

FEATURES

- 2.7W Into 4Ω from 5.0V power supply at THD+N = 10% (Typ.).
- 1.7W Into 8Ω from 5.0V power supply at THD+N = 10% (Typ.)
- 2.0V~5.5V Power supply.
- Low shutdown current.(TYP.=0.05uA)
- Low quiescent current.(TYP.=3.0mA)
- Thermal shutdown protection with auto recovery feature.
- Advanced power ON/OFF pop reduction.
- Lead free and green package available. (RoHS Compliant)
- Space Saving Package
 - 8-pin EMSOP package. (with thermal pad)
 - 8-pin MSOP package. (without thermal pad)

APPLICATION

- Boom Box.
- Education, Toy.
- MID,
- Portable electronic devices.
- Mobile Phones.

GENERAL DESCRIPTION

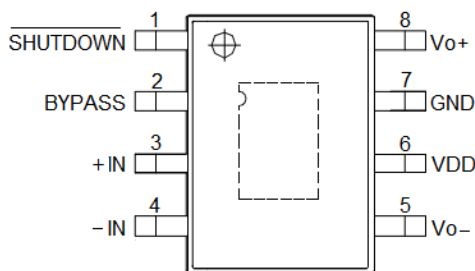
The LY8897 is a 2.7W audio power amplifier. It is capable of driving 4Ω speaker load at a continuous average output of 2.7W with less than 10% distortion (THD+N) from a 5.0V power supply and 8Ω speaker load at a continuous average output of 1.7W with less than 10% distortion (THD+N) from a 5.0V power supply.

The LY8897 primarily designed for high quality application in other portable communication device. And the LY8897 audio amplifier features low power consumption shutdown mode. It is achieved by driving the shutdown pin with logic low. And the LY8897 has an internal thermal shutdown protection feature.

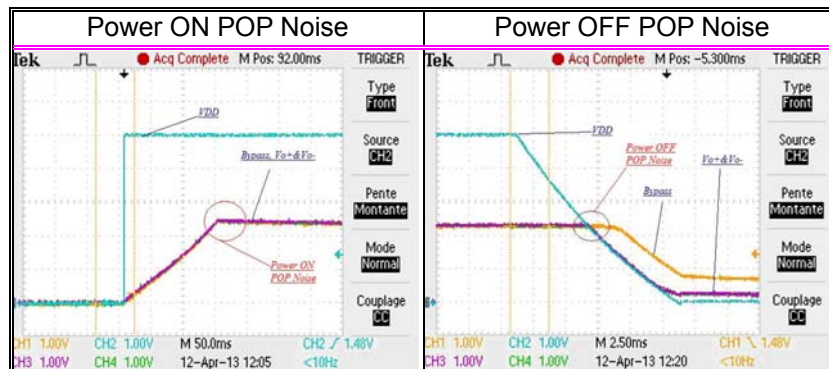
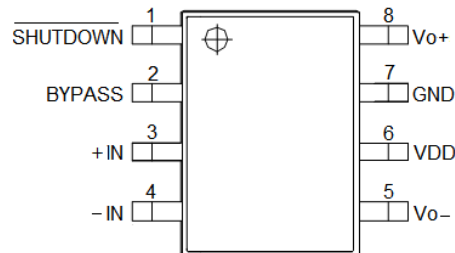
The LY8897 audio amplifier was designed specifically to provide high quality output power with a minimal amount of external components. The LY8897 does not require output capacitors, and the LY8897 is ideally suited for other low voltage applications or portable electronic devices where minimal power consumption is a primary requirement.

PIN CONFIGURATION

LY8897 EMSOP8 pin configuration (TOP VIEW)



LY8897 MSOP8 pin configuration (TOP VIEW)



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PIN DESCRIPTION

SYMBOL	Pin No.		DESCRIPTION
	EMSOP8	MSOP8	
SHUTDOWN	1	1	Shutdown the device. (when LOW level is shutdown mode)
BYPASS	2	2	Bypass pin
+IN	3	3	Positive Input
-IN	4	4	Negative Input
Vo1(-)	5	5	Negative output
V _{DD}	6	6	Power Supply
GND	7	7	Ground
Vo2(+)	8	8	Positive Output

ORDERING INFORMATION

Ordering Code	Speaker Channels	Pin/ Package	Output Power (THD+N=10%)	Input Type	Output Type
LY8897X	Mono	EMSOP8 (with thermal pad)	2.7W/4Ω @5.0V_BTL 1.7W/8Ω @5.0V_BTL	SE/ DF	BTL
LY8897U	Mono	MSOP8 (without thermal pad)	2.7W/4Ω @5.0V_BTL 1.7W/8Ω @5.0V_BTL	SE/ DF	BTL

