

## LOW POWER DIGITAL UHF PAGING RECEIVER

### GENERAL DESCRIPTION

The UAA2050T is a very low power UHF and VHF radio receiver circuit, primarily intended for use in paging receivers (27 MHz to 470 MHz) for wide-area digital paging systems employing direct FM non-return-to-zero (NRZ) frequency-shift keying (FSK) modulation.

Used in conjunction with the PCA5000T decoder for POCSAG paging systems, it offers an extremely advanced radio paging concept.

The radio receiver design is based on the offset receiver principle. The receiver provides fully filtered and squared data to drive the decoder and can be turned off completely by external inputs.

### Features

- Low noise preamplifier ensuring high RF sensitivity
- Few external components required
- Low current consumption
- Wide operating supply voltage range
- Power on/off mode selectable via the enable input (RE)
- Low battery voltage detector
- Crystal controlled receiver frequency (AFC)
- Fully compatible with all FSK modulated systems (512 and 1200 bits/s)
- Uses low cost crystal

### QUICK REFERENCE DATA

parameter	condition	symbol	min.	typ.	max.	unit
Supply voltage range		$V_P = V_{7-15}$	1.9	-	3.5	V
Supply voltage for preamplifier		$V_{23,24-15}$	1.0	-	3.5	V
Total supply current		$I_{tot}$	2.25	-	3.76	mA
Supply current OFF		$-I_{OFF}$	-	-	1.0	$\mu A$
RF sensitivity (RMS value)	$1 \times 10^{-2}$ bit error rate	$EMF/2$	-	0.18	0.25	$\mu V$
Preamplifier noise figure	note 1; $f = 470$ MHz	NF	-	4.5	-	dB
Operating ambient temperature range		$T_{amb}$	-10	-	+70	$^{\circ}C$

#### Note to the Quick reference data

1. Including the transforming network.

### PACKAGE OUTLINE

28-lead mini-pack, plastic (SO28; SOT136A).

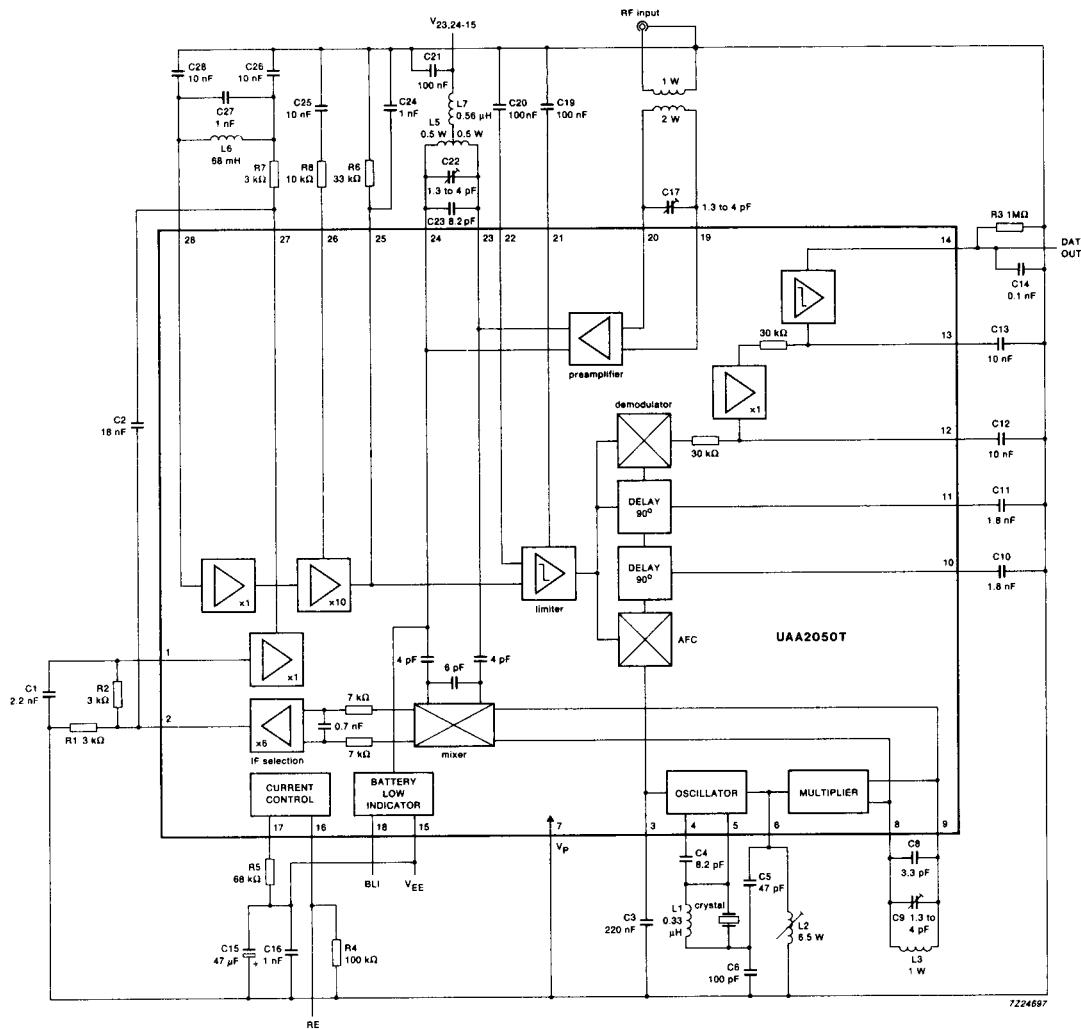


Fig.1 Block diagram (showing UHF external components).

