

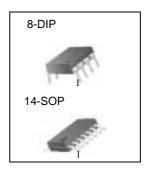
# UC3842/UC3843/UC3844/UC3845 SMPS Controller

#### **Features**

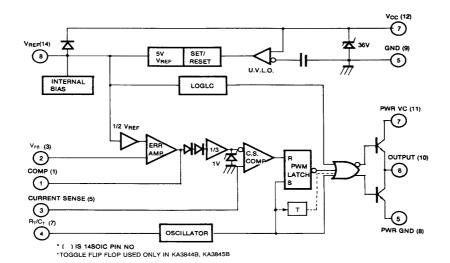
- Low Start Up Current
- Maximum Duty Clamp
- UVLO With Hysteresis
- Operating Frequency Up To 500KHz

### **Description**

The UC3842/UC3844/UC3844 are fixed frequency current-mode PWM controller. They are specially designed for Off - Line and DC-to-DC converter applications with minimum external components. These integrated circuits feature a trimmed oscillator for precise duty cycle control, a temperature compensated reference, high gain error amplifier. current sensing comparator, and a high current totempole output Ideally suited for driving a power MOSFET. Protection circuity Includes built in under-voltage lockout and current limiting. TheUC3842 and UC3844 have UVLO thresholds of 16V (on) and 10V (off) The UC3843 and UC3845 are 8.5V (on) and 7.9V (off) The UC3842 and UC3843 can operate within 100% duty cycle. The UC3844and UC3845 can operate with 50% duty cycle.



#### **Internal Block Diagram**



## **Absolute Maximum Ratings**

Parameter	Symbol	Value	Unit
Supply Voltage	Vcc	30	V
Output Current	lo	±1	A
Analog Inputs (Pin 2.3)	V(ANA)	-0.3 to 6.3	V
Error Amp Output Sink Current	ISINK (E.A)	10	mA
Power Dissipation (T <sub>A</sub> = 25°C)	PD	1	W

### **Electrical Characteristics**

(V<sub>CC</sub>=15V, R<sub>T</sub>=10K $\Omega$ , C<sub>T</sub>=3.3nF, T<sub>A</sub>= 0°C to +70°C, unless otherwise specified)

Parameter	Symbol	Conditions	Min.	Тур.	Max.	Unit
REFERENCE SECTION						
Reference Output Voltage	VREF	T <sub>J</sub> = 25°C, I <sub>REF</sub> = 1mA	4.90	5.00	5.10	V
Line Regulation	ΔVREF	12V≤VCC≤25V	-	6	20	mV
Load Regulation	$\Delta VREF$	1mA≤I <sub>REF</sub> ≤20mA	-	6	25	mV
Short Circuit Output Current	Isc	T <sub>A</sub> = 25°C	-	-100	-180	mA
OSCILLATOR SECTION						
Oscillation Frequency	f	T <sub>J</sub> = 25°C	47	52	57	KHz
Frequency Change with Voltage	Δf/ΔVCC	12V≤VCC≤25V	-	0.05	1	%
Oscillator Amplitude	Vosc	-	-	1.6	-	V <sub>P-P</sub>
ERROR AMPLIFIER SECTION				l		
Input Bias Current	IBIAS	-	-	-0.1	-2	μΑ
Input Voltage	VI(E>A)	V <sub>1</sub> = 2.5V	2.42	2.50	2.58	V
Open Loop Voltage Gain	Gvo	2V≤ V <sub>O</sub> ≤4V	65	90	-	dB
Power Supply Rejection Ratio	PSRR	12V≤ V <sub>CC</sub> ≤25V	60	70	-	dB
Output Sink Current	ISINK	V <sub>2</sub> = 2.7V, V <sub>1</sub> = 1.1V	2	7	-	mA
Output Source Current	ISOURCE	V <sub>2</sub> = 2.3V, V <sub>1</sub> = 5V	-0.6	-1.0	-	mA
High Output Voltage	Voн	$V_2 = 2.3V$ , $R_L = 15K\Omega$ to GND	5	6	-	V
Low Output Voltage	Vol	$V_2 = 2.7V$ , $R_L = 15K\Omega$ to Pin 8	-	0.8	1.1	V
CURRENT SENSE SECTION						
Gain	G∨	(Note 1 & 2)	2.85	3	3.15	V/V
Maximum Input Signal	VI(MAX)	V <sub>1</sub> = 5V(Note 1)	0.9	1	1.1	V
Power Supply Rejection Ratio	PSRR	12V≤ V <sub>CC</sub> ≤25V (Note 1)	-	70	-	dB
Input Bias Current	IBIAS	-	-	-3	-10	μΑ
OUTPUT SECTION				•		
Low Output Voltage	VoL	ISINK = 20mA	-	0.08	0.4	V
		ISINK = 200mA	-	1.4	2.2	V
High Output Voltage	Voн	ISOURCE = 20mA	13	13.5	-	V
		ISOURCE = 200mA	12	13.0	-	V
Rise Time	tR	T <sub>J</sub> = 25°C, C <sub>L</sub> = 1nF (Note 3)	-	45	150	ns
Fall Time	tF	T <sub>J</sub> = 25°C, C <sub>L</sub> = 1nF (Note 3)	-	35	150	ns
UNDER-VOLTAGE LOCKOUT SECTION						
Start Threshold	\/	UC3842/UC3844	14.5	16.0	17.5	V
V	VTH(ST)	UC3843/UC3845	7.8	8.4	9.0	V
Min. Operating Voltage	Vopr(MIN)	UC3842/UC3844	8.5	10.0	11.5	V
(After Turn On)	VOPK(IVIIN)	UC3843/UC3844	7.0	7.6	8.2	V

