# 64-Channel Serial to Parallel Converter with P-Channel Open Drain Controllable Output Current 

## Features

- HVCMOS ${ }^{\circledR}$ technology
- 5.0V CMOS Logic
- Output voltage up to -85 V
- Output current source control
- 16MHz equivalent data rate
- Latched data outputs
- Forward and reverse shifting options (DIR pin)
- Diode to VDD allows efficient power recovery


## General Description

The HV57009 is a low-voltage serial to high-voltage parallel converter with P-channel open drain outputs. This device has been designed for use as a driver for plasma panels.

The device has two parallel 32 -bit shift registers, permitting data rates twice the speed of one (they are clocked together). There are also 64 latches and control logic to perform the blanking of the outputs. $\mathrm{HV}_{\text {out }} 1$ is connected to the first stage of the first shift register through the blanking logic. Data is shifted through the shift registers on the logic low to high transition of the clock. The DIR pin causes CCW shifting when connected to VSS, and CW shifting when connected to VDD. A data output buffer is provided for cascading devices. This output reflects the current status of the last bit of the shift register ( $\mathrm{HV}_{\mathrm{ouT}} 64$ ). Operation of the shift register is not affected by the $\overline{\mathrm{LE}}$ (latch enable), or the $\overline{\mathrm{BL}}$ (blanking) inputs. Transfer of data from the shift registers to latches occurs when the $\overline{\mathrm{LE}}$ input is high. The data in the latches is stored when $\overline{\mathrm{LE}}$ is low.

The HV570 has 64 channels of output constant current sourcing capability. They are adjustable from 0.1 to 2.0 mA through one external resistor or a current source.

## Functional Block Diagram



Each SR (shift register) provides 32 outputs. SR1 supplies outputs 1 to 32 and SR2 supplies outputs 33 to 64 .

## Ordering Information

| Part Number | Package Option | Packing |
| :--- | :--- | :--- |
| HV57009PG-G | 80 -Lead PQFP | $66 /$ Tray |

-G denotes a lead (Pb)-free / RoHS compliant package

## Absolute Maximum Ratings

| Parameter | Value |
| :--- | ---: |
| Supply voltage, $\mathrm{V}_{\mathrm{DD}}{ }^{1}$ | -0.5 V to +7.5 V |
| Output voltage, $\mathrm{V}_{\mathrm{NN}}{ }^{1}$ | $\mathrm{~V}_{\mathrm{DD}}+0.5 \mathrm{~V}$ to -95 V |
| Logic input levels ${ }^{1}$ | -0.3 V to $\mathrm{V}_{\mathrm{DD}}+0.3 \mathrm{~V}$ |
| Ground current ${ }^{2}$ | 1.5 A |
| Continuous total power dissipation ${ }^{3}$ | 1200 mW |
| Operating temperature range | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |
| Storage temperature range | $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ |

Absolute Maximum Ratings are those values beyond which damage to the device may occur. Functional operation under these conditions is not implied. Continuous operation of the device at the absolute rating level may affect device reliability.

## Notes:

1. All voltages are referenced to $V_{\text {ss }}$
2. Duty cycle is limited by the total power dissipated in the package.
3. For operation above $25^{\circ} \mathrm{C}$ ambient derate linearly to maximum operating temperature at $20 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$.

## Pin Configuration



## Product Marking



Package may or may not include the following marks: Si or $4 i$
80-Lead PQFP
Typical Thermal Resistance

| Package | $\boldsymbol{\theta}_{j a}$ |
| :--- | :--- |
| 80 -Lead PQFP | $37^{\circ} \mathrm{C} / \mathrm{W}$ |

Recommended Operating Conditions

| Sym | Parameter | Min | Max | Units |
| :---: | :---: | :---: | :---: | :---: |
| $V_{D D}$ | Logic supply voltage | 4.5 | 5.5 | V |
| $\mathrm{HV}_{\text {OUT }}$ | HV output off voltage | -85 | $V_{D D}$ | V |
| $\mathrm{V}_{\text {IH }}$ | High-level input voltage | $\mathrm{V}_{\mathrm{DD}}-1.2 \mathrm{~V}$ | $V_{D D}$ | V |
| $\mathrm{V}_{\text {IL }}$ | Low-level input voltage | 0 | 1.2 | V |
| $\mathrm{f}_{\text {CLK }}$ | Clock frequency per register | DC | 8.0 4.5 | MHz |
| $\mathrm{T}_{\mathrm{A}}$ | Operating free-air temperature | -40 | +85 | ${ }^{\circ} \mathrm{C}$ |

Notes:
Power-up sequence should be the following:

1. Connect ground
2. Apply $V_{D D}$
3. Set all inputs to a known state

Power-down sequence should be the reverse of the above.