## DEVELOPMENT DATA

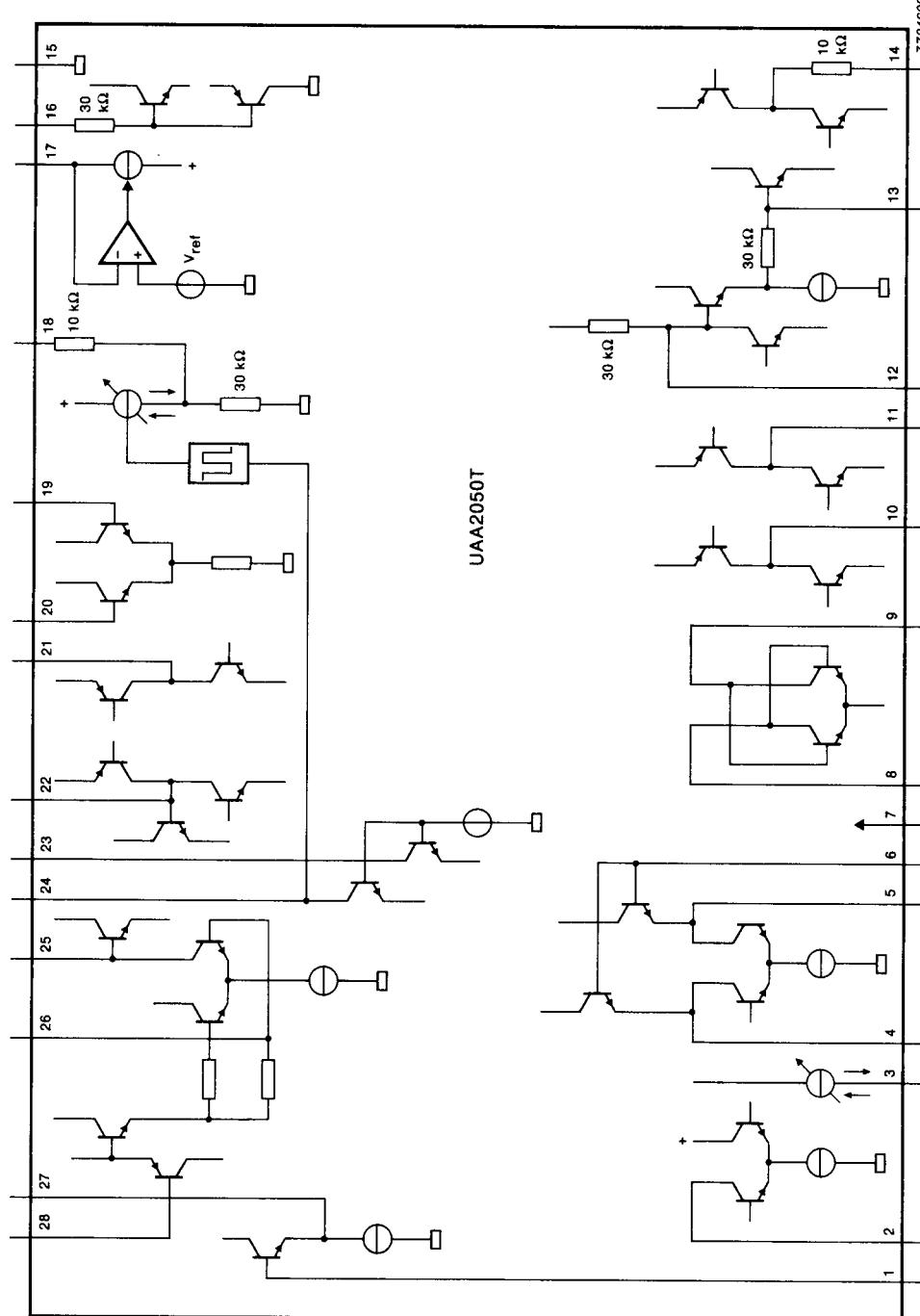


Fig.2 Pinning diagram and equivalent circuits.

**PINNING**

pin	mnemonic	description
1	IFF2	IF filter 2
2	IFF1	IF filter 1
3	AFC	AFC (test point 1)
4	OSC IN	oscillator input
5	OSC AFC	oscillator AFC range
6	OSC OUT	oscillator output
7	V <sub>P</sub>	supply voltage (positive)
8	MC1	multiplier coil
9	MC2	multiplier coil
10	AFCD	AFC delay
11	DEMODD	demodulator delay
12	DAT F1	data filter 1
13	DAT F2	data filter 2
14	DAT OUT	data output
15	V <sub>EE</sub>	supply voltage (negative)
16	RE	receiver enable input
17	CC	current control input
18	BLI	battery low indicator output
19	PREAMP1	preamplifier input 1
20	PREAMP2	preamplifier input 2
21	LC2	limiter decoupling 2
22	LC1	limiter decoupling 1
23	MIX1	mixer input 1 (preamplifier output)
24	MIX2	mixer input 2 (preamplifier output)
25	LIM IN	limiter input (test point 2)
26	IFF5	IF filter 5
27	IFF3	IF filter 3
28	IFF4	IF filter 4

**RATINGS**

Limiting values in accordance with the Absolute Maximum System (IEC134)

parameter	condition	symbol	min.	max.	unit
Supply voltage		$V_P = V_{7-15}$	-0.3	+5.0	V
Supply voltage for preamplifier		$V_{23,24-15}$	-0.3	+5.0	V
Operating ambient temperature range		$T_{amb}$	-10	+70	°C
Storage temperature range		$T_{stg}$	-55	+125	°C
Electrostatic handling; human body model*	except pins 19 and 20 only pins 19 and 20	$V_{ESD}$	-1000 -500	+1000 +1000	V V

\* Equivalent to discharging a 100 pF capacitor through a 1.5 kΩ series resistor.

