sub> Pos.		V <sub>O</sub>				V	
	V <sub>RR</sub> Neg.		V <sub>EE</sub> +1				V	
<b>DYNAMIC CHARACTERISTICS (2)</b>								
Input to Output Bandwidth			15				MHz	R <sub>L</sub> = 5k, C <sub>L</sub> = 20 pF V <sub>R</sub> = 1.6 V p-p, R <sub>L</sub> = 5k to V <sub>EE</sub> V <sub>R</sub> = 1.6 V p-p, R <sub>L</sub> = 5k to V <sub>EE</sub> V <sub>OUT</sub> =50mV > f <sub>c</sub> above code 16
Input to Output Settling Time (5)			150				ns	
Small Signal Voltage Reference	f <sub>r</sub>		15				MHz	V <sub>OUT</sub> =50mV > f <sub>c</sub> for all codes
Input to Output Bandwidth							MHz	
Small Signal Voltage Reference	f <sub>r</sub>		15				MHz	V <sub>OUT</sub> =50mV > f <sub>c</sub> for all codes
Input to Output Bandwidth							MHz	
Voltage Settling from V <sub>REF</sub> to V <sub>DAC</sub> Out	t <sub>sr</sub>		300				ns	V <sub>R</sub> =0 to V <sub>R</sub> =3V Step (6) to 1 LSB
Voltage Settling from Digital Code to V <sub>DAC</sub> Out	t <sub>sd</sub>		300				ns	ZS to FS to 1 LSB
V <sub>REF</sub> Feedthrough	F <sub>DT</sub>		TBD				dB	Codes=0 @ 1 MHz
Group Delay	GD		TBD				ns	
Harmonic Distortion	T <sub>HD</sub>		TBD				%	V <sub>REF</sub> =1MHz Sine 3V p-p
Channel-to-Channel Crosstalk	C <sub>T</sub>		TBD				dB	@ 1 MHz, single channel
Digital Feedthrough	Q		TBD				nVS	CLK to V <sub>OUT</sub>
Power Supply Rejection Ratio	PSRR		TBD				%/%	ΔV=±5%
<b>POWER CONSUMPTION</b>								
Positive Supply Current	I <sub>CC</sub>		8				mA	V <sub>REF</sub> = 0 V
Negative Supply Current	I <sub>EE</sub>		8				mA	V <sub>REF</sub> = 0 V
Power Dissipation	P <sub>DISS</sub>		80				mW	V <sub>REF</sub> = 0 V, Codes = all 1
<b>DIGITAL INPUT CHARACTERISTICS</b>								
Logic High (3)	V <sub>IH</sub>	2.4					V	
Logic Low (3)	V <sub>IL</sub>			0.8			V	
Input Current	I <sub>L</sub>			±10			μA	
Input Capacitance (2)	C <sub>L</sub>		8				pF	



## ELECTRICAL CHARACTERISTICS TABLE

Description	Symbol	25°C			Tmin to Tmax		Units	Conditions
		Min	Typ	Max	Min	Max		
<b>DIGITAL TIMING SPECIFICATIONS (2, 4)</b>								
Address to $\overline{\text{LD}}$ Setup	$t_{AS}$		70				ns	
Address to $\overline{\text{LD}}$ Hold	$t_{AH}$		0				ns	
Data to $\overline{\text{LD}}$ Setup	$t_{DS}$		70				ns	
Data to $\overline{\text{LD}}$ Hold	$t_{DH}$		0				ns	
$\overline{\text{LD}}$ Pulse Width	$t_{LD}$		70				ns	
PRESET Pulse Width	$t_{PR}$		60				ns	

## NOTES

- (1) Full Scale Range (FSR) is 3V.
- (2) Guaranteed but not production tested.
- (3) Digital input levels should not go below ground or exceed the positive supply voltage, otherwise damage may occur.
- (4) See Figure 1.
- (5) For reference input pulse:  $t_R = t_F \geq 100$  ns.

