

# DELPHI SERIES



## FEATURES

- High efficiency: 97% @ 24Vin, 12V/6A out
- Small size and low profile: (SIP) 50.8 x 12.7 x 9.5mm (2.00" x 0.50" x 0.37")
- Standard footprint
- Voltage and resistor-based trim
- Pre-bias startup
- No minimum load required
- Output voltage programmable from 5Vdc to 15Vdc via external resistor
- Fixed frequency operation (300KHz)
- Input UVLO, output OTP, OCP
- Remote ON/OFF
- Remote sense
- ISO 9001, TL 9000, ISO 14001, QS9000, OHSAS18001 certified manufacturing facility
- UL/cUL 60950-1 (US & Canada), and TUV (EN60950-1) - pending

## Delphi DNM24 series Non-Isolated Point of Load DC/DC Power Modules: 20-30Vin, 5-15V/10A out

The Delphi series DNM24S, 20~30V input, single output, non-isolated point of load DC/DC converters are the latest offering from a world leader in power systems technology and manufacturing — Delta Electronics, Inc. The DNM24S series provides a programmable output voltage from 5V to 15V through an external trimming resistor. This product family is available in SIP package and provides 10A of output current in an industry standard footprint and pinout. With creative design technology and optimization of component placement, these converters possess outstanding electrical and thermal performance and extremely high reliability under highly stressful operating conditions.

## OPTIONS

- Negative On/Off logic

## APPLICATIONS

- Telecom/DataCom
- Distributed power architectures
- Servers and workstations
- LAN/WAN applications
- Data processing applications

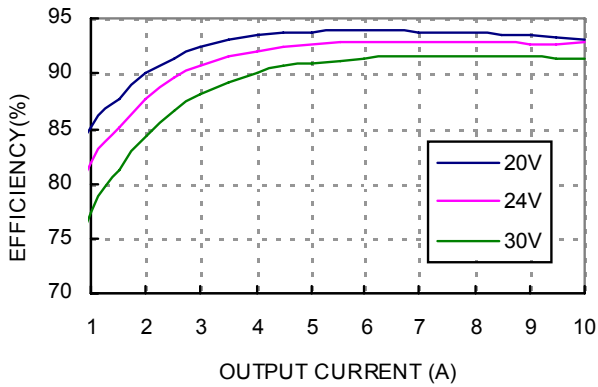
# TECHNICAL SPECIFICATIONS

T<sub>A</sub> = 25°C, airflow rate = 300 LFM, V<sub>in</sub> = 20Vdc and 30Vdc, nominal V<sub>out</sub> unless otherwise noted.