### **Electrical Characteristics (Continued)**

(V<sub>CC</sub>=15V, R<sub>T</sub>=10KΩ, C<sub>T</sub>=3.3nF, T<sub>A</sub>= 0°C to +70°C unless otherwise specified)

Parameter	Symbol	Conditions	Min.	Тур.	Max.	Unit
PWM SECTION					_	,
Max. Duty Cycle	D <sub>(max)</sub>	UC3842/UC3843	95	97	100	%
	D	UC3844/UC3845	47	48	50	%
Min. Duty Cycle	D(MIN)	-	-	-	0	%
TOTAL STANDBY CURRENT						
Start-Up Current	IST	-	-	0.45	1	mA
Operating Supply Current	ICC(OPR)	V <sub>3</sub> =V <sub>2</sub> =ON	-	14	17	mA
Zener Voltage	Vz	ICC = 25mA	30	38	-	V

Adjust VCC above the start threshould before setting at 15V

#### Note

- 1. Parameter measured at trip point of latch
- 2. Gain defined as:

$$A = \frac{\Delta V_1}{\Delta V_3} \qquad , 0 \le V3 \le 0.8V$$

3. These parameters, although guaranteed, are not 100 tested in production.

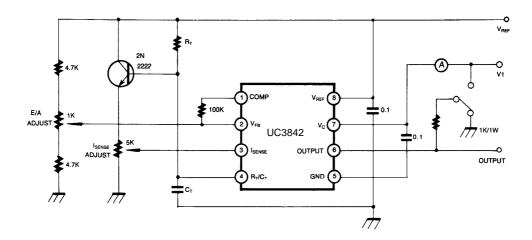


Figure 1. Open Loop Test Circuit

High peak currents associated with capacitive loads necessitate careful grounding techniques Timing and bypass capacitors should be connected close to pin 5 in a single point ground. The transistor and  $5K\Omega$  potentiometer are used to sample the oscillator waveform and apply an adjustable ramp to pin 3.

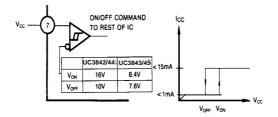
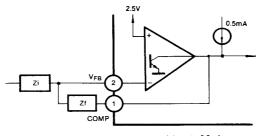


Figure 2. Under Voltage Lockout

During Under-Voltage Lock-Out, the output driver is biased to a high impedance state. Pin 6 should be shunted to ground with a bleeder resistor to prevent activating the power switch with output leakage current.



Error amp can source or sink up to 0.5mA

Figure 3. Error Amp Configuration

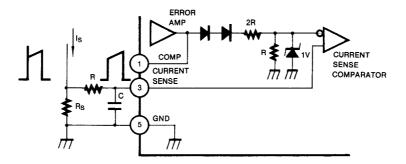


Figure 4. Current Sense Circuit

Peak current (I<sub>S</sub>) is determined by the formula:

$$I_{S}(MAX) = \frac{1.0V}{R_{S}}$$

A small RC filter may be required to suppress switch transients.

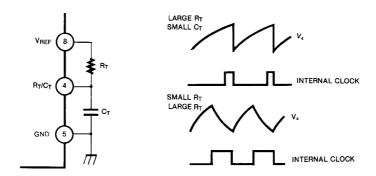


Figure 5. Oscillator Waveforms and Maximum Duty Cycle

Oscillator timing capacitor,  $C_T$ , is charged by  $V_{REF}$  through  $R_T$ , and discharged by an internal current source. During the discharge time, the internal clock signal blanks the output to the low state. Selection of  $R_T$  and  $C_T$  therefore determines both oscillator frequency and maximum duty cycle. Charge and discharge times are determined by the formulas:

$$\begin{split} t_c &= 0.55 \; R_T \; C_T \\ t_D &= \; R_T C_T I_n \! \! \left( \frac{0.0063 R_T - 2.7}{0.0063 R_T - 4} \right) \end{split}$$

Frequency, then, is:  $f=(t_c + t_d)^{-1}$ 

ForRT > 5K
$$\Omega$$
,f=  $\frac{1.8}{R_TC_T}$ 

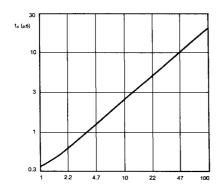


Figure 6. Oscillator Dead Time & Frequency

(Deadtime vs  $C_T RT > 5k\Omega$ )