Specifications are subject to change without notice

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## ABSOLUTE MAXIMUM RATINGS (1, 2) (TA = +25°C unless otherwise noted)

$V_{CC}$ to $V_{REFN}$ .....	+6.5 V	Operating Temperature Range	
$V_{EE}$ to $V_{REFN}$ .....	-6.5 V	Extended Industrial .....	-40°C to +85°C
$V_{CC}$ to DGND .....	+13.0 V	Maximum Junction Temperature .....	-65°C to 150°C
$V_{EE}$ to DGND .....	-6.5 V	Storage Temperature .....	150°C
$V_{REF}$ 1-4 to DGND, $V_{REFN}$ .....	$V_{CC}$ to $V_{EE}$	Lead Temperature (Soldering, 10 sec) .....	+300°C
$V_{OUT}$ 1-4 to DGND, $V_{REFN}$ .....	$V_{CC}$ to $V_{EE}$	Package Power Dissipation Rating @ 75°C	
Digital Input & Output Voltage to DGND	-0.5 to $V_{DD}$ +0.5 V	PDIP, SOIC, PLCC .....	1050mW
		Derates above 75°C .....	14mW/°C

## NOTES:

- (1) Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation at or above this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.
- (2) Any input pin which can see a value outside the absolute maximum ratings should be protected by Schottky diode clamps (HP5082-2835) from input pin to the supplies. All inputs have protection diodes which will protect the device from short transients outside the supplies of less than 100mA for less than 100 $\mu$ s.

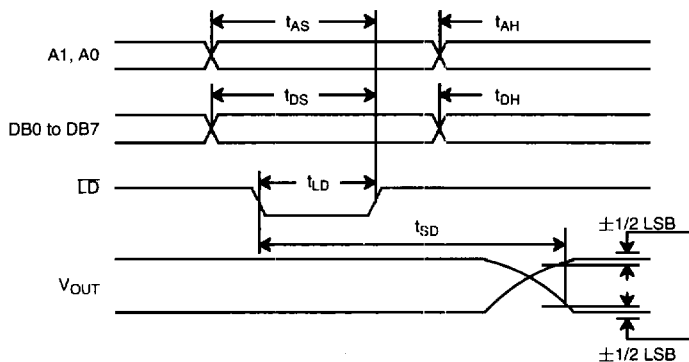


Figure 1. Timing Diagram

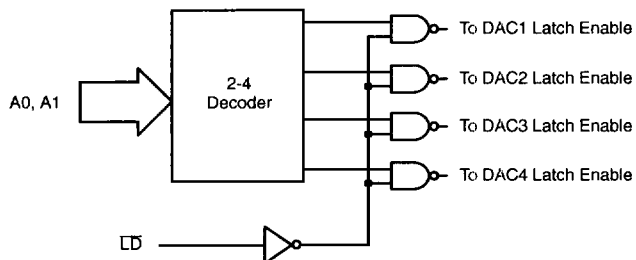


Figure 2. Input Control Logic (Simplified) Block Diagram

LD	A1	A0	Operation
L	L	L	DAC1 Transparent
↑	L	L	DAC1 Latched
L	L	H	DAC2 Transparent
↑	L	H	DAC2 Latched
L	H	L	DAC3 Transparent
↑	H	L	DAC3 Latched
L	H	H	DAC4 Transparent
↑	H	H	DAC4 Latched
H	X	X	No Operation

Table 1. Truth Table

D7 MSB	D6	D5	D4	D3	D2	D1	D0 LSB	DAC Output Voltage $V_{OI} = V_{REFN} + (V_{RI} - AGND) \left( \frac{D}{256} \right)$
0	0	0	0	0	0	0	0	$V_{REFN}$
0	0	0	0	0	0	0	1	$(V_{RI} - V_{REFN}) \left( \frac{1}{256} \right) + V_{REFN}$
⋮	⋮	⋮	⋮	⋮	⋮	⋮	⋮	⋮
1	1	1	1	1	1	1	0	$(V_{RI} - V_{REFN}) \left( \frac{254}{256} \right) + V_{REFN}$
1	1	1	1	1	1	1	1	$(V_{RI} - V_{REFN}) \left( \frac{255}{256} \right) + V_{REFN}$

Table 2. DAC Transfer Function  
Analog Output vs. Digital Code (With  $V_{REF}$  Shorted to INV)



## THEORY OF OPERATION

The MP7643 is a 4-channel multiplying D/A converter that incorporates a novel open loop architecture invented by MPS. The design produces the wider bandwidth, faster settling time, more constant group delay, and a lower noise operation compared to the conventional R-2R based architectures. This device is particularly useful in applications where multipliers are used to perform the gain adjustment function for high frequency analog signal conditioning. Multipliers produce higher noise and offset for low gain operation. This design allows for digital control of gain with constant and very low noise from the low gain through high gain ranges of operation.