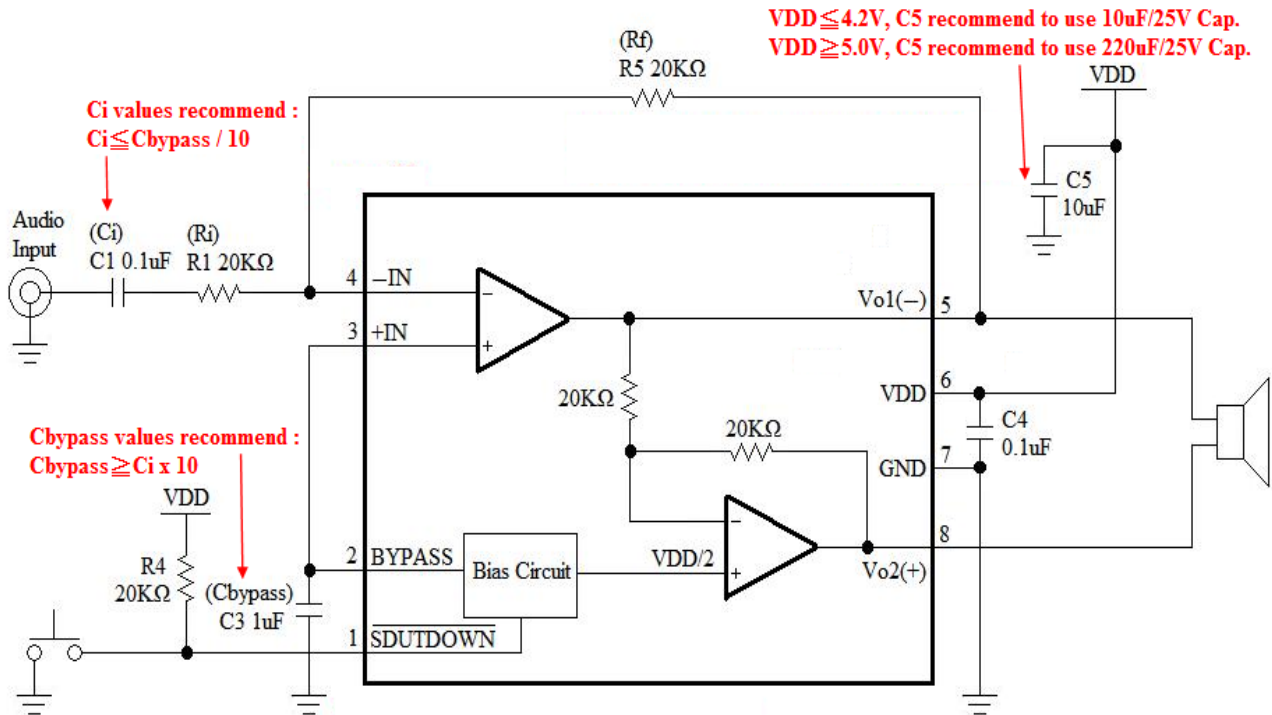
APPLICATION CIRCUIT


Figure 1. LY8897 application schematic with Single -Ended input

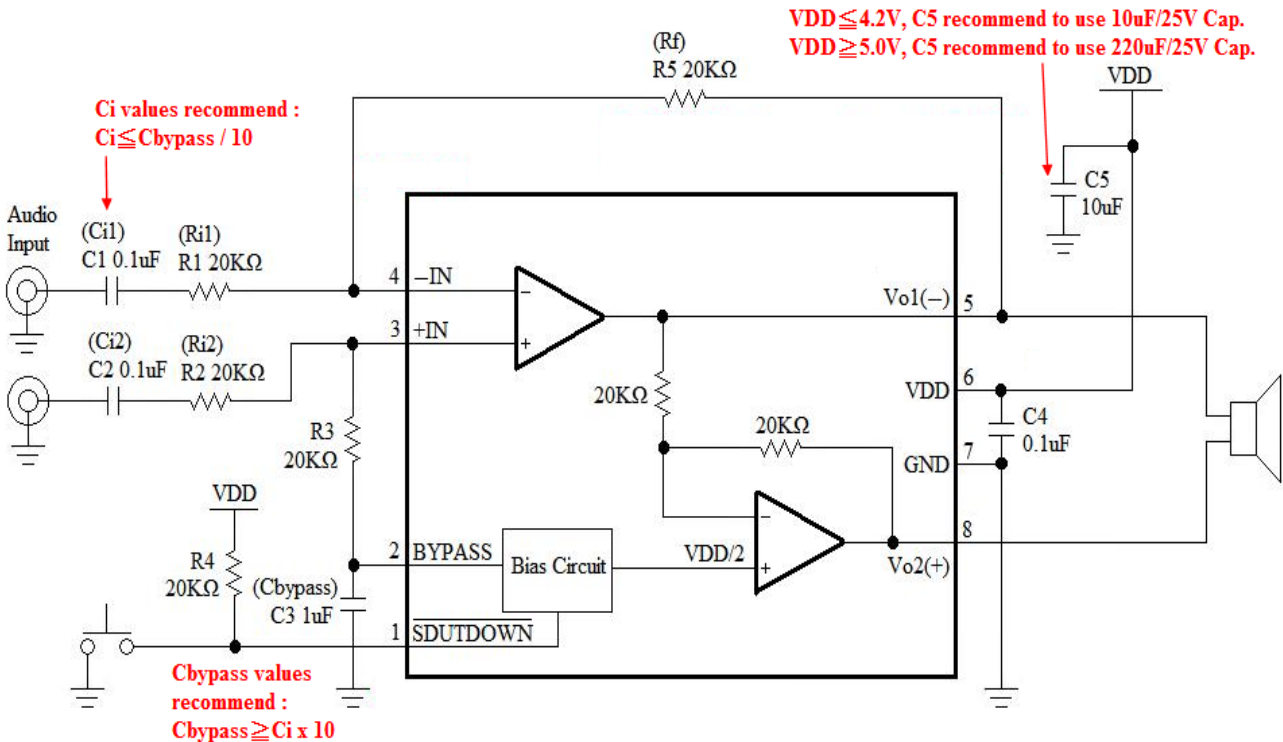


Figure 2. LY8897 application schematic with Differential input



ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	RATING	UNIT
Supply Voltage	V _{DD}	6.0	V
Operating Temperature	T _A	-40 to 85 (I grade)	°C
Input Voltage	V _I	-0.3V to V _{DD} +0.3V	V
Storage Temperature	T _{STG}	-65 to 150	°C
Power Dissipation	P _D	Internally Limited	W
ESD Susceptibility	V _{ESD}	2000	V
Junction Temperature	T _{JMAX}	150	°C
Soldering Temperature (under 10 sec)	T _{SOLDER}	260	°C

DC ELECTRICAL CHARACTERISTICS (T_A=25°C)

PARAMETER	SYMBOL	TEST CONDITION	MIN.	TYP. ^(*)	MAX.	UNIT
Power supply voltage	V _{DD}	--	2.0	-	5.5	V
Quiescent Current	I _Q	V _{DD} =5.0V, Load=4Ω.	-	3.0	9.0	mA
Shutdown Current	I _{SD}	V _{DD} =5.0V, V _{SD} =0V	-	0.05	2.0	μA
Shutdown Voltage Input High	V _{SDIH}	V _{DD} =5.0V, V _{SD Mode} = V _{DD}	1.2	-	-	V
Shutdown Voltage Input Low	V _{SDIL}	V _{DD} =5.0V, V _{SD Mode} = GND	-	-	0.4	
Output Offset Voltage	V _{OS}	V _{DD} =5.0V, Load=4Ω.	-	7.0	50.0	mV

(*2) Typical values are included for reference only and are not guaranteed or tested.

Typical values are measured at VCC = VCC(TYP.) and T_A = 25°C

OPERATING CHARACTERISTICS(1) (T_A=25°C)

PARAMETER	SYMBOL	TEST CONDITION	MIN.	TYP. ^(*)	MAX.	UNIT	
Power Supply Rejection Ratio	PSRR	V _{ripple} = 200mV sine p-p, V _{DD} =5.0V, R _L =4Ω, A _v =2. Input=GND.	1KHz	-	68.77	-	dB
			217Hz	-	55.60	-	
		V _{ripple} = 200mV sine p-p, V _{DD} =5.0V, R _L =4Ω, A _v =2. Input=Floating.	1KHz	-	68.31	-	
			217Hz	-	67.73	-	
Signal-to-noise ratio	SNR	V _{DD} =5.0V, R _L =8Ω, Input=GND, A _v =2, 1W=0dB,	A-weighting	-	94.26	-	dB
			None A-weighting	-	91.78	-	
Output voltage noise	V _n	V _{DD} =5.0V, A _v =2, f=20Hz~20KHz, Input=GND, R _L =4Ω SPK,	A-weighting	-	38.7	-	μV
			None A-weighting	-	51.5	-	
Thermal shutdown temperature	T _{SD}	Shutdown temp.	-	160	-	°C	
		Restore temp.	-	130	-		

(*2) Typical values are included for reference only and are not guaranteed or tested.

Typical values are measured at VCC = VCC(TYP.) and T_A = 25°C



OPERATING CHARACTERISTICS(2) (TA=25°C)

PARAMETER	SYMBOL	TEST CONDITION	MIN.	TYP. ^{*2}	MAX.	UNIT		
Output Power	Po	RL=4Ω ^{*3} THD=10%,f=1 kHz	VDD=5.5V	-	3.3	-	W	
			VDD=5.0V	-	2.7	-		
			VDD=3.7V	-	1.5	-		
			VDD=2.5V	-	0.6	-		
			VDD=2.0V	-	0.37	-		
			VDD=5.5V	-	2.5	-		
		RL=4Ω ^{*3} THD=1%,f=1 kHz	VDD=5.0V	-	2.0	-		
			VDD=3.7V	-	1.1	-		
			VDD=2.5V	-	0.5	-		
			VDD=2.0V	-	0.24	-		
			RL=8Ω THD=10%,f=1 kHz	VDD=5.5V	-	2.0		-
				VDD=5.0V	-	1.7		-
		VDD=3.7V		-	1.0	-		
		VDD=2.5V		-	0.4	-		
		RL=8Ω THD=1%,f=1 kHz	VDD=2.0V	-	0.25	-		
			VDD=5.5V	-	1.6	-		
			VDD=5.0V	-	1.3	-		
			VDD=3.7V	-	0.7	-		
VDD=2.5V	-	0.3	-					
	-	0.2	-					

(*2) Typical values are included for reference only and are not guaranteed or tested.

Typical values are measured at VCC = VCC(TYP.) and TA = 25°C

(*3) When driving 4Ω loads from 4.2V~5V power supply, the device must be mounted to a circuit board.



