

## 16Mx72 DDR SDRAM iNTEGRATED Plastic Encapsulated Microcircuit

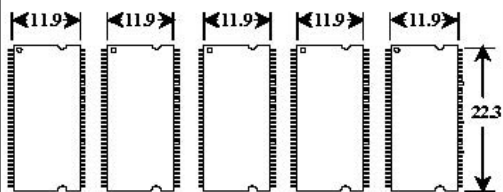
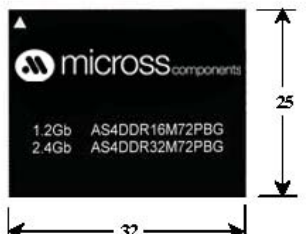
### FEATURES

- DDR SDRAM Data Rate = 200, 250, 266, 333Mbps
- Package:
  - 219 Plastic Ball Grid Array (PBGA), 32 x 25mm
- 2.5V  $\pm 0.2$ V core power supply
- 2.5V I/O (SSTL\_2 compatible)
- Differential clock inputs (CLK and CLK#)
- Commands entered on each positive CLK edge
- Internal pipelined double-data-rate (DDR) architecture; two data accesses per clock cycle
- Programmable Burst length: 2, 4 or 8
- Bidirectional data strobe (DQS) transmitted/received with data, i.e., source-synchronous data capture (one per byte)
- DQS edge-aligned with data for READs; center-aligned with data for WRITEs
- DLL to align DQ and DQS transitions with CLK
- Four internal banks for concurrent operation
- Two data mask (DM) pins for masking write data
- Programmable IOL/IOH option
- Auto precharge option
- Auto Refresh and Self Refresh Modes
- Industrial, Enhanced and Military Temperature Ranges
- Organized as 16M x 72/80
- Weight: AS4DDR16M72PBG = 3.50 grams typical

\* This product and or it's specifications is subject to change without notice..

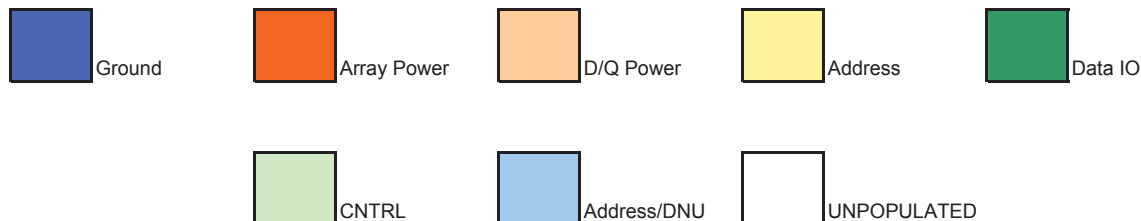
### BENEFITS

- 40% SPACE SAVINGS
- Reduced part count
- Reduced I/O count
  - 34% I/O Reduction
- Reduced trace lengths for lower parasitic capacitance
- Suitable for hi-reliability applications
- Laminate interposer for optimum TCE match
- Upgradeable to 32M x 72 density (AS4DDR32M72PBG)
- Meets or exceeds published specifications of White's W3E16M72S-XBX

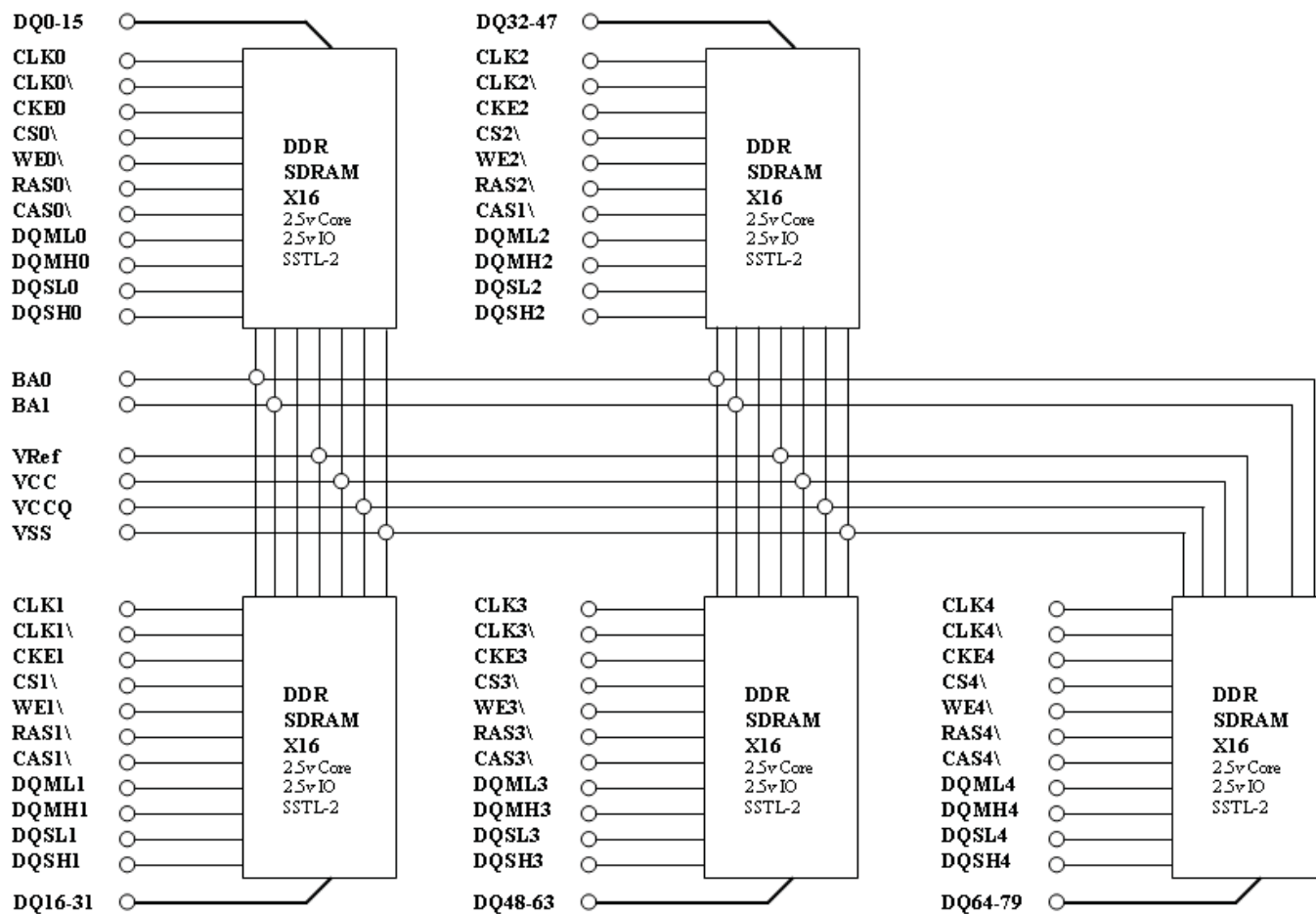
	Monolithic Solution	Integrated MCP Solution	
OPTIONS			SAVINGS
Area	5 x 265mm <sup>2</sup> = 1328mm <sup>2</sup> Plus	800mm <sup>2</sup>	40+ %
I/O Count	5 x 66 pins = 330 pins	219 Balls	34%

## SDRAM-DDR PINOUT TOP VIEW

	DQ0	DQ14	DQ15	VSS	VSS	A9	A10	A11	A8	VCCQ	VCCQ	DQ16	DQ17	DQ31	VSS
DQ1	DQ2	DQ12	DQ13	VSS	VSS	A0	A7	A6	A1	VCC	VCC	DQ18	DQ19	DQ29	DQ30
DQ3	DQ4	DQ10	DQ11	VCC	VCC	A2	A5	A4	A3	VSS	VSS	DQ20	DQ21	DQ27	DQ28
DQ6	DQ5	DQ8	DQ9	VCCQ	VSSQ	A12	DNU	DNU	DNU	VSS	VSS	DQ22	DQ23	DQ26	DQ25
DQ7	DQML0	VCC	DQMH0	DQSH3	DQSL0	DQSH0	BA0	BA1	DQSL1	DQSH1	VREF	DQML1	VSS	NC	DQ24
CA S0\	WE0\	VCC	CLK0	DQSL3							RAS1\	WE1\	VSS	DQMH1	CLK1
CS0\	RAS0\	VCC	CKE0	CLK0\							CAS\	CS1\	VSS	CLK1\	CKE1
VSS	VSS	VCC	VCCQ	VSS							VCC	VSS	VSS	VCCQ	VCC
VSS	VSS	VCC	VCCQ	VSS							VCC	VSS	VSS	VCCQ	VCC
CLK3\	CKE3	VCC	CS3\	DQSL4							CLK2\	CKE2	VSS	RAS2\	CS2\
NC	CLK3	VCC	CAS3\	RAS3\							DQSL2	CLK2	VSS	WE2\	CAS2\
DQ56	DQMH3	VCC	WE3\	DQML3	CKE4	DQMH4	CLK4	CAS4\	WE4\	RAS4\	CS4\	DQMH2	VSS	DQML2	DQ39
DQ57	DQ58	DQ55	DQ54	DQSH4	CLK4\	DQ73	DQ72	DQ71	DQ70	DQML4	DQSH2\	DQ41	DQ40	DQ37	DQ38
DQ60	DQ59	DQ53	DQ52	VSS	VSS	DQ75	DQ74	DQ69	DQ68	VCC	VCC	DQ43	DQ42	DQ36	DQ35
DQ62	DQ61	DQ51	DQ50	VCC	VCC	DQ77	DQ76	DQ67	DQ66	VSS	VSS	DQ45	DQ44	DQ34	DQ33
VSS	DQ63	DQ49	DQ48	VCCQ	VCCQ	DQ79	DQ78	DQ65	DQ64	VSS	VSS	DQ47	DQ46	DQ32	VCC



