

Advance Information

This document contains information on a product under development. The parametric information contains target parameters that are subject to change.

M02046

3.3/5V Limiting Amplifier for Applications to 1.25 Gbps

The M02046 is a highly integrated high-gain limiting amplifier that can be used with the same board layout and footprint as the MC2046-2. Featuring PECL outputs, the M02046 is intended for use in applications to 1.25 Gbps. Full output swing is achieved even at minimum input sensitivity. The M02046 can operate with a 3.3V or 5V supply.

Included in the M02046 is a programmable signal-level detector, allowing the user to set thresholds at which the logic outputs are enabled. The signal detect function has typically 2 dB (optical) of hysteresis which prevents chatter at low input levels. A squelch function, which turns off the output when no signal is present, is provided by externally connecting the LOS Status output to the JAM input.

The M02046 has a PECL Status output and the M02046C has a CMOS Status output. Both versions have a CMOS LOS output.

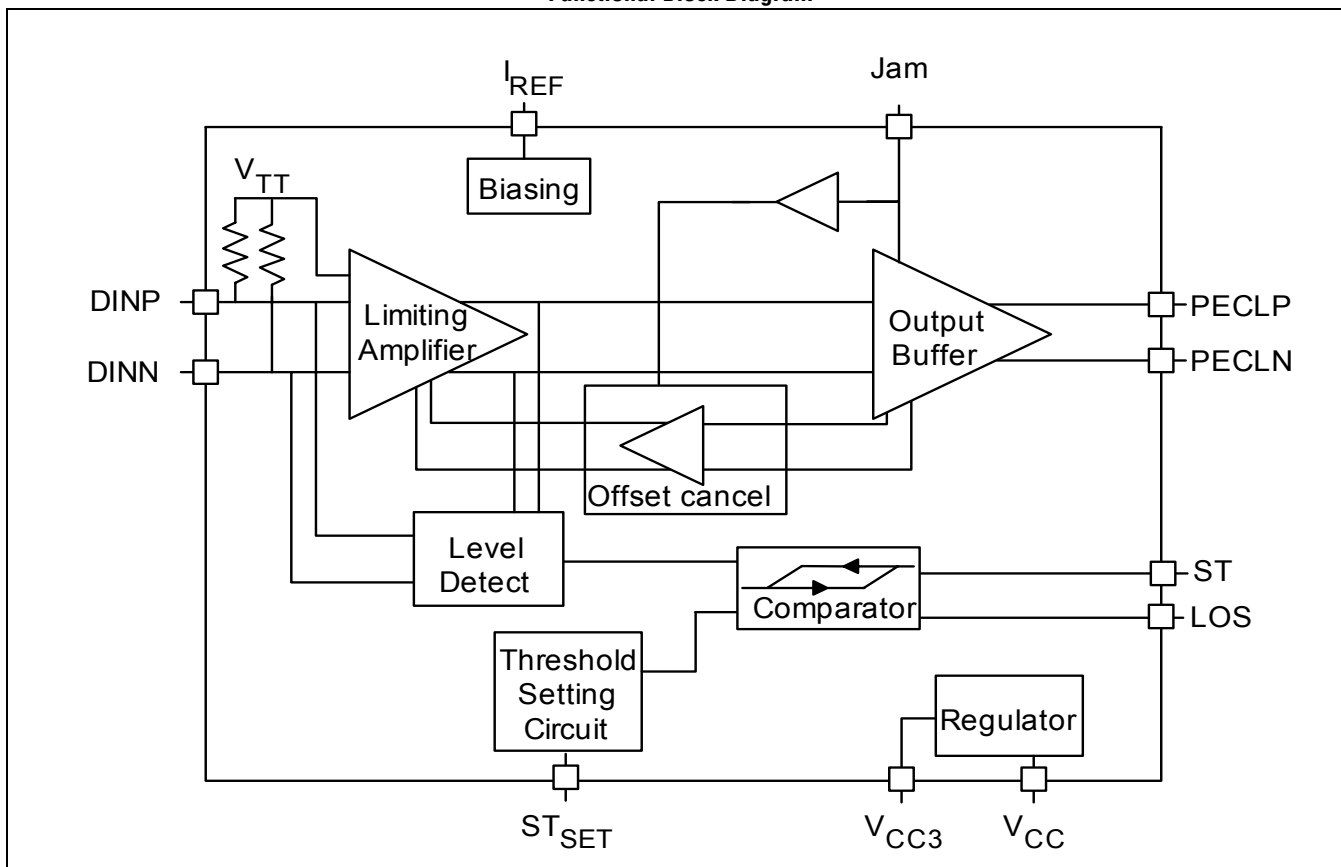
Applications

- 1.06 Gbps Fibre Channel
- 1.25 Gbps Ethernet
- 1.25 Gbps SDH/SONET

Features

- Pin compatible with the MC2046-2
- Operates with a 3.3 or 5V supply
- 2 mV input sensitivity at 1.25 Gbps
- Programmable input-signal level detect
- On-chip DC offset cancellation circuit
- CMOS and PECL Signal Detect output variants
- Output Jam Function
- Low power (< 200 mW (M02046C) at 3.3V)

Functional Block Diagram



Ordering Information

Part Number	Package	Operating Temperature
M02046-xx	PECL Status Output (CMOS LOS Output) in QSOP16 package	-40 °C to 85 °C
M02046C-xx	CMOS Status Output (CMOS LOS Output) in QSOP16 package	-40 °C to 85 °C
M02046-PEVM	Evaluation board with M02046 (PECL Status Output)	-40 °C to 85 °C
M02046-CEVM	Evaluation board with M02046C (CMOS Status Output)	-40 °C to 85 °C

Note: xx represents the revision number. Please contact your local sales office for correct digits.

Revision History

Revision	Level	Date	ASIC Revision	Description
A	Advanced	February 2004	-12	Initial Release
B	Advanced	May 2004	-12	Remove references to RSSI function