TYPICAL PERFORMANCE CHARACTERISTICS

Figure 3
THD+N vs. Output Power (@ $R_L=4\Omega$, $f=1\text{kHz}$, $A_v=10$)

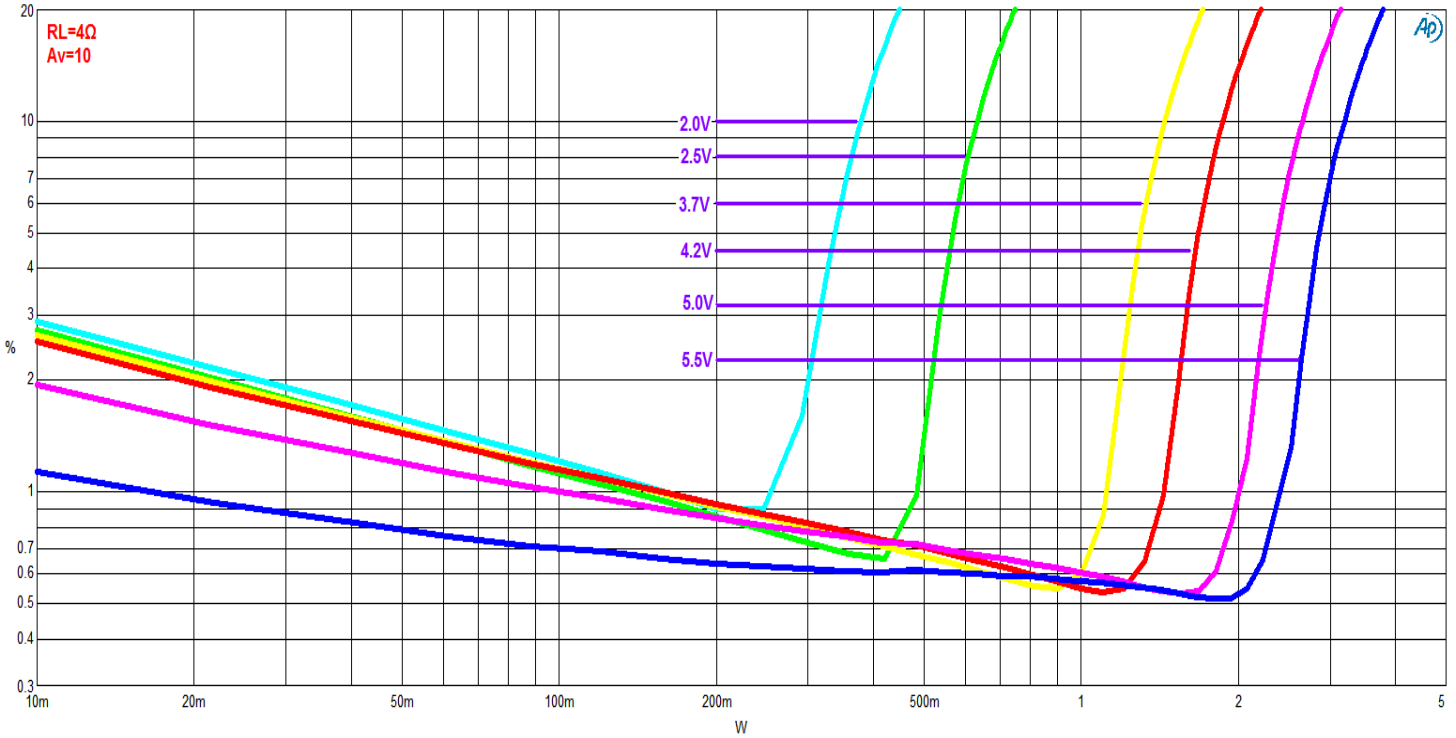


Figure 4
THD+N vs. Output Power (@ $R_L=8\Omega$, $f=1\text{kHz}$, $A_v=10$)

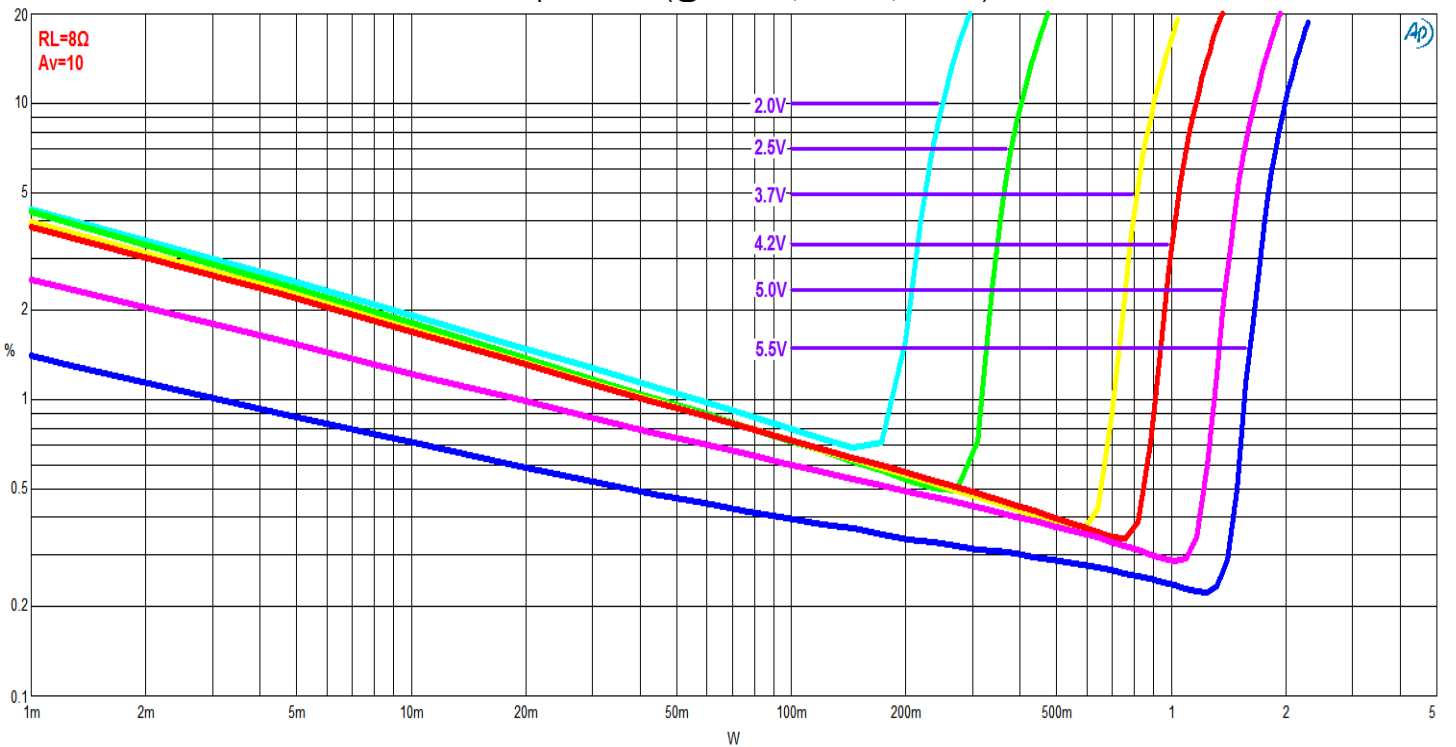


Figure 5
Power Supply Rejection Ratio (PSRR)