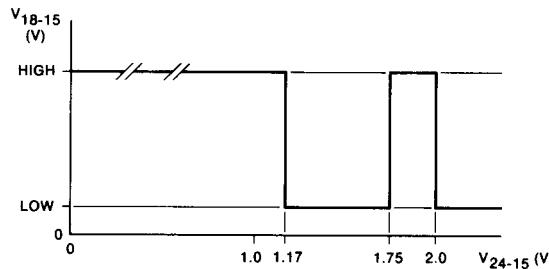
**DC CHARACTERISTICS**

$V_P = 2.0 \text{ V}$ ;  $V_{23,24-15} = 1.1 \text{ V}$ ; all voltages referenced to  $V_{EE}$ ;  $T_{amb} = -10 \text{ to } +55 \text{ }^\circ\text{C}$ ; typical values measured at  $T_{amb} = 25 \text{ }^\circ\text{C}$ ; test circuit as Fig.4; L2 and L3 short-circuited; unless otherwise specified

parameter	conditions	symbol	min.	typ.	max.	unit
<b>Supply</b>						
Supply voltage		$V_P = V_{7-15}$	1.9	2.0	3.5	V
Supply voltage for preamplifier	$V_{23,24-15} \leq V_P$	$V_{23,24-15}$	1.0	1.1	3.5	V
Supply current	$V_{RE} \geq V_P - 0.6 \text{ V}$	$I_7$	1.9	2.5	3.2	mA
Supply current for preamplifier	$V_{RE} \geq V_P - 0.6 \text{ V}$	$I_{23,24}$	0.35	0.45	0.56	mA
Supply current OFF	$V_{RE} \leq 0.4 \text{ V}$	$I_{OFF}$	—	—	1.0	$\mu\text{A}$
<b>Receiver enable (RE)</b>						
<b>POWER-OFF MODE:</b>						
input voltage		$V_{RE} = V_{16-15}$	—	—	0.4	V
input current	$V_{RE} = 0.4 \text{ V}$	$I_{16}$	—	—	1.0	$\mu\text{A}$
<b>POWER-ON MODE:</b>						
input voltage		$V_{RE} = V_{16-15}$	$V_P - 0.6 \text{ V}$	—	—	V
input current	$V_{RE} = 1.4 \text{ V}$	$I_{16}$	—	1.0	5.0	$\mu\text{A}$
<b>Data output (DAT OUT)</b>						
Output voltage HIGH	$I_{14} = \pm 10 \mu\text{A}$	$V_{14-15}$	$V_P - 0.7 \text{ V}$	—	—	V
Output voltage LOW	$I_{14} = \pm 10 \mu\text{A}$	$V_{14-15}$	—	—	0.5	V
<b>Battery voltage low indicator</b>						
Detection voltage	see Fig.3	$V_{DET}$ $V_{DET}$	1.10 1.90	1.17 2.00	1.24 2.10	V V
Output voltage HIGH	$1.0 \text{ V} \leq V_{24-15} \leq 1.10 \text{ V}; 1.85 \text{ V} \leq V_{24-15} \leq 1.90 \text{ V}; I_{18} = \pm 7 \mu\text{A}$	$V_{OH}$	$V_P - 0.5 \text{ V}$	—	—	V
Output voltage LOW	$1.24 \text{ V} \leq V_{24-15} \leq 1.65 \text{ V}; 2.10 \text{ V} \leq V_{24-15}; I_{18} = \pm 7 \mu\text{A}$	$V_{OL}$	—	—	0.5	V

**Note to the DC characteristics**

1.  $V_{18-15}$  goes HIGH if  $V_{24-15}$  is less than  $V_{DET}$ .



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Fig.3 Typical battery low indicator thresholds.

**AC CHARACTERISTICS (UHF)**

For AC test procedures refer to section 'TEST INFORMATION'.  $V_P = 2.0 \text{ V}$ ;  $V_{23,24-15} = 1.1 \text{ V}$ ; all voltages referenced to  $V_{EE}$ ;  $T_{amb} = -10 \text{ to } +55^\circ\text{C}$ ; typical values measured at  $T_{amb} = 25^\circ\text{C}$ ; test circuit as Fig.4 and printed-circuit board layout as Fig.7; test signal:  $f = 469.200 \text{ MHz}$ ; deviation =  $\pm 4.5 \text{ kHz}$ ; modulation = 256 Hz rectangular; channel spacing = 25 kHz; unless otherwise specified

parameter	conditions	symbol	min.	typ.	max.	unit
UHF sensitivity range (RMS value)	note 1; $1 \times 10^{-2}$ bit error rate	EMF/2	—	0.18	0.25	$\mu\text{V}$
		EMF/2	—	-122	-119	dBm
	$T_{amb} = -10 \text{ to } +70^\circ\text{C}$	EMF/2	—	—	-116	dBm
Adjacent channel selectivity		aa	60	67	—	dB
	$T_{amb} = -10 \text{ to } +70^\circ\text{C}$	aa	50	—	—	dB
Co-channel selectivity		ac	8	—	—	dB
Spurious response rejection		as	55	58	—	dB
Intermodulation response		IM	50	60	—	dB
Blocking	$\Delta f \geq \pm 4 \text{ MHz}$	EMF	80	83	—	$\text{dB}\mu\text{V}$
AFC lock-in range		$\pm \Delta f$	3	—	—	kHz
Preamplifier	see Fig.9					
Noise Figure	note 2	NF	—	4.5	—	dB
Third order intercept point		IP3	—	-20	—	dBm
Available power gain		$G_p$	—	6	—	dB
1 dB compression point (RMS value)		EMF/2	—	10	—	mV
VHF RF sensitivity range (RMS value)	see Fig.10; $f = 173.95 \text{ MHz}$ ; $1 \times 10^{-2}$ bit error rate	EMF/2	—	0.14	—	$\mu\text{V}$
		EMF/2	—	-124	—	dBm