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1.0 Functional Description

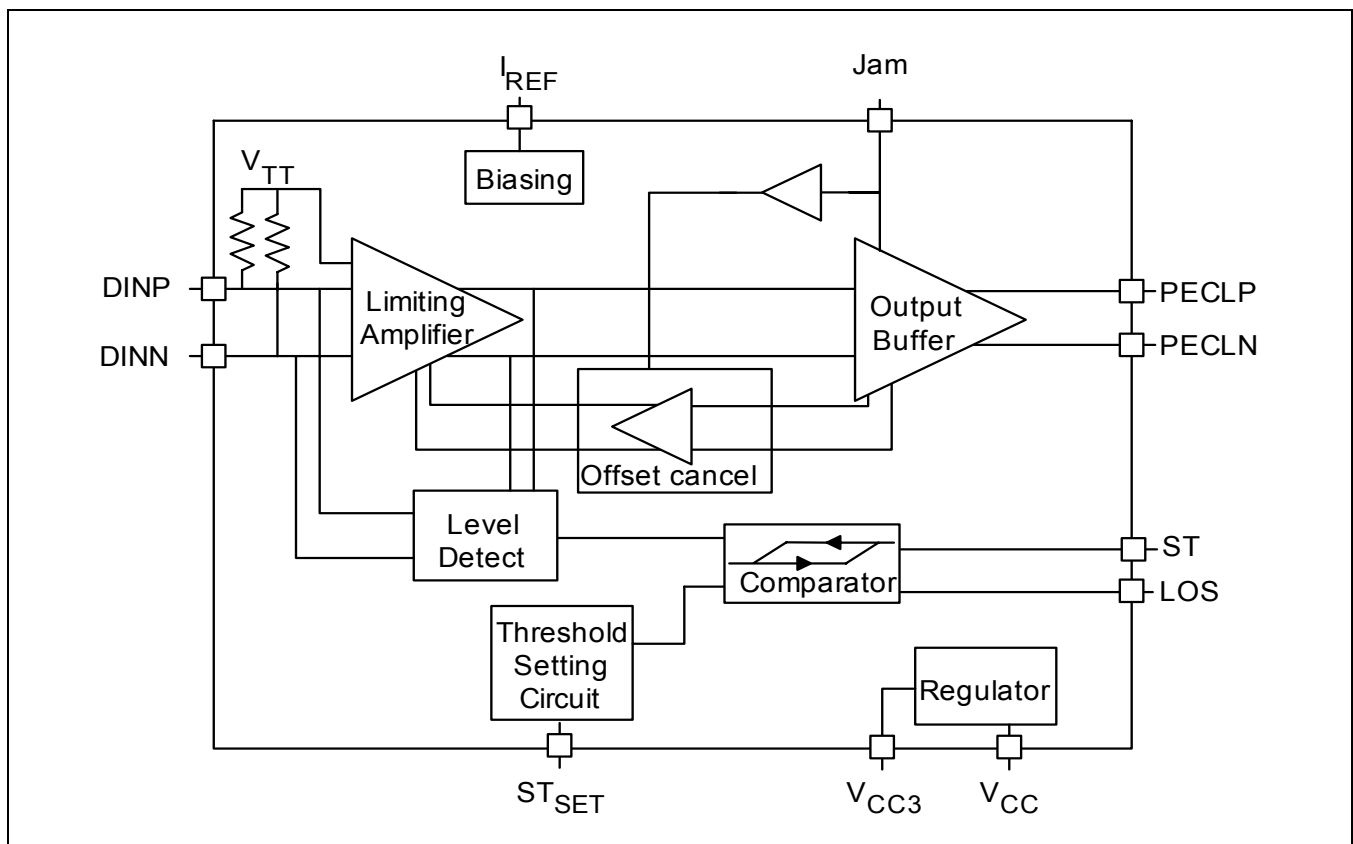
1.1 Overview

The M02046 is a highly integrated high-gain limiting amplifier that can be used with the same board layout and footprint as the MC2046-2. Featuring PECL outputs, the M02046 is intended for use in applications to 1.25 Gbps. Full output swing is achieved even at minimum input sensitivity. The M02046 can operate with a 3.3V or 5V supply.

Included in the M02046 is a programmable signal-level detector, allowing the user to set thresholds at which the logic outputs are enabled. The signal detect function has typically 2 dB (optical) of hysteresis which prevents chatter at low input levels. A squelch function, which turns off the output when no signal is present, is provided by externally connecting the LOS Status output to the JAM input.

The M02046 has a PECL Status output and the M02046C has a CMOS Status output. Both versions have a CMOS LOS output.

Figure 1-1. Block Diagram Example



1.2 Features

- Pin compatible with the MC2046-2
- Operates with a 3.3 or 5V supply
- 2 mV input sensitivity at 1.25 Gbps
- Programmable input-signal level detect
- On-chip DC offset cancellation circuit
- CMOS and PECL Signal Detect output variants
- Output Jam Function
- Low power (< 200 mW (M02046C) at 3.3V)

1.3 General Description

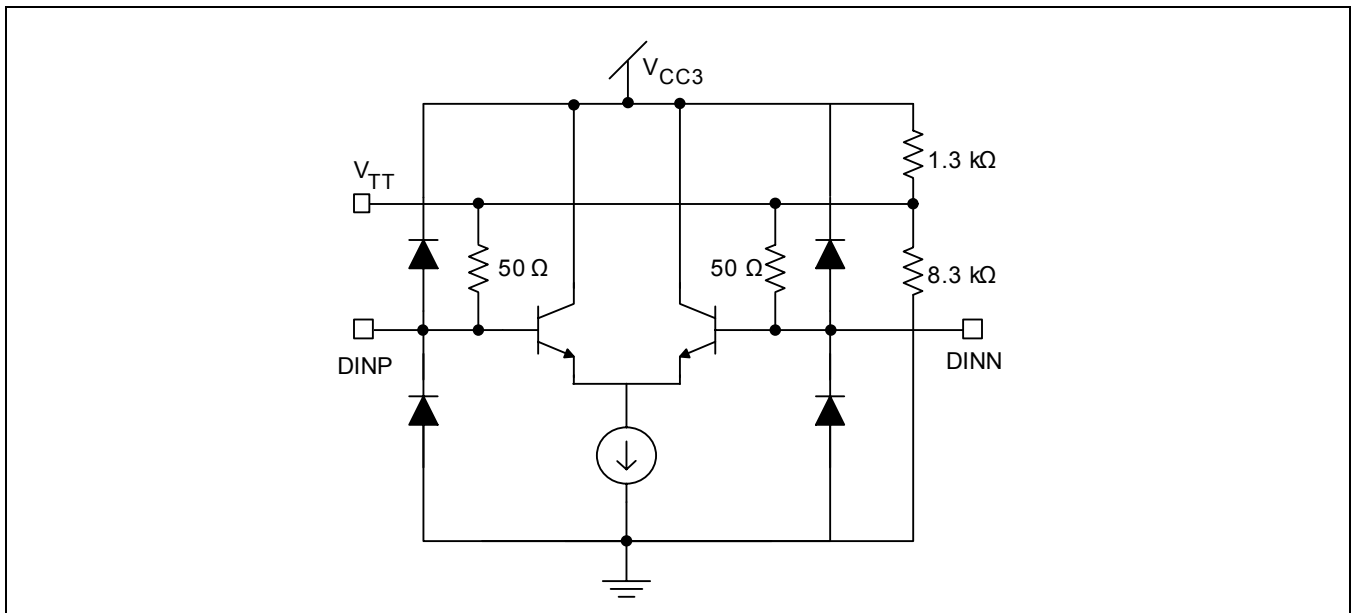
The M02046 is a high-gain limiting amplifier for applications up to 1.25 Gbps, and incorporates a limiting amplifier, an input signal level detection circuit and also a fully integrated DC-offset cancellation loop that does not require any external components. The M02046 features a PECL high-speed data outputs.

The M02046 provides the user with the flexibility to set the signal detect threshold and features either a PECL status output (M02046) or a CMOS status output (M02046C). Optional output buffer disable (squelch/jam) can be implemented using the JAM input.

1.3.1 Inputs

The data inputs are internally connected to V_{TT} via 50 Ω resistors, and generally need to be AC coupled. Referring to Figure 1-2, the nominal V_{TT} voltage is 2.85V because of the internal resistor divider to V_{CC3} , which means this is the DC potential on the data inputs. See the applications information section for further details on choosing the AC-coupling capacitor.

Figure 1-2. CML Data Inputs



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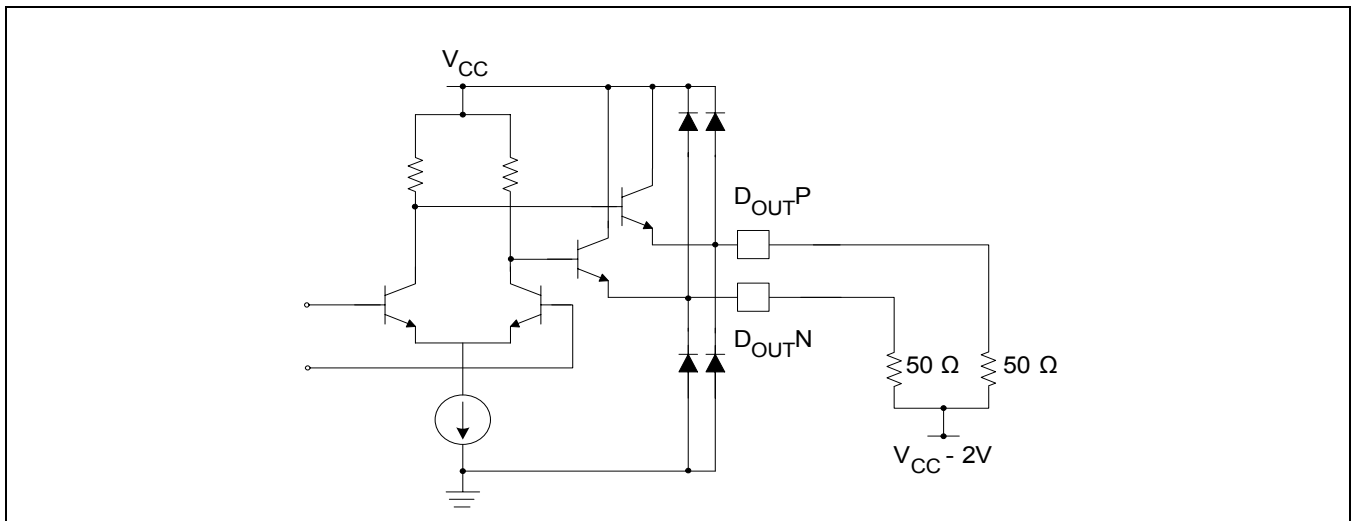
1.3.2 DC Offset Compensation

