

## N-CHANNEL 24V - 0.0052 Ω - 60A DPAK/IPAK STripFET™ III POWER MOSFET

TYPE	V <sub>DSS</sub>	R <sub>DS(on)</sub>	ID
STD90NH02L	24 V	< 0.006 Ω	60 A(2)
	)		\ <i>1</i>

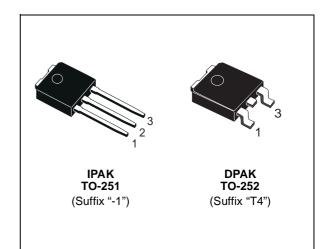
- TYPICAL  $R_{DS}(on) = 0.0052 \Omega @ 10 V$
- TYPICAL R<sub>DS</sub>(on) = 0.007 Ω @ 5 V
- R<sub>DS(ON)</sub> \* Qg INDUSTRY's BENCHMARK
- CONDUCTION LOSSES REDUCED
- SWITCHING LOSSES REDUCED
- LOW THRESHOLD DEVICE
- THROUGH-HOLE IPAK (TO-251) POWER PACKAGE IN TUBE (SUFFIX "-1")
- SURFACE-MOUNTING DPAK (TO-252) POWER PACKAGE IN TAPE & REEL (SUFFIX "T4")

#### DESCRIPTION

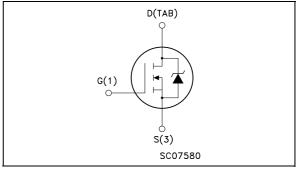
The **STD90NH02L** utilizes the latest advanced design rules of ST's proprietary STripFET<sup>TM</sup> technology. This is suitable fot the most demanding DC-DC converter application where high efficiency is to be achieved.

#### APPLICATIONS

 SPECIFICALLY DESIGNED AND OPTIMISED FOR HIGH EFFICIENCY DC/DC CONVERTES



#### INTERNAL SCHEMATIC DIAGRAM



#### **Ordering Information**

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SALES TYPE	MARKING	PACKAGE	PACKAGING
STD90NH02LT4	D90NH02L	TO-252	TAPE & REEL
STD90NH02L-1	D90NH02L	TO-251	TUBE

#### **ABSOLUTE MAXIMUM RATINGS**

Symbol	Parameter	Value	Unit	
V <sub>spike(1)</sub>	Drain-source Voltage Rating	30	V	
V <sub>DS</sub>	Drain-source Voltage (V <sub>GS</sub> = 0)	24	V	
V <sub>DGR</sub>	Drain-gate Voltage ( $R_{GS} = 20 \text{ k}\Omega$ )	24	V	
V <sub>GS</sub>	Gate- source Voltage	± 20	V	
I <sub>D</sub> (2)	Drain Current (continuous) at T <sub>C</sub> = 25°C	60	A	
I <sub>D</sub> (2)	Drain Current (continuous) at T <sub>C</sub> = 100°C	60	A	
I <sub>DM</sub> (3)	Drain Current (pulsed)	240	A	
P <sub>tot</sub>	Total Dissipation at $T_C = 25^{\circ}C$	95	W	
	Derating Factor	0.63	W/°C	
E <sub>AS</sub> <sup>(4)</sup>	Single Pulse Avalanche Energy	600	mJ	
T <sub>stg</sub>	Storage Temperature	-55 to 175	°C	
Τ <sub>i</sub>	Max. Operating Junction Temperature	-5510 175	C	

September 2003

#### THERMAL DATA

Rthj-case Rthj-amb	Thermal Resistance Junction-case Thermal Resistance Junction-ambient Maximum Lead Temperature For Soldering Purpose	Max Max	1.58 100 275	°C/W °C/W
11	Maximum Lead Temperature For Soldering Purpose		275	Ĵ

# **ELECTRICAL CHARACTERISTICS** (T<sub>CASE</sub> = 25 °C UNLESS OTHERWISE SPECIFIED) OFF

Symbol	Parameter	ameter Test Conditions		Тур.	Max.	Unit
V <sub>(BR)DSS</sub>	Drain-source Breakdown Voltage	I <sub>D</sub> = 25 mA, V <sub>GS</sub> = 0	24			V
IDSS	Zero Gate Voltage Drain Current (V <sub>GS</sub> = 0)	$V_{DS} = 20 V$ $V_{DS} = 20 V$ $T_{C} = 125^{\circ}C$			1 10	μΑ μΑ
I <sub>GSS</sub>	Gate-body Leakage Current (V <sub>DS</sub> = 0)	$V_{GS} = \pm 20V$			±100	nA