## FUNCTIONAL BLOCK DIAGRAM



## PIN DEFINITIONS / FUNCTIONAL DESCRIPTION

BGA Locations	SYMBOL	DESCRIPTION
F4, F16, G5, G15, K1, K12, L2, L13, N6, M8	CKx, CKx\	Clock: CKx and CKx\ are differential clock inputs. All address and control input signals are sampled on the crossing of the positive edge of CKx and negative edge of CKx\.
G4, G16, K2, K13, M6	CKEx	Output data (DQ's and DQS) is referenced to the crossings of the differential clock inputs.
G4, G16, K2, K13, M6	CKEx	Clock Enable: CKE controls the clock inputs. CKE high enables, CKE Low disables the clock input pins. Driving CKE Low provides PRECHARGE POWER-DOWN. CKE is synchronous for POWER-DOWN entry and exit, and for SELF REFRESH entry CKE is Asynchronous for SELF REFRESH exit and for disabling the outputs. CKE must be maintained HIGH throughout read and write accesses. Input buffers are disabled during POWER-DOWN Input buffers are disabled during SELF REFRESH. CKE is an SSTL-2 input but will detect an LVCMOS LOW level after VCC is applied
G1, G13, K4, K16, M12	CSx\	Chip Select: CSx\ enables the COMMAND register(s) of each of the five (5) contained words. All commands are masked with CSx\ is registered HIGH. CSx\ provides for external bank selection on systems with multiple banks. CSx\ is considered part of the COMMAND CODE.
F4, F16, G5, G15, K1, K12, L2, L13, N7, M9	RASx\, CASx\, Wex\	Command Inputs: RASx, CASx and Wex\ define the command being entered.
G4, G16, K2, K14, M7	DQMLx, DQMHx	Input Data Mask: DM is an input mask signal for write data. Input data is masked when DQMLx or Hx is sampled HIGH at time of a WRITE access. DM is sampled on both edges of DQSLx and DQSHx.
E8, E9	BA0, BA1	Bank Address Inputs: BA0, BA1, define which bank an ACTIVE READ, WRITE or PRECHARGE Command is being applied.
A7, A8, A9, A10, B7, B8, B9, B10, C7, C8, C9, C10, D7	A0-11, A12	Address Input: Provide the row address for Active commands, and the column address and auto precharge bit (A10) for READ / WRITE commands to select one location out of the memory array in the respective bank. A10 sampled during a PRECHARGE command determines whether the PRECHARGE applies to one bank or all banks. The address inputs also provide the op-code during a MODE RESISTER SET command.
A2, A3, A4, A13, A14, B1, B2, B3, B4, B13, B14, B15, B16, C1, C2, C3, C4, C13, C14, C15, C16, D1, D2, D3, D4, D13, D14, D15, D16, E1, E16, M1, M16, N1, N2, N3, N4, N13, N14, N15, N16, T2, T3, T4, T13, T14, T15, N7, N8, N9, N10, P7, P8, P9, P10, R7, R8, R9, R10, T7, T8, T9, T10	DQ0-79	Data I/O
E6, E7, E10, E11, F5, K5, L12, N5, N12, E5	DQSLX, DQSHX	Data Strobe: Output with read data, input with write data. DQS is edge-aligned with read data, centered in write data. It is used to capture data
B11, B12, C5, C6, E3, F3, G3, H3, H12, H16, J3, J12, J16, K3, L3, M3, P11, P12, R5, R6, T16	VCC	Core Power Supply
A11, A12, D5, D6, H4, H15, J4, J15, T5, T6	VCCQ	I/O Power Supply
A5, A6, A16, B5, B6, C11, C12, D11, D12, E14, F14, G14, H1, H2, H5, H13, H14, J1, J2, J5, J13, J14, K14, L14, P5, P6, R11, R12, T1, T11, T12, M14	VSS	Ground (Digital)
E12	VREF	SSTL-2 Reference Voltage

## GENERAL DESCRIPTION

The 1.2Gb DDR SDRAM MCM, is a high-speed CMOS, dynamic random-access, memory using 5 chips containing 268,435,456 bits. Each chip is internally configured as a quad-bank DRAM. Each of the chip's 67,108,864-bit banks is organized as 8,192 rows by 512 columns by 16 bits.

The 128MB(1.2Gb) DDR SDRAM MCM uses a DDR architecture to achieve high-speed operation. The double data rate architecture is essentially a 2n-prefetch architecture with an interface designed to transfer two data words per clock cycle at the I/O pins. A single read or write access for the 128MB DDR SDRAM effectively consists of a single 2n-bit wide, one-clock-cycle data transfer at the internal DRAM core and two corresponding n-bit wide, one-half-clock-cycle data transfers at the I/O pins.

A bidirectional data strobe (DQS) is transmitted externally, along with data, for use in data capture at the receiver. DQS is a strobe transmitted by the DDR SDRAM during READs and by the memory controller during WRITES. DQS is edgealigned with data for READs and center-aligned with data for WRITES. Each chip has two data strobes, one for the lower byte and one for the upper byte.

The 128MB DDR SDRAM operates from a differential clock (CLK and CLK#); the crossing of CLK going HIGH and CLK# going LOW will be referred to as the positive edge of CLK. Commands (address and control signals) are registered at every positive edge of CLK. Input data is registered on both edges of DQS, and output data is referenced to both edges of DQS, as well as to both edges of CLK.

Read and write accesses to the DDR SDRAM are burst oriented; accesses start at a selected location and continue for a programmed number of locations in a programmed sequence. Accesses begin with the registration of an ACTIVE command, which is then followed by a READ or WRITE command. The address bits registered coincident with the ACTIVE command are used to select the bank and row to be accessed. The address bits registered coincident with the READ or WRITE command are used to select the bank and the starting column location for the burst access.

The DDR SDRAM provides for programmable READ or WRITE burst lengths of 2, 4, or 8 locations. An auto precharge function may be enabled to provide a selftimed row precharge that is initiated at the end of the burst access.

The pipelined, multibank architecture of DDR SDRAMs allows for concurrent operation, thereby providing high effective bandwidth by hiding row precharge and activation time.

An auto refresh mode is provided, along with a powersaving power-down mode.

## FUNCTIONAL DESCRIPTION

Read and write accesses to the DDR SDRAM are burst oriented; accesses start at a selected location and continue for a programmed number of locations in a programmed sequence. Accesses begin with the registration of an ACTIVE command which is then followed by a READ or WRITE command. The address bits registered coincident with the ACTIVE command are used to select the bank and row to be accessed (BA0 and BA1 select the bank, A0-12 select the row). The address bits registered coincident with the READ or WRITE command are used to select the starting column location for the burst access.

Prior to normal operation, the SDRAM must be initialized. The following sections provide detailed information covering device initialization, register definition, command descriptions and device operation.

## INITIALIZATION

DDR SDRAMs must be powered up and initialized in a predefined manner. Operational procedures other than those specified may result in undefined operation. Power must first be applied to VCC and VCCQ simultaneously, and then to VREF (and to the system VTT). VTT must be applied after VCCQ to avoid device latch-up, which may cause permanent damage to the device. VREF can be applied any time after VCCQ but is expected to be nominally coincident with VTT. Except for CKE, inputs are not recognized as valid until after VREF is applied. CKE is an SSTL\_2 input but will detect an LVCMOS LOW level after VCC is applied. Maintaining an LVCMOS LOW level on CKE during powerup is required to ensure that the DQ and DQS outputs will be in the High-Z state, where they will remain until driven in normal operation (by a read access). After all power supply and reference voltages are stable, and the clock is stable, the DDR SDRAM requires a 200µs delay prior to applying an executable command.

Once the 200µs delay has been satisfied, a DESELECT or NOP command should be applied, and CKE should be brought HIGH. Following the NOP command, a PRECHARGE ALL command should be applied. Next a LOAD MODE REGISTER command should be issued for the extended mode register (BA1 LOW and BA0 HIGH) to enable the DLL, followed by another LOAD MODE REGISTER command to the mode register (BA0/BA1 both LOW) to reset the DLL and to program the operating parameters. Two-hundred clock cycles are required between the DLL reset and any READ command. A PRECHARGE ALL command should then be applied, placing the device in the all banks idle state.

Once in the idle state, two AUTO REFRESH cycles must be performed (trfc must be satisfied.) Additionally, a LOAD MODE REGISTER command for the mode register with the reset DLL bit deactivated (i.e., to program operating parameters without resetting the DLL) is required. Following these requirements, the DDR SDRAM is ready for normal operation.

## REGISTER DEFINITION

### MODE REGISTER

The Mode Register is used to define the specific mode of operation of the DDR SDRAM. This definition includes the selection of a burst length, a burst type, a CAS latency, and an operating mode, as shown in Figure 3. The Mode Register is programmed via the MODE REGISTER SET command (with BA0 = 0 and BA1 = 0) and will retain the stored information until it is programmed again or the device loses power. (Except for bit A8 which is self clearing).

Reprogramming the mode register will not alter the contents of the memory, provided it is performed correctly. The Mode Register must be loaded (reloaded) when all banks are idle and no bursts are in progress, and the controller must wait the specified time before initiating the subsequent operation. Violating either of these requirements will result in unspecified operation. Mode register bits A0-A2 specify the burst length, A3 specifies the type of burst (sequential or interleaved), A4-A6 specify the CAS latency, and A7-A12 specify the operating mode.

### BURST LENGTH

Read and write accesses to the DDR SDRAM are burst oriented, with the burst length being programmable, as shown in Figure 3. The burst length determines the maximum number of column locations that can be accessed for a given READ or WRITE command. Burst lengths of 2, 4 or 8 locations are available for both the sequential and the interleaved burst types.

Reserved states should not be used, as unknown operation or incompatibility with future versions may result.