The M02046 contain an internal DC autozero circuit that can remove the effect of DC offsets without using external components. This circuit is configured such that the feedback is effective only at frequencies well below the lowest frequency of interest. The low frequency cut off is typically 25 kHz.

1.3.3 Data Outputs

The M02046 features 100k/300k PECL compliant outputs as shown in [Figure 1-3](#). The outputs may be terminated using any standard AC or DC-coupling PECL termination technique. AC-coupling is used in applications where the average DC content of the data is zero e.g. SONET. The advantage of this approach is lower power consumption, no susceptibility to DC drive and compatibility with non-PECL interfaces.

Figure 1-3. PECL Data Outputs



1.3.4 Signal Detect (ST) and Loss of Signal (LOS)

The M02046 features input signal level detection over an extended range. Using an external resistor, R_{ST} , between pin ST_{SET} and V_{CC3} ([Figure 1-6](#)) the user can program the input signal threshold. The signal detect status is indicated on the both the Signal Detect (ST) and LOS output pins. The Status output is either PECL (M02046) or CMOS (M02046C). The PECL version is shown in [Figure 1-4](#) while [Figure 1-5](#) shows the ST output for the CMOS version of the device (and the LOS output for both versions of the device).

The ST (LOS) signal is active (not asserted) when the signal is above the threshold value. The signal detection circuitry has the equivalent of 4dB (typical) electrical hysteresis.

Figure 1-4. PECL ST Output (M02046)

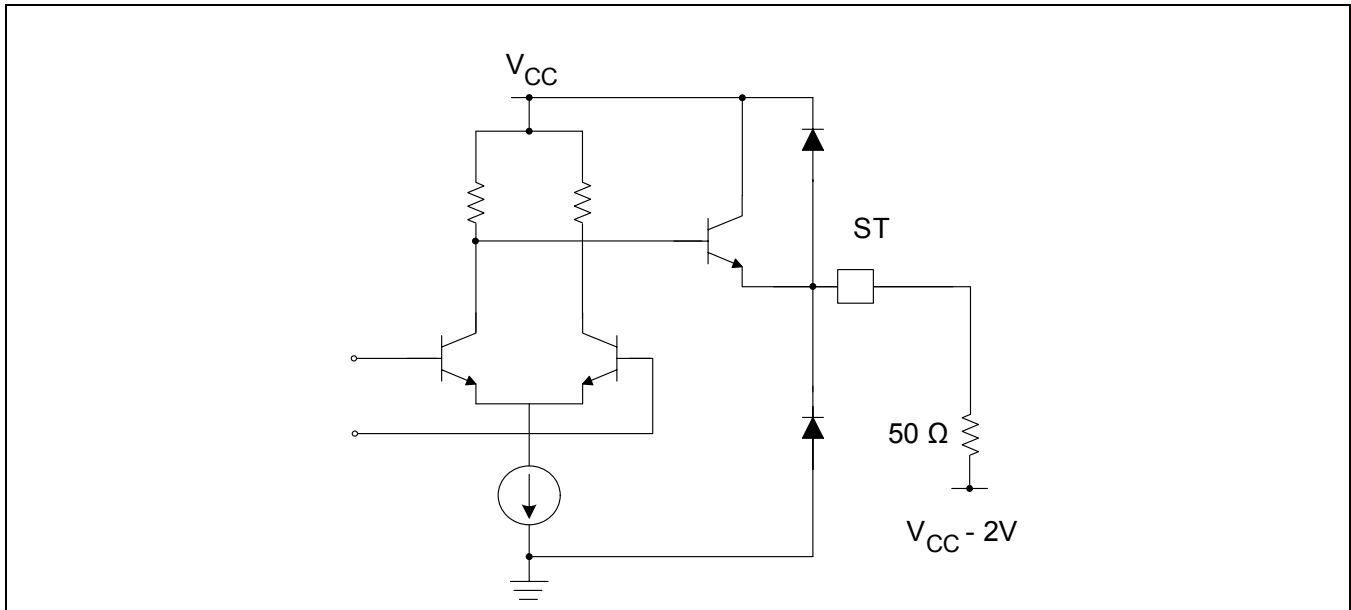
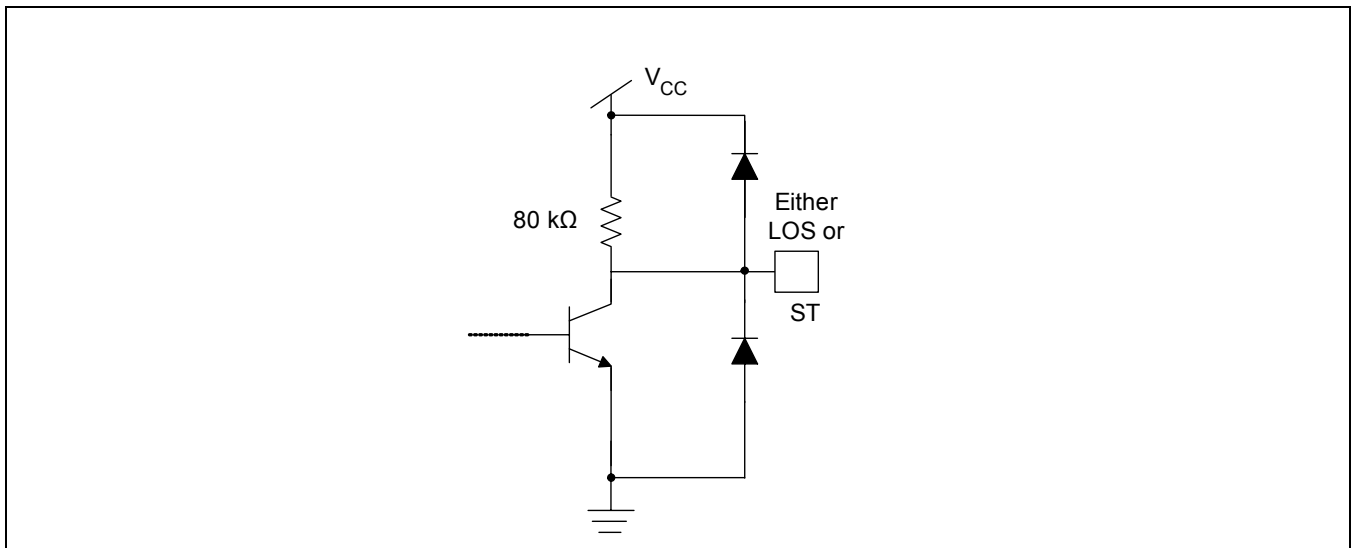


