

# HGTP20N60B3, HGTG20N60B3

November 1997

## 40A, 600V, UFS Series N-Channel IGBTs

### **Features**

- 40A, 600V at T<sub>C</sub> = 25°C
- 600V Switching SOA Capability
- · Short Circuit Rated
- Low Conduction Loss

### Ordering Information

PART NUMBER	PACKAGE	BRAND
HGTP20N60B3	TO-220AB	G20N60B3
HGTG20N60B3	TO-247	HG20N60B3

NOTE: When ordering, use the entire part number.

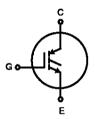
### Description

The HGTP20N60B3 and the HGTG20N60B3 are Generation III MOS gated high voltage switching devices combining the best features of MOSFETs and bipolar transistors. These devices have the high input impedance of a MOSFET and the low on-state conduction loss of a bipolar transistor. The much lower on-state voltage drop varies only moderately between 25°C and 150°C.

The IGBT is ideal for many high voltage switching applications operating at moderate frequencies where low conduction losses are essential, such as: AC and DC motor controls, power supplies and drivers for solenoids, relays and contactors.

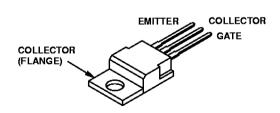
Formerly developmental type TA49050.

### Symbol

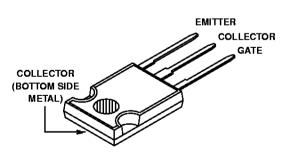


### **Pinouts**

JEDEC TO-220AB



### JEDEC STYLE TO-247



### HARRIS SEMICONDUCTOR IGBT PRODUCT IS COVERED BY ONE OR MORE OF THE FOLLOWING U.S. PATENTS

4,364,073	4,417,385	4,430,792	4,443,931	4,466,176	4,516,143	4,532,534	4,567,641
4,587,713	4,598,461	4,605,948	4,618,872	4,620,211	4,631,564	4,639,754	4,639,762
4,641,162	4,644,637	4,682,195	4,684,413	4,694,313	4,717,679	4,743,952	4,783,690
4,794,432	4,801,986	4,803,533	4,809,045	4,809,047	4,810,665	4,823,176	4,837,606
4,860,080	4,883,767	4,888,627	4,890,143	4,901,127	4,904,609	4,933,740	4,963,951
4,969,027							

### HGTP20N60B3, HGTG20N60B3

### Absolute Maximum Ratings T<sub>C</sub> = 25°C, Unless Otherwise Specified

	HGTP20N60B3 HGTG20N60B3	UNITS
Collector to Emitter Voltage	600	V
Collector to Gate Voltage, $R_{GE} = 1M\Omega$	600	V
Collector Current Continuous		
At T <sub>C</sub> = 25°C	40	Α
At T <sub>C</sub> = 110°C I <sub>C110</sub>	20	Α
Collector Current Pulsed (Note 1)	160	Α
Gate to Emitter Voltage Continuous	±20	٧
Gate to Emitter Voltage Pulsed	± <b>30</b>	V
Switching Safe Operating Area at T <sub>C</sub> = 150°CSSOA	30A at 600V	
Power Dissipation Total at T <sub>C</sub> = 25°C	165	W
Power Dissipation Derating T <sub>C</sub> > 25°C	1.32	W/°C
Operating and Storage Junction Temperature Range	-40 to 150	°C
Maximum Lead Temperature for SolderingTL	260	°C
Short Circuit Withstand Time (Note 2) at V <sub>GE</sub> = 15V	4	μs
Short Circuit Withstand Time (Note 2) at V <sub>GE</sub> = 10V	10	μs