DC Electrical Characteristics (All voltages are referenced to $V_{s s} V_{s s}=0, T_{A}=25^{\circ} \mathrm{C}$ )

| Sym | Parameter |  | Min | Max | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{I}_{\mathrm{DD}}$ | $\mathrm{V}_{\text {DD }}$ supply current |  | - | 15 | mA | $\mathrm{V}_{\mathrm{DD}}=\mathrm{V}_{\mathrm{DD}}$ max, $\mathrm{f}_{\mathrm{CLK}}=8.0 \mathrm{MHz}$ |
| $\mathrm{I}_{\text {NN }}$ | High voltage supply current |  | - | -10 | $\mu \mathrm{A}$ | Outputs off, $\mathrm{HV}_{\text {OUT }}=-85 \mathrm{~V}$ (total of all outputs) |
| $\mathrm{I}_{\text {DDQ }}$ | Quiescent $\mathrm{V}_{\text {D }}$ supply current |  | - | 100 | $\mu \mathrm{A}$ | $\begin{aligned} & \text { All inputs }=V_{\text {D }} \\ & \text { except }+I N=V_{S S}=G N D \end{aligned}$ |
| $\mathrm{V}_{\text {OH }}$ | High level output | Data Out | $\mathrm{V}_{\mathrm{DD}}-0.5 \mathrm{~V}$ | - | V | $\mathrm{I}_{0}=-100 \mu \mathrm{~A}$ |
|  |  | HV ${ }_{\text {OUT }}$ | +1.0 | $\mathrm{V}_{\mathrm{DD}}$ | V | $\mathrm{I}_{0}=-2.0 \mathrm{~mA}$ |
| $\mathrm{V}_{\mathrm{OL}}$ | Low level output | Data Out | - | +0.5 | V | $\mathrm{I}_{0}=100 \mu \mathrm{~A}$ |
| $\mathrm{I}_{\mathrm{H}}$ | High-level logic input current |  | - | 1.0 | $\mu \mathrm{A}$ | $\mathrm{V}_{I H}=\mathrm{V}_{\text {DD }}$ |
| $1 /$ | Low-level logic input current |  | - | -1.0 | $\mu \mathrm{A}$ | $\mathrm{V}_{\mathrm{LL}}=0 \mathrm{~V}$ |
| $\mathrm{I}_{\mathrm{cs}}$ | High output source current |  | - | -2.0 | mA | $\mathrm{V}_{\text {REF }}=2.0 \mathrm{~V}, \mathrm{R}_{\text {EXT }}=1.0 \mathrm{~K} \Omega \text {, }$ <br> see Figures 1a and 1b |
|  |  |  | -0.1 | - |  | $\mathrm{V}_{\text {REF }}=0.1 \mathrm{~V}, \mathrm{R}_{\text {EXT }}=1.0 \mathrm{~K} \Omega \text {, }$ <br> see Figures 1a and 1b |
| $\Delta l_{\text {cs }}$ | HV output source current for $\mathrm{I}_{\text {REF }}=2.0 \mathrm{~mA}$ |  | - | 10 | \% | $\mathrm{V}_{\text {REF }}=2.0 \mathrm{~V}, \mathrm{R}_{\text {EXT }}=1.0 \mathrm{~K} \Omega$ |

Note:
Current going out of the chip is considered negative.
AC Electrical Characteristics (Logic signal inputs and data inputs have $t_{f} t_{t} \leq 5 n s[10 \%$ and $90 \%$ points] for measurements)