**Notes to the AC characteristics**

1. A simple digital method of performing an approximate bit error rate (BER) measurement with a counter is shown in Fig.5. At high signal levels ( $10 \mu\text{V}$ ) the counter should read the exact frequency of the data input to the signal generator (256 Hz). As the signal level is reduced, errors occur at the receiver output and effectively changes the output frequency read by the counter (error duration is nearly always less than one bit length). The input signal level ( $V_{ref}$ ) is reduced until the bit error rate is  $\leq 1 \times 10^{-2}$  (5 bit errors for 512 bit/s system). The frequency from the data output signal will change from 256 Hz (512 bit/s system) to 261 Hz. This RF-level is the reference for the following tests ( $V_{ref}$ ).
2. Including the transforming network.

**TEST INFORMATION****Tuning procedure for AC tests**

1. After performing the DC tests, prepare the device for AC testing (as shown in Fig. 4).
2. Connect pin 3 to a voltage source of  $V_{3.7} = -0.5$  V. Measure the multiplier frequency with a counter or spectrum analyzer connected to the link winding of L3. Tune L2 to set the crystal oscillator to a frequency of:

$$\frac{\text{Received frequency} + 2.2 \text{ kHz}}{5} (\pm 100 \text{ Hz})$$

For a received frequency of  $f = 469.200$  MHz, the oscillator frequency ( $f_{osc}$ ) = 93.840 MHz.

3. Remove the test voltage source and turn on the signal generator ( $f = 469.200$  MHz; deviation =  $\pm 4.5$  kHz; rectangular 256 Hz modulation; RF input level = 1 mV).

**Note** During the following tests the RF signal generator level should be reduced as the receiver is tuned, to ensure the peak-to-peak output voltage at pin 25 lies between 20 mV and 100 mV.

4. Tune C9 (multiplier) to obtain a peak audio output voltage on pin 25.
5. Tune C22 (Mixer input) to obtain a peak audio output voltage on pin 25.
6. Disconnect the frequency counter from the multiplier output. Measure the voltage at pin 3 and check if it is within the range of -0.48 V to -0.52 V. If it is outside this range then adjust L2 (oscillator) until it is within the limits.
7. Check with an oscilloscope that clean data is appearing on the data output (pin 14) and proceed with the AC tests.

**Test conditions**

The data output signal corresponds to the sensitivity definition.

**Where:**

- $f_1$  is the modulated test signal
- $f_2$  is the unmodulated test signal
- $f_{cs}$  is the channel spacing (25 kHz)
- $V_1$  is the signal generator 1 output
- $V_2$  is the signal generator 2 output
- $V_{ref}$  is defined in 'Notes to the AC characteristics'.

1. Adjacent channel selectivity (see Fig.6);
  - generator 1: modulated test signal;  $V_1 = V_{ref} + 3 \text{ dB}$
  - generator 2: unmodulated test signal;  $V_2 = V_1 + 60 \text{ dB}$  ( $f_2 = f_1 \pm \Delta f_{cs}$ ).
2. Co-channel selectivity (see Fig.6);
  - generator 1: modulated test signal;  $V_1 = V_{ref} + 3 \text{ dB}$
  - generator 2: unmodulated test signal;  $V_2 = V_1 - 8 \text{ dB}$  ( $f_2 = f_1 \pm 3 \text{ kHz}$  maximum).
3. Spurious response rejection (see Fig.6);
  - generator 1: modulated test signal;  $V_1 = V_{ref} + 3 \text{ dB}$
  - generator 2: unmodulated test signal;  $V_2 = V_1 + 55 \text{ dB}$  ( $f_2 = 100 \text{ kHz}$  to  $1 \text{ GHz}$ ;  $|f_2 - f_1| \geq 2\Delta f_{cs}$ ).
4. Intermodulation response (see Fig.6);
  - generator 1: modulated test signal;  $f_1 = 469.2 \text{ MHz} \pm 2 \times n \times \Delta f_{cs}$   $n = 1 \text{ to } 4$ .
  - generator 2: unmodulated test signal;  $f_2 = 469.2 \text{ MHz} \pm n \times \Delta f_{cs}$ ,  $n = 1 \text{ to } 4$ .

Output voltages of both generators increase with the same level, until a data output signal corresponding to the sensitivity definition is reached. The level must be 50 dB above the  $V_{ref} + 3 \text{ dB}$  level.