#### ON (5)

Symbol	Parameter	Test Conditions		Min.	Тур.	Max.	Unit
V <sub>GS(th)</sub>	Gate Threshold Voltage	$V_{DS} = V_{GS}$	I <sub>D</sub> = 250 μA	1	1.8	2.5	V
R <sub>DS(on)</sub>	Static Drain-source On Resistance	V <sub>GS</sub> = 10 V V <sub>GS</sub> = 5 V	I <sub>D</sub> = 30 A I <sub>D</sub> = 15 A		0.0052 0.007	0.006 0.011	Ω Ω

#### DYNAMIC

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
g <sub>fs</sub> <sup>(5)</sup>	Forward Transconductance	V <sub>DS</sub> = 10 V I <sub>D</sub> = 30 A		40		S
C <sub>iss</sub> C <sub>oss</sub> C <sub>rss</sub>	Input Capacitance Output Capacitance Reverse Transfer Capacitance	$V_{DS} = 15V f = 1 MHz V_{GS} = 0$		2850 800 120		pF pF pF
R <sub>G</sub>	Gate Input Resistance	f = 1 MHz Gate DC Bias = 0 Test Signal Level = 20 mV Open Drain		1		Ω



#### ELECTRICAL CHARACTERISTICS (continued)

#### SWITCHING ON

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
t <sub>d(on)</sub> t <sub>r</sub>	Turn-on Delay Time Rise Time			13 75		ns ns
Q <sub>g</sub> Q <sub>gs</sub> Q <sub>gd</sub>	Total Gate Charge Source Gate Charge Gate-Drain Charge	V <sub>DD</sub> = 10 V I <sub>D</sub> = 60 A V <sub>GS</sub> = 10 V		47.5 10 7	64	nC nC nC
Q <sub>oss</sub> (6)	Output Charge	V <sub>DS</sub> = 16 V V <sub>GS</sub> = 0 V		18.8		nC
Q <sub>gls</sub> (7)	Third-quadrant Gate Charge	V <sub>DS</sub> < 0 V V <sub>GS</sub> = 10 V		44		nC

#### SWITCHING OFF

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
t <sub>d(off)</sub> t <sub>f</sub>	Turn-off Delay Time Fall Time	$ \begin{array}{ll} V_{DD} = 10 \ V & I_D = 30 \ A \\ R_G = 4.7 \Omega, & V_{GS} = 10 \ V \\ (\text{Resistive Load, Figure 3}) \end{array} $		50 18	24.3	ns ns

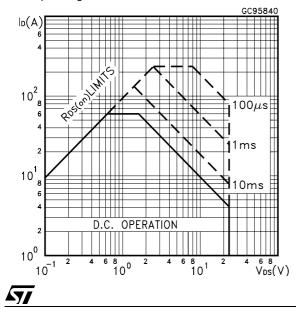
#### SOURCE DRAIN DIODE

Symbol	Parameter	Parameter Test Conditions		Min.	Тур.	Max.	Unit
I <sub>SD</sub> I <sub>SDM</sub>	Source-drain Current Source-drain Current (pulsed)					60 240	A A
V <sub>SD</sub> (5)	Forward On Voltage	I <sub>SD</sub> = 30 A	$V_{GS} = 0$			1.3	V
t <sub>rr</sub> Q <sub>rr</sub> I <sub>RRM</sub>	Reverse Recovery Time Reverse Recovery Charge Reverse Recovery Current	$I_{SD} = 60 \text{ A}$ $V_{DD} = 16 \text{ V}$ (see test circu	di/dt = 100A/µs T <sub>j</sub> = 150°C it, Figure 5)		35 35 2	47 47	ns nC A

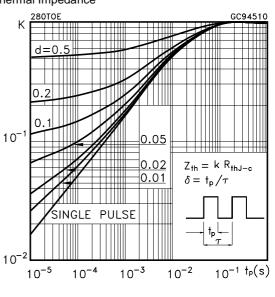
 $^{(1)}$  Garanted when external Rg=4.7  $\Omega$  and  $t_f < t_{fmax}.$   $^{(2)}$  Value limited by wire bonding  $^{(3)}$  Pulse width limited by safe operating area.  $^{(4)}$  Starting  $T_j$  = 25 °C,  $I_D$  = 30A,  $V_{DD}$  = 15V  $\,$ .