### Linearity Characteristics

Each DAC achieves  $DNL \leq \pm 0.5$  LSB (typ),  $INL \leq \pm 1$  LSB (typ), and gain error  $\leq \pm 1.5\%$ . Since all 4 channel D/A converters are fabricated on the same IC, the linearity matching and gain matching of  $\pm 0.5\%$  (typ) is achieved.

### AC and Transient Settling Characteristics

The novel subranging architecture delivers a 15 MHz (typ.)  $-3$  dB bandwidth. With all codes = 1 and a 1.6 V step impulse at  $V_{REF}(1-4)$ , the analog output settles to 8 bits of accuracy in typically 150 ns (with  $R_L = 5k$  to  $V_{EE}$ ). Also with  $V_{REF} = 3$  V or  $-3$  V and a FS to ZS or ZS to FS code change, the respective analog output settles to 8 bits typically in 300 ns. Note that the AC performance specifications also match to within  $\pm 0.5\%$  (typ.) between all 4 channels. The above AC and transient performance is achieved with each channel consuming only 20 mW (typ.) with either  $\pm 5$  V or 0 V to 10 V supplies.

### Digital Interface

The MP7643 allows direct interface to most microprocessor buses without additional I/O circuitry. *Figure 1.* and *Figure 2.* describe the operation, specification and interface characteristics of the logic port.

The address bits A0 and A1 determine which D/A channel is selected. When  $\overline{LD}$  input is low the respective latch of the D/A is enabled (digital input data becomes transparent to the latch and the selected DAC channel), and digital data is loaded into the selected DAC.

### Power Supplies and Voltage Reference DC Voltage Ranges

For the single supply operation,  $V_{CC} = +10$  V,  $V_{DD} = +5$  V, and  $V_{EE} = GND = 0$  V. The  $V_{OUT}$  1-4 and  $V_{REF}$  1-4 range would be  $V_{CC} - 1.3$  V ( $10 - 1.8 = 8.2$  V) to  $V_{EE} + 1.5$  V ( $0 + 1.5 = 1.5$  V).  $V_{REFN}$  is the equivalent of AGND for this DAC. In this mode  $V_{REFN}$  can be set at  $(V_{CC} + V_{EE})/2 = (10 + 0)/2 = 5$  V.  $V_{REFN}$  DC range can, however, be set from  $V_{EE} + 1.5 = 1.5$  V to  $V_{CC} - 1.5 = 8.2$  V. Refer to *Table 2.* for the relationship equations.

For the dual supply operation,  $V_{CC} = +5$ ,  $V_{DD} = +5$ , and  $V_{EE} = -5$  V. The  $V_{OUT}$  1-4 and  $V_{REF}$  1-4 range would be  $V_{CC} - 1.8$  V ( $5 - 1.8 = 3.2$  V) to  $V_{EE} + 1.5$  V ( $-5 + 1.5 = -3.5$  V). In this mode  $V_{REFN}$  can be set to  $(V_{CC} + V_{EE})/2 = (5 - 5)/2 = 0$  V. However,  $V_{REFN}$  DC range can be set from  $V_{EE} + 1.5 = -3.5$  V to  $V_{CC} - 1.8 = +3.2$  V. Refer to *Table 2.* for the relationship equations.

### About the INV Input and its DC Voltage Range

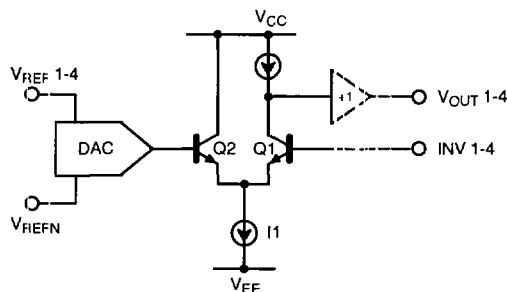


Figure 3. Simplified Block Diagram

As noted in the specification table, the max DC value of the INV input pin is  $V_O$ . *Figure 3.* shows a simplified block diagram of the internal circuitry around INV. If  $V_{INV}$  exceeds  $V_O$ , Q1 will saturate and the amp and consequently the DAC becomes non-functional.

The min DC range of INV is limited to  $V_{be}(Q1)$  and  $V_{CE}(sat)$  of  $I_1$ . Therefore,  $INV(\text{min-DC}) = V_{EE} + 1$  V.