When a READ or WRITE command is issued, a block of columns equal to the burst length is effectively selected. All accesses for that burst take place within this block, meaning that the burst will wrap within the block if a boundary is reached. The block is uniquely selected by A1-Ai when the burst length is set to two; by A2-Ai when the burst length is set to four (where Ai is the most significant column address for a given configuration); and by A3-Ai when the burst length is set to eight. The remaining (least significant) address bit(s) is (are) used to select the starting location within the block. The programmed burst length applies to both READ and WRITE bursts.

### BURST TYPE

Accesses within a given burst may be programmed to be either sequential or interleaved; this is referred to as the burst type and is selected via bit M3.

The ordering of accesses within a burst is determined by the burst length, the burst type and the starting column address, as shown in Table 1.

**TABLE 1 - BURST DEFINITION**

Burst Length	Starting Column Address	Order of Accesses Within a Burst	
		Type = Sequential	Type = Interleaved
2	A0		
	0	0-1	0-1
	1	1-0	1-0
4	A1 A0		
	0 0	0-1-2-3	0-1-2-3
	0 1	1-2-3-0	1-0-3-2
	1 0	2-3-0-1	2-3-0-1
	1 1	3-0-1-2	3-2-1-0
8	A2 A1 A0		
	0 0 0	0-1-2-3-4-5-6-7	0-1-2-3-4-5-6-7
	0 0 1	1-2-3-4-5-6-7-0	1-0-3-2-5-4-7-6
	0 1 0	2-3-4-5-6-7-0-1	2-3-0-1-6-7-4-5
	0 1 1	3-4-5-6-7-0-1-2	3-2-1-0-7-6-5-4
	1 0 0	4-5-6-7-0-1-2-3	4-5-6-7-0-1-2-3
	1 0 1	5-6-7-0-1-2-3-4	5-4-7-6-1-0-3-2
	1 1 0	6-7-0-1-2-3-4-5	6-7-4-5-2-3-0-1
	1 1 1	7-0-1-2-3-4-5-6	7-6-5-4-3-2-1-0

#### NOTES:

- For a burst length of two, A1-Ai select two-data-element block; A0 selects the starting column within the block.
- For a burst length of four, A2-Ai select four-data-element block; A0-1 select the starting column within the block.
- For a burst length of eight, A3-Ai select eight-data-element block; A0-2 select the starting column within the block.
- Whenever a boundary of the block is reached within a given sequence above, the following access wraps within the block.

### READ LATENCY

The READ latency is the delay, in clock cycles, between the registration of a READ command and the availability of the first bit of output data. The latency can be set to 2 or 2.5 clocks.

If a READ command is registered at clock edge n, and the latency is m clocks, the data will be available by clock edge n+m. Table 2 below indicates the operating frequencies at which each CAS latency setting can be used.

Reserved states should not be used as unknown operation or incompatibility with future versions may result.

**TABLE 2 - CAS LATENCY**

SPEED	ALLOWABLE OPERATING FREQUENCY (MHz)	
	CAS LATENCY=2	CAS LATENCY=2.5
-10	≤ 75	≤ 100
-8	≤ 100	≤ 125
-75	≤ 100	≤ 133
-6	≤ 100	≤ 166



## OPERATING MODE

The normal operating mode is selected by issuing a MODE REGISTER SET command with bits A7-A12 each set to zero, and bits A0-A6 set to the desired values. A DLL reset is initiated by issuing a MODE REGISTER SET command with bits A7 and A9-A12 each set to zero, bit A8 set to one, and bits A0-A6 set to the desired values. Although not required, JEDEC specifications recommend when a LOAD MODE REGISTER command is issued to reset the DLL, it should always be followed by a LOAD MODE REGISTER command to select normal operating mode.

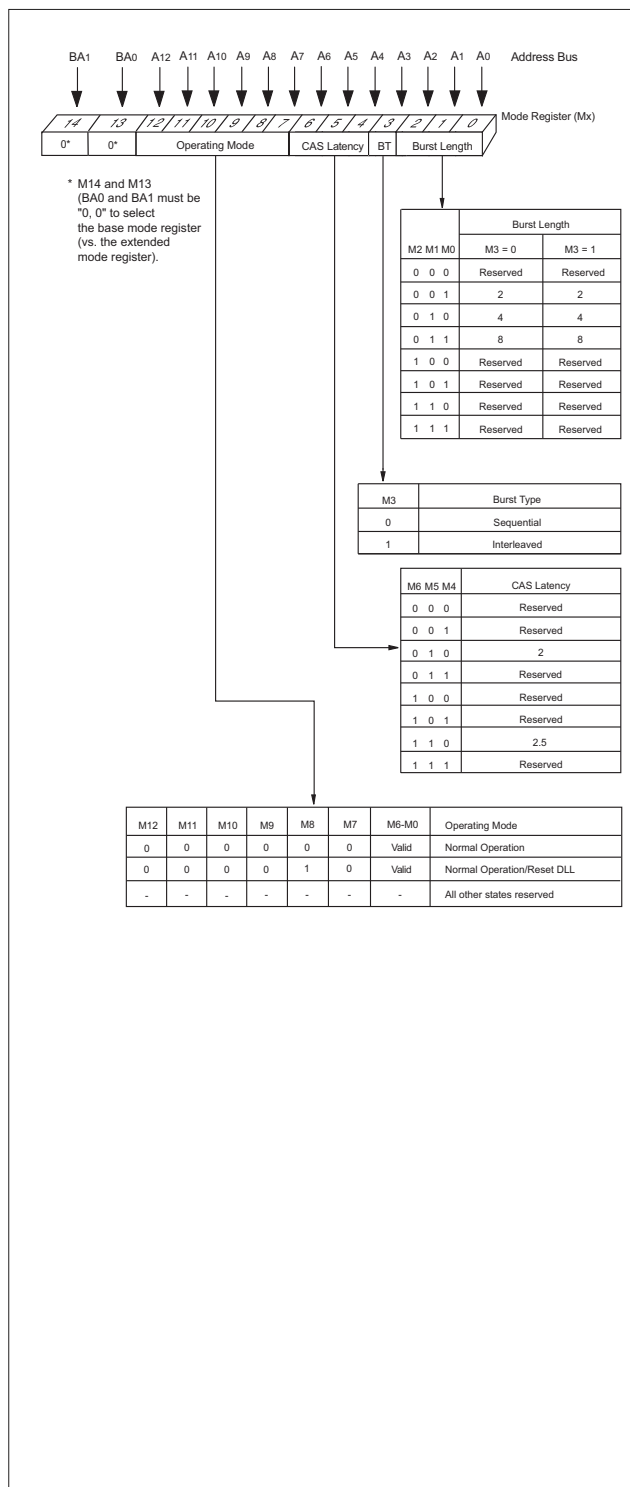
All other combinations of values for A7-A12 are reserved for future use and/or test modes. Test modes and reserved states should not be used because unknown operation or incompatibility with future versions may result.

## EXTENDED MODE REGISTER

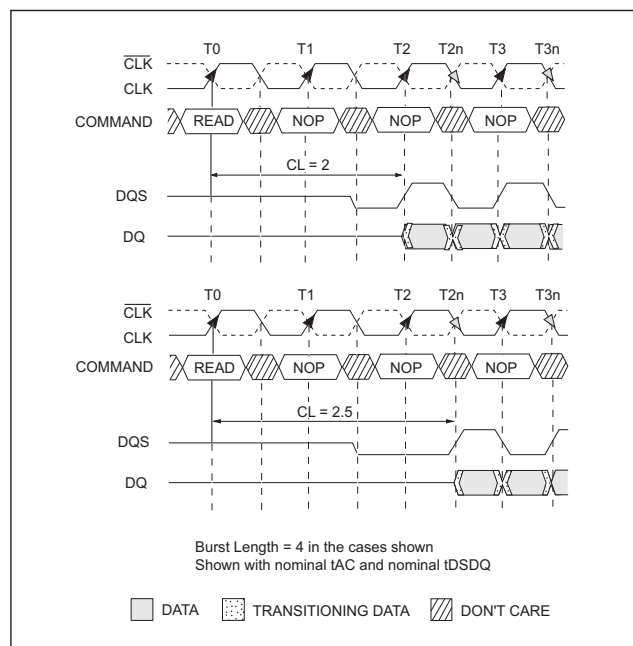
The extended mode register controls functions beyond those controlled by the mode register; these additional functions are DLL enable/disable, output drive strength, and QFC#. These functions are controlled via the bits shown in Figure 3. The extended mode register is programmed via the LOAD MODE REGISTER command to the mode register (with BA0 = 1 and BA1 = 0) and will retain the stored information until it is programmed again or the device loses power. The enabling of the DLL should always be followed by a LOAD MODE REGISTER command to the mode register (BA0/BA1 both LOW) to reset the DLL.

The extended mode register must be loaded when all banks are idle and no bursts are in progress, and the controller must wait the specified time before initiating any subsequent operation. Violating either of these requirements could result in unspecified operation.

**FIGURE 1 - MODE REGISTER DEFINITION**  
**FIGURE 3 - MODE REGISTER DEFINITION**



## FIGURE 2 - CAS LATENCY



## OUTPUT DRIVE STRENGTH

The normal full drive strength for all outputs are specified to be SSTL2, Class II. The DDR SDRAM supports an option for reduced drive. This option is intended for the support of the lighter load and/or point-to-point environments. The selection of the reduced drive strength will alter the DQs and DQSs from SSTL2, Class II drive strength to a reduced drive strength, which is approximately 54 percent of the SSTL2, Class II drive strength.

## DLL ENABLE/DISABLE

The DLL must be enabled for normal operation. DLL enable is required during power-up initialization and upon returning to normal operation after having disabled the DLL for the purpose of debug or evaluation. (When the device exits self refresh mode, the DLL is enabled automatically.) Any time the DLL is enabled, 200 clock cycles must occur before a READ command can be issued.