Figure 1-5. CMOS ST (M02046C) and LOS (M02046 and M02046C) Output

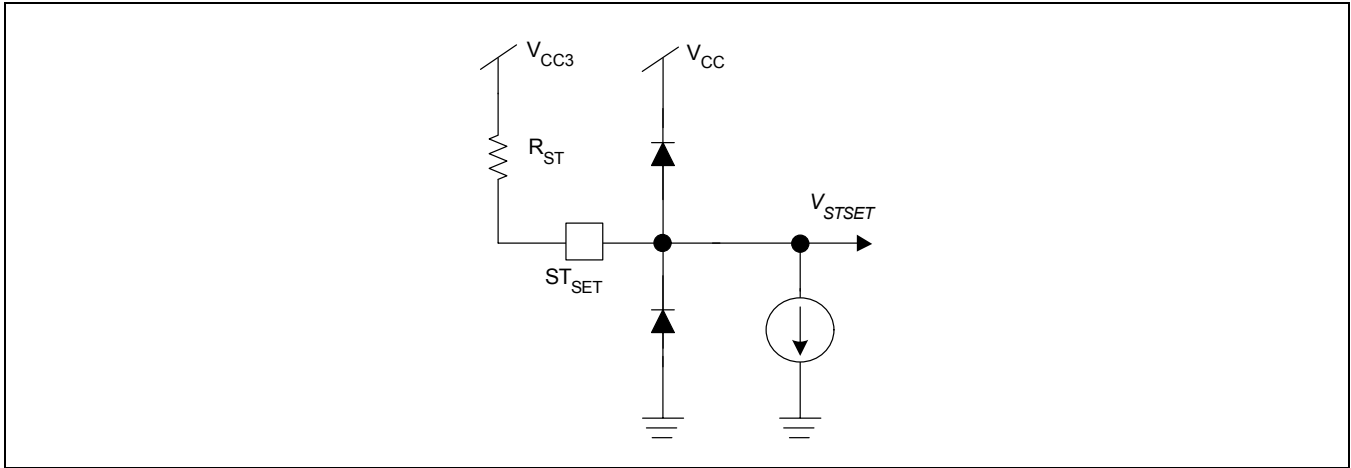


R_{ST} establishes a threshold voltage at the ST_{SET} pin as shown in Figure 1-6. Internally, the input signal level is monitored by the Level Detector which creates a DC voltage proportional to the input signal peak to peak value. The voltage at ST_{SET} is internally compared to the signal level from the Level Detector. When the Level Detect voltage is less than $V_{(STSET)}$, LOS is asserted and will stay asserted until the input signal level increases by a pre-defined amount of hysteresis. When the input level increases by more than this hysteresis above $V_{(STSET)}$, LOS is deasserted. See the applications information section for the selection of R_{ST} .

Note that ST_{SET} can be left open if the loss of signal detector function is not required. In this case LOS would be low.

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Figure 1-6. STset Input

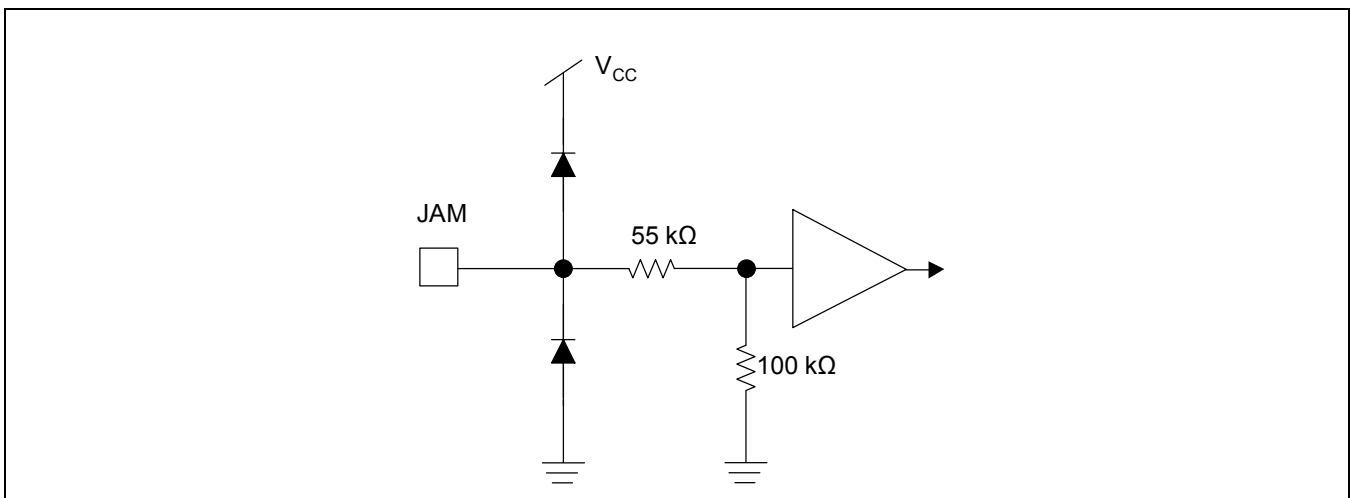


1.3.5 JAM Function

When asserted, the active high power down (JAM) pin forces the outputs to a logic “one” state. This ensures that no data is propagated through the system. The loss of signal detection circuit can be used to automatically force the data outputs to a high state when the input signal falls below the threshold. The function is normally used to allow data to propagate only when the signal is above the user’s bit-error-rate requirement. It therefore inhibits the data outputs toggling due to noise when there is no signal present (“squelch”).

In order to implement this function, LOS should be connected to the JAM pin shown in Figure 1-7, thus forcing the data outputs to a logic “one” state when the signal falls below the threshold.

Figure 1-7. JAM Input



1.3.6 Voltage Regulation

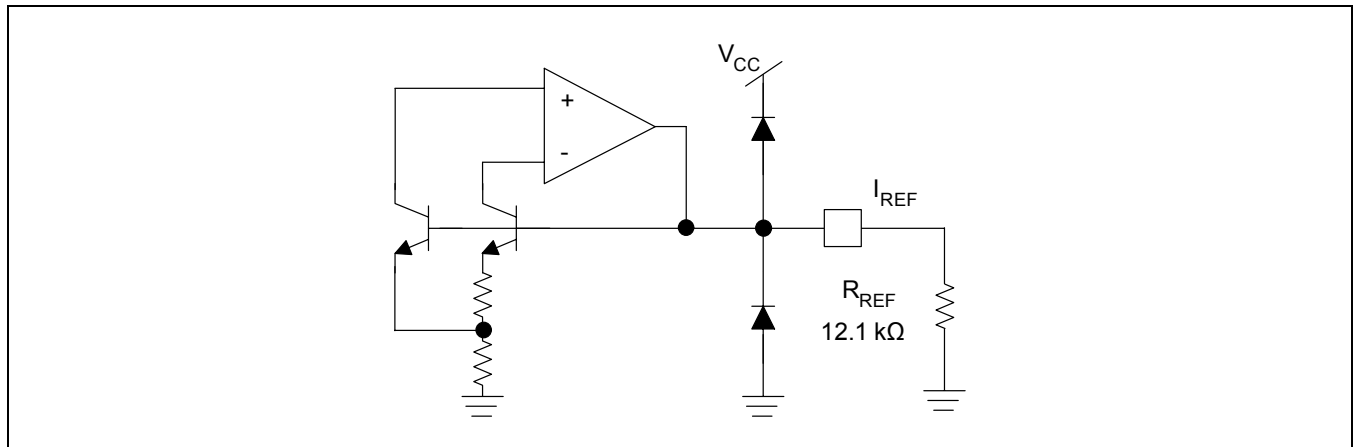
The M02046 contain an on-chip voltage regulator to allow both 5V and 3.3V operation. When used at 5V, the on-chip regulator is enabled and the digital inputs and outputs are compatible with TTL 5V logic levels.

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1.3.7 Reference Current Generation

The M02046 contain an accurate on-chip bias circuit that requires an external 12.1 kΩ 1% resistor, R_{REF}, from pin I_{REF} to ground to define an on-chip reference current.