 $\stackrel{(5)}{=}$  Pulsed: Pulse duration = 300 µs, duty cycle 1.5 %.  $\stackrel{(6)}{=}$   $Q_{oss} = C_{oss}^*\Delta$  Vin ,  $C_{oss} = C_{gd} + C_{ds}$  . See Appendix A  $^{(7)}$  Gate charge for synchronous operation

Safe Operating Area

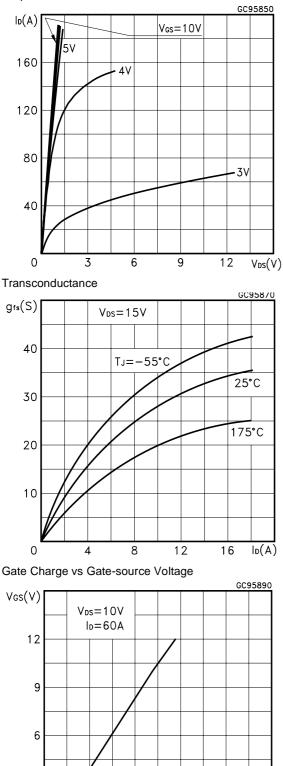


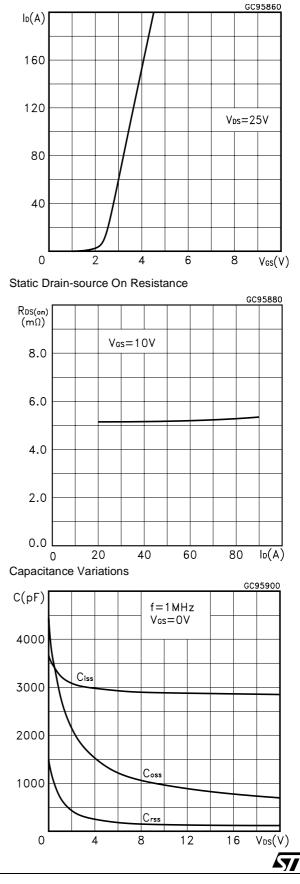
Thermal Impedance



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#### **Output Characteristics**





Transfer Characteristics

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0

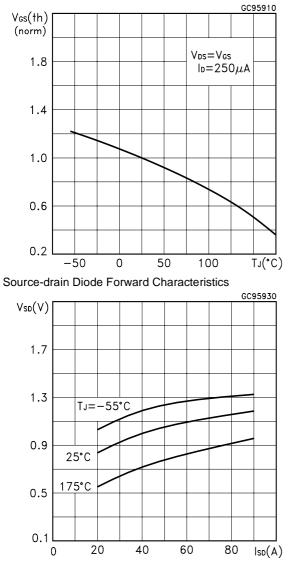
20

40

60

80

Q₀(nC)



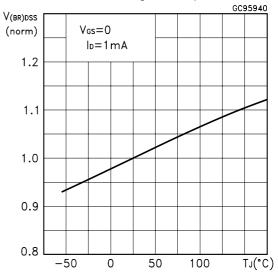
Normalized Gate Threshold Voltage vs Temperature

GC95920 Ros(on)  $V_{GS} = 10V$ (norm) 10=30A 1.6 1.4 1.2 1.0 0.8 -50 0 50 100 (D°)LT

Normalized on Resistance vs Temperature

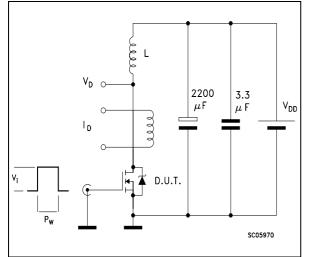




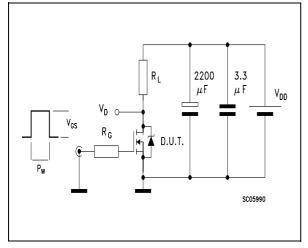


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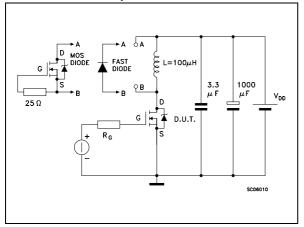
#### Fig. 1: Unclamped Inductive Load Test Circuit



