

# Features

- Specially Optimized for the EMI
- Internal 500V Power MOSFET
- Active PFC for High Power Factor and Low THD
- Critical Conduction Mode Operation
- System Efficiency up to 95%
- Without auxiliary winding
- Excellent Line and Load Regulation
- ±3% LED Output Current Accuracy
- Thermal Regulation Function
- LED Short/Open Protection
- Current Sensing Resistor Open Protection
- Auto Fault Recovery
- Available in SOP7 Package

## **Applications**

- E27 LED Bulb, Spot Light
- PAR30, PAR38 Lamp
- T8/T10 LED String
- Other LED Lighting

# Description

The WS3411 is a high precision non-isolated buck driver with active PFC, specially designed for universal input offline constant current LED lighting. The driver with on-chip PFC circuit achieves high power factor and low THD. Operating in critical conduction mode, the power MOSFET switching loss is reduced and the inductor is fully utilized.

Non-isolated Buck APFC Offline LED Driver

The WS3411 integrates a 500V power MOSFET. With few external components, the LED output current can be precisely controlled.

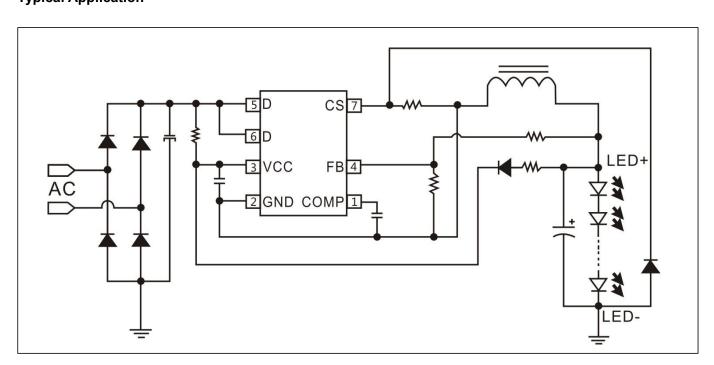
The WS3411 offers rich protection functions to improve the system reliability, including LED open circuit protection, LED short circuit protection, CS resistor open circuit protection and cycle by cycle current limit. All the protection functions are auto-recovery. The system reliability is further improved by the thermal regulation function. The output current is reduced when the driver is over temperature.

The WS3411 has ultra low start up current and ultra low operating current.

The WS3411 is available in SOP7 Package

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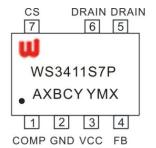
## **Typical Application**





## **Pin Configuration and Marking Information**

The WS3411 is available in SOP7 Package, the top marking is shown as below:

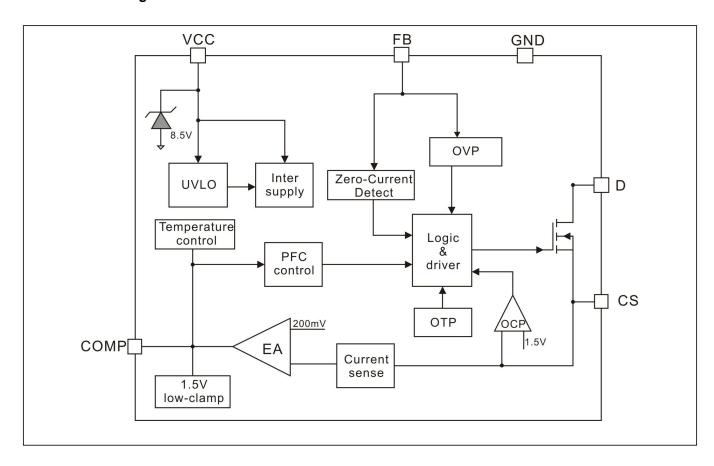


WS3411S7P
A: Product Code
X: Internal Code
BCY: Internal Code For QC
YMX: D/C

### **Pin Definition**

Name	Pin No.	Description	
COMP	4	Loop Compensation Node. This pin connects a capacitor to GND for stabilization of the	
COIVIE		control loop, achieving accurate LED current, high Power Factor and low THD.	
GND	2	Ground.	
VCC	3	Power Supply Pin. Connect a bypass capacitor from this pin to GND.	
ED. 4		Feedback Voltage Input Pin. This pin detects the inductor demagnetization signal and the	
FB	4	output voltage.	
D	5/6	Internal HV Power MOSFET Drain.	
CS	7	Current Sense Pin. Connect a resistor to GND to sense the inductor current.	