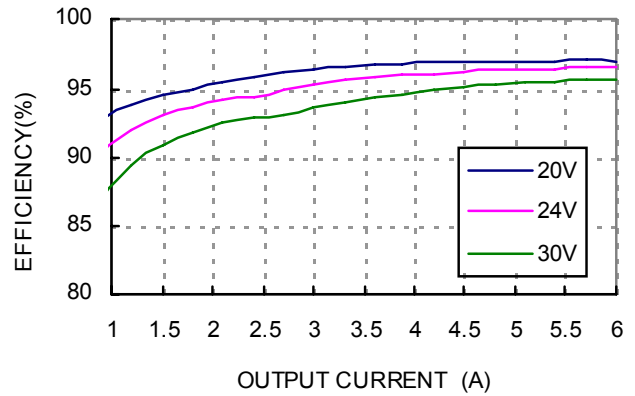
PARAMETER	NOTES and CONDITIONS	DNM24S0B0R10NFB			
		Min.	Typ.	Max.	Units
<b>ABSOLUTE MAXIMUM RATINGS</b>					
Input Voltage (Continuous)		0		36	Vdc
Operating Temperature	Refer to Figure 27 for the measuring point	-40		125	°C
Storage Temperature		-55		125	°C
<b>INPUT CHARACTERISTICS</b>					
Operating Input Voltage		20	24	30	V
Input Under-Voltage Lockout					
Turn-On Voltage Threshold				19	V
Turn-Off Voltage Threshold		17			V
Maximum Input Current	V <sub>in</sub> =V <sub>in,min</sub> to V <sub>in,max</sub> , I <sub>o</sub> =6A, V <sub>o</sub> =12V			4.5	A
No-Load Input Current	V <sub>in</sub> =24V, I <sub>o</sub> =Min Load, V <sub>o</sub> =12V		70		mA
Off Converter Input Current	V <sub>in</sub> =24V, Off Converter			3	mA
Inrush Transient	V <sub>in</sub> = V <sub>in,min</sub> to V <sub>in,max</sub> , I <sub>o</sub> =I <sub>o,min</sub> to I <sub>o,max</sub>			1	A <sup>2</sup> S
Recommended Input Fuse				15	A
<b>OUTPUT CHARACTERISTICS</b>					
Output Voltage Set Point	V <sub>in</sub> =24V, I <sub>o</sub> =I <sub>o,max</sub>	-2.0	V <sub>o,set</sub>	+2.0	% V <sub>o,set</sub>
Output Voltage Adjustable Range		5		15	V
<b>Output Voltage Regulation</b>					
Over Line	V <sub>in</sub> =V <sub>in,min</sub> to V <sub>in,max</sub>			0.4	% V <sub>o,set</sub>
Over Load	I <sub>o</sub> =I <sub>o,min</sub> to I <sub>o,max</sub>			0.4	% V <sub>o,set</sub>
Over Temperature	T <sub>a</sub> = -40°C to 85°C		0.5	1	% V <sub>o,set</sub>
Total Output Voltage Range	Over sample load, line and temperature	-3		+3	% V <sub>o,set</sub>
<b>Output Voltage Ripple and Noise</b>					
5Hz to 20MHz bandwidth					
Peak-to-Peak	V <sub>in</sub> =min to max, I <sub>o</sub> =min to max 1μF ceramic, 10μF Tan		80	160	mV
RMS	V <sub>in</sub> =min to max, I <sub>o</sub> =min to max 1μF ceramic, 10μF Tan		30	60	mV
<b>Output Current Range</b>					
V <sub>in</sub> =24V, V <sub>o</sub> =5V		0		10	A
V <sub>in</sub> =24V, V <sub>o</sub> =12V		0		6	A
V <sub>in</sub> =24V, V <sub>o</sub> =15V		0		4.5	A
Output Voltage Over-shoot at Start-up	V <sub>in</sub> =min to max, I <sub>o</sub> =I <sub>o,max</sub>			3	% V <sub>o,set</sub>
Output Short-Circuit Current (Hiccup mode)	I <sub>o,s/c</sub>		20		Adc
<b>DYNAMIC CHARACTERISTICS</b>					
<b>Dynamic Load Response</b>					
10μF Tan & 1μF ceramic load cap, 5A/μs, V <sub>in</sub> =24V, V <sub>o</sub> =12V, I <sub>o,max</sub> =6A, no external out capacitor					
Positive Step Change in Output Current	50% I <sub>o,max</sub> to 100% I <sub>o,max</sub>		280		mVpk
Negative Step Change in Output Current	100% I <sub>o,max</sub> to 50% I <sub>o,max</sub>		280		mVpk
Settling Time to 10% of Peak Deviation			50		μs
<b>Dynamic Load Response</b>					
10μF Tan & 1μF ceramic load cap, 5A/μs, V <sub>in</sub> =24V, V <sub>o</sub> =12V, I <sub>o,max</sub> =6A, 2x150uF OS-CON capacitor					
Positive Step Change in Output Current	50% I <sub>o,max</sub> to 100% I <sub>o,max</sub>		130		mVpk
Negative Step Change in Output Current	100% I <sub>o,max</sub> to 50% I <sub>o,max</sub>		130		mVpk
Settling Time to 10% of Peak Deviation			50		μs
<b>Turn-On Transient</b>					
Turn-On Transient	I <sub>o</sub> =I <sub>o,max</sub>				
Start-Up Time, From On/Off Control	V <sub>on/off</sub> , V <sub>o</sub> =10% of V <sub>o,set</sub>	2	4	8	ms
Start-Up Time, From Input	V <sub>in</sub> =V <sub>in,min</sub> , V <sub>o</sub> =10% of V <sub>o,set</sub>	2	4	8	ms
Output Voltage Rise Time	Time for V <sub>o</sub> to rise from 10% to 90% of V <sub>o,set</sub>	2	5	9	ms
<b>Maximum Output Startup Capacitive Load</b>					
Full load; ESR ≥ 1mΩ				1000	μF
Full load; ESR ≥ 10mΩ				2000	μF
<b>EFFICIENCY</b>					
V <sub>o</sub> =5V	V <sub>in</sub> =24V, I <sub>o</sub> =I <sub>o,max</sub>		93.0		%
V <sub>o</sub> =12V	V <sub>in</sub> =24V, I <sub>o</sub> =I <sub>o,max</sub>		97.0		%
V <sub>o</sub> =15V	V <sub>in</sub> =24V, I <sub>o</sub> =I <sub>o,max</sub>		97.0		%
<b>FEATURE CHARACTERISTICS</b>					
<b>Switching Frequency</b>					
ON/OFF Control, (Negative logic)			300		kHz
<b>Logic Low Voltage</b>					
Logic Low Voltage	Module On, V <sub>on/off</sub>	-0.3		1.2	V
Logic High Voltage	Module Off, V <sub>on/off</sub>	2.5		V <sub>in,max</sub>	V
<b>Logic Low Current</b>					
Logic Low Current	Module On, I <sub>on/off</sub>		10	30	uA
Logic High Current	Module Off, I <sub>on/off</sub>			1	mA
<b>ON/OFF Control, (Positive Logic)</b>					
Logic High Voltage	Module On, V <sub>on/off</sub>	V <sub>in</sub> -2.5		V <sub>in,max</sub>	V
Logic Low Voltage	Module Off, V <sub>on/off</sub>	-0.3		1.2	V
Logic High Current	Module On, I <sub>on/off</sub>		10	30	uA
Logic Low Current	Module Off, I <sub>on/off</sub>			1	mA
Remote Sense Range				0.5	V
<b>GENERAL SPECIFICATIONS</b>					
MTBF	I <sub>o</sub> =I <sub>o,max</sub> , T <sub>a</sub> =25°C		TBD		M hours
Weight			12		grams
Over-Temperature Shutdown	Refer to Figure 27 for the measuring point		125		°C