**Fig. 3:** Switching Times Test Circuits For Resistive Load



**Fig. 5:** Test Circuit For Inductive Load Switching And Diode Recovery Times



#### Fig. 2: Unclamped Inductive Waveform

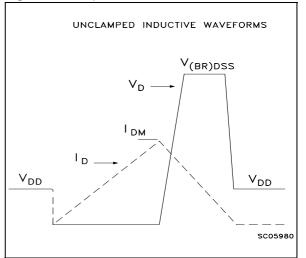
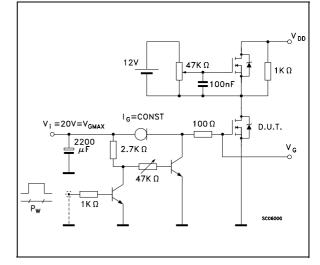


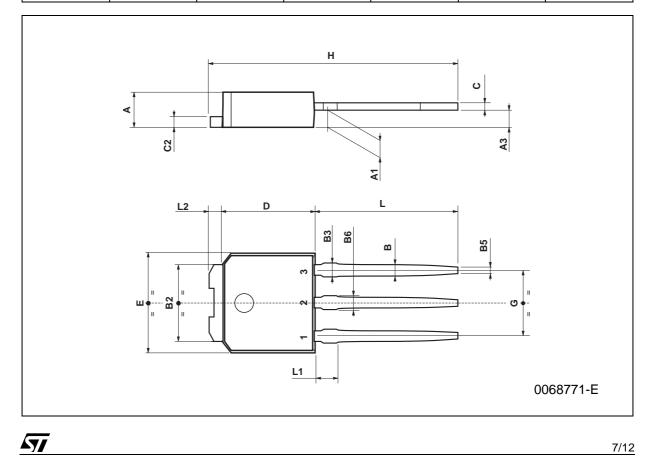
Fig. 4: Gate Charge test Circuit



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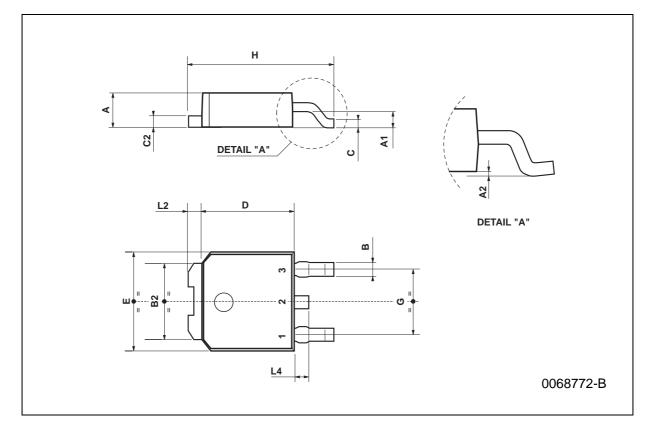
	mm			inch		
MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
2.2		2.4	0.086		0.094	
0.9		1.1	0.035		0.043	
0.7		1.3	0.027		0.051	
0.64		0.9	0.025		0.031	
5.2		5.4	0.204		0.212	
		0.85			0.033	
	0.3			0.012		
		0.95			0.037	
0.45		0.6	0.017		0.023	
0.48		0.6	0.019		0.023	
6		6.2	0.236		0.244	
6.4		6.6	0.252		0.260	
4.4		4.6	0.173		0.181	
15.9		16.3	0.626		0.641	
9		9.4	0.354		0.370	
0.8		1.2	0.031		0.047	
	0.9 0.7 0.64 5.2 0.45 0.45 0.48 6 6 6.4 4.4 15.9 9	0.9   0.7   0.64   5.2   0.3   0.45   0.48   6   6.4   4.4   15.9   9	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	$\begin{array}{c c c c c c c c c c c c c c c c c c c $	





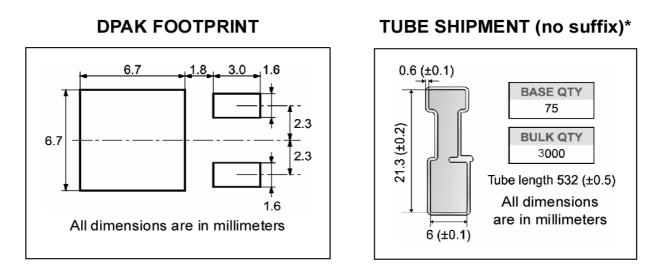
DIM.		mm			inch	
Dini.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
А	2.2		2.4	0.086		0.094
A1	0.9		1.1	0.035		0.043
A2	0.03		0.23	0.001		0.009
В	0.64		0.9	0.025		0.035
B2	5.2		5.4	0.204		0.212
С	0.45		0.6	0.017		0.023
C2	0.48		0.6	0.019		0.023
D	6		6.2	0.236		0.244
E	6.4		6.6	0.252		0.260
G	4.4		4.6	0.173		0.181
Н	9.35		10.1	0.368		0.397
L2		0.8			0.031	
L4	0.6		1	0.023		0.039