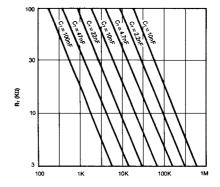


Figure 7. Timing Resistance vs Frequency

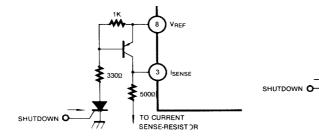


Figure 8. Shutdown Techniques

Shutdown of the UC3842 can be accomplished by two methods; either raise pin 3 above 1V or pull pin 1 below a voltage two diode drops above ground. Either method causes the output of the PWM comparator to be high (refer to block diagram). The PWM latch is reset dominant so that the output will remain low until the next clock cycle after the shutdown condition at pins 1 and/or 3 is removed. In one example, an externally latched shutdown may be accomplished by adding an SOR which will be reset by cycling Voc below the lower UVLO threshold. At this point the reference turns off, allowing the SCR to reset.

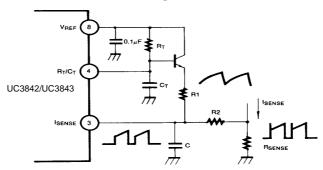


Figure 9. Slope Compensation

A fraction of the oscillator ramp can be resistively summed with the current sense signal to provide slope compensation for converters requiring duty cycles over 50%. Note that capacitor, C, forms a filter with R2 to suppress the leading edge switch spikes.

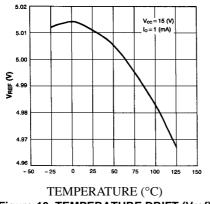


Figure 10. TEMPERATURE DRIFT (Vref)

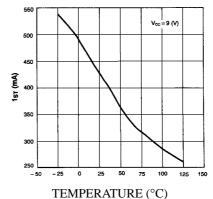


Figure 11. TEMPERATURE DRIFT (Ist)

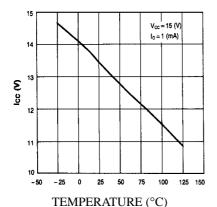
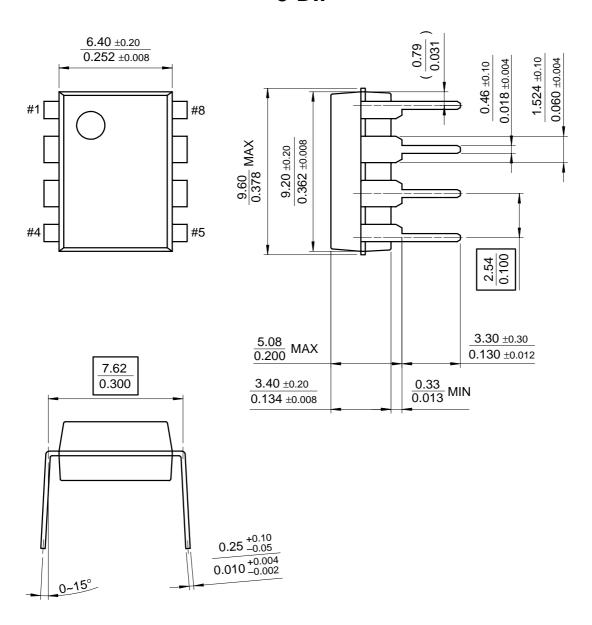


Figure 12. TEMPERATURE DRIFT (Icc)

#### **Mechanical Dimensions**

#### Package

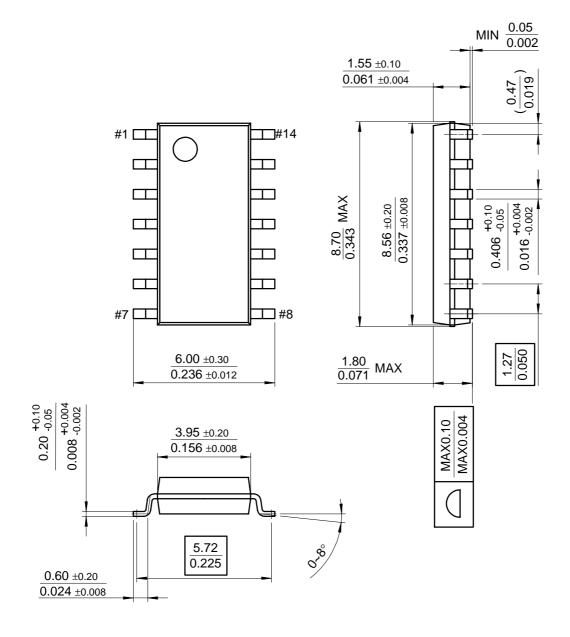
# 8-DIP



#### **Mechanical Dimensions** (Continued)

#### **Package**

### 14-SOP



## **Ordering Information**

Product Number	Package	Operating Temperature
UC3842N		
UC3843N	8 DIP	
UC3844N	-	
UC3845N		0 ~ + 70°C
UC3842D	- 14 SOP	0~+70 C
UC3843D		
UC3844D		
UC3845D		

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