5. Blocking (see Fig.6);
  - generator 1: modulated test signal;  $V_1 (\text{EMF}) = 3 \text{ dB}\mu\text{V}$
  - generator 2: unmodulated test signal;  $V_2 (\text{EMF}) = 80 \text{ dB}\mu\text{V}$  ( $f_2 = f_1 \pm \Delta f$ ;  $\Delta f \geq 4 \text{ MHz}$ )
6. AFC lock-in-range (see Fig.5);
  - generator 1: modulated test signal;  $V_1 = V_{ref} + 3 \text{ dB}$  ( $f_1 = 469.2 \text{ MHz} \pm 3 \text{ kHz}$ )

## DEVELOPMENT DATA

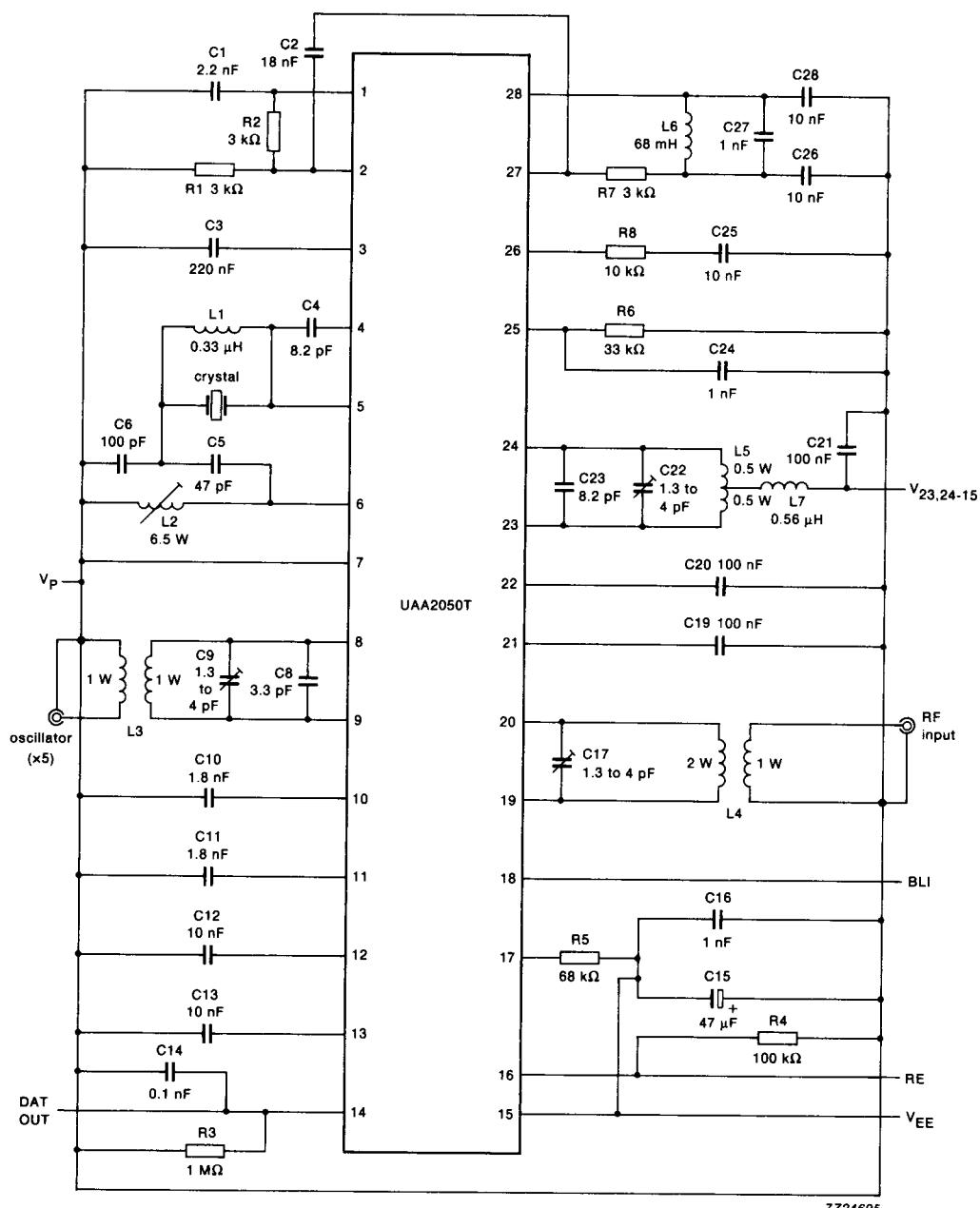


Fig.4 Test circuit (UHF).

**Crystal data**

Test frequency for the receiver chip is 469.2 MHz, crystal frequency is 93.838 MHz.

Static capacitance C0 (max.) = 4 pF

Dynamic capacitance C1 = 0.4 fF  $\pm 20\%$

Dynamic resistance R1 (max.) = 75  $\Omega$

Temperature drift =  $\pm 5 \times 10^{-6}$ .

**Inductors**

L3, L4 and L5: wound with 0.9 mm silvered wire, without pot or core and diameter of 4.5 mm.

L1 = 0.33  $\mu$ H chip inductor (at 25 MHz minimum value of Q = 12)

L2 = 6.5 turns

L3 = 1 turn

L4 = 2 turns

L5 = 1 turn

L6 = inductor, Philips microchoke (at 10 kHz minimum value of Q = 10)

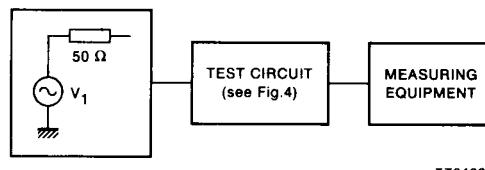


Fig.5 Test figure A.