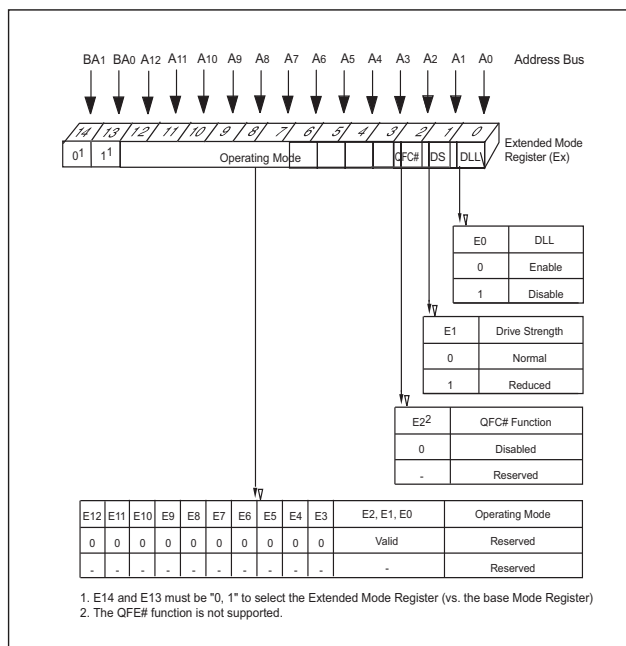
## COMMANDS

The Truth Table provides a quick reference of available commands. This is followed by a written description of each command.

## DESELECT

The DSELECT function (CS# HiGH) prevents new commands from being executed by the DDR SDRAM. The SDRAM is effectively deselected. Operations already in progress are not affected.

### FIGURE 3 - EXTENDED MODE REGISTER DEFINITION



## NO OPERATION (NOP)

The NO OPERATION (NOP) command is used to perform a NOP to the selected DDR SDRAM (CS# is LOW). This prevents unwanted commands from being registered during idle or wait states. Operations already in progress are not affected.

## LOAD MODE REGISTER

The Mode Registers are loaded via inputs A0-12. The LOAD MODE REGISTER command can only be issued when all banks are idle, and a subsequent executable command cannot be issued until tMRD is met.

## ACTIVE

The **ACTIVE** command is used to open (or activate) a row in a particular bank for a subsequent access. The value on the BA0, BA1 inputs selects the bank, and the address provided on inputs A0-12 selects the row. This row remains active (or open) for accesses until a **PRECHARGE** command is issued to that bank. A **PRECHARGE** command must be issued before opening a different row in the same bank.



### TRUTH TABLE - COMMANDS (NOTE 1)

NAME (FUNCTION)	CS#	RAS#	CAS#	WE#	ADDR
DESELECT (NOP) (9)	H	X	X	X	X
NO OPERATION (NOP) (9)	L	H	H	H	X
ACTIVE (Select bank and activate row) (3)	L	L	H	H	Bank/Row
READ (Select bank and column, and start READ burst) (4)	L	H	L	H	Bank/Col
WRITE (Select bank and column, and start WRITE burst) (4)	L	H	L	L	Bank/Col
BURST TERMINATE (8)	L	H	H	L	X
PRECHARGE (Deactivate row in bank or banks) (5)	L	L	H	L	Code
AUTO REFRESH or SELF REFRESH (Enter self refresh mode) (6,7)	L	L	L	H	X
LOAD MODE REGISTER (2)	L	L	L	L	Op-Code

### TRUTH TABLE - DM OPERATION

NAME (FUNCTION)	DM	DQs
WRITE ENABLE (10)	L	Valid
WRITE INHIBIT (10)	H	X

#### NOTES:

1. CKE is HIGH for all commands shown except SELF REFRESH.
2. A0-12 define the op-code to be written to the selected Mode Register. BA0, BA1 select either the mode register (0, 0) or the extended mode register (1, 0).
3. A0-12 provide row address, and BA0, BA1 provide bank address.
4. A0-8 provide column address; A10 HIGH enables the auto precharge feature (non-persistent), while A10 LOW disables the auto precharge feature; BA0, BA1 provide bank address.
5. A10 LOW: BA0, BA1 determine the bank being precharged. A10 HIGH: All banks precharged and BA0, BA1 are "Don't Care."

6. This command is AUTO REFRESH if CKE is HIGH; SELF REFRESH if CKE is LOW.
7. Internal refresh counter controls row addressing; all inputs and I/Os are "Don't Care" except for CKE.
8. Applies only to read bursts with auto precharge disabled; this command is undefined (and should not be used) for READ bursts with auto precharge enabled and for WRITE bursts.
9. Deselect and NOP are functionally interchangeable.
10. Used to mask write data; provided coincident with the corresponding data.

## READ

The READ command is used to initiate a burst read access to an active row. The value on the BA0, BA1 inputs selects the bank, and the address provided on inputs A0-8 selects the starting column location. The value on input A10 determines whether or not AUTO PRECHARGE is used. If AUTO PRECHARGE is selected, the row being accessed will be precharged at the end of the READ burst; if AUTO PRECHARGE is not selected, the row will remain open for subsequent accesses.

## WRITE

The WRITE command is used to initiate a burst write access to an active row. The value on the BA0, BA1 inputs selects the bank, and the address provided on inputs A0-8 selects the starting column location. The value on input A10 determines whether or not AUTO PRECHARGE is used. If AUTO PRECHARGE is selected, the row being accessed will be precharged at the end of the WRITE burst; if AUTO PRECHARGE is not selected, the row will remain open for subsequent accesses. Input data appearing on the D/Qs is written to the memory array subject to the DQM input logic level appearing coincident with the data. If a given DQM signal is registered LOW, the corresponding data will be written to memory; if the DQM signal is registered HIGH, the corresponding data inputs will be ignored, and a WRITE will not be executed to that byte/column location.

## PRECHARGE

The PRECHARGE command is used to deactivate the open row in a particular bank or the open row in all banks. The bank(s) will be available for a subsequent row access a specified time ( $t_{RP}$ ) after the PRECHARGE command is issued. Except in the case of concurrent auto precharge, where a READ or WRITE command to a different bank is allowed as long as it does not interrupt the data transfer in the current bank and does not violate any other timing parameters. Input A10 determines whether one or all banks are to be precharged, and in the case where only one bank is to be precharged, inputs BA0, BA1 select the bank. Otherwise BA0, BA1 are treated as "Don't Care." Once a bank has been precharged, it is in the idle state and must be activated prior to any READ or WRITE commands being issued to that bank. A PRECHARGE command will be treated as a NOP if there is no open row in that bank (idle state), or if the previously open row is already in the process of precharging.

## AUTO PRECHARGE

AUTO PRECHARGE is a feature which performs the same individual-bank PRECHARGE function described above, but without requiring an explicit command. This is accomplished by using A10 to enable AUTO PRECHARGE in conjunction with a specific READ or WRITE command. A precharge of the bank/row that is addressed with the READ or WRITE command is automatically performed upon completion of the READ or WRITE burst. AUTO PRECHARGE is nonpersistent in that it is either enabled or disabled for each individual READ or WRITE command. The device supports concurrent auto precharge if the command to the other bank does not interrupt the data transfer to the current bank.

AUTO PRECHARGE ensures that the precharge is initiated at the earliest valid stage within a burst. This “earliest valid stage” is determined as if an explicit precharge command was issued at the earliest possible time, without violating  $t_{RAS}$  (MIN). The user must not issue another command to the same bank until the precharge time ( $t_{RP}$ ) is completed. This is determined as if an explicit PRECHARGE command was issued at the earliest possible time, without violating  $t_{RAS}$  (MIN).

## BURST TERMINATE

The BURST TERMINATE command is used to truncate READ bursts (with auto precharge disabled). The most recently registered READ command prior to the BURST TERMINATE command will be truncated. The open page which the READ burst was terminated from remains open.

## AUTO REFRESH

AUTO REFRESH is used during normal operation of the DDR SDRAM and is analogous to CAS#-BEFORE-RAS# (CBR) REFRESH in conventional DRAMs. This command is nonpersistent, so it must be issued each time a refresh is required.

The addressing is generated by the internal refresh controller. This makes the address bits “Don’t Care” during an AUTO REFRESH command. Each DDR SDRAM requires AUTO REFRESH cycles at an average interval of  $7.8125\mu s$  (maximum).

To allow for improved efficiency in scheduling and switching between tasks, some flexibility in the absolute refresh interval is provided. A maximum of eight AUTO REFRESH commands can be posted to any given DDR SDRAM, meaning that the maximum absolute interval between any AUTO REFRESH command and the next AUTO REFRESH command is  $9 \times 7.8125\mu s$  ( $70.3\mu s$ ). This maximum absolute interval is to allow future support for DLL updates internal to the DDR SDRAM to be restricted to AUTO REFRESH cycles, without allowing excessive drift in  $t_{AC}$  between updates.

Although not a JEDEC requirement, to provide for future functionality features, CKE must be active (High) during the AUTO REFRESH period. The AUTO REFRESH period begins when the AUTO REFRESH command is registered and ends  $t_{RFC}$  later.

## SELF REFRESH\*

The SELF REFRESH command can be used to retain data in the DDR SDRAM, even if the rest of the system is powered down. When in the self refresh mode, the DDR SDRAM retains data without external clocking. The SELF REFRESH command is initiated like an AUTO REFRESH command except CKE is disabled (LOW). The DLL is automatically disabled upon entering SELF REFRESH and is automatically enabled upon exiting SELF REFRESH (200 clock cycles must then occur before a READ command can be issued). Input signals except CKE are “Don’t Care” during SELF REFRESH.

The procedure for exiting self refresh requires a sequence of commands. First, CLK must be stable prior to CKE going back HIGH. Once CKE is HIGH, the DDR SDRAM must have NOP commands issued for  $t_{XSNR}$ , because time is required for the completion of any internal refresh in progress.