Figure 1-8. Reference Current Generation



1.3.8 Connecting V_{CC} and V_{CC3}

For 5V operation, the V_{CC} pin is connected to an appropriate 5V ± 7.5% supply. No potential should be applied to the V_{CC3} pin. The only connection to V_{CC3} should be R_{ST} as shown in Figure 1-6.

When V_{CC} = 5V all logic outputs and the data outputs are 5V compatible. For low power operation, V_{CC} and V_{CC3} should be connected to an appropriate 3.3V ± 7.5% supply. In this case all I/Os are 3.3V compatible.

1.3.9 Choosing an Input AC-Coupling Capacitor

When AC-coupling the input the coupling capacitor should be of sufficient value to pass the lowest frequencies of interest, bearing in mind the number of consecutive identical bits, and the input resistance of the part. For SONET data, a good rule of thumb is to chose a coupling capacitor that has a cut-off frequency less than 1/(10,000) of the input data rate. For example, for 1.25 Gbps data, the coupling capacitor should be chosen as:

$$f_{\text{CUTOFF}} \leq (1.25 \times 10^9 / 10 \times 10^3) = 125 \text{ kHz}$$

The -3 dB cutoff frequency of the low pass filter at the 50 Ω input is found as:

$$f_{3\text{dB}} = 1 / (2 * \pi * 50 \Omega * C_{\text{AC}})$$

so solving for C where $f_{3\text{dB}} = f_{\text{CUTOFF}}$

$$C_{\text{AC}} = 1 / (2 * \pi * 50 \Omega * f_{\text{CUTOFF}}) \tag{EQ.1}$$

and in this case the minimum capacitor is 25 nF.

For Ethernet or Fibre Channel, there are less consecutive bits in the data, and the recommended cut-off frequency is 1/(1,000) of the input data rate. This results in a minimum capacitor of 2.5 nF for 1.25 Gbps Ethernet.

Multirate applications down to 155 Mbps

In this case, the input coupling capacitor needs to be large enough to pass 15 kHz ($155 \times 10^6 / 10,000$) which results in a capacitor value of 0.2 μ F. However, because this low pass frequency is close to the 25 kHz low pass frequency of the internal DC servo loop, it is preferable to use a larger input coupling capacitor such as 1 μ F which provides an input cutoff frequency of 3.1 kHz. This separates the two poles sufficiently to allow them to be considered independent. This capacitor should also have a 20 nF capacitor in parallel to pass the higher frequency data (in the multirate application) without distortion.

In all cases, a high quality coupling capacitor should be used as to pass the high frequency content of the input data stream.

1.3.10 Setting the Signal Detect Level

Using [Figure 1-9](#), the value for R_{ST} is chosen to set the LOS threshold at the desired value. The resulting hysteresis is also shown in [Figure 1-9](#).

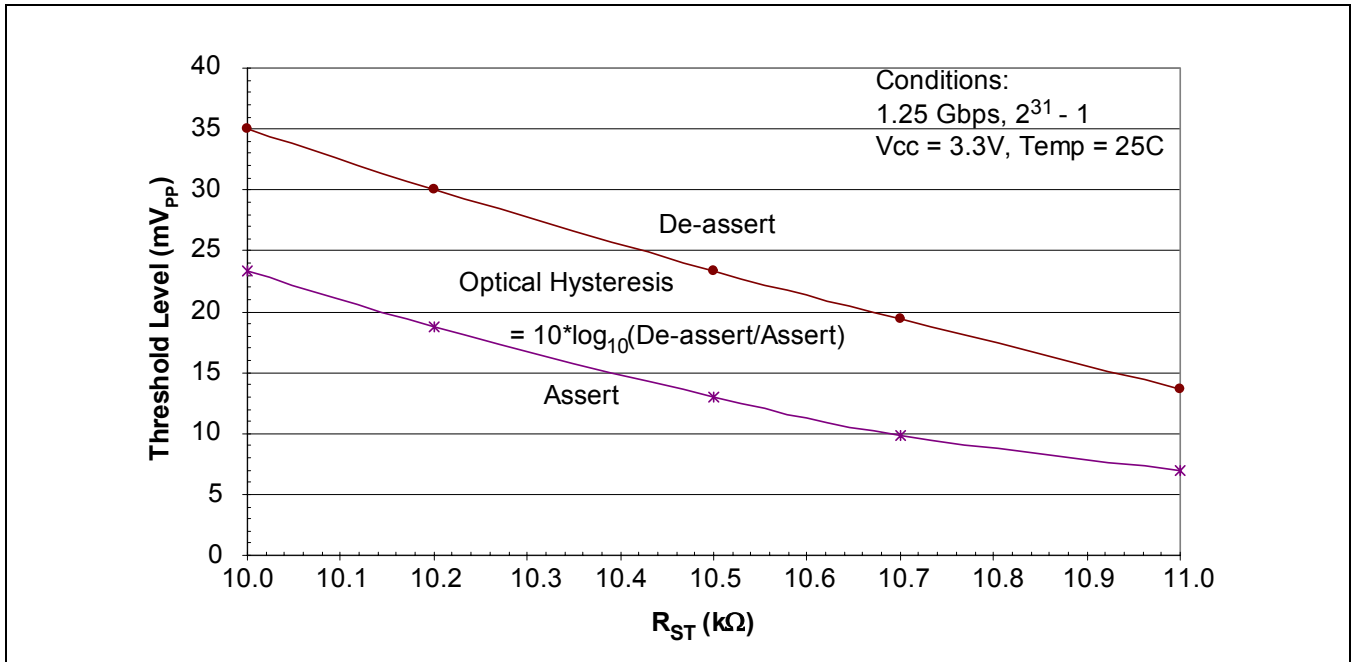
From [Figure 1-9](#), it is apparent that small variations in R_{LOS} cause significant variation in the LOS threshold level, particularly for low input signal levels. This is because of the logarithmic relationship between the internal level detect voltage and the input signal level. It is recommended that a 1% resistor be used for R_{ST} and that allowance is provided for LOS variation, particularly when the LOS threshold is near the sensitivity limit of the M02046.

Example R_{ST} resistor values are given in [Table 1-1](#).

Table 1-1. R_{ST} Resistor Values

VIN (mV pp) differential	R_{ST} (k Ω)
6.9	11.0
9.9	10.7
13.0	10.5
18.8	10.2
23.4	10.0

Figure 1-9. Loss of Signal Characteristic



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1.3.11 PECLP, PECLN, and ST_{PECL} (M02046) Termination

The data and ST (in the PECL version of the part) outputs of the M02046 are PECL compatible. For the high speed PECLP and PECLN outputs any standard AC or DC-coupling termination technique can be used. Figure 1-10 and Figure 1-11 illustrate typical AC and DC terminations.

AC-coupling is used in applications where the average DC content of the data is zero e.g. SONET. The advantage of this approach is lower power consumption, no susceptibility to DC drift and compatibility with non-PECL interfaces. Figure 1-10 shows the circuit configuration and Table 1-2 lists the resistor values. If using transmission lines other than 50 Ω, the shunt terminating resistance Z_T should equal twice the impedance of the transmission line (Z₀).

DC-coupling can be used when driving PECL interfaces and has the advantage of a reduced component count. A Thevenin termination is used at the receive end to give a 50 Ω load and the correct DC bias. Figure 1-11 shows the circuit configuration and Table 1-2 the resistor values.

Alternatively, if available, terminating to V_{CC} - 2V as shown in Figure 1-12 has the advantage that the resistance value is the same for 3.3 V and 5 V operation and it also has performance advantages at high data rates.

In the M02046, ST is a PECL output. It is recommended that it be DC terminated in a PECL load, preferably as shown in Figure 1-12, but the decoupling capacitor on the load is not required as this is a DC output.