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

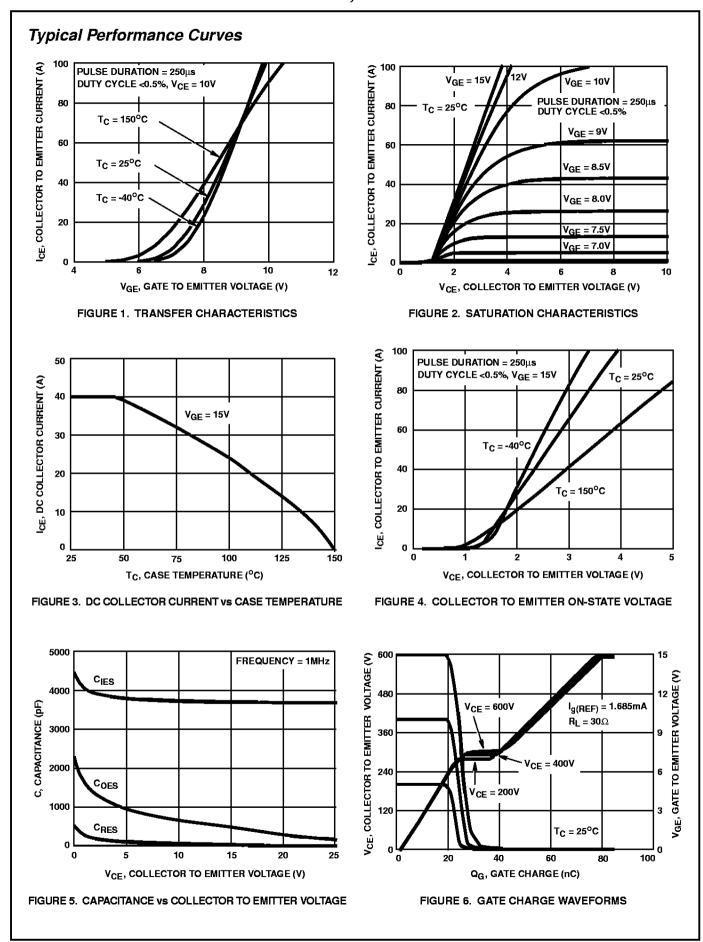
- 1. Repetitive Rating: Pulse width limited by maximum junction temperature. 2.  $V_{CE}$  = 360V,  $T_{C}$  = 125°C,  $R_{GE}$  = 25 $\Omega$ .

## $\textbf{Electrical Specifications} \quad \textbf{T}_{C} = 25^{o}\text{C, Unless Otherwise Specified}$

PARAMETER	SYMBOL	TEST CONDITIONS		MIN	TYP	MAX	UNITS
Collector to Emitter Breakdown Voltage	BV <sub>CES</sub>	$I_C = 250 \mu A, V_{GE} = 0 V$		600	-	-	٧
Collector to Emitter Leakage Current	ICES	V <sub>CE</sub> = BV <sub>CES</sub>	T <sub>C</sub> = 25°C	-	-	250	μΑ
			T <sub>C</sub> = 150°C	-	-	1.0	mA
Collector to Emitter Saturation Voltage	V <sub>CE(SAT)</sub>	I <sub>C</sub> = I <sub>C110</sub> , V <sub>GE</sub> = 15V	T <sub>C</sub> = 25°C	-	1.8	2.0	٧
			T <sub>C</sub> = 150°C	-	2.1	2.5	٧
Gate to Emitter Threshold Voltage	V <sub>GE(TH)</sub>	I <sub>C</sub> = 250μA, V <sub>CE</sub> = V <sub>GE</sub>		3.0	5.0	6.0	٧
Gate to Emitter Leakage Current	IGES	V <sub>GE</sub> = ±20V		-	-	±100	nA
Switching SOA	SSOA	$T_C = 150^{\circ}C$ $V_{GE} = 15V$ $R_G = 10\Omega$ $L = 45\mu H$	V <sub>CE</sub> = 480V	100	-	-	Α
			V <sub>CE</sub> = 600V	30	-	-	А
Gate to Emitter Plateau Voltage	V <sub>GEP</sub>	I <sub>C</sub> = I <sub>C110</sub> , V <sub>CE</sub> = 0.5 BV <sub>CES</sub>		-	8.0	-	٧
On-State Gate Charge	Q <sub>G(ON)</sub>	I <sub>C</sub> = I <sub>C110</sub> , V <sub>CE</sub> = 0.5 BV <sub>CES</sub>	V <sub>GE</sub> = 15V	-	80	105	пC
			V <sub>GE</sub> = 20V	-	105	135	пC
Current Turn-On Delay Time	t <sub>d</sub> (ON)I	T <sub>C</sub> = 150°C	I <sub>CE</sub> = I <sub>C110</sub>		25	-	ns
Current Rise Time	t <sub>rl</sub>	CE =   C110   VCE = 0.8 BVCES			20	-	ns
Current Turn-Off Delay Time	td(OFF)I	$V_{GE} = 0.8 \text{ BVCES}$ $V_{GE} = 15V$ $R_{G} = 10\Omega$ $L = 100\mu\text{H}$		-	220	275	ns
Current Fall Time	t <sub>fl</sub>			-	140	175	ns
Turn-On Energy	E <sub>ON</sub>			-	475	-	μ٦
Turn-Off Energy (Note 3)	E <sub>OFF</sub>	1	-	1050	-	μIJ	
Thermal Resistance	R <sub>eJC</sub>			-	-	0.76	°C/W