A simple algorithm for meeting both refresh and DLL requirements is to apply NOPs for 200 clock cycles before applying any other command.

\* Self refresh available in commercial and industrial temperatures only.

**ABSOLUTE MAXIMUM RATINGS**

Parameter		Unit
Voltage on $V_{CC}$ , $V_{CCQ}$ Supply relative to $V_{SS}$	-1 ot 3.6	V
Voltage on I/O pins relative to $V_{SS}$	-1 ot 3.6	V
Operating Temperature $T_A$ (Mil)	-55 to +125	°C
Operating Temperature $T_A$ (Ind)	-40 to +85	°C
Storage Temperature, Plastic	-55 to +150	°C

Note: Stress greater than those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions greater than those indicated in the operational conditions for extended periods may affect reliability.

**CAPACITANCE (NOTE 13)**

Parameter	Symbol	Max	Unit
Input Capacitance: CLK	$C_{11}$	8	pF
Addresses, BA0-1 Input Capacitance	CA	30	pF
Input Capacitance: All other input-only pins	$C_{12}$	9	pF
Input/Output Capacitance: I/O's	$C_{10}$	12	pF

## DC ELECTRICAL CHARACTERISTICS AND OPERATING CONDITIONS (NOTES 1,6)

 $V_{CC}, V_{CCQ} = +2.5V \pm 0.2V; -55^{\circ}C \pm 0.2V, -55^{\circ}C \leq T_A \leq +125^{\circ}C$ 

Parameter / Condition	Symbol	Min	Max	Units
Supply Voltage	$V_{CC}$	2.3	2.7	V
I/O Supply Voltage	$V_{CCQ}$	2.3	2.7	V
Input Leakage Current: Any input $0V \leq V_{IN} \leq V_{CC}$ (All other pins not under test = 0V)	$I_I$	-2	2	$\mu A$
Input Leakage Address Current (All other pins not under test = 0V)	$I_{I_A}$	-10	10	$\mu A$
Output Leakage Current: I/O's are disabled; $0V \leq V_{OUT} \leq V_{CC}$	$I_{OZ}$	-5	5	$\mu A$
Output Levels: Full drive option				
High Current ( $V_{OUT} = V_{CCQ} - 0.373V$ , minimum $V_{REF}$ , minimum $V_{TT}$ )	$I_{OH}$	-12	-	mA
Low Current ( $V_{OUT} = 0.373V$ , maximum $V_{REF}$ , maximum $V_{TT}$ )	$I_{OL}$	12	-	mA
Output Levels: Reduced drive option				
High Current ( $V_{OUT} = V_{CCQ} - 0.763V$ , minimum $V_{REF}$ , minimum $V_{TT}$ )	$I_{OHR}$	-9	-	mA
Low Current ( $V_{OUT} = 0.763V$ , maximum $V_{REF}$ , maximum $V_{TT}$ )	$I_{OLR}$	9	-	mA
I/O Reference Voltage (6)	$V_{REF}$	$0.49 \times V_{CCQ}$	$0.51 \times V_{CCQ}$	V
I/O Termination Voltage (53)	$V_{TT}$	$V_{REF} - 0.04$	$V_{REF} + 0.04$	V

## AC INPUT OPERATING CONDITIONS (NOTES 1,6)14, 28, 40

 $V_{CC}, V_{CCQ} = +2.5V \pm 0.2V; -55^{\circ}C \pm 0.2V, -55^{\circ}C \leq T_A \leq +125^{\circ}C$ 

Parameter / Condition	Symbol	Min	Max	Units
Input High (Logic 1) Voltage:	$V_{IH} (AC)$	$V_{REF} + 0.310$	-	V
Input Low (Logic 0) Voltage:	$V_{IL} (AC)$	-	$V_{REF} - 0.310$	V

## $I_{CC}$ SPECIFICATIONS AND CONDITIONS (NOTES 1-5, 10, 12, 14)

 $V_{CC}, V_{CCQ} = +2.5V \pm 0.2V; -55^{\circ}C \pm 0.2V, -55^{\circ}C \leq T_A \leq +125^{\circ}C$ 

Parameter / Condition	Symbol	Max			
		333, 266 250 Mbps	200 Mbps	Units	
OPERATING CURRENT: One bank, Active-Precharge; $t_{RC} = t_{RC} \text{ (MIN)}$ ; $t_{CK} = t_{CK} \text{ (MIN)}$ ; DQ, DM, and DQS inputs changing once per clock cycle; Address and control inputs changing once every two clock cycles; (22, 48)	$I_{CC0}$	625	600	mA	
OPERATING CURRENT: One bank, Active-Read- Precharge; Burst=2; $t_{RC} = t_{RC} \text{ (MIN)}$ ; $t_{CK} = t_{CK} \text{ (MIN)}$ ; $I_{OUT} = 0\text{mA}$ ; Address and control inputs changing once every two clock cycles; (22, 48)	$I_{CC1}$	850	775	mA	
PRECHARGE POWER-DOWN STANDBY CURRENT: All banks idle; Power-down mode; $t_{CK}=t_{CK} \text{ (MIN)}$ ; CKE=LOW; (23, 32, 50)	$I_{CC2P}$	20	20	mA	
IDLE STANDBY CURRENT: CS#=HIGH, All banks idle; $t_{CK}=t_{CK} \text{ (MIN)}$ ; CKE=HIGH; Address and other control inputs changing once per clock cycle. $V_{IN}=V_{REF}$ for DQ, DQS and DM (51)	$I_{CC2F}$	225	225	mA	
ACTIVE POWER-DOWN STANDBY CURRENT: One bank active; Power-down mode; $t_{CK}=t_{CK} \text{ (MIN)}$ ; CKE=LOW (23, 32, 50)	$I_{CC3P}$	150	150	mA	
ACTIVE STANDBY CURRENT: CS#=HIGH; CKE=HIGH; One bank; Active Precharge; $t_{RC}=t_{RAS} \text{ (MAX)}$ ; $t_{CK}=t_{CK} \text{ (MIN)}$ ; DQ, DM and DQS inputs changin twice per clock cycle; Address and other control inputs changing once per clock cycle (22)	$I_{CC3N}$	250	250	mA	
OPERATING CURRENT: Burst=2; Reads; Continuous burst; One bank active; Address and control inputs changing once per clock cycle; $t_{CK}=t_{CK} \text{ (MIN)}$ ; $I_{OUT}=0\text{mA}$ (22,48)	$I_{CC4R}$	925	925	mA	
OPERATING CURRENT: Burst=2; Writes; Continuous burst; One bank active; Address and control inputs changing once per clock cycle; $t_{CK}=t_{CK} \text{ (MIN)}$ ; DQ, DM and DQS inputs changin twice per clock cycle (22)	$I_{CC4W}$	800	800	mA	
AUTO REFRESH CURRENT	$t_{REF}=t_{RC} \text{ (MIN)}$ (27, 50)	$I_{CC5}$	1225	1225	mA
	$t_{REF}=7.8125 \mu\text{s}$ (27, 50)	$I_{CC5A}$	30	30	mA
SELF REFRESH CURRENT: CKE $\leq 0.2\text{V}$	Standard (11)	$I_{CC6}$	20	20	mA
OPERATING CURRENT: Four bank interleaving DEADs (BL=4) with auto precharge, $t_{RC}=t_{RC} \text{ (MIN)}$ ; $t_{CK}=t_{CK} \text{ (MIN)}$ ; Addresses and control inputs change only during Active READ or WRITE commands. (22, 49)	$I_{CC7}$	2000	2000	mA	