| Sym | Parameter | Min | Max | Units | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{f}_{\text {CLK }}$ | Clock frequency | DC | 8.0 | MHz | Per register |
|  |  |  | 4.5 |  | When cascading devices |
| $\mathrm{t}_{\mathrm{wL}}, \mathrm{t}_{\mathrm{wH}}$ | Clock width high or low | 62 | - | ns | --- |
| $\mathrm{t}_{\text {su }}$ | Data set-up time before clock rises | 20 | - | ns | --- |
| $\mathrm{t}_{\mathrm{H}}$ | Data hold time after clock rises | 15 | - | ns | --- |
| $\mathrm{t}_{\text {ON }}, \mathrm{t}_{\text {OFF }}$ | Time from latch enable to $\mathrm{HV}_{\text {out }}$ | - | 500 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}$ |
| $\mathrm{t}_{\text {DHL }}$ | Delay time clock to data high to low | - | 150 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}$ |
| $\mathrm{t}_{\mathrm{DLH}}$ | Delay time clock to data low to high | - | 150 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}$ |
| $\mathrm{t}_{\text {DLE }}$ | Delay time clock to $\overline{L E}$ low to high | 45 | - | ns | --- |
| $t_{\text {WLE }}$ | LE pulse width | 25 | - | ns | --- |
| $\mathrm{t}_{\text {sLE }}$ | $\overline{\text { LE }}$ set-up time before clock rises | 0 | - | ns | --- |
| $\mathrm{t}_{\mathrm{r}}, \mathrm{t}_{\mathrm{f}}$ | Max. allowable clock rise and fall time (10\% and 90\% points) | - | 100 | ns | --- |

## Input and Output Equivalent Circuits



## Shift Register Operation



## Switching Waveforms



## Function Table

| Function | Inputs |  |  |  |  | Outputs |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Data In | CLK | LE | BL | DIR | Shift Reg | HV Outputs | Data Out |
| All O/P high | X | X | X | L | X | * | ON | * |
| Data falls through (latches transparent) | L | _ ${ }^{-}$ | H | H | X | L.....L | ON | L |
|  | H | _ ${ }^{-}$ | H | H | X | H..... H | OFF | H |
| Data stored in latches | X | X | L | H | X | * | Inversion of stored data | * |
| I/O relation | $\mathrm{D}_{10} 1$ 1-2A | _ ${ }^{-}$ | H | H | H | $\mathrm{Q}_{\mathrm{n}} \rightarrow \mathrm{Q}_{\mathrm{n}+1}$ | New ON or OFF | $\mathrm{D}_{10} 1-2 \mathrm{~B}$ |
|  | $\mathrm{D}_{110} 1-2 \mathrm{~A}$ | _ $\uparrow^{-}$ | L | H | H | $Q_{n} \rightarrow Q_{n+1}$ | Previous ON or OFF | $\mathrm{D}_{110} 1-2 \mathrm{~B}$ |
|  | $\mathrm{D}_{110} 1-2 \mathrm{~B}$ | _- ${ }^{-}$ | L | H | L | $\mathrm{Q}_{\mathrm{n}} \rightarrow \mathrm{Q}_{\mathrm{n}-1}$ | Previous ON or OFF | $\mathrm{D}_{10} 1-2 \mathrm{~A}$ |
|  | $\mathrm{D}_{110} 1-2 \mathrm{~B}$ | _ ${ }^{-}$ | H | H | L | $\mathrm{Q}_{\mathrm{n}} \rightarrow \mathrm{Q}_{\mathrm{n}-1}$ | New ON or OFF | $\mathrm{D}_{10} 1$ 1-2A |

## Note:

* $=$ dependent on previous stage's state. See Figure 7 for DIN and DOUT pin designation for CW and CCW shift.
$H=V_{D D}$ (Logic) $V_{N N}$ (HV Outputs)
$L=V_{S S}$


## Typical Current Programing Circuits



Figure 1a: Negative Control


Figure 1b: Positive Control
*Required if $R_{E X T}>10 \mathrm{~K} \Omega$ or $R_{E X T}$ is replaced by a constant current source.

Since:

$$
I_{\text {OUT }}=I_{\text {REF }}=\left|V_{R E F}\right| / R_{E X T}
$$

Therefore:

$$
\begin{aligned}
& \text { If } I_{\text {OUT }}=2.0 \mathrm{~mA} \text { and } V_{\text {REF }}=-5.0 \mathrm{~V} \rightarrow R_{\text {EXT }}=2.5 \mathrm{~K} \Omega . \\
& \text { If } I_{\text {OUT }}=1.0 \mathrm{~mA} \text { and } R_{E X T}=1.0 \mathrm{~K} \Omega \rightarrow V_{R E F}=-1.0 \mathrm{~V} .
\end{aligned}
$$

If $R_{E X T}>10 K \Omega$, add series network $R_{D}$ and $C_{D}$ to ground for stability as shown.

This control method behaves linearly as long as the operational amplifier is not saturated. However, it requires a negative power source and needs to provide a current $I_{\text {REF }}=I_{\text {OUT }}$ for each HV570 chip being controlled.