#### NOTE:

<sup>3.</sup> Turn-Off Energy Loss (EOFF) is defined as the integral of the instantaneous power loss starting at the trailing edge of the input pulse and ending at the point where the collector current equals zero (I<sub>CE</sub> = 0A). The HGTP20N60B3 and HGTG20N60B3 were tested per JEDEC standard No. 24-1 Method for Measurement of Power Device Turn-Off Switching Loss. This test method produces the true total Turn-Off Energy Loss. Turn-on losses include diode losses.



### Typical Performance Curves (Continued)

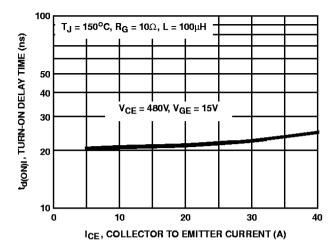


FIGURE 7. TURN-ON DELAY TIME AS A FUNCTION OF COLLECTOR TO EMITTER CURRENT

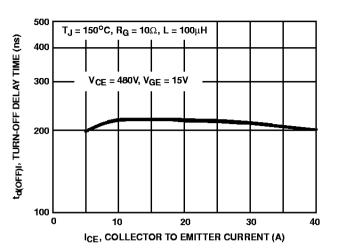


FIGURE 8. TURN-OFF DELAY TIME AS A FUNCTION OF COLLECTOR TO EMITTER CURRENT

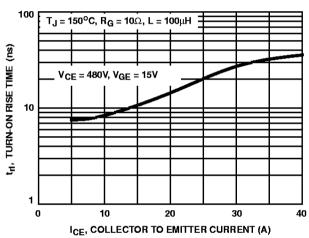


FIGURE 9. TURN-ON RISE TIME AS A FUNCTION OF COLLECTOR TO EMITTER CURRENT

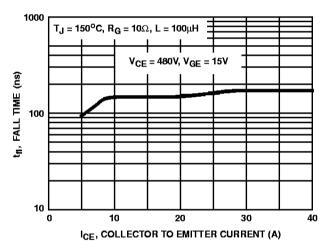


FIGURE 10. TURN-OFF FALL TIME AS A FUNCTION OF COLLECTOR TO EMITTER CURRENT

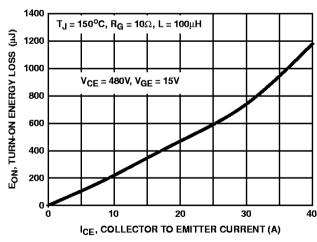


FIGURE 11. TURN-ON ENERGY LOSS AS A FUNCTION OF COLLECTOR TO EMITTER CURRENT

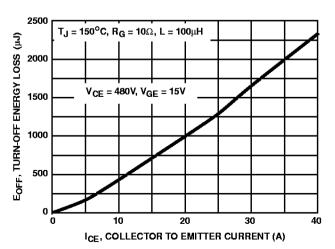


FIGURE 12. TURN-OFF ENERGY LOSS AS A FUNCTION OF COLLECTOR TO EMITTER CURRENT

### Typical Performance Curves (Continued)

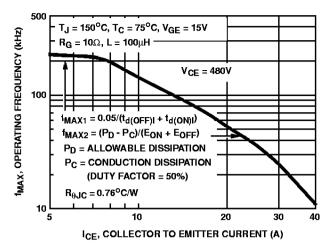


FIGURE 13. OPERATING FREQUENCY AS A FUNCTION OF COLLECTOR TO EMITTER CURRENT

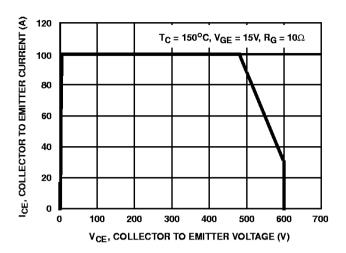


FIGURE 14. SWITCHING SAFE OPERATING AREA

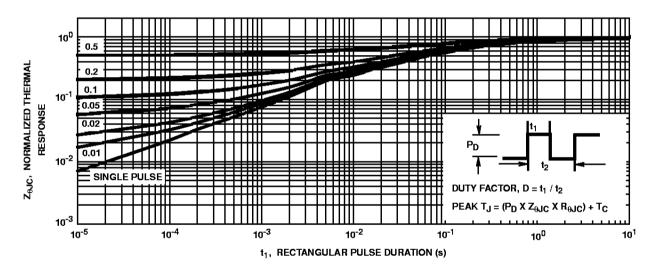


FIGURE 15. IGBT NORMALIZED TRANSIENT THERMAL RESPONSE, JUNCTION TO CASE

### Test Circuit and Waveform

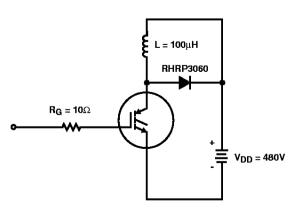


FIGURE 16. INDUCTIVE SWITCHING TEST CIRCUIT

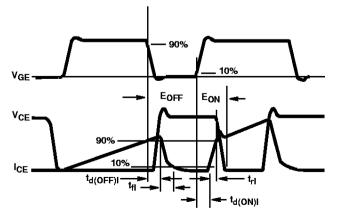


FIGURE 17. SWITCHING TEST WAVEFORMS

### Handling Precautions for IGBTs

Insulated Gate Bipolar Transistors are susceptible to gateinsulation damage by the electrostatic discharge of energy through the devices. When handling these devices, care should be exercised to assure that the static charge built in the handler's body capacitance is not discharged through the device. With proper handling and application procedures, however, IGBTs are currently being extensively used in production by numerous equipment manufacturers in military, industrial and consumer applications, with virtually no damage problems due to electrostatic discharge. IGBTs can be handled safely if the following basic precautions are taken:

- 1. Prior to assembly into a circuit, all leads should be kept shorted together either by the use of metal shorting springs or by the insertion into conductive material such as "ECCOSORBD™ LD26" or equivalent.