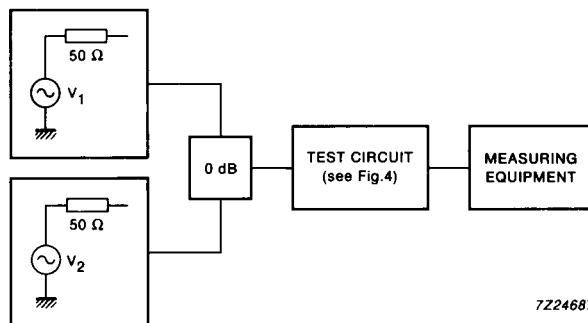
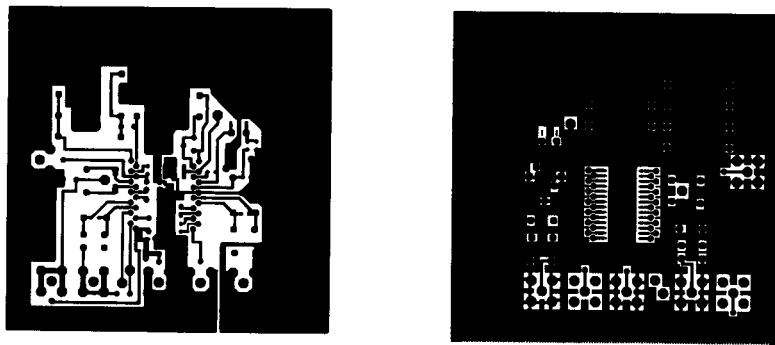


Fig.6 Test figure B.



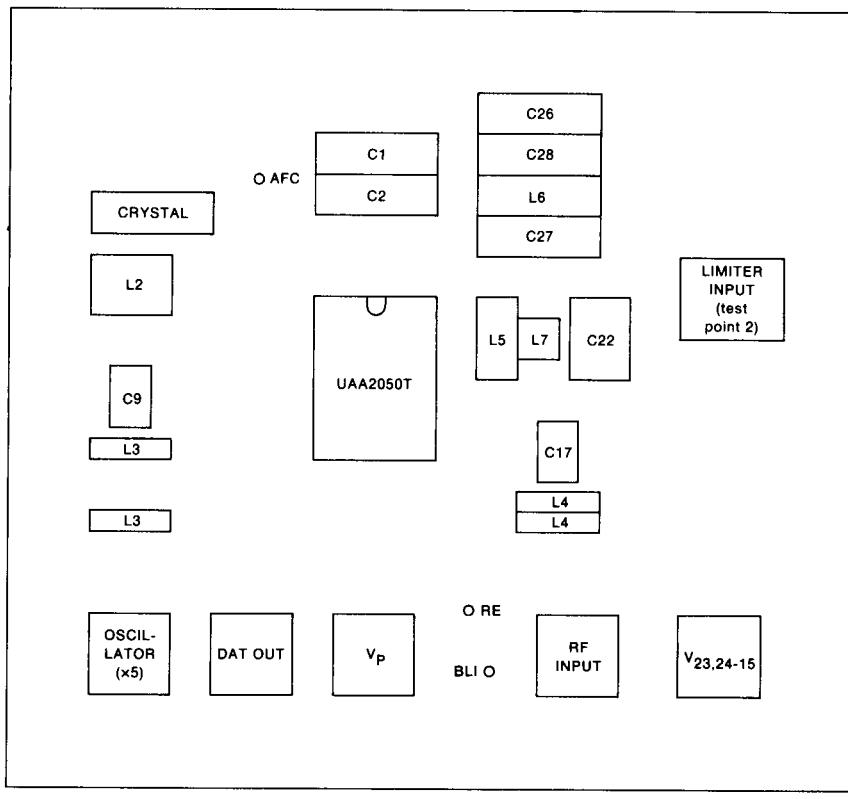
(a) underside

(b) top side

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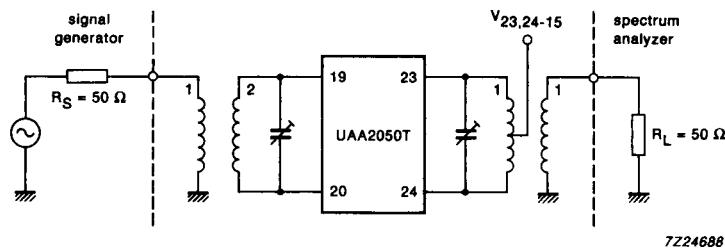
Fig. 7 Printed-circuit board for UHF range (see Fig.4).

## DEVELOPMENT DATA



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Fig. 8 Printed-circuit board for UHF range (component arrangement).



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Fig.9 Test circuit for the preamplifier.