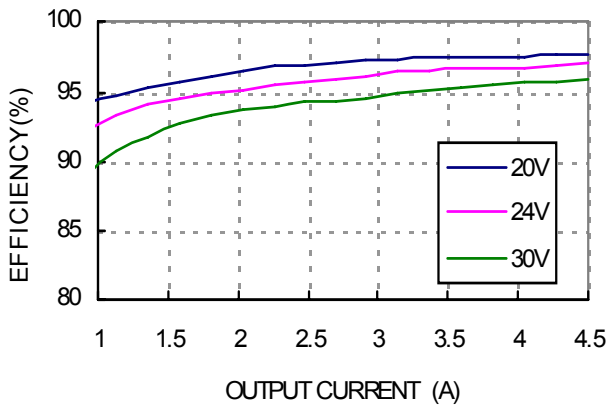
# ELECTRICAL CHARACTERISTICS CURVES



**Figure 1:** Converter efficiency vs. output current (5.0V output voltage)



**Figure 2:** Converter efficiency vs. output current (12V output voltage)



**Figure 3:** Converter efficiency vs. output current (15V output voltage)

# ELECTRICAL CHARACTERISTICS CURVES

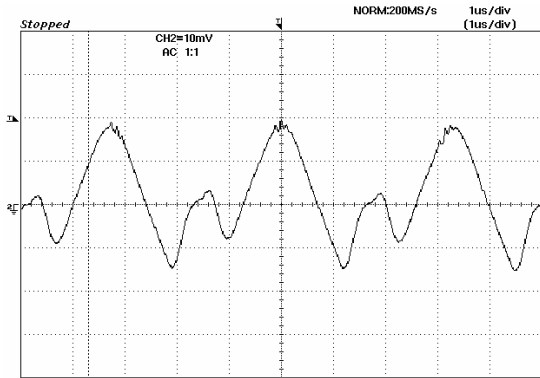


Figure 4: Output ripple & noise at 24Vin, 5.0V/10A out

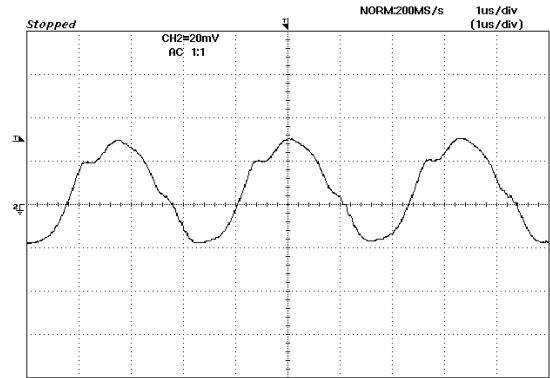


Figure 5: Output ripple & noise at 24Vin, 12V/6A out

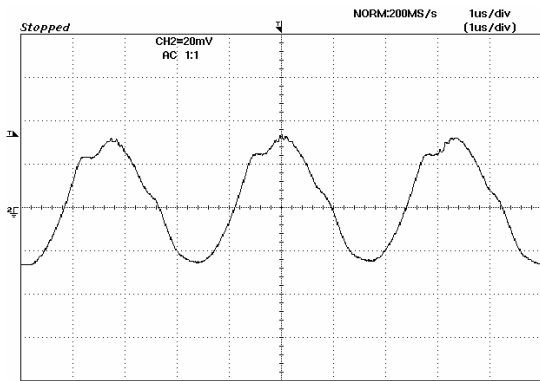


Figure 6: Output ripple & noise at 24Vin, 15V/4.5A out

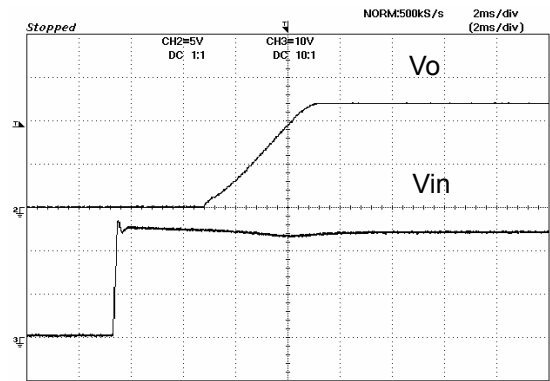


Figure 7: Turn on delay time at 24vin, 12V/6A out

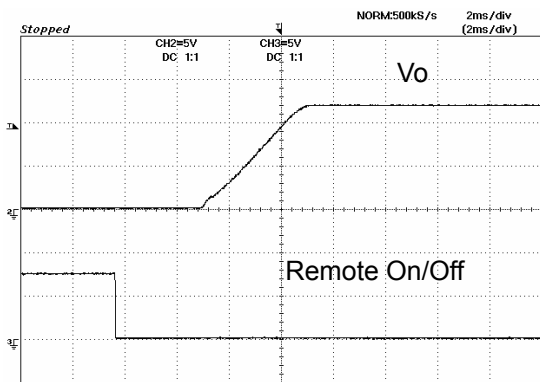


Figure 8: Turn on delay time at Remote On/Off, 12V/6A out

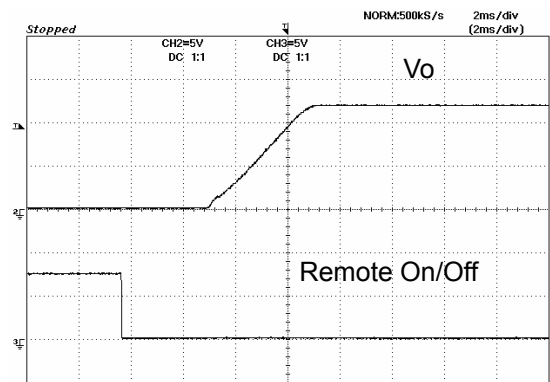
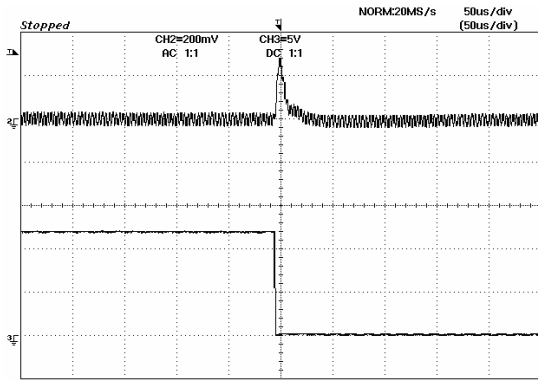