If $\mathrm{HV}_{\text {OUT }} \geq+1.0 \mathrm{~V}$, the $\mathrm{HV}_{\text {OUT }}$ cascade may no longer operate as a perfect current source, and the output current will
diminish. This effect depends on the magnitude of the output current.
Given $\mathrm{I}_{\mathrm{OUT}}$ and $\mathrm{V}_{\text {REF }}$, the $\mathrm{R}_{\mathrm{EXT}}$ can be calculated by using:

$$
R_{E X T}=V_{\text {REF }} / I_{\text {REF }}=V_{\text {REF }} / I_{\text {OUT }}
$$

The intersection of a set of $I_{\text {OUT }}$ and $V_{\text {REF }}$ values can be located in the graph shown below. The value picked for $\mathrm{R}_{\mathrm{EXT}}$ must always be in the shaded area for linear operation. This control method has the advantage that $\mathrm{V}_{\text {REF }}$ is positive, and draws only leakage current. If $R_{\text {EXT }}>10 \mathrm{~K} \Omega$, add series network $R_{D}$ and $C_{D}$ to ground for stability as shown.

## Note:

Lower reference current $I_{\text {REP }}$ results in higher distortion, $\Delta l_{C S}$, on the output.


## Pin Function

| Pin \# | Function | Pin \# | Function | Pin \# | Function | Pin \# | Function |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | HV ${ }_{\text {out }} 24$ | 21 | $\mathrm{HV}_{\text {out }} 4$ | 41 | HV ${ }_{\text {OUT }} 64$ | 61 | $\mathrm{HV}_{\text {Out }} 44$ |
| 2 | HV ${ }_{\text {Out }} 23$ | 22 | $\mathrm{HV}_{\text {OUT }} 3$ | 42 | $\mathrm{HV}_{\text {OUT }} 63$ | 62 | $\mathrm{HV}_{\text {Out }} 43$ |
| 3 | $\mathrm{HV}_{\text {out }} 22$ | 23 | $\mathrm{HV}_{\text {out }}{ }^{2}$ | 43 | $\mathrm{HV}_{\text {Out }} 62$ | 63 | $\mathrm{HV}_{\text {Out }} 42$ |
| 4 | $\mathrm{HV}_{\text {out }} 21$ | 24 | $\mathrm{HV}_{\text {out }} 1$ | 44 | $\mathrm{HV}_{\text {OUT }} 61$ | 64 | $\mathrm{HV}_{\text {Out }} 41$ |
| 5 | HV ${ }_{\text {out }} 20$ | 25 | $\mathrm{D}_{110} 1 \mathrm{~A}$ | 45 | $\mathrm{HV}_{\text {Out }} 60$ | 65 | HV ${ }_{\text {OUT }} 40$ |
| 6 | $\mathrm{HV}_{\text {OUT }} 19$ | 26 | $\mathrm{D}_{110} 2 \mathrm{~A}$ | 46 | $\mathrm{HV}_{\text {Out }} 59$ | 66 | $\mathrm{HV}_{\text {Out }} 39$ |
| 7 | HV OUT 18 | 27 | NC | 47 | $\mathrm{HV}_{\text {Out }} 58$ | 67 | $\mathrm{HV}_{\text {Out }} 38$ |
| 8 | $\mathrm{HV}_{\text {OUT }} 17$ | 28 | NC | 48 | $\mathrm{HV}_{\text {Out }} 57$ | 68 | $\mathrm{HV}_{\text {OUT }} 37$ |
| 9 | HV out 16 | 29 | $\overline{\text { LE }}$ | 49 | $\mathrm{HV}_{\text {Out }} 56$ | 69 | $\mathrm{HV}_{\text {Out }} 36$ |
| 10 | $\mathrm{HV}_{\text {OUT }} 15$ | 30 | CLK | 50 | $\mathrm{HV}_{\text {Out }} 55$ | 70 | $\mathrm{HV}_{\text {OUT }} 35$ |
| 11 | $\mathrm{HV} \mathrm{out}^{14}$ | 31 | $\overline{\text { BL }}$ | 51 | $\mathrm{HV}_{\text {OuT }} 54$ | 71 | $\mathrm{HV}_{\text {Out }} 34$ |
| 12 | HV ${ }_{\text {OUT }} 13$ | 32 | VSS | 52 | $\mathrm{HV}_{\text {OUT }} 53$ | 72 | $\mathrm{HV}_{\text {OUT }} 33$ |
| 13 | $\mathrm{HV}_{\text {OUT }} 12$ | 33 | DIR | 53 | $\mathrm{HV}_{\text {out }} 52$ | 73 | $\mathrm{HV}_{\text {Out }} 32$ |
| 14 | HV ${ }_{\text {OUT }} 11$ | 34 | VDD | 54 | $\mathrm{HV}_{\text {Out }} 51$ | 74 | $\mathrm{HV}_{\text {Out }} 31$ |
| 15 | HV ${ }_{\text {OUT }} 10$ | 35 | -IN | 55 | $\mathrm{HV}_{\text {Out }} 50$ | 75 | $\mathrm{HV}_{\text {Out }} 30$ |
| 16 | $\mathrm{HV}_{\text {out }} 9$ | 36 | $\mathrm{D}_{110} 2 \mathrm{~B}$ | 56 | $\mathrm{HV}_{\text {out }} 49$ | 76 | $\mathrm{HV}_{\text {out }} 29$ |
| 17 | HV ${ }_{\text {Out }} 8$ | 37 | $\mathrm{D}_{10} 1 \mathrm{~B}$ | 57 | $\mathrm{HV}_{\text {out }} 48$ | 77 | $\mathrm{HV}_{\text {out }} 28$ |
| 18 | HV ${ }_{\text {Out }} 7$ | 38 | NC | 58 | $\mathrm{HV}_{\text {OUT }} 47$ | 78 | $\mathrm{HV}_{\text {Out }} 27$ |
| 19 | $\mathrm{HV}_{\text {Out }}{ }^{6}$ | 39 | +IN | 59 | $\mathrm{HV}_{\text {out }} 46$ | 79 | $\mathrm{HV}_{\text {Out }} 26$ |
| 20 | $\mathrm{HV}_{\text {Out }} 5$ | 40 | VBP | 60 | $\mathrm{HV}_{\text {Out }} 45$ | 80 | $\mathrm{HV}_{\text {out }} 25$ |