Table 1-2. PECL Termination Resistor Values

Supply	Output Impedance	R _{PULL-DOWN}	Z _T	R _{TA} / R _{TB}	R _T / R _B
5 V	50 Ω	270 Ω	100 Ω	2.7 kΩ / 7.8 kΩ	82 Ω / 130 Ω
3.3 V	50 Ω	150 Ω	100 Ω	2.7 kΩ / 4.3 kΩ	130 Ω / 82 Ω

Figure 1-10. AC-Coupled PECL Termination

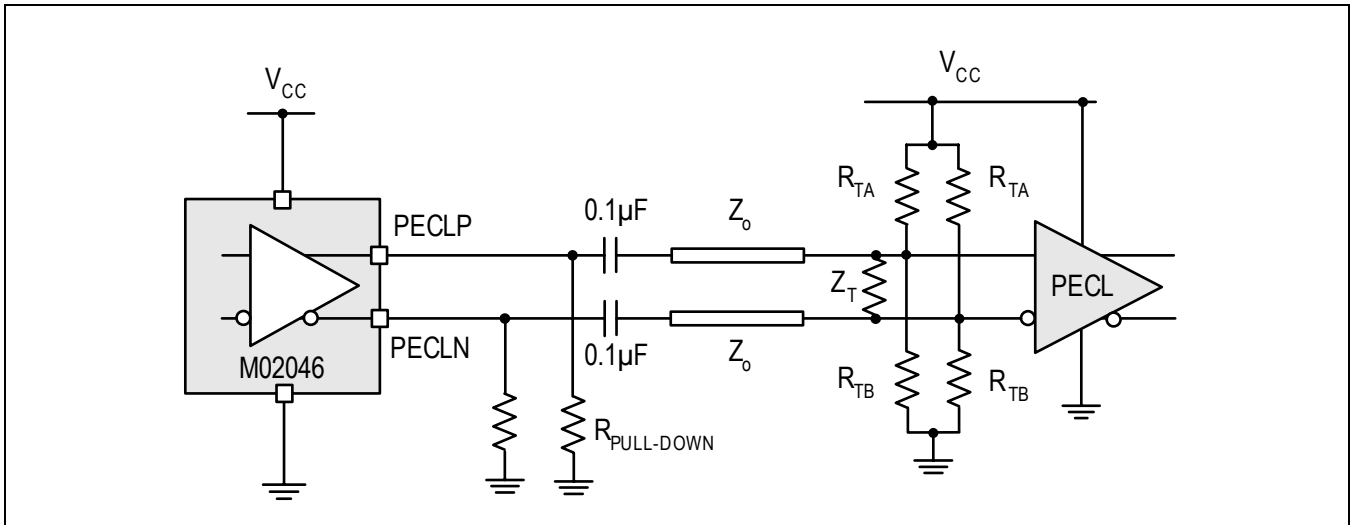


Figure 1-11. DC-Coupled PECL Termination

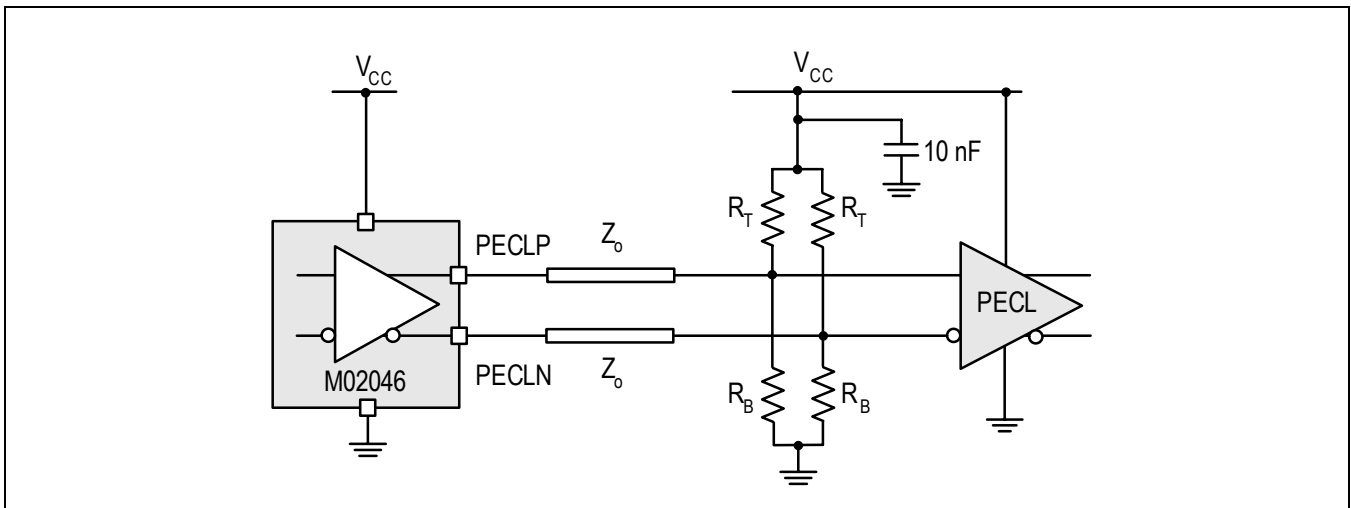
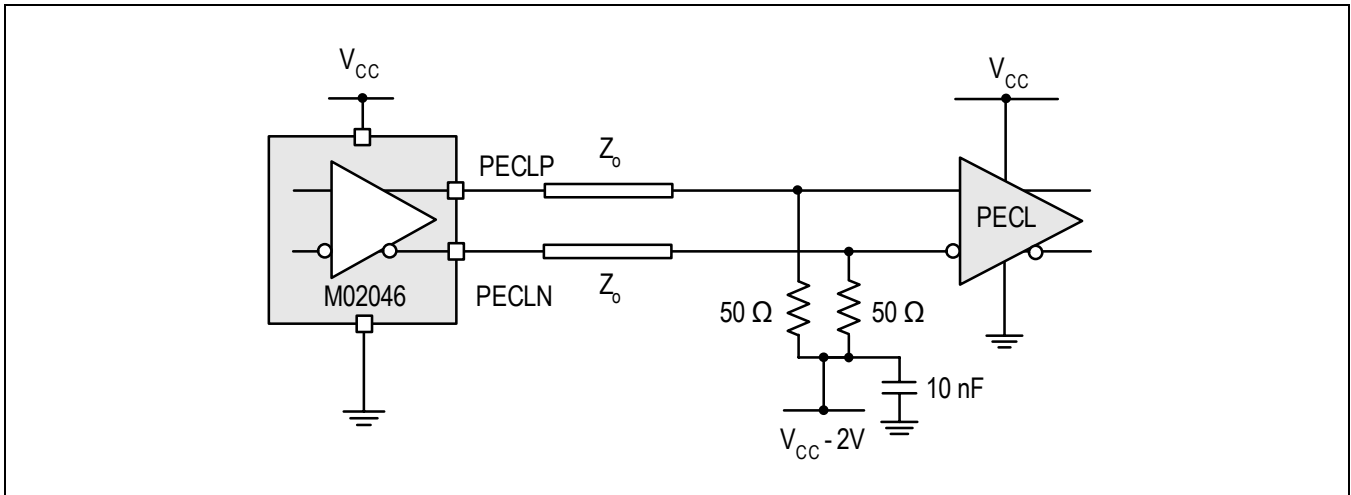