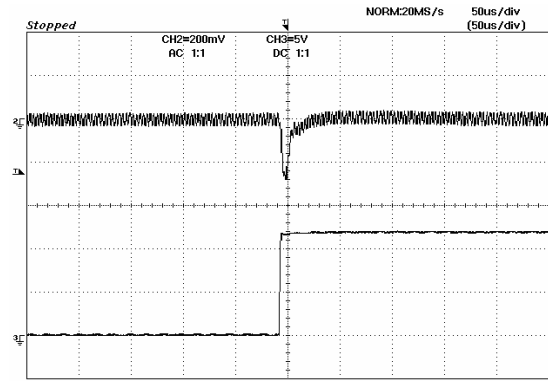
Figure 9: Turn on Using Remote On/Off with external capacitors ( $C_o = 2000 \mu F$ ), 12V/6A out



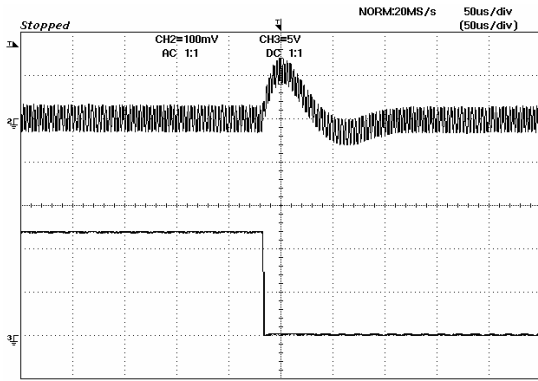
# ELECTRICAL CHARACTERISTICS CURVES



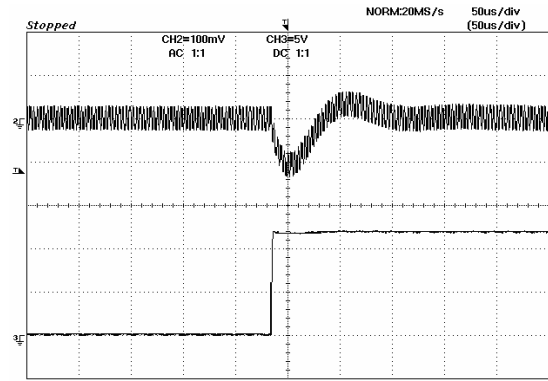
**Figure 10:** Typical transient response to step load change at  $5A/\mu S$  from 100% to 50% of  $I_o$ , max at  $24V_{in}$ ,  $12.0V_{out}$  ( $C_{out} = 1\mu F$  ceramic,  $10\mu F$  tantalum)



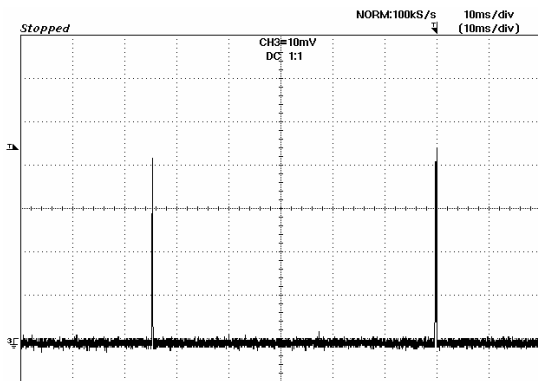
**Figure 11:** Typical transient response to step load change at  $5A/\mu S$  from 50% to 100% of  $I_o$ , max at  $24V_{in}$ ,  $12.0V_{out}$  ( $C_{out} = 1\mu F$  ceramic,  $10\mu F$  tantalum)



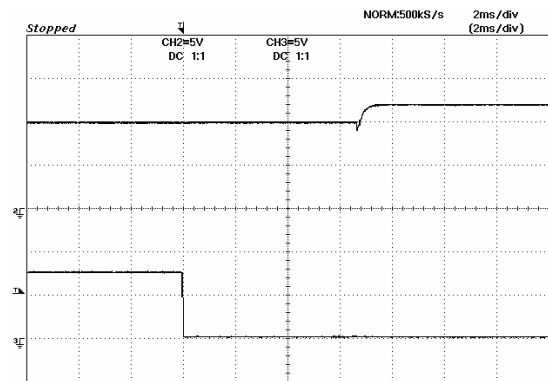
**Figure 12:** Typical transient response to step load change at  $5A/\mu S$  from 100% to 50% of  $I_o$ , max at  $24V_{in}$ ,  $12.0V_{out}$  ( $C_{out} = 1\mu F$  ceramic,  $10\mu F$  tantalum) with external  $2 \times 150\mu F$  OS-CON capacitors



**Figure 13:** Typical transient response to step load change at  $5A/\mu S$  from 50% to 100% of  $I_o$ , max at  $24V_{in}$ ,  $12.0V_{out}$  ( $C_{out} = 1\mu F$  ceramic,  $10\mu F$  tantalum) with external  $2 \times 150\mu F$  OS-CON capacitors