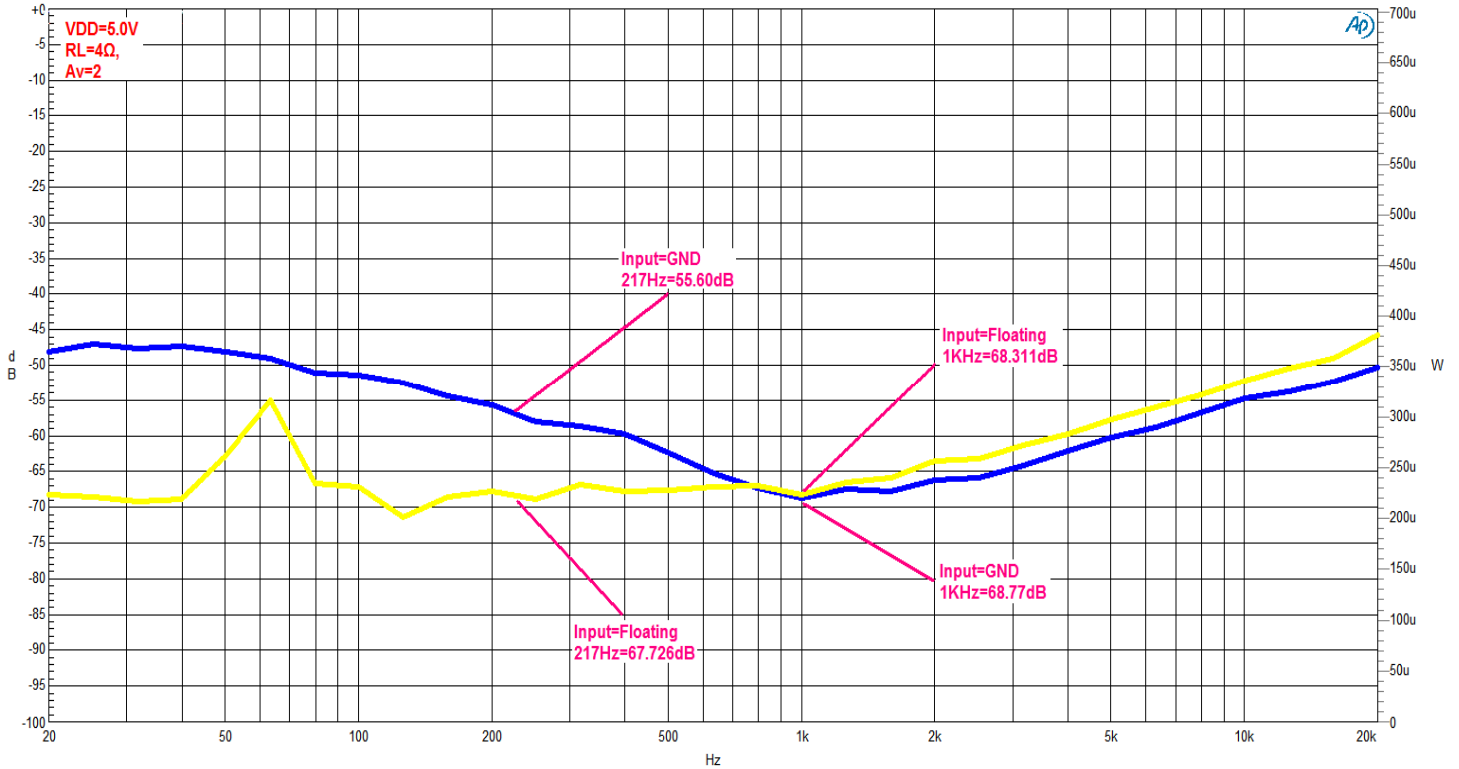


Figure 6
Frequency response

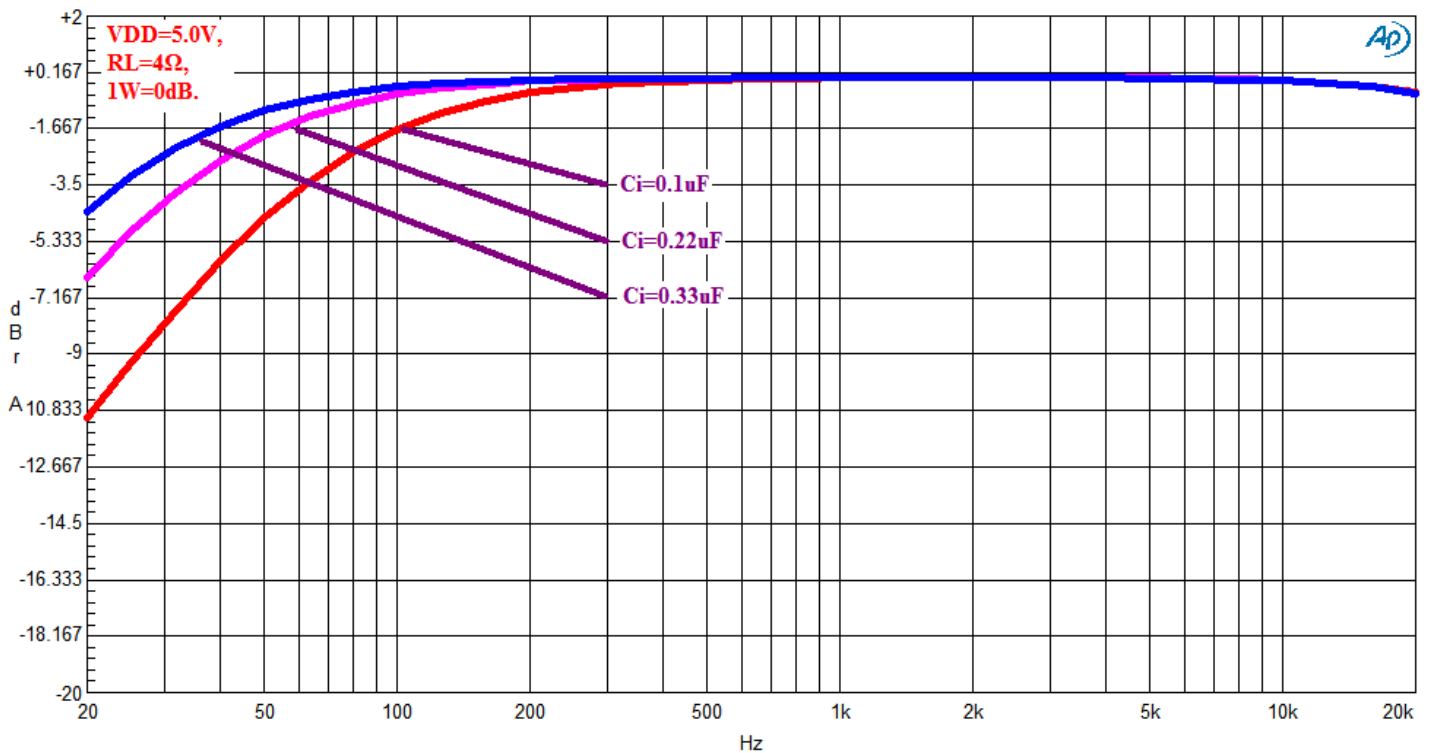
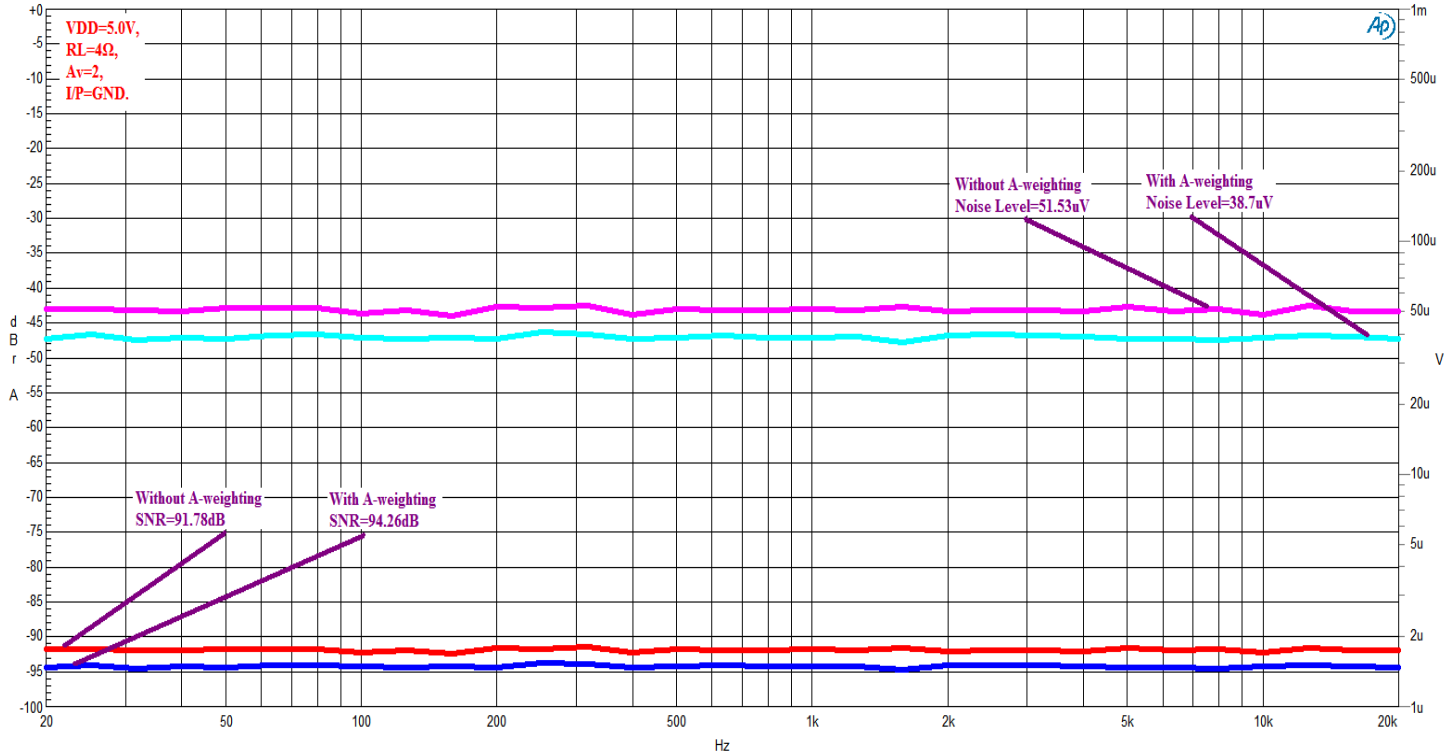
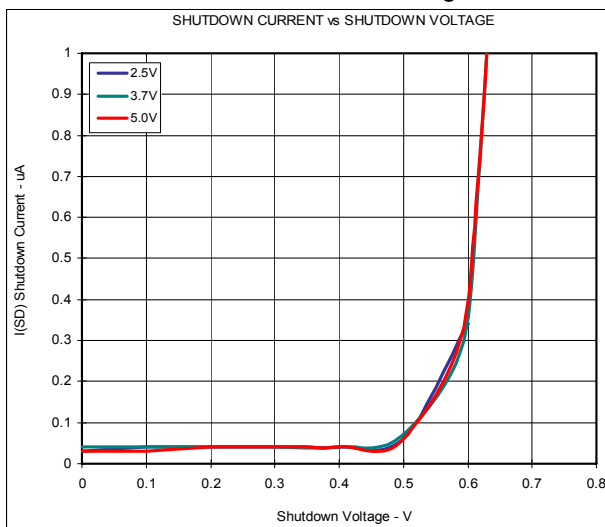