Figure 1-12. Alternative PECL Termination



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1.4 Pin Definitions

Table 1-3. Pin Descriptions

QSOP Pin#	Name	Function
1	ST _{SET}	Loss of signal threshold setting input. Connect a 1% resistor between this pin and V _{CC3} to set loss of signal threshold.
2	V _{CC3}	Power supply input for 3.3V applications or the output of the internally regulated 3.3V voltage when V _{CC} = 5V. Connect directly to supply for 3.3V applications (internal regulator not in use). Do not connect to power supply if V _{CC} = 5V.
3	GND	Ground.
4	DINP	Non-inverting data input. Internally terminated with 50Ω to V _{TT} .
5	DINN	Inverting data input. Internally terminated with 50Ω to V _{TT} .
6	NC	No Connect. Leave Floating.
7	V _{CC3}	Power supply input for 3.3V applications or the output of the internally regulated 3.3V voltage when V _{CC} = 5V. Connect directly to supply for 3.3V applications (internal regulator not in use). Do not connect to power supply if V _{CC} = 5V.
8	JAM	Output disable. When high, data outputs are disabled (with non-inverting output held high and inverting output held low). Connect to LOS output to disable outputs with loss of signal. Outputs are enabled when JAM is low or floating. Internal 150kΩ resistor to ground.
9	LOS	Loss of signal output. Goes high when input signal falls below threshold set by ST _{SET} . This output is an open collector TTL with internal 80kΩ pull-up resistor to V _{CC} . Leave floating if not used.
10	ST	Signal detect output. Goes high when input signal amplitude is above threshold set by ST _{SET} . In M02046C, this output is an open collector TTL with internal 80kΩ pull-up resistor to V _{CC} ; In M02046, this output is PECL. Leave floating if not used.
11	GND	Ground.
12	PECLN	Inverting data output (PECL).
13	PECLP	Non-inverting data output (PECL).
14	V _{CC}	Power supply. Connect to either +5V or +3.3V.
15	NC	No Connect. Leave Floating.
16	I _{REF}	Internal reference current. Must be connected to ground through a 12.1kΩ 1% resistor.

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2.0 Product Specification

2.1 Absolute Maximum Ratings

These are the absolute maximum ratings at or beyond which the IC can be expected to fail or be damaged. Reliable operation at these extremes for any length of time is not implied.

NOTE:

The package bottom must be adequately grounded to ensure correct thermal and electrical performance, and it is recommended that vias are inserted through to a lower ground plane.

Table 2-1. Absolute Maximum Ratings

Symbol	Parameter	Rating	Units
V_{CC}	Power supply voltage (V_{CC} -GND)	-0.5 to +6V	V
T_{STG}	Storage temperature	-65 to +150	°C
PECLP, PECLN, ST_{PECL}	PECL Output pins voltage	$V_{CC} - 2$ to $V_{CC} + 0.4$	V
$I(PECLP)$, $I(PECLN)$, $I(ST_{PECL})$	PECL Output pins maximum continuous current (delivered to load)	30	mA
$ DINP - DINN $	Data input pins differential voltage	0.80	V
DINP, DINN	Data input pins voltage meeting $ DINP - DINN $ requirement	GND to $V_{CC3} + 0.4$	V
ST_{SET}	Signal detect threshold setting pin voltage	GND to $V_{CC} + 0.4$	V
JAM	Output enable pin voltage	GND to $V_{CC} + 0.4$	V
ST_{CMOS} , LOS	CMOS Status Output pins voltage	GND to $V_{CC} + 0.4$	V
I_{REF}	Current into Reference input	+0 to -120	μ A
$I(LOS)$, $I(ST_{CMOS})$	Current into CMOS Status Output pins	+1500 to -100	μ A

2.2 Recommended Operating Conditions

Table 2-2. Recommended Operating Conditions

Parameter	Rating	Units
Power supply: (V_{CC} -GND) (apply no potential to V_{CC3}) or (V_{CC3} -GND) (connect V_{CC} to same potential as V_{CC3})	+5V \pm 7.5% or +3.3V \pm 7.5%	V
Junction temperature	-40 to +110	°C
Operating ambient	-40 to +85	°C

2.3 DC Characteristics

$V_{CC} = +3.3V \pm 7.5\%$ or $+5V \pm 7.5\%$, $T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$, unless otherwise noted.

Typical specifications are for $V_{CC} = 3.3V$, $T_A = 25^\circ\text{C}$, unless otherwise noted.

Table 2-3. DC Characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Units
I_{CC}	Supply Current	Includes PECL load		52	TBD	mA
$V_{OUTL_{pecl}}$	PECL Output Low Voltage (PECLP, PECLN)	Single ended; 50 Ω load to $V_{CC} - 2V$	$V_{CC}-1.81$	$V_{CC}-1.71$	$V_{CC}-1.62$	V
$V_{OUTH_{pecl}}$	PECL Output High Voltage (PECLP, PECLN)	Single ended; 50 Ω load to $V_{CC} - 2V$	$V_{CC}-1.025$	$V_{CC}-0.952$	$V_{CC}-0.88$	V
	Differential Input Resistance	Measured between DINP and DINN	85	100	115	Ω
	CMOS ST ⁽¹⁾ , LOS ^(1,2) Output High Voltage	External 4.7-10k Ω pull up to V_{CC}	2.4	V_{CC}		V
	CMOS ST ⁽¹⁾ , LOS ^(1,2) Output Low Voltage	External 4.7-10k Ω pull up to V_{CC}	0		0.4	V
	PECL ST Output High Voltage ⁽²⁾	ST terminated 50 Ω to $V_{CC} - 2V$	$V_{CC}-1.025$	$V_{CC}-0.952$	$V_{CC}-0.88$	V
	PECL ST Output Low Voltage ⁽²⁾	ST terminated 50 Ω to $V_{CC} - 2V$	$V_{CC}-1.81$	$V_{CC}-1.71$	$V_{CC}-1.62$	V
	JAM Input High Voltage		2.7		5.5	V
	JAM Input Low Voltage				0.8	V

Notes:

1. M02046C
2. M02046

2.4 AC Characteristics

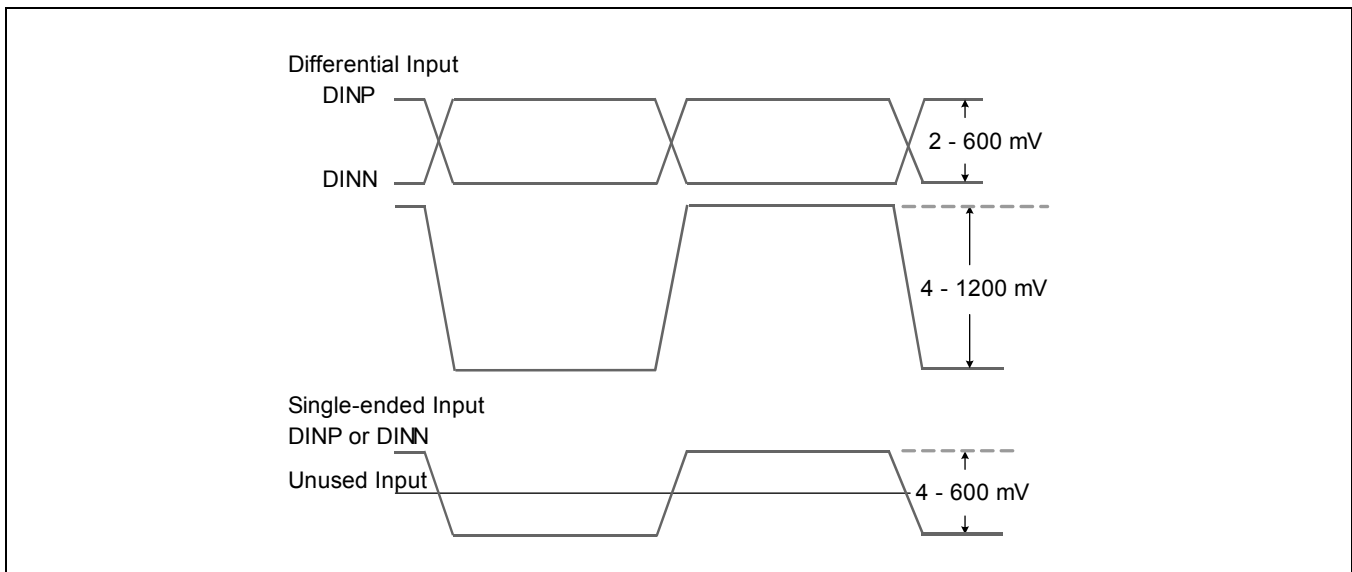
$V_{CC} = +3.3V \pm 7.5\%$ or $+5V \pm 7.5\%$, $T_A = -40^\circ C$ to $+85^\circ C$, input bit rate = 2.5 Gbps 2^{23} -1 PRBS unless otherwise noted. Typical specifications are for $V_{CC} = 3.3V$, $T_A = 25^\circ C$, unless otherwise noted.