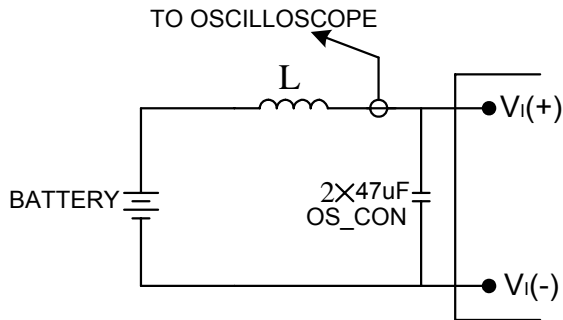


**Figure 14:** Output short circuit current  $24V_{in}$ ,  $5.0V_{out}$  ( $20A/div$ )



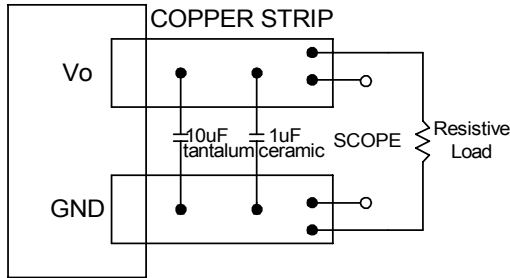
**Figure 15:** Turn on with Prebias  $24V_{in}$ ,  $12V/0A_{out}$ ,  $V_{bias} = 10.2V_{dc}$

## TEST CONFIGURATIONS



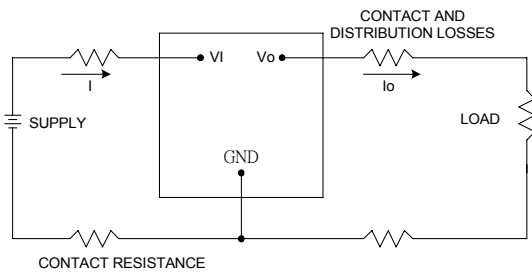
Note: Input reflected-ripple current is measured with a simulated source inductance. Current is measured at the input of the module.

**Figure 16:** Input reflected-ripple test setup



Note: Use a 10µF tantalum and 1µF capacitor. Scope measurement should be made using a BNC connector.

**Figure 17:** Peak-peak output noise and startup transient measurement test setup



**Figure 18:** Output voltage and efficiency measurement test setup

Note: All measurements are taken at the module terminals. When the module is not soldered (via socket), place Kelvin connections at module terminals to avoid measurement errors due to contact resistance.

$$\eta = \left( \frac{V_o \times I_o}{V_i \times I_i} \right) \times 100 \%$$

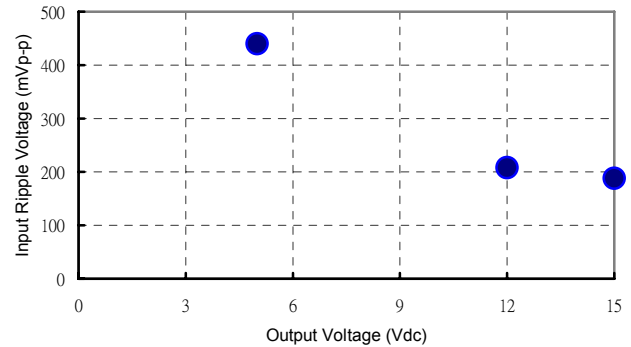
## DESIGN CONSIDERATIONS

### Input Source Impedance

To maintain low-noise and ripple at the input voltage, it is critical to use low ESR capacitors at the input to the module. Figure 19 shows the input ripple voltage (mVp-p) for various output models using 2x47 µF low ESR OS-CON capacitors (SANYO P/N:35SVPD47M, 47µF/35V or equivalent).

The input capacitance should be able to handle an AC Ripple current of at least:

$$I_{rms} = I_{out} \sqrt{\frac{V_{out}}{V_{in}} \left( 1 - \frac{V_{out}}{V_{in}} \right)} \quad A_{rms}$$



**Figure 19:** Input ripple voltage for various Output models,  $V_o=5V$   $I_o = 10A$  ·  $V_o=12V$   $I_o = 6A$  and  $V_o=15V$   $I_o = 4.5A$  ( $C_{in} = 2x47\mu F$  OS-CON capacitors at the input)

## DESIGN CONSIDERATIONS (CON.)

The power module should be connected to a low ac-impedance input source. Highly inductive source impedances can affect the stability of the module. An input capacitance must be placed close to the modules input pins to filter ripple current and ensure module stability in the presence of inductive traces that supply the input voltage to the module.

### Safety Considerations

For safety-agency approval the power module must be installed in compliance with the spacing and separation requirements of the end-use safety agency standards.

For the converter output to be considered meeting the requirements of safety extra-low voltage (SELV), the input must meet SELV requirements. The power module has extra-low voltage (ELV) outputs when all inputs are ELV.

The input to these units is to be provided with a maximum 15A of glass type fast-acting fuse in the ungrounded lead.

## FEATURES DESCRIPTIONS

### Remote On/Off

The DNM series power modules have an On/Off pin for remote On/Off operation. Both positive and negative On/Off logic options are available in the DNM series power modules.

For positive logic module, connect an open collector (NPN) transistor or open drain (N channel) MOSFET between the On/Off pin and the GND pin (see figure 20). Positive logic On/Off signal turns the module ON during the logic high and turns the module OFF during the logic low. When the positive On/Off function is not used, leave the pin floating or tie to  $V_{in}$  (module will be On).