**ELECTRICAL CHARACTERISTICS AND RECOMMENDED AC OPERATING CHARACTERISTICS  
(NOTES 1-5, 14-17, 33)**

Parameter		Symbol	-6, 333 [266] Mbps @CL=2.5 [CL=2]		-75, 266 [250] Mbps @CL=2.5 [CL=2]		Units
			Min	Max	Min	Max	
Access window of DQs from CLK/CLK#		t <sub>AC</sub>	-0.70	+0.70	-0.75	+0.75	ns
CLK high-level width (30)		t <sub>CH</sub>	0.45	0.55	0.45	0.55	t <sub>CK</sub>
CLK low-level width (30)		t <sub>CL</sub>	0.45	0.55	0.45	0.55	t <sub>CK</sub>
Clock cycle time	CL=2.5 (45, 52)	t <sub>CK</sub> (2.5)	6	13	7.5	13	ns
	CL=2 (45, 52)	t <sub>CK</sub> (2)	7.5	13	8	13	ns
DQ and DM input hold time relative to DQS (26, 31)		t <sub>DH</sub>	0.45		0.5		ns
DQ and DM input setup time relative to DQS (26,31)		t <sub>DS</sub>	0.45		0.5		ns
DQ and DM input pulse width (for each input) (31)		t <sub>DIPW</sub>	1.75		1.75		ns
Access window of DQS from CLK/CLK#		t <sub>DQSCK</sub>	-0.6	+0.6	-0.75	+0.75	ns
DQS input high pulse width		t <sub>DQSH</sub>	0.35		0.35		t <sub>CK</sub>
DQS input high pulse width		t <sub>DQSL</sub>	0.35		0.35		t <sub>CK</sub>
DQS-DQ skew, DQS to last valid, per group, per access (25,26)		t <sub>DQSQ</sub>		0.45		0.5	ns
Write command to first DQS latching transition		t <sub>DQSS</sub>	0.75	1.25	0.75	1.25	t <sub>CK</sub>
DQS falling edge to CLK rising - setup time		t <sub>DSS</sub>	0.2		0.2		t <sub>CK</sub>
DQS falling edge to CLK rising - hold time		t <sub>DSH</sub>	0.2		0.2		t <sub>CK</sub>
Half clock period (34)		t <sub>HP</sub>	t <sub>CH</sub> , t <sub>CL</sub>		t <sub>CH</sub> , t <sub>CL</sub>		ns
Data-out high-impedance window from CLK/CLK# (18, 42)		t <sub>HZ</sub>		+0.7		+0.75	ns
Data-out low-impedance window from CLK/CLK# (18, 43)		t <sub>LZ</sub>	-0.7		-0.75		ns
Address and control input hold time (fast slew rate) (14)		t <sub>IHF</sub>	0.75		0.9		ns
Address and control input setup time (fast slew rate) (14)		t <sub>ISF</sub>	0.75		0.9		ns
Address and control input hold time (slow slew rate) (14)		t <sub>IHS</sub>	0.8		1.0		ns
Address and control input setup time (slow slew rate) (14)		t <sub>ISS</sub>	0.8		1.0		ns
LOAD MODE REGISTER command cycle time		t <sub>MRD</sub>	12		15		ns
DQ-DQS hold, DQS to first DQ to go non-valid, per access (25, 26)		t <sub>QH</sub>	t <sub>HP</sub> -t <sub>QHS</sub>		t <sub>HP</sub> -t <sub>QHS</sub>		ns
Data hold skew factor		t <sub>QHS</sub>		0.55		1	ns
ACTIVE to PRECHARGE command (35)		t <sub>RAS</sub>	42	70,000	40	120,000	ns
ACTIVE to READ with Auto precharge command		t <sub>RAP</sub>	15		15		ns
ACTIVE to ACTIVE/AUTO REFRESH command period		t <sub>RC</sub>	60		60		ns
AUTO REFRESH command period (50)		t <sub>RFC</sub>	72		75		ns
ACTIVE to READ or WRITE delay		t <sub>RCD</sub>	15		15		ns
PRECHARGE command period		t <sub>RP</sub>	15		15		ns
DQS read preamble (42)		t <sub>RPRE</sub>	0.9	1.1	0.9	1.1	t <sub>CK</sub>
DQS read postamble		t <sub>RPST</sub>	0.4	0.6	0.4	0.6	t <sub>CK</sub>
Active bank a to ACTIVE bank b command		t <sub>RRD</sub>	12		15		ns
DQS write preamble		t <sub>WPRE</sub>	0.25		0.25		t <sub>CK</sub>
DQS write preamble setup time (20,21)		t <sub>WPRES</sub>	0		0		ns
DQS write postamble (19)		t <sub>WPST</sub>	0.4	0.6	0.4	0.6	t <sub>CK</sub>
Write recovery time		t <sub>WR</sub>	15		15		ns
Internal WRITE to READ command delay		t <sub>WTR</sub>	1		1		t <sub>CK</sub>
Data valid output window (25)		NA	t <sub>QH</sub> - t <sub>DQSQ</sub>		t <sub>QH</sub> - t <sub>DQSQ</sub>		ns
REFRESH to REFRESH command interval (Commercial & Industrial temp		t <sub>REFC</sub>		70.3		70.3	μs
REFRESH to REFRESH command interval (Military temp only) (23)		t <sub>REFC</sub>		35		35	μs
Average periodic refresh interval (Commercial & Industrial temp only) (23)		t <sub>REFI</sub>		7.8		7.8	μs
Average periodic refresh interval (Military temp only) (23)		t <sub>REFI</sub>		3.9		3.9	μs
Terminating voltage delay to Vcc (53)		t <sub>VTD</sub>	0		0		ns
Exit SELF REFRESH to non-READ command		t <sub>XSNR</sub>	75		75		ns
Exit SELF REFRESH to READ command		t <sub>XSRD</sub>	200		200		t <sub>CK</sub>

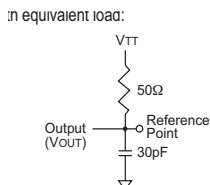
**ELECTRICAL CHARACTERISTICS AND RECOMMENDED AC OPERATING CHARACTERISTICS**  
**(NOTES 1-5, 14-17, 33)**

Parameter	Symbol	-8, 250 [200] Mbps @CL=2.5 [CL=2]		-10, 200 [167] Mbps @CL=2.5 [CL=2]		Units
		Min	Max	Min	Max	
Access window of DQs from CLK/CLK#	$t_{AC}$	-0.8	+0.8	-0.8	+0.8	ns
CLK high-level width (30)	$t_{CH}$	0.45	0.55	0.45	0.55	$t_{CK}$
CLK low-level width (30)	$t_{CL}$	0.45	0.55	0.45	0.55	$t_{CK}$
Clock cycle time	CL=2.5 (45, 52) $t_{CK} (2.5)$	8	13	10	13	ns
	CL=2 (45, 52) $t_{CK} (2)$	10	13	13	15	ns
DQ and DM input hold time relative to DQS (26, 31)	$t_{DH}$	0.6		0.6		ns
DQ and DM input setup time relative to DQS (26,31)	$t_{DS}$	0.6		0.6		ns
DQ and DM input pulse width (for each input) (31)	$t_{DIPW}$	2		2		ns
Access window of DQS from CLK/CLK#	$t_{DQSCK}$	-0.8	+0.8	-0.8	+0.8	ns
DQS input high pulse width	$t_{DQSH}$	0.35		0.35		$t_{CK}$
DQS input high pulse width	$t_{DQSL}$	0.35		0.35		$t_{CK}$
DQS-DQ skew, DQS to last valid, per group, per access (25,26)	$t_{DQSQ}$		0.6		0.6	ns
Write command to first DQS latching transition	$t_{DQSS}$	0.75	1.25	0.75	1.25	$t_{CK}$
DQS falling edge to CLK rising - setup time	$t_{DSS}$	0.2		0.2		$t_{CK}$
DQS falling edge to CLK rising - hold time	$t_{DSH}$	0.2		0.2		$t_{CK}$
Half clock period (34)	$t_{HP}$	$t_{CH}, t_{CL}$		$t_{CH}, t_{CL}$		ns
Data-out high-impedance window from CLK/CLK# (18, 42)	$t_{HZ}$		+0.8		+0.8	ns
Data-out low-impedance window from CLK/CLK# (18, 43)	$t_{LZ}$	-0.8		-0.8		ns
Address and control input hold time (fast slew rate) (14)	$t_{IHF}$	1.1		1.1		ns
Address and control input setup time (fast slew rate) (14)	$t_{ISF}$	1.1		1.1		ns
Address and control input hold time (slow slew rate) (14)	$t_{IHS}$	1.1		1.1		ns
Address and control input setup time (slow slew rate) (14)	$t_{ISS}$	1.1		1.1		ns
LOAD MODE REGISTER command cycle time	$t_{MRD}$	16		16		ns
DQ-DQS hold, DQS to first DQ to go non-valid, per access (25, 26)	$t_{QH}$	$t_{HP}-t_{QHS}$		$t_{HP}-t_{QHS}$		ns
Data hold skew factor	$t_{QHS}$		1		1	ns
ACTIVE to PRECHARGE command (35)	$t_{RAS}$	40	120,000	40	120,000	ns
ACTIVE to READ with Auto precharge command	$t_{RAP}$	20		20		ns
ACTIVE to ACTIVE/AUTO REFRESH command period	$t_{RC}$	70		70		ns
AUTO REFRESH command period (50)	$t_{RFC}$	80		80		ns
ACTIVE to READ or WRITE delay	$t_{RCD}$	20		20		ns
PRECHARGE command period	$t_{RP}$	20		20		ns
DQS read preamble (42)	$t_{RPRE}$	0.9	1.1	0.9	1.1	$t_{CK}$
DQS read postamble	$t_{RPST}$	0.4	0.6	0.4	0.6	$t_{CK}$
Active bank a to ACTIVE bank b command	$t_{RRD}$	15		15		ns
DQS write preamble	$t_{WPRES}$	0.25		0.25		$t_{CK}$
DQS write preamble setup time (20,21)	$t_{WPRES}$	0		0		ns
DQS write postamble (19)	$t_{WPST}$	0.4	0.6	0.4	0.6	$t_{CK}$
Write recovery time	$t_{WR}$	15		15		ns
Internal WRITE to READ command delay	$t_{WTR}$	1		1		$t_{CK}$
Data valid output window (25)	NA	$t_{QH}-t_{DQSQ}$		$t_{QH}-t_{DQSQ}$		ns
REFRESH to REFRESH command interval (Commercial & Industrial temp)	$t_{REFC}$		70.3		70.3	$\mu s$
REFRESH to REFRESH command interval (Military temp only) (23)	$t_{REFC}$		35		35	$\mu s$
Average periodic refresh interval (Commercial & Industrial temp only) (23)	$t_{REFI}$		7.8		7.8	$\mu s$
Average periodic refresh interval (Military temp only) (23)	$t_{REFI}$		3.9		3.9	$\mu s$
Terminating voltage delay to Vcc (53)	$t_{VTD}$	0		0		ns
Exit SELF REFRESH to non-READ command	$t_{XSNR}$	80		80		ns
Exit SELF REFRESH to READ command	$t_{XSRD}$	200		200		$t_{CK}$



## NOTES:

1. All voltages referenced to  $V_{SS}$ .
2. Tests for AC timing,  $I_{CC}$ , and electrical AC and DC characteristics may be conducted at nominal reference/supply voltage levels, but the related specifications and device operation are guaranteed for the full voltage range specified.
3. Outputs measured with equivalent load:



4. AC timing and  $I_{CC}$  tests may use a  $V_{IL}$ -to- $V_{IH}$  swing of up to 1.5V in the test environment, but input timing is still referenced to  $V_{REF}$  (or to the crossing point for CLK/CLK#), and parameter specifications are guaranteed for the specified AC input levels under normal use conditions. The minimum slew rate for the input signals used to test the device is 1V/ns in the range between  $V_{IL}(AC)$  and  $V_{IH}(AC)$ .
5. The AC and DC input level specifications are as defined in the SSTL\_2 Standard (i.e., the receiver will effectively switch as a result of the signal crossing the AC input level, and will remain in that state as long as the signal does not ring back above [below] the DC input LOW [HIGH] level).
6.  $V_{REF}$  is expected to equal  $V_{CCQ}/2$  of the transmitting device and to track variations in the DC level of the same. Peak-to-peak noise (noncommon mode) on  $V_{REF}$  may not exceed  $\pm 2$  percent of the DC value. Thus, from  $V_{CCQ}/2$ ,  $V_{REF}$  is allowed  $\pm 25$ mV for DC error and an additional  $\pm 25$ mV for AC noise. This measurement is to be taken at the nearest  $V_{REF}$  by-pass capacitor.
7.  $V_{TT}$  is not applied directly to the device.  $V_{TT}$  is a system supply for signal termination resistors, is expected to be set equal to  $V_{REF}$  and must track variations in the DC level of  $V_{REF}$ .
8.  $V_{ID}$  is the magnitude of the difference between the input level on CLK and the input level on CLK#.
9. The value of  $V_{IX}$  and  $V_{MP}$  are expected to equal  $V_{CCQ}/2$  of the transmitting device and must track variations in the DC level of the same.
10.  $I_{CC}$  is dependent on output loading and cycle rates. Specified values are obtained with minimum cycle time with the outputs open.
11. Enables on-chip refresh and address counters.
12.  $I_{CC}$  specifications are tested after the device is properly initialized, and is averaged at the defined cycle rate.
13. This parameter is not tested but guaranteed by design.  $t_A = 25^\circ\text{C}$ ,  $f = 1$  MHz
14. Command/Address input slew rate = 0.5V/ns. For 266 MHz with slew rates 1V/ns and faster,  $t_{IS}$  and  $t_{IH}$  are reduced to 900ps. If the slew rate is less than 0.5V/ns, timing must be derated:  $t_{IS}$  has an additional 50ps per each 100mV/ns reduction in slew rate from the 500mV/ns.  $t_{IH}$  has 0ps added, that is, it remains constant. If the level, slew rate exceeds 4.5V/ns, functionality is uncertain.

15. The CLK/CLK# input reference level (for timing referenced to CLK/CLK#) is the point at which CLK and CLK# cross; the input reference level for signals other than CLK/CLK# is  $V_{REF}$ .
16. Inputs are not recognized as valid until  $V_{REF}$  stabilizes. Exception: during the period before  $V_{REF}$  stabilizes,  $CKE < 0.3 \times V_{CCQ}$  is recognized as LOW.
17. The output timing reference level, as measured at the timing reference point indicated in Note 3, is  $V_{TT}$ .
18.  $t_{HZ}$  and  $t_{LZ}$  transitions occur in the same access time windows as valid data transitions. These parameters are not referenced to a specific voltage level, but specify when the device output is no longer driving (HZ) or begins driving (LZ).
19. The maximum limit for this parameter is not a device limit. The device will operate with a greater value for this parameter, but system performance (bus turnaround) will degrade accordingly.
20. This is not a device limit. The device will operate with a negative value, but system performance could be degraded due to bus turnaround.
21. It is recommended that DQS be valid (HIGH or LOW) on or before the WRITE command. The case shown (DQS going from High-Z to logic LOW) applies when no WRITES were previously in progress on the bus. If a previous WRITE was in progress, DQS could be HIGH during this time, depending on  $t_{DQSS}$ .
22. MIN ( $t_{RC}$  or  $t_{RFC}$ ) for  $I_{CC}$  measurements is the smallest multiple of  $t_{CK}$  that meets the minimum absolute value for the respective parameter.  $t_{RAS}$  (MAX) for  $I_{CC}$  measurements is the largest multiple of  $t_{CK}$  that meets the maximum absolute value for  $t_{RAS}$ .
23. The refresh period 64ms. This equates to an average refresh rate of However, an AUTO REFRESH command must be asserted at least once every 70.3 $\mu$ s; burst refreshing or posting by the DRAM controller greater than eight refresh cycles is not allowed.
24. The I/O capacitance per DQS and DQ byte/group will not differ by more than this maximum amount for any given device.
25. The valid data window is derived by achieving other specifications -  $t_{HP}$  ( $t_{CK}/2$ ),  $t_{DQSQ}$ , and  $t_{QH}$  ( $t_{QH} = t_{HP} - t_{QHS}$ ). The data valid window derates directly proportional with the clock duty cycle and a practical data valid window can be derived. The clock is allowed a maximum duty cycle variation of 45/55. Functionality is uncertain when operating beyond a 45/55 ratio. The data valid window derating curves are provided
26. Referenced to each output group: LDQS with DQ0-DQ7; and UDQS with DQ8-DQ15 of each chip.
27. This limit is actually a nominal value and does not result in a fail value.  $CKE$  is HIGH below for duty cycles ranging between 50/50 and 45/55. during REFRESH command period ( $t_{RFC}$  [MIN]) else  $CKE$  is LOW (i.e., during standby)
28. To maintain a valid level, the transitioning edge of the input must:
  - a) Sustain a constant slew rate from the current AC level through to the target AC level,  $V_{IL}(AC)$  or  $V_{IH}(AC)$ .
  - b) Reach at least the target AC level.
  - c) After the AC target level is reached, continue to maintain at least the target DC level,  $V_{IL}(DC)$  or  $V_{IH}(DC)$ .

FIGURE A - PULL-DOWN CHARACTERISTICS

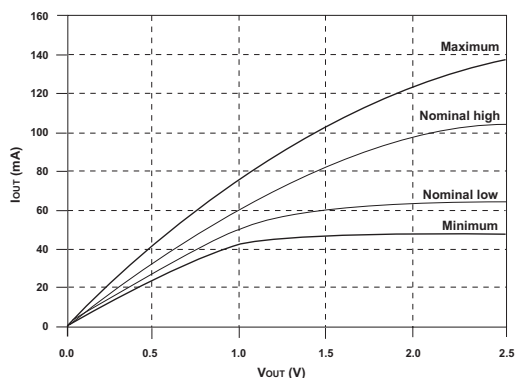
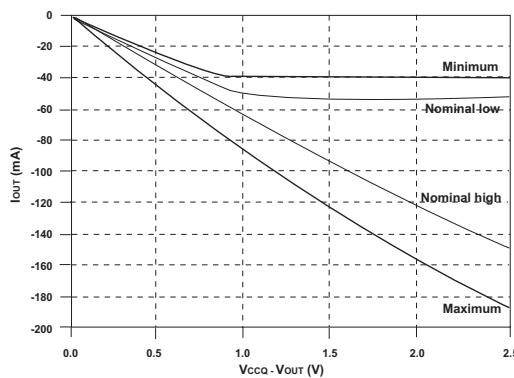


FIGURE B - PULL-UP CHARACTERISTICS



29. The Input capacitance per pin group will not differ by more than this maximum amount for any given device.
30. CLK and CLK# input slew rate must be  $> 1\text{V/ns}$  ( $> 2\text{V/ns}$  differentially).
31. DQ and DM input slew rates must not deviate from DQS by more than 10%. If the DQ/DM/DQS slew rate is less than  $0.5\text{V/ns}$ , timing must be derated: 50ps must be added to  $t_{DS}$  and  $t_{DH}$  for each  $100\text{mV/ns}$  reduction in slew rate. If slew rate exceeds  $4\text{V/ns}$ , functionality is uncertain.
32. Vcc must not vary more than 4% if CKE is not active while any bank is active.
33. The clock is allowed up to  $\pm 150\text{ps}$  of jitter. Each timing parameter is allowed to vary by the same amount.
34.  $t_{HP}$  min is the lesser of  $t_{CL}$  minimum and  $t_{CH}$  minimum actually applied to the device CLK and CLK# inputs, collectively during bank active.
35. READs and WRITEs with auto precharge are not allowed to be issued until  $t_{RAS}(\text{MIN})$  can be satisfied prior to the internal precharge command being issued.
36. Any positive glitch must be less than 1/3 of the clock and not more than +400mV or 2.9 volts, whichever is less. Any negative glitch must be less than 1/3 of the clock cycle and not exceed either -300mV or 2.2 volts, whichever is more positive.
37. Normal Output Drive Curves:
  - a) The full variation in driver pull-down current from minimum to maximum process, temperature and voltage will lie within the outer bounding lines of the V-I curve of Figure A.
  - b) The variation in driver pull-down current within nominal limits of voltage and temperature is expected, but not guaranteed, to lie within the inner bounding lines of the V-I curve of Figure A.
  - c) The full variation in driver pull-up current from minimum to maximum process, temperature and voltage will lie within the outer bounding lines of the V-I curve of Figure B.
  - d) The variation in driver pull-up current within nominal limits of voltage and temperature is expected, but not guaranteed, to lie within the inner bounding lines of the V-I curve of Figure B.
  - e) The full variation in the ratio of the maximum to minimum pull-up and pull-down current should be between .71 and 1.4, for device drain-to-source voltages from 0.1V to 1.0 Volt, and at the same voltage and temperature.
  - f) The full variation in the ratio of the nominal pull-up to pull-down current should be unity  $\pm 10\%$ , for device drain-to-source voltages from 0.1V to 1.0 Volt.
38. Reduced Output Drive Curves:
  - a) The full variation in driver pull-down current from minimum to maximum process, temperature and voltage will lie within the outer bounding lines of the V-I curve of Figure C.
  - b) The variation in driver pull-down current within nominal limits of voltage and temperature is expected, but not guaranteed, to lie within the inner bounding lines of the V-I curve of Figure C.
  - c) The full variation in driver pull-up current from minimum to maximum process, temperature and voltage will lie within the outer bounding lines of the V-I curve of Figure D.
39. The voltage levels used are derived from a minimum Vcc level and the referenced test load. In practice, the voltage levels obtained from a properly terminated bus will provide significantly different voltage values.
40.  $V_{IH}$  overshoot:  $V_{IH}(\text{MAX}) = V_{CCQ} + 1.5\text{V}$  for a pulse width  $\leq 3\text{ns}$  and the pulse width can not be greater than 1/3 of the cycle rate.
41. Vcc and Vcco must track each other.
42. This maximum value is derived from the referenced test load. In practice, the values obtained in a typical terminated design may reflect up to 310ps less for  $t_{HZ}(\text{MAX})$  and the last DVW.  $t_{HZ}(\text{MAX})$  will prevail over  $t_{DSCCK}(\text{MAX}) + t_{RPST}(\text{MAX})$  condition.  $t_{LZ}(\text{MIN})$  will prevail over  $t_{DSCCK}(\text{MIN}) + t_{RPST}(\text{MAX})$  condition.
43. For slew rates greater than  $1\text{V/ns}$  the (LZ) transition will start about 310ps earlier.
44. During initialization, Vcco, VTT, and VREF must be equal to or less than Vcc + 0.3V. Alternatively, VTT may be 1.35V maximum during power up, even if Vcc/Vcco are 0 volts, provided a minimum of 42 ohms of series resistance is used between the VTT supply and the input pin.
45. The current part operates below the slowest JEDEC operating frequency of 83 MHz. As such, future die may not reflect this option.
46. Reserved for future use.
47. Reserved for future use.
48. Random addressing changing 50% of data changing at every transfer.
49. Random addressing changing 100% of data changing at every transfer.
50. CKE must be active (high) during the entire time a refresh command is executed. That is, from the time the AUTO REFRESH command is registered, CKE must be active at each rising clock edge, until  $t_{RFC}$  has been satisfied.
51. Icc2N specifies the DQ, DQS, and DM to be driven to a valid high or low logic level. Icc2O is similar to Icc2F except Icc2O specifies the address and control inputs to remain stable. Although Icc2F, Icc2N, and Icc2O are similar, Icc2F is "worst case."
52. Whenever the operating frequency is altered, not including jitter, the DLL is required to be reset. This is followed by 200 clock cycles before any READ command.
53. VTT is not applied directly to the device; however,  $t_{VTD}$  should be greater than or equal to zero to avoid device latch-up. Vcco, VTT and VREF must be equal to or less than Vcc + 0.3V. Alternatively VTT may be 1.35V max during power-up even if Vcc/Vcco are 0V, provided a minimum of 42  $\Omega$  of series resistance is used between the VTT supply and the input pin. Once initialized, VREF must always be powered within the specified range.