Table 2-4. AC Characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Units
$V_{IN(MIN)}$	Differential Input Sensitivity	1.25 Gbps, BER < 10^{-12}			4	mV
$V_{I(MAX)}$	Input Overload	BER < 10^{-12} , differential input 1.25 Gbps	1200			mV
		BER < 10^{-12} , single-ended input, 1.25 Gbps	600			mV
v_n	RMS Input Referred Noise			185		μV_{RMS}
V_{SD}	Signal Detect Programmable Range	Differential inputs	4		100	mV
HYS	Signal Detect Hysteresis	(electrical); signal detect level set to 20 mV _{pp}	2	4	6	dB
BW_{LF}	Small-Signal -3dB Low Frequency Cutoff	Excluding AC coupling capacitors		25		kHz
DJ	Deterministic Jitter	K28.5 pattern at 1.25 Gbps			25	ps
RJ	Random Jitter	10 mV _{pp} input		3		ps _{RMS}
t_r / t_f	Data Output Rise and Fall Times	20% to 80%; outputs terminated into 50Ω; 10 mV _{pp} input		185	300	ps
T_{LOS_ON}	Time from LOS state until LOS output is asserted	LOS assert time after 1 V _{pp} input signal is turned off; signal detect level set to 10 mV			100	μs
T_{LOS_OFF}	Time from non-LOS state until LOS is deasserted	LOS deassert time after input crosses signal detect level; signal detect set to 10 mV with applied input signal of 20 mV _{pp}			100	μs

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Figure 2-1. Data Input Requirements



2.5 Package Specification

Figure 2-3. M02046 Pinout - 16 Pin QSOP Top View

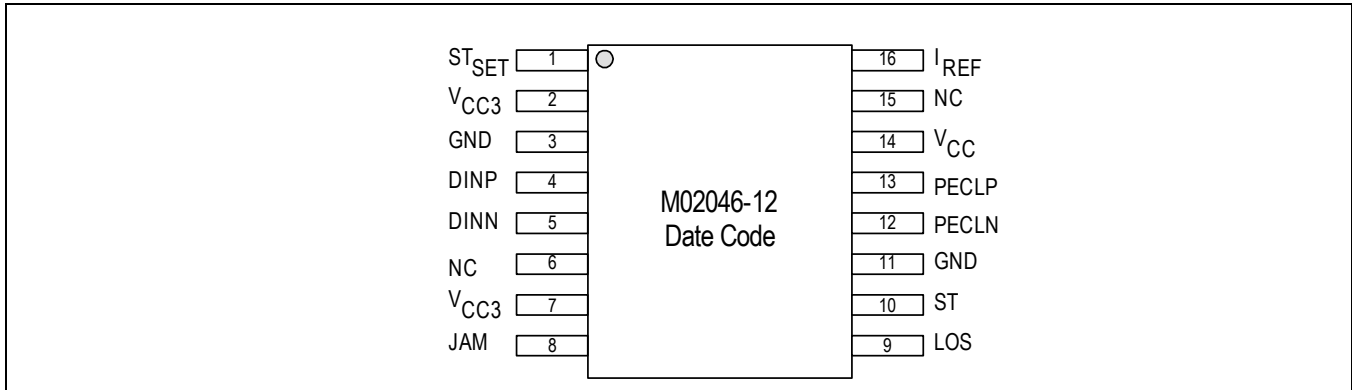
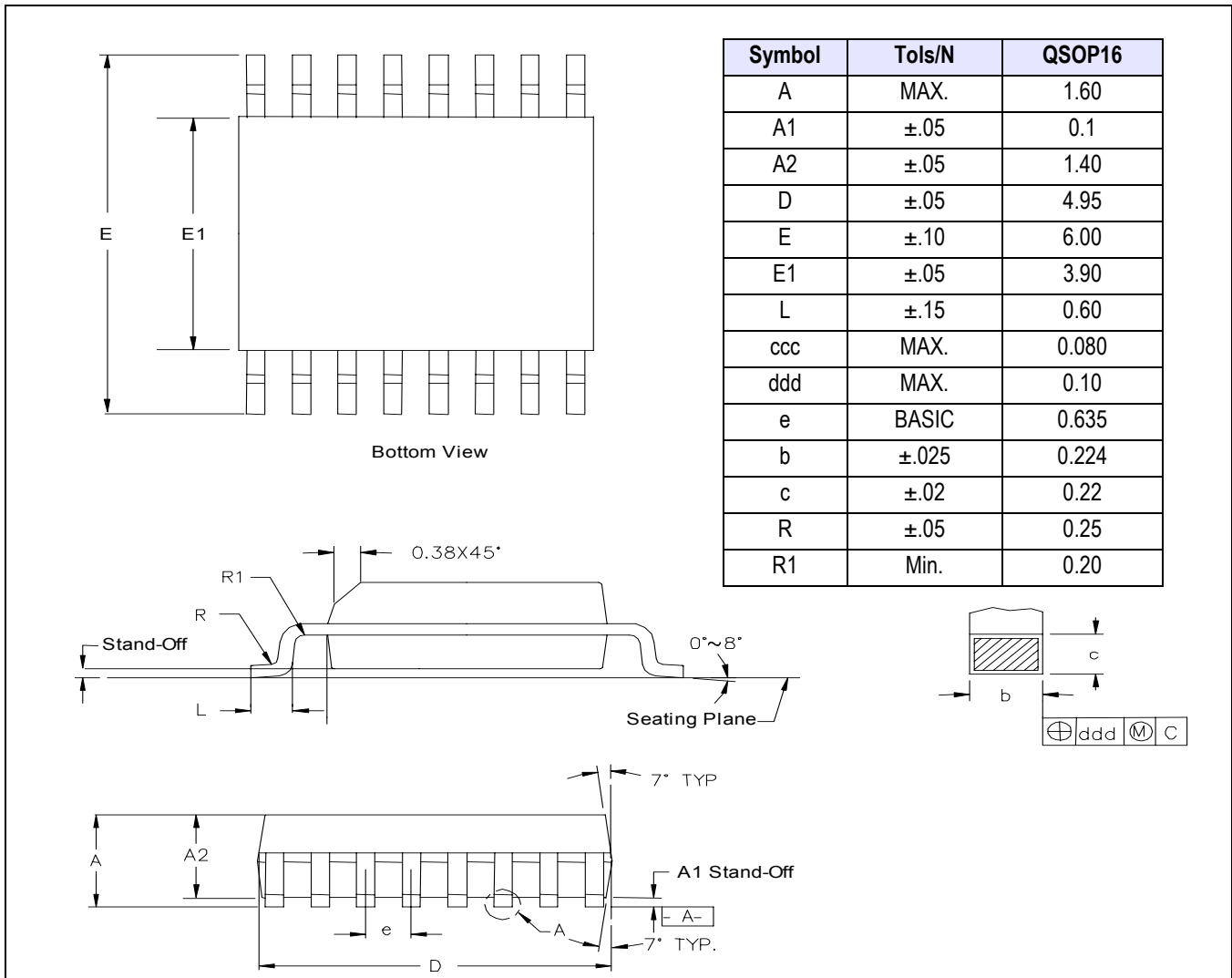


Figure 2-4. Package Information



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