For negative logic module, the On/Off pin is pulled high with an external pull-up resistor (see figure 21). Negative logic On/Off signal turns the module OFF during logic high and turns the module ON during logic low. If the negative On/Off function is not used, leave the pin floating or tie to GND. (module will be On)

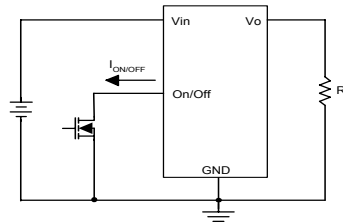


Figure 20: Positive remote On/Off implementation

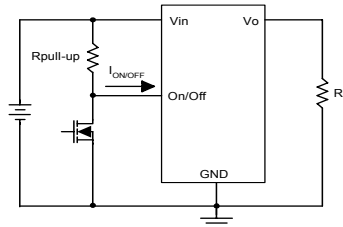


Figure 21: Negative remote On/Off implementation

### Over-Current Protection

To provide protection in an output over load fault condition, the unit is equipped with internal over-current protection. When the over-current protection is triggered, the unit enters hiccup mode. The units operate normally once the fault condition is removed.



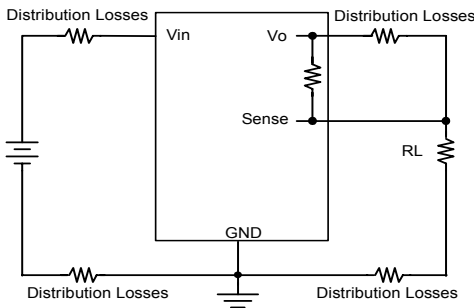
## FEATURES DESCRIPTIONS (CON.)

### Over-Temperature Protection

The over-temperature protection consists of circuitry that provides protection from thermal damage. If the temperature exceeds the over-temperature threshold the module will shut down. The module will try to restart after shutdown. If the over-temperature condition still exists during restart, the module will shut down again. This restart trial will continue until the temperature is within specification

### Remote Sense

The DNM provide  $V_o$  remote sensing to achieve proper regulation at the load points and reduce effects of distribution losses on output line. In the event of an open remote sense line, the module shall maintain local sense regulation through an internal resistor. The module shall correct for a total of 0.1V of loss. The remote sense line impedance shall be  $< 10\Omega$ .



**Figure 22:** Effective circuit configuration for remote sense operation

### Output Voltage Programming

The output voltage of the DNM can be programmed to any voltage between 5.021Vdc and 15.0Vdc by connecting one resistor (shown as  $R_{trim}$  in Figure 23) between the TRIM and GND pins of the module. Without this external resistor, the output voltage of the module is 5.021 Vdc. To calculate the value of the resistor  $R_{trim}$  for a particular output voltage  $V_o$ , please use the following equation:

$$R_{trim} = \left( \frac{10500}{V_o - 5.021} - 1000 \right) \cdot \Omega$$

$R_{trim}$  is the external resistor in  $\Omega$   
 $V_o$  is the desired output voltage

DS\_DNM24SIP10\_08142008

For example, to program the output voltage of the DNM module to 12Vdc,  $R_{trim}$  is calculated as follows:

$$R_{trim} = \left( \frac{10500}{6.979} - 1000 \right) \cdot \Omega$$

$$R_{trim} = 504.514\Omega$$

DNM can also be programmed by applying a voltage between the TRIM and GND pins (Figure 24). The following equation can be used to determine the value of  $V_{trim}$  needed for a desired output voltage  $V_o$ :

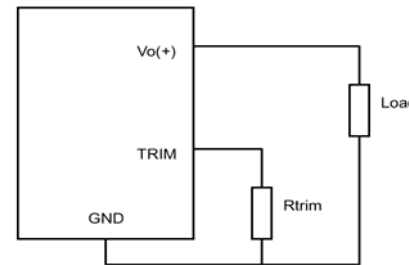
$$V_{trim} = 0.7 - [(V_o - 5.021) \cdot 0.0667]$$

$V_{trim}$  is the external voltage in V  
 $V_o$  is the desired output voltage

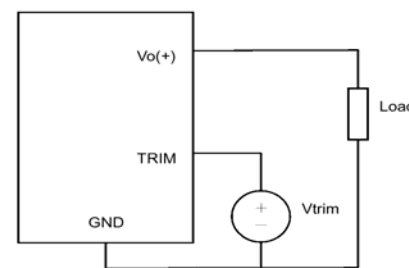
For example, to program the output voltage of a DNM module to 12 Vdc,  $V_{trim}$  is calculated as follows

$$V_{trim} = 0.7 - (6.979 \cdot 0.0667)$$

$$V_{trim} = 0.2345V$$



**Figure 23:** Circuit configuration for programming output voltage using an external resistor



**Figure 24:** Circuit Configuration for programming output voltage using external voltage source



## FEATURE DESCRIPTIONS (CON.)

Table 1 provides Rtrim values required for some common output voltages, while Table 2 provides value of external voltage source, Vtrim, for the same common output voltages. By using a  $\pm 0.5\%$  tolerance trim resistor, set point tolerance of  $\pm 2\%$  can be achieved as specified in the electrical specification.

**Table 1**

VO (V)	Rtrim ( $\Omega$ )
5.021	Open
12	504.514
15	52.21

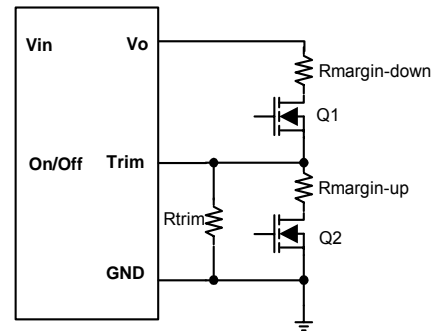
**Table 2**

VO (V)	Vtrim (V)
5.021	Open
12	0.2345
15	0.0344

The amount of power delivered by the module is the voltage at the output terminals multiplied by the output current. When using the trim feature, the output voltage of the module can be increased, which at the same output current would increase the power output of the module. Care should be taken to ensure that the maximum output power of the module must not exceed the maximum rated power ( $V_{o.set} \times I_{o.max} \leq P_{max}$ ).