- 2. When devices are removed by hand from their carriers, the hand being used should be grounded by any suitable means - for example, with a metallic wristband.
- 3. Tips of soldering irons should be grounded.
- 4. Devices should never be inserted into or removed from circuits with power on.
- 5. Gate Voltage Rating Never exceed the gate-voltage rating of  $V_{\mbox{\scriptsize GEM}}$ . Exceeding the rated  $V_{\mbox{\scriptsize GE}}$  can result in permanent damage to the oxide layer in the gate region.
- 6. Gate Termination The gates of these devices are essentially capacitors. Circuits that leave the gate opencircuited or floating should be avoided. These conditions can result in turn-on of the device due to voltage buildup on the input capacitor due to leakage currents or pickup.
- 7. Gate Protection These devices do not have an internal monolithic zener diode from gate to emitter. If gate protection is required an external zener is recommended.

### Operating Frequency Information

Operating frequency information for a typical device (Figure 13) is presented as a guide for estimating device performance for a specific application. Other typical frequency vs collector current (ICF) plots are possible using the information shown for a typical unit in Figures 4, 7, 8, 11 and 12. The operating frequency plot (Figure 13) of a typical device shows f<sub>MAX1</sub> or f<sub>MAX2</sub> whichever is smaller at each point. The information is based on measurements of a typical device and is bounded by the maximum rated junction temperature.

 $f_{MAX1}$  is defined by  $f_{MAX1} = 0.05/(t_{d(OFF)} + t_{d(ON)})$ . Deadtime (the denominator) has been arbitrarily held to 10% of the on- state time for a 50% duty factor. Other definitions are possible.  $t_{d(OFF)I}$  and  $t_{d(ON)I}$  are defined in Figure 17.

Device turn-off delay can establish an additional frequency limiting condition for an application other than T<sub>JMAX</sub>. td(OFF) is important when controlling output ripple under a lightly loaded condition.

 $f_{MAX2}$  is defined by  $f_{MAX2} = (P_D - P_C)/(E_{OFF} + E_{ON})$ . The allowable dissipation (P<sub>D</sub>) is defined by P<sub>D</sub> =  $(T_{JMAX} - T_C)/R_{BJC}$ .

The sum of device switching and conduction losses must not exceed Pp. A 50% duty factor was used (Figure 13) and the conduction losses (P<sub>C</sub>) are approximated by  $P_C = (V_{CE} \times I_{CE})/2$ .

EON and EOFF are defined in the switching waveforms shown in Figure 17. EON is the integral of the instantaneous power loss (ICE x VCE) during turn-on and EOFF is the integral of the instantaneous power loss (ICE x VCE) during turn-off. All tail losses are included in the calculation for  $E_{OFF}$ ; i.e. the collector current equals zero ( $I_{CF} = 0$ ).

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