## **Internal Block Diagram**





# **Ordering Information**

Package	Marking	Part Number
7-Pin SOP-7, Pb-free	WS3411S7P	WS3411S7P

# **Recommended Operation Conditions**

Part Number	Package	Input voltage Maximum ou power	
WS3411S7P	SOP7	175VAC-264VAC	80V/150mA
W3341137F		90VAC-264VAC	50V/150mA

# **Absolute Maximum Ratings**

symbol	paramete	Range	unit
V <sub>CC</sub>	Vcc pin voltage	Vcc pin voltage -0.3~8	
V <sub>D</sub>	Internal HV MOSFET Drain to Source voltage	-0.3~500	V
V <sub>CS</sub>	Current sense pin input voltage	-0.3~7	V
$V_{FB}$	Feedback pin input voltage	-0.3~7	V
V <sub>COMP</sub>	Compensation pin voltage	-0.3~7	V
P <sub>DMAX</sub>	Power dissipation	0.5	W
TJ	Operating junction temperature	165	$^{\circ}$ C
T <sub>STG</sub>	Storage temperature range	-55~165	$^{\circ}$

**Note :** Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device.



# **Electrical Characteristics** (Unless otherwise specified, VCC=8V and TA =25 $\,^{\circ}$ C)

Symbol	Parameter	Conditions	Min	Тур	Max	Units
Supply V	oltage Section			<u> </u>		_
V <sub>cc_clamp</sub>	VCC Clamp Voltage		7.5	8.0	8.5	V
I <sub>cc_clamp</sub>	VCC Clamp Current			5		mA
V <sub>CC_ST</sub>	VCC Turn On Threshold	VCC Rising	7.07	7.57	8.07	V
V <sub>uvlo_HYS</sub>	VCC Turn Off hysteresis	VCC Falling		1.32		V
I <sub>st</sub>	VCC Startup Current	VCC <vcc_st-0.5v< td=""><td></td><td>70</td><td>100</td><td>uA</td></vcc_st-0.5v<>		70	100	uA
I <sub>op</sub>	VCC Operating Current			400		uA
Current S	Sense Section					
	Over current			4		.,
Vocp	Protectionthreshold			1		V
+	Leading Edge Blanking Time			0.50		
$T_LEB$	for Current Sense			350		ns
Td	Switch off Delay Time			200		ns
Compens	sation Section		<u>'</u>	<u>'</u>	<u> </u>	<u> </u>
$V_{REF}$	Internal Reference Voltage		194	200	206	mV
V <sub>CL</sub>	COMP Low Clamp Voltage			1.5		V
V <sub>CH</sub>	COMP High Clamp Voltage			4		V
Internal c	Iriver					
T <sub>OFF_MIN</sub>	Minimum Off Time			3		us
T <sub>ON_MAX</sub>	Maximum On Time			40		us
Feedback	Section					
.,	FB Over Voltage Protection			4.0		V
$V_{FB}$	Threshold			1.6		v
	FB Falling Edge			0.0		
V <sub>ZCD</sub>	Threshold Voltage			0.2		V
Power M	OSFET					
D	Static Drain-source			8	10	Ω
R <sub>DSON</sub>	On-resistance			0	10	1 12
$V_{DS}$	Drain-Source Breakdown		500			V
<b>v</b> DS	Voltage		300			v
Thermal	Regulation Section					
T <sub>REG</sub>	Thermal Regulation			150		°C
	Temperature			130		



## **Application Information**

The WS3411 is a high precision Active PFC driver integrating 500V power MOSFET, specially designed for non-isolated buck offline constant current LED lighting. Operating in critical conduction mode, the driver achieves high power factor, low THD and high efficiency.

### Start Up

After system powered up, the VCC pin capacitor is charged up by the start up resistor. When the VCC pin voltage reaches the turn on threshold, the internal circuits start working. The COMP pin voltage is pulled up to 1.5V quickly, then the WS3411starts switching. The system works at 10k Hz frequency at the beginning, the COMP voltage rises up gradually, and the inductor peak current also rises up. The LED current hence achieves a soft start without overshoot. After the output voltage is built up, the VCC power is supplied by the output voltage through a diode.

The maximum value of the startup current is 100uA. For the application of 85VAC—264VAC, the startup resistor can be calculated by the equation:

$$R_{st} < \frac{V_{in\_\min}}{I_{st\_\max}} \approx \frac{85V}{100uA} = 850K\Omega$$

### **Constant Current Control**

The WS3411 utilizes floating ground structure. The inductor current is sensed during the whole switching cycle, thus it achieves high precision output current control and also excellent line and load regulation.

The current in LED can be calculated by the equation:

$$I_O = \frac{V_{CS}}{R_{CS}} = \frac{200mV}{Rcs}$$

Rcs is the current sensing resistor.