### Voltage Margining

Output voltage margining can be implemented in the DNM modules by connecting a resistor,  $R_{margin-up}$ , from the Trim pin to the ground pin for margining-up the output voltage and by connecting a resistor,  $R_{margin-down}$ , from the Trim pin to the output pin for margining-down. Figure 25 shows the circuit configuration for output voltage margining. If unused, leave the trim pin unconnected. A calculation tool is available from the evaluation procedure which computes the values of  $R_{margin-up}$  and  $R_{margin-down}$  for a specific output voltage and margin percentage.



**Figure 25:** Circuit configuration for output voltage margining

## THERMAL CONSIDERATIONS

Thermal management is an important part of the system design. To ensure proper, reliable operation, sufficient cooling of the power module is needed over the entire temperature range of the module. Convection cooling is usually the dominant mode of heat transfer.

Hence, the choice of equipment to characterize the thermal performance of the power module is a wind tunnel.

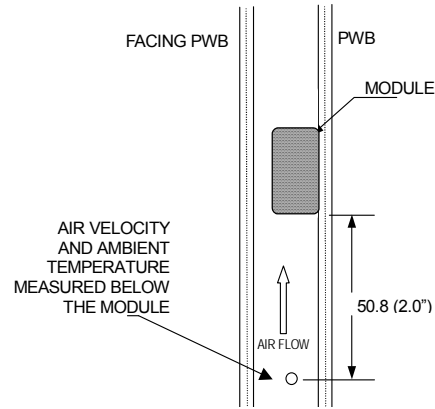
### Thermal Testing Setup

Delta's DC/DC power modules are characterized in heated vertical wind tunnels that simulate the thermal environments encountered in most electronics equipment. This type of equipment commonly uses vertically mounted circuit cards in cabinet racks in which the power modules are mounted.

The following figure shows the wind tunnel characterization setup. The power module is mounted on a test PWB and is vertically positioned within the wind tunnel. The height of this fan duct is constantly kept at 25.4mm (1").

### Thermal De-rating

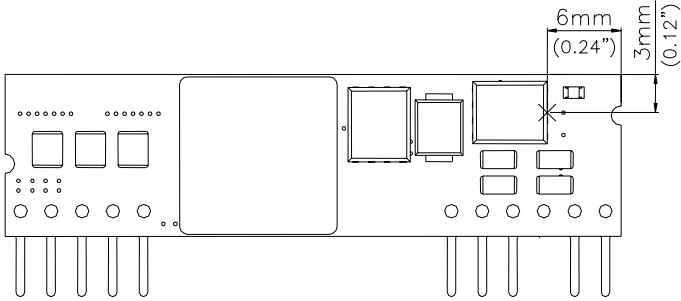
Heat can be removed by increasing airflow over the module. To enhance system reliability, the power module should always be operated below the maximum operating temperature. If the temperature exceeds the maximum module temperature, reliability of the unit may be affected.



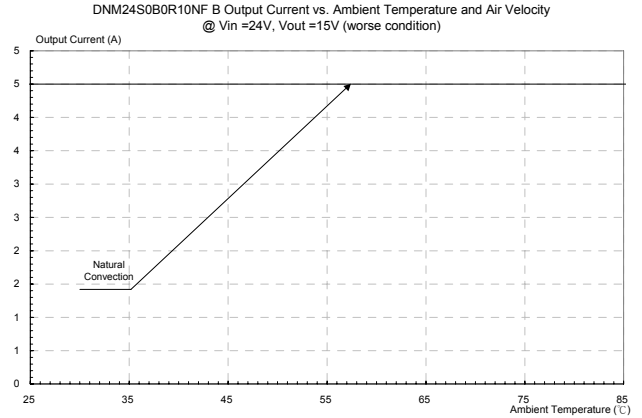
Note: Wind Tunnel Test Setup Figure Dimensions are in millimeters and (Inches)