## Notes:

1. Pin designation for $D I R=V D D$.
2. A $0.1 \mu F$ capacitor is needed between VDD and VBP (pin 40) for better output current stability and to prevent transient cross-coupling between outputs. See Figures 1a and 1b.

## 80-Lead PQFP Package Outline (PG)

### 20.00x14.00mm body, 3.40 mm height (max), 0.80 mm pitch, 3.90 mm footprint



## Top View



Note:

1. A Pin 1 identifier must be located in the index area indicated. The Pin 1 identifier can be: a molded mark/identifier; an embedded metal marker; or a printed indicator.

| Symbol |  | A | A1 | A2 | b | D | D1 | E | E1 | e | L | L1 | L2 | $\boldsymbol{\theta}$ | 01 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Dimension (mm) | MIN | 2.80* | 0.25 | 2.55 | 0.30 | 23.65* | 19.80* | 17.65* | 13.80* | $\begin{aligned} & 0.80 \\ & \text { BSC } \end{aligned}$ | 0.73 | $\begin{aligned} & 1.95 \\ & \text { REF } \end{aligned}$ | $\begin{aligned} & 0.25 \\ & \text { BSC } \end{aligned}$ | $0^{\circ}$ | $5^{\circ}$ |
|  | NOM | - | - | 2.80 | - | 23.90 | 20.00 | 17.90 | 14.00 |  | 0.88 |  |  | $3.5^{\circ}$ | - |
|  | MAX | 3.40 | 0.50* | 3.05 | 0.45 | 24.15* | 20.20* | 18.15* | 14.20* |  | 1.03 |  |  | $7^{\circ}$ | $16^{\circ}$ |

JEDEC Registration MO-112, Variation CB-1, Issue B, Sept. 1995.

* This dimension is not specified in the JEDEC drawing.


## Drawings not to scale.

Supertex Doc. \#: DSPD-80PQFPPG, Version C041309.
(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to http://www.supertex.com/packaging.html.)

[^0]
[^0]:    Supertex inc. does not recommend the use of its products in life support applications, and will not knowingly sell them for use in such applications unless it receives an adequate "product liability indemnification insurance agreement." Supertex inc. does not assume responsibility for use of devices described, and limits its liability to the replacement of the devices determined defective due to workmanship. No responsibility is assumed for possible omissions and inaccuracies. Circuitry and specifications are subject to change without notice. For the latest product specifications refer to the Supertex inc. (website: http//www.supertex.com)