FIGURE C - PULL-DOWN CHARACTERISTICS

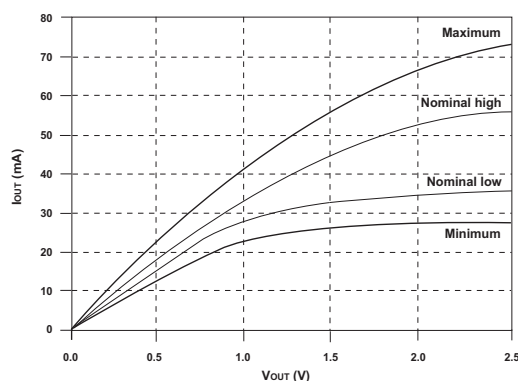
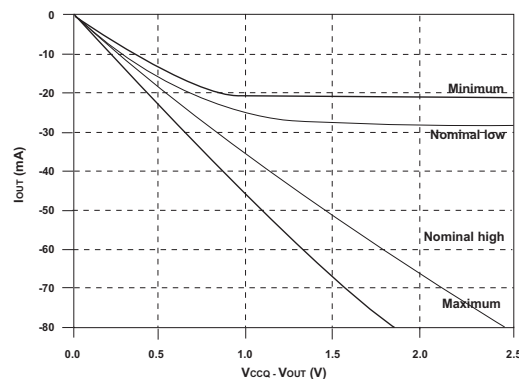
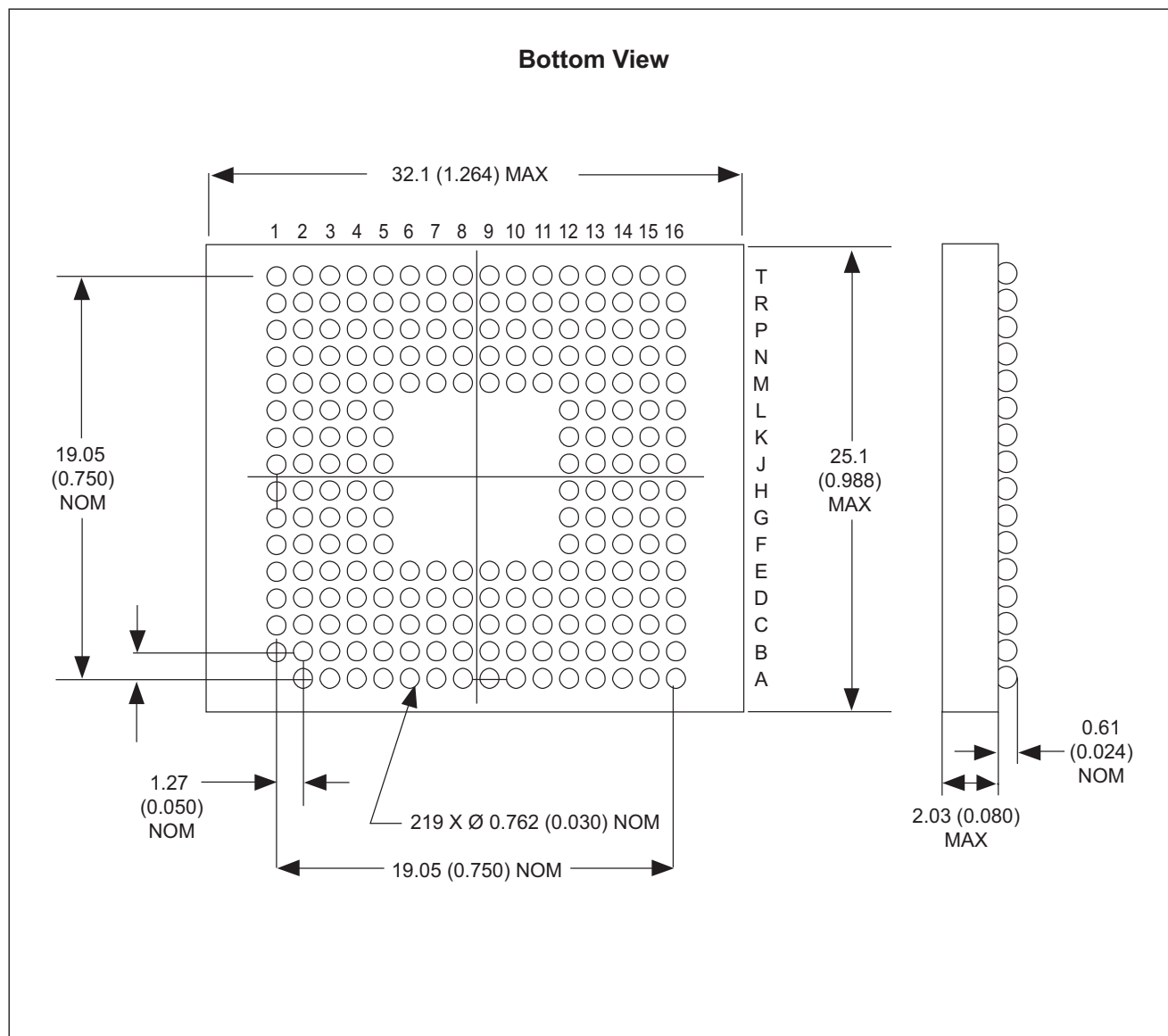


FIGURE D - PULL-UP CHARACTERISTICS



**PACKAGE DIMENSION: 219 PLASTIC BALL GRID ARRAY (PBGA)**



ALL LINEAR DIMENSIONS ARE MILLIMETERS AND PARENTHETICALLY IN INCHES

## ORDERING INFORMATION

Part Number	Core Freq.	Data Transfer Rate	Package	Process
AS4DDR16M72-6/IT	166 MHz	333 Mbps	219-PBGA	Industrial
AS4DDR16M72-75/IT	133 MHz	266 Mbps	219-PBGA	Industrial
AS4DDR16M72-8/IT	125 MHz	250 Mbps	219-PBGA	Industrial
AS4DDR16M72-10/IT	100 MHz	200 Mbps	219-PBGA	Industrial
AS4DDR16M72-75/ET	133 MHz	266 Mbps	219-PBGA	Enhanced
AS4DDR16M72-8/ET	125 MHz	250 Mbps	219-PBGA	Enhanced
AS4DDR16M72-10/ET	100 MHz	200 Mbps	219-PBGA	Enhanced
AS4DDR16M72-75/XT	133 MHz	266 Mbps	219-PBGA	Military
AS4DDR16M72-8/XT	125 MHz	250 Mbps	219-PBGA	Military
AS4DDR16M72-10/XT	100 MHz	200 Mbps	219-PBGA	Military

**DOCUMENT TITLE**

16M X 72 DDR SDRAM Multi-Chip Package

**REVISION HISTORY**

<b><u>Rev #</u></b>	<b><u>History</u></b>	<b><u>Release Date</u></b>	<b><u>Status</u></b>
0.3	Pre-Release	March 2005	Advanced
1.0	RELEASE	December 2005	Advanced
2.0	RELEASE	June 2006	Preliminary
2.1	Updated Order Chart "Extended" to "Military" Temp	June 2009	Release
2.2	Updated Micross information	January 2010	Release