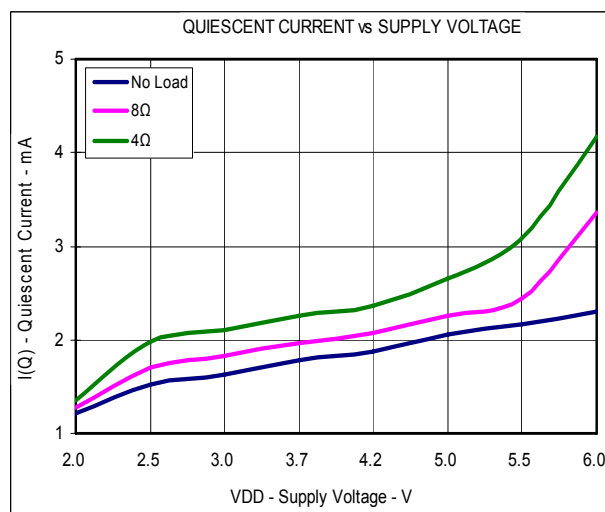
Figure 7
SNR vs. Noise Level



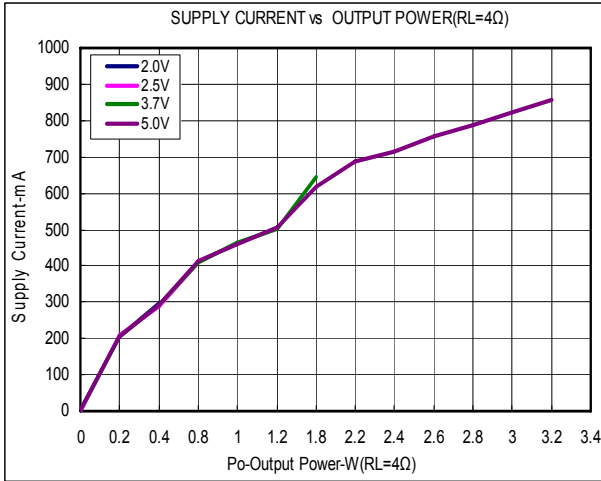
SD Current vs. SD Voltage



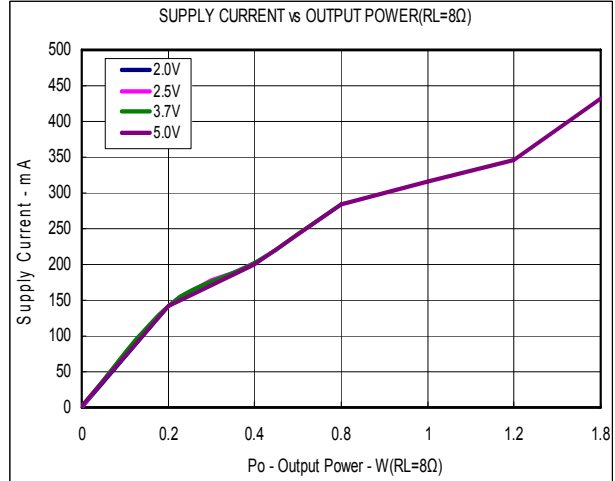
Quiescent Current vs. Supply voltage



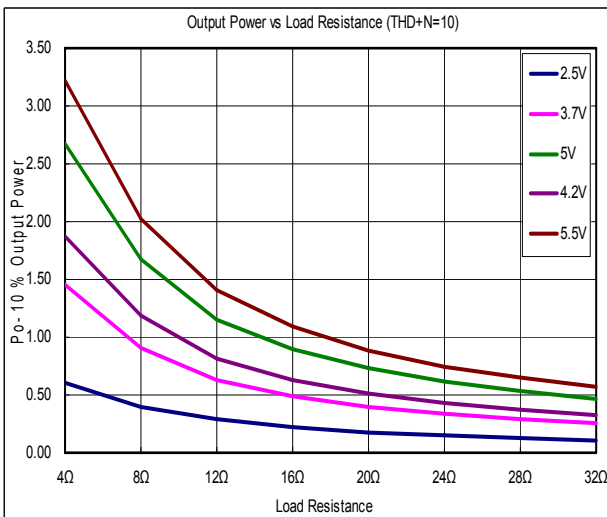
V_{DD} Current vs. Output Power (RL=4Ω)



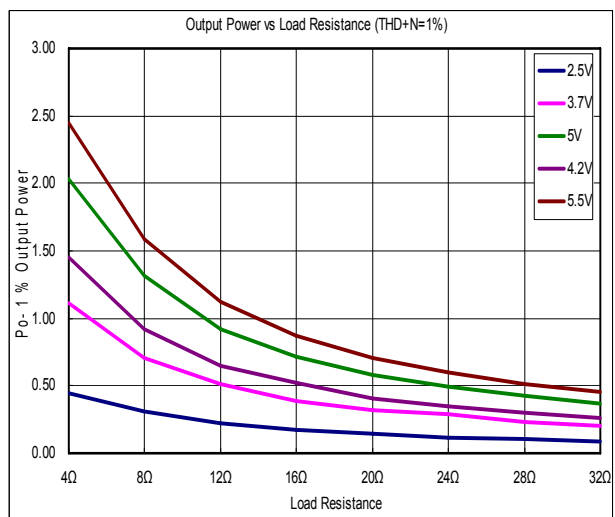
V_{DD} Current vs. Output Power (RL=8Ω)



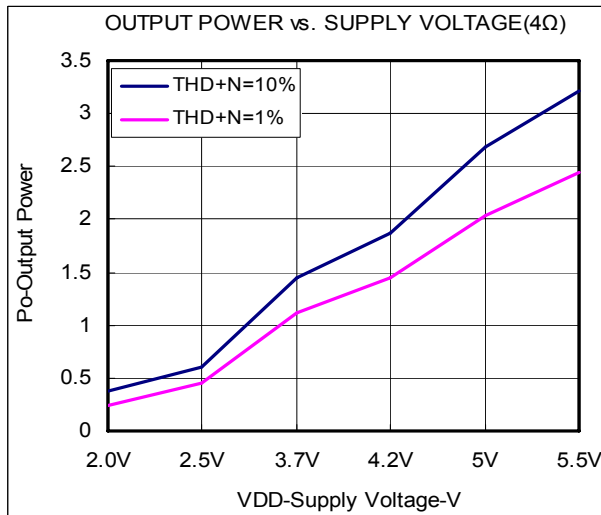
Load Resistance vs. Output Power (THD+N=10%)



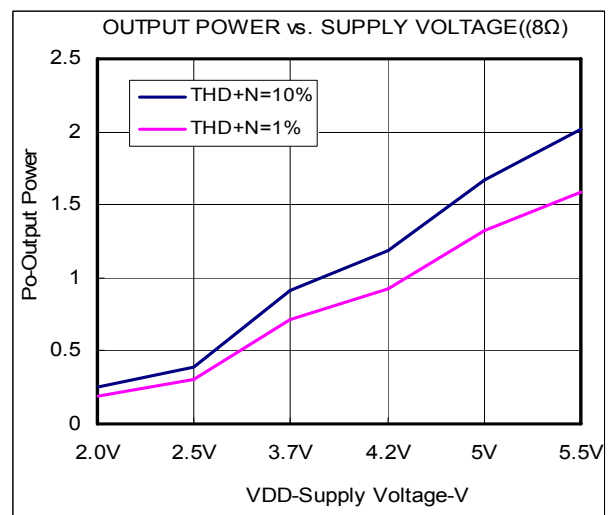
Load Resistance vs. Output Power (THD+N=1%)