## DEVELOPMENT DATA

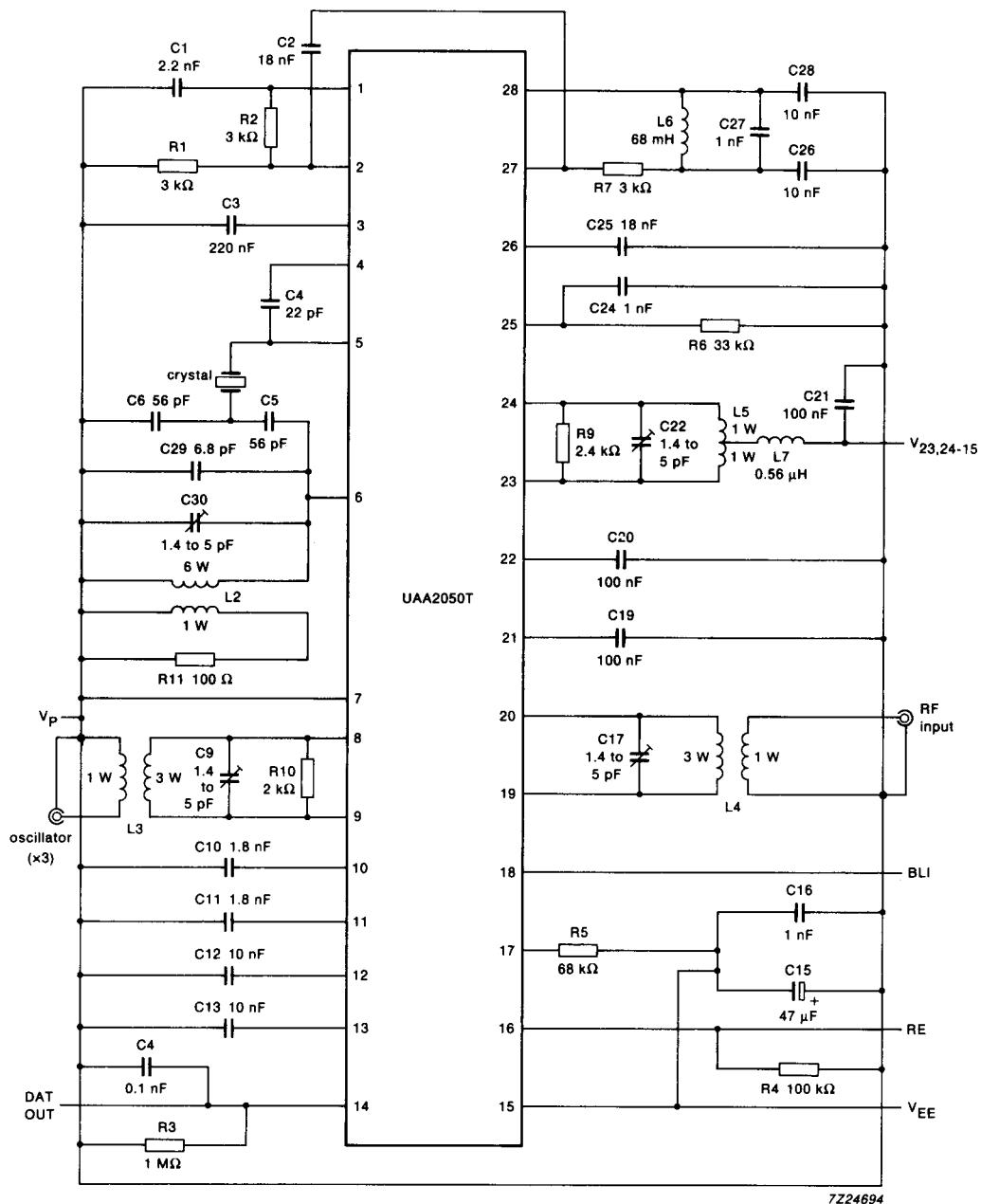


Fig.10 Test circuit (VHF).

## Crystal data

Test frequency for the receiver chip is 173.95 MHz, crystal frequency is 57.9859 MHz.

## Inductors

L2, L3, L4 and L5: wound with 0.3 mm enamelled copper wire on Toko 4.5 mm diameter former, without pot or core.  
Screening cans are also used.

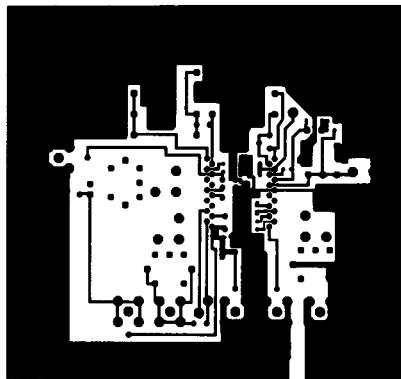
L2 = 6 turns, 2 turns per groove; link winding, 1 turn over the centre of the other winding

L3 = 3 turns, 1 turns per groove; link winding, 1 turn at the bottom of the former

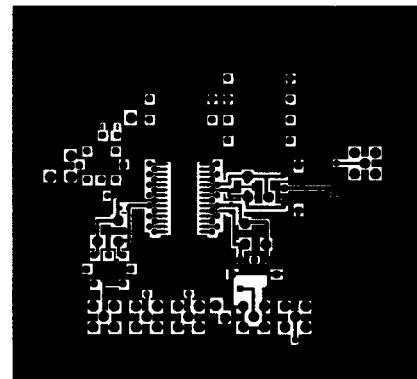
L4 = 3 turns, 1 turns per groove; link winding, 1 turn at the centre of the other winding

L5 = 2 turns

L6 = inductor, Philips microchoke (at 10 kHz minimum value of  $Q = 10$ )



(a) underside



(b) top side

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Fig. 11 Printed-circuit board for VHF range (see Fig.10).