#### **Feedback Network**

The WS3411 senses the output current zero crossing information through the feedback network, the FB falling threshold voltage is set to 0.2V with 0.15V hysteresis. The FB pin is also used to detect output OVP, the threshold voltage is 1.6V. The ratio of FB upper resistor to lower resistor can be set as:

$$\frac{R_{FBL}}{R_{FBL} + R_{FBH}} = \frac{1.6V}{V_{OVP}}$$

Where.

RFBL: The lower resistor of the feedback network

RFBH: The upper resistor of the feedback network

VOVP: Output over voltage setting point

The minimum VOVP should be set about 1.3Vo.It is recommended that the FB lower resistor is set to  $3.3K\Omega$ — $5.1K\Omega$ paralleled with a 22pF ceramic capacitor.

### **Protection Functions**

The WS3411 offers rich protection functions to improve the system reliability.

When the LED is open circuit, the output voltage will gradually rise up. The output voltage is sensed by the FB pin when power MOSFET is turned off. When FB voltage reaches the OVP threshold, it will trigger fault logic and the system stops switching.

When the LED is shorted circuit, the system will work under 10k Hz switching frequency. Meanwhile, the output voltage is low and the VCC pin cannot be charged up by the output voltage, so the VCC pin voltage will gradually decrease and finally reaches the UVLO threshold.

After the system enters into fault condition, the VCC voltage will decrease until it reaches UVLO threshold. Then the system will re-start again. If the fault condition is removed, the system will resume normal operation.

When the output is short circuit or the inductor is saturated, the CS peak voltage will be relatively high. When CS voltage reaches the internal limitation (1V), the power MOSFET will be turned off instantaneously. This cycle by cycle current limitation can help protecting the power MOSFET, the inductor and the output diode.

### **Thermal Regulation**

The WS3411 integrates thermal regulation function. When the system is over temperature, the output current is gradually reduced; the output power and thermal dissipation are also reduced. The system temperature is regulated and the system reliability is improved. The thermal regulation temperature is set to 150°C internally.

### **PCB Layouts**

The following guidelines should be followed in WS3411 PCB layout:

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## **Bypass Capacitor**

The bypass capacitor on VCC pin should be as close as possible to the VCC and GND pins.

#### **Ground Path**

The power ground path for current sense resistor should be short and wide, and it should be as close as possible to the IC ground (pin 2), otherwise the LED output current accuracy maybe affected. The IC signal ground for COMP and FB components should be connected to the IC GND pin with short traces and should be away from the power ground path.

## The Area of Power Loop

The area of main current loop should be as small as possible to reduce EMI radiation.

### **FB Pin**

The feedback resistor divider should be as close as possible to the FB pin, and the trace must keeps away from dynamic node of the inductor (DRAIN pin trace, the positive pin and the negative pin of Vo trace), otherwise the FB pin OVP function might have risk to be mis-triggered by the system noise.

### **DRAIN Pin**

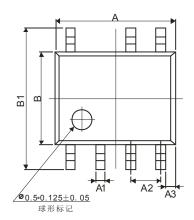
To increase the copper area of DRAIN pin for better thermal dissipation. However, too large copper area may compromise EMI performance.

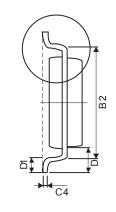
### The material of PCB

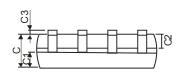
Avoid choosing the material that is easy to absorb the moisture, just like paper copper-clad laminates.

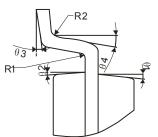


# Package Information SOP7Package Outline Dimensions









	Winsemi			
O	Dimensions in Millimeters		Dimensions in Inches	
Symbol	Min	Max	Min	Max
Α	4.70	5.10	0.185	0.201
В	3.70	4.10	0.146	0.161
С	1.30	1.50	0.051	0.059
A1	0.35	0.48	0.014	0.019
A2	1.27TYP		0.05TYP	
A3	0.345TYP		0.014TYP	
B1	5.80	6.20	0.228 0.244	
B2	5.00TYP		0.197TYP	
C1	0.55	0.70	0.022	0.028
C2	0.55	0.70	0.022	0.028
C3	0.05	0.225	0.002	0.009
C4	0.203TYP		0.008TYP	
D	1.05TYP 0.041TYP		1TYP	
D1	0.40	0.80	0.016	0.031



## NOTE:

- 1.We strongly recommend customers check carefully on the trademark when buying our product, if there is any question, please don't be hesitate to contact us.
- 2. Please do not exceed the absolute maximum ratings of the device when circuit designing.
- 3. Winsemi Microelectronics Co., Ltd reserved the right to make changes in this specification sheet and is subject to change without prior notice.

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