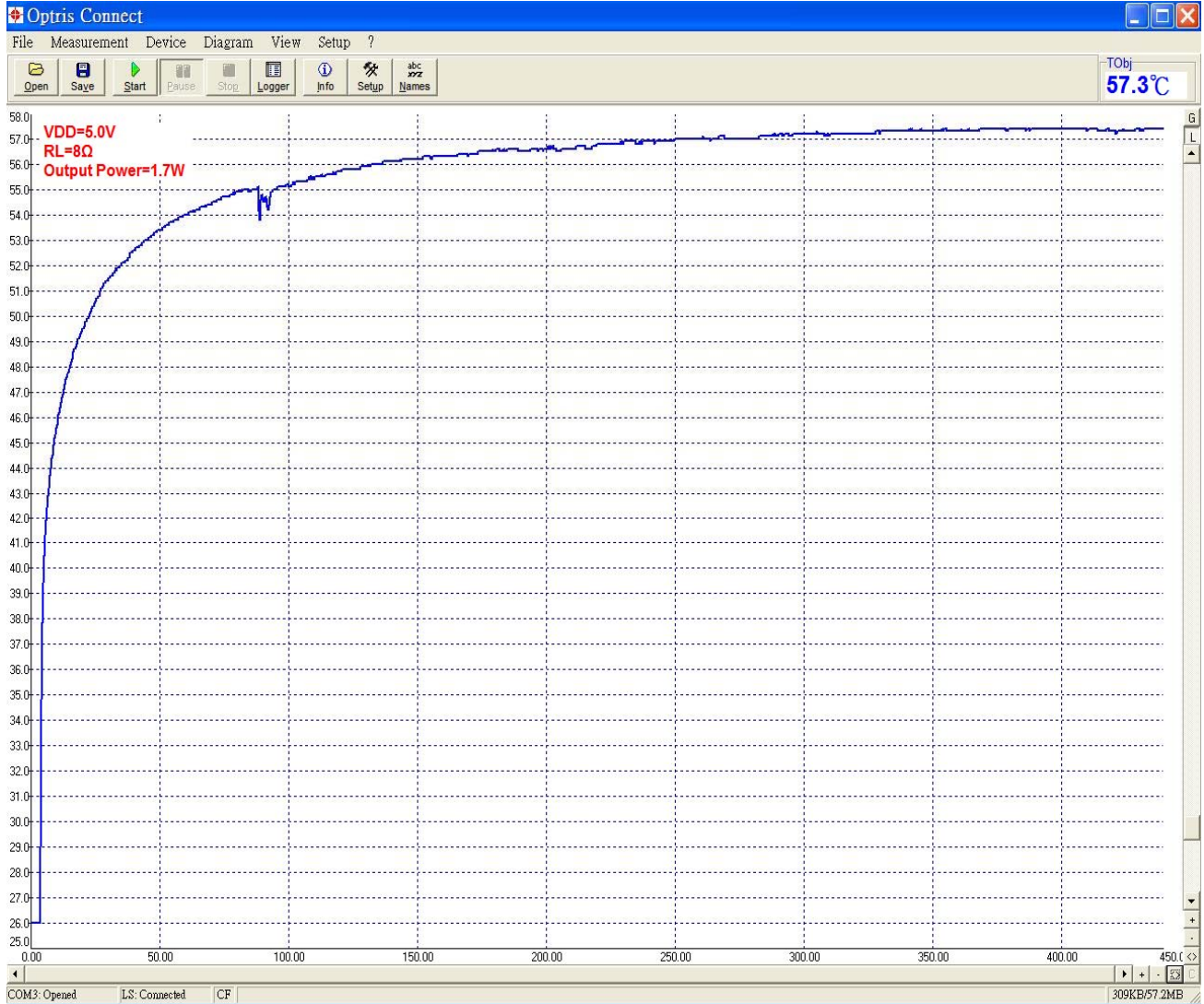
V_{DD} vs. Output Power (RL=4Ω)



V_{DD} vs. Output Power (RL=8Ω)



Output Power vs. Temperature



APPLICATION INFORMATION

Bridged configuration explanation

As shown in *Figure 1 and 2*, the LY8897 has two operational amplifiers internally, allowing for a few different amplifier configurations. The first amplifier's gain is externally configurable, while the second amplifier is internally fixed in a unity-gain, inverting configuration. The closed-loop gain of the first amplifier is set by selecting the ratio of R_f to R_{IN} while the second amplifier's gain is fixed by the two internal 20k Ω resistors. *Figure 1 and 2* shows that the output of amplifier one serves as the input to amplifier two which results in both amplifiers producing signals identical in magnitude, but out of phase by 180°. Consequently, the differential gain for the IC is

$$A_{VD} = 2 \times (R_f / R_{IN}) \dots\dots\dots(1)$$

By driving the load differentially through outputs V_{o1} and V_{o2} , an amplifier configuration commonly referred to as "bridged mode" is established. Bridged mode operation is different from the classical single-ended amplifier configuration where one side of the load is connected to ground.

A bridge amplifier design has a few distinct advantages over the single-ended configuration, as it provides differential drive to the load, thus doubling output swing for a specified supply voltage. Four times the output power is possible as compared to a single-ended amplifier under the same conditions.

This increase in attainable output power assumes that the amplifier is not current limited or clipped. In order to choose an amplifier's closed-loop gain without causing excessive clipping, please refer to the Audio Power Amplifier Design section.

A bridge configuration, such as the one used in the LY8897, also creates a second advantage over single-ended amplifiers. Since the differential outputs, V_{o1} and V_{o2} , are biased at half-supply, no net DC voltage exists across the load. This eliminates the need for an output coupling capacitor which is required in a single supply, single-ended amplifier configuration. Without an output coupling capacitor, the half-supply bias across the load would result in both increased internal IC power dissipation and also possible loudspeaker damage.

Input Capacitors (C_i)

The LY8897 input capacitors and input resistors form a high-pass filter with the corner frequency, f_c , determined in equation Equation 2.

$$f_c = \frac{1}{2\pi R_i C_i} \dots\dots\dots(2)$$

Equation 3 is reconfigured to solve for the input coupling capacitance.

$$C_i = \frac{1}{2\pi R_i f_c} \dots\dots\dots(3)$$

For example

In the table 1 shows the external components. R_{in} in connect with C_{in} to create a high-pass filter.

Table 1. Typical Component Values

Reference	Description				Note
R_i	20K Ω				1% tolerance resistors
C_i	0.1uF	0.22uF	0.33uF	0.47uF	80%/-20% non polarized
corner frequency	79.57Hz	36.17Hz	20.4Hz	16.93Hz	

$$C_i = 1 / (2\pi R_i f_c)$$

$C_i = 1 / (2\pi \times 20K\Omega \times 20Hz) = 0.1\mu F$, One would likely choose a value of 0.1 μF as this value is commonly used.