DEVELOPMENT DATA

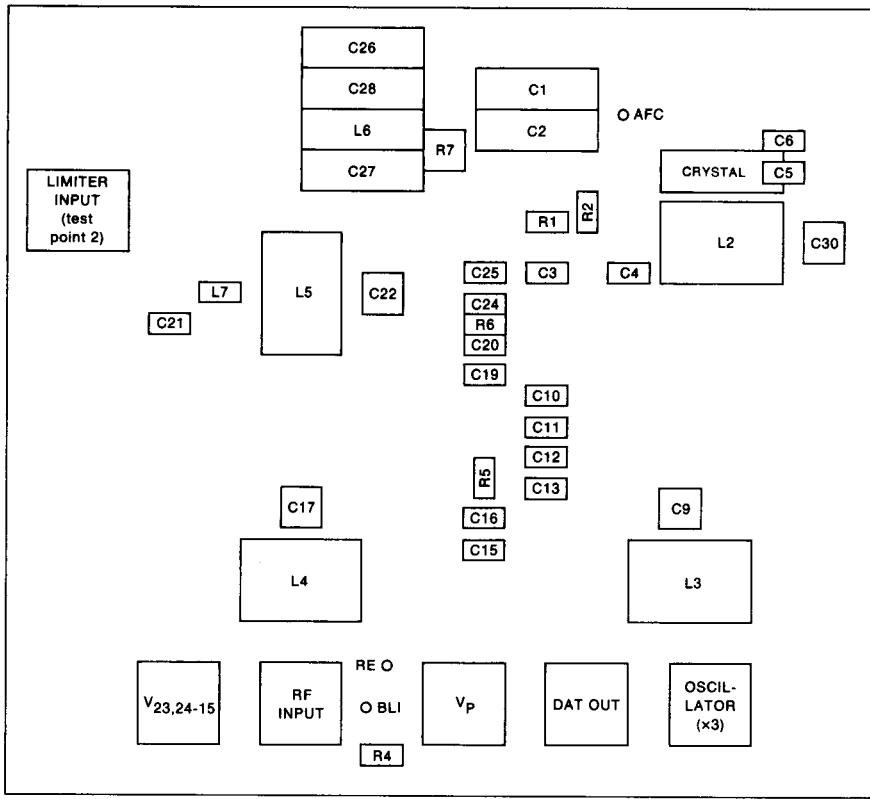


Fig. 12 Printed-circuit board for VHF range (component arrangement).

## APPLICATION INFORMATION

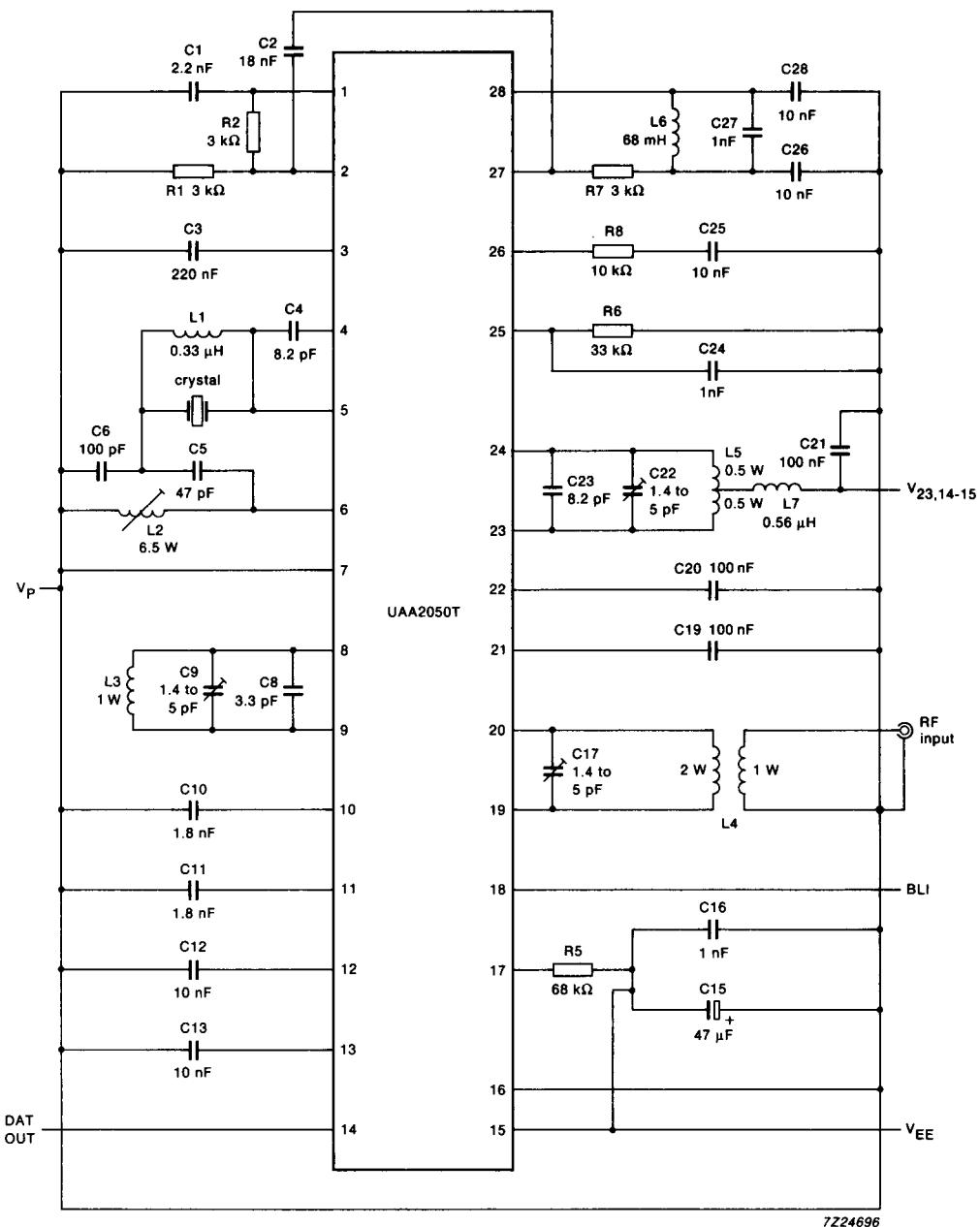


Fig.13 Application diagram (UHF).

## Note to Fig.13

For information concerning the crystal oscillator and inductors, refer to Fig.4.