Figure 26: Wind tunnel test setup

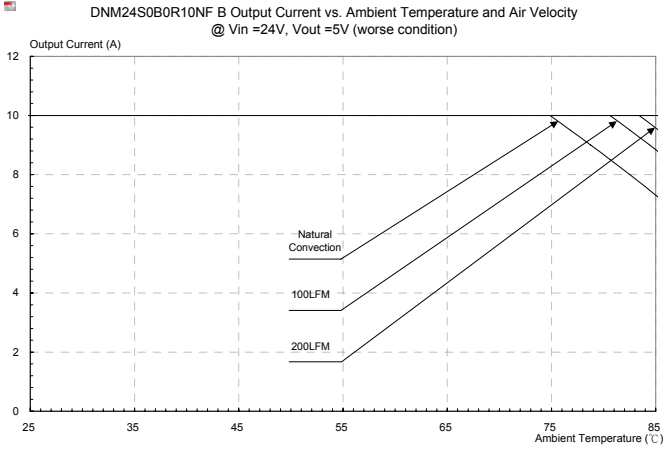
# THERMAL CURVES



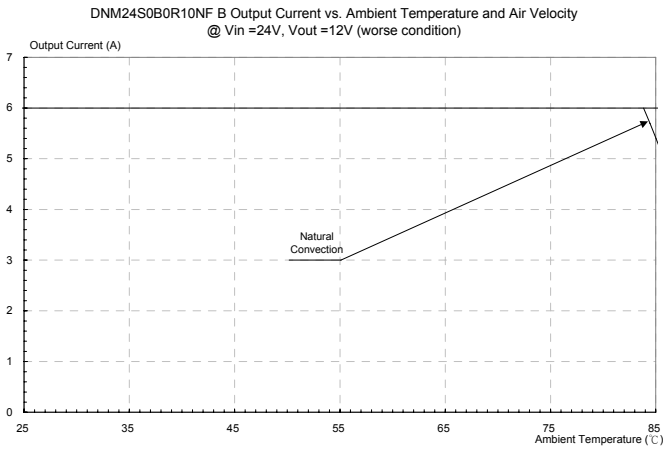
**Figure 27:** Temperature measurement location  
The allowed maximum hot spot temperature is defined at 125°C



**Figure 30:** Output current vs. ambient temperature and air velocity @ Vin=24V, Vo=15V (Either Orientation)



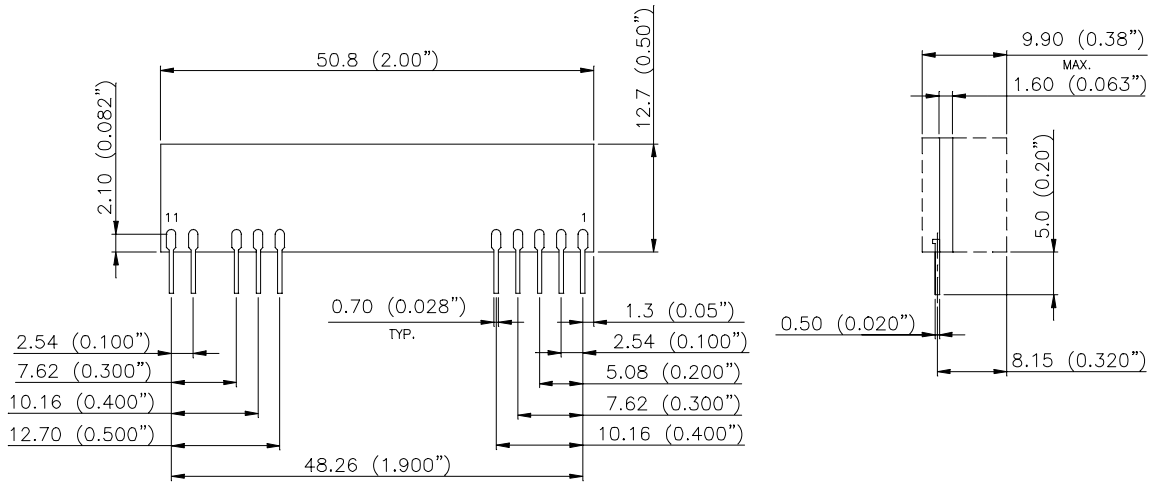
**Figure 28:** Output current vs. ambient temperature and air velocity @ Vin=24V, Vo=5.0V (Either Orientation)



**Figure 29:** Output current vs. ambient temperature and air velocity @ Vin=24V, Vo=12V (Either Orientation)

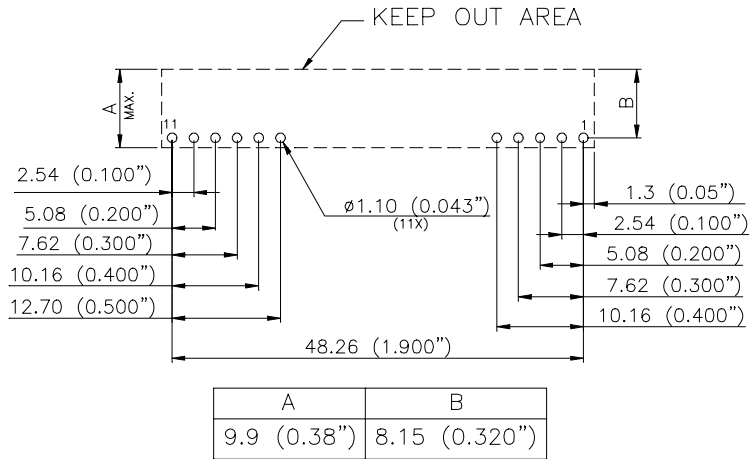
# MECHANICAL DRAWING

## SIP PACKAGE



BACK VIEW

SIDE VIEW



## RECOMMENDED P.W.B LAYOUT

### NOTES:

DIMENSIONS ARE IN MILLIMETERS AND (INCHES)

TOLERANCES: X.Xmm $\pm$ 0.5mm(X.XX in. $\pm$ 0.02 in.)

X.XXmm $\pm$ 0.25mm(X.XXX in. $\pm$ 0.010 in.)

## PART NUMBERING SYSTEM

DNM	24	S	0B0	R	10	N	F	B
Product Series	Input Voltage	Numbers of Outputs	Output Voltage	Package Type	Output Current	On/Off logic		Option Code
DNM ~ 10A	24 - 20~30V	S - Single	0B0 - Programmable	R - SIP	10 - 10A	N- negative P- positive	F- RoHS 6/6 (Lead Free)	B - No tracking pin

## MODEL LIST

Model Name	Packaging	Input Voltage	Output Voltage	Output Current	On/Off logic	Efficiency 24Vin @ 100% load
DNM24S0B0R10NFB	SIP	20V ~ 30V	5.0V ~ 15.0V	4.5A~10A	Negative	97% (12V/6A)

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