Note that it is important to C_i must be 10 times smaller than the bypass capacitor to reduce clicking and popping noise from power on/off and entering and leaving shutdown. After sizing C_i for a given cutoff frequency, size the bypass capacitor to 10 times that of the input capacitor.

$$C_i \leq C_{bypass} / 10$$

Bypass Capacitor Value Selection

Besides minimizing the input capacitor size, careful consideration should be paid to value of C_{bypass} , the capacitor connected to the BYPASS pin. Since C_{bypass} determines how fast the LY8897 settles to quiescent operation, its value is critical when minimizing turn-on pops. The slower the LY8897's outputs ramp to their quiescent DC voltage (nominally 1/2 VDD), the smaller the turn-on pop. Choosing C_{bypass} value equal 1.0 to 10 μF along with a small value of C_i (in the range of 1 nF to 0.39 μF), produces a click-less and pop-less shutdown function. As discussed above, choosing C_i no larger than necessary for the desired bandwidth helps minimize clicks and pops. Therefore, increasing the bypass capacitor reduces clicking and popping noise from power on/off and entering and leaving shutdown. To have minimal pop, C_{bypass} should be 10 times larger than C_i .

$$C_{bypass} \geq C_i \times 10$$

Table 2. CBYPASS Reference Component Values

PARAMETER	SYMBOL	TEST CONDITION	MIN.	TYP.	MAX.	UNIT	
Start-up time from shutdown	Z_i	V _{DD} =5.0V, C _i =0.1 μF , R _i =20K Ω , A _v =10	C _{bypass} = 10 μf	-	940	-	ms
			C _{bypass} = 4.7 μf	-	550	-	
			C _{bypass} = 2.2 μf	-	190	-	
			C _{bypass} = 1.0 μf	-	158	-	
		V _{DD} =3.7V, C _i =0.1 μF , R _i =20K Ω , A _v =10	C _{bypass} = 10 μf	-	780	-	
			C _{bypass} = 4.7 μf	-	480	-	
			C _{bypass} = 2.2 μf	-	180	-	
			C _{bypass} = 1.0 μf	-	126	-	

Power Supply Bypassing Capacitor

As with any amplifier, proper supply bypassing is critical for low noise performance and high power supply rejection. The capacitor location on both the bypass and power supply pins should be as close to the device as possible.

The LY8897 is a mono class AB audio amplifier that requires adequate power supply decoupling to ensure the efficiency is high and total harmonic distortion (THD) is low. For higher frequency transients, spikes, or digital hash on the line, a good low equivalent-series-resistance ceramic capacitor or electrolytic capacitors, typically 10~220 μF , placed as close as possible to the device VDD lead works best. Placing 0.1 μF decoupling capacitor close to the LY8897 is very important for the efficiency of the class AB amplifier, because any resistance or inductance in the trace between the device and the capacitor can cause a loss in efficiency and protect device damage. For filtering lower-frequency noise signals, a 10.0 μF or greater capacitor placed near the audio power amplifier would also help, but it is not required in most applications because of the high PSRR of this device.

Shutdown Function

When the LY8897 not in use. The device will be to turn off the amplifier to reduce power consumption. When logic low is applied to the shutdown pin, this shutdown feature will turns the amplifier off. By switching the shutdown pin connected to GND, the device supply current draw will be minimized in idle mode. The pin cannot be left floating due to the internal did not pull-up.

Over-Heat Protection

The LY8897 has a built-in over-heat protection circuit , it will turn off all power output when the chip temperature over 160°C , the chip will return to normal operation automatically after the temperature cool down to 130°C .

PCB LAYOUT

All the external components must place very close to the LY8897. The input resistors need to be very close to the LY8897 input pins so noise does not couple on the high impedance nodes between the input resistors and the input amplifier of the LY8897. Then place the decoupling capacitor Cs, close to the LY8897 is important for the efficiency of the class AB amplifier. Any resistance or inductance in the trace between the device and the capacitor can cause a loss in efficiency.

Making the high current traces going to VDD, GND, VO+ and VO- pins of the LY8897 should be as wide as possible to minimize trace resistance. If these traces are too thin, the LY8897's performance and output power will decrease. The input traces do not need to be wide, but do need to run side-by-side to enable common-mode noise cancellation.

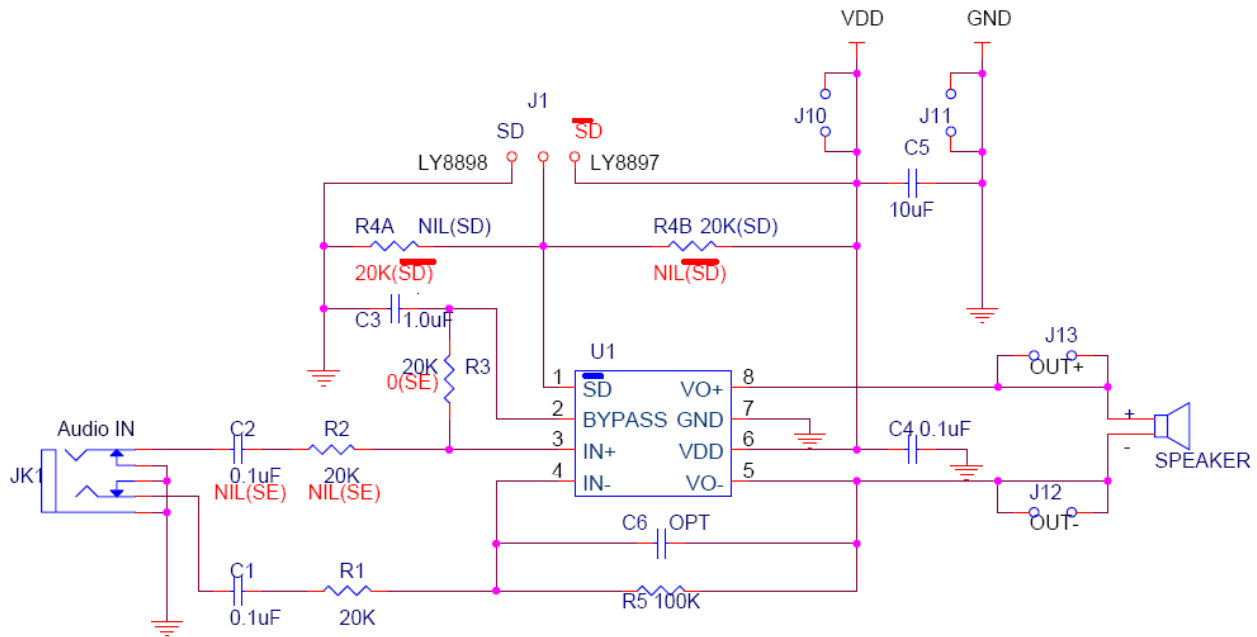
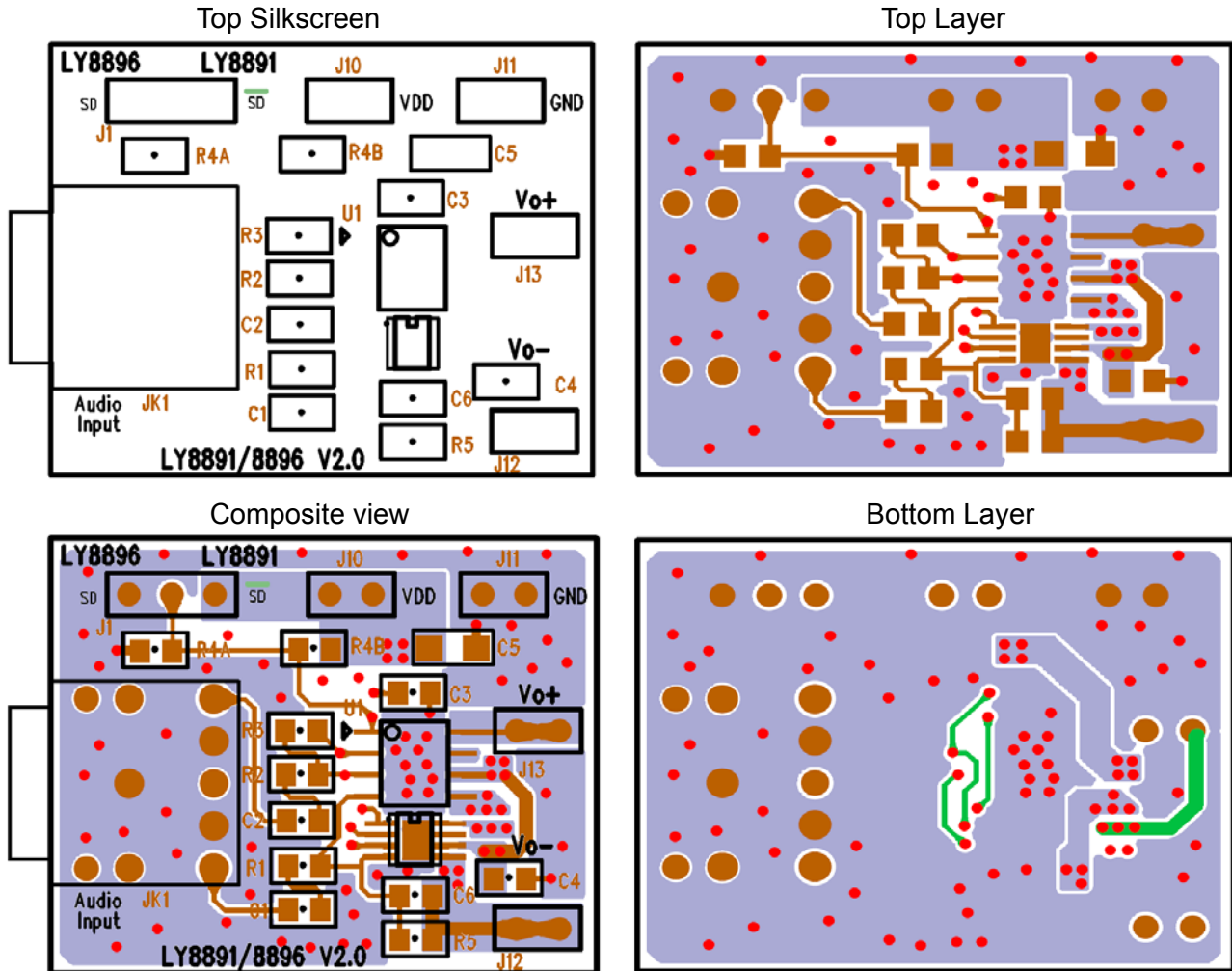
DEMO BOARD INFORMATION
Demo Board Application Circuit :