## TO-252 (DPAK) MECHANICAL DATA

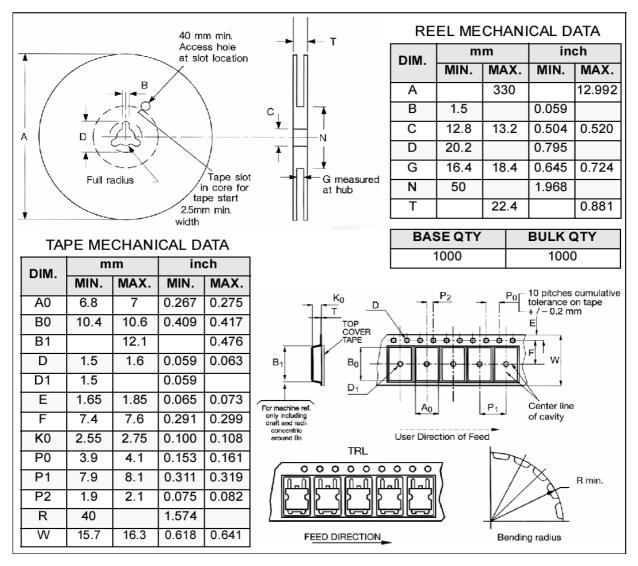


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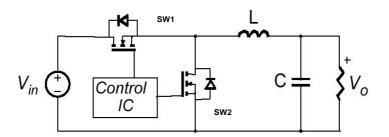


## TAPE AND REEL SHIPMENT (suffix "T4")\*



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### APPENDIX A Buck Converter: Power Losses Estimation



The power losses associated with the FETs in a Synchronous Buck converter can be estimated using the equations shown in the table below. The formulas give a good approximation, for the sake of performance comparison, of how different pairs of devices affect the converter efficiency. However a very important parameter, the working temperature, is not considered. The real device behavior is really dependent on how the heat generated inside the devices is emoved to allow for a safer working junction temperature.

The low side (SW2) device requires:

- Very low R<sub>DS(on)</sub> to reduce conduction losses
- Small Q<sub>gls</sub> to reduce the gate charge losses
- Small C<sub>oss</sub> to reduce losses due to output capacitance
- Small Q<sub>rr</sub> to reduce losses on SW<sub>1</sub> during its turn-on
- The  $C_{gd}/C_{gs}$  ratio lower than  $V_{th}/V_{gg}$  ratio especially with low drain to source voltage to avoid the cross conduction phenomenon;

The high side (SW1) device requires:

• Small  $R_g$  and  $L_s$  to allow higher gate current peak and to limit the voltage feedback on the gate

**Δ**Υ/

- Small Qg to have a faster commutation and to reduce gate charge losses
- Low R<sub>DS(on)</sub> to reduce the conduction losses.

		High Side Switch (SW1)	Low Side Switch (SW2)
Pconduct	ion	$R_{_{DS(on)SW1}}*I_{L}^{2}*d$	$R_{DS(on)SW2} * I_L^2 * (1-d)$
Pswitchin	ıg	$V_{in} * (Q_{gsth(SW1)} + Q_{gd(SW1)}) * f * \frac{I_L}{I_g}$	Zero Voltage Switching
P <sub>diode</sub>	Recovery	Not Applicable	$^{1}V_{in} * Q_{rr(SW2)} * f$
	Conduction	Not Applicable	$V_{f(SW2)} * I_L * t_{deadtime} * f$
Pgate(Qc	)	$Q_{g(SW1)} * V_{gg} * f$	$\mathbf{Q}_{\mathrm{gls(SW2)}} * \mathbf{V}_{\mathrm{gg}} * \mathbf{f}$
P <sub>Qoss</sub>		$\frac{V_{in} * Q_{oss(SW1)} * f}{2}$	$\frac{V_{in} * Q_{oss(SW2)} * f}{2}$

Parameter	Meaning
d	Duty-cycle
Qgsth	Post threshold gate charge
Q <sub>gls</sub>	Third quadrant gate charge
Pconduction	On state losses
Pswitching	On-off transition losses
Pdiode	Conduction and reverse recovery diode losses
Pgate	Gate drive losses
PQoss	Output capacitance losses

<sup>&</sup>lt;sup>1</sup> Dissipated by SW1 during turn-on

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