Figure 8. LY8897 Demo Board Application Circuit

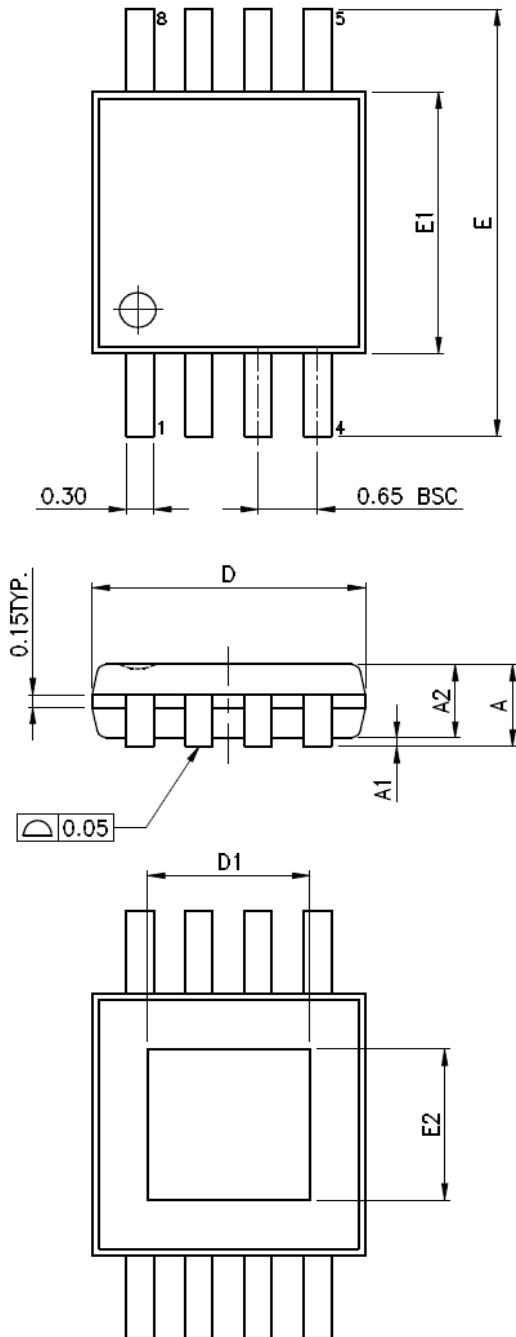
Demo Board BOM List :
LY8897 V2.0 BOM List (Stereo Mode)

No.	Description	Reference	Note
1	Resistor, 20KΩ	R1, R4A (R2,R3 DF input only)	1/16W,1%
2	Resistor, 100KΩ	R5	1/16W,1%
3	Capacitor, 0.1uF	C1, C4 (C2 DF input only)	80%/-20%, non polarized
4	Capacitor, 1.0uF	C3	80%/-20%, non polarized
5	Capacitor, 10.0uF	C5	25V,105°C,8x11,EC Cap.
6	IC	U1	LY8897X or LY8897U
7	1*2 Pin Header	J1	Pitch 2.54 mm

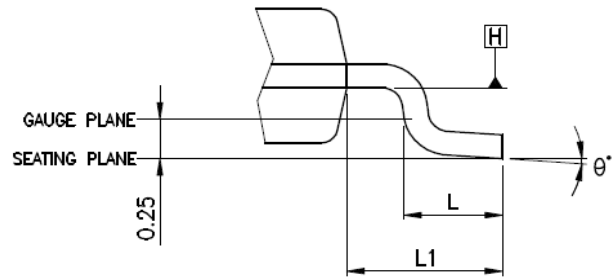
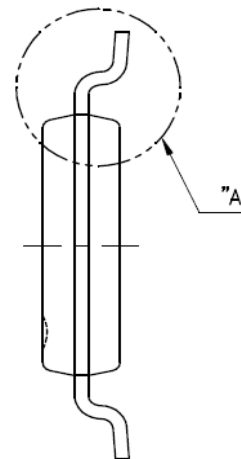
Demo Board Artwork :



(*3) When driving 4Ω loads from 4.2V~5V power supply, the device must be mounted to a circuit board.

PACKAGE OUTLINE DIMENSION
8 Pin EMSOP Package Outline Dimension


THERMALLY ENHANCED VARIATIONS ONLY

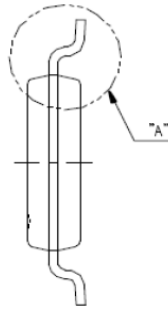
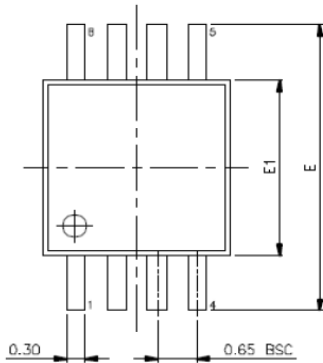

DETAIL A


SYMBOLS	MIN.	NOM.	MAX.
A	—	—	1.10
A1	0.00	—	0.15
A2	0.75	0.85	0.95
D	3.00 BSC		
E	4.90 BSC		
E1	3.00 BSC		
L	0.40	0.60	0.80
L1	0.95 REF		
θ°	0	—	8

UNIT : MM

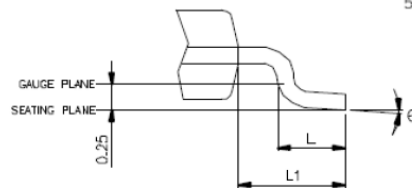
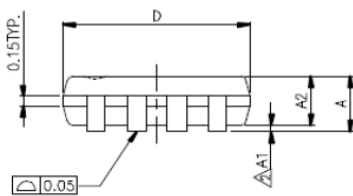
THERMALLY ENHANCED DIMENSIONS(SHOWN IN MM)

PAD SIZE	E2		D1	
	MIN.	MAX.	MIN.	MAX.
68X70E	1.38	1.73	1.42	1.78

8 pin 25.6 mil MSOP Package Outline Dimension


SYMBOLS	MIN.	NOM.	MAX.
A	—	—	1.10
A1	0.00	—	0.15
A2	0.75	0.85	0.95
D	3.00 BSC		
E	4.90 BSC		
E1	3.00 BSC		
L	0.40	0.60	0.80
L1	0.95 REF		
θ°	0	—	8

UNIT : MM


NOTES:

1. JEDEC OUTLINE : MO-187 AA
2. DIMENSION 'D' DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.15 PER SIDE.
3. DIMENSION 'E1' DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION. INTERLEAD FLASH OR PROTRUSION SHALL NOT EXCEED 0.25 PER SIDE.
4. DIMENSION '0.22' DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.08 MM TOTAL IN EXCESS OF THE '0.22' DIMENSION AT MAXIMUM MATERIAL CONDITION. DAMBAR CANNOT BE LOCATED ON THE LOWER RADIUS OF THE FOOT. MINIMUM SPAC BETWEEN PROTRUSION AND ADJACENT LEAD IS 0.07 MM.
5. DIMENSIONS 'D' AND 'E1' TO BE DETERMINED AT DATUM PLANE \square .